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Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of [Embedded - Microprocessors](#)

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details

Product Status	Obsolete
Core Processor	PowerPC e500
Number of Cores/Bus Width	2 Core, 32-Bit
Speed	1.067GHz
Co-Processors/DSP	Signal Processing; SPE
RAM Controllers	DDR2, DDR3
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100/1000Mbps (4)
SATA	-
USB	-
Voltage - I/O	1.5V, 1.8V, 2.5V, 3.3V
Operating Temperature	0°C ~ 105°C (TA)
Security Features	-
Package / Case	1023-BFBGA, FCBGA
Supplier Device Package	1023-FCBGA (33x33)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mpc8572lvtarld

- Ability to launch DMA from single write transaction
- Serial RapidIO interface unit
 - Supports *RapidIO Interconnect Specification, Revision 1.2*
 - Both 1x and 4x LP-serial link interfaces
 - Long- and short-haul electricals with selectable pre-compensation
 - Transmission rates of 1.25, 2.5, and 3.125 Gbaud (data rates of 1.0, 2.0, and 2.5 Gbps) per lane
 - Auto-detection of 1x- and 4x-mode operation during port initialization
 - Link initialization and synchronization
 - Large and small size transport information field support selectable at initialization time
 - 34-bit addressing
 - Up to 256 bytes data payload
 - All transaction flows and priorities
 - Atomic set/clr/inc/dec for read-modify-write operations
 - Generation of IO_READ_HOME and FLUSH with data for accessing cache-coherent data at a remote memory system
 - Receiver-controlled flow control
 - Error detection, recovery, and time-out for packets and control symbols as required by the RapidIO specification
 - Register and register bit extensions as described in part VIII (Error Management) of the RapidIO specification
 - Hardware recovery only
 - Register support is not required for software-mediated error recovery.
 - Accept-all mode of operation for fail-over support
 - Support for RapidIO error injection
 - Internal LP-serial and application interface-level loopback modes
 - Memory and PHY BIST for at-speed production test
- RapidIO-compliant message unit
 - 4 Kbytes of payload per message
 - Up to sixteen 256-byte segments per message
 - Two inbound data message structures within the inbox
 - Capable of receiving three letters at any mailbox
 - Two outbound data message structures within the outbox
 - Capable of sending three letters simultaneously
 - Single segment multicast to up to 32 devIDs
 - Chaining and direct modes in the outbox
 - Single inbound doorbell message structure
 - Facility to accept port-write messages

Figure 1 shows the MPC8572E block diagram.

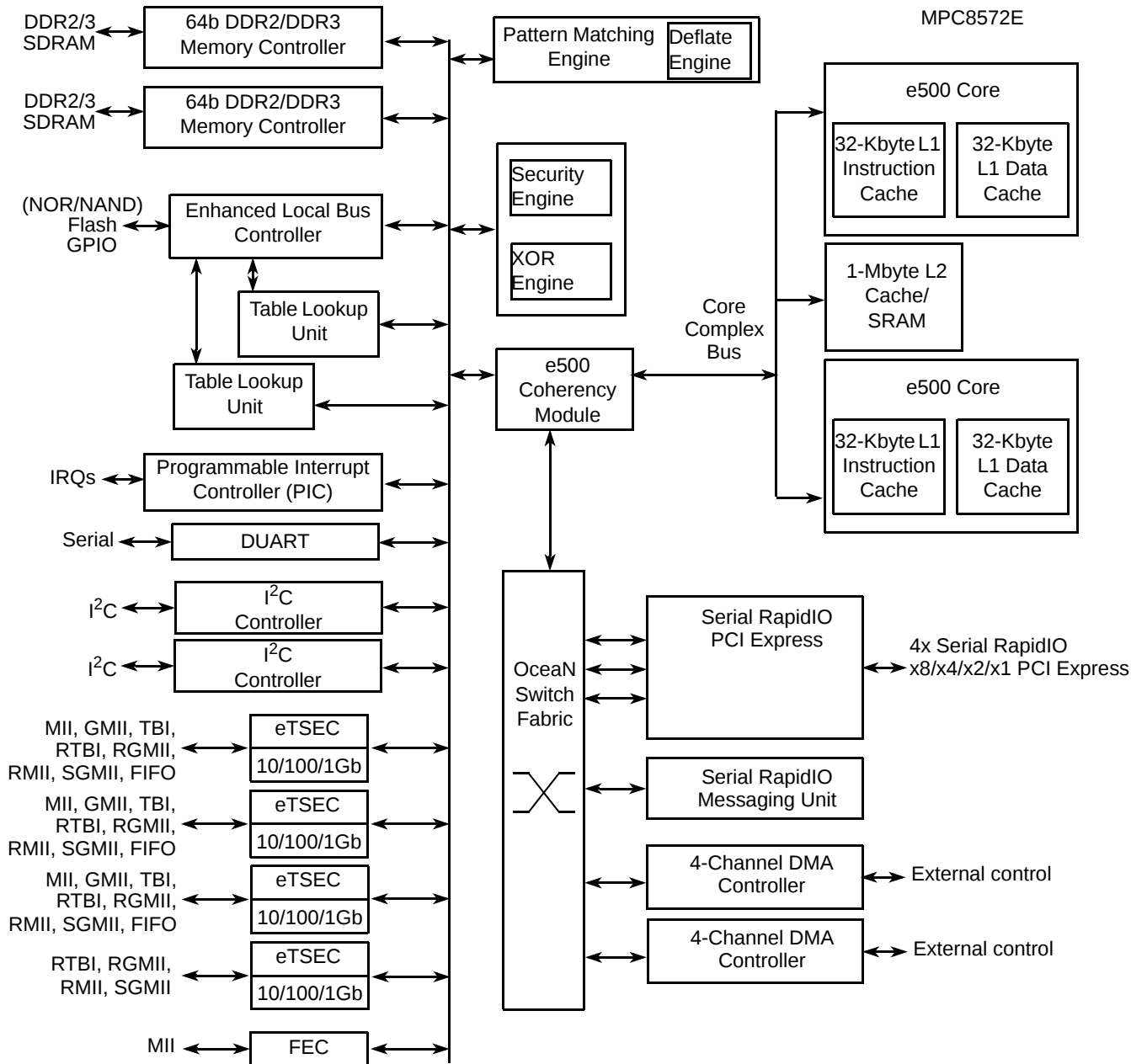


Figure 1. MPC8572E Block Diagram

2 Electrical Characteristics

This section provides the AC and DC electrical specifications for the MPC8572E. The MPC8572E is currently targeted to these specifications. Some of these specifications are independent of the I/O cell, but are included for a more complete reference. These are not purely I/O buffer design specifications.

Table 8. DDRCLK AC Timing Specifications (continued)

At recommended operating conditions with OV_{DD} of $3.3V \pm 5\%$.

Parameter/Condition	Symbol	Min	Typical	Max	Unit	Notes
DDRCLK jitter	—	—	—	+/- 150	ps	4, 5, 6

Notes:

- Caution:** The DDR complex clock to DDRCLK ratio settings must be chosen such that the resulting DDR complex clock frequency does not exceed the maximum or minimum operating frequencies. Refer to [Section 19.4, “DDR/DDRCLK PLL Ratio,”](#) for ratio settings.
- Rise and fall times for DDRCLK are measured at 0.6 V and 2.7 V.
- Timing is guaranteed by design and characterization.
- This represents the total input jitter—short term and long term—and is guaranteed by design.
- The DDRCLK driver's closed loop jitter bandwidth should be <500 kHz at -20 dB. The bandwidth must be set low to allow cascade-connected PLL-based devices to track DDRCLK drivers with the specified jitter.
- For spread spectrum clocking, guidelines are +0% to -1% down spread at a modulation rate between 20 kHz and 60 kHz on DDRCLK.

