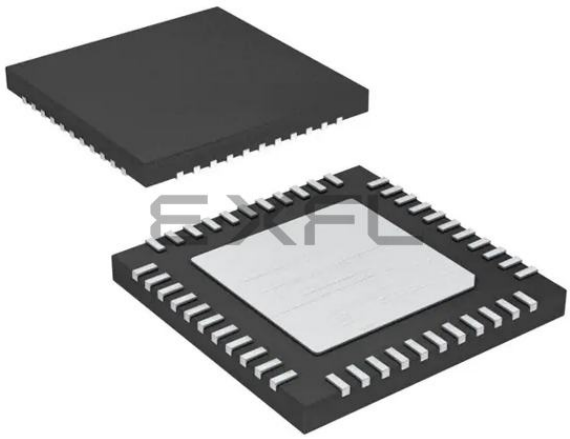

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Details	
Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	40MHz
Connectivity	CANbus, I²C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, HLVD, POR, PWM, WDT
Number of I/O	36
Program Memory Size	64KB (32K x 16)
Program Memory Type	FLASH
EEPROM Size	1K x 8
RAM Size	3.25K x 8
Voltage - Supply (Vcc/Vdd)	2V ~ 5.5V
Data Converters	A/D 11x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	44-VQFN Exposed Pad
Supplier Device Package	44-QFN (8x8)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic18lf4680-i-ml

TABLE 1-3: PIC18F4585/4680 PINOUT I/O DESCRIPTIONS (CONTINUED)

Pin Name	Pin Number			Pin Type	Buffer Type	Description
	PDIP	QFN	TQFP			
						PORTA is a bidirectional I/O port.
RA0/AN0/CVREF	2	19	19			
RA0				I/O	TTL	Digital I/O.
AN0				I	Analog	Analog input 0.
CVREF				O	Analog	Analog comparator reference output.
RA1/AN1	3	20	20			
RA1				I/O	TTL	Digital I/O.
AN1				I	Analog	Analog input 1.
RA2/AN2/VREF-	4	21	21			
RA2				I/O	TTL	Digital I/O.
AN2				I	Analog	Analog input 2.
VREF-				I	Analog	A/D reference voltage (low) input.
RA3/AN3/VREF+	5	22	22			
RA3				I/O	TTL	Digital I/O.
AN3				I	Analog	Analog input 3.
VREF+				I	Analog	A/D reference voltage (high) input.
RA4/T0CKI	6	23	23			
RA4				I/O	TTL	Digital I/O.
T0CKI				I	ST	Timer0 external clock input.
RA5/AN4/$\overline{SS}$/HLVDIN	7	24	24			
RA5				I/O	TTL	Digital I/O.
AN4				I	Analog	Analog input 4.
$\overline{SS}$				I	TTL	SPI slave select input.
HLVDIN				I	Analog	High/Low-Voltage Detect input.
RA6						See the OSC2/CLKO/RA6 pin.
RA7						See the OSC1/CLKI/RA7 pin.

Legend: TTL = TTL compatible input; CMOS = CMOS compatible input or output; ST = Schmitt Trigger input with CMOS levels; I = Input; O = Output; P = Power

TABLE 4-4: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)

Register	Applicable Devices				Power-on Reset, Brown-out Reset	MCLR Resets, WDT Reset, RESET Instruction, Stack Resets	Wake-up via WDT or Interrupt
TXB0D5	2585	2680	4585	4680	xxxx xxxx	uuuu uuuu	uuuu uuuu
TXB0D4	2585	2680	4585	4680	xxxx xxxx	uuuu uuuu	uuuu uuuu
TXB0D3	2585	2680	4585	4680	xxxx xxxx	uuuu uuuu	uuuu uuuu
TXB0D2	2585	2680	4585	4680	xxxx xxxx	uuuu uuuu	uuuu uuuu
TXB0D1	2585	2680	4585	4680	xxxx xxxx	uuuu uuuu	uuuu uuuu
TXB0D0	2585	2680	4585	4680	xxxx xxxx	uuuu uuuu	uuuu uuuu
TXB0DLC	2585	2680	4585	4680	-x-- xxxx	-u-- uuuu	-u-- uuuu
TXB0EIDL	2585	2680	4585	4680	xxxx xxxx	uuuu uuuu	uuuu uuuu
TXB0EIDH	2585	2680	4585	4680	xxxx xxxx	uuuu uuuu	-uuu uuuu
TXB0SIDL	2585	2680	4585	4680	xxx- x-xx	uuu- u-uu	uuu- u-uu
TXB0SIDH	2585	2680	4585	4680	xxxx xxxx	uuuu uuuu	uuuu uuuu
TXB0CON	2585	2680	4585	4680	0000 0-00	0000 0-00	uuuu u-uu
TXB1D7	2585	2680	4585	4680	xxxx xxxx	uuuu uuuu	uuuu uuuu
TXB1D6	2585	2680	4585	4680	xxxx xxxx	uuuu uuuu	uuuu uuuu
TXB1D5	2585	2680	4585	4680	xxxx xxxx	uuuu uuuu	uuuu uuuu
TXB1D4	2585	2680	4585	4680	xxxx xxxx	uuuu uuuu	uuuu uuuu
TXB1D3	2585	2680	4585	4680	xxxx xxxx	uuuu uuuu	uuuu uuuu
TXB1D2	2585	2680	4585	4680	xxxx xxxx	uuuu uuuu	uuuu uuuu
TXB1D1	2585	2680	4585	4680	xxxx xxxx	uuuu uuuu	uuuu uuuu
TXB1D0	2585	2680	4585	4680	xxxx xxxx	uuuu uuuu	uuuu uuuu
TXB1DLC	2585	2680	4585	4680	-x-- xxxx	-u-- uuuu	-u-- uuuu
TXB1EIDL	2585	2680	4585	4680	xxxx xxxx	uuuu uuuu	uuuu uuuu
TXB1EIDH	2585	2680	4585	4680	xxxx xxxx	uuuu uuuu	uuuu uuuu
TXB1SIDL	2585	2680	4585	4680	xxx- x-xx	uuu- u-uu	uuu- uu-u
TXB1SIDH	2585	2680	4585	4680	xxxx xxxx	uuuu uuuu	-uuu uuuu
TXB1CON	2585	2680	4585	4680	0000 0-00	0000 0-00	uuuu u-uu
TXB2D7	2585	2680	4585	4680	xxxx xxxx	uuuu uuuu	0uuu uuuu
TXB2D6	2585	2680	4585	4680	xxxx xxxx	uuuu uuuu	0uuu uuuu
TXB2D5	2585	2680	4585	4680	xxxx xxxx	uuuu uuuu	0uuu uuuu
TXB2D4	2585	2680	4585	4680	xxxx xxxx	uuuu uuuu	0uuu uuuu
TXB2D3	2585	2680	4585	4680	xxxx xxxx	uuuu uuuu	0uuu uuuu
TXB2D2	2585	2680	4585	4680	xxxx xxxx	uuuu uuuu	0uuu uuuu

Legend: u = unchanged, x = unknown, - = unimplemented bit, read as '0', q = value depends on condition. Shaded cells indicate conditions do not apply for the designated device.

