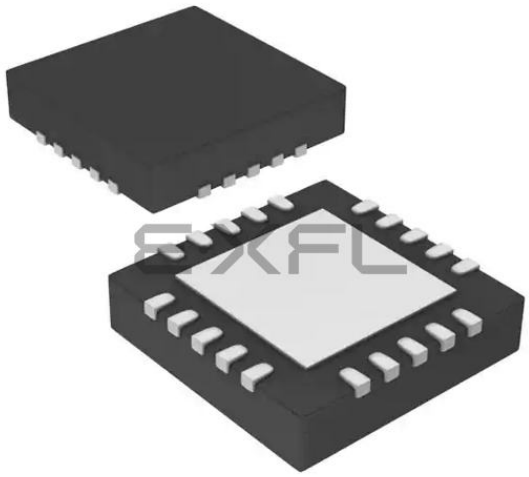

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Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	20MHz
Connectivity	-
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	17
Program Memory Size	3.5KB (2K x 14)
Program Memory Type	FLASH
EEPROM Size	256 x 8
RAM Size	128 x 8
Voltage - Supply (Vcc/Vdd)	2V ~ 5.5V
Data Converters	A/D 14x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	20-VFQFN Exposed Pad
Supplier Device Package	20-QFN (4x4)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic16hv785-i-ml

QFN (4x4x0.9) Pin Diagram

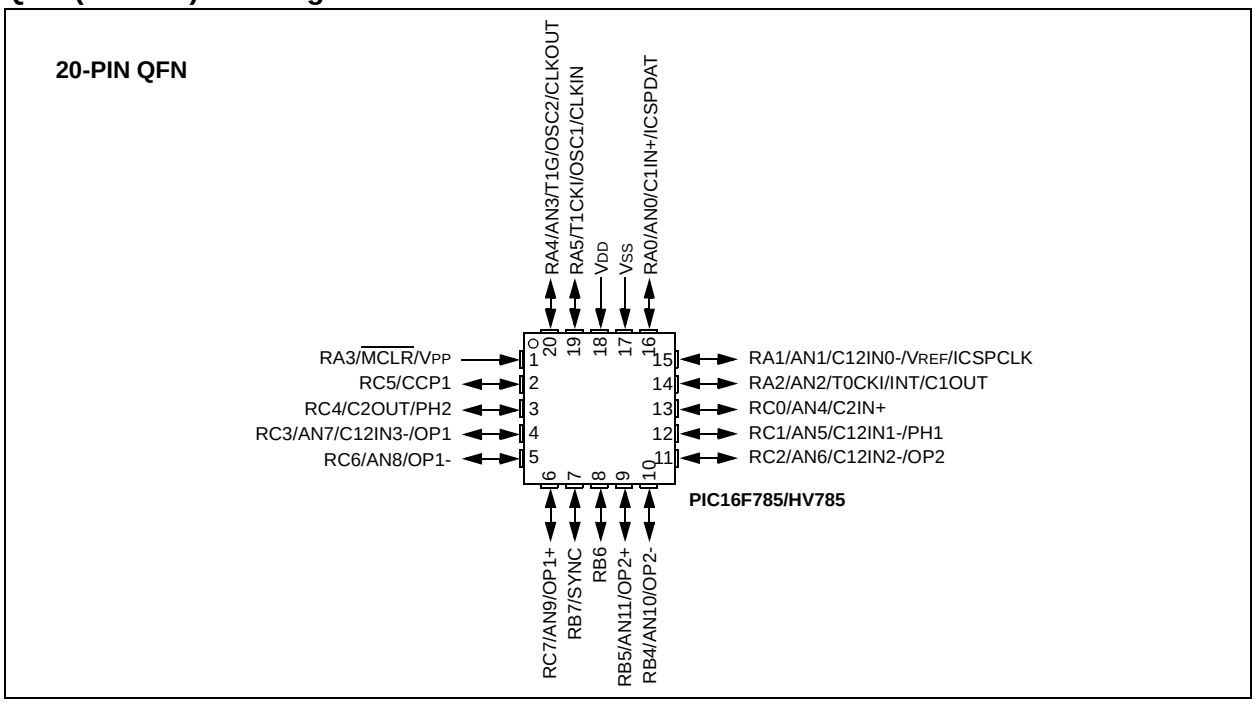


TABLE 2: QFN PIN SUMMARY

I/O	Pin	Analog	Comp.	Op Amps	PWM	Timers	CCP	Interrupt	Pull-ups	Basic
RA0	16	AN0	C1IN+	—	—	—	—	IOC	Y	ICSPDAT
RA1	15	AN1/VREF	C12IN0-	—	—	—	—	IOC	Y	ICSPCLK
RA2	14	AN2	C1OUT	—	—	T0CKI	—	INT/IOC	Y	—
RA3(1)	1	—	—	—	—	—	—	IOC	Y	$\overline{\text{MCLR}}$/VPP
RA4	20	AN3	—	—	—	T1G	—	IOC	Y	OSC2/CLKOUT
RA5	19	—	—	—	—	T1CKI	—	IOC	Y	OSC1/CLKIN
RB4	10	AN10	—	OP2-	—	—	—	—	—	—
RB5	9	AN11	—	OP2+	—	—	—	—	—	—
RB6(2)	8	—	—	—	—	—	—	—	—	—
RB7	7	—	—	—	SYNC	—	—	—	—	—
RC0	13	AN4	C2IN+	—	—	—	—	—	—	—
RC1	12	AN5	C12IN1-	—	PH1	—	—	—	—	—
RC2	11	AN6	C12IN2-	OP2	—	—	—	—	—	—
RC3	4	AN7	C12IN3-	OP1	—	—	—	—	—	—
RC4	3	—	C2OUT	—	PH2	—	—	—	—	—
RC5	2	—	—	—	—	—	CCP1	—	—	—
RC6	5	AN8	—	OP1-	—	—	—	—	—	—
RC7	6	AN9	—	OP1+	—	—	—	—	—	—
—	18	—	—	—	—	—	—	—	—	VDD
—	17	—	—	—	—	—	—	—	—	VSS

Note 1: Input only.

2: Open drain.

2.2.2.2 OPTION_REG Register

The Option register is a readable and writable register, which contains various control bits to configure the TMR0/WDT prescaler, the external RA2/INT interrupt, the TMR0 and the weak pull-ups on PORTA.

Note:	To achieve a 1:1 prescaler assignment for TMR0, assign the prescaler to the WDT by setting PSA bit to '1' in the OPTION Register. See **Section 5.4 "Prescaler"**.

REGISTER 2-2: OPTION_REG: OPTION REGISTER

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
$\overline{\text{RAPU}}$	INTEDG	T0CS	T0SE	PSA	PS2	PS1	PS0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 **$\overline{\text{RAPU}}$:** PORTA Pull-up Enable bit
1 = PORTA pull-ups are disabled
0 = PORTA pull-ups are enabled by individual port latch values in WPUA register

bit 6 **INTEDG:** Interrupt Edge Select bit
1 = Interrupt on rising edge of RA2/AN2/T0CKI/INT/C1OUT pin
0 = Interrupt on falling edge of RA2/AN2/T0CKI/INT/C1OUT pin

bit 5 **T0CS:** TMR0 Clock Source Select bit
1 = Transition on RA2/AN2/T0CKI/INT/C1OUT pin
0 = Internal instruction cycle clock (CLKOUT)

bit 4 **T0SE:** TMR0 Source Edge Select bit
1 = Increment on high-to-low transition on RA2/AN2/T0CKI/INT/C1OUT pin
0 = Increment on low-to-high transition on RA2/AN2/T0CKI/INT/C1OUT pin

bit 3 **PSA:** Prescaler Assignment bit
1 = Prescaler is assigned to the WDT
0 = Prescaler is assigned to the Timer0 module

bit 2-0 **PS<2:0>:** Prescaler Rate Select bits

Bit Value	TMR0 Rate	WDT Rate(1)
000	1 : 2	1 : 1
001	1 : 4	1 : 2
010	1 : 8	1 : 4
011	1 : 16	1 : 8
100	1 : 32	1 : 16
101	1 : 64	1 : 32
110	1 : 128	1 : 64
111	1 : 256	1 : 128

Note 1: A dedicated 16-bit WDT postscaler is available for the PIC16F785/HV785. See **Section 15.5 "Watchdog Timer (WDT)"** for more information.