4.5 Platform to eTSEC FIFO Restrictions

Note the following eTSEC FIFO mode maximum speed restrictions based on platform (CCB) frequency.

For FIFO GMII modes (both 8 and 16 bit) and 16-bit encoded FIFO mode:

FIFO TX/RX clock frequency $\leq$ platform clock (CCB) frequency/4.2

For example, if the platform (CCB) frequency is 533 MHz, the FIFO TX/RX clock frequency should be no more than 127 MHz.

For 8-bit encoded FIFO mode:

FIFO TX/RX clock frequency $\leq$ platform clock (CCB) frequency/3.2

For example, if the platform (CCB) frequency is 533 MHz, the FIFO TX/RX clock frequency should be no more than 167 MHz.

4.6 Other Input Clocks

For information on the input clocks of other functional blocks of the platform, such as SerDes and eTSEC, see the respective sections of this document.

5 RESET Initialization

[Table 9](#) describes the AC electrical specifications for the RESET initialization timing.

Table 9. RESET Initialization Timing Specifications

Parameter/Condition	Min	Max	Unit	Notes
Required assertion time of $\overline{HRESET}$	100	—	μs	2
Minimum assertion time for $\overline{SRESET}$	3	—	SYSCLKs	1

7.2 DUART AC Electrical Specifications

Table 22 provides the AC timing parameters for the DUART interface.

Table 22. DUART AC Timing Specifications

At recommended operating conditions with OV_{DD} of $3.3V \pm 5\%$.

Parameter	Value	Unit	Notes
Minimum baud rate	$f_{CCB}/1,048,576$	baud	1, 2
Maximum baud rate	$f_{CCB}/16$	baud	1, 2, 3
Oversample rate	16	—	1, 4

Notes:

1. Guaranteed by design
2. f_{CCB} refers to the internal platform clock frequency.
3. Actual attainable baud rate is limited by the latency of interrupt processing.
4. The middle of a start bit is detected as the 8th sampled 0 after the 1-to-0 transition of the start bit. Subsequent bit values are sampled each 16th sample.

8 Ethernet: Enhanced Three-Speed Ethernet (eTSEC)

This section provides the AC and DC electrical characteristics for the enhanced three-speed Ethernet controller.

8.1 Enhanced Three-Speed Ethernet Controller (eTSEC) (10/100/1000 Mbps)—FIFO/GMII/MII/TBI/RGMII/RTBI/RMII Electrical Characteristics

The electrical characteristics specified here apply to all FIFO mode, gigabit media independent interface (GMII), media independent interface (MII), ten-bit interface (TBI), reduced gigabit media independent interface (RGMII), reduced ten-bit interface (RTBI), and reduced media independent interface (RMII) signals except management data input/output (MDIO) and management data clock (MDC), and serial gigabit media independent interface (SGMII). The RGMII, RTBI and FIFO mode interfaces are defined for 2.5 V, while the GMII, MII, RMII, and TBI interfaces can operate at both 2.5 V and 3.3V.

The GMII, MII, or TBI interface timing is compliant with IEEE 802.3. The RGMII and RTBI interfaces follow the Reduced Gigabit Media-Independent Interface (RGMII) Specification Version 1.3 (12/10/2000). The RMII interface follows the RMII Consortium RMII Specification Version 1.2 (3/20/1998).

The electrical characteristics for MDIO and MDC are specified in [Section 9, “Ethernet Management Interface Electrical Characteristics.”](#)

The electrical characteristics for SGMII is specified in [Section 8.3, “SGMII Interface Electrical Characteristics.”](#) The SGMII interface conforms (with exceptions) to the Serial-GMII Specification Version 1.8.

Figure 11 shows the GMII receive AC timing diagram.

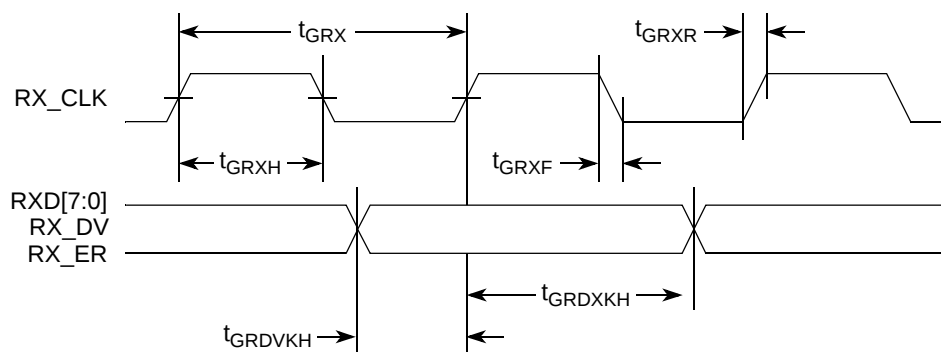


Figure 11. GMII Receive AC Timing Diagram

8.2.3 MII AC Timing Specifications

This section describes the MII transmit and receive AC timing specifications.

8.2.3.1 MII Transmit AC Timing Specifications

Table 29 provides the MII transmit AC timing specifications.

Table 29. MII Transmit AC Timing Specifications

At recommended operating conditions with V_{DD}/V_{DD} of 2.5/ 3.3 V $\pm$ 5%.

Parameter/Condition	Symbol ¹	Min	Typ	Max	Unit
TX_CLK clock period 10 Mbps	t_{MTX}^2	—	400	—	ns
TX_CLK clock period 100 Mbps	t_{MTX}	—	40	—	ns
TX_CLK duty cycle	t_{MTXH}/t_{MTX}	35	—	65	%
TX_CLK to MII data TXD[3:0], TX_ER, TX_EN delay	t_{MTKHDx}	1	5	15	ns
TX_CLK data clock rise (20%-80%)	t_{MTXR}^2	1.0	—	4.0	ns
TX_CLK data clock fall (80%-20%)	t_{MTXF}^2	1.0	—	4.0	ns

Notes:

- The symbols used for timing specifications herein follow the pattern of $t_{(first\ two\ letters\ of\ functional\ block)(signal)(state)\ (reference)(state)}$ for inputs and $t_{(first\ two\ letters\ of\ functional\ block)(reference)(state)(signal)(state)}$ for outputs. For example, t_{MTKHDx} symbolizes MII transmit timing (MT) for the time t_{MTX} clock reference (K) going high (H) until data outputs (D) are invalid (X). Note that, in general, the clock reference symbol representation is based on two to three letters representing the clock of a particular functional. For example, the subscript of t_{MTX} represents the MII(M) transmit (TX) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).
- Guaranteed by design.

Table 36. RMII Receive AC Timing Specifications (continued)

At recommended operating conditions with LV_{DD}/TV_{DD} of 2.5/ 3.3 V $\pm$ 5%.

