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"[Embedded - Microcontrollers](#)" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "[Embedded - Microcontrollers](#)"

Details

| | |
|----------------------------|---|
| Product Status | Obsolete |
| Core Processor | PIC |
| Core Size | 8-Bit |
| Speed | 4MHz |
| Connectivity | UART/USART |
| Peripherals | Brown-out Detect/Reset, POR, PWM, WDT |
| Number of I/O | 16 |
| Program Memory Size | 1.75KB (1K x 14) |
| Program Memory Type | FLASH |
| EEPROM Size | 128 x 8 |
| RAM Size | 224 x 8 |
| Voltage - Supply (Vcc/Vdd) | 2V ~ 5.5V |
| Data Converters | - |
| Oscillator Type | Internal |
| Operating Temperature | -40°C ~ 85°C (TA) |
| Mounting Type | Surface Mount |
| Package / Case | 18-SOIC (0.295", 7.50mm Width) |
| Supplier Device Package | 18-SOIC |
| Purchase URL | https://www.e-xfl.com/product-detail/microchip-technology/pic16lf627t-04i-so |

PIC16F62X

NOTES:

TABLE 2-1: PIC16F62X PINOUT DESCRIPTION

| Name | Function | Input Type | Output Type | Description |
|-----------------|----------|------------|-------------|--|
| RA0/AN0 | RA0 | ST | CMOS | Bi-directional I/O port |
| | AN0 | AN | — | Analog comparator input |
| RA1/AN1 | RA1 | ST | CMOS | Bi-directional I/O port |
| | AN1 | AN | — | Analog comparator input |
| RA2/AN2/VREF | RA2 | ST | CMOS | Bi-directional I/O port |
| | AN2 | AN | — | Analog comparator input |
| | VREF | — | AN | VREF output |
| RA3/AN3/CMP1 | RA3 | ST | CMOS | Bi-directional I/O port |
| | AN3 | AN | — | Analog comparator input |
| | CMP1 | — | CMOS | Comparator 1 output |
| RA4/T0CKI/CMP2 | RA4 | ST | OD | Bi-directional I/O port |
| | T0CKI | ST | — | Timer0 clock input |
| | CMP2 | — | OD | Comparator 2 output |
| RA5/MCLR/VPP | RA5 | ST | — | Input port |
| | MCLR | ST | — | Master clear |
| | VPP | — | — | Programming voltage input. When configured as MCLR, this pin is an active low RESET to the device. Voltage on MCLR/VPP must not exceed VDD during normal device operation. |
| RA6/OSC2/CLKOUT | RA6 | ST | CMOS | Bi-directional I/O port |
| | OSC2 | XTAL | — | Oscillator crystal output. Connects to crystal or resonator in Crystal Oscillator mode. |
| | CLKOUT | — | CMOS | In ER/INTRC mode, OSC2 pin can output CLKOUT, which has 1/4 the frequency of OSC1 |
| RA7/OSC1/CLKIN | RA7 | ST | CMOS | Bi-directional I/O port |
| | OSC1 | XTAL | — | Oscillator crystal input |
| | CLKIN | ST | — | External clock source input. ER biasing pin. |
| RB0/INT | RB0 | TTL | CMOS | Bi-directional I/O port. Can be software programmed for internal weak pull-up. |
| | INT | ST | — | External interrupt. |
| RB1/RX/DT | RB1 | TTL | CMOS | Bi-directional I/O port. Can be software programmed for internal weak pull-up. |
| | RX | ST | — | USART receive pin |
| | DT | ST | CMOS | Synchronous data I/O. |
| RB2/TX/CK | RB2 | TTL | CMOS | Bi-directional I/O port. |
| | TX | — | CMOS | USART transmit pin |
| | CK | ST | CMOS | Synchronous clock I/O. Can be software programmed for internal weak pull-up. |
| RB3/CCP1 | RB3 | TTL | CMOS | Bi-directional I/O port. Can be software programmed for internal weak pull-up. |
| | CCP1 | ST | CMOS | Capture/Compare/PWM I/O |