3.2 Clock Source Modes

Clock Source modes can be classified as external or internal.

- External Clock modes rely on external circuitry for the clock source. Examples are oscillator modules (EC mode), quartz crystal resonators or ceramic resonators (LP, XT, and HS modes) and resistor-capacitor (RC mode) circuits.
- Internal clock sources are contained internally within the PIC16F785/HV785. The PIC16F785/HV785 has two internal oscillators; the 8 MHz High-frequency Internal Oscillator (HFINTOSC) and 31 kHz Low-frequency Internal Oscillator (LFINTOSC).

The system clock can be selected between external or internal clock sources via the System Clock Selection (SCS) bit (see **Section 3.5 "Clock Switching"**).

3.3 External Clock Modes

3.3.1 OSCILLATOR START-UP TIMER (OST)

When the PIC16F785/HV785 is configured for any of the Crystal Oscillator modes (LP, XT or HS), the Oscillator Start-up Timer (OST) is enabled, which extends the Reset period to allow the oscillator additional time to stabilize. The OST counts 1024 clock periods present on the OSC1 pin following a Power-on Reset (POR), a wake from Sleep, or when the Power-up Timer (PWRT) has expired (if the PWRT is enabled). During this time, the program counter does not increment and program execution is suspended. The OST ensures that the oscillator circuit, using a quartz crystal resonator or ceramic resonator, has started and is providing a stable system clock to the PIC16F785/HV785. Table 3-1 shows examples where the oscillator delay is invoked.

In order to minimize latency between external oscillator start-up and code execution, the Two-Speed Clock Start-up mode can be selected (see **Section 3.6 "Two-Speed Clock Start-up Mode"**).

TABLE 3-1: OSCILLATOR DELAY EXAMPLES

Switch From	Switch To	Frequency	Oscillator Delay	Comments
Sleep/POR	INTRC INTOSC	31 kHz 125 kHz-8 MHz	5 μs-10 μs (approx.) CPU Start-up(1)	Following a wake-up from Sleep mode or POR, CPU start-up is invoked to allow the CPU to become ready for code execution.
Sleep	EC, RC	DC – 20 MHz		
LFINTOSC (31 kHz)	EC, RC	DC – 20 MHz		
Sleep/POR	LP, XT, HS	31 kHz-20 MHz	1024 Clock Cycles (OST)	
LFINTOSC (31 kHz)	INTOSC	125 kHz-8 MHz	1 μs (approx.)	

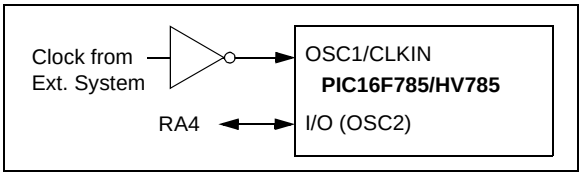
Note 1: The 5 μs-10 μs start-up delay is based on a 1 MHz System Clock.

3.3.2 EC MODE

The External Clock (EC) mode allows an externally generated logic level as the system clock source. When operating in this mode, an external clock source is connected to OSC1 pin and the RA4 pin is available for general purpose I/O. Figure 3-2 shows the pin connections for EC mode.

The Oscillator Start-up Timer (OST) is disabled when EC mode is selected. Therefore, there is no delay in operation after a Power-on Reset (POR) or wake-up from Sleep. Because the PIC16F785/HV785 design is fully static, stopping the external clock input will have the effect of halting the device while leaving all data intact. Upon restarting the external clock, the device will resume operation as if no time had elapsed.

FIGURE 3-2: EXTERNAL CLOCK (EC) MODE OPERATION

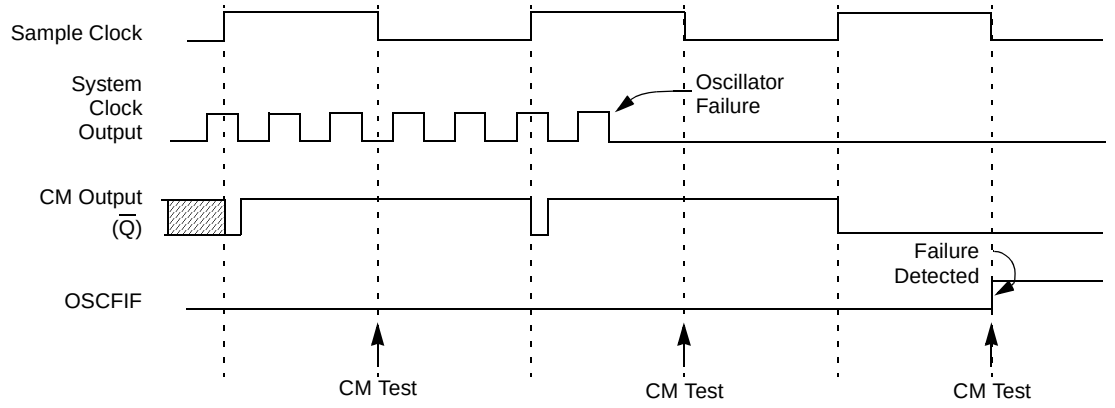


3.7.1 FAIL-SAFE CONDITION CLEARING

The Fail-Safe condition is cleared after a Reset, the execution of a SLEEP instruction, or a modification of the SCS bit. While in Fail-Safe condition, the PIC16F785/HV785 uses the internal oscillator as the system clock source. The IRCF bits in the OSCCON Register can be modified to adjust the internal oscillator frequency without exiting the Fail-Safe condition.

The Fail-Safe condition must be cleared before the OSFIF flag can be cleared.

FIGURE 3-9: FSCM TIMING DIAGRAM



Note: The system clock is normally at a much higher frequency than the sample clock. The relative frequencies in this example have been chosen for clarity.

3.7.2 RESET OR WAKE-UP FROM SLEEP

The FSCM is designed to detect oscillator failure at any point after the device has exited a Reset or Sleep condition and the Oscillator Start-up Timer (OST) has expired. If the external clock is EC or RC mode, monitoring will begin immediately following these events.

For LP, XT or HS mode, the external oscillator may require a start-up time considerably longer than the FSCM sample clock time; a false clock failure may be detected (see Figure 3-9). To prevent this, the internal oscillator is automatically configured as the system clock and functions until the external clock is stable (the OST has timed out). This is identical to Two-Speed Start-up mode. Once the external oscillator is stable, the LFINTOSC returns to its role as the FSCM source.

Note:	Due to the wide range of oscillator start-up times, the Fail-Safe circuit is not active during oscillator start-up (i.e., after exiting Reset or Sleep). After an appropriate amount of time, the user should check the OSTS bit in the OSCCON Register to verify the oscillator start-up and system clock switchover has successfully completed.

8.3.3 OPERATION IN SLEEP MODE

In Sleep mode, all clock sources are disabled. Timer2 will not increment and the state of the module will not change. If the RC5/CCP1 pin is driving a value, it will continue to drive that value. When the device wakes up, it will continue from this state.

8.3.3.1 OPERATION WITH FAIL-SAFE CLOCK MONITOR

If the Fail-Safe Clock Monitor is enabled, a clock failure will force the CCP to be clocked from the internal oscillator clock source, which may have a different clock frequency than the primary clock.

See **Section 3.0 "Clock Sources"** for additional details.

8.3.4 EFFECTS OF RESET

Any Reset will force all ports to Input mode and the CCP registers to their Reset states.