Parameter/Condition	Symbol ¹	Min	Typ	Max	Unit
RXD[1:0], CRS_DV, RX_ER hold time to TSEc _n _TX_CLK rising edge	t_{RMRDX}	2.0	—	—	ns

Note:

- The symbols used for timing specifications herein follow the pattern of $t_{(first\ two\ letters\ of\ functional\ block)(signal)(state)\ (reference)(state)}$ for inputs and $t_{(first\ two\ letters\ of\ functional\ block)(reference)(state)(signal)(state)}$ for outputs. For example, t_{MRDVKH} symbolizes MII receive timing (MR) with respect to the time data input signals (D) reach the valid state (V) relative to the t_{MRX} clock reference (K) going to the high (H) state or setup time. Also, t_{MRDXKL} symbolizes MII receive timing (GR) with respect to the time data input signals (D) went invalid (X) relative to the t_{MRX} clock reference (K) going to the low (L) state or hold time. Note that, in general, the clock reference symbol representation is based on three letters representing the clock of a particular functional. For example, the subscript of t_{MRX} represents the MII (M) receive (RX) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).

Figure 20 provides the AC test load for eTSEC.

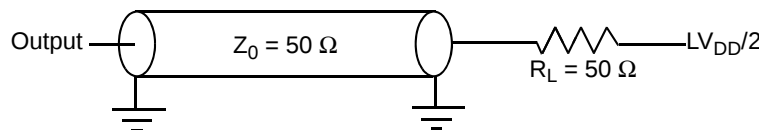


Figure 20. eTSEC AC Test Load

Figure 21 shows the RMII receive AC timing diagram.

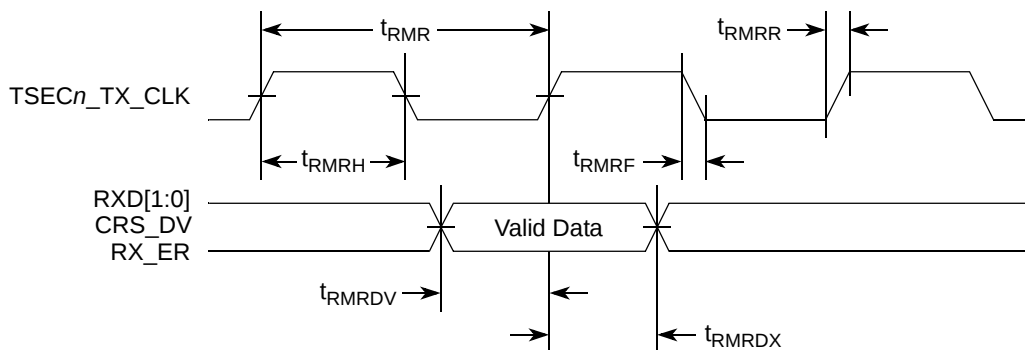


Figure 21. RMII Receive AC Timing Diagram

8.3 SGMII Interface Electrical Characteristics

Each SGMII port features a 4-wire AC-Coupled serial link from the dedicated SerDes 2 interface of MPC8572E as shown in Figure 22, where C_{TX} is the external (on board) AC-Coupled capacitor. Each output pin of the SerDes transmitter differential pair features 50- Ω output impedance. Each input of the SerDes receiver differential pair features 50- Ω on-die termination to SGND_SRDS2 (xcvss). The reference circuit of the SerDes transmitter and receiver is shown in Figure 54.

When an eTSEC port is configured to operate in SGMII mode, the parallel interface's output signals of this eTSEC port can be left floating. The input signals should be terminated based on the guidelines

8.3.4 SGMII AC Timing Specifications

This section describes the SGMII transmit and receive AC timing specifications. Transmitter and receiver characteristics are measured at the transmitter outputs (SD2_TX[n] and $\overline{\text{SD2_TX}}[n]$) or at the receiver inputs (SD2_RX[n] and $\overline{\text{SD2_RX}}[n]$) as depicted in Figure 25, respectively.

8.3.4.1 SGMII Transmit AC Timing Specifications

Table 40 provides the SGMII transmit AC timing targets. A source synchronous clock is not provided.

Table 40. SGMII Transmit AC Timing Specifications

At recommended operating conditions with $XV_{DD_SRDS2} = 1.1V \pm 5\%$.

Parameter	Symbol	Min	Typ	Max	Unit	Notes
Deterministic Jitter	JD	—	—	0.17	UI p-p	—
Total Jitter	JT	—	—	0.35	UI p-p	—
Unit Interval	UI	799.92	800	800.08	ps	1
V _{OD} fall time (80%-20%)	t _{fall}	50	—	120	ps	—
V _{OD} rise time (20%-80%)	t _{rise}	50	—	120	ps	—

Notes:

- Each UI is 800 ps $\pm$ 100 ppm.

8.3.4.2 SGMII Receive AC Timing Specifications

Table 41 provides the SGMII receive AC timing specifications. Source synchronous clocking is not supported. Clock is recovered from the data. Figure 24 shows the SGMII receiver input compliance mask eye diagram.

Table 41. SGMII Receive AC Timing Specifications

At recommended operating conditions with $XV_{DD_SRDS2} = 1.1V \pm 5\%$.

Parameter	Symbol	Min	Typ	Max	Unit	Notes
Deterministic Jitter Tolerance	JD	0.37	—	—	UI p-p	1
Combined Deterministic and Random Jitter Tolerance	JDR	0.55	—	—	UI p-p	1
Sinusoidal Jitter Tolerance	JSIN	0.1	—	—	UI p-p	1
Total Jitter Tolerance	JT	0.65	—	—	UI p-p	1
Bit Error Ratio	BER	—	—	10 ⁻¹²	—	—
Unit Interval	UI	799.92	800	800.08	ps	2
AC Coupling Capacitor	C _{TX}	5	—	200	nF	3

Notes:

- Measured at receiver.
- Each UI is 800 ps $\pm$ 100 ppm.
- The external AC coupling capacitor is required. It is recommended to be placed near the device transmitter outputs.
- See *RapidIO 1x/4x LP Serial Physical Layer Specification* for interpretation of jitter specifications.

8.4 eTSEC IEEE Std 1588™ AC Specifications

Figure 26 shows the data and command output timing diagram.

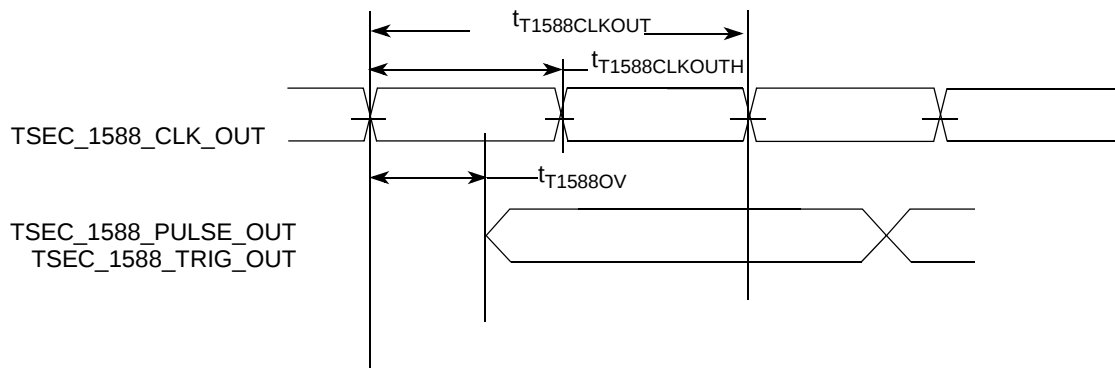


Figure 26. eTSEC IEEE 1588 Output AC Timing

¹ The output delay is count starting rising edge if $t_{T1588CLKOUT}$ is non-inverting. Otherwise, it is count starting falling edge.