Note 1: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).

2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).

3: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.

4: See Table 4-3 for Reset value for specific condition.

5: Bits 6 and 7 of PORTA, LATA and TRISA are enabled, depending on the oscillator mode selected. When not enabled as PORTA pins, they are disabled and read '0'.

6: This register reads all '0's until the ECAN™ technology is set up in Mode 1 or Mode 2.

5.6.3 MAPPING THE ACCESS BANK IN INDEXED LITERAL OFFSET MODE

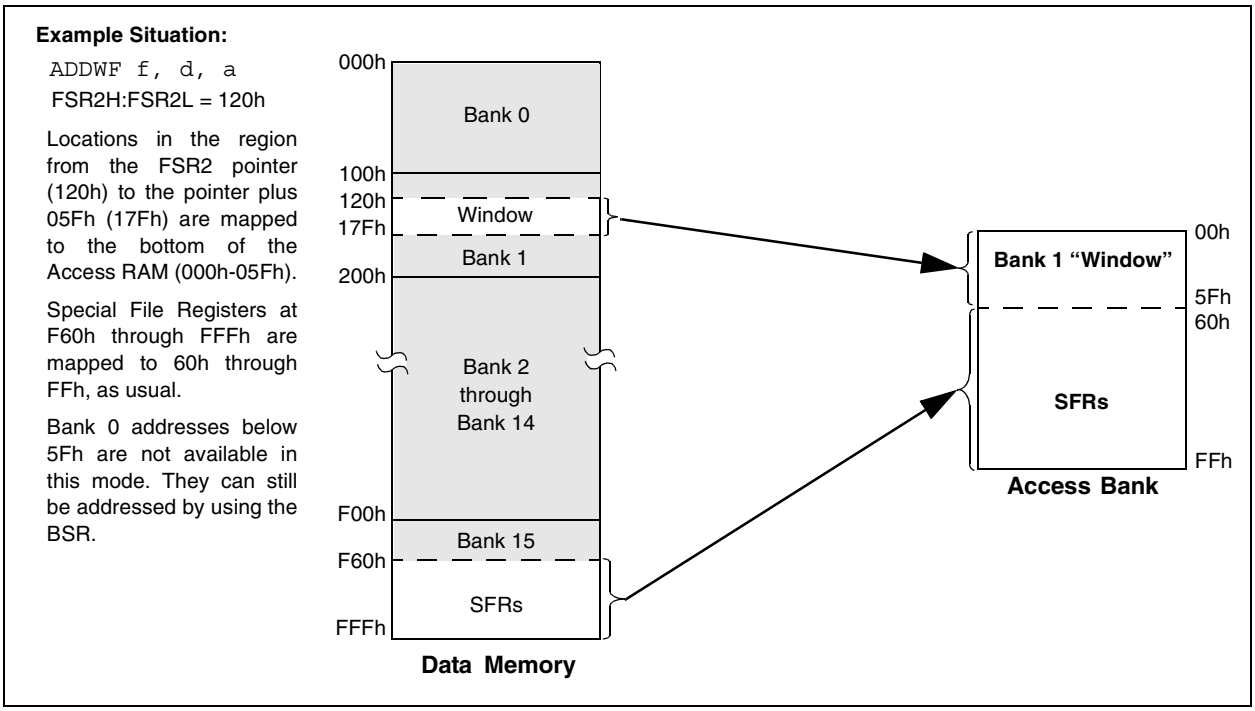
The use of Indexed Literal Offset Addressing mode effectively changes how the lower half of Access RAM (00h to 7Fh) is mapped. Rather than containing just the contents of the bottom half of Bank 0, this mode maps the contents from Bank 0 and a user defined “window” that can be located anywhere in the data memory space. The value of FSR2 establishes the lower boundary of the addresses mapped into the window, while the upper boundary is defined by FSR2 plus 95 (5Fh). Addresses in the Access RAM above 5Fh are mapped as previously described (see **Section 5.3.2 “Access Bank”**). An example of Access Bank remapping in this addressing mode is shown in Figure 5-9.

Remapping of the Access Bank applies *only* to operations using the Indexed Literal Offset mode. Operations that use the BSR (Access RAM bit is ‘1’) will continue to use direct addressing as before. Any indirect or indexed operation that explicitly uses any of the indirect file operands (including FSR2) will continue to operate as standard indirect addressing. Any instruction that uses the Access Bank, but includes a register address of greater than 05Fh, will use direct addressing and the normal Access Bank map.

5.6.4 BSR IN INDEXED LITERAL OFFSET MODE

Although the Access Bank is remapped when the extended instruction set is enabled, the operation of the BSR remains unchanged. Direct addressing using the BSR to select the data memory bank operates in the same manner as previously described.

FIGURE 5-9: REMAPPING THE ACCESS BANK WITH INDEXED LITERAL OFFSET ADDRESSING



9.0 INTERRUPTS

The PIC18F2585/2680/4585/4680 devices have multiple interrupt sources and an interrupt priority feature that allows each interrupt source to be assigned a high priority level or a low priority level. The high priority interrupt vector is at 000008h and the low priority interrupt vector is at 000018h. High priority interrupt events will interrupt any low priority interrupts that may be in progress.

There are ten registers which are used to control interrupt operation. These registers are:

- RCON
- INTCON
- INTCON2
- INTCON3
- PIR1, PIR2, PIR3
- PIE1, PIE2, PIE3
- IPR1, IPR2, IPR3

It is recommended that the Microchip header files supplied with MPLAB® IDE be used for the symbolic bit names in these registers. This allows the assembler/compiler to automatically take care of the placement of these bits within the specified register.

Each interrupt source has three bits to control its operation. The functions of these bits are:

- Flag bit to indicate that an interrupt event occurred
- Enable bit that allows program execution to branch to the interrupt vector address when the flag bit is set
- Priority bit to select high priority or low priority

The interrupt priority feature is enabled by setting the IPEN bit (RCON<7>). When interrupt priority is enabled, there are two bits which enable interrupts globally. Setting the GIEH bit (INTCON<7>) enables all interrupts that have the priority bit set (high priority). Setting the GIEL bit (INTCON<6>) enables all interrupts that have the priority bit cleared (low priority). When the interrupt flag, enable bit and appropriate global interrupt enable bit are set, the interrupt will vector immediately to address 000008h or 000018h, depending on the priority bit setting. Individual interrupts can be disabled through their corresponding enable bits.

When the IPEN bit is cleared (default state), the interrupt priority feature is disabled and interrupts are compatible with PIC® mid-range devices. In Compatibility mode, the interrupt priority bits for each source have no effect. INTCON<6> is the PEIE bit, which enables/disables all peripheral interrupt sources. INTCON<7> is the GIE bit, which enables/disables all interrupt sources. All interrupts branch to address 000008h in Compatibility mode.

