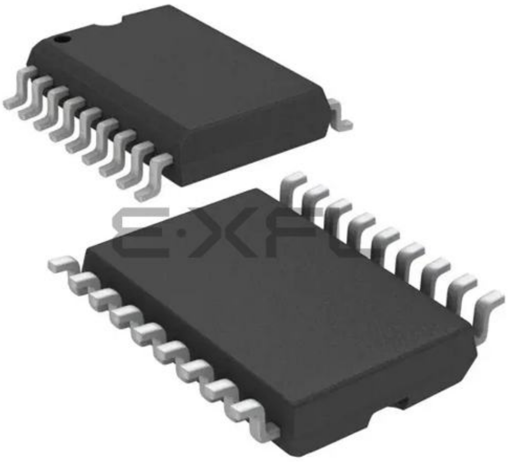

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Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	20MHz
Connectivity	UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	16
Program Memory Size	1.75KB (1K x 14)
Program Memory Type	FLASH
EEPROM Size	128 x 8
RAM Size	224 x 8
Voltage - Supply (Vcc/Vdd)	3V ~ 5.5V
Data Converters	-
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	18-SOIC (0.295", 7.50mm Width)
Supplier Device Package	18-SOIC
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic16f627t-20i-so

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PIC16F62X

FLASH-Based 8-Bit CMOS Microcontrollers

Devices Included in this Data Sheet:

- PIC16F627
- PIC16F628

Referred to collectively as PIC16F62X

High Performance RISC CPU:

- Only 35 instructions to learn
- All single cycle instructions (200 ns), except for program branches which are two-cycle
- Operating speed:
  - DC - 20 MHz clock input
  - DC - 200 ns instruction cycle

Device	Memory		
	FLASH Program	RAM Data	EEPROM Data
PIC16F627	1024 x 14	224 x 8	128 x 8
PIC16F628	2048 x 14	224 x 8	128 x 8

- Interrupt capability
- 16 special function hardware registers
- 8-level deep hardware stack
- Direct, Indirect and Relative addressing modes

Peripheral Features:

- 16 I/O pins with individual direction control
- High current sink/source for direct LED drive
- Analog comparator module with:
  - Two analog comparators
  - Programmable on-chip voltage reference (V_{REF}) module
  - Programmable input multiplexing from device inputs and internal voltage reference
  - Comparator outputs are externally accessible
- Timer0: 8-bit timer/counter with 8-bit programmable prescaler
- Timer1: 16-bit timer/counter with external crystal/clock capability
- Timer2: 8-bit timer/counter with 8-bit period register, prescaler and postscaler
- Capture, Compare, PWM (CCP) module
  - Capture is 16-bit, max. resolution is 12.5 ns
  - Compare is 16-bit, max. resolution is 200 ns
  - PWM max. resolution is 10-bit
- Universal Synchronous/Asynchronous Receiver/Transmitter USART/SCI
- 16 Bytes of common RAM

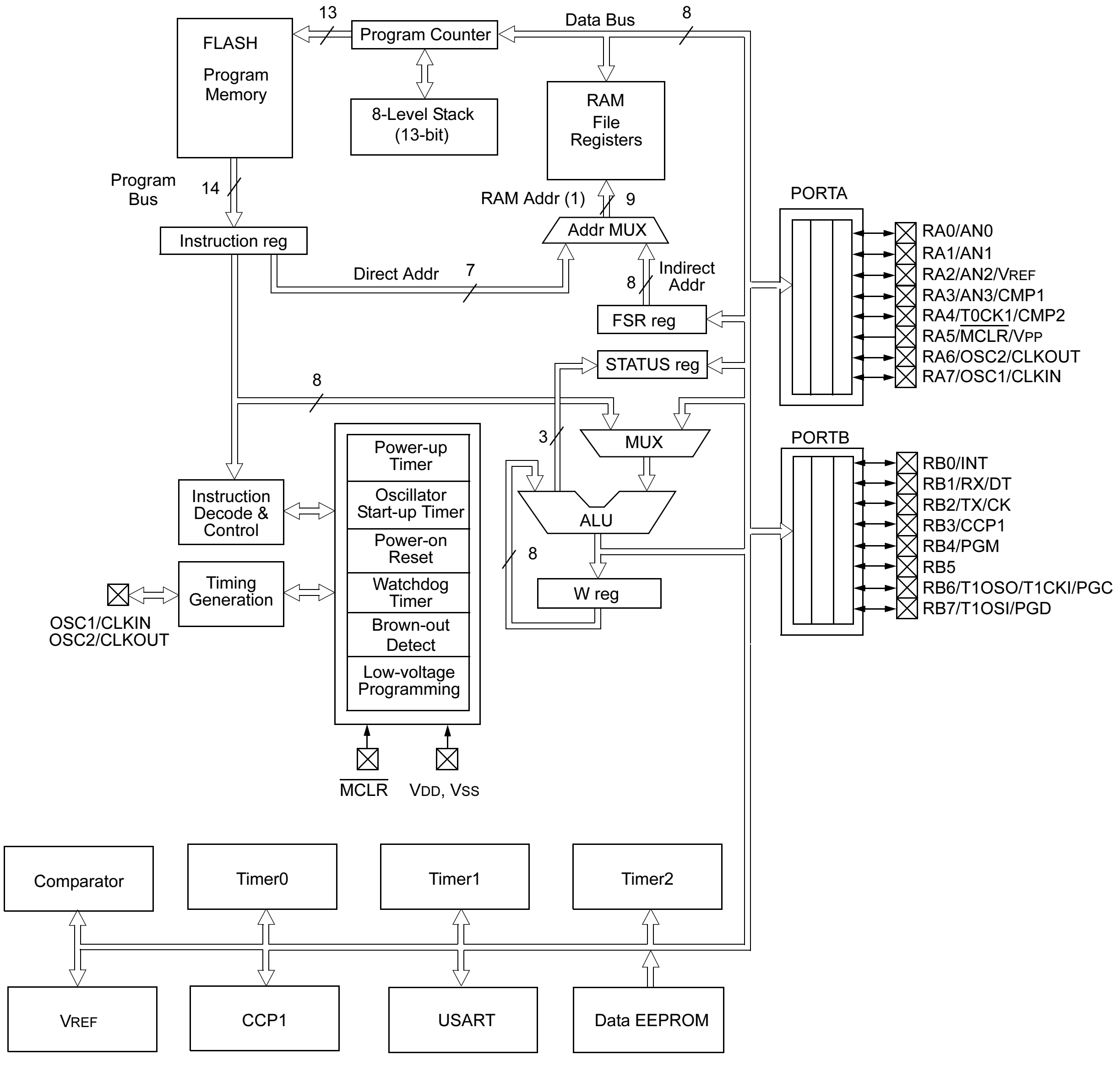
Special Microcontroller Features:

- Power-on Reset (POR)
- Power-up Timer (PWRT) and Oscillator Start-up Timer (OST)
- Brown-out Detect (BOD)
- Watchdog Timer (WDT) with its own on-chip RC oscillator for reliable operation
- Multiplexed $\overline{MCLR}$-pin
- Programmable weak pull-ups on PORTB
- Programmable code protection
- Low voltage programming
- Power saving SLEEP mode
- Selectable oscillator options
  - FLASH configuration bits for oscillator options
  - ER (External Resistor) oscillator
    - Reduced part count
  - Dual speed INTRC
    - Lower current consumption
  - EC External Clock input
  - XT Oscillator mode
  - HS Oscillator mode
  - LP Oscillator mode
- In-circuit Serial Programming™ (via two pins)
- Four user programmable ID locations

CMOS Technology:

- Low power, high speed CMOS FLASH technology
- Fully static design
- Wide operating voltage range
  - PIC16F627 - 3.0V to 5.5V
  - PIC16F628 - 3.0V to 5.5V
  - PIC16LF627 - 2.0V to 5.5V
  - PIC16LF628 - 2.0V to 5.5V
- Commercial, industrial and extended temperature range
- Low power consumption
  - < 2.0 mA @ 5.0V, 4.0 MHz
  - 15 μA typical @ 3.0V, 32 kHz
  - < 1.0 μA typical standby current @ 3.0V

FIGURE 2-1: BLOCK DIAGRAM



Note 1: Higher order bits are from the STATUS register.