Figure 27 shows the data and command input timing diagram.

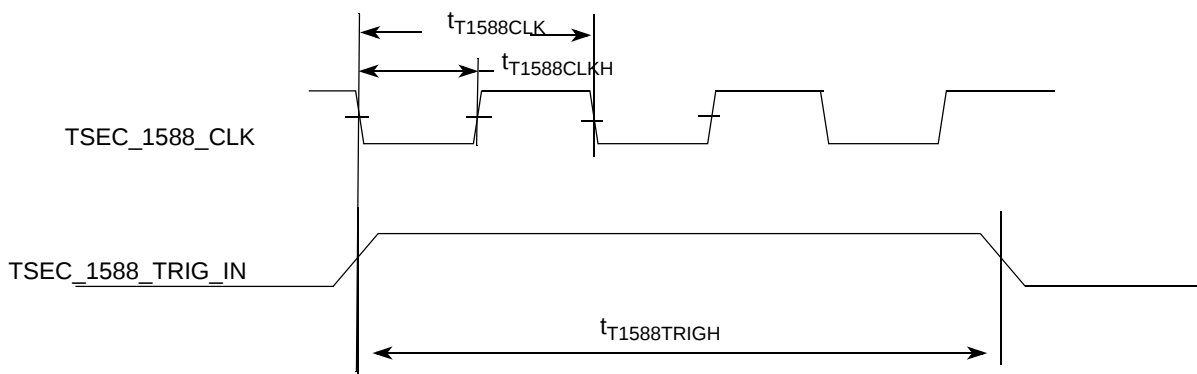


Figure 27. eTSEC IEEE 1588 Input AC timing

Table 42 provides the IEEE 1588 AC timing specifications.

Table 42. eTSEC IEEE 1588 AC Timing Specifications

At recommended operating conditions with LV_{DD}/TV_{DD} of $3.3\text{ V} \pm 5\%$ or $2.5\text{ V} \pm 5\%$

Parameter/Condition	Symbol	Min	Typ	Max	Unit	Note
TSEC_1588_CLK clock period	$t_{T1588CLK}$	3.3	—	$T_{TX_CLK}^{*9}$	ns	1
TSEC_1588_CLK duty cycle	$t_{T1588CLKH} / t_{T1588CLK}$	40	50	60	%	—
TSEC_1588_CLK peak-to-peak jitter	$t_{T1588CLKINJ}$	—	—	250	ps	—
Rise time eTSEC_1588_CLK (20%–80%)	$t_{T1588CLKINR}$	1.0	—	2.0	ns	—
Fall time eTSEC_1588_CLK (80%–20%)	$t_{T1588CLKINF}$	1.0	—	2.0	ns	—
TSEC_1588_CLK_OUT clock period	$t_{T1588CLKOUT}$	$2 \cdot t_{T1588CLK}$	—	—	ns	—

Table 43. MII Management DC Electrical Characteristics (LV_{DD}/TV_{DD}=3.3 V) (continued)

Parameter	Symbol	Min	Max	Unit	Notes
Input low current (LV _{DD} /TV _{DD} = Max, V _{IN} = 0.5 V)	I _{IL}	−600	—	μA	—

Note:

1. EC1_MDC and EC1_MDIO operate on LV_{DD}.
2. EC3_MDC & EC3_MDIO and EC5_MDC & EC5_MDIO operate on TV_{DD}.
3. Note that the symbol V_{IN}, in this case, represents the LV_{IN} and TV_{IN} symbol referenced in [Table 1](#).

Table 44. MII Management DC Electrical Characteristics (LV_{DD}/TV_{DD}=2.5 V)

Parameters	Symbol	Min	Max	Unit	Notes
Supply voltage 2.5 V	LV _{DD} /TV _{DD}	2.37	2.63	V	1,2
Output high voltage (LV _{DD} /TV _{DD} = Min, I _{OH} = −1.0 mA)	V _{OH}	2.00	LV _{DD} /TV _{DD} + 0.3	V	—
Output low voltage (LV _{DD} /TV _{DD} = Min, I _{OL} = 1.0 mA)	V _{OL}	GND − 0.3	0.40	V	—
Input high voltage	V _{IH}	1.70	LV _{DD} /TV _{DD} + 0.3	V	—
Input low voltage	V _{IL}	−0.3	0.70	V	—
Input high current (V _{IN} = LV _{DD} , V _{IN} = TV _{DD})	I _{IH}	—	10	μA	1, 2,3
Input low current (V _{IN} = GND)	I _{IL}	−15	—	μA	3

Note:

1. EC1_MDC and EC1_MDIO operate on LV_{DD}.
2. EC3_MDC & EC3_MDIO and EC5_MDC & EC5_MDIO operate on TV_{DD}.
3. Note that the symbol V_{IN}, in this case, represents the LV_{IN} and TV_{IN} symbols referenced in [Table 1](#).

9.2 MII Management AC Electrical Specifications

[Table 45](#) provides the MII management AC timing specifications. There are three sets of Ethernet management signals (EC1_MDC and EC1_MDIO, EC3_MDC and EC3_MDIO, EC5_MDC and EC5_MDIO). These are not explicitly shown in the table or in the figure following.

Table 45. MII Management AC Timing Specifications

At recommended operating conditions with LV_{DD}/TV_{DD} of 3.3 V ± 5% or 2.5 V ± 5%.

Parameter/Condition	Symbol ¹	Min	Typ	Max	Unit	Notes
ECn_MDC frequency	f _{MDC}	0.9	2.5	9.3	MHz	2, 3
ECn_MDC period	t _{MDC}	107.5	—	1120	ns	—
ECn_MDC clock pulse width high	t _{MDCH}	32	—	—	ns	—
ECn_MDC to ECn_MDIO delay	t _{MDKHDX}	10	—	16*t _{plb_clk}	ns	5

Table 45. MII Management AC Timing Specifications (continued)

At recommended operating conditions with V_{DD}/TV_{DD} of 3.3 V $\pm$ 5% or 2.5 V $\pm$ 5%.

Parameter/Condition	Symbol ¹	Min	Typ	Max	Unit	Notes
ECn_MDIO to ECn_MDC setup time	t_{MDDVKH}	5	—	—	ns	—
ECn_MDIO to ECn_MDC hold time	t_{MDDXKH}	0	—	—	ns	—
ECn_MDC rise time	t_{MDCR}	—	—	10	ns	4
ECn_MDC fall time	t_{MDHF}	—	—	10	ns	4

Notes:

1. The symbols used for timing specifications herein follow the pattern of $t_{(first\ two\ letters\ of\ functional\ block)(signal)(state)}$ (reference)(state) for inputs and $t_{(first\ two\ letters\ of\ functional\ block)(reference)(state)(signal)(state)}$ for outputs. For example, t_{MDKHDX} symbolizes management data timing (MD) for the time t_{MDC} from clock reference (K) high (H) until data outputs (D) are invalid (X) or data hold time. Also, t_{MDDVKH} symbolizes management data timing (MD) with respect to the time data input signals (D) reach the valid state (V) relative to the t_{MDC} clock reference (K) going to the high (H) state or setup time. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).
2. This parameter is dependent on the eTSEC system clock speed, which is half of the Platform Frequency (f_{CCB}). The actual ECn_MDC output clock frequency for a specific eTSEC port can be programmed by configuring the MgmtClk bit field of MPC8572E's MIIMCFG register, based on the platform (CCB) clock running for the device. The formula is: Platform Frequency (CCB)/(2*Frequency Divider determined by MIIMCFG[MgmtClk] encoding selection). For example, if MIIMCFG[MgmtClk] = 000 and the platform (CCB) is currently running at 533 MHz, $f_{MDC} = 533/(2*4*8) = 533/64 = 8.3$ MHz. That is, for a system running at a particular platform frequency (f_{CCB}), the ECn_MDC output clock frequency can be programmed between maximum $f_{MDC} = f_{CCB}/64$ and minimum $f_{MDC} = f_{CCB}/448$. Refer to MPC8572E reference manual's MIIMCFG register section for more detail.
3. The maximum ECn_MDC output clock frequency is defined based on the maximum platform frequency for MPC8572E (600 MHz) divided by 64, while the minimum ECn_MDC output clock frequency is defined based on the minimum platform frequency for MPC8572E (400 MHz) divided by 448, following the formula described in Note 2 above. The typical ECn_MDC output clock frequency of 2.5 MHz is shown for reference purpose per IEEE 802.3 specification.
4. Guaranteed by design.
5. t_{plb_clk} is the platform (CCB) clock.