When an interrupt is responded to, the global interrupt enable bit is cleared to disable further interrupts. If the IPEN bit is cleared, this is the GIE bit. If interrupt priority levels are used, this will be either the GIEH or GIEL bit. High priority interrupt sources can interrupt a low priority interrupt. Low priority interrupts are not processed while high priority interrupts are in progress.

The return address is pushed onto the stack and the PC is loaded with the interrupt vector address (000008h or 000018h). Once in the Interrupt Service Routine, the source(s) of the interrupt can be determined by polling the interrupt flag bits. The interrupt flag bits must be cleared in software before re-enabling interrupts to avoid recursive interrupts.

The “return from interrupt” instruction, `RETFIE`, exits the interrupt routine and sets the GIE bit (GIEH or GIEL if priority levels are used), which re-enables interrupts.

For external interrupt events, such as the INT pins or the PORTB input change interrupt, the interrupt latency will be three to four instruction cycles. The exact latency is the same for one or two-cycle instructions. Individual interrupt flag bits are set, regardless of the status of their corresponding enable bit or the GIE bit.

Note:	Do not use the `MOVFF` instruction to modify any of the interrupt control registers while **any** interrupt is enabled. Doing so may cause erratic microcontroller behavior.

REGISTER 9-12: IPR3: PERIPHERAL INTERRUPT PRIORITY REGISTER 3

Mode 0

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
IRXIP	WAKIP	ERRIP	TXB2IP	TXB1IP(1)	TXB0IP(1)	RXB1IP	RXB0IP

Mode 1, 2

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
IRXIP	WAKIP	ERRIP	TXBnIP	TXB1IP(1)	TXB0IP(1)	RXBnIP	FIFOWMIP
bit 7							bit 0

bit 7 **IRXIP:** CAN Invalid Received Message Interrupt Priority bit
1 = High priority
0 = Low priority

bit 6 **WAKIP:** CAN bus Activity Wake-up Interrupt Priority bit
1 = High priority
0 = Low priority

bit 5 **ERRIP:** CAN bus Error Interrupt Priority bit
1 = High priority
0 = Low priority

bit 4 When CAN is in Mode 0:
TXB2IP: CAN Transmit Buffer 2 Interrupt Priority bit
1 = High priority
0 = Low priority
When CAN is in Mode 1 or 2:
TXBnIP: CAN Transmit Buffer Interrupt Priority bit
1 = High priority
0 = Low priority

bit 3 **TXB1IP:** CAN Transmit Buffer 1 Interrupt Priority bit(1)
1 = High priority
0 = Low priority

bit 2 **TXB0IP:** CAN Transmit Buffer 0 Interrupt Priority bit(1)
1 = High priority
0 = Low priority

bit 1 When CAN is in Mode 0:
RXB1IP: CAN Receive Buffer 1 Interrupt Priority bit
1 = High priority
0 = Low priority
When CAN is in Mode 1 or 2:
RXBnIP: CAN Receive Buffer Interrupts Priority bit
1 = High priority
0 = Low priority

bit 0 When CAN is in Mode 0:
RXB0IP: CAN Receive Buffer 0 Interrupt Priority bit
1 = High priority
0 = Low priority
When CAN is in Mode 1:
Unimplemented: Read as '0'
When CAN is in Mode 2:
FIFOWMIP: FIFO Watermark Interrupt Priority bit
1 = High priority
0 = Low priority

Note 1: In CAN Mode 1 and 2, this bit is forced to '0'.

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

NOTES:

TABLE 16-3: REGISTERS ASSOCIATED WITH ECCP1 MODULE AND TIMER1 TO TIMER3

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	49
RCON	IPEN	SBOREN	—	$\overline{RI}$	$\overline{TO}$	$\overline{PD}$	$\overline{POR}$	$\overline{BOR}$	50
IPR1	PSPIP	ADIP	RCIP	TXIP	SSPIP	CCP1IP	TMR2IP	TMR1IP	52
PIR1	PSPIF	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	52
PIE1	PSPIE	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	52
IPR2	OSCFIP	CMIP(3)	—	EEIP	BCLIP	HLVDIP	TMR3IP	ECCP1IP(3)	51
PIR2	OSCFIF	CMIF(3)	—	EEIF	BCLIF	HLVDIF	TMR3IF	ECCP1IF(3)	51
PIE2	OSCFIE	CMIE(3)	—	EEIE	BCLIE	HLVDIE	TMR3IE	ECCP1IE(3)	52
TRISB	PORTB Data Direction Register								52
TRISC	PORTC Data Direction Register								52
TRISD(1)	PORTD Data Direction Register								52
TMR1L	Holding Register for the Least Significant Byte of the 16-bit TMR1 Register								50
TMR1H	Holding Register for the Most Significant Byte of the 16-bit TMR1 Register								50
T1CON	RD16	T1RUN	T1CKPS1	T1CKPS0	T1OSCEN	$\overline{T1SYNC}$	TMR1CS	TMR1ON	50
TMR2	Timer2 Module Register								50
T2CON	—	T2OUTPS3	T2OUTPS2	T2OUTPS1	T2OUTPS0	TMR2ON	T2CKPS1	T2CKPS0	50
PR2	Timer2 Period Register								50
TMR3L	Holding Register for the Least Significant Byte of the 16-bit TMR3 Register								51
TMR3H	Holding Register for the Most Significant Byte of the 16-bit TMR3 Register								51
T3CON	RD16	T3ECCP1(1)	T3CKPS1	T3CKPS0	T3CCP1(1)	$\overline{T3SYNC}$	TMR3CS	TMR3ON	51
ECCPR1L(2)	Enhanced Capture/Compare/PWM Register 1 (LSB)								51
ECCPR1H(2)	Enhanced Capture/Compare/PWM Register 1 (MSB)								51
ECCP1CON(2)	EPWM1M1	EPWM1M0	EDC1B1	EDC1B0	ECCP1M3	ECCP1M2	ECCP1M1	ECCP1M0	51
ECCP1AS(2)	ECCPASE	ECCPAS2	ECCPAS1	ECCPAS0	PSSAC1	PSSAC0	PSSBD1(2)	PSSBD0(2)	51
ECCP1DEL(2)	PRSEN	PDC6(2)	PDC5(2)	PDC4(2)	PDC3(2)	PDC2(2)	PDC1(2)	PDC0(2)	51

Legend: — = unimplemented, read as '0'. Shaded cells are not used during ECCP1 operation.

Note 1: These bits are available on PIC18F4X8X devices only.

2: These bits or registers are unimplemented in PIC18F2X8X devices; always maintain these bit clear.

3: These bits are available on PIC18F4X8X and reserved on PIC18F2X8X devices.

