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Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details

Product Status	Active
Number of LABs/CLBs	896
Number of Logic Elements/Cells	8064
Total RAM Bits	294912
Number of I/O	97
Number of Gates	400000
Voltage - Supply	1.14V ~ 1.26V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 100°C (TJ)
Package / Case	144-LQFP
Supplier Device Package	144-TQFP (20x20)
Purchase URL	https://www.e-xfl.com/product-detail/xilinx/xc3s400-4tq144i

According to [Figure 7](#), the clock line OTCLK1 connects the CK inputs of the upper registers on the output and three-state paths. Similarly, OTCLK2 connects the CK inputs for the lower registers on the output and three-state paths. The upper and lower registers on the input path have independent clock lines: ICLK1 and ICLK2. The enable line OCE connects the CE inputs of the upper and lower registers on the output path. Similarly, TCE connects the CE inputs for the register pair on the three-state path and ICE does the same for the register pair on the input path. The Set/Reset (SR) line entering the IOB is common to all six registers, as is the Reverse (REV) line.

Each storage element supports numerous options in addition to the control over signal polarity described in the IOB Overview section. These are described in [Table 6](#).

Table 6: Storage Element Options

Option Switch	Function	Specificity
FF/Latch	Chooses between an edge-sensitive flip-flop or a level-sensitive latch	Independent for each storage element.
SYNC/ASYN	Determines whether SR is synchronous or asynchronous	Independent for each storage element.
SRHIGH/SRLOW	Determines whether SR acts as a Set, which forces the storage element to a logic "1" (SRHIGH) or a Reset, which forces a logic "0" (SRLOW).	Independent for each storage element, except when using FDDR. In the latter case, the selection for the upper element (OFF1 or TFF2) applies to both elements.
INIT1/INIT0	In the event of a Global Set/Reset, after configuration or upon activation of the GSR net, this switch decides whether to set or reset a storage element. By default, choosing SRLOW also selects INIT0; choosing SRHIGH also selects INIT1.	Independent for each storage element, except when using FDDR. In the latter case, selecting INIT0 for one element applies to both elements (even though INIT1 is selected for the other).

Double-Data-Rate Transmission

Double-Data-Rate (DDR) transmission describes the technique of synchronizing signals to both the rising and falling edges of the clock signal. Spartan-3 devices use register-pairs in all three IOB paths to perform DDR operations.

The pair of storage elements on the IOB's Output path (OFF1 and OFF2), used as registers, combine with a special multiplexer to form a DDR D-type flip-flop (FDDR). This primitive permits DDR transmission where output data bits are synchronized to both the rising and falling edges of a clock. It is possible to access this function by placing either an FDDRSE or an FDDRCPE component or symbol into the design. DDR operation requires two clock signals (50% duty cycle), one the inverted form of the other. These signals trigger the two registers in alternating fashion, as shown in [Figure 8](#). Commonly, the Digital Clock Manager (DCM) generates the two clock signals by mirroring an incoming signal, then shifting it 180 degrees. This approach ensures minimal skew between the two signals.

The storage-element-pair on the Three-State path (TFF1 and TFF2) can also be combined with a local multiplexer to form an FDDR primitive. This permits synchronizing the output enable to both the rising and falling edges of a clock. This DDR operation is realized in the same way as for the output path.

The storage-element-pair on the input path (IFF1 and IFF2) allows an I/O to receive a DDR signal. An incoming DDR clock signal triggers one register and the inverted clock signal triggers the other register. In this way, the registers take turns capturing bits of the incoming DDR data signal.

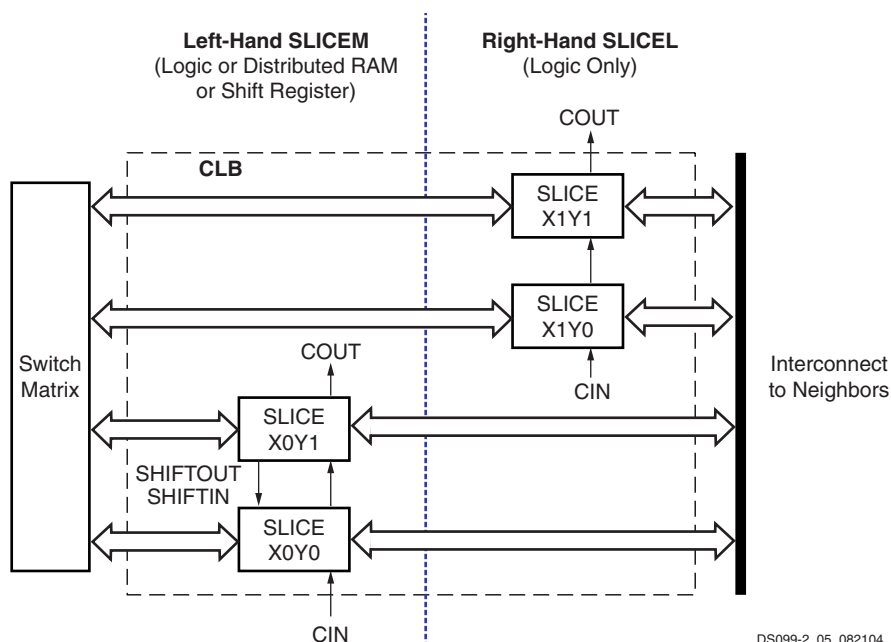


Figure 11: Arrangement of Slices within the CLB

Elements Within a Slice

All four slices have the following elements in common: two logic function generators, two storage elements, wide-function multiplexers, carry logic, and arithmetic gates, as shown in [Figure 12, page 24](#). Both the left-hand and right-hand slice pairs use these elements to provide logic, arithmetic, and ROM functions. Besides these, the left-hand pair supports two additional functions: storing data using Distributed RAM and shifting data with 16-bit registers. [Figure 12](#) is a diagram of the left-hand slice; therefore, it represents a superset of the elements and connections to be found in all slices. See [Function Generator, page 25](#) for more information.

The RAM-based function generator—also known as a Look-Up Table or LUT—is the main resource for implementing logic functions. Furthermore, the LUTs in each left-hand slice pair can be configured as Distributed RAM or a 16-bit shift register. For information on the former, refer to the chapter entitled “Using Look-Up Tables as Distributed RAM” in [UG331](#); for information on the latter, refer to the chapter entitled “Using Look-Up Tables as Shift Registers” in [UG331](#). The function generators located in the upper and lower portions of the slice are referred to as the “G” and “F”, respectively.

The storage element, which is programmable as either a D-type flip-flop or a level-sensitive latch, provides a means for synchronizing data to a clock signal, among other uses. The storage elements in the upper and lower portions of the slice are called FFY and FFX, respectively.

Wide-function multiplexers effectively combine LUTs in order to permit more complex logic operations. Each slice has two of these multiplexers with F5MUX in the lower portion of the slice and F1MUX in the upper portion. Depending on the slice, F1MUX takes on the name F6MUX, F7MUX, or F8MUX. For more details on the multiplexers, refer to the chapter entitled “Using Dedicated Multiplexers” in [UG331](#).

The carry chain, together with various dedicated arithmetic logic gates, support fast and efficient implementations of math operations. The carry chain enters the slice as CIN and exits as COUT. Five multiplexers control the chain: CYINIT, CY0F, and CYMUXF in the lower portion as well as CY0G and CYMUXG in the upper portion. The dedicated arithmetic logic includes the exclusive-OR gates XORG and XORF (upper and lower portions of the slice, respectively) as well as the AND gates GAND and FAND (upper and lower portions, respectively). For more details on the carry logic, refer to the chapter entitled “Using Carry and Arithmetic Logic” in [UG331](#).