Figure 28 shows the MII management AC timing diagram.

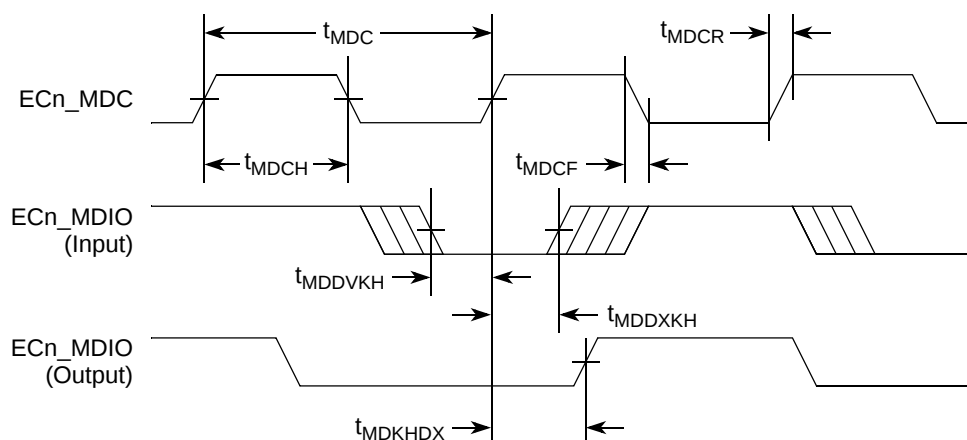

Figure 28. MII Management Interface Timing Diagram

Figure 30 through Figure 35 show the local bus signals.

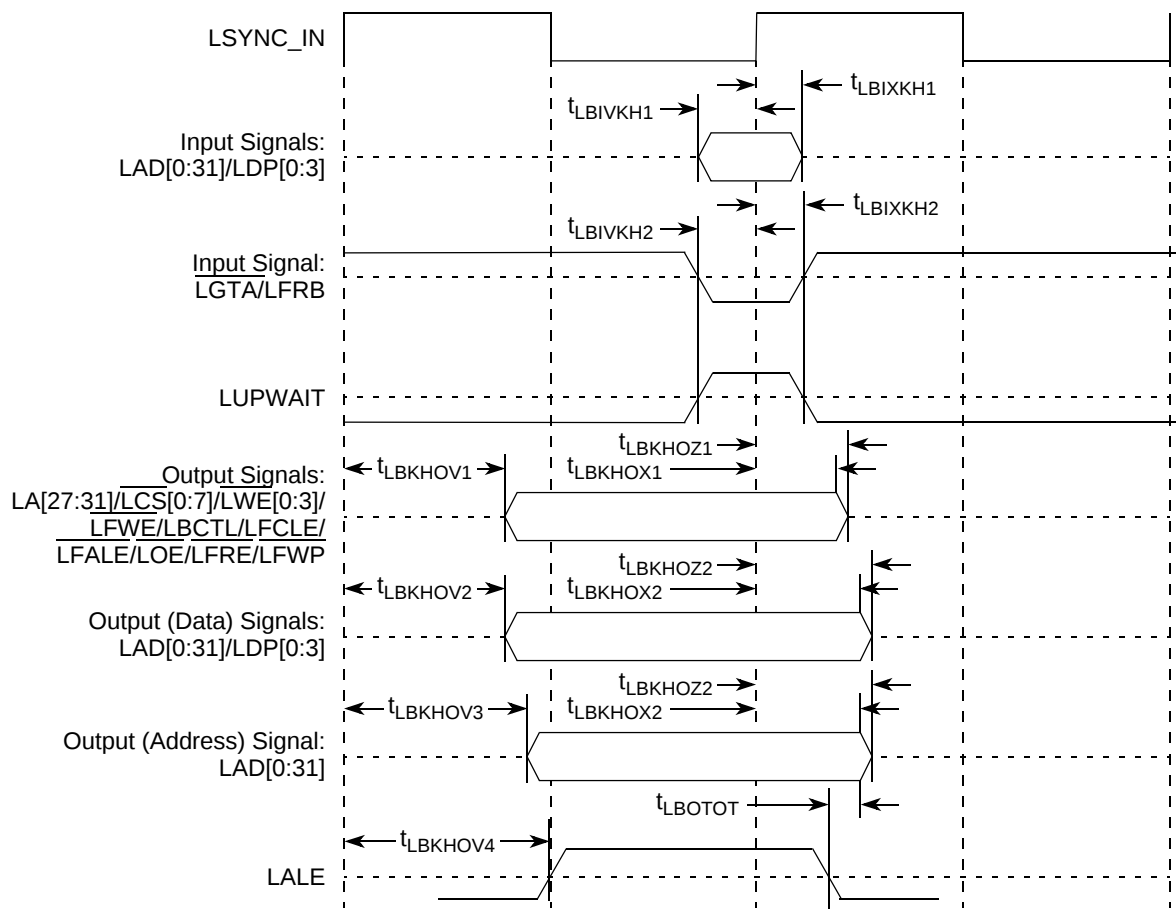


Figure 30. Local Bus Signals, Non-Special Signals Only (PLL Enabled)

Table 52 describes the general timing parameters of the local bus interface at $BV_{DD} = 3.3 \text{ V DC}$ with PLL disabled.

Table 52. Local Bus General Timing Parameters—PLL Bypassed

At recommended operating conditions with BV_{DD} of $3.3 \text{ V} \pm 5\%$

Parameter	Symbol ¹	Min	Max	Unit	Notes
Local bus cycle time	t_{LBK}	12	—	ns	2
Local bus duty cycle	t_{LBKH}/t_{LBK}	43	57	%	—
Internal launch/capture clock to LCLK delay	t_{LBKHK1}	2.3	4.0	ns	—
Input setup to local bus clock (except LGTA/LUPWAIT)	$t_{LBIVKH1}$	5.8	—	ns	4, 5
LGTA/LUPWAIT input setup to local bus clock	$t_{LBIVKL2}$	5.7	—	ns	4, 5
Input hold from local bus clock (except LGTA/LUPWAIT)	$t_{LBIXKH1}$	-1.3	—	ns	4, 5

11 Programmable Interrupt Controller

In IRQ edge trigger mode, when an external interrupt signal is asserted (according to the programmed polarity), it must remain asserted for at least 3 system clocks (SYSCLK periods).

12 JTAG

This section describes the AC electrical specifications for the IEEE 1149.1 (JTAG) interface of the MPC8572E.