TABLE 18-3: BAUD RATES FOR ASYNCHRONOUS MODES (CONTINUED)

BAUD RATE (K)	SYNC = 0, BRGH = 0, BRG16 = 1											
	FOSC = 40.000 MHz			FOSC = 20.000 MHz			FOSC = 10.000 MHz			FOSC = 8.000 MHz		
	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)
0.3	0.300	0.00	8332	0.300	0.02	4165	0.300	0.02	2082	300	-0.04	1665
1.2	1.200	0.02	2082	1.200	-0.03	1041	1.200	-0.03	520	1201	-0.16	415
2.4	2.402	0.06	1040	2.399	-0.03	520	2.404	0.16	259	2403	-0.16	207
9.6	9.615	0.16	259	9.615	0.16	129	9.615	0.16	64	9615	-0.16	51
19.2	19.231	0.16	129	19.231	0.16	64	19.531	1.73	31	19230	-0.16	25
57.6	58.140	0.94	42	56.818	-1.36	21	56.818	-1.36	10	55555	3.55	8
115.2	113.636	-1.36	21	113.636	-1.36	10	125.000	8.51	4	—	—	—

BAUD RATE (K)	SYNC = 0, BRGH = 0, BRG16 = 1								
	FOSC = 4.000 MHz			FOSC = 2.000 MHz			FOSC = 1.000 MHz		
	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)
0.3	0.300	0.04	832	300	-0.16	415	300	-0.16	207
1.2	1.202	0.16	207	1201	-0.16	103	1201	-0.16	51
2.4	2.404	0.16	103	2403	-0.16	51	2403	-0.16	25
9.6	9.615	0.16	25	9615	-0.16	12	—	—	—
19.2	19.231	0.16	12	—	—	—	—	—	—
57.6	62.500	8.51	3	—	—	—	—	—	—
115.2	125.000	8.51	1	—	—	—	—	—	—

BAUD RATE (K)	SYNC = 0, BRGH = 1, BRG16 = 1 or SYNC = 1, BRG16 = 1											
	FOSC = 40.000 MHz			FOSC = 20.000 MHz			FOSC = 10.000 MHz			FOSC = 8.000 MHz		
	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)
0.3	0.300	0.00	33332	0.300	0.00	16665	0.300	0.00	8332	300	-0.01	6665
1.2	1.200	0.00	8332	1.200	0.02	4165	1.200	0.02	2082	1200	-0.04	1665
2.4	2.400	0.02	4165	2.400	0.02	2082	2.402	0.06	1040	2400	-0.04	832
9.6	9.606	0.06	1040	9.596	-0.03	520	9.615	0.16	259	9615	-0.16	207
19.2	19.193	-0.03	520	19.231	0.16	259	19.231	0.16	129	19230	-0.16	103
57.6	57.803	0.35	172	57.471	-0.22	86	58.140	0.94	42	57142	0.79	34
115.2	114.943	-0.22	86	116.279	0.94	42	113.636	-1.36	21	117647	-2.12	16

BAUD RATE (K)	SYNC = 0, BRGH = 1, BRG16 = 1 or SYNC = 1, BRG16 = 1								
	FOSC = 4.000 MHz			FOSC = 2.000 MHz			FOSC = 1.000 MHz		
	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)
0.3	0.300	0.01	3332	300	-0.04	1665	300	-0.04	832
1.2	1.200	0.04	832	1201	-0.16	415	1201	-0.16	207
2.4	2.404	0.16	415	2403	-0.16	207	2403	-0.16	103
9.6	9.615	0.16	103	9615	-0.16	51	9615	-0.16	25
19.2	19.231	0.16	51	19230	-0.16	25	19230	-0.16	12
57.6	58.824	2.12	16	55555	3.55	8	—	—	—
115.2	111.111	-3.55	8	—	—	—	—	—	—

18.2.2 EUSART ASYNCHRONOUS RECEIVER

The receiver block diagram is shown in Figure 18-6. The data is received on the RX pin and drives the data recovery block. The data recovery block is actually a high-speed shifter operating at x16 times the baud rate, whereas the main receive serial shifter operates at the bit rate or at FOSC. This mode would typically be used in RS-232 systems.

To set up an Asynchronous Reception:

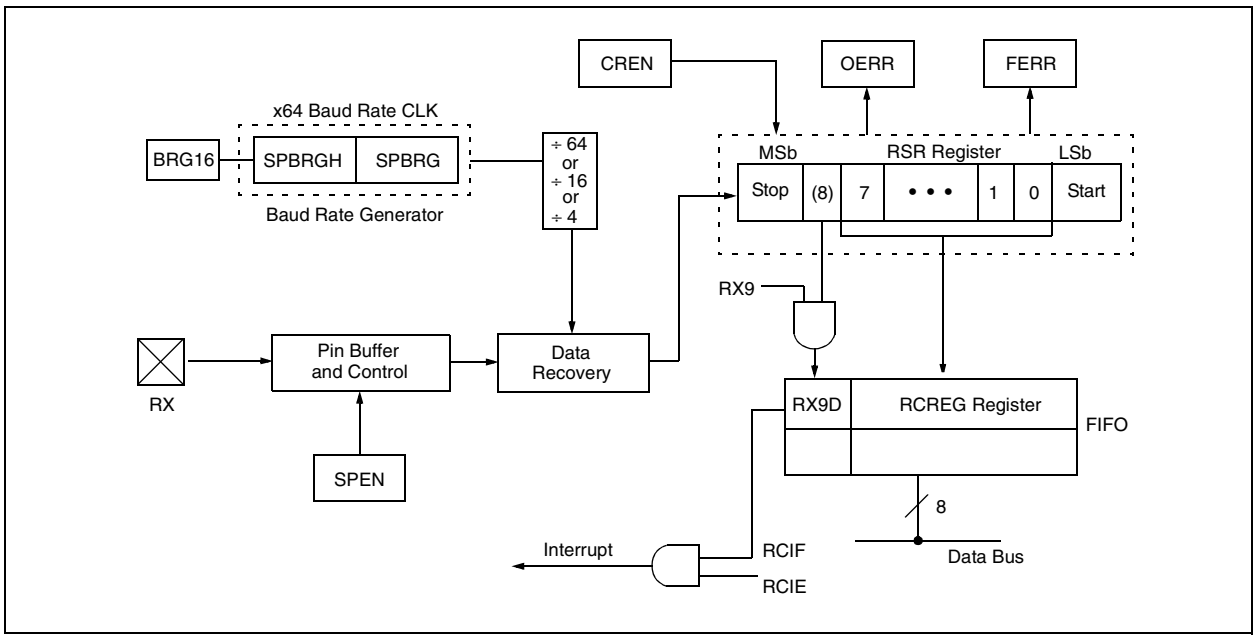
1. Initialize the SPBRGH:SPBRG registers for the appropriate baud rate. Set or clear the BRGH and BRG16 bits, as required, to achieve the desired baud rate.
2. Enable the asynchronous serial port by clearing bit SYNC and setting bit SPEN.
3. If interrupts are desired, set enable bit RCIE.
4. If 9-bit reception is desired, set bit RX9.
5. Enable the reception by setting bit CREN.
6. Flag bit RCIF will be set when reception is complete and an interrupt will be generated if enable bit RCIE was set.
7. Read the RCSTA register to get the 9th bit (if enabled) and determine if any error occurred during reception.
8. Read the 8-bit received data by reading the RCREG register.
9. If any error occurred, clear the error by clearing enable bit CREN.
10. If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