8.3.5 SETUP FOR PWM OPERATION

The following steps should be taken when configuring the CCP module for PWM operation:

1. Configure the PWM pin (RC5/CCP1) as an input by setting the TRISC<5> bit.
2. Set the PWM period by loading the PR2 register.
3. Configure the CCP module for the PWM mode by loading the CCP1CON register with the appropriate values.
4. Set the PWM duty cycle by loading the CCPR1L register and CCP1CON<5:4> bits.
5. Configure and start TMR2:
   - Clear the TMR2 interrupt flag bit by clearing the TMR2IF bit of the PIR1 Register.
   - Set the TMR2 prescale value by loading the T2CKPS bits of the T2CON Register.
   - Enable Timer2 by setting the TMR2ON bit of the T2CON Register.
6. Enable PWM output after a new PWM cycle has started:
   - Wait until TMR2 overflows (TMR2IF bit is set).
   - Enable the RC5/CCP1 pin output by clearing the TRISC<5> bit.

TABLE 8-4: REGISTERS ASSOCIATED WITH CCP AND TIMER2

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
CCP1CON	—	—	DC1B1	DC1B0	CCP1M3	CCP1M2	CCP1M1	CCP1M0	0000 0000	0000 0000
CCPR1L	Capture/Compare/PWM Register 1 Low Byte								xxxx xxxx	uuuu uuuu
CCPR1H	Capture/Compare/PWM Register 1 High Byte								xxxx xxxx	uuuu uuuu
INTCON	GIE	PEIE	T0IE	INTE	RAIE	T0IF	INTF	RAIF	0000 0000	0000 0000
PIE1	EEIE	ADIE	CCP1IE	C2IE	C1IE	OSFIE	TMR2IE	TMR1IE	0000 0000	0000 0000
PIR1	EEIF	ADIF	CCP1IF	C2IF	C1IF	OSFIF	TMR2IF	TMR1IF	0000 0000	0000 0000
PR2	Timer2 Module Period Register								1111 1111	1111 1111
T2CON	—	TOUTPS3	TOUTPS2	TOUTPS1	TOUTPS0	TMR2ON	T2CKPS1	T2CKPS0	-000 0000	-000 0000
TMR2	Timer2 Module Register								0000 0000	0000 0000
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	--11 1111	--11 1111

Legend: – = Unimplemented locations, read as '0', u = unchanged, x = unknown. Shaded cells are not used by the CCP or Timer2 modules.

9.2 Comparator Outputs

The comparator outputs are read through the CM1CON0, COM2CON0 or CM2CON1 registers. CM1CON0 and CM2CON0 each contain the individual comparator output of Comparator 1 and Comparator 2, respectively. CM2CON2 contains a mirror copy of both comparator outputs facilitating a simultaneous read of both comparators. These bits are read-only. The comparator outputs may also be directly output to the RA2/AN2/T0CKI/INT/C1OUT and RC4/C2OUT/PH2 I/O pins. When enabled, multiplexers in the output path of the RA2 and RC4 pins will switch and the output of each pin will be the unsynchronized output of the comparator. The uncertainty of each of the comparators is related to the input offset voltage and the response time given in the specifications. Figure 9-1 and Figure 9-2 show the output block diagrams for Comparators 1 and 2, respectively.

The TRIS bits will still function as an output enable/ disable for the RA2/AN2/T0CKI/INT/C1OUT and RC4/ C2OUT/PH2 pins while in this mode.

The polarity of the comparator outputs can be changed using the C1POL and C2POL bits of the CMxCON0 Register.

Timer1 gate source can be configured to use the T1G pin or Comparator 2 output as selected by the T1GSS bit of the CM2CON1 Register. The Timer1 gate feature can be used to time the duration or interval of analog events. The output of Comparator 2 can also be synchronized with Timer1 by setting the C2SYNC bit of the CM2CON1 Register. When enabled, the output of Comparator 2 is latched on the falling edge of the Timer1 clock source. If a prescaler is used with Timer1, Comparator 2 is latched after the prescaler. To prevent a race condition, the Comparator 2 output is latched on the falling edge of the Timer1 clock source and Timer1 increments on the rising edge of its clock source. See the Comparator 2 Block Diagram (Figure 9-2) and the Timer1 Block Diagram (Figure 6-1) for more information.

It is recommended to synchronize Comparator 2 with Timer1 by setting the C2SYNC bit when Comparator 2 is used as the Timer1 gate source. This ensures Timer1 does not miss an increment if Comparator 2 changes during an increment.

9.3 Comparator Interrupts

The comparator interrupt flags are set whenever there is a change in the output value of its respective comparator. Software will need to maintain information about the status of the output bits, as read from CM2CON0<7:6>, to determine the actual change that has occurred. The CxIF bits, PIR1<4:3>, are the Comparator Interrupt Flags. Each comparator interrupt bit must be reset in software by clearing it to '0'. Since it is also possible to write a '1' to this register, a simulated interrupt may be initiated.

The CxIE bits of the PIE1 Register and the PEIE bit of the INTCON Register must be set to enable the interrupts. In addition, the GIE bit must also be set. If any of these bits are cleared, the interrupt is not enabled, though the CxIF bits will still be set if an interrupt condition occurs.

The comparator interrupt of the PIC16F785/HV785 differs from previous designs in that the interrupt flag is set by the mismatch edge and not the mismatch level. This means that the interrupt flag can be reset without the additional step of reading or writing the CMxCON0 register to clear the mismatch registers. When the mismatch registers are not cleared, an interrupt will not occur when the comparator output returns to the previous state. When the mismatch registers are cleared, an interrupt will occur when the comparator returns to the previous state.

Note 1: If a change in the CMxCON0 register (CxOUT) should occur when a read operation is being executed (start of the Q2 cycle), then the CxIF of the PIR1 Register interrupt flag may not get set.

2: When either comparator is first enabled, bias circuitry in the Comparator module may cause an invalid output from the comparator until the bias circuitry is stable. Allow about 1 μs for bias settling then clear the mismatch condition and interrupt flags before enabling comparator interrupts.

9.4 Effects of Reset

A Reset forces all registers to their Reset state. This disables both comparators.

13.0 TWO-PHASE PWM

The two-phase PWM (Pulse Width Modulator) is a stand-alone peripheral that supports:

- Single or dual-phase PWM
- Single complementary output PWM with overlap/ delay
- Sync input/output to cascade devices for additional phases

Setting either, or both, of the PH1EN or PH2EN bits of the PWMCON0 register will activate the PWM module (see Register 13-1). If PH1 is used then TRISC<1> must be cleared to configure the pin as an output. The same is true for TRISC<4> when using PH2. Both PH1EN and PH2EN must be set when using Complementary mode.

13.1 PWM Period

The PWM period is derived from the main clock (FOSC), the PWM prescaler and the period counter (see Figure 13-1). The prescale bits of the PWMP Register, (see Register 13-2) determine the value of the clock divider which divides the system clock (FOSC) to the pwm_clk. This pwm_clk is used to drive the PWM counter. In Master mode, the PWM counter is reset when the count reaches the period count of the PER Register, (see Register 13-2), which determines the frequency of the PWM. The relationship between the PWM frequency, prescale and period count is shown in Equation 13-1.

EQUATION 13-1: PWM FREQUENCY

$$PWM_{FREQ} = \frac{F_{OSC}}{(2^{PWMP} \bullet (PER + 1)}$$

The maximum PWM frequency is FOSC/2, since the period count must be greater than zero.

In Slave mode, the period counter is reset by the SYNC input, which is the master device period counter reset. For proper operation, the slave period count should be equal to or greater than that of the master.

13.2 PWM Phase

Each enabled phase output is driven active when the phase counter matches the corresponding PWM phase count in the PH Register (see Register 13-3 and Register 13-4). The phase output remains true until terminated by a feedback signal from either of the comparators or the auto-shutdown activates.

Phase granularity is a function of the period count value. For example, if PER<4:0> = 3, each output can be shifted in 90° steps (see Equation 13-2).