TABLE 3-2: SPECIAL FUNCTION REGISTERS SUMMARY BANK 1

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR Reset(1)	Details on Page
Bank 1											
80h	INDF	Addressing this location uses contents of FSR to address data memory (not a physical register)								xxxx xxxx	25
81h	OPTION	$\overline{RBPU}$	INTEDG	T0CS	T0SE	PSA	PS2	PS1	PS0	1111 1111	20
82h	PCL	Program Counter's (PC) Least Significant Byte								0000 0000	25
83h	STATUS	IRP	RP1	RP0	$\overline{TO}$	$\overline{PD}$	Z	DC	C	0001 1xxx	19
84h	FSR	Indirect data memory address pointer								xxxx xxxx	25
85h	TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	1111 1111	29
86h	TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	1111 1111	34
87h	—	Unimplemented								—	—
88h	—	Unimplemented								—	—
89h	—	Unimplemented								—	—
8Ah	PCLATH	—	—	—	Write buffer for upper 5 bits of program counter					---0 0000	25
8Bh	INTCON	GIE	PEIE	T0IE	INTE	RBIE	T0IF	INTF	RBIF	0000 000x	21
8Ch	PIE1	EEIE	CMIE	RCIE	TXIE	—	CCP1IE	TMR2IE	TMR1IE	0000 -000	22
8Dh	—	Unimplemented								—	—
8Eh	PCON	—	—	—	—	OSCF	—	$\overline{POR}$	$\overline{BOD}$	---- 1-0x	24
8Fh	—	Unimplemented								—	—
90h	—	Unimplemented								—	—
91h	—	Unimplemented								—	—
92h	PR2	Timer2 Period Register								1111 1111	50
93h	—	Unimplemented								—	—
94h	—	Unimplemented								—	—
95h	—	Unimplemented								—	—
96h	—	Unimplemented								—	—
97h	—	Unimplemented								—	—
98h	TXSTA	CSRC	TX9	TXEN	SYNC	—	BRGH	TRMT	TX9D	0000 -010	69
99h	SPBRG	Baud Rate Generator Register								0000 0000	69
9Ah	EEDATA	EEPROM data register								xxxx xxxx	87
9Bh	EEADR	—	EEPROM address register							xxxx xxxx	87
9Ch	EECON1	—	—	—	—	WRERR	WREN	WR	RD	---- x000	87
9Dh	EECON2	EEPROM control register 2 (not a physical register)								--------	87
9Eh	—	Unimplemented								—	—
9Fh	VRCON	VREN	VROE	VRR	—	VR3	VR2	VR1	VR0	000- 0000	59

Legend: — = Unimplemented locations read as ‘0’, u = unchanged, x = unknown, q = value depends on condition, shaded = unimplemented

Note 1: For the Initialization Condition for Registers Tables, refer to Table 14-7 and Table 14-8 on page 98.

3.2.2.2 OPTION Register

The OPTION register is a readable and writable register which contains various control bits to configure the TMR0/WDT prescaler, the external RB0/INT interrupt, TMR0, and the weak pull-ups on PORTB.

> **Note:** To achieve a 1:1 prescaler assignment for TMR0, assign the prescaler to the WDT (PSA = 1). See Section 6.3.1

REGISTER 3-2: OPTION REGISTER (ADDRESS: 81h, 181h)

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
$\overline{\text{RBPU}}$	INTEDG	T0CS	T0SE	PSA	PS2	PS1	PS0
bit 7							bit 0

bit 7 **$\overline{\textbf{RBPU}}$**: PORTB Pull-up Enable bit
- 1 = PORTB pull-ups are disabled
- 0 = PORTB pull-ups are enabled by individual port latch values

bit 6 **INTEDG**: Interrupt Edge Select bit
- 1 = Interrupt on rising edge of RB0/INT pin
- 0 = Interrupt on falling edge of RB0/INT pin

bit 5 **T0CS**: TMR0 Clock Source Select bit
- 1 = Transition on RA4/T0CKI pin
- 0 = Internal instruction cycle clock (CLKOUT)

bit 4 **T0SE**: TMR0 Source Edge Select bit
- 1 = Increment on high-to-low transition on RA4/T0CKI pin
- 0 = Increment on low-to-high transition on RA4/T0CKI pin

bit 3 **PSA**: Prescaler Assignment bit
- 1 = Prescaler is assigned to the WDT
- 0 = Prescaler is assigned to the Timer0 module

bit 2-0 **PS2:PS0**: Prescaler Rate Select bits

Bit Value	TMR0 Rate	WDT Rate
000	1 : 2	1 : 1
001	1 : 4	1 : 2
010	1 : 8	1 : 4
011	1 : 16	1 : 8
100	1 : 32	1 : 16
101	1 : 64	1 : 32
110	1 : 128	1 : 64
111	1 : 256	1 : 128

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’	
-n = Value at POR	’1’ = Bit is set	’0’ = Bit is cleared	x = Bit is unknown

3.2.2.6 PCON Register

The PCON register contains flag bits to differentiate between a Power-on Reset, an external $\overline{\text{MCLR}}$ Reset, WDT Reset or a Brown-out Detect.

> **Note:** $\overline{\text{BOD}}$ is unknown on Power-on Reset. It must then be set by the user and checked on subsequent RESETS to see if $\overline{\text{BOD}}$ is cleared, indicating a brown-out has occurred. The $\overline{\text{BOD}}$ STATUS bit is a “don't care” and is not necessarily predictable if the brown-out circuit is disabled (by clearing the BODEN bit in the Configuration word).

REGISTER 3-6: PCON REGISTER (ADDRESS: 0Ch)

U-0	U-0	U-0	U-0	R/W-1	U-0	R/W-q	R/W-q
—	—	—	—	OSCF	—	$\overline{\text{POR}}$	$\overline{\text{BOD}}$
bit 7							bit 0

bit 7-4 **Unimplemented:** Read as '0'

bit 3 **OSCF**: INTRC/ER oscillator frequency
1 = 4 MHz typical(1)
0 = 37 KHz typical

bit 2 **Unimplemented:** Read as '0'

bit 1 **$\overline{\text{POR}}$**: Power-on Reset STATUS bit
1 = No Power-on Reset occurred
0 = A Power-on Reset occurred (must be set in software after a Power-on Reset occurs)

bit 0 **$\overline{\text{BOD}}$**: Brown-out Detect STATUS bit
1 = No Brown-out Reset occurred
0 = A Brown-out Reset occurred (must be set in software after a Brown-out Reset occurs)

Note 1: When in ER Oscillator mode, setting OSCF = 1 will cause the oscillator frequency to change to the frequency specified by the external resistor.

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’	
-n = Value at POR	’1’ = Bit is set	’0’ = Bit is cleared	x = Bit is unknown

6.0 TIMER0 MODULE

The Timer0 module timer/counter has the following features:

- 8-bit timer/counter
- Readable and writable
- 8-bit software programmable prescaler
- Internal or external clock select
- Interrupt on overflow from FFh to 00h
- Edge select for external clock

Figure 6-1 is a simplified block diagram of the Timer0 module. Additional information available in the PICmicro™ Mid-Range MCU Family Reference Manual, DS31010A.

Timer mode is selected by clearing the T0CS bit (OPTION<5>). In Timer mode, the TMR0 will increment every instruction cycle (without prescaler). If Timer0 is written, the increment is inhibited for the following two cycles. The user can work around this by writing an adjusted value to TMR0.

Counter mode is selected by setting the T0CS bit. In this mode Timer0 will increment either on every rising or falling edge of pin RA4/T0CKI. The incrementing edge is determined by the source edge (T0SE) control bit (OPTION<4>). Clearing the T0SE bit selects the rising edge. Restrictions on the external clock input are discussed in detail in Section 6.2.

The prescaler is shared between the Timer0 module and the Watchdog Timer. The prescaler assignment is controlled in software by the control bit PSA (OPTION<3>). Clearing the PSA bit will assign the prescaler to Timer0. The prescaler is not readable or writable. When the prescaler is assigned to the Timer0 module, prescale value of 1:2, 1:4,..., 1:256 are selectable. Section 6.3 details the operation of the prescaler.

6.1 TIMER0 Interrupt

Timer0 interrupt is generated when the TMR0 register timer/counter overflows from FFh to 00h. This overflow sets the T0IF bit. The interrupt can be masked by clearing the T0IE bit (INTCON<5>). The T0IF bit (INTCON<2>) must be cleared in software by the Timer0 module interrupt service routine before re-enabling this interrupt. The Timer0 interrupt cannot wake the processor from SLEEP since the timer is shut off during SLEEP.