18.2.3 SETTING UP 9-BIT MODE WITH ADDRESS DETECT

This mode would typically be used in RS-485 systems. To set up an Asynchronous Reception with Address Detect Enable:

1. Initialize the SPBRGH:SPBRG registers for the appropriate baud rate. Set or clear the BRGH and BRG16 bits, as required, to achieve the desired baud rate.
2. Enable the asynchronous serial port by clearing the SYNC bit and setting the SPEN bit.
3. If interrupts are required, set the RCEN bit and select the desired priority level with the RCIP bit.
4. Set the RX9 bit to enable 9-bit reception.
5. Set the ADDEN bit to enable address detect.
6. Enable reception by setting the CREN bit.
7. The RCIF bit will be set when reception is complete. The interrupt will be Acknowledged if the RCIE and GIE bits are set.
8. Read the RCSTA register to determine if any error occurred during reception, as well as read bit 9 of data (if applicable).
9. Read RCREG to determine if the device is being addressed.
10. If any error occurred, clear the CREN bit.
11. If the device has been addressed, clear the ADDEN bit to allow all received data into the receive buffer and interrupt the CPU.

FIGURE 18-6: EUSART RECEIVE BLOCK DIAGRAM



18.2.5 BREAK CHARACTER SEQUENCE

The Enhanced EUSART module has the capability of sending the special Break character sequences that are required by the LIN bus standard. The Break character transmit consists of a Start bit, followed by twelve '0' bits and a Stop bit. The frame Break character is sent whenever the SENDB and TXEN bits (TXSTA<3> and TXSTA<5>) are set while the Transmit Shift register is loaded with data. Note that the value of data written to TXREG will be ignored and all '0's will be transmitted.

The SENDB bit is automatically reset by hardware after the corresponding Stop bit is sent. This allows the user to preload the transmit FIFO with the next transmit byte following the Break character (typically, the Sync character in the LIN specification).

Note that the data value written to the TXREG for the Break character is ignored. The write simply serves the purpose of initiating the proper sequence.

The TRMT bit indicates when the transmit operation is active or Idle, just as it does during normal transmission. See Figure 18-10 for the timing of the Break character sequence.

18.2.5.1 Break and Sync Transmit Sequence

The following sequence will send a message frame header made up of a Break, followed by an Auto-Baud Sync byte. This sequence is typical of a LIN bus master.

1. Configure the EUSART for the desired mode.
2. Set the TXEN and SENDB bits to set up the Break character.
3. Load the TXREG with a dummy character to initiate transmission (the value is ignored).
4. Write '55h' to TXREG to load the Sync character into the transmit FIFO buffer.
5. After the Break has been sent, the SENDB bit is reset by hardware. The Sync character now transmits in the preconfigured mode.

When the TXREG becomes empty, as indicated by the TXIF, the next data byte can be written to TXREG.

18.2.6 RECEIVING A BREAK CHARACTER

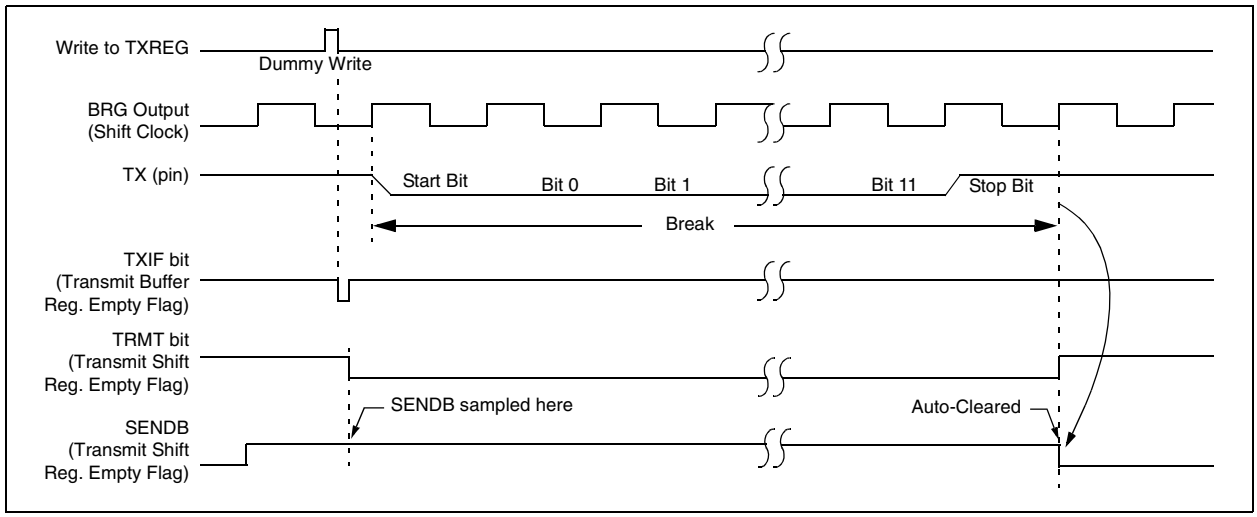
The Enhanced USART module can receive a Break character in two ways.

The first method forces configuration of the baud rate at a frequency of 9/13 the typical speed. This allows for the Stop bit transition to be at the correct sampling location (13 bits for Break versus Start bit and 8 data bits for typical data).

The second method uses the auto-wake-up feature described in **Section 18.2.4 "Auto-Wake-up on Sync Break Character"**. By enabling this feature, the EUSART will sample the next two transitions on RX/DT, cause an RCIF interrupt and receive the next data byte followed by another interrupt.

Note that following a Break character, the user will typically want to enable the Auto-Baud Rate Detect feature. For both methods, the user can set the ABD bit once the TXIF interrupt is observed.

FIGURE 18-10: SEND BREAK CHARACTER SEQUENCE



18.4 EUSART Synchronous Slave Mode

Synchronous Slave mode is entered by clearing bit, CSRC (TXSTA<7>). This mode differs from the Synchronous Master mode in that the shift clock is supplied externally at the CK pin (instead of being supplied internally in Master mode). This allows the device to transfer or receive data while in any low-power mode.

18.4.1 EUSART SYNCHRONOUS SLAVE TRANSMIT

The operation of the Synchronous Master and Slave modes are identical, except in the case of the Sleep mode.