Main Logic Paths

Central to the operation of each slice are two nearly identical data paths, distinguished using the terms *top* and *bottom*. The description that follows uses names associated with the bottom path. (The top path names appear in parentheses.) The basic path originates at an interconnect-switch matrix outside the CLB. Four lines, F1 through F4 (or G1 through G4 on the

Table 22: Status Logic Signals

Signal	Direction	Description
RST	Input	A High resets the entire DCM to its initial power-on state. Initializes the DLL taps for a delay of zero. Sets the LOCKED output Low. This input is asynchronous.
STATUS[7:0]	Output	The bit values on the STATUS bus provide information regarding the state of DLL and PS operation
LOCKED	Output	Indicates that the CLKIN and CLKFB signals are in phase by going High. The two signals are out-of-phase when Low.

Table 23: DCM STATUS Bus

Bit	Name	Description
0	Phase Shift Overflow	A value of 1 indicates a phase shift overflow when one of two conditions occurs: Incrementing (or decrementing) TPS beyond 255/256 of a CLKIN cycle. The DLL is producing its maximum possible phase shift (i.e., all delay taps are active). ⁽¹⁾
1	CLKIN Input Stopped Toggling	A value of 1 indicates that the CLKIN input signal is not toggling. A value of 0 indicates toggling. This bit functions only when the CLKFB input is connected. ⁽²⁾
2	CLKFX/CLKFX180 Output Stopped Toggling	A value of 1 indicates that the CLKFX or CLKFX180 output signals are not toggling. A value of 0 indicates toggling. This bit functions only when using the Digital Frequency Synthesizer (DFS).
3:7	Reserved	—

Notes:

1. The DLL phase shift with all delay taps active is specified as the parameter FINE_SHIFT_RANGE.
2. If only the DFS clock outputs are used, but none of the DLL clock outputs, this bit will not go High when the CLKIN signal stops.

Table 24: Status Attributes

Attribute	Description	Values
STARTUP_WAIT	Delays transition from configuration to user mode until lock condition is achieved.	TRUE, FALSE

Stabilizing DCM Clocks Before User Mode

It is possible to delay the completion of device configuration until after the DLL has achieved a lock condition using the STARTUP_WAIT attribute described in [Table 24](#). This option ensures that the FPGA does not enter user mode—i.e., begin functional operation—until all system clocks generated by the DCM are stable. In order to achieve the delay, it is necessary to set the attribute to TRUE as well as set the BitGen option LCK_cycle to one of the six cycles making up the Startup phase of configuration. The selected cycle defines the point at which configuration will halt until the LOCKED output goes High.

Global Clock Network

Spartan-3 devices have eight Global Clock inputs called GCLK0 - GCLK7. These inputs provide access to a low-capacitance, low-skew network that is well-suited to carrying high-frequency signals. The Spartan-3 FPGAs clock network is shown in [Figure 23](#). GCLK0 through GCLK3 are located in the center of the bottom edge. GCLK4 through GCLK7 are located in the center of the top edge.

Eight Global Clock Multiplexers (also called BUFGMUX elements) are provided that accept signals from Global Clock inputs and route them to the internal clock network as well as DCMs. Four BUFGMUX elements are located in the center of the bottom edge, just above the GCLK0 - GCLK3 inputs. The remaining four BUFGMUX elements are located in the center of the top edge, just below the GCLK4 - GCLK7 inputs.

Pairs of BUFGMUX elements share global inputs, as shown in [Figure 24](#). For example, the GCLK4 and GCLK5 inputs both potentially connect to BUFGMUX4 and BUFGMUX5 located in the upper right center. A differential clock input uses a pair of GCLK inputs to connect to a single BUFGMUX element.

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Table 33: General DC Characteristics of User I/O, Dual-Purpose, and Dedicated Pins

Symbol	Description	Test Conditions	Min	Typ	Max	Units
$I_L^{(2)(4)}$	Leakage current at User I/O, Dual-Purpose, and Dedicated pins	Driver is Hi-Z, $V_{IN} = 0V$ or V_{CCO} max, sample-tested	$V_{CCO} \geq 3.0V$	—	—	± 25 μA
			$V_{CCO} < 3.0V$	—	—	± 10 μA
$I_{RPU}^{(3)}$	Current through pull-up resistor at User I/O, Dual-Purpose, and Dedicated pins	$V_{IN} = 0V$, $V_{CCO} = 3.3V$	—0.84	—	—2.35	mA
		$V_{IN} = 0V$, $V_{CCO} = 3.0V$	—0.69	—	—1.99	mA
		$V_{IN} = 0V$, $V_{CCO} = 2.5V$	—0.47	—	—1.41	mA
		$V_{IN} = 0V$, $V_{CCO} = 1.8V$	—0.21	—	—0.69	mA
		$V_{IN} = 0V$, $V_{CCO} = 1.5V$	—0.13	—	—0.43	mA
		$V_{IN} = 0V$, $V_{CCO} = 1.2V$	—0.06	—	—0.22	mA
$R_{PU}^{(3)}$	Equivalent resistance of pull-up resistor at User I/O, Dual-Purpose, and Dedicated pins, derived from I_{RPU}	$V_{CCO} = 3.0V$ to $3.465V$	1.27	—	4.11	k Ω
		$V_{CCO} = 2.3V$ to $2.7V$	1.15	—	3.25	k Ω
		$V_{CCO} = 1.7V$ to $1.9V$	2.45	—	9.10	k Ω
		$V_{CCO} = 1.4V$ to $1.6V$	3.25	—	12.10	k Ω
		$V_{CCO} = 1.14$ to $1.26V$	5.15	—	21.00	k Ω
$I_{RPD}^{(3)}$	Current through pull-down resistor at User I/O, Dual-Purpose, and Dedicated pins	$V_{IN} = V_{CCO}$	0.37	—	1.67	mA
$R_{PD}^{(3)}$	Equivalent resistance of pull-down resistor at User I/O, Dual-Purpose, and Dedicated pins, driven from I_{RPD}	$V_{IN} = V_{CCO} = 3.0V$ to $3.465V$	1.75	—	9.35	k Ω
		$V_{IN} = V_{CCO} = 2.3V$ to $2.7V$	1.35	—	7.30	k Ω
		$V_{IN} = V_{CCO} = 1.7V$ to $1.9V$	1.00	—	5.15	k Ω
		$V_{IN} = V_{CCO} = 1.4V$ to $1.6V$	0.85	—	4.35	k Ω
		$V_{IN} = V_{CCO} = 1.14$ to $1.26V$	0.68	—	3.465	k Ω
R_{DCI}	Value of external reference resistor to support DCI I/O standards		20	—	100	Ω
I_{REF}	V_{REF} current per pin	$V_{CCO} \geq 3.0V$	—	—	± 25	μA
		$V_{CCO} < 3.0V$	—	—	± 10	μA
C_{IN}	Input capacitance		3	—	10	pF

Notes:

- The numbers in this table are based on the conditions set forth in [Table 32](#).
- The I_L specification applies to every I/O pin throughout power-on as long as the voltage on that pin stays between the absolute V_{IN} minimum and maximum values ([Table 28](#)). For hot-swap applications, at the time of card connection, be sure to keep all I/O voltages within this range before applying V_{CCO} power. Consider applying V_{CCO} power before connecting the signal lines, to avoid turning on the ESD protection diodes, shown in Module 2: [Figure 7, page 11](#). When the FPGA is completely unpowered, the I/O pins are high impedance, but there is a path through the upper and lower ESD protection diodes.
- This parameter is based on characterization. The pull-up resistance $R_{PU} = V_{CCO} / I_{RPU}$. The pull-down resistance $R_{PD} = V_{IN} / I_{RPD}$. Spartan-3 family values for both resistances are stronger than they have been for previous FPGA families.
- For single-ended signals that are placed on a differential-capable I/O, V_{IN} of $-0.2V$ to $-0.3V$ is supported but can cause increased leakage between the two pins. See the *Parasitic Leakage* section in [UG331, Spartan-3 Generation FPGA User Guide](#).

