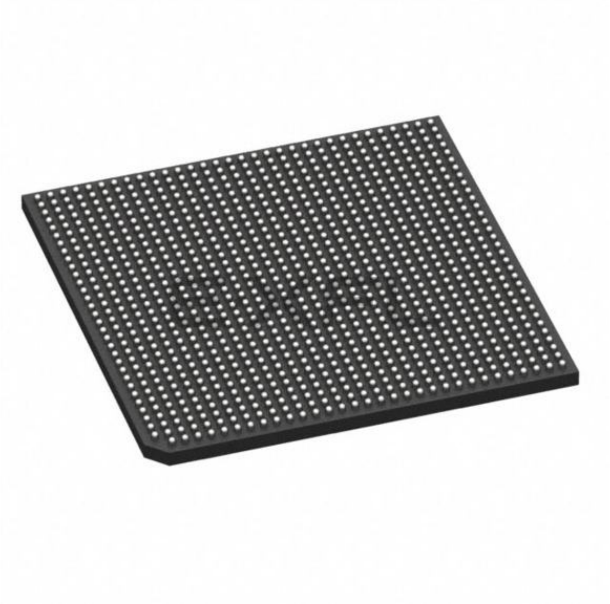

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Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Obsolete
Core Processor	PowerPC e500
Number of Cores/Bus Width	2 Core, 32-Bit
Speed	1.333GHz
Co-Processors/DSP	Signal Processing; SPE, Security; SEC
RAM Controllers	DDR2, DDR3
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100/1000Mbps (4)
SATA	-
USB	-
Voltage - I/O	1.5V, 1.8V, 2.5V, 3.3V
Operating Temperature	0°C ~ 105°C (TA)
Security Features	Cryptography, Random Number Generator
Package / Case	1023-BFBGA, FCBGA
Supplier Device Package	1023-FCBGA (33x33)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/ppc8572evtaulc

Figure 1 shows the MPC8572E block diagram.

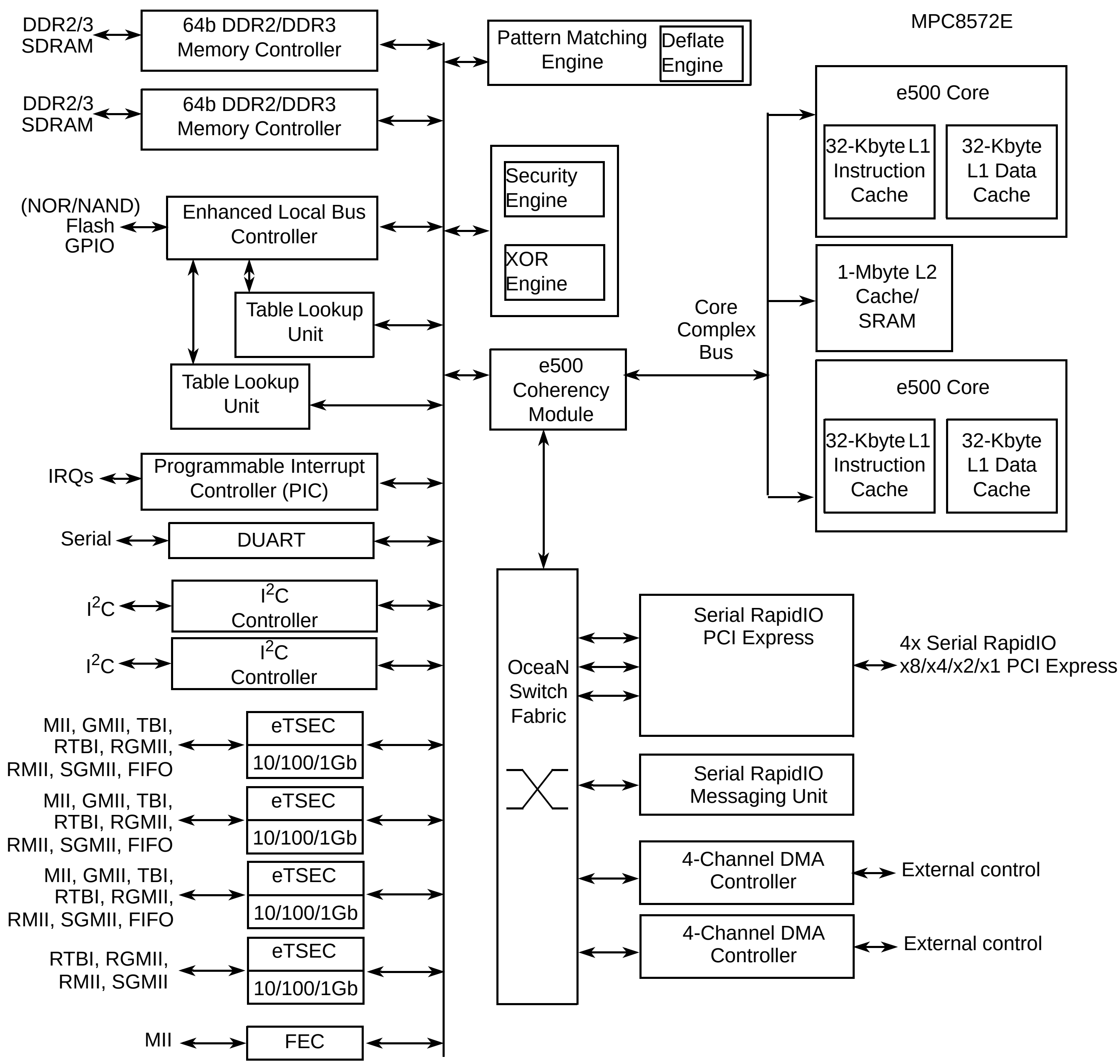


Figure 1. MPC8572E Block Diagram

2 Electrical Characteristics

This section provides the AC and DC electrical specifications for the MPC8572E. The MPC8572E is currently targeted to these specifications. Some of these specifications are independent of the I/O cell, but are included for a more complete reference. These are not purely I/O buffer design specifications.