EQUATION 13-2: PHASE RESOLUTION

$$Phase_{DEG} = \frac{360}{(PER + 1)}$$

13.3 PWM Duty Cycle

Each PWM output is driven inactive, terminating the drive period, by asynchronous feedback through the internal comparators. The duty cycle resolution is in effect infinitely adjustable. Either or both comparators can be used to reset the PWM by setting the corresponding comparator enable bit (CxEN, see Register 13-3). Duty cycles of 100% can be obtained by suppressing the feedback which would otherwise terminate the pulse.

The comparator outputs can be “held off”, or blanked, by enabling the corresponding BLANK bit (BLANKx, see Register 13-1) for each phase. The blank bit disables the comparator outputs for 1/2 of a system clock (FOSC), thus ensuring at least TOSC/2 active time for the PWM output. Blanking avoids early termination of the PWM output which may result due to switching transients at the beginning of the cycle.

13.4 Master/Slave Operation

Multiple chips can operate together to achieve additional phases by operating one as the master and the others as slaves. When the PWM is configured as a master, the RB7/SYNC pin is an output and generates a high output for one pwm_clk period at the end of each PWM period (see Figure 13-4).

When the PWM is configured as a slave, the RB7/ SYNC pin is an input. The high input from a master in this configuration resets the PWM period counter which synchronizes the slave unit at the end of each PWM period. Proper operation of a slave device requires a common external FOSC clock source to drive the master and slave. The PWM prescale value of the slave device must also be identical to that of the master. As mentioned previously, the slave period count value must be greater than or equal to that of the master.

The PWM Counter will be reset and held at zero when both PH1EN and PH2EN of the PWMCON0 Register are false. If the PWM is configured as a slave, the PWM Counter will remain reset at zero until the first SYNC input is received.

REGISTER 13-3: PWMPH1: PWM PHASE 1 CONTROL REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
POL	C2EN	C1EN	PH4	PH3	PH2	PH1	PH0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 **POL:** PH1 Output Polarity bit
1 = PH1 Pin is active-low
0 = PH1 Pin is active-high

bit 6 **C2EN:** Comparator 2 Enable bit
When COMOD<1:0> = 00(1)
1 = PH1 is reset when C2OUT goes high
0 = PH1 ignores Comparator 2
When COMOD<1:0> = X1(1)
1 = Complementary drive terminates when C2OUT goes high
0 = Comparator 2 is ignored
When COMOD<1:0> = 10(1)
C2EN has no effect

bit 5 **C1EN:** Comparator 1 Enable bit
When COMOD<1:0> = 00(1)
1 = PH1 is reset when C1OUT goes high
0 = PH1 ignores Comparator 1
When COMOD<1:0> = X1(1)
1 = Complementary drive terminates when C1OUT goes high
0 = Comparator 1 is ignored
When COMOD<1:0> = 10(1)
C1EN has no effect

bit 4-0 **PH<4:0>:** PWM Phase bits
When COMOD<1:0> = 00(1)
00000 = PH1 starts 1 pwm_clk period after falling edge of SYNC pulse. All other PH1 delays are expressed relative to this time.
00001 = PH1 is delayed by 1 pwm_clk pulse
••••• = •••
11111 = PH1 is delayed by 31 pwm_clk pulses
When COMOD<1:0> = X1 or 1X(1)
00000 = Complementary drive starts 1 pwm_clk period after falling edge of SYNC pulse. All other delays are expressed relative to this time.
00001 = Complementary drive start is delayed by 1 pwm_clk pulse
••••• = •••
11111 = Complementary drive start is delayed by 31 pwm_clk pulses

Note 1: See PWMCON1 register (Register 13-5).

14.2 Reading the EEPROM Data Memory

To read a data memory location, the user must write the address to the EEADR register and then set control bit RD of the EECON1 Register, as shown in Example 14-1. The data is available, in the very next cycle, in the EEDAT register. Therefore, it can be read in the next instruction. EEDAT holds this value until another read, or until it is written to by the user (during a write operation).

EXAMPLE 14-1: DATA EEPROM READ

```
BSF     STATUS,RP0    ;Bank 1
BCF     STATUS,RP1    ;
MOVLW   CONFIG_ADDR   ;
MOVWF   EEADR         ;Address to read
BSF     EECON1,RD     ;EE Read
MOVF    EEDAT,W       ;Move data to W
```

14.3 Writing to the EEPROM Data Memory

To write an EEPROM data location, the user must first write the address to the EEADR register and the data to the EEDAT register. Then the user must follow a specific sequence to initiate the write for each byte, as shown in Example 14-2.

EXAMPLE 14-2: DATA EEPROM WRITE

```
        BSF     STATUS,RP0   ;Bank 1
        BCF     STATUS,RP1   ;
        BSF     EECON1,WREN  ;Enable write
        BCF     INTCON,GIE   ;Disable INTs
        BTFSC   INTCON,GIE   ;See AN-576
        GOTO    $-2          ;
Required
Sequence
        MOVLW   55h          ;Unlock write
        MOVWF   EECON2       ;
        MOVLW   AAh          ;
        MOVWF   EECON2       ;
        BSF     EECON1,WR    ;Start the write
        BSF     INTCON,GIE   ;Enable INTs
```

The write will not initiate if the sequence in Example 14-2 is not followed exactly (write 55h to EECON2, write AAh to EECON2, then set WR bit) for each byte. It is strongly recommended that interrupts be disabled during this code segment. A cycle count is executed during the required sequence. Any number that is not equal to the required cycles to execute the required sequence will prevent the data from being written into the EEPROM.

Additionally, the WREN bit in EECON1 must be set to enable write. This mechanism prevents accidental writes to data EEPROM due to errant (unexpected) code execution (i.e., lost programs). The user should keep the WREN bit clear at all times, except when updating the EEPROM. The WREN bit is not cleared by hardware.

After a write sequence has been initiated, clearing the WREN bit will not affect this write cycle. The WR bit will be inhibited from being set unless the WREN bit is set.

At the completion of the write cycle, the WR bit is cleared in the hardware and the EE Write Complete Interrupt Flag bit (EEIF) is set. The user can either enable this interrupt or poll this bit. The EEIF bit of the PIR1 Register must be cleared by software.

14.4 Write Verify

Depending on the application, good programming practice may dictate that the value written to the data EEPROM should be verified (see Example 14-3) to the desired value to be written.

EXAMPLE 14-3: WRITE VERIFY

```
BSF     STATUS,RP0   ;Bank 1
BCF     STATUS,RP1   ;
MOVF    EEDAT,W      ;EEDAT not changed
                      from previous write
BSF     EECON1,RD    ;YES, Read the
                     ; value written
XORWF   EEDAT,W      ;
BTFSS   STATUS,Z     ;Is data the same
GOTO    WRITE_ERR    ;No, handle error
                     ;Yes, continue
```

14.4.1 USING THE DATA EEPROM

The data EEPROM is a high-endurance, byte addressable array that has been optimized for the storage of frequently changing information (e.g., program variables or other data that are updated often). When variables in one section change frequently, while variables in another section do not change, it is possible to exceed the total number of write cycles to the EEPROM (specification D124) without exceeding the total number of write cycles to a single byte (specifications D120 and D120A). If this is the case, then a refresh of the array must be performed. For this reason, variables that change infrequently (such as constants, IDs, calibration, etc.) should be stored in Flash program memory.

14.5 Protect Against Spurious Write

There are conditions when the user may not want to write to the data EEPROM memory. To protect against spurious EEPROM writes, various mechanisms have been built in. On power-up, WREN is cleared. Also, the Power-up Timer (64 ms duration) prevents EEPROM write.

The write initiate sequence and the WREN bit helps prevent an accidental write during a brown-out, power glitch and software malfunction.