Table 53 provides the JTAG AC timing specifications as defined in Figure 37 through Figure 39.

Table 53. JTAG AC Timing Specifications (Independent of SYSCLK) ¹

At recommended operating conditions with OV_{DD} of 3.3 V $\pm$ 5%.

Parameter	Symbol ²	Min	Max	Unit	Notes
JTAG external clock frequency of operation	f_{JTG}	0	33.3	MHz	—
JTAG external clock cycle time	t_{JTG}	30	—	ns	—
JTAG external clock pulse width measured at 1.4 V	t_{JTKHKL}	15	—	ns	—
JTAG external clock rise and fall times	t_{JTGR} & t_{JTGF}	0	2	ns	6
$\overline{TRST}$ assert time	t_{TRST}	25	—	ns	3
Input setup times:				ns	
Boundary-scan data TMS, TDI	t_{JTDVKH} t_{JTIVKH}	4 0	— —		4
Input hold times:				ns	
Boundary-scan data TMS, TDI	t_{JTDXKH} t_{JTIXKH}	20 25	— —		4
Valid times:				ns	
Boundary-scan data TDO	t_{JTKLDV} t_{JTKLOV}	4 4	20 25		5
Output hold times:				ns	
Boundary-scan data TDO	t_{JTKLDX} t_{JTKLOX}	30 30	— —		5

Table 62. Differential Transmitter (TX) Output Specifications (continued)

Symbol	Parameter	Min	Nominal	Max	Units	Comments
$V_{TX-DC-CM}$	The TX DC Common Mode Voltage	0	—	3.6	V	The allowed DC Common Mode voltage under any conditions. See Note 6.
$I_{TX-SHORT}$	TX Short Circuit Current Limit		—	90	mA	The total current the Transmitter can provide when shorted to its ground
$T_{TX-IDLE-MIN}$	Minimum time spent in Electrical Idle	50	—	—	UI	Minimum time a Transmitter must be in Electrical Idle Utilized by the Receiver to start looking for an Electrical Idle Exit after successfully receiving an Electrical Idle ordered set
$T_{TX-IDLE-SET-TO-IDLE}$	Maximum time to transition to a valid Electrical idle after sending an Electrical Idle ordered set	—	—	20	UI	After sending an Electrical Idle ordered set, the Transmitter must meet all Electrical Idle Specifications within this time. This is considered a debounce time for the Transmitter to meet Electrical Idle after transitioning from L0.
$T_{TX-IDLE-TO-DIFF-DATA}$	Maximum time to transition to valid TX specifications after leaving an Electrical idle condition	—	—	20	UI	Maximum time to meet all TX specifications when transitioning from Electrical Idle to sending differential data. This is considered a debounce time for the TX to meet all TX specifications after leaving Electrical Idle
$RL_{TX-DIFF}$	Differential Return Loss	12	—	—	dB	Measured over 50 MHz to 1.25 GHz. See Note 4
RL_{TX-CM}	Common Mode Return Loss	6	—	—	dB	Measured over 50 MHz to 1.25 GHz. See Note 4
$Z_{TX-DIFF-DC}$	DC Differential TX Impedance	80	100	120	Ω	TX DC Differential mode Low Impedance
Z_{TX-DC}	Transmitter DC Impedance	40	—	—	Ω	Required TX D+ as well as D- DC Impedance during all states
$L_{TX-SKEW}$	Lane-to-Lane Output Skew	—	—	500 + 2 UI	ps	Static skew between any two Transmitter Lanes within a single Link
C_{TX}	AC Coupling Capacitor	75	—	200	nF	All Transmitters shall be AC coupled. The AC coupling is required either within the media or within the transmitting component itself. See Note 8.

16.5.1 Compliance Test and Measurement Load

The AC timing and voltage parameters must be verified at the measurement point, as specified within 0.2 inches of the package pins, into a test/measurement load shown in Figure 57.

NOTE

The allowance of the measurement point to be within 0.2 inches of the package pins is meant to acknowledge that package/board routing may benefit from D+ and D– not being exactly matched in length at the package pin boundary.

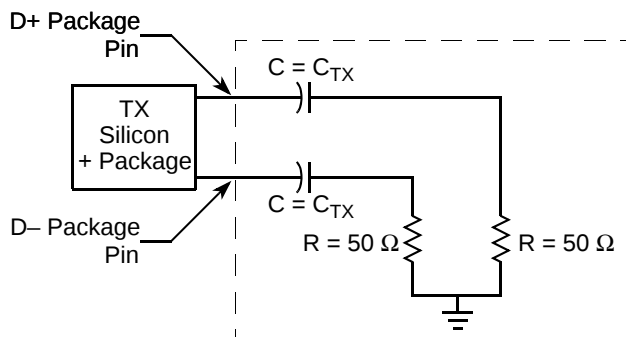


Figure 57. Compliance Test/Measurement Load

17 Serial RapidIO

This section describes the DC and AC electrical specifications for the RapidIO interface of the MPC8572E for the LP-Serial physical layer. The electrical specifications cover both single and multiple-lane links. Two transmitters (short run and long run) and a single receiver are specified for each of three baud rates, 1.25, 2.50, and 3.125 GBaud.

Two transmitter specifications allow for solutions ranging from simple board-to-board interconnect to driving two connectors across a backplane. A single receiver specification is given that accepts signals from both the short run and long run transmitter specifications.

The short run transmitter should be used mainly for chip-to-chip connections on either the same printed circuit board or across a single connector. This covers the case where connections are made to a mezzanine (daughter) card. The minimum swings of the short run specification reduce the overall power used by the transceivers.

The long run transmitter specifications use larger voltage swings that are capable of driving signals across backplanes. This allows a user to drive signals across two connectors and a backplane. The specifications allow a distance of at least 50 cm at all baud rates.

All unit intervals are specified with a tolerance of ± 100 ppm. The worst case frequency difference between any transmit and receive clock is 200 ppm.

To ensure interoperability between drivers and receivers of different vendors and technologies, AC coupling at the receiver input must be used.

Table 69. Long Run Transmitter AC Timing Specifications—2.5 GBaud

Characteristic	Symbol	Range		Unit	Notes
		Min	Max		
Output Voltage,	V_O	-0.40	2.30	Volts	Voltage relative to COMMON of either signal comprising a differential pair
Differential Output Voltage	V_{DIFFPP}	800	1600	mV p-p	—
Deterministic Jitter	J_D	—	0.17	UI p-p	—
Total Jitter	J_T	—	0.35	UI p-p	—
Multiple output skew	S_{MO}	—	1000	ps	Skew at the transmitter output between lanes of a multilane link
Unit Interval	UI	400	400	ps	+/- 100 ppm

Table 70. Long Run Transmitter AC Timing Specifications—3.125 GBaud

Characteristic	Symbol	Range		Unit	Notes
		Min	Max		
Output Voltage,	V_O	-0.40	2.30	Volts	Voltage relative to COMMON of either signal comprising a differential pair
Differential Output Voltage	V_{DIFFPP}	800	1600	mV p-p	—
Deterministic Jitter	J_D	—	0.17	UI p-p	—
Total Jitter	J_T	—	0.35	UI p-p	—
Multiple output skew	S_{MO}	—	1000	ps	Skew at the transmitter output between lanes of a multilane link
Unit Interval	UI	320	320	ps	+/- 100 ppm

For each baud rate at which an LP-Serial transmitter is specified to operate, the output eye pattern of the transmitter shall fall entirely within the unshaded portion of the transmitter output compliance mask shown in [Figure 58](#) with the parameters specified in [Figure 71](#) when measured at the output pins of the device and the device is driving a $100\ \Omega$ +/-5% differential resistive load. The output eye pattern of an LP-Serial transmitter that implements pre-emphasis (to equalize the link and reduce inter-symbol interference) need only comply with the Transmitter Output Compliance Mask when pre-emphasis is disabled or minimized.