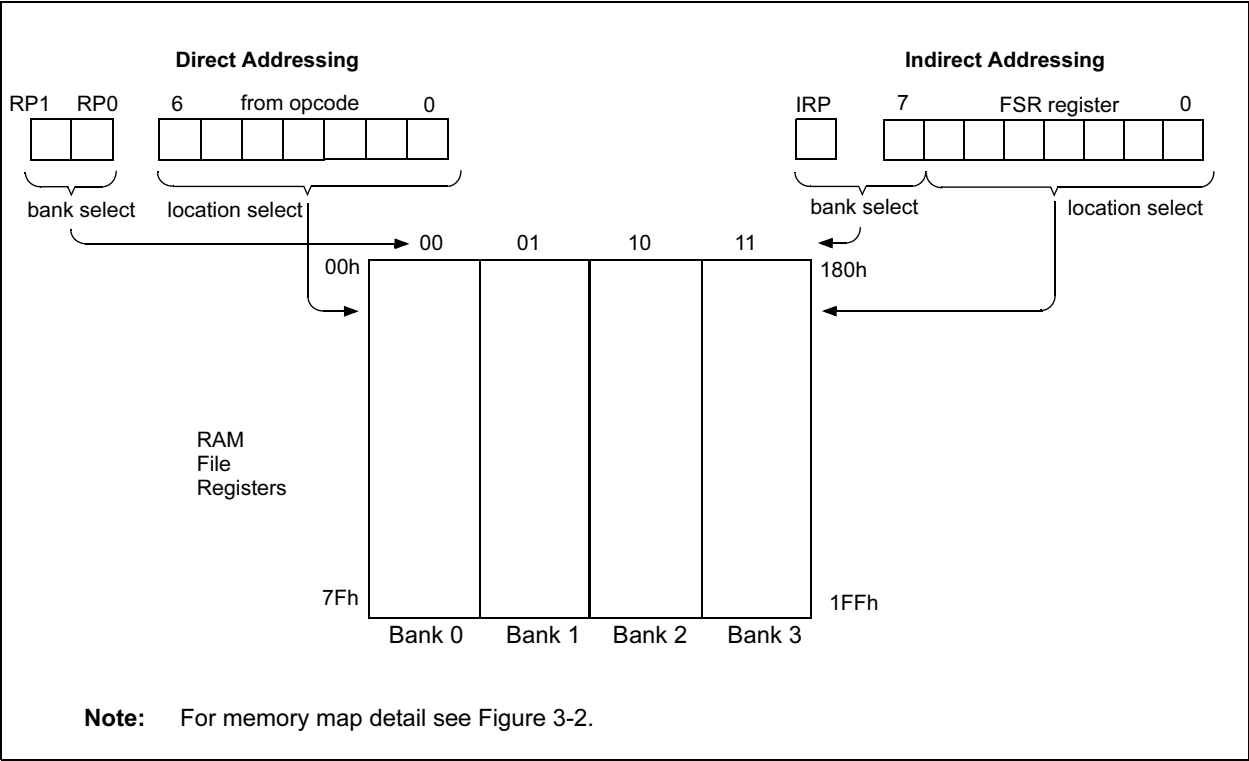
Legend: O = Output
 — = Not used
 TTL = TTL Input

CMOS = CMOS Output
 I = Input
 OD = Open Drain Output

P = Power
 ST = Schmitt Trigger Input
 AN = Analog

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FIGURE 3-4: DIRECT/INDIRECT ADDRESSING PIC16F62X



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NOTES:

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FIGURE 5-10: BLOCK DIAGRAM OF RB2/TX/CK PIN