TABLE 15-4: INITIALIZATION CONDITION FOR REGISTERS (CONTINUED)

Register	Address	Power-on Reset	$\overline{\text{MCLR}}$ Reset WDT Reset Brown-out Reset(1)	Wake-up from Sleep through interrupt Wake-up from Sleep through WDT Time-out
VRCON	99h	000- 0000	000- 0000	uuu- uuuu
EEDAT	9Ah	0000 0000	0000 0000	uuuu uuuu
EEADR	9Bh	0000 0000	0000 0000	uuuu uuuu
EECON1	9Ch	---- x000	---- q000	---- uuuu
EECON2	9Dh	---- ----	---- ----	---- ----
ADRESL	9Eh	xxxx xxxx	uuuu uuuu	uuuu uuuu
ADCON1	9Fh	-000 ----	-000 ----	-uuu ----
PWMCON1	110h	-000 0000	-000 0000	-uuu uuuu
PWMCON0	111h	0000 0000	0000 0000	uuuu uuuu
PWMCLK	112h	0000 0000	0000 0000	uuuu uuuu
PWMPH1	113h	0000 0000	0000 0000	uuuu uuuu
PWMPH2	114h	0000 0000	0000 0000	uuuu uuuu
CM1CON0	119h	0000 0000	0000 0000	uuuu uuuu
CM2CON0	11Ah	0000 0000	0000 0000	uuuu uuuu
CM2CON1	11Bh	00-- --10	00-- --10	uu-- --uu
OPA1CON	11Ch	0--- ----	0--- ----	u--- ----
OPA2CON	11Dh	0--- ----	0--- ----	u--- ----

Legend: u = unchanged, x = unknown, – = unimplemented bit, reads as '0', q = value depends on condition.

Note 1: If V_{DD} goes too low, Power-on Reset will be activated and registers will be affected differently.

2: One or more bits in INTCON and/or PIR1 will be affected (to cause wake-up).

3: When the wake-up is due to an interrupt and the GIE bit is set, the PC is loaded with the interrupt vector (0004h).

4: See Table 15-5 for Reset value for specific condition.

5: If Reset was due to brown-out, then bit 0 = 0. All other Resets will cause bit 0 = u.

6: Analog channels read 0 but data latches are unknown.

7: Analog channels read 0 but data latches are unchanged.

15.3 Interrupts

The PIC16F785/HV785 has 11 sources of interrupt:

- External Interrupt RA2/INT
- TMR0 Overflow Interrupt
- PORTA Change Interrupt
- 2 Comparator Interrupts
- A/D Interrupt
- Timer1 Overflow Interrupt
- Timer2 Match Interrupt
- EEPROM Data Write Interrupt
- Fail-Safe Clock Monitor Interrupt
- CCP Interrupt

The Interrupt Control register (INTCON) and Peripheral Interrupt register (PIR1) record individual interrupt requests in flag bits. The INTCON register also has individual and global interrupt enable bits.

A Global Interrupt Enable bit, GIE of the INTCON Register enables (if set) all unmasked interrupts, or disables (if cleared) all interrupts. Individual interrupts can be disabled through their corresponding enable bits in INTCON register and PIE1 register. GIE is cleared on Reset.

The Return from Interrupt instruction, RETFIE, exits interrupt routine, as well as sets the GIE bit, which re-enables unmasked interrupts.

The following interrupt flags are contained in the INTCON register:

- INT Pin Interrupt
- PORTA Change Interrupt
- TMR0 Overflow Interrupt

The peripheral interrupt flags are contained in the special register PIR1. The corresponding interrupt enable bit is contained in special register PIE1.

The following interrupt flags are contained in the PIR1 register:

- EEPROM Data Write Interrupt
- A/D Interrupt
- 2 Comparator Interrupts
- Timer1 Overflow Interrupt
- Timer2 Match Interrupt
- Fail-Safe Clock Monitor Interrupt
- CCP Interrupt

When an interrupt is serviced:

- The GIE is cleared to disable any further interrupt
- The return address is PUSHed onto the stack
- The PC is loaded with 0004h

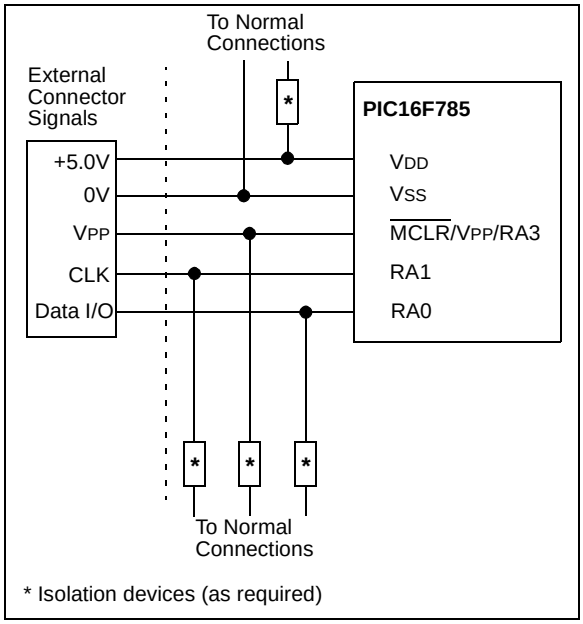
For external interrupt events, such as the INT pin or PORTA change interrupt, the interrupt latency will be three or four instruction cycles. The exact latency depends upon when the interrupt event occurs (see Figure 15-8). The latency is the same for one or two-cycle instructions. Once in the Interrupt Service Routine, the source(s) of the interrupt can be determined by polling the interrupt flag bits. The interrupt flag bit(s) must be cleared in software before re-enabling interrupts to avoid multiple interrupt requests.

Note 1: Individual interrupt flag bits are set, regardless of the status of their corresponding mask bit or the GIE bit.

2: When an instruction that clears the GIE bit is executed, any interrupts that were pending for execution in the next cycle are ignored. The interrupts, which were ignored, are still pending to be serviced when the GIE bit is set again.

For additional information on Timer1, Timer2, comparators, A/D, Data EEPROM or CCP modules, refer to the respective peripheral section.

FIGURE 15-11: TYPICAL IN-CIRCUIT SERIAL PROGRAMMING CONNECTION



15.10 In-Circuit Debugger

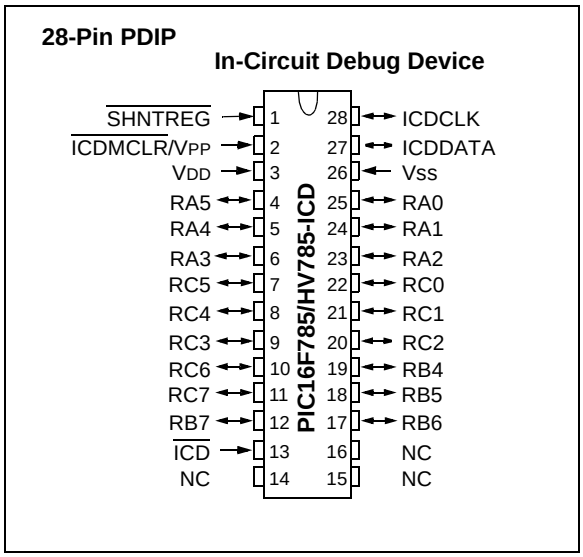
- In-circuit debugging requires clock, data and $\overline{MCLR}$ pins. A special 28-pin PIC16F785-ICD device is used with MPLAB® ICD 2 to provide separate clock, data and $\overline{MCLR}$ pins so that no pins are lost for these functions, leaving all 18 of the PIC16F785/HV785 I/O pins available to the user during debug operation.
- This special ICD device is mounted on the top of a header and its signals are routed to the MPLAB ICD 2 connector. On the bottom of the header is a 20-pin socket that plugs into the user's target via the 20-pin stand-off connector.
- When the $\overline{ICD}$ pin on the PIC16F785-ICD device is held low, the In-Circuit Debugger functionality is enabled. This function allows simple debugging functions when used with MPLAB ICD 2. When the microcontroller has this feature enabled, some of the resources are not available for general use. Table 15-9 shows which features are consumed by the background debugger.

TABLE 15-9: DEBUGGER RESOURCES

Resource	Description
I/O pins	ICDCLK, ICDDATA
Stack	1 level
Data RAM	65h-70h, F0h
Program Memory	Address 0h must be NOP 700h-7FFh

For more information, see "*MPLAB® ICD 2 In-Circuit Debugger User's Guide*" (DS51331), available on Microchip's web site (www.microchip.com).