6.2 Using Timer0 with External Clock

When an external clock input is used for Timer0, it must meet certain requirements. The external clock requirement is due to internal phase clock (TOSC) synchronization. Also, there is a delay in the actual incrementing of Timer0 after synchronization.

6.2.1 EXTERNAL CLOCK SYNCHRONIZATION

When no prescaler is used, the external clock input is the same as the prescaler output. The synchronization of T0CKI with the internal phase clocks is accomplished by sampling the prescaler output on the Q2 and Q4 cycles of the internal phase clocks (Figure 6-1). Therefore, it is necessary for T0CKI to be high for at least 2TOSC (and a small RC delay of 20 ns) and low for at least 2TOSC (and a small RC delay of 20 ns). Refer to the electrical specification of the desired device.

When a prescaler is used, the external clock input is divided by the asynchronous ripple-counter type prescaler so that the prescaler output is symmetrical. For the external clock to meet the sampling requirement, the ripple-counter must be taken into account. Therefore, it is necessary for T0CKI to have a period of at least 4TOSC (and a small RC delay of 40 ns) divided by the prescaler value. The only requirement on T0CKI high and low time is that they do not violate the minimum pulse width requirement of 10 ns. Refer to parameters 40, 41 and 42 in the electrical specification of the desired device. See Table 17-7.

The code example in Example 9-1 depicts the steps required to configure the Comparator module. RA3 and RA4 are configured as digital output. RA0 and RA1 are configured as the V- inputs and RA2 as the V+ input to both comparators.

EXAMPLE 9-1: INITIALIZING COMPARATOR MODULE

```
FLAG_REG EQU         0X20
CLRF     FLAG_REG    ;Init flag register
CLRF     PORTA       ;Init PORTA
MOVF     CMCON, W    ;Load comparator bits
ANDLW    0xC0        ;Mask comparator bits
IORWF    FLAG_REG,F  ;Store bits in flag register
MOVLW    0x03        ;Init comparator mode
MOVWF    CMCON       ;CM<2:0> = 011
BSF      STATUS,RP0  ;Select Bank1
MOVLW    0x07        ;Initialize data direction
MOVWF    TRISA       ;Set RA<2:0> as inputs
                     ;RA<4:3> as outputs
                     ;TRISA<7:5> always read '0'
BCF      STATUS,RP0  ;Select Bank 0
CALL     DELAY10     ;10µs delay
MOVF     CMCON,F     ;Read CMCON to end change condition
BCF      PIR1,CMIF   ;Clear pending interrupts
BSF      STATUS,RP0  ;Select Bank 1
BSF      PIE1,CMIE   ;Enable comparator interrupts
BCF      STATUS,RP0  ;Select Bank 0
BSF      INTCON,PEIE ;Enable peripheral interrupts
BSF      INTCON,GIE  ;Global interrupt enable
```

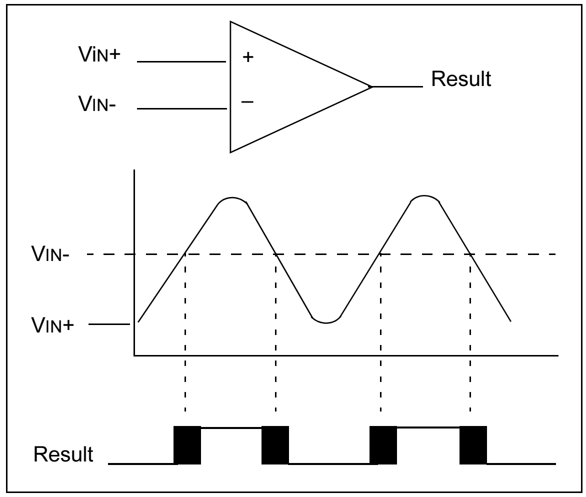
9.2 Comparator Operation

A single comparator is shown in Figure 9-2 along with the relationship between the analog input levels and the digital output. When the analog input at VIN+ is less than the analog input VIN-, the output of the comparator is a digital low level. When the analog input at VIN+ is greater than the analog input VIN-, the output of the comparator is a digital high level. The shaded areas of the output of the comparator in Figure 9-2 represent the uncertainty due to input offsets and response time.

9.3 Comparator Reference

An external or internal reference signal may be used depending on the Comparator Operating mode. The analog signal that is present at VIN- is compared to the signal at VIN+, and the digital output of the comparator is adjusted accordingly (Figure 9-2).

FIGURE 9-2: SINGLE COMPARATOR



9.3.1 EXTERNAL REFERENCE SIGNAL

When external voltage references are used, the comparator module can be configured to have the comparators operate from the same or different reference sources. However, threshold detector applications may require the same reference. The reference signal must be between VSS and VDD, and can be applied to either pin of the comparator(s).

9.3.2 INTERNAL REFERENCE SIGNAL

The Comparator module also allows the selection of an internally generated voltage reference for the comparators. Section 10.0, Voltage Reference Manual, contains a detailed description of the Voltage Reference module that provides this signal. The internal reference signal is used when the comparators are in mode CM<2:0>=010 (Figure 9-1). In this mode, the internal voltage reference is applied to the VIN+ pin of both comparators.

9.4 Comparator Response Time

Response time is the minimum time, after selecting a new reference voltage or input source, before the comparator output is ensured to have a valid level. If the internal reference is changed, the maximum delay of the internal voltage reference must be considered when using the comparator outputs. Otherwise the maximum delay of the comparators should be used (Table 17-1**).**

EXAMPLE 10-1: VOLTAGE REFERENCE CONFIGURATION

```
MOVLW    0x02          ; 4 Inputs Muxed
MOVWF    CMCON         ; to 2 comps.
BSF      STATUS,RP0    ; go to Bank 1
MOVLW    0x07          ; RA3-RA0 are
MOVWF    TRISA         ; outputs
MOVLW    0xA6          ; enable VREF
MOVWF    VRCON         ; low range
                       ; set VR<3:0>=6
BCF      STATUS,RP0    ; go to Bank 0
CALL     DELAY10       ; 10µs delay
```

10.2 Voltage Reference Accuracy/Error

The full range of VSS to VDD cannot be realized due to the construction of the module. The transistors on the top and bottom of the resistor ladder network (Figure 10-1) keep VREF from approaching VSS or VDD. The Voltage Reference is VDD derived and therefore, the VREF output changes with fluctuations in VDD. The tested absolute accuracy of the Voltage Reference can be found in Table 17-2.

10.3 Operation During SLEEP

When the device wakes-up from SLEEP through an interrupt or a Watchdog Timer timeout, the contents of the VRCON register are not affected. To minimize current consumption in SLEEP mode, the Voltage Reference should be disabled.

10.4 Effects of a RESET

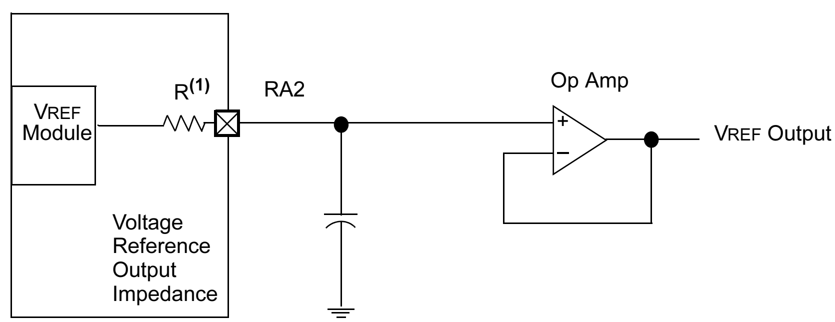
A device RESET disables the Voltage Reference by clearing bit VREN (VRCON<7>). This RESET also disconnects the reference from the RA2 pin by clearing bit VROE (VRCON<6>) and selects the high voltage range by clearing bit VRR (VRCON<5>). The VREF value select bits, VRCON<3:0>, are also cleared.

10.5 Connection Considerations

The Voltage Reference module operates independently of the Comparator module. The output of the reference generator may be connected to the RA2 pin if the TRISA<2> bit is set and the VROE bit, VRCON<6>, is set. Enabling the Voltage Reference output onto the RA2 pin with an input signal present will increase current consumption. Connecting RA2 as a digital output with VREF enabled will also increase current consumption.

The RA2 pin can be used as a simple D/A output with limited drive capability. Due to the limited drive capability, a buffer must be used in conjunction with the Voltage Reference output for external connections to VREF. Figure 10-2 shows an example buffering technique.