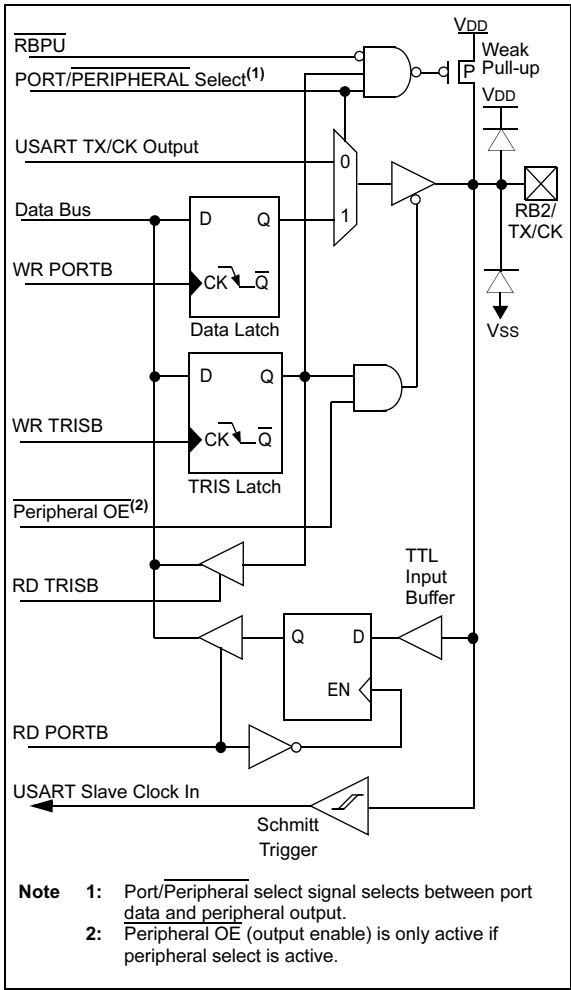
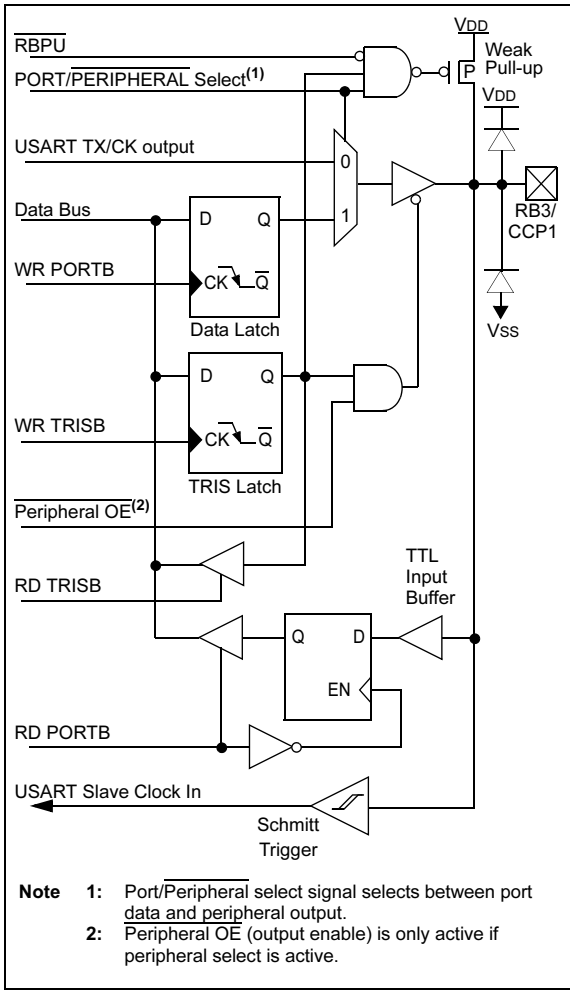


FIGURE 5-11: BLOCK DIAGRAM OF RB3/CCP1 PIN



6.0 TIMER0 MODULE

The Timer0 module timer/counter has the following features:

- 8-bit timer/counter
- Readable and writable
- 8-bit software programmable prescaler
- Internal or external clock select
- Interrupt on overflow from FFh to 00h
- Edge select for external clock

Figure 6-1 is a simplified block diagram of the Timer0 module. Additional information available in the PICmicro™ Mid-Range MCU Family Reference Manual, DS31010A.

Timer mode is selected by clearing the T0CS bit (OPTION<5>). In Timer mode, the TMR0 will increment every instruction cycle (without prescaler). If Timer0 is written, the increment is inhibited for the following two cycles. The user can work around this by writing an adjusted value to TMR0.

Counter mode is selected by setting the T0CS bit. In this mode Timer0 will increment either on every rising or falling edge of pin RA4/T0CKI. The incrementing edge is determined by the source edge (T0SE) control bit (OPTION<4>). Clearing the T0SE bit selects the rising edge. Restrictions on the external clock input are discussed in detail in Section 6.2.

The prescaler is shared between the Timer0 module and the Watchdog Timer. The prescaler assignment is controlled in software by the control bit PSA (OPTION<3>). Clearing the PSA bit will assign the prescaler to Timer0. The prescaler is not readable or writable. When the prescaler is assigned to the Timer0 module, prescale value of 1:2, 1:4,..., 1:256 are selectable. Section 6.3 details the operation of the prescaler.

6.1 TIMER0 Interrupt

Timer0 interrupt is generated when the TMR0 register timer/counter overflows from FFh to 00h. This overflow sets the T0IF bit. The interrupt can be masked by clearing the T0IE bit (INTCON<5>). The T0IF bit (INTCON<2>) must be cleared in software by the Timer0 module interrupt service routine before re-enabling this interrupt. The Timer0 interrupt cannot wake the processor from SLEEP since the timer is shut off during SLEEP.