Table 11. DDR2 SDRAM Interface DC Electrical Characteristics for GV_{DD}(typ) = 1.8 V (continued)

Parameter/Condition	Symbol	Min	Max	Unit	Notes
Output low current (V_{OUT} = 0.280 V)	I_{OL}	13.4	—	mA	—

Notes:
1. GV_{DD} is expected to be within 50 mV of the DRAM GV_{DD} at all times.
2. $MV_{REF}n$ is expected to be equal to 0.5 × GV_{DD}, and to track GV_{DD} DC variations as measured at the receiver. Peak-to-peak noise on $MV_{REF}n$ may not exceed ±2% of the DC value.
3. V_{TT} is not applied directly to the device. It is the supply to that far end signal termination is made and is expected to be equal to $MV_{REF}n$. This rail should track variations in the DC level of $MV_{REF}n$.
4. Output leakage is measured with all outputs disabled, 0 V ≤ V_{OUT} ≤ GV_{DD}.

Table 12 provides the recommended operating conditions for the DDR SDRAM controller of the MPC8572E when interfacing to DDR3 SDRAM.

Table 12. DDR3 SDRAM Interface DC Electrical Characteristics for GV_{DD}(typ) = 1.5 V

Parameter/Condition	Symbol	Min	Typical	Max	Unit
I/O supply voltage	GV_{DD}	1.425	1.575	V	1
I/O reference voltage	$MV_{REF}n$	0.49 × GV_{DD}	0.51 × GV_{DD}	V	2
Input high voltage	V_{IH}	$MV_{REF}n$ + 0.100	GV_{DD}	V	—
Input low voltage	V_{IL}	GND	$MV_{REF}n$ – 0.100	V	—
Output leakage current	I_{OZ}	–50	50	μA	3

Notes:
1. GV_{DD} is expected to be within 50 mV of the DRAM GV_{DD} at all times.
2. $MV_{REF}n$ is expected to be equal to 0.5 × GV_{DD}, and to track GV_{DD} DC variations as measured at the receiver. Peak-to-peak noise on $MV_{REF}n$ may not exceed ±1% of the DC value.
3. Output leakage is measured with all outputs disabled, 0 V ≤ V_{OUT} ≤ GV_{DD}.

Table 13 provides the DDR SDRAM controller interface capacitance for DDR2 and DDR3.

Table 13. DDR2 and DDR3 SDRAM Interface Capacitance for GV_{DD}(typ)=1.8 V and 1.5 V

Parameter/Condition	Symbol	Min	Typical	Max	Unit
Input/output capacitance: DQ, DQS, $\overline{DQS}$	C_{IO}	6	8	pF	1, 2
Delta input/output capacitance: DQ, DQS, $\overline{DQS}$	C_{DIO}	—	0.5	pF	1, 2

Note:
1. This parameter is sampled. GV_{DD} = 1.8 V ± 0.090 V (for DDR2), f = 1 MHz, T_A = 25°C, V_{OUT} = GV_{DD}/2, V_{OUT} (peak-to-peak) = 0.2 V.
2. This parameter is sampled. GV_{DD} = 1.5 V ± 0.075 V (for DDR3), f = 1 MHz, T_A = 25°C, V_{OUT} = GV_{DD}/2, V_{OUT} (peak-to-peak) = 0.175 V.

Table 18. DDR2 and DDR3 SDRAM Interface Output AC Timing Specifications (continued)

At recommended operating conditions with GV_{DD} of 1.8 V ± 5% for DDR2 or 1.5 V ± 5% for DDR3.

Parameter	Symbol [1]	Min	Max	Unit	Notes
800 MHz		0.917	—		
667 MHz		1.10	—		
533 MHz		1.48	—		
400 MHz		1.95	—		
ADDR/CMD output hold with respect to MCK	t_{DDKHAX}			ns	3
800 MHz		0.917	—		
667 MHz		1.10	—		
533 MHz		1.48	—		
400 MHz		1.95	—		
$\overline{MCS}$[n] output setup with respect to MCK	t_{DDKHCS}			ns	3
800 MHz		0.917	—		
667 MHz		1.10	—		
533 MHz		1.48	—		
400 MHz	t_{DDKHCS}	1.95	—	ns	3
$\overline{MCS}$[n] output hold with respect to MCK	t_{DDKHCX}			ns	3
800 MHz		0.917	—		
667 MHz		1.10	—		
533 MHz		1.48	—		
400 MHz		1.95	—		
MCK to MDQS Skew	t_{DDKHMH}			ns	4
800 MHz		–0.375	0.375		
<= 667 MHz		–0.6	0.6		
MDQ/MECC/MDM output setup with respect to MDQS	t_{DDKHDS}, t_{DDKLDS}			ps	5
800 MHz		375	—		
667 MHz		450	—		
533 MHz		538	—		
400 MHz		700	—		
MDQ/MECC/MDM output hold with respect to MDQS	t_{DDKHDX}, t_{DDKLDX}			ps	5
800 MHz		375	—		
667 MHz		450	—		

Table 24. MII, GMII, RMII, RGMII, TBI, RTBI, and FIFO DC Electrical Characteristics (continued)

Parameters	Symbol	Min	Max	Unit	Notes
Input high current (V_{IN} = LV_{DD}, V_{IN} = TV_{DD})	I_{IH}	—	10	μA	1, 2,3
Input low current (V_{IN} = GND)	I_{IL}	–15	—	μA	3

Note:

1 LV_{DD} supports eTSECs 1 and 2.
2 TV_{DD} supports eTSECs 3 and 4 or FEC.
3 Note that the symbol V_{IN}, in this case, represents the LV_{IN} and TV_{IN} symbols referenced in Table 1.

8.2 FIFO, GMII, MII, TBI, RGMII, RMII, and RTBI AC Timing Specifications

The AC timing specifications for FIFO, GMII, MII, TBI, RGMII, RMII and RTBI are presented in this section.

8.2.1 FIFO AC Specifications

The basis for the AC specifications for the eTSEC's FIFO modes is the double data rate RGMII and RTBI specifications, because they have similar performance and are described in a source-synchronous fashion like FIFO modes. However, the FIFO interface provides deliberate skew between the transmitted data and source clock in GMII fashion.