If two words are written to the TXREG and then the SLEEP instruction is executed, the following will occur:

a) The first word will immediately transfer to the TSR register and transmit.
b) The second word will remain in the TXREG register.
c) Flag bit TXIF will not be set.
d) When the first word has been shifted out of TSR, the TXREG register will transfer the second word to the TSR and flag bit TXIF will now be set.
e) If enable bit TXIE is set, the interrupt will wake the chip from Sleep. If the global interrupt is enabled, the program will branch to the interrupt vector.

To set up a Synchronous Slave Transmission:

1. Enable the synchronous slave serial port by setting bits SYNC and SPEN and clearing bit CSRC.
2. Clear bits CREN and SREN.
3. If interrupts are desired, set enable bit TXIE.
4. If 9-bit transmission is desired, set bit TX9.
5. Enable the transmission by setting enable bit TXEN.
6. If 9-bit transmission is selected, the ninth bit should be loaded in bit TX9D.
7. Start transmission by loading data to the TXREGx register.
8. If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

TABLE 18-9: REGISTERS ASSOCIATED WITH SYNCHRONOUS SLAVE TRANSMISSION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	49
PIR1	PSPIF(1)	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	52
PIE1	PSPIE(1)	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	52
IPR1	PSPIP(1)	ADIP	RCIP	TXIP	SSPIP	CCP1IP	TMR2IP	TMR1IP	52
RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	51
TXREG	EUSART Transmit Register								51
TXSTA	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	51
BAUDCON	ABDOVF	RCIDL	—	SCKP	BRG16	—	WUE	ABDEN	51
SPBRGH	EUSART Baud Rate Generator Register High Byte								51
SPBRG	EUSART Baud Rate Generator Register Low Byte								51

Legend: — = unimplemented, read as '0'. Shaded cells are not used for synchronous slave transmission.

Note 1: Reserved in PIC18F2X8X devices; always maintain these bits clear.

REGISTER 19-3: ADCON2: A/D CONTROL REGISTER 2

R/W-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
ADFM	—	ACQT2	ACQT1	ACQT0	ADCS2	ADCS1	ADCS0
bit 7							bit 0

bit 7 **ADFM:** A/D Result Format Select bit
1 = Right justified
0 = Left justified

bit 6 **Unimplemented:** Read as '0'

bit 5-3 **ACQT2:ACQT0:** A/D Acquisition Time Select bits
111 = 20 TAD
110 = 16 TAD
101 = 12 TAD
100 = 8 TAD
011 = 6 TAD
010 = 4 TAD
001 = 2 TAD
000 = 0 TAD(1)

bit 2-0 **ADCS2:ADCS0:** A/D Conversion Clock Select bits
111 = FRC (clock derived from A/D RC oscillator)(1)
110 = FOSC/64
101 = FOSC/16
100 = FOSC/4
011 = FRC (clock derived from A/D RC oscillator)(1)
010 = FOSC/32
001 = FOSC/8
000 = FOSC/2

Note 1: If the A/D FRC clock source is selected, a delay of one TCY (instruction cycle) is added before the A/D clock starts. This allows the SLEEP instruction to be executed before starting a conversion.

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

EXAMPLE 23-2: WIN AND ICODE BITS USAGE IN INTERRUPT SERVICE ROUTINE TO ACCESS TX/RX BUFFERS (CONTINUED)

```
ErrorInterrupt
    BCF     PIR3, ERRIF                 ; Clear the interrupt flag
    …                                   ; Handle error.
    RETFIE
TXB2Interrupt
    BCF     PIR3, TXB2IF                ; Clear the interrupt flag
    GOTO    AccessBuffer
TXB1Interrupt
    BCF     PIR3, TXB1IF                ; Clear the interrupt flag
    GOTO    AccessBuffer
TXB0Interrupt
    BCF     PIR3, TXB0IF                ; Clear the interrupt flag
    GOTO    AccessBuffer
RXB1Interrupt
    BCF     PIR3, RXB1IF                ; Clear the interrupt flag
    GOTO    Accessbuffer
RXB0Interrupt
    BCF     PIR3, RXB0IF                ; Clear the interrupt flag
    GOTO    AccessBuffer
AccessBuffer                            ; This is either TX or RX interrupt
    ; Copy CANSTAT.ICODE bits to CANCON.WIN bits
    MOVF    TempCANCON, W               ; Clear CANCON.WIN bits before copying
                                        ; new ones.
    ANDLW   B'11110001'                 ; Use previously saved CANCON value to
                                        ; make sure same value.
    MOVWF   TempCANCON                  ; Copy masked value back to TempCANCON
    MOVF    TempCANSTAT, W              ; Retrieve ICODE bits
    ANDLW   B'00001110'                 ; Use previously saved CANSTAT value
                                        ; to make sure same value.
    IORWF   TempCANCON                  ; Copy ICODE bits to WIN bits.
    MOVFF   TempCANCON, CANCON          ; Copy the result to actual CANCON
    ; Access current buffer…
    ; User code
    ; Restore CANCON.WIN bits
    MOVF    CANCON, W                   ; Preserve current non WIN bits
    ANDLW   B'11110001'
    IORWF   TempCANCON                  ; Restore original WIN bits
    ; Do not need to restore CANSTAT - it is read-only register.
    ; Return from interrupt or check for another module interrupt source
```

23.2.2 DEDICATED CAN TRANSMIT BUFFER REGISTERS

This section describes the dedicated CAN Transmit Buffer registers and their associated control registers.

REGISTER 23-5: TXBnCON: TRANSMIT BUFFER n CONTROL REGISTERS [0 ≤ n ≤ 2]

Mode 0

R/C-0	R-0	R-0	R-0	R/W-0	U-0	R/W-0	R/W-0
TXBIF	TXABT(1)	TXLARB(1)	TXERR(1)	TXREQ(2)	—	TXPRI1(3)	TXPRI0(3)

Mode 1, 2

R/C-0	R-0	R-0	R-0	R/W-0	U-0	R/W-0	R/W-0
TXBIF	TXABT(1)	TXLARB(1)	TXERR(1)	TXREQ(2)	—	TXPRI1(3)	TXPRI0(3)
bit 7							bit 0

bit 7 **TXBIF:** Transmit Buffer Interrupt Flag bit
1 = Transmit buffer has completed transmission of message and may be reloaded
0 = Transmit buffer has not completed transmission of a message

bit 6 **TXABT:** Transmission Aborted Status bit(1)
1 = Message was aborted
0 = Message was not aborted

bit 5 **TXLARB:** Transmission Lost Arbitration Status bit(1)
1 = Message lost arbitration while being sent
0 = Message did not lose arbitration while being sent

bit 4 **TXERR:** Transmission Error Detected Status bit(1)
1 = A bus error occurred while the message was being sent
0 = A bus error did not occur while the message was being sent

bit 3 **TXREQ:** Transmit Request Status bit(2)
1 = Requests sending a message. Clears the TXABT, TXLARB and TXERR bits.
0 = Automatically cleared when the message is successfully sent

bit 2 **Unimplemented:** Read as '0'

bit 1-0 **TXPRI1:TXPRI0:** Transmit Priority bits(3)
11 = Priority Level 3 (highest priority)
10 = Priority Level 2
01 = Priority Level 1
00 = Priority Level 0 (lowest priority)

Note 1: This bit is automatically cleared when TXREQ is set.