FIGURE 15-12: 28-PIN ICD PINOUT



RETFIE **Return from Interrupt**

Syntax: [*label*] RETFIE

Operands: None

Operation: TOS → PC,
1 → GIE

Status Affected: None

Encoding:

00	0000	0000	1001

Description: Return from Interrupt. Stack is POPed and Top-of-Stack (TOS) is loaded in the PC. Interrupts are enabled by setting Global Interrupt Enable bit, GIE of the INTCON Register. This is a two-cycle instruction.

RETLW **Return with Literal in W**

Syntax: [*label*] RETLW k

Operands: 0 ≤ k ≤ 255

Operation: k → (W);
TOS → PC

Status Affected: None

Encoding:

11	01xx	kkkk	kkkk

Description: The W register is loaded with the eight-bit literal ‘k’. The program counter is loaded from the top of the stack (the return address). This is a two-cycle instruction.

RETURN **Return from Subroutine**

Syntax: [*label*] RETURN

Operands: None

Operation: TOS → PC

Status Affected: None

Description: Return from subroutine. The stack is POPed and the top of the stack (TOS) is loaded into the program counter. This is a two-cycle instruction.

RLF **Rotate Left f through Carry**

Syntax: [*label*] RLF f,d

Operands: 0 ≤ f ≤ 127
d ∈ [0,1]

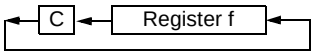
Operation: See description below

Status Affected: C

Encoding:

00	1101	dfff	ffff

Description: The contents of register ‘f’ are rotated one bit to the left through the Carry Flag. If ‘d’ is ‘0’, the result is placed in the W register. If ‘d’ is ‘1’, the result is stored back in register ‘f’.



RRF **Rotate Right f through Carry**

Syntax: [*label*] RRF f,d

Operands: 0 ≤ f ≤ 127
d ∈ [0,1]

Operation: See description below

Status Affected: C

Encoding:

00	1100	dfff	ffff

Description: The contents of register ‘f’ are rotated one bit to the right through the Carry Flag. If ‘d’ is ‘0’, the result is placed in the W register. If ‘d’ is ‘1’ the result is placed back in register ‘f’.

SLEEP **Go into Standby mode**

Syntax: [*label*] SLEEP

Operands: None

Operation: 00h → WDT,
0 → WDT prescaler,
1 → $\overline{TO}$,
0 → $\overline{PD}$

Status Affected: $\overline{TO}$, $\overline{PD}$

Encoding:

00	0000	0110	0011

Description: The power-down Status bit, $\overline{PD}$ is cleared. Time out Status bit, $\overline{TO}$ is set. Watchdog Timer and its prescaler are cleared.
The processor is put into Sleep mode with the oscillator stopped.

18.0 DEVELOPMENT SUPPORT

The PIC® microcontrollers are supported with a full range of hardware and software development tools:

- Integrated Development Environment
  - MPLAB® IDE Software
- Assemblers/Compilers/Linkers
  - MPASM™ Assembler
  - MPLAB C18 and MPLAB C30 C Compilers
  - MPLINK™ Object Linker/ MPLIB™ Object Librarian
  - MPLAB ASM30 Assembler/Linker/Library
- Simulators
  - MPLAB SIM Software Simulator
- Emulators
  - MPLAB ICE 2000 In-Circuit Emulator
  - MPLAB REAL ICE™ In-Circuit Emulator
- In-Circuit Debugger
  - MPLAB ICD 2
- Device Programmers
  - PICSTART® Plus Development Programmer
  - MPLAB PM3 Device Programmer
  - PICkit™ 2 Development Programmer
- Low-Cost Demonstration and Development Boards and Evaluation Kits

18.1 MPLAB Integrated Development Environment Software

The MPLAB IDE software brings an ease of software development previously unseen in the 8/16-bit microcontroller market. The MPLAB IDE is a Windows® operating system-based application that contains:

- A single graphical interface to all debugging tools
  - Simulator
  - Programmer (sold separately)
  - Emulator (sold separately)
  - In-Circuit Debugger (sold separately)
- A full-featured editor with color-coded context
- A multiple project manager
- Customizable data windows with direct edit of contents
- High-level source code debugging
- Visual device initializer for easy register initialization
- Mouse over variable inspection
- Drag and drop variables from source to watch windows
- Extensive on-line help
- Integration of select third party tools, such as HI-TECH Software C Compilers and IAR C Compilers

The MPLAB IDE allows you to:

- Edit your source files (either assembly or C)
- One touch assemble (or compile) and download to PIC MCU emulator and simulator tools (automatically updates all project information)
- Debug using:
  - Source files (assembly or C)
  - Mixed assembly and C
  - Machine code

MPLAB IDE supports multiple debugging tools in a single development paradigm, from the cost-effective simulators, through low-cost in-circuit debuggers, to full-featured emulators. This eliminates the learning curve when upgrading to tools with increased flexibility and power.

19.0 ELECTRICAL SPECIFICATIONS

Absolute Maximum Ratings(†)

Ambient temperature under bias ..... -40 to +125°C
Storage temperature ..... -65°C to +150°C
Voltage on V_{DD} with respect to V_{SS} ..... -0.3 to +6.5V
Voltage on $\overline{MCLR}$ with respect to Vss ..... -0.3 to +13.5V
Voltage on RB6 open-drain pin with respect to Vss ..... -0.3 to +8.5V
Voltage on all other pins with respect to V_{SS} ..... -0.3V to (V_{DD} + 0.3V)
Total power dissipation(1) (PDIP and SOIC) ..... 800 mW
Total power dissipation(1) (SSOP) ..... 600 mW
Maximum current out of V_{SS} pin ..... 300 mA
Maximum current into V_{DD} pin ..... 250 mA
Input clamp current, I_{IK} (V_I < 0 or V_I > V_{DD}) ..... ±20 mA
Output clamp current, I_{OK} (Vo < 0 or Vo > V_{DD}) ..... ±20 mA
Maximum output current sunk by any I/O pin ..... 25 mA
Maximum output current sourced by any I/O pin ..... 25 mA
Maximum current sunk by PORTA, PORTB, and PORTC (combined) ..... 200 mA
Maximum current sourced PORTA, PORTB, and PORTC (combined) ..... 200 mA

Note 1: Power dissipation is calculated as follows: $P_{DIS} = V_{DD} \times \{I_{DD} - \sum I_{OH}\} + \sum \{(V_{DD} - V_{OH}) \times I_{OH}\} + \sum (Vol \times I_{OL})$.

† **NOTICE**: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operation listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

Note: Voltage spikes below V_{SS} at the $\overline{MCLR}$ pin, inducing currents greater than 80 mA, may cause latch-up. Thus, a series resistor of 50-100Ω should be used when applying a "low" level to the $\overline{MCLR}$ pin, rather than pulling this pin directly to V_{SS}.

19.2 DC Characteristics: PIC16F785/HV785-I (Industrial)(1), (2)

DC CHARACTERISTICS		Standard Operating Conditions (unless otherwise stated) Operating temperature -40°C ≤ TA ≤ +85°C for industrial					
Param No.	Device Characteristics	Min	Typ†	Max	Units	Conditions	
						VDD	
D010	**Supply Current (IDD)**	—	11	23	µA	2.0	FOSC = 32 kHz LP Oscillator mode
		—	18	38	µA	3.0	
		—	35	75	µA	5.0	
D011		—	140	240	µA	2.0	FOSC = 1 MHz XT Oscillator mode
		—	220	380	µA	3.0	
		—	380	550	µA	5.0	
D012		—	260	360	µA	2.0	FOSC = 4 MHz XT Oscillator mode
		—	420	650	µA	3.0	
		—	0.8	1.1	mA	5.0	
D013		—	130	220	µA	2.0	FOSC = 1 MHz EC Oscillator mode
		—	215	360	µA	3.0	
		—	360	520	µA	5.0	
D014		—	220	340	µA	2.0	FOSC = 4 MHz EC Oscillator mode
		—	375	550	µA	3.0	
		—	0.65	1	mA	5.0	
D015		—	8	20	µA	2.0	FOSC = 31 kHz INTRC mode
		—	16	40	µA	3.0	
		—	31	65	µA	5.0	
D016		—	340	450	µA	2.0	FOSC = 4 MHz INTOSC mode
		—	500	700	µA	3.0	
		—	800	1200	µA	5.0	
D017		—	230	400	µA	2.0	FOSC = 4 MHz EXTRC mode
		—	400	680	µA	3.0	
		—	0.63	1.1	mA	5.0	
D018		—	2.6	3.25	mA	4.5	FOSC = 20 MHz HS Oscillator mode
		—	2.8	3.35	mA	5.0	

† Data in 'Typ' column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: The test conditions for all IDD measurements in Active Operation mode are: OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD; $\overline{MCLR}$ = VDD; WDT disabled.