Table 35: Recommended Operating Conditions for User I/Os Using Single-Ended Standards

Signal Standard (IOSTANDARD)	V _{CCO}			V _{REF}			V _{IL}	V _{IH}
	Min (V)	Nom (V)	Max (V)	Min (V)	Nom (V)	Max (V)	Max (V)	Min (V)
GTL ⁽³⁾	—	—	—	0.74	0.8	0.86	V _{REF} – 0.05	V _{REF} + 0.05
GTL_DCI	—	1.2	—	0.74	0.8	0.86	V _{REF} – 0.05	V _{REF} + 0.05
GTLP ⁽³⁾	—	—	—	0.88	1	1.12	V _{REF} – 0.1	V _{REF} + 0.1
GTLP_DCI	—	1.5	—	0.88	1	1.12	V _{REF} – 0.1	V _{REF} + 0.1
HSLVDCI_15	1.4	1.5	1.6	—	0.75	—	V _{REF} – 0.1	V _{REF} + 0.1
HSLVDCI_18	1.7	1.8	1.9	—	0.9	—	V _{REF} – 0.1	V _{REF} + 0.1
HSLVDCI_25	2.3	2.5	2.7	—	1.25	—	V _{REF} – 0.1	V _{REF} + 0.1
HSLVDCI_33	3.0	3.3	3.465	—	1.65	—	V _{REF} – 0.1	V _{REF} + 0.1
HSTL_I, HSTL_I_DCI	1.4	1.5	1.6	0.68	0.75	0.9	V _{REF} – 0.1	V _{REF} + 0.1
HSTL_III, HSTL_III_DCI	1.4	1.5	1.6	—	0.9	—	V _{REF} – 0.1	V _{REF} + 0.1
HSTL_I_18, HSTL_I_DCI_18	1.7	1.8	1.9	0.8	0.9	1.1	V _{REF} – 0.1	V _{REF} + 0.1
HSTL_II_18, HSTL_II_DCI_18	1.7	1.8	1.9	—	0.9	—	V _{REF} – 0.1	V _{REF} + 0.1
HSTL_III_18, HSTL_III_DCI_18	1.7	1.8	1.9	—	1.1	—	V _{REF} – 0.1	V _{REF} + 0.1
LVC MOS12	1.14	1.2	1.3	—	—	—	0.37V _{CCO}	0.58V _{CCO}
LVC MOS15, LVDCI_15, LVDCI_DV2_15	1.4	1.5	1.6	—	—	—	0.30V _{CCO}	0.70V _{CCO}
LVC MOS18, LVDCI_18, LVDCI_DV2_18	1.7	1.8	1.9	—	—	—	0.30V _{CCO}	0.70V _{CCO}
LVC MOS25 ^(4,5) , LVDCI_25, LVDCI_DV2_25 ⁽⁴⁾	2.3	2.5	2.7	—	—	—	0.7	1.7
LVC MOS33, LVDCI_33, LVDCI_DV2_33 ⁽⁴⁾	3.0	3.3	3.465	—	—	—	0.8	2.0
LVTTL	3.0	3.3	3.465	—	—	—	0.8	2.0
PCI33_3 ⁽⁷⁾	3.0	3.3	3.465	—	—	—	0.30V _{CCO}	0.50V _{CCO}
SSTL18_I, SSTL18_I_DCI	1.7	1.8	1.9	0.833	0.900	0.969	V _{REF} – 0.125	V _{REF} + 0.125
SSTL18_II	1.7	1.8	1.9	0.833	0.900	0.969	V _{REF} – 0.125	V _{REF} + 0.125
SSTL2_I, SSTL2_I_DCI	2.3	2.5	2.7	1.15	1.25	1.35	V _{REF} – 0.15	V _{REF} + 0.15
SSTL2_II, SSTL2_II_DCI	2.3	2.5	2.7	1.15	1.25	1.35	V _{REF} – 0.15	V _{REF} + 0.15

Notes:

- Descriptions of the symbols used in this table are as follows:
V_{CCO} – the supply voltage for output drivers as well as LVC MOS, LVTTL, and PCI inputs
V_{REF} – the reference voltage for setting the input switching threshold
V_{IL} – the input voltage that indicates a Low logic level
V_{IH} – the input voltage that indicates a High logic level
- For device operation, the maximum signal voltage (V_{IH} max) may be as high as V_{IN} max. See [Table 28](#).
- Because the GTL and GTLP standards employ open-drain output buffers, V_{CCO} lines do not supply current to the I/O circuit, rather this current is provided using an external pull-up resistor connected from the I/O pin to a termination voltage (V_{TT}). Nevertheless, the voltage applied to the associated V_{CCO} lines must always be at or above V_{TT} and I/O pad voltages.
- There is approximately 100 mV of hysteresis on inputs using LVC MOS25 or LVC MOS33 standards.
- All dedicated pins (M0-M2, CCLK, PROG_B, DONE, HSWAP_EN, TCK, TDI, TDO, and TMS) use the LVC MOS standard and draw power from the V_{CCAUX} rail (2.5V). The dual-purpose configuration pins (DIN/D0, D1-D7, CS_B, RDWR_B, BUSY/DOUT, and INIT_B) use the LVC MOS standard before the user mode. For these pins, apply 2.5V to the V_{CCO} Bank 4 and V_{CCO} Bank 5 rails at power-on and throughout configuration. For information concerning the use of 3.3V signals, see [3.3V-Tolerant Configuration Interface, page 47](#).
- The Global Clock Inputs (GCLK0-GCLK7) are dual-purpose pins to which any signal standard can be assigned.
- For more information, see [XAPP457](#).

Table 41: System-Synchronous Pin-to-Pin Setup and Hold Times for the IOB Input Path

Symbol	Description	Conditions	Device	Speed Grade		Units
				-5	-4	
				Min	Min	
Setup Times						
T_{PSDCM}	When writing to the Input Flip-Flop (IFF), the time from the setup of data at the Input pin to the active transition at a Global Clock pin. The DCM is in use. No Input Delay is programmed.	LVCMOS25 ⁽²⁾ , IOBDELAY = NONE, with DCM ⁽⁴⁾	XC3S50	2.37	2.71	ns
			XC3S200	2.13	2.35	ns
			XC3S400	2.15	2.36	ns
			XC3S1000	2.58	2.95	ns
			XC3S1500	2.55	2.91	ns
			XC3S2000	2.59	2.96	ns
			XC3S4000	2.76	3.15	ns
			XC3S5000	2.69	3.08	ns
T_{PSFD}	When writing to IFF, the time from the setup of data at the Input pin to an active transition at the Global Clock pin. The DCM is not in use. The Input Delay is programmed.	LVCMOS25 ⁽²⁾ , IOBDELAY = IFD, without DCM	XC3S50	3.00	3.46	ns
			XC3S200	2.63	3.02	ns
			XC3S400	2.50	2.87	ns
			XC3S1000	3.50	4.03	ns
			XC3S1500	3.78	4.35	ns
			XC3S2000	4.98	5.73	ns
			XC3S4000	5.25	6.05	ns
			XC3S5000	5.37	6.18	ns
Hold Times						
T_{PHDCM}	When writing to IFF, the time from the active transition at the Global Clock pin to the point when data must be held at the Input pin. The DCM is in use. No Input Delay is programmed.	LVCMOS25 ⁽³⁾ , IOBDELAY = NONE, with DCM ⁽⁴⁾	XC3S50	−0.45	−0.40	ns
			XC3S200	−0.12	−0.05	ns
			XC3S400	−0.12	−0.05	ns
			XC3S1000	−0.43	−0.38	ns
			XC3S1500	−0.45	−0.40	ns
			XC3S2000	−0.47	−0.42	ns
			XC3S4000	−0.61	−0.56	ns
			XC3S5000	−0.62	−0.57	ns