2: While TXREQ is set, Transmit Buffer registers remain read-only. Clearing this bit in software while the bit is set will request a message abort.

3: These bits define the order in which transmit buffers will be transferred. They do not alter the CAN message identifier.

Legend:			
C = Clearable bit	R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

23.9.2 TIME QUANTA

As already mentioned, the Time Quanta is a fixed unit derived from the oscillator period and baud rate prescaler. Its relationship to TBIT and the Nominal Bit Rate is shown in Example 23-6.

EXAMPLE 23-6: CALCULATING TQ, NOMINAL BIT RATE AND NOMINAL BIT TIME

TQ (µs) = (2 * (BRP + 1))/FOSC (MHz)

TBIT (µs) = TQ (µs) * number of TQ per bit interval

Nominal Bit Rate (bits/s) = 1/TBIT

This frequency (FOSC) refers to the effective frequency used. If, for example, a 10 MHz external signal is used along with a PLL, then the effective frequency will be 4 x 10 MHz which equals 40 MHz.

CASE 1:

For FOSC = 16 MHz, BRP<5:0> = 00h and Nominal Bit Time = 8 TQ:

TQ = (2 * 1)/16 = 0.125 µs (125 ns)

TBIT = 8 * 0.125 = 1 µs (10^{-6}s)

Nominal Bit Rate = $1/10^{-6} = 10^6$ bits/s (1 Mb/s)

CASE 2:

For FOSC = 20 MHz, BRP<5:0> = 01h and Nominal Bit Time = 8 TQ:

TQ = (2 * 2)/20 = 0.2 µs (200 ns)

TBIT = 8 * 0.2 = 1.6 µs ($1.6 * 10^{-6}$s)

Nominal Bit Rate = $1/1.6 * 10^{-6}$s = 625,000 bits/s (625 Kb/s)

CASE 3:

For FOSC = 25 MHz, BRP<5:0> = 3Fh and Nominal Bit Time = 25 TQ:

TQ = (2 * 64)/25 = 5.12 µs

TBIT = 25 * 5.12 = 128 µs ($1.28 * 10^{-4}$s)

Nominal Bit Rate = $1/1.28 * 10^{-4}$ = 7813 bits/s (7.8 Kb/s)

The frequencies of the oscillators in the different nodes must be coordinated in order to provide a system wide specified nominal bit time. This means that all oscillators must have a TOSC that is an integral divisor of TQ. It should also be noted that although the number of TQ is programmable from 4 to 25, the usable minimum is 8 TQ. There is no assurance that a bit time of less than 8 TQ in length will operate correctly.

23.9.3 SYNCHRONIZATION SEGMENT

This part of the bit time is used to synchronize the various CAN nodes on the bus. The edge of the input signal is expected to occur during the sync segment. The duration is 1 TQ.

23.9.4 PROPAGATION SEGMENT

This part of the bit time is used to compensate for physical delay times within the network. These delay times consist of the signal propagation time on the bus line and the internal delay time of the nodes. The length of the Propagation Segment can be programmed from 1 TQ to 8 TQ by setting the PRSEG2:PRSEG0 bits.

23.9.5 PHASE BUFFER SEGMENTS

The phase buffer segments are used to optimally locate the sampling point of the received bit within the nominal bit time. The sampling point occurs between Phase Segment 1 and Phase Segment 2. These segments can be lengthened or shortened by the resynchronization process. The end of Phase Segment 1 determines the sampling point within a bit time. Phase Segment 1 is programmable from 1 TQ to 8 TQ in duration. Phase Segment 2 provides delay before the next transmitted data transition and is also programmable from 1 TQ to 8 TQ in duration. However, due to IPT requirements, the actual minimum length of Phase Segment 2 is 2 TQ, or it may be defined to be equal to the greater of Phase Segment 1 or the Information Processing Time (IPT). The sampling point should be as late as possible or approximately 80% of the bit time.

23.9.6 SAMPLE POINT

The sample point is the point of time at which the bus level is read and the value of the received bit is determined. The sampling point occurs at the end of Phase Segment 1. If the bit timing is slow and contains many TQ, it is possible to specify multiple sampling of the bus line at the sample point. The value of the received bit is determined to be the value of the majority decision of three values. The three samples are taken at the sample point and twice before, with a time of TQ/2 between each sample.

23.9.7 INFORMATION PROCESSING TIME

The Information Processing Time (IPT) is the time segment starting at the sample point that is reserved for calculation of the subsequent bit level. The CAN specification defines this time to be less than or equal to 2 TQ. The PIC18F2585/2680/4585/4680 devices define this time to be 2 TQ. Thus, Phase Segment 2 must be at least 2 TQ long.

27.2 DC Characteristics: Power-Down and Supply Current PIC18F2585/2680/4585/4680 (Industrial) PIC18LF2585/2680/4585/4680 (Industrial) (Continued)

PIC18LF2585/2680/4585/4680 (Industrial)		**Standard Operating Conditions (unless otherwise stated)** Operating temperature -40°C ≤ TA ≤ +85°C for industrial					
PIC18F2585/2680/4585/4680 (Industrial, Extended)		**Standard Operating Conditions (unless otherwise stated)** Operating temperature -40°C ≤ TA ≤ +85°C for industrial -40°C ≤ TA ≤ +125°C for extended					
Param No.	**Device**	**Typ**	**Max**	**Units**	**Conditions**		
	Supply Current (IDD)(2,3)						
	PIC18LFX585/X680	410.00	550.00	µA	-40°C	VDD = 2.0V	FOSC = 1 MHz (**PRI_RUN**, EC oscillator)
		420.00	550.00	µA	+25°C		
		420.00	550.00	µA	+85°C		
	PIC18LFX585/X680	0.87	0.88	mA	-40°C	VDD = 3.0V	
		0.77	0.88	mA	+25°C		
		0.72	0.88	mA	+85°C		
	All devices	1.90	3.00	mA	-40°C	VDD = 5.0V	
		1.60	3.00	mA	+25°C		
		1.50	3.00	mA	+85°C		
	PIC18FX585/X680	1.50	3.30	mA	+125°C		
	PIC18LFX585/X680	1.40	2.20	mA	-40°C	VDD = 2.0V	FOSC = 4 MHz (**PRI_RUN**, EC oscillator)
		1.40	2.20	mA	+25°C		
		1.40	2.20	mA	+85°C		
	PIC18LFX585/X680	2.30	3.30	mA	-40°C	VDD = 3.0V	
		2.30	3.30	mA	+25°C		
		2.30	3.30	mA	+85°C		
	All devices	4.50	6.60	mA	-40°C	VDD = 5.0V	
		4.30	6.60	mA	+25°C		
		4.30	6.60	mA	+85°C		
	PIC18FX585/X680	5.00	7.70	mA	+125°C		

Legend: Shading of rows is to assist in readability of the table.