6.2 Using Timer0 with External Clock

When an external clock input is used for Timer0, it must meet certain requirements. The external clock requirement is due to internal phase clock (Tosc) synchronization. Also, there is a delay in the actual incrementing of Timer0 after synchronization.

6.2.1 EXTERNAL CLOCK SYNCHRONIZATION

When no prescaler is used, the external clock input is the same as the prescaler output. The synchronization of T0CKI with the internal phase clocks is accomplished by sampling the prescaler output on the Q2 and Q4 cycles of the internal phase clocks (Figure 6-1). Therefore, it is necessary for T0CKI to be high for at least 2Tosc (and a small RC delay of 20 ns) and low for at least 2Tosc (and a small RC delay of 20 ns). Refer to the electrical specification of the desired device.

When a prescaler is used, the external clock input is divided by the asynchronous ripple-counter type prescaler so that the prescaler output is symmetrical. For the external clock to meet the sampling requirement, the ripple-counter must be taken into account. Therefore, it is necessary for T0CKI to have a period of at least 4Tosc (and a small RC delay of 40 ns) divided by the prescaler value. The only requirement on T0CKI high and low time is that they do not violate the minimum pulse width requirement of 10 ns. Refer to parameters 40, 41 and 42 in the electrical specification of the desired device. See Table 17-7.

6.3.1 SWITCHING PRESCALER ASSIGNMENT

The prescaler assignment is fully under software control (i.e., it can be changed “on the fly” during program execution). Use the instruction sequences, shown in Example 6-1, when changing the prescaler assignment from Timer0 to WDT, to avoid an unintended device RESET.

EXAMPLE 6-1: CHANGING PRESCALER (TIMER0→WDT)

```
BCF     STATUS, RP0    ;Skip if already in
                        ;Bank 0
CLRWDT                     ;Clear WDT
CLRWF   TMR0           ;Clear TMR0 & Prescaler
BSF     STATUS, RP0    ;Bank 1
MOVLW   '00101111'b    ;These 3 lines
                        ;(5, 6, 7)
MOVWF   OPTION_REG     ;are required only
                        ;if desired PS<2:0>
                        ;are
CLRWDT                     ;000 or 001
MOVLW   '00101xxx'b    ;Set Postscaler to
MOVWF   OPTION_REG     ;desired WDT rate
BCF     STATUS, RP0    ;Return to Bank 0
```

To change prescaler from the WDT to the TMR0 module use the sequence shown in Example 6-2. This precaution must be taken even if the WDT is disabled.

EXAMPLE 6-2: CHANGING PRESCALER (WDT→TIMER0)

```
CLRWDT                     ;Clear WDT and
                        ;prescaler
BSF     STATUS, RP0
MOVLW   b'xxxx0xxx'    ;Select TMR0, new
                        ;prescale value and
                        ;clock source
MOVWF   OPTION_REG
BCF     STATUS, RP0
```

TABLE 6-1: REGISTERS ASSOCIATED WITH TIMER0

| Address | Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Value on POR | Value on All Other RESETS |
|-----------------------|-----------------------|------------------------|--------|--------|--------|--------|--------|--------|--------|--------------|---------------------------|
| 01h | TMR0 | Timer0 module register | | | | | | | | xxxx xxxx | uuuu uuuu |
| 0Bh/8Bh/ 10Bh/18Bh | INTCON | GIE | PEIE | T0IE | INTE | RBIE | T0IF | INTF | RBIF | 0000 000x | 0000 000u |
| 81h, 181h | OPTION ⁽²⁾ | <u>RBPU</u> | INTEDG | T0CS | T0SE | PSA | PS2 | PS1 | PS0 | 1111 1111 | 1111 1111 |
| 85h | TRISA | TRISA7 | TRISA6 | TRISA5 | TRISA4 | TRISA3 | TRISA2 | TRISA1 | TRISA0 | 1111 1111 | 1111 1111 |

Legend: — = Unimplemented locations, read as '0', u = unchanged, x = unknown

Note 1: Shaded bits are not used by TMR0 module.

2: Option is referred by OPTION_REG in MPLAB.