Table 50: Recommended Number of Simultaneously Switching Outputs per V_{CCO}/GND Pair (Cont'd)

Signal Standard (IOSTANDARD)			Package				
			VQ100	TQ144	PQ208	CP132	FT256, FG320, FG456, FG676, FG900, FG1156
LVCMOS33	Slow	2	34	24	24	52	76
		4	17	14	14	26	46
		6	17	11	11	26	27
		8	10	10	10	13	20
		12	9	9	9	13	13
		16	8	8	8	8	10
		24	8	8	8	8	9
	Fast	2	20	20	20	26	44
		4	15	15	15	15	26
		6	11	11	11	13	16
		8	10	10	10	10	12
		12	8	8	8	8	10
		16	8	8	8	8	8
		24	7	7	7	7	7
LVDCI_33			10	10	10	10	10
LVDCI_DV2_33			10	10	10	10	10
HSLVDCI_33			10	10	10	10	10
LVTTTL	Slow	2	34	25	25	52	60
		4	17	16	16	26	41
		6	17	15	15	26	29
		8	12	12	12	13	22
		12	10	10	10	13	13
		16	10	10	10	10	11
		24	8	8	8	8	9
	Fast	2	20	20	20	26	34
		4	13	13	13	13	20
		6	11	11	11	13	15
		8	10	10	10	10	12
		12	9	9	9	9	10
		16	8	8	8	8	9
		24	7	7	7	7	7

Table 59: Switching Characteristics for the DLL (Cont'd)

Symbol	Description	Frequency Mode / FCLKIN Range	Device	Speed Grade				Units
				-5		-4		
				Min	Max	Min	Max	
Lock Time								
LOCK_DLL	When using the DLL alone: The time from deassertion at the DCM's Reset input to the rising transition at its LOCKED output. When the DCM is locked, the CLKIN and CLKFB signals are in phase	$18\text{ MHz} \leq F_{\text{CLKIN}} \leq 30\text{ MHz}$	All	—	2.88	—	2.88	ms
		$30\text{ MHz} < F_{\text{CLKIN}} \leq 40\text{ MHz}$		—	2.16	—	2.16	ms
		$40\text{ MHz} < F_{\text{CLKIN}} \leq 50\text{ MHz}$		—	1.20	—	1.20	ms
		$50\text{ MHz} < F_{\text{CLKIN}} \leq 60\text{ MHz}$		—	0.60	—	0.60	ms
		$F_{\text{CLKIN}} > 60\text{ MHz}$		—	0.48	—	0.48	ms
Delay Lines								
DCM_TAP	Delay tap resolution	All	All	30.0	60.0	30.0	60.0	ps

Notes:

- The numbers in this table are based on the operating conditions set forth in [Table 32](#) and [Table 58](#).
- DLL specifications apply when any of the DLL outputs (CLK0, CLK90, CLK180, CLK270, CLK2X, CLK2X180, or CLKDV) are in use.
- Only mask revision 'E' and later devices (see [Mask and Fab Revisions, page 58](#)) and all revisions of the XC3S50 and the XC3S1000 support DLL feedback using the CLK2X output. For all other Spartan-3 devices, use feedback from the CLK0 output (instead of the CLK2X output) and set the CLK_FEEDBACK attribute to 1X.
- Indicates the maximum amount of output jitter that the DCM adds to the jitter on the CLKIN input.
- This specification only applies if the attribute DUTY_CYCLE_CORRECTION = TRUE.

Digital Frequency Synthesizer (DFS)
Table 60: Recommended Operating Conditions for the DFS

Symbol		Description	Frequency Mode	Speed Grade				Units
				-5		-4		
				Min	Max	Min	Max	
Input Frequency Ranges ⁽²⁾								
F _{CLKIN}	CLKIN_FREQ_FX	Frequency for the CLKIN input	All	1	280	1	280	MHz
Input Clock Jitter Tolerance ⁽³⁾								
CLKIN_CYC_JITT_FX_LF		Cycle-to-cycle jitter at the CLKIN input	Low	—	±300	—	±300	ps
CLKIN_CYC_JITT_FX_HF			High	—	±150	—	±150	ps
CLKIN_PER_JITT_FX		Period jitter at the CLKIN input	All	—	±1	—	±1	ns

Notes:

- DFS specifications apply when either of the DFS outputs (CLKFX or CLKFX180) are used.
- If both DFS and DLL outputs are used on the same DCM, follow the more restrictive CLKIN_FREQ_DLL specifications in [Table 58](#).
- CLKIN input jitter beyond these limits may cause the DCM to lose lock.

Table 71: Dual-Purpose Pins Used in Master or Slave Serial Mode

Pin Name	Direction	Description
DIN	Input	Serial Data Input: During the Master or Slave Serial configuration modes, DIN is the serial configuration data input, and all data is synchronized to the rising CCLK edge. After configuration, this pin is available as a user I/O. This signal is located in Bank 4 and its output voltage determined by VCCO_4. The BitGen option Persist permits this pin to retain its configuration function in the User mode.
DOUT	Output	Serial Data Output: In a multi-FPGA design where all the FPGAs use serial mode, connect the DOUT output of one FPGA—in either Master or Slave Serial mode—to the DIN input of the next FPGA—in Slave Serial mode—so that configuration data passes from one to the next, in daisy-chain fashion. This “daisy chain” permits sequential configuration of multiple FPGAs. This signal is located in Bank 4 and its output voltage determined by VCCO_4. The BitGen option Persist permits this pin to retain its configuration function in the User mode.
INIT_B	Bidirectional (open-drain)	Initializing Configuration Memory/Configuration Error: Just after power is applied, the FPGA produces a Low-to-High transition on this pin indicating that initialization (<i>i.e.</i> , clearing) of the configuration memory has finished. Before entering the User mode, this pin functions as an open-drain output, which requires a pull-up resistor in order to produce a High logic level. In a multi-FPGA design, tie (wire AND) the INIT_B pins from all FPGAs together so that the common node transitions High only after all of the FPGAs have been successfully initialized. Externally holding this pin Low beyond the initialization phase delays the start of configuration. This action stalls the FPGA at the configuration step just before the mode select pins are sampled. During configuration, the FPGA indicates the occurrence of a data (<i>i.e.</i> , CRC) error by asserting INIT_B Low. This signal is located in Bank 4 and its output voltage determined by VCCO_4. The BitGen option Persist permits this pin to retain its configuration function in the User mode.