When the eTSEC is configured for FIFO modes, all clocks are supplied from external sources to the relevant eTSEC interface. That is, the transmit clock must be applied to the eTSEC*n*'s TSEC*n*_TX_CLK, while the receive clock must be applied to pin TSEC*n*_RX_CLK. The eTSEC internally uses the transmit clock to synchronously generate transmit data and outputs an echoed copy of the transmit clock back on the TSEC*n*_GTX_CLK pin (while transmit data appears on TSEC*n*_TXD[7:0], for example). It is intended that external receivers capture eTSEC transmit data using the clock on TSEC*n*_GTX_CLK as a source-synchronous timing reference. Typically, the clock edge that launched the data can be used, because the clock is delayed by the eTSEC to allow acceptable set-up margin at the receiver. Note that there is a relationship between the maximum FIFO speed and the platform (CCB) frequency. For more information see Section 4.5, "Platform to eTSEC FIFO Restrictions."

Table 25 and Table 26 summarize the FIFO AC specifications.

Table 25. FIFO Mode Transmit AC Timing Specification

At recommended operating conditions with LV_{DD}/TV_{DD} of 2.5V ± 5%

Parameter/Condition	Symbol	Min	Typ	Max	Unit
TX_CLK, GTX_CLK clock period[1]	t_{FIT}	5.3	8.0	100	ns
TX_CLK, GTX_CLK duty cycle	t_{FITH}/t_{FIT}	45	50	55	%

Table 52. Local Bus General Timing Parameters—PLL Bypassed (continued)

At recommended operating conditions with BV_{DD} of 3.3 V ± 5%

Parameter	Symbol [1]	Min	Max	Unit	Notes
$\overline{\text{LGTA}}$/LUPWAIT input hold from local bus clock	$t_{LBIXKL2}$	-1.3	—	ns	4, 5
LALE output negation to high impedance for LAD/LDP (LATCH hold time)	t_{LBOTOT}	1.5	—	ns	6
Local bus clock to output valid (except LAD/LDP and LALE)	$t_{LBKLOV1}$	—	–0.3	ns	
Local bus clock to data valid for LAD/LDP	$t_{LBKLOV2}$	—	–0.1	ns	4
Local bus clock to address valid for LAD	$t_{LBKLOV3}$	—	0.0	ns	4
Local bus clock to LALE assertion	$t_{LBKLOV4}$	—	0.0	ns	4
Output hold from local bus clock (except LAD/LDP and LALE)	$t_{LBKLOX1}$	–3.3	—	ns	4
Output hold from local bus clock for LAD/LDP	$t_{LBKLOX2}$	–3.3	—	ns	4
Local bus clock to output high Impedance (except LAD/LDP and LALE)	$t_{LBKLOZ1}$	—	0.2	ns	7
Local bus clock to output high impedance for LAD/LDP	$t_{LBKLOZ2}$	—	0.2	ns	7

Notes:

1. The symbols used for timing specifications herein follow the pattern of $t_{\text{(First two letters of functional block)(signal)(state) (reference)(state)}}$ for inputs and $t_{\text{(First two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, $t_{LBIXKH1}$ symbolizes local bus timing (LB) for the input (I) to go invalid (X) with respect to the time the t_{LBK} clock reference (K) goes high (H), in this case for clock one(1). Also, t_{LBKHOX} symbolizes local bus timing (LB) for the t_{LBK} clock reference (K) to go high (H), with respect to the output (O) going invalid (X) or output hold time.
2. All timings are in reference to local bus clock for PLL bypass mode. Timings may be negative with respect to the local bus clock because the actual launch and capture of signals is done with the internal launch/capture clock, which precedes LCLK by t_{LBKHKT}.
3. Maximum possible clock skew between a clock LCLK[m] and a relative clock LCLK[n]. Skew measured between complementary signals at BV_{DD}/2.
4. All signals are measured from BVDD/2 of the rising edge of local bus clock for PLL bypass mode to 0.4 x BVDD of the signal in question for 3.3-V signaling levels.
5. Input timings are measured at the pin.
6. t_{LBOTOT} is a measurement of the minimum time between the negation of LALE and any change in LAD. t_{LBOTOT} is programmed with the LBCR[AHD] parameter.
7. For purposes of active/float timing measurements, the Hi-Z or off state is defined to be when the total current delivered through the component pin is less than or equal to the leakage current specification.

NOTE

In PLL bypass mode, LCLK[n] is the inverted version of the internal clock with the delay of t_{LBKHKT}. In this mode, signals are launched at the rising edge of the internal clock and are captured at the falling edge of the internal clock with the exception of $\overline{\text{LGTA}}$/LUPWAIT (which is captured on the rising edge of the internal clock).

6. Differential Waveform

1. The differential waveform is constructed by subtracting the inverting signal ($\overline{\mathrm{SD}n\_\mathrm{TX}}$, for example) from the non-inverting signal (SDn_TX, for example) within a differential pair. There is only one signal trace curve in a differential waveform. The voltage represented in the differential waveform is not referenced to ground. Refer to Figure 52 as an example for differential waveform.
2. Common Mode Voltage, V_{cm}

   The Common Mode Voltage is equal to one half of the sum of the voltages between each conductor of a balanced interchange circuit and ground. In this example, for SerDes output, $V_{cm\_out} = (V_{SDn\_TX} + V_{\overline{SDn\_TX}})/2 = (A + B) / 2$, which is the arithmetic mean of the two complimentary output voltages within a differential pair. In a system, the common mode voltage may often differ from one component's output to the other's input. Sometimes, it may be even different between the receiver input and driver output circuits within the same component. It is also referred as the DC offset in some occasion.