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to VDD or VSS and all features that add delta current disabled (such as WDT, Timer1 Oscillator, BOR, etc.).

2: The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption. The test conditions for all IDD measurements in active operation mode are:
OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;
$\overline{MCLR}$ = VDD; WDT enabled/disabled as specified.

3: For RC oscillator configurations, current through REXT is not included. The current through the resistor can be estimated by the formula Ir = VDD/2REXT (mA) with REXT in kΩ.

4: Standard low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.

TABLE 27-7: PLL CLOCK TIMING SPECIFICATIONS (V_{DD} = 4.2V TO 5.5V)

Param No.	Sym	Characteristic	Min	Typ†	Max	Units	Conditions
F10	F_{OSC}	Oscillator Frequency Range	4	—	10	MHz	HS mode only
F11	F_{SYS}	On-Chip VCO System Frequency	16	—	40	MHz	HS mode only
F12	t_{rc}	PLL Start-up Time (Lock Time)	—	—	2	ms	
F13	Δ_{CLK}	CLKO Stability (Jitter)	-2	—	+2	%	

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

TABLE 27-8: AC CHARACTERISTICS: INTERNAL RC ACCURACY
PIC18F2585/2680/4585/4680 (INDUSTRIAL)
PIC18LF2585/2680/4585/4680 (INDUSTRIAL)

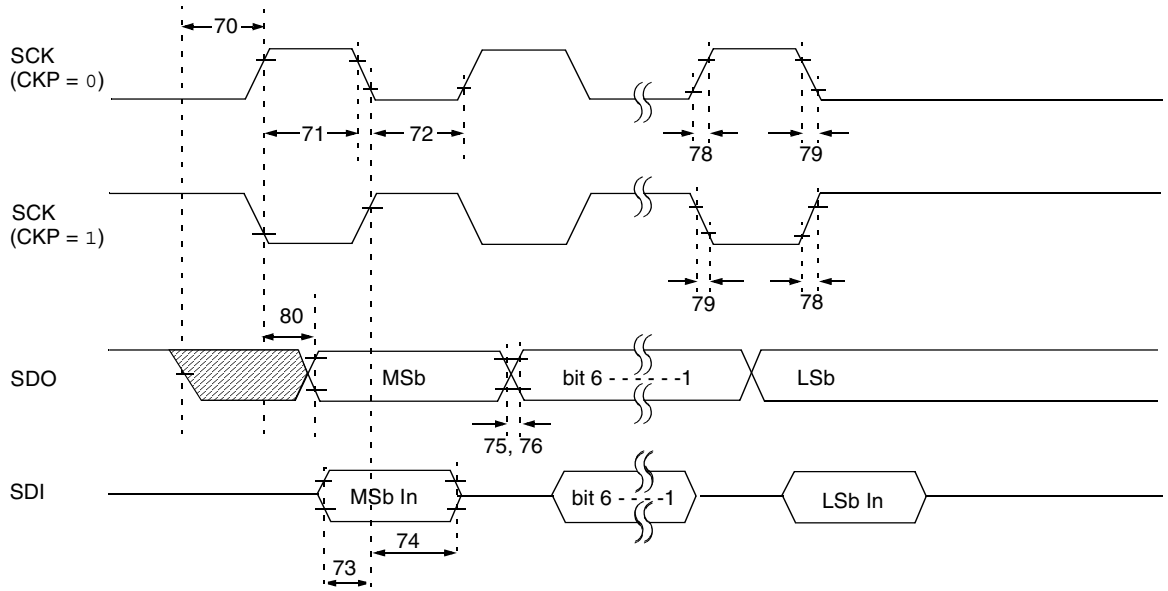
PIC18LF2585/2680/4585/4680 (Industrial)	**Standard Operating Conditions (unless otherwise stated)** Operating temperature -40°C ≤ T_A ≤ +85°C for industrial						
PIC18F2585/2680/4585/4680 (Industrial)	**Standard Operating Conditions (unless otherwise stated)** Operating temperature -40°C ≤ T_A ≤ +85°C for industrial						
Param No.	**Device**	**Min**	**Typ**	**Max**	**Units**	**Conditions**	
	INTOSC Accuracy @ Freq = 8 MHz, 4 MHz, 2 MHz, 1 MHz, 500 kHz, 250 kHz, 125 kHz(1)						
	PIC18LFX585/X680	-2	+/-1	2	%	+25°C	V_{DD} = 2.7-3.3V
		-5	—	5	%	-10°C to +85°C	V_{DD} = 2.7-3.3V
		-10	+/-1	10	%	-40°C to +85°C	V_{DD} = 2.7-3.3V
	PIC18FX585/X680	-2	+/-1	2	%	+25°C	V_{DD} = 4.5-5.5V
		-5	—	5	%	-10°C to +85°C	V_{DD} = 4.5-5.5V
		-10	+/-1	10	%	-40°C to +85°C	V_{DD} = 4.5-5.5V
	INTRC Accuracy @ Freq = 31 kHz(2)						
	PIC18LFX585/X680	26.562	—	35.938	kHz	-40°C to +85°C	V_{DD} = 2.7-3.3V
	PIC18FX585/X680	26.562	—	35.938	kHz	-40°C to +85°C	V_{DD} = 4.5-5.5V

Legend: Shading of rows is to assist in readability of the table.

Note 1: Frequency calibrated at 25°C. OSCTUNE register can be used to compensate for temperature drift.

2: INTRC frequency after calibration.

FIGURE 27-12: EXAMPLE SPI MASTER MODE TIMING (CKE = 0)



Note: Refer to Figure 27-4 for load conditions.

TABLE 27-14: EXAMPLE SPI MODE REQUIREMENTS (MASTER MODE, CKE = 0)

Param No.	Symbol	Characteristic		Min	Max	Units	Conditions
73	TDIV2SCH, TDIV2SCL	Setup Time of SDI Data Input to SCK Edge		100	—	ns	
74	TSCH2DIL, TSCL2DIL	Hold Time of SDI Data Input to SCK Edge		100	—	ns	
75	TDOR	SDO Data Output Rise Time	PIC18**F**XXXX	—	25	ns	
			PIC18**LF**XXXX	—	45	ns	VDD = 2.0V
76	TDOF	SDO Data Output Fall Time		—	25	ns	
78	TSCR	SCK Output Rise Time	PIC18**F**XXXX	—	25	ns	
			PIC18**LF**XXXX	—	45	ns	VDD = 2.0V
79	TSCF	SCK Output Fall Time		—	25	ns	
80	TSCH2DOV, TSCL2DOV	SDO Data Output Valid after SCK Edge	PIC18**F**XXXX	—	50	ns	
			PIC18**LF**XXXX	—	100	ns	VDD = 2.0V

V

W

X