FIGURE 10-2: VOLTAGE REFERENCE OUTPUT BUFFER EXAMPLE



Note 1: R is dependent upon the Voltage Reference Configuration VRCON<3:0> and VRCON<5>.

TABLE 10-1: REGISTERS ASSOCIATED WITH VOLTAGE REFERENCE

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value On POR	Value On All Other RESETS
9Fh	VRCON	VREN	VROE	VRR	—	VR3	VR2	VR1	VR0	000- 0000	000- 0000
1Fh	CMCON	C2OUT	C1OUT	C2INV	C1INV	CIS	CM2	CM1	CM0	0000 0000	0000 0000
85h	TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	1111 1111	1111 1111

Note 1: — = Unimplemented, read as ‘0’.

12.0 UNIVERSAL SYNCHRONOUS/ ASYNCHRONOUS RECEIVER/ TRANSMITTER (USART) MODULE

The Universal Synchronous Asynchronous Receiver Transmitter (USART) module is one of the two serial I/O modules. (USART is also known as a Serial Communications Interface or SCI). The USART can be configured as a full duplex asynchronous system that can communicate with peripheral devices such as CRT terminals and personal computers, or it can be configured as a half duplex synchronous system that can communicate with peripheral devices such as A/D or D/A integrated circuits, Serial EEPROMs etc.

The USART can be configured in the following modes:

- Asynchronous (full duplex)
- Synchronous - Master (half duplex)
- Synchronous - Slave (half duplex)

Bit SPEN (RCSTA<7>), and bits TRISB<2:1>, have to be set in order to configure pins RB2/TX/CK and RB1/RX/DT as the Universal Synchronous Asynchronous Receiver Transmitter.

REGISTER 12-1: TXSTA: TRANSMIT STATUS AND CONTROL REGISTER (ADDRESS: 98h)

R/W-0	R/W-0	R/W-0	R/W-0	U-0	R/W-0	R-1	R/W-0
CSRC	TX9	TXEN	SYNC	—	BRGH	TRMT	TX9D
bit 7							bit 0

bit 7 **CSRC**: Clock Source Select bit
Asynchronous mode
Don't care
Synchronous mode
1 = Master mode (Clock generated internally from BRG)
0 = Slave mode (Clock from external source)

bit 6 **TX9**: 9-bit Transmit Enable bit
1 = Selects 9-bit transmission
0 = Selects 8-bit transmission

bit 5 **TXEN**: Transmit Enable bit(1)
1 = Transmit enabled
0 = Transmit disabled

bit 4 **SYNC**: USART Mode Select bit
1 = Synchronous mode
0 = Asynchronous mode

bit 3 **Unimplemented:** Read as '0'

bit 2 **BRGH**: High Baud Rate Select bit
Asynchronous mode
1 = High speed
0 = Low speed
Synchronous mode
Unused in this mode

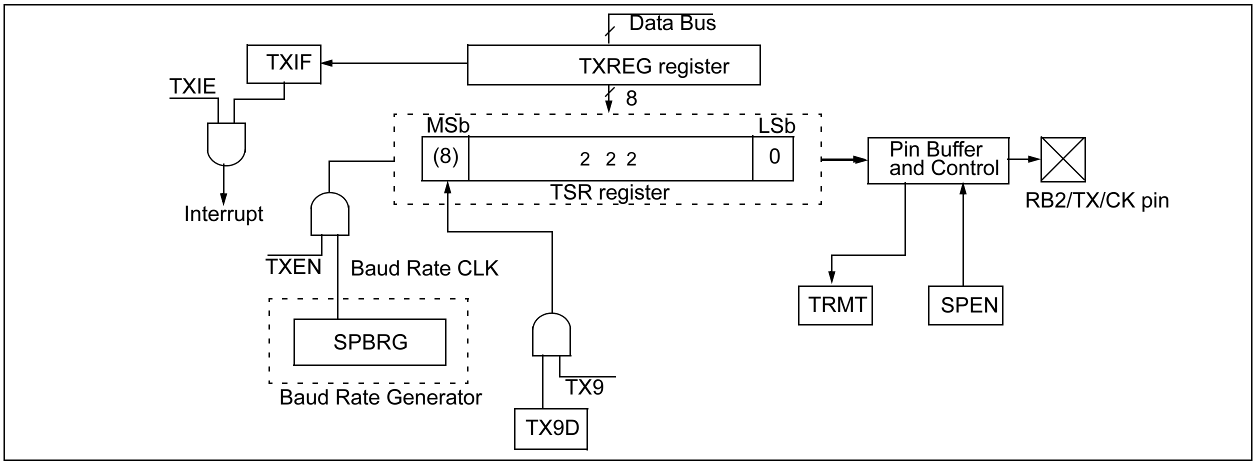
bit 1 **TRMT**: Transmit Shift Register STATUS bit
1 = TSR empty
0 = TSR full

bit 0 **TX9D**: 9th bit of transmit data. Can be PARITY bit.

Note 1: SREN/CREN overrides TXEN in SYNC mode.

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

FIGURE 12-5: USART TRANSMIT BLOCK DIAGRAM



Steps to follow when setting up an Asynchronous Transmission:

1. Initialize the SPBRG register for the appropriate baud rate. If a high speed baud rate is desired, set bit BRGH. (Section 12.1)
2. Enable the asynchronous serial port by clearing bit SYNC and setting bit SPEN.
3. If interrupts are desired, then set enable bit TXIE.
4. If 9-bit transmission is desired, then set transmit bit TX9.
5. Enable the transmission by setting bit TXEN, which will also set bit TXIF.
6. If 9-bit transmission is selected, the ninth bit should be loaded in bit TX9D.
7. Load data to the TXREG register (starts transmission).

FIGURE 12-6: ASYNCHRONOUS TRANSMISSION

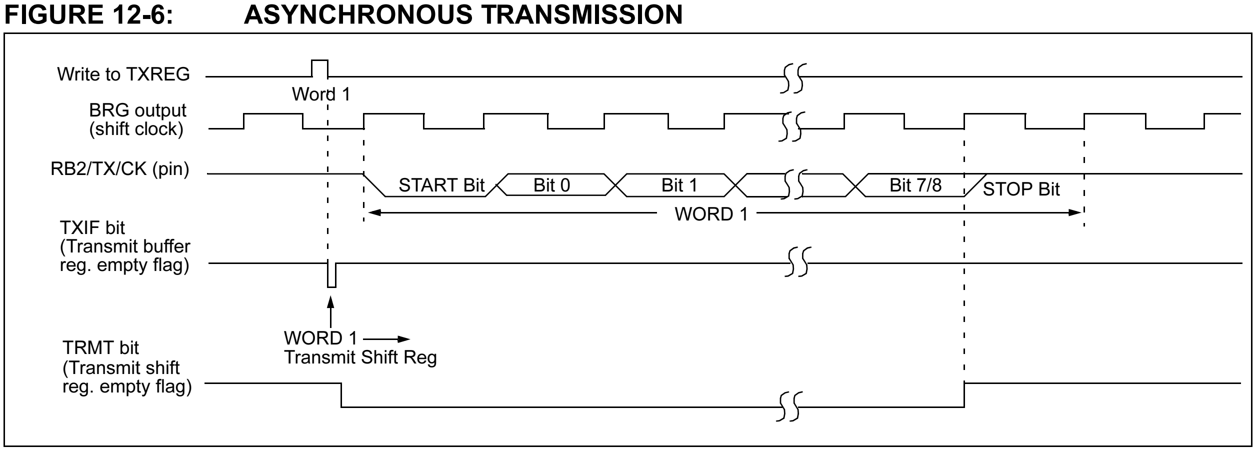
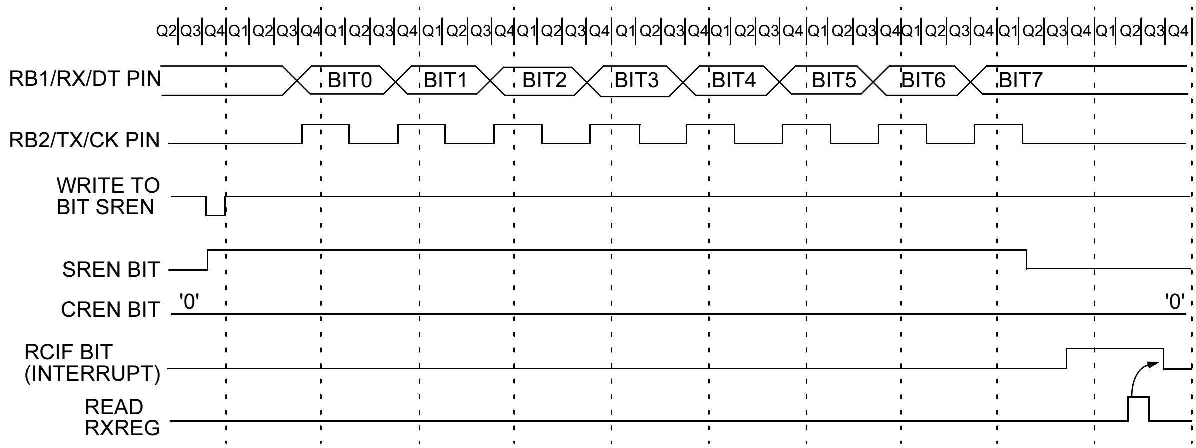


FIGURE 12-14: SYNCHRONOUS RECEPTION (MASTER MODE, SREN)



Note 1: Timing diagram demonstrates Sync Master mode with bit SREN = ‘1’ and bit BRG = ‘0’.