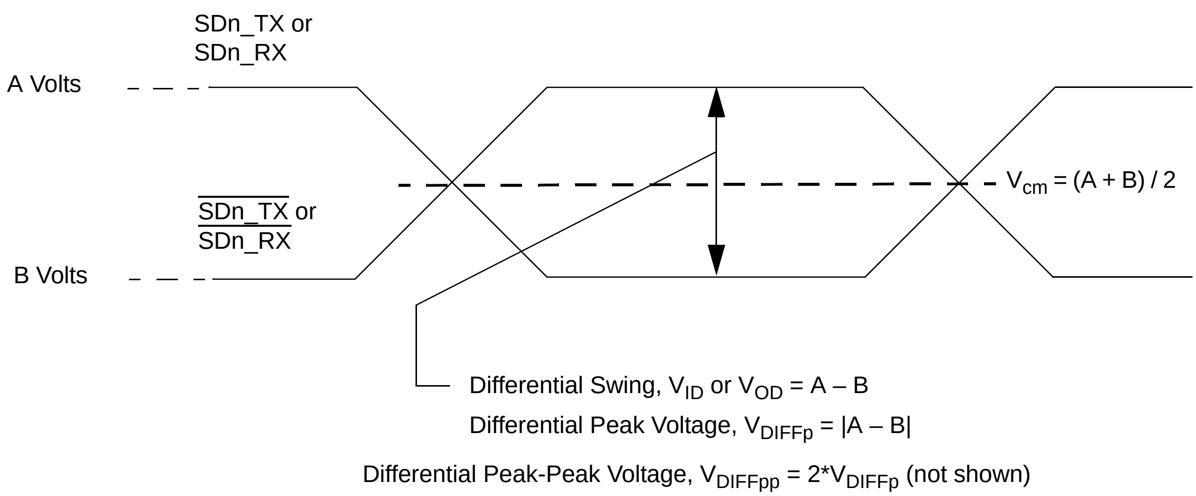


Figure 43. Differential Voltage Definitions for Transmitter or Receiver

To illustrate these definitions using real values, consider the case of a CML (Current Mode Logic) transmitter that has a common mode voltage of 2.25 V and each of its outputs, TD and $\overline{\mathrm{TD}}$, has a swing that goes between 2.5 V and 2.0 V. Using these values, the peak-to-peak voltage swing of each signal (TD or $\overline{\mathrm{TD}}$) is 500 mV p-p, which is referred as the single-ended swing for each signal. In this example, because the differential signaling environment is fully symmetrical, the transmitter output's differential swing (V_{OD}) has the same amplitude as each signal's single-ended swing. The differential output signal ranges between 500 mV and –500 mV, in other words, V_{OD} is 500 mV in one phase and –500 mV in the other phase. The peak differential voltage (V_{DIFFp}) is 500 mV. The peak-to-peak differential voltage ($V_{DIFFp\text{-}p}$) is 1000 mV p-p.

15.2 SerDes Reference Clocks

The SerDes reference clock inputs are applied to an internal PLL whose output creates the clock used by the corresponding SerDes lanes. The SerDes reference clocks inputs are SD1_REF_CLK and

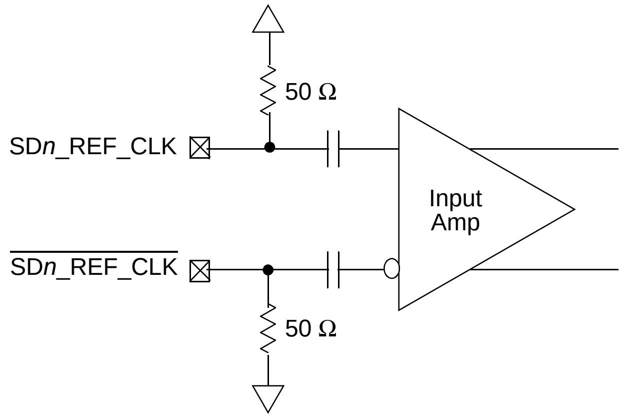


Figure 44. Receiver of SerDes Reference Clocks

15.2.2 DC Level Requirement for SerDes Reference Clocks

The DC level requirement for the MPC8572E SerDes reference clock inputs is different depending on the signaling mode used to connect the clock driver chip and SerDes reference clock inputs as described below.

- **Differential Mode**
  - The input amplitude of the differential clock must be between 400mV and 1600mV differential peak-peak (or between 200mV and 800mV differential peak). In other words, each signal wire of the differential pair must have a single-ended swing less than 800mV and greater than 200mV. This requirement is the same for both external DC-coupled or AC-coupled connection.
  - For **external DC-coupled** connection, as described in Section 15.2.1, "SerDes Reference Clock Receiver Characteristics," the maximum average current requirements sets the requirement for average voltage (common mode voltage) to be between 100 mV and 400 mV. Figure 45 shows the SerDes reference clock input requirement for DC-coupled connection scheme.
  - For **external AC-coupled** connection, there is no common mode voltage requirement for the clock driver. Because the external AC-coupling capacitor blocks the DC level, the clock driver and the SerDes reference clock receiver operate in different command mode voltages. The SerDes reference clock receiver in this connection scheme has its common mode voltage set to SGND_SRDS*n*. Each signal wire of the differential inputs is allowed to swing below and above the command mode voltage (SGND_SRDS*n*). Figure 46 shows the SerDes reference clock input requirement for AC-coupled connection scheme.
- **Single-ended Mode**
  - The reference clock can also be single-ended. The SD*n*_REF_CLK input amplitude (single-ended swing) must be between 400 mV and 800 mV peak-peak (from Vmin to Vmax) with $\overline{\text{SD}n\text{\_REF\_CLK}}$ either left unconnected or tied to ground.
  - The SD*n*_REF_CLK input average voltage must be between 200 and 400 mV. Figure 47 shows the SerDes reference clock input requirement for single-ended signaling mode.

Figure 49 shows the SerDes reference clock connection reference circuits for LVDS type clock driver. Because LVDS clock driver's common mode voltage is higher than the MPC8572E SerDes reference clock input's allowed range (100 to 400mV), AC-coupled connection scheme must be used. It assumes the LVDS output driver features 50-Ω termination resistor. It also assumes that the LVDS transmitter establishes its own common mode level without relying on the receiver or other external component.