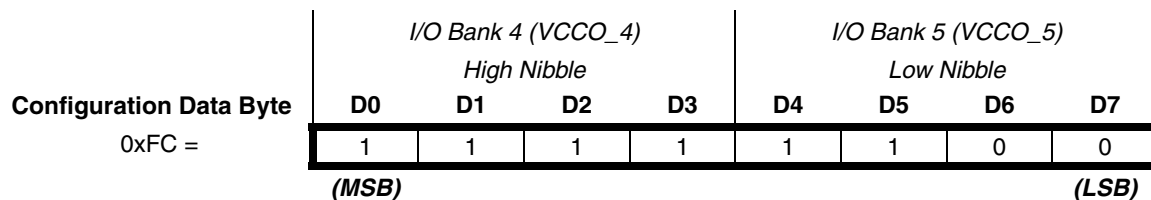


Figure 41: Configuration Data Byte Mapping to D0-D7 Bits

Parallel Configuration Modes (SelectMAP)

This section describes the dual-purpose configuration pins used during the Master and Slave Parallel configuration modes, sometimes also called the SelectMAP modes. In both Master and Slave Parallel configuration modes, D0-D7 form the byte-wide configuration data input. See Table 75 for Mode Select pin settings required for Parallel modes.

As shown in Figure 41, D0 is the most-significant bit while D7 is the least-significant bit. Bits D0-D3 form the high nibble of the byte and bits D4-D7 form the low nibble.

In the Parallel configuration modes, both the VCCO_4 and VCCO_5 voltage supplies are required and must both equal the voltage of the attached configuration device, typically either 2.5V or 3.3V.

Assert Low both the chip-select pin, CS_B, and the read/write control pin, RDWR_B, to write the configuration data byte presented on the D0-D7 pins to the FPGA on a rising-edge of the configuration clock, CCLK. The order of CS_B and RDWR_B does not matter, although RDWR_B must be asserted throughout the configuration process. If RDWR_B is de-asserted during configuration, the FPGA aborts the configuration operation.

After configuration, these pins are available as general-purpose user I/O. However, the SelectMAP configuration interface is optionally available for debugging and dynamic reconfiguration. To use these SelectMAP pins after configuration, set the Persist bitstream generation option.

The Readback debugging option, for example, requires the Persist bitstream generation option. During Readback mode, assert CS_B Low, along with RDWR_B High, to read a configuration data byte from the FPGA to the D0-D7 bus on a rising CCLK edge. During Readback mode, D0-D7 are output pins.

In all the cases, the configuration data and control signals are synchronized to the rising edge of the CCLK clock signal.

Table 72: Dual-Purpose Configuration Pins for Parallel (SelectMAP) Configuration Modes (Cont'd)

Pin Name	Direction	Description								
BUSY	Output	<p>Configuration Data Rate Control for Parallel Mode:</p> <p>In the Slave and Master Parallel modes, BUSY throttles the rate at which configuration data is loaded. BUSY is only necessary if CCLK operates at greater than 50 MHz. Ignore BUSY for frequencies of 50 MHz and below.</p> <p>When BUSY is Low, the FPGA accepts the next configuration data byte on the next rising CCLK edge for which CS_B and RDWR_B are Low. When BUSY is High, the FPGA ignores the next configuration data byte. The next configuration data value must be held or reloaded until the next rising CCLK edge when BUSY is Low. When CS_B is High, BUSY is in a high impedance state.</p> <table><thead><tr><th>BUSY</th><th>Function</th></tr></thead><tbody><tr><td>0</td><td>The FPGA is ready to accept the next configuration data byte.</td></tr><tr><td>1</td><td>The FPGA is busy processing the current configuration data byte and is not ready to accept the next byte.</td></tr><tr><td>Hi-Z</td><td>If CS_B is High, then BUSY is high impedance.</td></tr></tbody></table> <p>This signal is located in Bank 4 and its output voltage is determined by VCCO_4. The BitGen option Persist permits this pin to retain its configuration function in the User mode.</p>	BUSY	Function	0	The FPGA is ready to accept the next configuration data byte.	1	The FPGA is busy processing the current configuration data byte and is not ready to accept the next byte.	Hi-Z	If CS_B is High, then BUSY is high impedance.
BUSY	Function									
0	The FPGA is ready to accept the next configuration data byte.									
1	The FPGA is busy processing the current configuration data byte and is not ready to accept the next byte.									
Hi-Z	If CS_B is High, then BUSY is high impedance.									
INIT_B	Bidirectional (open-drain)	<p>Initializing Configuration Memory/Configuration Error (active-Low):</p> <p>See description under Serial Configuration Modes, page 112.</p>								

JTAG Configuration Mode

In the JTAG configuration mode all dual-purpose configuration pins are unused and behave exactly like user-I/O pins, as shown in [Table 79](#). See [Table 75](#) for Mode Select pin settings required for JTAG mode.

Dual-Purpose Pin I/O Standard During Configuration

During configuration, the dual-purpose pins default to CMOS input and output levels for the associated VCCO voltage supply pins. For example, in the Parallel configuration modes, both VCCO_4 and VCCO_5 are required. If connected to +2.5V, then the associated pins conform to the LVCMOS25 I/O standard. If connected to +3.3V, then the pins drive LVCMOS output levels and accept either LVTTL or LVCMOS input levels.

Dual-Purpose Pin Behavior After Configuration

After the configuration process completes, these pins, if they were borrowed during configuration, become user-I/O pins available to the application. If a dual-purpose configuration pin is not used during the configuration process—*i.e.*, the parallel configuration pins when using serial mode—then the pin behaves exactly like a general-purpose I/O. See [I/O Type: Unrestricted, General-purpose I/O Pins](#) section.

DCI: User I/O or Digitally Controlled Impedance Resistor Reference Input

These pins are individual user-I/O pins unless one of the I/O standards used in the bank requires the Digitally Controlled Impedance (DCI) feature. If DCI is used, then 1% precision resistors connected to the VRP_# and VRN_# pins match the impedance on the input or output buffers of the I/O standards that use DCI within the bank. The '#' character in the pin name indicates the associated I/O bank and is an integer, 0 through 7.

There are two DCI pins per I/O bank, except in the CP132 and TQ144 packages, which do not have any DCI inputs for Bank 5.

VRP and VRN Impedance Resistor Reference Inputs

The 1% precision impedance-matching resistor attached to the VRP_# pin controls the pull-up impedance of PMOS transistor in the input or output buffer. Consequently, the VRP_# pin must connect to ground. The 'P' character in "VRP" indicates that this pin controls the I/O buffer's PMOS transistor impedance. The VRP_# pin is used for both single and split termination.

CP132: 132-Ball Chip-Scale Package

Note: The CP132 and CPG132 packages are discontinued. See www.xilinx.com/support/documentation/spartan-3.htm#19600.

The pinout and footprint for the XC3S50 in the 132-ball chip-scale package, CP132, appear in [Table 89](#) and [Figure 45](#).

All the package pins appear in [Table 89](#) and are sorted by bank number, then by pin name. Pins that form a differential I/O pair appear together in the table. The table also shows the pin number for each pin and the pin type, as defined earlier.

The CP132 footprint has eight I/O banks. However, the voltage supplies for the two I/O banks along an edge are connected together internally. Consequently, there are four output voltage supplies, labeled VCCO_TOP, VCCO_RIGHT, VCCO_BOTTOM, and VCCO_LEFT.