12.5 USART Synchronous Slave Mode

Synchronous Slave mode differs from the Master mode in the fact that the shift clock is supplied externally at the RB2/TX/CK pin (instead of being supplied internally in Master mode). This allows the device to transfer or receive data while in SLEEP mode. Slave mode is entered by clearing bit CSRC (TXSTA<7>).

12.5.1 USART SYNCHRONOUS SLAVE TRANSMIT

The operation of the Synchronous Master and Slave modes are identical except in the case of the SLEEP mode.

If two words are written to the TXREG and then the `SLEEP` instruction is executed, the following will occur:

a) The first word will immediately transfer to the TSR register and transmit.
b) The second word will remain in TXREG register.
c) Flag bit TXIF will not be set.
d) When the first word has been shifted out of TSR, the TXREG register will transfer the second word to the TSR and flag bit TXIF will now be set.
e) If enable bit TXIE is set, the interrupt will wake the chip from SLEEP and if the global interrupt is enabled, the program will branch to the interrupt vector (0004h).

Steps to follow when setting up a Synchronous Slave Transmission:

1. Enable the synchronous slave serial port by setting bits SYNC and SPEN and clearing bit CSRC.
2. Clear bits CREN and SREN.
3. If interrupts are desired, then set enable bit TXIE.
4. If 9-bit transmission is desired, then set bit TX9.
5. Enable the transmission by setting enable bit TXEN.
6. If 9-bit transmission is selected, the ninth bit should be loaded in bit TX9D.
7. Start transmission by loading data to the TXREG register.

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12.5.2 USART SYNCHRONOUS SLAVE RECEPTION

The operation of the Synchronous Master and Slave modes is identical except in the case of the SLEEP mode. Also, bit SREN is a don't care in Slave mode.

If receive is enabled, by setting bit CREN, prior to the `SLEEP` instruction, then a word may be received during SLEEP. On completely receiving the word, the RSR register will transfer the data to the RCREG register and if enable bit RCIE bit is set, the interrupt generated will wake the chip from SLEEP. If the global interrupt is enabled, the program will branch to the interrupt vector (0004h).

Steps to follow when setting up a Synchronous Slave Reception:

1. Enable the synchronous master serial port by setting bits SYNC and SPEN and clearing bit CSRC.
2. If interrupts are desired, then set enable bit RCIE.
3. If 9-bit reception is desired, then set bit RX9.
4. To enable reception, set enable bit CREN.
5. Flag bit RCIF will be set when reception is complete and an interrupt will be generated, if enable bit RCIE was set.
6. Read the RCSTA register to get the ninth bit (if enabled) and determine if any error occurred during reception.
7. Read the 8-bit received data by reading the RCREG register.
8. If any error occurred, clear the error by clearing bit CREN.

TABLE 12-11: REGISTERS ASSOCIATED WITH SYNCHRONOUS SLAVE TRANSMISSION

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR	Value on all other RESETS
0Ch	PIR1	EEIF	CMIF	RCIF	TXIF	—	CCP1IF	TMR2IF	TMR1IF	0000 -000	0000 -000
18h	RCSTA	SPEN	RX9	SREN	CREN	ADEN	FERR	OERR	RX9D	0000 -00x	0000 -00x
19h	TXREG	USART Transmit Register								0000 0000	0000 0000
8Ch	PIE1	EEIE	CMIE	RCIE	TXIE	—	CCP1IE	TMR2IE	TMR1IE	0000 -000	0000 -000
98h	TXSTA	CSRC	TX9	TXEN	SYNC	—	BRGH	TRMT	TX9D	0000 -010	0000 -010
99h	SPBRG	Baud Rate Generator Register								0000 0000	0000 0000

Legend: x = unknown, - = unimplemented read as '0'. Shaded cells are not used for Synchronous Slave Transmission.

TABLE 12-12: REGISTERS ASSOCIATED WITH SYNCHRONOUS SLAVE RECEPTION

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR	Value on all other RESETS
0Ch	PIR1	EEIF	CMIF	RCIF	TXIF	—	CCP1IF	TMR2IF	TMR1IF	0000 -000	0000 -000
18h	RCSTA	SPEN	RX9	SREN	CREN	ADEN	FERR	OERR	RX9D	0000 -00x	0000 -00x
1Ah	RCREG	USART Receive Register								0000 0000	0000 0000
8Ch	PIE1	EEIE	CMIE	RCIE	TXIE	—	CCP1IE	TMR2IE	TMR1IE	0000 -000	0000 -000
98h	TXSTA	CSRC	TX9	TXEN	SYNC	—	BRGH	TRMT	TX9D	0000 -010	0000 -010
99h	SPBRG	Baud Rate Generator Register								0000 0000	0000 0000

Legend: x = unknown, - = unimplemented read as '0'. Shaded cells are not used for Synchronous Slave Reception.

FIGURE 14-2: EXTERNAL PARALLEL RESONANT CRYSTAL OSCILLATOR CIRCUIT

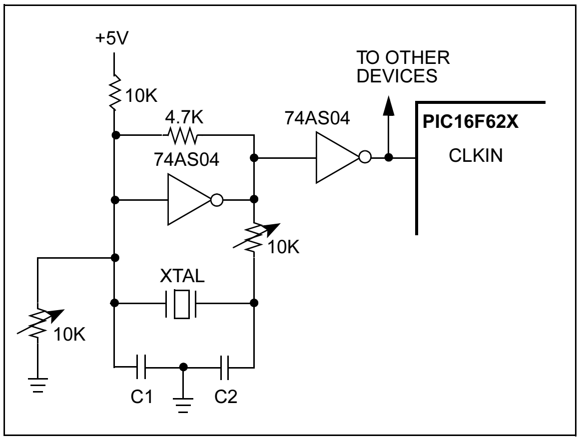
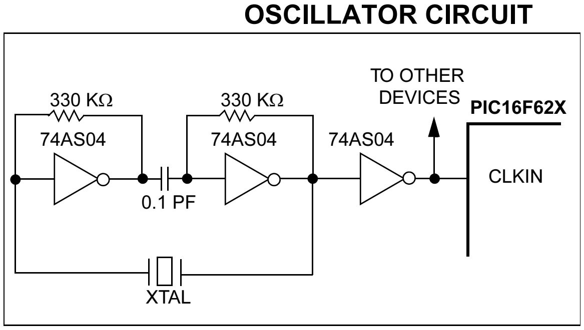


Figure 14-3 shows a series resonant oscillator circuit. This circuit is also designed to use the fundamental frequency of the crystal. The inverter performs a 180° phase shift in a series resonant oscillator circuit. The 330 kΩ resistors provide the negative feedback to bias the inverters in their linear region.

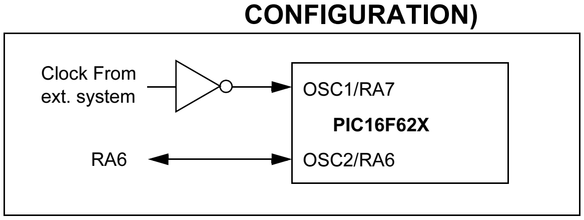
FIGURE 14-3: EXTERNAL SERIES RESONANT CRYSTAL OSCILLATOR CIRCUIT



14.2.4 EXTERNAL CLOCK IN

For applications, where a clock is already available elsewhere, users may directly drive the PIC16F62X provided that this external clock source meets the AC/DC timing requirements listed in Section 17.4. Figure 14-4 shows how an external clock circuit should be configured.