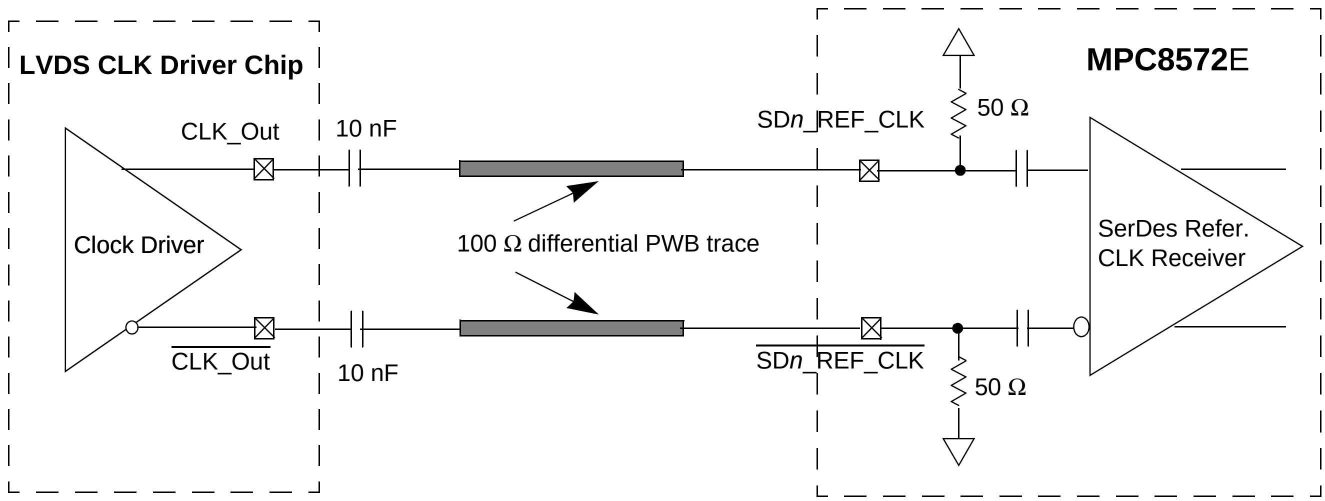


Figure 49. AC-Coupled Differential Connection with LVDS Clock Driver (Reference Only)

Figure 50 shows the SerDes reference clock connection reference circuits for LVPECL type clock driver. Because LVPECL driver's DC levels (both common mode voltages and output swing) are incompatible with MPC8572E SerDes reference clock input's DC requirement, AC-coupling must be used. Figure 50 assumes that the LVPECL clock driver's output impedance is 50Ω. R1 is used to DC-bias the LVPECL outputs prior to AC-coupling. Its value could be ranged from 140Ω to 240Ω depending on clock driver vendor's requirement. R2 is used together with the SerDes reference clock receiver's 50-Ω termination resistor to attenuate the LVPECL output's differential peak level such that it meets the MPC8572E SerDes reference clock's differential input amplitude requirement (between 200mV and 800mV differential peak). For example, if the LVPECL output's differential peak is 900mV and the desired SerDes reference clock input amplitude is selected as 600mV, the attenuation factor is 0.67, which requires R2 = 25Ω. Consult

- Section 17, “Serial RapidIO”

Note that external AC Coupling capacitor is required for the above three serial transmission protocols with the capacitor value defined in specification of each protocol section.

16 PCI Express

This section describes the DC and AC electrical specifications for the PCI Express bus of the MPC8572E.

16.1 DC Requirements for PCI Express SD1_REF_CLK and $\overline{\text{SD1\_REF\_CLK}}$

For more information, see Section 15.2, “SerDes Reference Clocks.”

16.2 AC Requirements for PCI Express SerDes Reference Clocks

Table 61 lists AC requirements.

Table 61. SD1_REF_CLK and $\overline{\text{SD1\_REF\_CLK}}$ AC Requirements

Symbol	Parameter Description	Min	Typical	Max	Units	Notes
t_{REF}	REFCLK cycle time	—	10	—	ns	1
t_{REFCJ}	REFCLK cycle-to-cycle jitter. Difference in the period of any two adjacent REFCLK cycles	—	—	100	ps	—
t_{REFPJ}	Phase jitter. Deviation in edge location with respect to mean edge location	–50	—	50	ps	—

Notes:
1. Typical cycle time is based on PCI Express Card Electromechanical Specification Revision 1.0a.

16.3 Clocking Dependencies

The ports on the two ends of a link must transmit data at a rate that is within 600 parts per million (ppm) of each other at all times. This is specified to allow bit rate clock sources with a +/– 300 ppm tolerance.

16.4 Physical Layer Specifications

The following is a summary of the specifications for the physical layer of PCI Express on this device. For further details as well as the specifications of the transport and data link layer, Use the PCI Express Base Specification. REV. 1.0a document.

16.4.1 Differential Transmitter (TX) Output

Table 62 defines the specifications for the differential output at all transmitters (TXs). The parameters are specified at the component pins.

Table 62. Differential Transmitter (TX) Output Specifications (continued)

Symbol	Parameter	Min	Nominal	Max	Units	Comments
$T_{crosslink}$	Crosslink Random Timeout	0	—	1	ms	This random timeout helps resolve conflicts in crosslink configuration by eventually resulting in only one Downstream and one Upstream Port. See Note 7.

Notes:

1. No test load is necessarily associated with this value.
2. Specified at the measurement point into a timing and voltage compliance test load as shown in Figure 57 and measured over any 250 consecutive TX UIs. (Also refer to the transmitter compliance eye diagram shown in Figure 55.)
3. A $T_{TX\text{-}EYE}$ = 0.70 UI provides for a total sum of deterministic and random jitter budget of $T_{TX\text{-}JITTER\text{-}MAX}$ = 0.30 UI for the Transmitter collected over any 250 consecutive TX UIs. The $T_{TX\text{-}EYE\text{-}MEDIAN\text{-}to\text{-}MAX\text{-}JITTER}$ median is less than half of the total TX jitter budget collected over any 250 consecutive TX UIs. It should be noted that the median is not the same as the mean. The jitter median describes the point in time where the number of jitter points on either side is approximately equal as opposed to the averaged time value.
4. The Transmitter input impedance shall result in a differential return loss greater than or equal to 12 dB and a common mode return loss greater than or equal to 6 dB over a frequency range of 50 MHz to 1.25 GHz. This input impedance requirement applies to all valid input levels. The reference impedance for return loss measurements is 50 ohms to ground for both the D+ and D- line (that is, as measured by a Vector Network Analyzer with 50 ohm probes—see Figure 57). Note that the series capacitors C_{TX} is optional for the return loss measurement.
5. Measured between 20-80% at transmitter package pins into a test load as shown in Figure 57 for both $V_{TX\text{-}D+}$ and $V_{TX\text{-}D-}$.
6. See Section 4.3.1.8 of the PCI Express Base Specifications Rev 1.0a.
7. See Section 4.2.6.3 of the PCI Express Base Specifications Rev 1.0a.
8. MPC8572E SerDes transmitter does not have C_{TX} built-in. An external AC Coupling capacitor is required.

16.4.2 Transmitter Compliance Eye Diagrams

The TX eye diagram in Figure 55 is specified using the passive compliance/test measurement load (see Figure 57) in place of any real PCI Express interconnect + RX component.