Pinout Table

Table 89: CP132 Package Pinout

Bank	XC3S50 Pin Name	CP132 Ball	Type
0	IO_L01N_0/VRP_0	A3	DCI
0	IO_L01P_0/VRN_0	C4	DCI
0	IO_L27N_0	C5	I/O
0	IO_L27P_0	B5	I/O
0	IO_L30N_0	B6	I/O
0	IO_L30P_0	A6	I/O
0	IO_L31N_0	C7	I/O
0	IO_L31P_0/VREF_0	B7	VREF
0	IO_L32N_0/GCLK7	A7	GCLK
0	IO_L32P_0/GCLK6	C8	GCLK
1	IO_L01N_1/VRP_1	A13	DCI
1	IO_L01P_1/VRN_1	B13	DCI
1	IO_L27N_1	C11	I/O
1	IO_L27P_1	A12	I/O
1	IO_L28N_1	A11	I/O
1	IO_L28P_1	B11	I/O
1	IO_L31N_1/VREF_1	C9	VREF
1	IO_L31P_1	A10	I/O
1	IO_L32N_1/GCLK5	A8	GCLK
1	IO_L32P_1/GCLK4	A9	GCLK
2	IO_L01N_2/VRP_2	D12	DCI
2	IO_L01P_2/VRN_2	C14	DCI
2	IO_L20N_2	E12	I/O
2	IO_L20P_2	E13	I/O
2	IO_L21N_2	E14	I/O
2	IO_L21P_2	F12	I/O
2	IO_L23N_2/VREF_2	F13	VREF
2	IO_L23P_2	F14	I/O
2	IO_L24N_2	G12	I/O

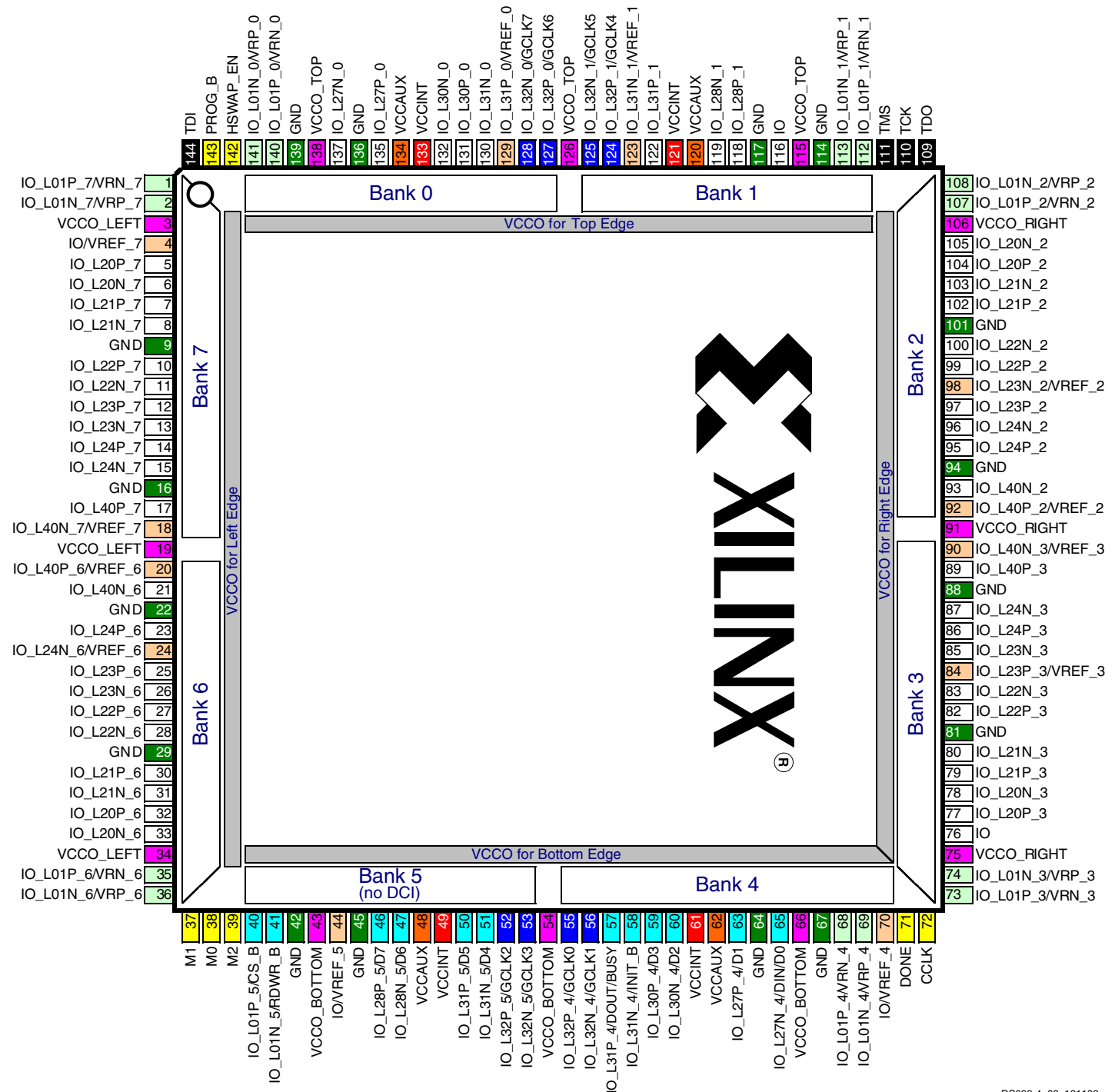
User I/Os by Bank

Table 92 indicates how the available user-I/O pins are distributed between the eight I/O banks on the TQ144 package.

Table 92: User I/Os Per Bank in TQ144 Package

Package Edge	I/O Bank	Maximum I/O	All Possible I/O Pins by Type				
			I/O	DUAL	DCI	VREF	GCLK
Top	0	10	5	0	2	1	2
	1	9	4	0	2	1	2
Right	2	14	10	0	2	2	0
	3	15	11	0	2	2	0
Bottom	4	11	0	6	2	1	2
	5	9	0	6	0	1	2
Left	6	14	10	0	2	2	0
	7	15	11	0	2	2	0

TQ144 Footprint



DS099-4_08_121103

Figure 46: TQ144 Package Footprint (Top View). Note pin 1 indicator in top-left corner and logo orientation.

51	I/O: Unrestricted, general-purpose user I/O	12	DUAL: Configuration pin, then possible user I/O	12	VREF: User I/O or input voltage reference for bank
14	DCI: User I/O or reference resistor input for bank	8	GCLK: User I/O or global clock buffer input	12	VCCO: Output voltage supply for bank
7	CONFIG: Dedicated configuration pins	4	JTAG: Dedicated JTAG port pins	4	VCCINT: Internal core voltage supply (+1.2V)
0	N.C.: No unconnected pins in this package	16	GND: Ground	4	VCCAUX: Auxiliary voltage supply (+2.5V)

Table 100: FG456 Package Pinout (Cont'd)