FIGURE 14-4: EXTERNAL CLOCK INPUT OPERATION (EC, HS, XT OR LP OSC CONFIGURATION)



14.2.5 ER OSCILLATOR

For timing insensitive applications, the ER (External Resistor) Clock mode offers additional cost savings. Only one external component, a resistor to VSS, is needed to set the operating frequency of the internal oscillator. The resistor draws a DC bias current which controls the oscillation frequency. In addition to the resistance value, the oscillator frequency will vary from unit to unit, and as a function of supply voltage and temperature. Since the controlling parameter is a DC current and not a capacitance, the particular package type and lead frame will not have a significant effect on the resultant frequency.

Figure 14-5 shows how the controlling resistor is connected to the PIC16F62X. For REXT values below 10k, the oscillator operation becomes sensitive to temperature. For very high REXT values (e.g., 1M), the oscillator becomes sensitive to leakage and may stop completely. Thus, we recommend keeping REXT between 10k and 1M.

FIGURE 14-5: EXTERNAL RESISTOR

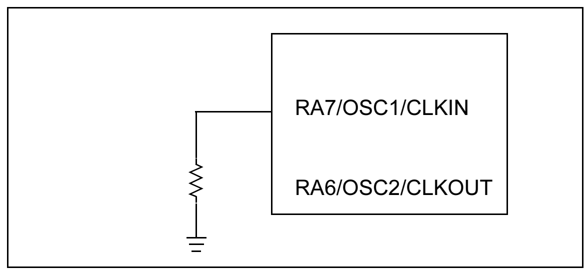


Table 14-3 shows the relationship between the resistance value and the operating frequency.

TABLE 14-3: RESISTANCE AND FREQUENCY RELATIONSHIP

Resistance	Frequency
0	10.4 MHz
1K	10 MHz
10K	7.4 MHz
20K	5.3 MHz
47K	3 MHz
100K	1.6 MHz
220K	800 kHz
470K	300 kHz
1M	200 kHz

The ER Oscillator mode has two options that control the unused OSC2 pin. The first allows it to be used as a general purpose I/O port. The other configures the pin as an output providing the FOSC signal (internal clock divided by 4) for test or external synchronization purposes.

14.12 In-Circuit Serial Programming

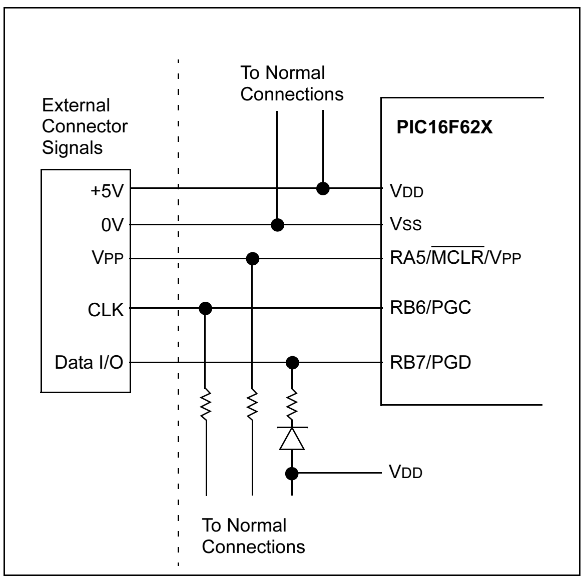
The PIC16F62X microcontrollers can be serially programmed while in the end application circuit. This is simply done with two lines for clock and data, and three other lines for power, ground, and the programming voltage. This allows customers to manufacture boards with unprogrammed devices, and then program the microcontroller just before shipping the product. This also allows the most recent firmware, or a custom firmware to be programmed.

The device is placed into a Program/Verify mode by holding the RB6 and RB7 pins low while raising the $\overline{MCLR}$ (V_{PP}) pin from V_{IL} to V_{IHH} (see programming specification). RB6 becomes the programming clock and RB7 becomes the programming data. Both RB6 and RB7 are Schmitt Trigger inputs in this mode.

After RESET, to place the device into Programming/Verify mode, the program counter (PC) is at location 00h. A 6-bit command is then supplied to the device. Depending on the command, 14 bits of program data are then supplied to or from the device, depending if the command was a load or a read. For complete details of serial programming, please refer to the Programming Specifications.

A typical in-circuit serial programming connection is shown in Figure 14-18.

FIGURE 14-18: TYPICAL IN-CIRCUIT SERIAL PROGRAMMING CONNECTION



14.13 Low Voltage Programming

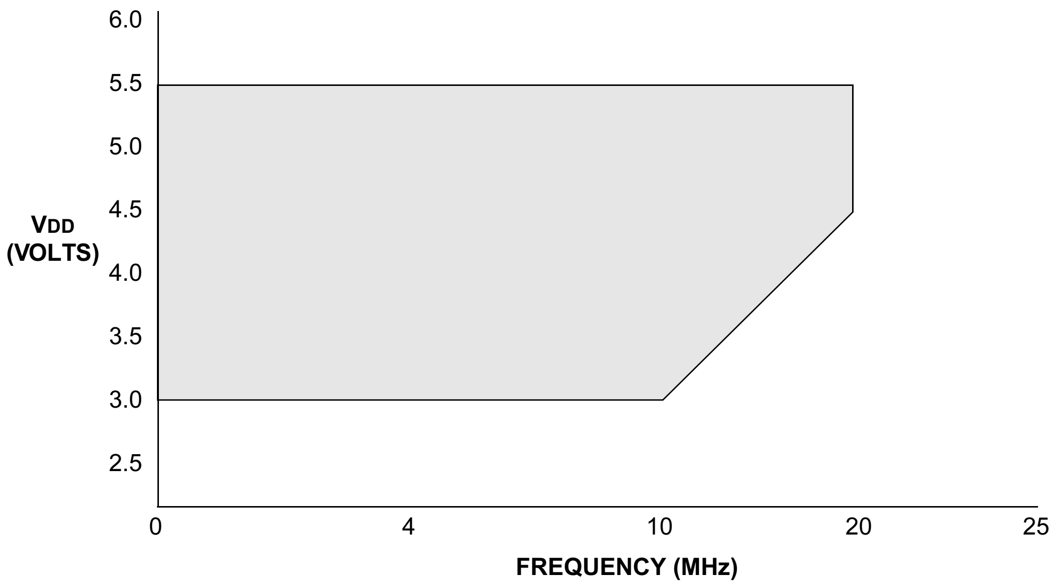
The LVP bit of the configuration word, enables the low voltage programming. This mode allows the microcontroller to be programmed via ICSP using only a 5V source. This mode removes the requirement of V_{IHH} to be placed on the $\overline{MCLR}$ pin. The LVP bit is normally erased to '1', which enables the low voltage programming. In this mode, the RB4/PGM pin is dedicated to the programming function and ceases to be a general purpose I/O pin. The device will enter Programming mode when a '1' is placed on the RB4/PGM pin. The HV Programming mode is still available by placing V_{IHH} on the $\overline{MCLR}$ pin.

> **Note 1:** While in this mode, the RB4 pin can no longer be used as a general purpose I/O pin.
>
> **2:** V_{DD} must be 5.0V $\pm$10% during erase/program operations while in low voltage Programming mode.

If Low voltage Programming mode is not used, the LVP bit can be programmed to a '0', and RB4/PGM becomes a digital I/O pin. To program the device, V_{IHH} must be placed onto $\overline{MCLR}$ during programming. The LVP bit may only be programmed when programming is entered with V_{IHH} on $\overline{MCLR}$. The LVP bit cannot be programmed when programming is entered with RB4/PGM.