There are two eye diagrams that must be met for the transmitter. Both eye diagrams must be aligned in time using the jitter median to locate the center of the eye diagram. The different eye diagrams differ in voltage depending whether it is a transition bit or a de-emphasized bit. The exact reduced voltage level of the de-emphasized bit is always relative to the transition bit.

The eye diagram must be valid for any 250 consecutive UIs.

A recovered TX UI is calculated over 3500 consecutive unit intervals of sample data. The eye diagram is created using all edges of the 250 consecutive UI in the center of the 3500 UI used for calculating the TX UI.

NOTE

It is recommended that the recovered TX UI is calculated using all edges in the 3500 consecutive UI interval with a fit algorithm using a minimization merit function (that is, least squares and median deviation fits).

Table 63. Differential Receiver (RX) Input Specifications (continued)

Symbol	Parameter	Min	Nominal	Max	Units	Comments
$L_{RX-SKEW}$	Total Skew	—	—	20	ns	Skew across all lanes on a Link. This includes variation in the length of SKP ordered set (for example, COM and one to five SKP Symbols) at the RX as well as any delay differences arising from the interconnect itself.

Notes:
1. No test load is necessarily associated with this value.
2. Specified at the measurement point and measured over any 250 consecutive UIs. The test load in Figure 57 should be used as the RX device when taking measurements (also refer to the Receiver compliance eye diagram shown in Figure 56). If the clocks to the RX and TX are not derived from the same reference clock, the TX UI recovered from 3500 consecutive UI must be used as a reference for the eye diagram.
3. A T_{RX-EYE} = 0.40 UI provides for a total sum of 0.60 UI deterministic and random jitter budget for the Transmitter and interconnect collected any 250 consecutive UIs. The $T_{RX-EYE-MEDIAN-to-MAX-JITTER}$ specification ensures a jitter distribution in which the median and the maximum deviation from the median is less than half of the total. UI jitter budget collected over any 250 consecutive TX UIs. It should be noted that the median is not the same as the mean. The jitter median describes the point in time where the number of jitter points on either side is approximately equal as opposed to the averaged time value. If the clocks to the RX and TX are not derived from the same reference clock, the TX UI recovered from 3500 consecutive UI must be used as the reference for the eye diagram.
4. The Receiver input impedance shall result in a differential return loss greater than or equal to 15 dB with the D+ line biased to 300 mV and the D- line biased to -300 mV and a common mode return loss greater than or equal to 6 dB (no bias required) over a frequency range of 50 MHz to 1.25 GHz. This input impedance requirement applies to all valid input levels. The reference impedance for return loss measurements for is 50 ohms to ground for both the D+ and D- line (that is, as measured by a Vector Network Analyzer with 50 ohm probes - see Figure 57). Note: that the series capacitors CTX is optional for the return loss measurement.
5. Impedance during all LTSSM states. When transitioning from a Fundamental Reset to Detect (the initial state of the LTSSM) there is a 5 ms transition time before Receiver termination values must be met on all un-configured Lanes of a Port.
6. The RX DC Common Mode Impedance that exists when no power is present or Fundamental Reset is asserted. This helps ensure that the Receiver Detect circuit does not falsely assume a Receiver is powered on when it is not. This term must be measured at 300 mV above the RX ground.
7. It is recommended that the recovered TX UI is calculated using all edges in the 3500 consecutive UI interval with a fit algorithm using a minimization merit function. Least squares and median deviation fits have worked well with experimental and simulated data.

Table 66. Short Run Transmitter AC Timing Specifications—2.5 GBaud (continued)

Characteristic	Symbol	Range		Unit	Notes
		Min	Max		
Multiple Output skew	S_{MO}	—	1000	ps	Skew at the transmitter output between lanes of a multilane link
Unit Interval	UI	400	400	ps	+/- 100 ppm

Table 67. Short Run Transmitter AC Timing Specifications—3.125 GBaud

Characteristic	Symbol	Range		Unit	Notes
		Min	Max		
Output Voltage,	V_O	–0.40	2.30	Volts	Voltage relative to COMMON of either signal comprising a differential pair
Differential Output Voltage	V_{DIFFPP}	500	1000	mV p-p	—
Deterministic Jitter	J_D	—	0.17	UI p-p	—
Total Jitter	J_T	—	0.35	UI p-p	—
Multiple output skew	S_{MO}	—	1000	ps	Skew at the transmitter output between lanes of a multilane link
Unit Interval	UI	320	320	ps	+/– 100 ppm

Table 68. Long Run Transmitter AC Timing Specifications—1.25 GBaud

Characteristic	Symbol	Range		Unit	Notes
		Min	Max		
Output Voltage,	V_O	–0.40	2.30	Volts	Voltage relative to COMMON of either signal comprising a differential pair
Differential Output Voltage	V_{DIFFPP}	800	1600	mV p-p	—
Deterministic Jitter	J_D	—	0.17	UI p-p	—
Total Jitter	J_T	—	0.35	UI p-p	—
Multiple output skew	S_{MO}	—	1000	ps	Skew at the transmitter output between lanes of a multilane link
Unit Interval	UI	800	800	ps	+/- 100 ppm

Table 74. Receiver AC Timing Specifications—3.125 GBaud

Characteristic	Symbol	Range		Unit	Notes
		Min	Max		
Differential Input Voltage	V_{IN}	200	1600	mV p-p	Measured at receiver
Deterministic Jitter Tolerance	J_D	0.37	—	UI p-p	Measured at receiver
Combined Deterministic and Random Jitter Tolerance	J_{DR}	0.55	—	UI p-p	Measured at receiver
Total Jitter Tolerance[1]	J_T	0.65	—	UI p-p	Measured at receiver
Multiple Input Skew	S_{MI}	—	22	ns	Skew at the receiver input between lanes of a multilane link
Bit Error Rate	BER	—	10^{-12}	—	—
Unit Interval	UI	320	320	ps	+/- 100 ppm

Note:

1. Total jitter is composed of three components, deterministic jitter, random jitter and single frequency sinusoidal jitter. The sinusoidal jitter may have any amplitude and frequency in the unshaded region of Figure 59. The sinusoidal jitter component is included to ensure margin for low frequency jitter, wander, noise, crosstalk and other variable system effects.

link shall be active in both the transmit and receive directions, and opposite ends of the links shall use asynchronous clocks. Four lane implementations shall use CJPAT as defined in Annex 48A. Single lane implementations shall use the CJPAT sequence specified in Annex 48A for transmission on lane 0. The amount of data represented in the eye shall be adequate to ensure that the bit error ratio is less than 10^{-12}. The eye pattern shall be measured with AC coupling and the compliance template centered at 0 Volts differential. The left and right edges of the template shall be aligned with the mean zero crossing points of the measured data eye. The load for this test shall be 100 Ω resistive +/– 5% differential to 2.5 GHz.