Bank	3S400 Pin Name	3S1000, 3S1500, 3S2000 Pin Name	FG456 Pin Number	Type
6	N.C. (◆)	IO_L28N_6	R5	I/O
6	N.C. (◆)	IO_L28P_6	P6	I/O
6	N.C. (◆)	IO_L29N_6	R2	I/O
6	N.C. (◆)	IO_L29P_6	R1	I/O
6	N.C. (◆)	IO_L31N_6	P5	I/O
6	N.C. (◆)	IO_L31P_6	P4	I/O
6	N.C. (◆)	IO_L32N_6	P2	I/O
6	N.C. (◆)	IO_L32P_6	P1	I/O
6	N.C. (◆)	IO_L33N_6	N6	I/O
6	N.C. (◆)	IO_L33P_6	N5	I/O
6	IO_L34N_6/VREF_6	IO_L34N_6/VREF_6	N4	VREF
6	IO_L34P_6	IO_L34P_6	N3	I/O
6	IO_L35N_6	IO_L35N_6	N2	I/O
6	IO_L35P_6	IO_L35P_6	N1	I/O
6	IO_L38N_6	IO_L38N_6	M6	I/O
6	IO_L38P_6	IO_L38P_6	M5	I/O
6	IO_L39N_6	IO_L39N_6	M4	I/O
6	IO_L39P_6	IO_L39P_6	M3	I/O
6	IO_L40N_6	IO_L40N_6	M2	I/O
6	IO_L40P_6/VREF_6	IO_L40P_6/VREF_6	M1	VREF
6	VCCO_6	VCCO_6	M7	VCCO
6	VCCO_6	VCCO_6	N7	VCCO
6	VCCO_6	VCCO_6	P7	VCCO
6	VCCO_6	VCCO_6	R3	VCCO
6	VCCO_6	VCCO_6	R6	VCCO
7	IO	IO	C2	I/O
7	IO_L01N_7/VRP_7	IO_L01N_7/VRP_7	C3	DCI
7	IO_L01P_7/VRN_7	IO_L01P_7/VRN_7	C4	DCI
7	IO_L16N_7	IO_L16N_7	D1	I/O
7	IO_L16P_7/VREF_7	IO_L16P_7/VREF_7	C1	VREF
7	IO_L17N_7	IO_L17N_7	E4	I/O
7	IO_L17P_7	IO_L17P_7	D4	I/O
7	IO_L19N_7/VREF_7	IO_L19N_7/VREF_7	D3	VREF
7	IO_L19P_7	IO_L19P_7	D2	I/O
7	IO_L20N_7	IO_L20N_7	F4	I/O
7	IO_L20P_7	IO_L20P_7	E3	I/O
7	IO_L21N_7	IO_L21N_7	E1	I/O
7	IO_L21P_7	IO_L21P_7	E2	I/O
7	IO_L22N_7	IO_L22N_7	G6	I/O
7	IO_L22P_7	IO_L22P_7	F5	I/O

User I/Os by Bank

Table 101 indicates how the available user-I/O pins are distributed between the eight I/O banks for the XC3S400 in the FG456 package. Similarly, Table 102 shows how the available user-I/O pins are distributed between the eight I/O banks for the XC3S1000, XC3S1500, and XC3S2000 in the FG456 package.

Table 101: User I/Os Per Bank for XC3S400 in FG456 Package

Edge	I/O Bank	Maximum I/O	All Possible I/O Pins by Type				
			I/O	DUAL	DCI	VREF	GCLK
Top	0	35	27	0	2	4	2
	1	35	27	0	2	4	2
Right	2	31	25	0	2	4	0
	3	31	25	0	2	4	0
Bottom	4	35	21	6	2	4	2
	5	35	21	6	2	4	2
Left	6	31	25	0	2	4	0
	7	31	25	0	2	4	0

Table 102: User I/Os Per Bank for XC3S1000, XC3S1500, and XC3S2000 in FG456 Package

Edge	I/O Bank	Maximum I/O	All Possible I/O Pins by Type				
			I/O	DUAL	DCI	VREF	GCLK
Top	0	40	31	0	2	5	2
	1	40	31	0	2	5	2
Right	2	43	37	0	2	4	0
	3	43	37	0	2	4	0
Bottom	4	41	26	6	2	5	2
	5	40	25	6	2	5	2
Left	6	43	37	0	2	4	0
	7	43	37	0	2	4	0

Table 107: FG900 Package Pinout (Cont'd)

Bank	XC3S2000 Pin Name	XC3S4000, XC3S5000 Pin Name	FG900 Pin Number	Type
2	VCCO_2	VCCO_2	J28	VCCO
2	VCCO_2	VCCO_2	N28	VCCO
3	IO	IO	AB25	I/O
3	IO_L01N_3/VRP_3	IO_L01N_3/VRP_3	AH30	DCI
3	IO_L01P_3/VRN_3	IO_L01P_3/VRN_3	AH29	DCI
3	IO_L02N_3/VREF_3	IO_L02N_3/VREF_3	AG28	VREF
3	IO_L02P_3	IO_L02P_3	AG27	I/O
3	IO_L03N_3	IO_L03N_3	AG30	I/O
3	IO_L03P_3	IO_L03P_3	AG29	I/O
3	IO_L04N_3	IO_L04N_3	AF30	I/O
3	IO_L04P_3	IO_L04P_3	AF29	I/O
3	IO_L05N_3	IO_L05N_3	AE26	I/O
3	IO_L05P_3	IO_L05P_3	AF27	I/O
3	IO_L06N_3	IO_L06N_3	AE29	I/O
3	IO_L06P_3	IO_L06P_3	AE28	I/O
3	IO_L07N_3	IO_L07N_3	AD28	I/O
3	IO_L07P_3	IO_L07P_3	AD27	I/O
3	IO_L08N_3	IO_L08N_3	AD30	I/O
3	IO_L08P_3	IO_L08P_3	AD29	I/O
3	IO_L09N_3	IO_L09N_3	AC24	I/O
3	IO_L09P_3/VREF_3	IO_L09P_3/VREF_3	AD25	VREF
3	IO_L10N_3	IO_L10N_3	AC26	I/O
3	IO_L10P_3	IO_L10P_3	AC25	I/O
3	IO_L11N_3	IO_L11N_3	AC28	I/O
3	IO_L11P_3	IO_L11P_3	AC27	I/O
3	IO_L13N_3/VREF_3	IO_L13N_3/VREF_3	AC30	VREF
3	IO_L13P_3	IO_L13P_3	AC29	I/O
3	IO_L14N_3	IO_L14N_3	AB27	I/O
3	IO_L14P_3	IO_L14P_3	AB26	I/O
3	IO_L15N_3	IO_L15N_3	AB30	I/O
3	IO_L15P_3	IO_L15P_3	AB29	I/O
3	IO_L16N_3	IO_L16N_3	AA22	I/O
3	IO_L16P_3	IO_L16P_3	AB23	I/O
3	IO_L17N_3	IO_L17N_3	AA25	I/O
3	IO_L17P_3/VREF_3	IO_L17P_3/VREF_3	AA24	VREF
3	IO_L19N_3	IO_L19N_3	AA29	I/O
3	IO_L19P_3	IO_L19P_3	AA28	I/O
3	IO_L20N_3	IO_L20N_3	Y21	I/O
3	IO_L20P_3	IO_L20P_3	AA21	I/O
3	IO_L21N_3	IO_L21N_3	Y24	I/O

Table 110: FG1156 Package Pinout (Cont'd)