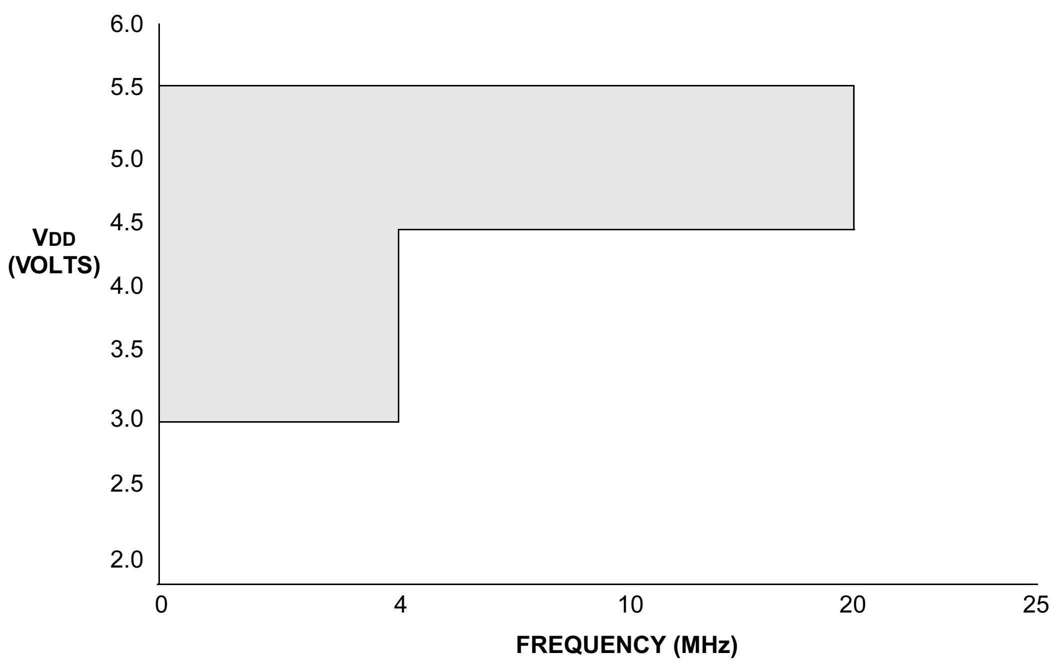
It should be noted, that once the LVP bit is programmed to 0, High voltage Programming mode can be used to program the device.

FIGURE 17-1: PIC16F62X VOLTAGE-FREQUENCY GRAPH, 0°C ≤ TA ≤ +70°C



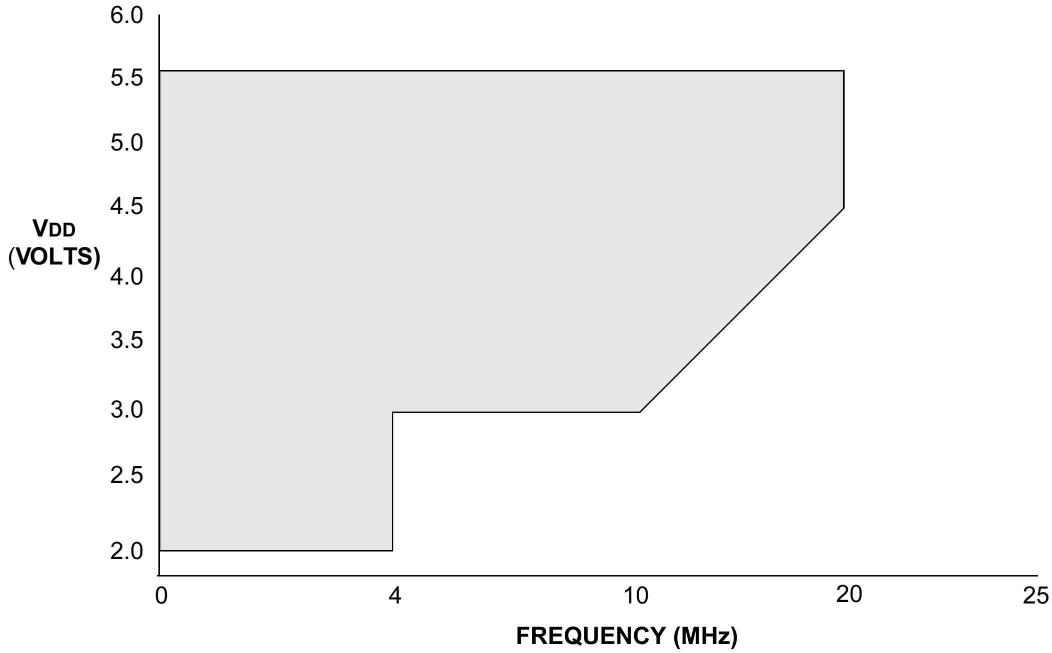
Note 1: The shaded region indicates the permissible combinations of voltage and frequency.

FIGURE 17-2: PIC16F62X VOLTAGE-FREQUENCY GRAPH, -40°C ≤ TA < 0°C, +70°C < TA ≤ 85°C



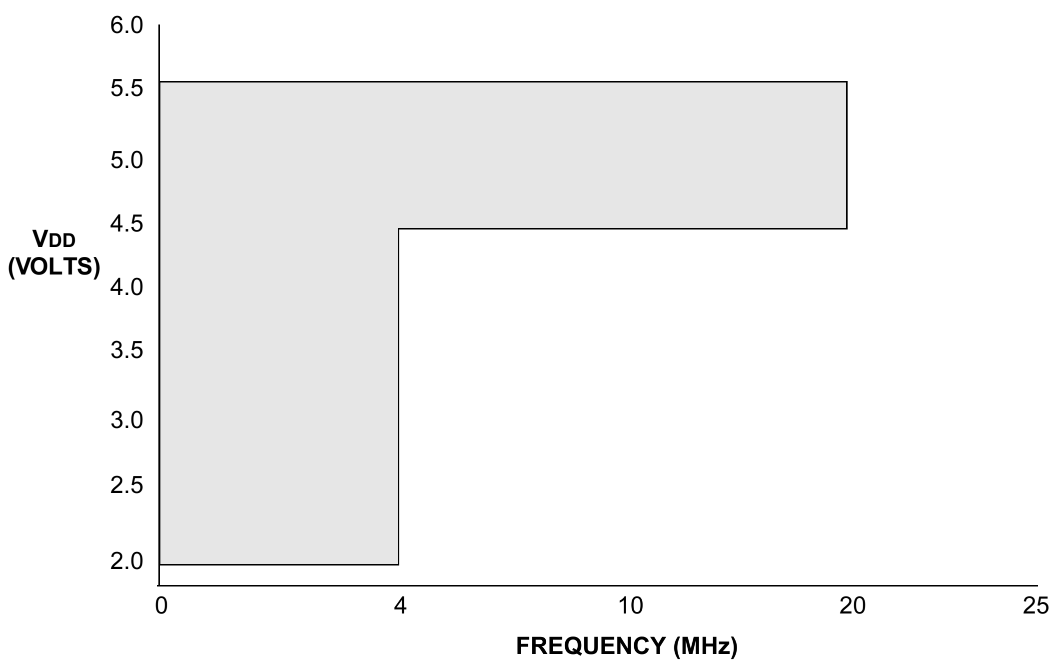
Note 1: The shaded region indicates the permissible combinations of voltage and frequency.

FIGURE 17-3: PIC16LF62X VOLTAGE-FREQUENCY GRAPH, 0°C ≤ TA ≤ +70°C



Note 1: The shaded region indicates the permissible combinations of voltage and frequency.

FIGURE 17-4: PIC16LF62X VOLTAGE-FREQUENCY GRAPH, -40°C ≤ TA < 0°C, +70°C < TA ≤ 85°C



Note 1: The shaded region indicates the permissible combinations of voltage and frequency.

TABLE 17-1: COMPARATOR SPECIFICATIONS

	Operating Conditions: 3.0V < VDD <5.5V, -40°C < TA < +125°C, unless otherwise stated.						
Param No.	**Characteristics**	**Sym**	**Min**	**Typ**	**Max**	**Units**	**Comments**
D300	Input offset voltage	VIOFF	—	±5.0	±10	mV	
D301*	Input Common mode voltage	VICM	0	—	VDD - 1.5	V	
D302*	Common Mode Rejection Ratio	CMRR	55	—	—	db	
300* 300A	Response Time(1)	TRESP	—	150	400 600	ns ns	16F62X 16LF62X
301	Comparator Mode Change to Output Valid*	TMC2OV	—	—	10	μs	

* These parameters are characterized but not tested.

Note 1: Response time measured with one comparator input at (VDD - 1.5)/2 while the other input transitions from VSS to VDD.

TABLE 17-2: VOLTAGE REFERENCE SPECIFICATIONS

	Operating Conditions: 3.0V < VDD < 5.5V, -40°C < TA < +125°C, unless otherwise stated.						
Spec No.	**Characteristics**	**Sym**	**Min**	**Typ**	**Max**	**Units**	**Comments**
D310	Resolution	VRES	VDD/24	—	VDD/32	LSb	
D311	Absolute Accuracy	VRaa	— —	— —	1/4 1/2	LSb LSb	Low Range (VRR = 1) High Range (VRR = 0)
D312*	Unit Resistor Value (R)	VRur	—	2k	—	Ω	
310*	Settling Time(1)	Tset	—	—	10	μs	

* These parameters are characterized but not tested.