2: The supply current is mainly a function of the operating voltage and frequency. Other factors such as I/O pin loading and switching rate, oscillator type, internal code execution pattern and temperature also have an impact on the current consumption.

3: The peripheral current is the sum of the base IDD or IPD and the additional current consumed when this peripheral is enabled. The peripheral ∆ current can be determined by subtracting the base IDD or IPD current from this limit. Max values should be used when calculating total current consumption.

4: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to VDD. When A/D is off, it will not consume any current other than leakage current. the power-down current spec includes any such leakage from the A/D module.

19.5 Timing Parameter Symbology

The timing parameter symbols have been created with one of the following formats:

1. TppS2ppS
2. TppS

T			
F	Frequency	T	Time

Lowercase letters (pp) and their meanings:

pp			
cc	CCP1	osc	OSC1
ck	CLKOUT	rd	$\overline{RD}$
cs	$\overline{CS}$	rw	$\overline{RD}$ or $\overline{WR}$
di	SDI	sc	SCK
do	SDO	ss	$\overline{SS}$
dt	Data in	t0	T0CKI
io	I/O port	t1	T1CKI
mc	$\overline{MCLR}$	wr	$\overline{WR}$

Uppercase letters and their meanings:

S			
F	Fall	P	Period
H	High	R	Rise
I	Invalid (High-impedance)	V	Valid
L	Low	Z	High-impedance

FIGURE 19-2: LOAD CONDITIONS

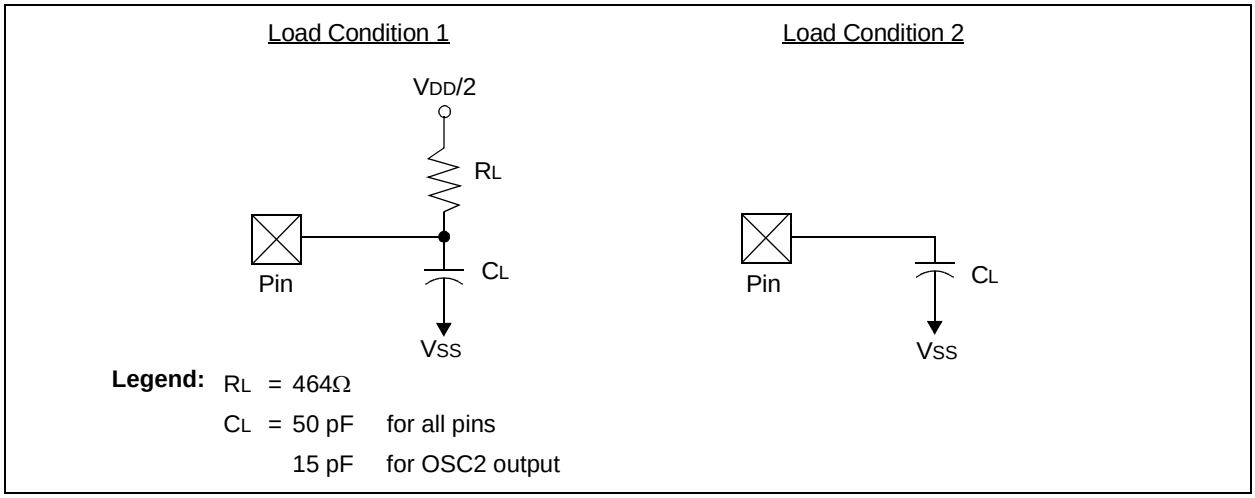


FIGURE 19-7: TIMER0 AND TIMER1 EXTERNAL CLOCK TIMINGS

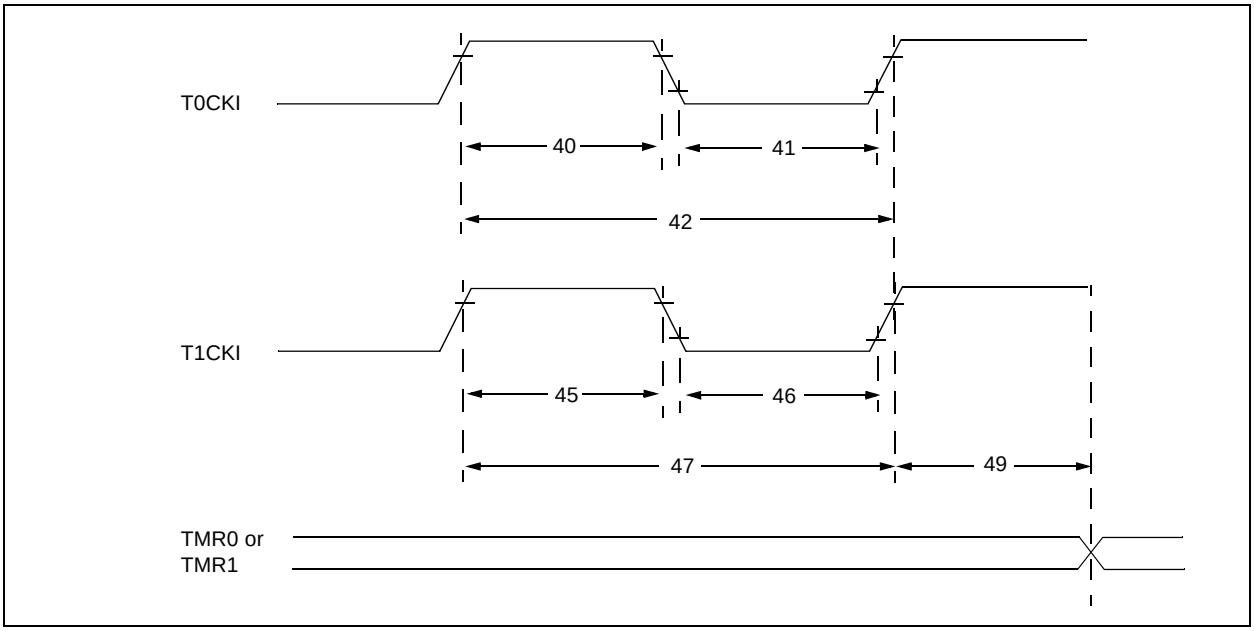


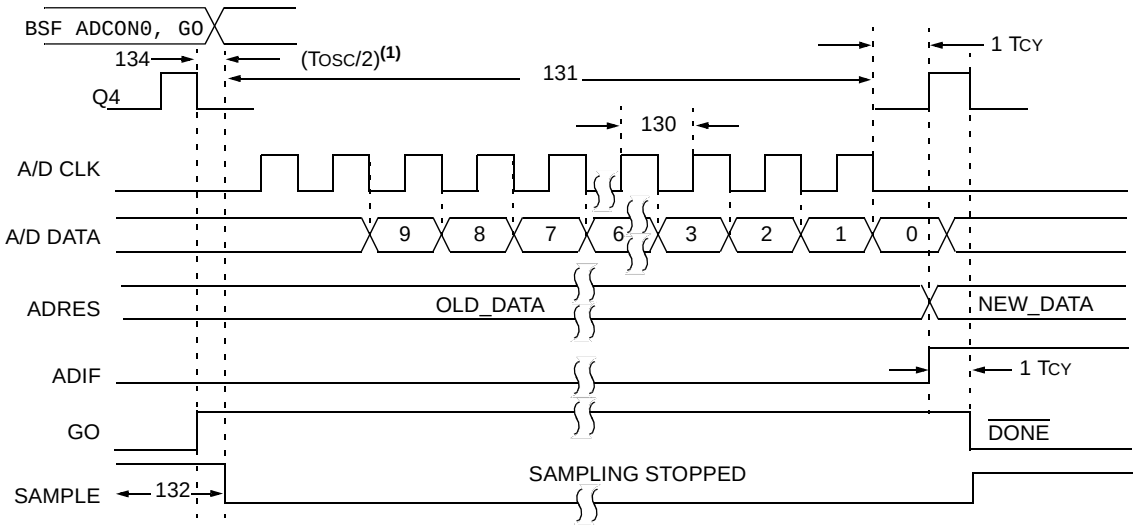
TABLE 19-5: TIMER0 AND TIMER1 EXTERNAL CLOCK REQUIREMENTS