17.8.2 Jitter Test Measurements

For the purpose of jitter measurement, the effects of a single-pole high pass filter with a 3 dB point at (Baud Frequency)/1667 is applied to the jitter. The data pattern for jitter measurements is the Continuous Jitter Test Pattern (CJPAT) pattern defined in Annex 48A of IEEE 802.3ae. All lanes of the LP-Serial link shall be active in both the transmit and receive directions, and opposite ends of the links shall use asynchronous clocks. Four lane implementations shall use CJPAT as defined in Annex 48A. Single lane implementations shall use the CJPAT sequence specified in Annex 48A for transmission on lane 0. Jitter shall be measured with AC coupling and at 0 Volts differential. Jitter measurement for the transmitter (or for calibration of a jitter tolerance setup) shall be performed with a test procedure resulting in a BER curve such as that described in Annex 48B of IEEE 802.3ae.

17.8.3 Transmit Jitter

Transmit jitter is measured at the driver output when terminated into a load of 100 Ω resistive +/– 5% differential to 2.5 GHz.

17.8.4 Jitter Tolerance

Jitter tolerance is measured at the receiver using a jitter tolerance test signal. This signal is obtained by first producing the sum of deterministic and random jitter defined in Section 17.6, “Receiver Specifications,” and then adjusting the signal amplitude until the data eye contacts the 6 points of the minimum eye opening of the receive template shown in Figure 60 and Table 75. Note that for this to occur, the test signal must have vertical waveform symmetry about the average value and have horizontal symmetry (including jitter) about the mean zero crossing. Eye template measurement requirements are as defined above. Random jitter is calibrated using a high pass filter with a low frequency corner at 20 MHz and a 20 dB/decade roll-off below this. The required sinusoidal jitter specified in Section 17.6, “Receiver Specifications,” is then added to the signal and the test load is replaced by the receiver being tested.

18 Package Description

This section describes package parameters, pin assignments, and dimensions.

Table 76. MPC8572E Pinout Listing (continued)

Signal	Signal Name	Package Pin Number	Pin Type	Power Supply	Notes
Power and Ground Signals					
GND	Ground	A18, A25, A29, C3, C6, C9, C12, C15, C20, C22, E5, E8, E11, E14, F3, G7, G10, G13, G16, H5, H21, J3, J9, J12, J18, K7, L5, L13, L15, L16, L21, M3, M9, M12, M14, M16, M18, N7, N13, N15, N17, N19, N21, N23, P5, P12, P14, P16, P20, P22, R3, R9, R11, R13, R15, R17, R19, R21, R23, R26, T7, T12, T14, T16, T18, T20, T22, T30, U5, U11, U13, U15, U16, U17, U19, U21, U23, U25, V3, V9, V12, V14, V16, V18, V20, V22, W7, W11, W13, W15, W17, W19, W21, W27, W32, Y5, Y12, Y14, Y16, Y18, Y20, AA3, AA9, AA13, AA15, AA17, AA19, AA21, AA30, AB7, AB26, AC5, AC11, AC13, AD3, AD9, AD14, AD17, AD22, AE7, AE13, AF5, AF11, AG3, AG9, AG15, AG19, AH7, AH13, AH22, AJ5, AJ11, AJ17, AK3, AK9, AK15, AK24, AL7, AL13, AL19, AL26	—	—	—
XGND_SRDS1	SerDes Transceiver Pad GND (xpadvss)	C23, C27, D23, D25, E23, E26, F23, F24, G23, G27, H23, H25, J23, J26, K23, K24, L27, M25	—	—	—
XGND_SRDS2	SerDes Transceiver Pad GND (xpadvss)	AD23, AD25, AE23, AE27, AF23, AF24, AG23, AG26, AH23, AH25, AJ27	—	—	—

Table 76. MPC8572E Pinout Listing (continued)

Signal	Signal Name	Package Pin Number	Pin Type	Power Supply	Notes
SD1_IMP_CAL_RX	SerDes1 Rx Impedance Calibration	B32	I	200Ω (±1%) to GND	—
SD1_IMP_CAL_TX	SerDes1 Tx Impedance Calibration	T32	I	100Ω (±1%) to GND	—
SD1_PLL_TPA	SerDes1 PLL Test Point Analog	J30	O	AVDD_SRDS analog	17
SD2_IMP_CAL_RX	SerDes2 Rx Impedance Calibration	AC32	I	200Ω (±1%) to GND	—
SD2_IMP_CAL_TX	SerDes2 Tx Impedance Calibration	AM32	I	100Ω (±1%) to GND	—
SD2_PLL_TPA	SerDes2 PLL Test Point Analog	AH30	O	AVDD_SRDS analog	17
TEMP_ANODE	Temperature Diode Anode	AA31	—	internal diode	14
TEMP_CATHODE	Temperature Diode Cathode	AB31	—	internal diode	14
No Connection Pins					

Figure 62 shows the PLL power supply filter circuits.

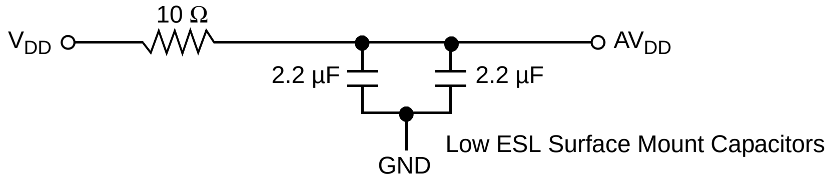
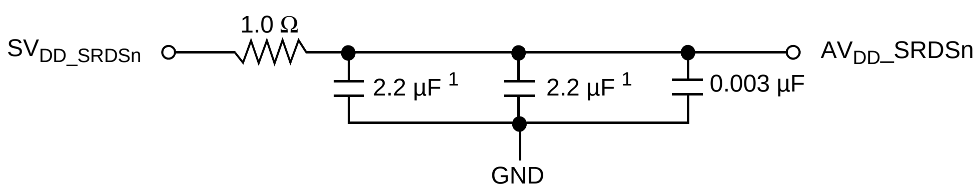


Figure 62. PLL Power Supply Filter Circuit

NOTE

It is recommended to have the minimum number of vias in the AV_{DD} trace for board layout. For example, zero vias might be possible if the AV_{DD} filter is placed on the component side. One via might be possible if it is placed on the opposite of the component side. Additionally, all traces for AV_{DD} and the filter components should be low impedance, 10 to 15 mils wide and short. This includes traces going to GND and the supply rails they are filtering.