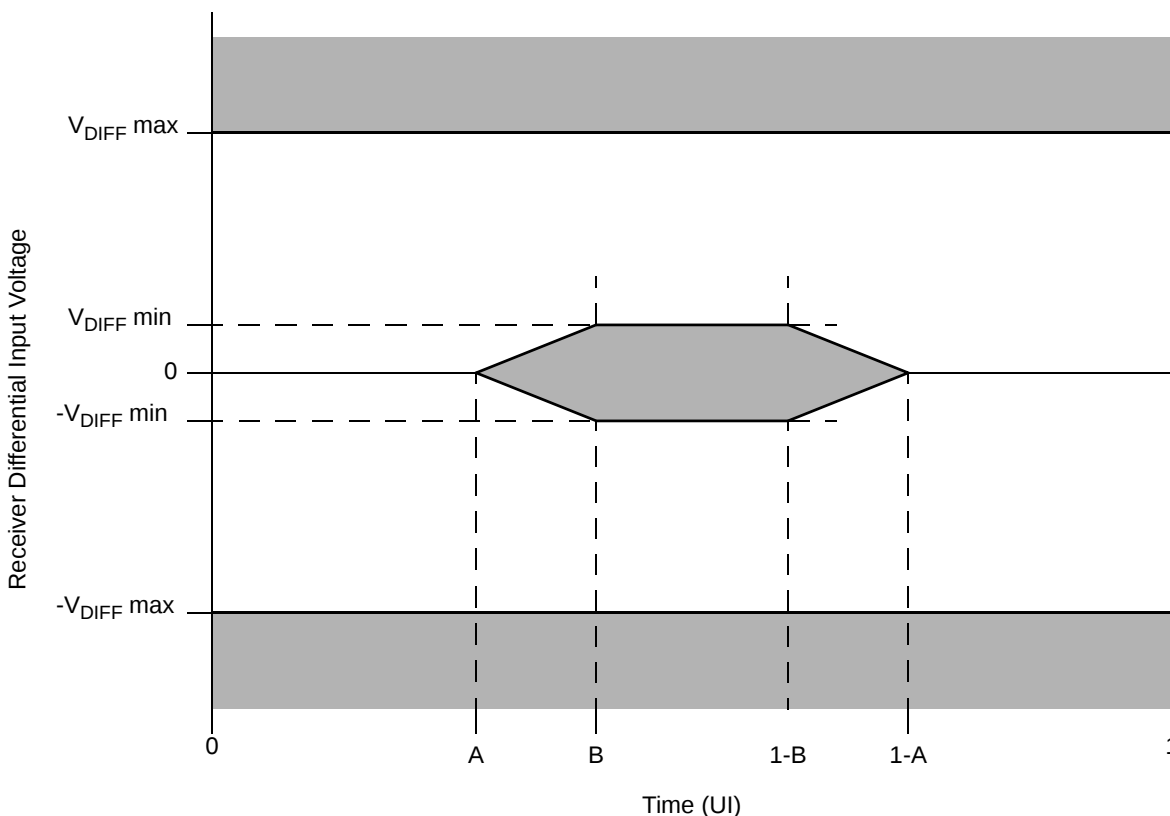


Figure 60. Receiver Input Compliance Mask

Table 75. Receiver Input Compliance Mask Parameters Exclusive of Sinusoidal Jitter

Receiver Type	$V_{DIFFmin}$ (mV)	$V_{DIFFmax}$ (mV)	A (UI)	B (UI)
1.25 GBaud	100	800	0.275	0.400
2.5 GBaud	100	800	0.275	0.400
3.125 GBaud	100	800	0.275	0.400

17.8 Measurement and Test Requirements

Because the LP-Serial electrical specification are guided by the XAUI electrical interface specified in Clause 47 of IEEE 802.3ae-2002, the measurement and test requirements defined here are similarly guided by Clause 47. Additionally, the CJPAT test pattern defined in Annex 48A of IEEE 802.3ae-2002 is specified as the test pattern for use in eye pattern and jitter measurements. Annex 48B of IEEE 802.3ae-2002 is recommended as a reference for additional information on jitter test methods.

17.8.1 Eye Template Measurements

For the purpose of eye template measurements, the effects of a single-pole high pass filter with a 3 dB point at (Baud Frequency)/1667 is applied to the jitter. The data pattern for template measurements is the Continuous Jitter Test Pattern (CJPAT) defined in Annex 48A of IEEE 802.3ae. All lanes of the LP-Serial

link shall be active in both the transmit and receive directions, and opposite ends of the links shall use asynchronous clocks. Four lane implementations shall use CJPAT as defined in Annex 48A. Single lane implementations shall use the CJPAT sequence specified in Annex 48A for transmission on lane 0. The amount of data represented in the eye shall be adequate to ensure that the bit error ratio is less than 10^{-12} . The eye pattern shall be measured with AC coupling and the compliance template centered at 0 Volts differential. The left and right edges of the template shall be aligned with the mean zero crossing points of the measured data eye. The load for this test shall be 100 Ω resistive $\pm$ 5% differential to 2.5 GHz.

17.8.2 Jitter Test Measurements

For the purpose of jitter measurement, the effects of a single-pole high pass filter with a 3 dB point at (Baud Frequency)/1667 is applied to the jitter. The data pattern for jitter measurements is the Continuous Jitter Test Pattern (CJPAT) pattern defined in Annex 48A of IEEE 802.3ae. All lanes of the LP-Serial link shall be active in both the transmit and receive directions, and opposite ends of the links shall use asynchronous clocks. Four lane implementations shall use CJPAT as defined in Annex 48A. Single lane implementations shall use the CJPAT sequence specified in Annex 48A for transmission on lane 0. Jitter shall be measured with AC coupling and at 0 Volts differential. Jitter measurement for the transmitter (or for calibration of a jitter tolerance setup) shall be performed with a test procedure resulting in a BER curve such as that described in Annex 48B of IEEE 802.3ae.

17.8.3 Transmit Jitter

Transmit jitter is measured at the driver output when terminated into a load of 100 Ω resistive $\pm$ 5% differential to 2.5 GHz.

17.8.4 Jitter Tolerance

Jitter tolerance is measured at the receiver using a jitter tolerance test signal. This signal is obtained by first producing the sum of deterministic and random jitter defined in [Section 17.6, “Receiver Specifications,”](#) and then adjusting the signal amplitude until the data eye contacts the 6 points of the minimum eye opening of the receive template shown in [Figure 60](#) and [Table 75](#). Note that for this to occur, the test signal must have vertical waveform symmetry about the average value and have horizontal symmetry (including jitter) about the mean zero crossing. Eye template measurement requirements are as defined above. Random jitter is calibrated using a high pass filter with a low frequency corner at 20 MHz and a 20 dB/decade roll-off below this. The required sinusoidal jitter specified in [Section 17.6, “Receiver Specifications,”](#) is then added to the signal and the test load is replaced by the receiver being tested.

18 Package Description

This section describes package parameters, pin assignments, and dimensions.