Note 1: Settling time measured while VRR = 1 and VR<3:0> transitions from `0000` to `1111`.

.

Note:	The graphs and tables provided in this section are for design guidance and are not tested.

FIGURE 18-2: MAXIMUM IDD VS FOSC OVER VDD (HS MODE)

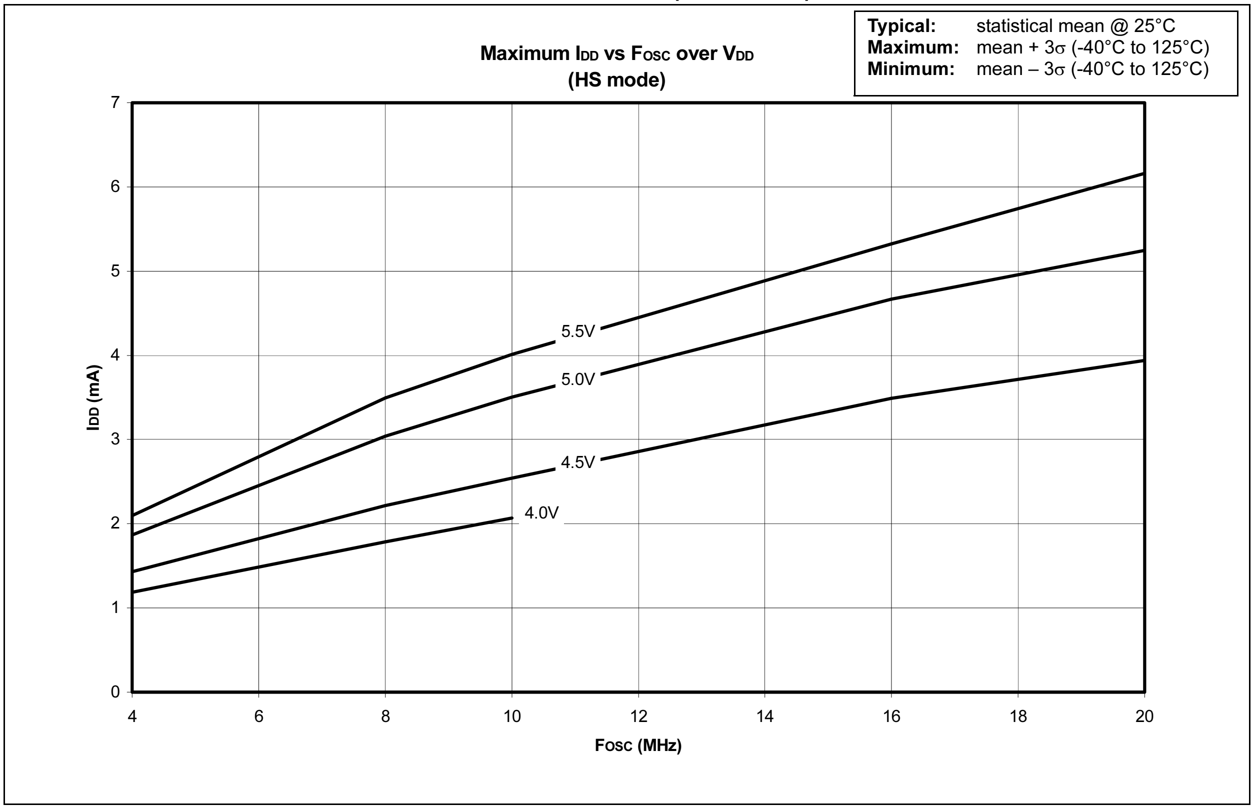
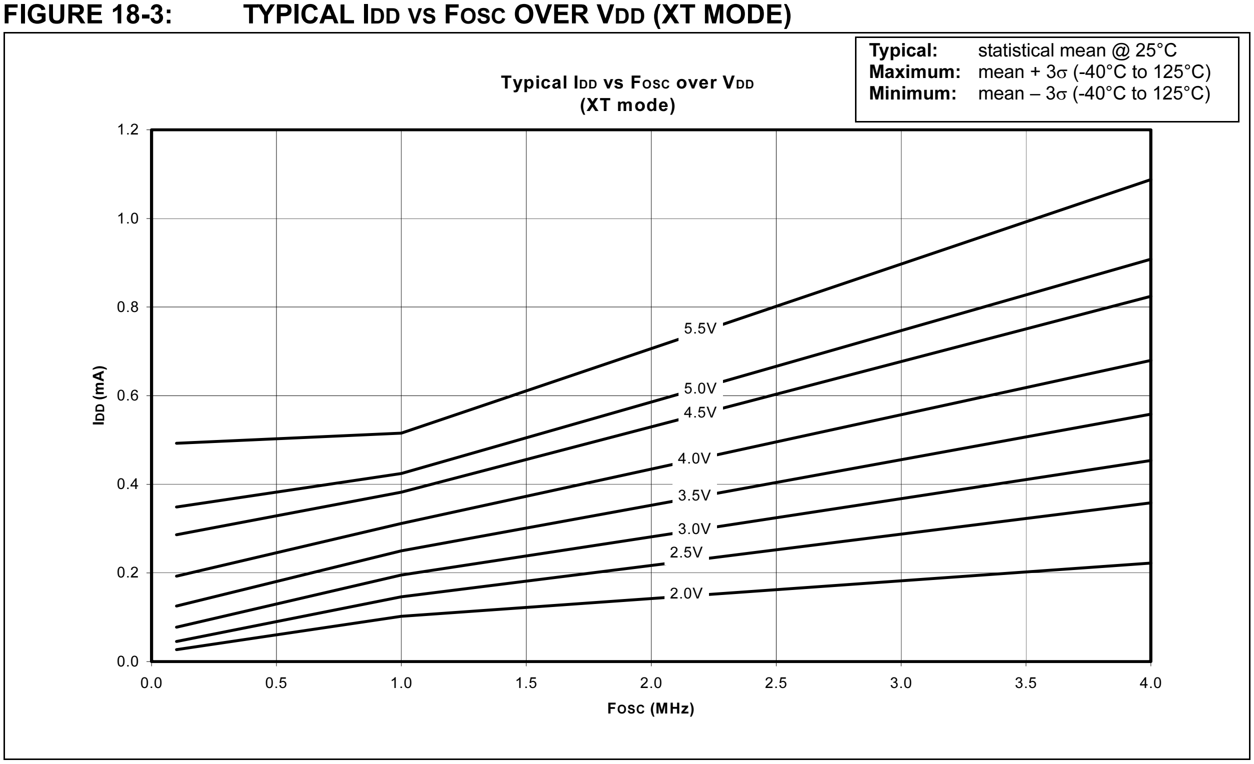


FIGURE 18-3: TYPICAL IDD VS FOSC OVER VDD (XT MODE)



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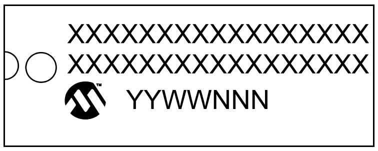
19.0 PACKAGING INFORMATION

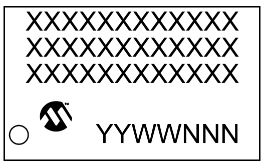
19.1 Package Marking Information

18-LEAD PDIP

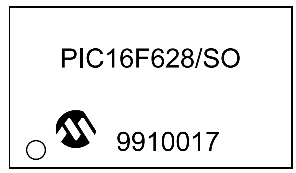


EXAMPLE

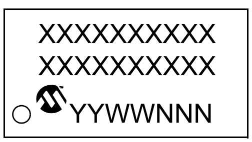
18-LEAD SOIC (.300")



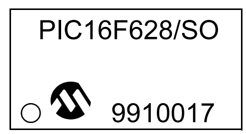
EXAMPLE



20-LEAD SSOP



EXAMPLE



Legend: MM...M	Microchip part number information
XX...X	Customer specific information(1)
YY	Year code (last 2 digits of calendar year)
WW	Week code (week of January 1 is week ‘01’)
NNN	Alphanumeric traceability code

Note: In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line thus limiting the number of available characters for customer specific information.

* Standard OTP marking consists of Microchip part number, year code, week code, facility code, mask rev#, and assembly code. For OTP marking beyond this, certain price adders apply. Please check with your Microchip Sales Office. For QTP devices, any special marking adders are included in QTP price.