The AV_{DD}_SRDSn signal provides power for the analog portions of the SerDesn PLL. To ensure stability of the internal clock, the power supplied to the PLL is filtered using a circuit similar to the one shown in following figure. For maximum effectiveness, the filter circuit is placed as closely as possible to the AV_{DD}_SRDSn ball to ensure it filters out as much noise as possible. The ground connection should be near the AV_{DD}_SRDSn ball. The 0.003-µF capacitor is closest to the ball, followed by the two 2.2 µF capacitors, and finally the 1 Ω resistor to the board supply plane. The capacitors are connected from AV_{DD}_SRDSn to the ground plane. Use ceramic chip capacitors with the highest possible self-resonant frequency. All traces should be kept short, wide and direct.



1. An 0805 sized capacitor is recommended for system initial bring-up.

Figure 63. SerDes PLL Power Supply Filter

NOTE

AV_{DD}_SRDSn should be a filtered version of SV_{DD}_SRDS*n*.

NOTE

Signals on the SerDesn interface are fed from the XV_{DD}_SRDS*n* power plane.

21.3 Decoupling Recommendations

Due to large address and data buses, and high operating frequencies, the device can generate transient power surges and high frequency noise in its power supply, especially while driving large capacitive loads.

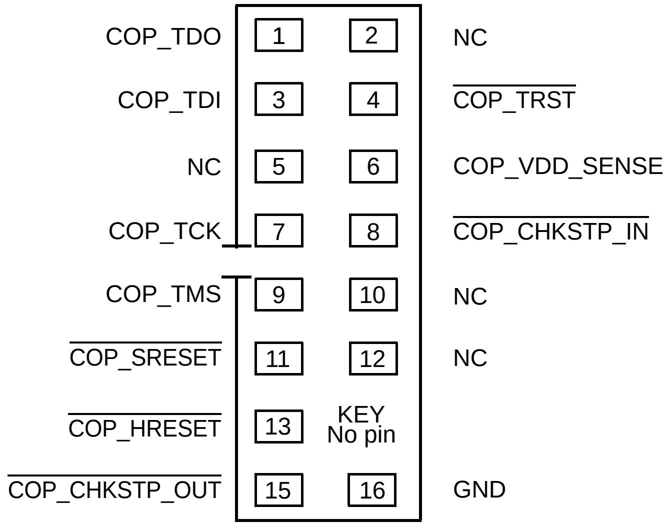


Figure 65. COP Connector Physical Pinout

22 Ordering Information

Ordering information for the parts fully covered by this specification document is provided in Section 22.1, “Part Numbers Fully Addressed by this Document.”

22.1 Part Numbers Fully Addressed by this Document

Table 86 through Table 88 provide the Freescale part numbering nomenclature for the MPC8572E. Note that the individual part numbers correspond to a maximum processor core frequency. For available frequencies, contact your local Freescale sales office. In addition to the processor frequency, the part numbering scheme also includes an application modifier which may specify special application conditions. Each part number also contains a revision code which refers to the die mask revision number.

Table 86. Part Numbering Nomenclature—Rev 2.2.1

MPC	***nnnn***	***e***	***t***	***l***	***pp***	***ffm***	***r***
Product Code[1]	**Part Identifier**	**Security Engine**	**Temperature**	**Power**	**Package Sphere Type[2]**	**Processor Frequency/ DDR Data Rate[3]**	**Silicon Revision**
MPC PPC	8572	E = Included	Blank = 0 to 105°C C = –40 to 105°C	Blank = Standard L = Low	PX = Leaded, FC-PBGA VT = Pb-free, FC-PBGA[4] VJ = Fully Pb-free FC-PBGA[5]	AVN = 1500-MHz processor; 800 MT/s DDR data rate AUL = 1333-MHz processor; 667 MT/s DDR data rate ATL = 1200-MHz processor; 667 MT/s DDR data rate ARL = 1067-MHz processor; 667 MT/s DDR data rate	E = Ver. 2.2.1 (SVR = 0x80E8_0022) SEC included
		Blank = Not included					E = Ver. 2.2.1 (SVR = 0x80E0_0022) SEC not included

Notes:

1 MPC stands for “Qualified.”
PPC stands for “Prototype”

2 See Section 18, “Package Description,” for more information on the available package types.

3 Processor core frequencies supported by parts addressed by this specification only. Not all parts described in this specification support all core frequencies. Additionally, parts addressed by part number specifications may support other maximum core frequencies.

4. The VT part number is ROHS-compliant with the permitted exception of the C4 die bumps.

5. The VJ part number is entirely lead-free. This includes the C4 die bumps.

[3] Processor core frequencies supported by parts addressed by this specification only. Not all parts described in this specification support all core frequencies. Additionally, parts addressed by part number specifications may support other maximum core frequencies.

22.2 Part Marking

Parts are marked as the example shown in Figure 67.

FC-PBGA

Notes:

MMMMMM is the 6-digit mask number.
ATWLYYWW is the traceability code.
CCCCC is the country of assembly. This space is left blank if parts are assembled in the United States.

Figure 67. Part Marking for FC-PBGA Device

Table 89 explains line four of Figure 67.

Table 89. Meaning of Last Line of Part Marking

Digit	Description
A	Assembly Site E Oak Hill Q KLM
WL	Lot number
YY	Year assembled
WW	Work week assembled

23 Document Revision History

Table 90 provides a revision history for the MPC8572E hardware specification.

Table 90. Document Revision History

Rev. Number	Date	Substantive Change(s)
7	03/2016	• Updated Section 22.2, "Part Marking," changed the five-digit mask number to six digits.