Bank	XC3S4000 Pin Name	XC3S5000 Pin Name	FG1156 Pin Number	Type
2	IO_L19N_2	IO_L19N_2	M29	I/O
2	IO_L19P_2	IO_L19P_2	M30	I/O
2	IO_L20N_2	IO_L20N_2	M31	I/O
2	IO_L20P_2	IO_L20P_2	M32	I/O
2	IO_L21N_2	IO_L21N_2	M26	I/O
2	IO_L21P_2	IO_L21P_2	N25	I/O
2	IO_L22N_2	IO_L22N_2	N27	I/O
2	IO_L22P_2	IO_L22P_2	N28	I/O
2	IO_L23N_2/VREF_2	IO_L23N_2/VREF_2	N31	VREF
2	IO_L23P_2	IO_L23P_2	N32	I/O
2	IO_L24N_2	IO_L24N_2	N24	I/O
2	IO_L24P_2	IO_L24P_2	P24	I/O
2	IO_L26N_2	IO_L26N_2	P29	I/O
2	IO_L26P_2	IO_L26P_2	P30	I/O
2	IO_L27N_2	IO_L27N_2	P31	I/O
2	IO_L27P_2	IO_L27P_2	P32	I/O
2	IO_L28N_2	IO_L28N_2	P33	I/O
2	IO_L28P_2	IO_L28P_2	P34	I/O
2	IO_L29N_2	IO_L29N_2	R24	I/O
2	IO_L29P_2	IO_L29P_2	R25	I/O
2	IO_L30N_2	IO_L30N_2	R28	I/O
2	IO_L30P_2	IO_L30P_2	R29	I/O
2	IO_L31N_2	IO_L31N_2	R31	I/O
2	IO_L31P_2	IO_L31P_2	R32	I/O
2	IO_L32N_2	IO_L32N_2	R33	I/O
2	IO_L32P_2	IO_L32P_2	R34	I/O
2	IO_L33N_2	IO_L33N_2	R26	I/O
2	IO_L33P_2	IO_L33P_2	T25	I/O
2	IO_L34N_2/VREF_2	IO_L34N_2/VREF_2	T28	VREF
2	IO_L34P_2	IO_L34P_2	T29	I/O
2	IO_L35N_2	IO_L35N_2	T32	I/O
2	IO_L35P_2	IO_L35P_2	T33	I/O
2	IO_L37N_2	IO_L37N_2	U27	I/O
2	IO_L37P_2	IO_L37P_2	U28	I/O
2	IO_L38N_2	IO_L38N_2	U29	I/O
2	IO_L38P_2	IO_L38P_2	U30	I/O
2	IO_L39N_2	IO_L39N_2	U31	I/O
2	IO_L39P_2	IO_L39P_2	U32	I/O
2	IO_L40N_2	IO_L40N_2	U33	I/O
2	IO_L40P_2/VREF_2	IO_L40P_2/VREF_2	U34	VREF

Table 110: FG1156 Package Pinout (Cont'd)

Bank	XC3S4000 Pin Name	XC3S5000 Pin Name	FG1156 Pin Number	Type
4	IO_L01N_4/VRP_4	IO_L01N_4/VRP_4	AN32	DCI
4	IO_L01P_4/VRN_4	IO_L01P_4/VRN_4	AP32	DCI
4	IO_L02N_4	IO_L02N_4	AN31	I/O
4	IO_L02P_4	IO_L02P_4	AP31	I/O
4	IO_L03N_4	IO_L03N_4	AM30	I/O
4	IO_L03P_4	IO_L03P_4	AN30	I/O
4	IO_L04N_4	IO_L04N_4	AN27	I/O
4	IO_L04P_4	IO_L04P_4	AP27	I/O
4	IO_L05N_4	IO_L05N_4	AH26	I/O
4	IO_L05P_4	IO_L05P_4	AJ26	I/O
4	IO_L06N_4/VREF_4	IO_L06N_4/VREF_4	AL26	VREF
4	IO_L06P_4	IO_L06P_4	AM26	I/O
4	IO_L07N_4	IO_L07N_4	AF25	I/O
4	IO_L07P_4	IO_L07P_4	AG25	I/O
4	IO_L08N_4	IO_L08N_4	AH25	I/O
4	IO_L08P_4	IO_L08P_4	AJ25	I/O
4	IO_L09N_4	IO_L09N_4	AL25	I/O
4	IO_L09P_4	IO_L09P_4	AM25	I/O
4	IO_L10N_4	IO_L10N_4	AN25	I/O
4	IO_L10P_4	IO_L10P_4	AP25	I/O
4	IO_L11N_4	IO_L11N_4	AD23	I/O
4	IO_L11P_4	IO_L11P_4	AE23	I/O
4	IO_L12N_4	IO_L12N_4	AF23	I/O
4	IO_L12P_4	IO_L12P_4	AG23	I/O
4	IO_L13N_4	IO_L13N_4	AJ23	I/O
4	IO_L13P_4	IO_L13P_4	AK23	I/O
4	IO_L14N_4	IO_L14N_4	AL23	I/O
4	IO_L14P_4	IO_L14P_4	AM23	I/O
4	IO_L15N_4	IO_L15N_4	AN23	I/O
4	IO_L15P_4	IO_L15P_4	AP23	I/O
4	IO_L16N_4	IO_L16N_4	AG22	I/O
4	IO_L16P_4	IO_L16P_4	AH22	I/O
4	IO_L17N_4	IO_L17N_4	AL22	I/O
4	IO_L17P_4	IO_L17P_4	AM22	I/O
4	IO_L18N_4	IO_L18N_4	AD21	I/O
4	IO_L18P_4	IO_L18P_4	AE21	I/O
4	IO_L19N_4	IO_L19N_4	AG21	I/O
4	IO_L19P_4	IO_L19P_4	AH21	I/O
4	IO_L20N_4	IO_L20N_4	AJ21	I/O
4	IO_L20P_4	IO_L20P_4	AK21	I/O

Table 110: FG1156 Package Pinout (Cont'd)

Bank	XC3S4000 Pin Name	XC3S5000 Pin Name	FG1156 Pin Number	Type
5	VCCO_5	VCCO_5	AJ13	VCCO
5	VCCO_5	VCCO_5	AL11	VCCO
5	VCCO_5	VCCO_5	AL16	VCCO
5	VCCO_5	VCCO_5	AM4	VCCO
5	VCCO_5	VCCO_5	AM8	VCCO
5	VCCO_5	VCCO_5	AN13	VCCO
6	IO	IO	AH1	I/O
6	IO	IO	AH2	I/O
6	IO	IO	V9	I/O
6	IO	IO	V10	I/O
6	IO_L01N_6/VRP_6	IO_L01N_6/VRP_6	AM2	DCI
6	IO_L01P_6/VRN_6	IO_L01P_6/VRN_6	AM1	DCI
6	IO_L02N_6	IO_L02N_6	AL2	I/O
6	IO_L02P_6	IO_L02P_6	AL1	I/O
6	IO_L03N_6/VREF_6	IO_L03N_6/VREF_6	AK3	VREF
6	IO_L03P_6	IO_L03P_6	AK2	I/O
6	IO_L04N_6	IO_L04N_6	AJ4	I/O
6	IO_L04P_6	IO_L04P_6	AJ3	I/O
6	IO_L05N_6	IO_L05N_6	AJ2	I/O
6	IO_L05P_6	IO_L05P_6	AJ1	I/O
6	IO_L06N_6	IO_L06N_6	AH6	I/O
6	IO_L06P_6	IO_L06P_6	AH5	I/O
6	IO_L07N_6	IO_L07N_6	AG6	I/O
6	IO_L07P_6	IO_L07P_6	AG5	I/O
6	IO_L08N_6	IO_L08N_6	AG2	I/O
6	IO_L08P_6	IO_L08P_6	AG1	I/O
6	IO_L09N_6/VREF_6	IO_L09N_6/VREF_6	AF7	VREF
6	IO_L09P_6	IO_L09P_6	AF6	I/O
6	IO_L10N_6	IO_L10N_6	AG4	I/O
6	IO_L10P_6	IO_L10P_6	AF4	I/O
6	IO_L11N_6	IO_L11N_6	AF3	I/O
6	IO_L11P_6	IO_L11P_6	AF2	I/O
6	IO_L12N_6	IO_L12N_6	AF8	I/O
6	IO_L12P_6	IO_L12P_6	AE9	I/O
6	IO_L13N_6	IO_L13N_6	AE8	I/O
6	IO_L13P_6/VREF_6	IO_L13P_6/VREF_6	AE7	VREF
6	IO_L14N_6	IO_L14N_6	AE6	I/O
6	IO_L14P_6	IO_L14P_6	AE5	I/O
6	IO_L15N_6	IO_L15N_6	AE4	I/O
6	IO_L15P_6	IO_L15P_6	AE3	I/O