Param No.	Sym	Characteristic			Min	Typ†	Max	Units	Conditions
40*	TT0H	T0CKI High Pulse Width		No Prescaler	0.5 TCY + 20	—	—	ns	
				With Prescaler	10	—	—	ns	
41*	TT0L	T0CKI Low Pulse Width		No Prescaler	0.5 TCY + 20	—	—	ns	
				With Prescaler	10	—	—	ns	
42*	TT0P	T0CKI Period			Greater of: 20 or TCY + 40 / N	—	—	ns	N = prescale value (2, 4, ..., 256)
45*	TT1H	T1CKI High Time	Synchronous, No Prescaler		0.5 TCY + 20	—	—	ns	
			Synchronous, with Prescaler		15	—	—	ns	
			Asynchronous		30	—	—	ns	
46*	TT1L	T1CKI Low Time	Synchronous, No Prescaler		0.5 TCY + 20	—	—	ns	
			Synchronous, with Prescaler		15	—	—	ns	
			Asynchronous		30	—	—	ns	
47*	TT1P	T1CKI Input Period	Synchronous		Greater of: 30 or TCY + 40 / N	—	—	ns	N = prescale value (1, 2, 4, 8)
			Asynchronous		60	—	—	ns	
48	FT1	Timer1 oscillator input frequency range (oscillator enabled by setting bit T1OSCEN)			DC	—	200*	kHz	
49	TCKEZTMR1	Delay from external clock edge to timer increment			2 TOSC*	—	7 TOSC*	—	

* These parameters are characterized but not tested.

† Data in 'Typ' column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

FIGURE 19-9: PIC16F785/HV785 A/D CONVERSION TIMING (NORMAL MODE)



Note 1: If the A/D clock source is selected as RC, a time of TCY is added before the A/D clock starts. This allows the SLEEP instruction to be executed.

TABLE 19-16: PIC16F785/HV785 A/D CONVERSION REQUIREMENTS

Param No.	Sym	Characteristic	Min	Typ†	Max	Units	Conditions
130	TAD	A/D Clock Period	1.6	—	—	µs	TOSC-based, VREF ≥ 3.0V
			3.0*	—	—	µs	TOSC-based, VREF full range
130	TAD	A/D Internal RC Oscillator Period					ADCS<1:0> = 11 (RC mode)
			3.0*	6.0	9.0*	µs	At VDD = 2.5V
			2.0*	4.0	6.0*	µs	At VDD = 5.0V
131	TCNV	Conversion Time (not including Acquisition Time)(1)	—	11	—	TAD	Set GO bit to new data in A/D result register
132	TACQ	Acquisition Time	**(Note 2)**	11.5	—	µs	
			5*	—	—	µs	The minimum time is the amplifier settling time. This may be used if the "new" input voltage has not changed by more than 1 LSb (i.e., 4.1 mV @ 4.096V) from the last sampled voltage (as stored on CHOLD).
134	TGO	Q4 to A/D Clock Start	—	TOSC/2	—	—	If the A/D clock source is selected as RC, a time of TCY is added before the A/D clock starts. This allows the SLEEP instruction to be executed.

* These parameters are characterized but not tested.

† Data in 'Typ' column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: ADRESH and ADRESL registers may be read on the following TCY cycle.

2: See **Section 12.2 "A/D Acquisition Requirements"** for minimum conditions.

FIGURE 20-42: TYPICAL VP6 REFERENCE VOLTAGE DISTRIBUTION (5V, 125°C)

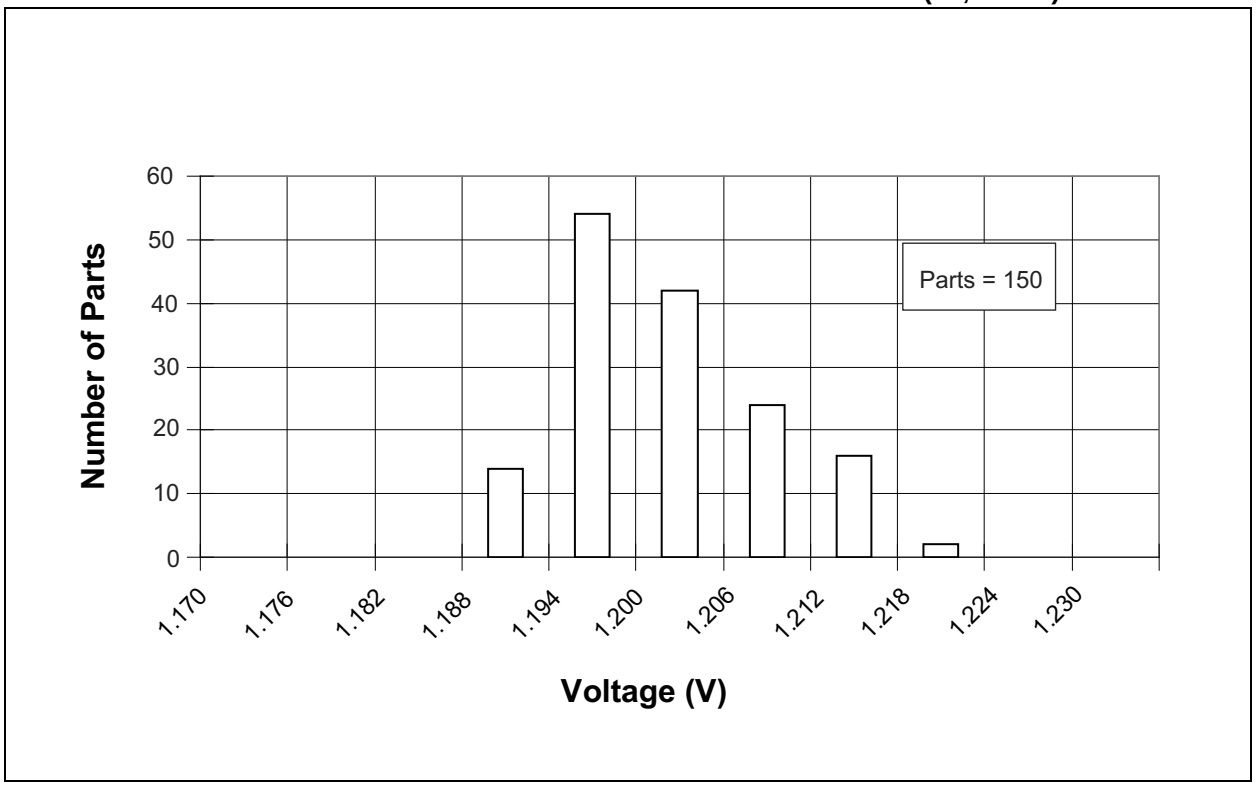


FIGURE 20-43: TYPICAL VP6 REFERENCE VOLTAGE DISTRIBUTION (5V, -40°C)