18.1 Package Parameters for the MPC8572E FC-PBGA

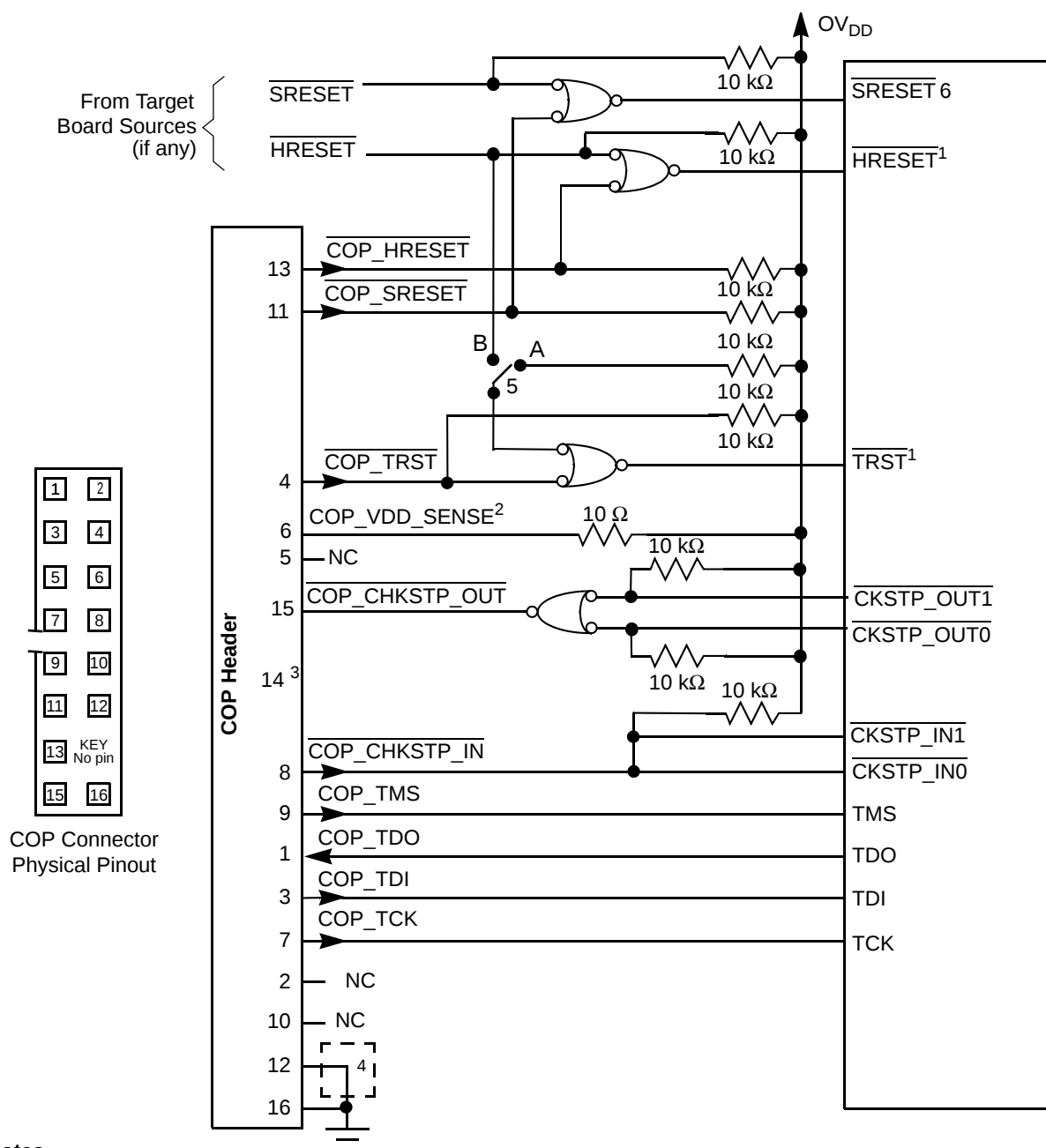
The package parameters are as provided in the following list. The package type is 33 mm × 33 mm, 1023 flip chip plastic ball grid array (FC-PBGA).

Package outline	33 mm × 33 mm
Interconnects	1023
Ball Pitch	1 mm
Ball Diameter (Typical)	0.6 mm
Solder Balls	63% Sn
	37% Pb
Solder Balls (Lead-Free)	96.5% Sn
	3.5% Ag

Table 76. MPC8572E Pinout Listing (continued)

Signal	Signal Name	Package Pin Number	Pin Type	Power Supply	Notes
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25. When operating in DDR2 mode, connect Dn_MDIC[0] to ground through 18.2-Ω (full-strength mode) or 36.4-Ω (half-strength mode) precision 1% resistor, and connect Dn_MDIC[1] to GVDD through 18.2-Ω (full-strength mode) or 36.4-Ω (half-strength mode) precision 1% resistor. When operating in DDR3 mode, connect Dn_MDIC[0] to ground through 20-Ω (full-strength mode) or 40-Ω (half-strength mode) precision 1% resistor, and connect Dn_MDIC[1] to GVDD through 20-Ω (full-strength mode) or 40-Ω (half-strength mode) precision 1% resistor. These pins are used for automatic calibration of the DDR IOs.
26. These pins should be connected to XVDD_SRDS1.
27. These pins should be pulled to ground (XGND_SRDS1) through a 300-Ω (±10%) resistor.
28. These pins should be left floating.
29. These pins should be pulled up to TVDD through a 2–10 KΩ resistor.
30. These pins have other manufacturing or debug test functions. It is recommended to add both pull-up resistor pads to OVDD and pull-down resistor pads to GND on board to support future debug testing when needed.
31. DDRCLK input is only required when the MPC8572E DDR controller is running in asynchronous mode. When the DDR controller is configured to run in synchronous mode via POR setting `cfg_ddr_pll[0:2]=111`, the DDRCLK input is not required. It is recommended to tie it off to GND when DDR controller is running in synchronous mode. See the *MPC8572E PowerQUICC™ III Integrated Host Processor Family Reference Manual* Rev.0, Table 4-3 in section 4.2.2 “Clock Signals”, section 4.4.3.2 “DDR PLL Ratio” and Table 4-10 “DDR Complex Clock PLL Ratio” for more detailed description regarding DDR controller operation in asynchronous and synchronous modes.
32. EC_GTX_CLK125 is a 125-MHz input clock shared among all eTSEC ports in the following modes: GMII, TBI, RGMII and RTBI. If none of the eTSEC ports is operating in these modes, the EC_GTX_CLK125 input can be tied off to GND.
33. These pins should be pulled to ground (GND).
34. These pins are sampled at POR for General Purpose configuration use by software. Their value has no impact on the functionality of the hardware.



Notes:

1. The COP port and target board should be able to independently assert $\overline{\text{HRESET}}$ and $\overline{\text{TRST}}$ to the processor to fully control the processor as shown here.
2. Populate this with a 10 Ω resistor for short-circuit/current-limiting protection.
3. The KEY location (pin 14) is not physically present on the COP header.
4. Although pin 12 is defined as a No-Connect, some debug tools may use pin 12 as an additional GND pin for improved signal integrity.
5. This switch is included as a precaution for BSDL testing. The switch should be closed to position A during BSDL testing to avoid accidentally asserting the TRST line. If BSDL testing is not being performed, this switch should be closed to position B.
6. Asserting $\overline{\text{SRESET}}$ causes a machine check interrupt to the e500 cores.