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REGISTER 12-2: RCSTA: RECEIVE STATUS AND CONTROL REGISTER (ADDRESS: 18h)

| R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R-0 | R-0 | R-x |
|-------|-------|-------|-------|-------|------|------|-------|
| SPEN | RX9 | SREN | CREN | ADEN | FERR | OERR | RX9D |
| bit 7 | | | | | | | bit 0 |

- bit 7 **SPEN**: Serial Port Enable bit
(Configures RB1/RX/DT and RB2/TX/CK pins as serial port pins when bits TRISB<2:17> are set)
1 = Serial port enabled
0 = Serial port disabled
- bit 6 **RX9**: 9-bit Receive Enable bit
1 = Selects 9-bit reception
0 = Selects 8-bit reception
- bit 5 **SREN**: Single Receive Enable bit
Asynchronous mode:
Don't care
Synchronous mode - master:
1 = Enables single receive
0 = Disables single receive
This bit is cleared after reception is complete.
Synchronous mode - slave:
Unused in this mode
- bit 4 **CREN**: Continuous Receive Enable bit
Asynchronous mode:
1 = Enables continuous receive
0 = Disables continuous receive
Synchronous mode:
1 = Enables continuous receive until enable bit CREN is cleared (CREN overrides SREN)
0 = Disables continuous receive
- bit 3 **ADEN**: Address Detect Enable bit
Asynchronous mode 9-bit (RX9 = 1):
1 = Enables address detection, enable interrupt and load of the receive buffer when RSR<8> is set
0 = Disables address detection, all bytes are received, and ninth bit can be used as PARITY bit
Asynchronous mode 8-bit (RX9=0):
Unused in this mode
Synchronous mode
Unused in this mode
- bit 2 **FERR**: Framing Error bit
1 = Framing error (Can be updated by reading RCREG register and receive next valid byte)
0 = No framing error
- bit 1 **OERR**: Overrun Error bit
1 = Overrun error (Can be cleared by clearing bit CREN)
0 = No overrun error
- bit 0 **RX9D**: 9th bit of received data (Can be PARITY bit)

Legend:

| | | |
|-------------------|------------------|--|
| R = Readable bit | W = Writable bit | U = Unimplemented bit, read as '0' |
| -n = Value at POR | '1' = Bit is set | '0' = Bit is cleared x = Bit is unknown |

Steps to follow when setting up an Asynchronous Reception:

1. Initialize the SPBRG register for the appropriate baud rate. If a high speed baud rate is desired, set bit BRGH. (Section 12.1).
2. Enable the asynchronous serial port by clearing bit SYNC, and setting bit SPEN.
3. If interrupts are desired, then set enable bit RCIE.
4. If 9-bit reception is desired, then set bit RX9.
5. Enable the reception by setting bit CREN.
6. Flag bit RCIF will be set when reception is complete and an interrupt will be generated if enable bit RCIE was set.
7. Read the RCSTA register to get the ninth bit (if enabled) and determine if any error occurred during reception.
8. Read the 8-bit received data by reading the RCREG register.
9. If any error occurred, clear the error by clearing enable bit CREN.

TABLE 12-7: REGISTERS ASSOCIATED WITH ASYNCHRONOUS RECEPTION

| Address | Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Value on POR | Value on all other RESETS |
|---------|-------|------------------------------|-------|-------|-------|-------|--------|--------|--------|--------------|---------------------------|
| 0Ch | PIR1 | EEIF | CMIF | RCIF | TXIF | — | CCP1IF | TMR2IF | TMR1IF | 0000 -000 | 0000 -000 |
| 18h | RCSTA | SPEN | RX9 | SREN | CREN | ADEN | FERR | OERR | RX9D | 0000 -00x | 0000 -00x |
| 1Ah | RCREG | USART Receive Register | | | | | | | | 0000 0000 | 0000 0000 |
| 8Ch | PIE1 | EEIE | CMIE | RCIE | TXIE | — | CCP1IE | TMR2IE | TMR1IE | 0000 -000 | 0000 -000 |
| 98h | TXSTA | CSRC | TX9 | TXEN | SYNC | — | BRGH | TRMT | TX9D | 0000 -010 | 0000 -010 |
| 99h | SPBRG | Baud Rate Generator Register | | | | | | | | 0000 0000 | 0000 0000 |

Legend: x = unknown, - = unimplemented locations read as '0'. Shaded cells are not used for Asynchronous Reception.

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TABLE 12-9: REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER TRANSMISSION

| Address | Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Value on POR | Value on all other RESETS |
|---------|-------|------------------------------|-------|-------|-------|-------|--------|--------|--------|--------------|---------------------------|
| 0Ch | PIR1 | EEIF | CMIF | RCIF | TXIF | — | CCP1IF | TMR2IF | TMR1IF | 0000 -000 | 0000 -000 |
| 18h | RCSTA | SPEN | RX9 | SREN | CREN | ADEN | FERR | OERR | RX9D | 0000 -00x | 0000 -00x |
| 19h | TXREG | USART Transmit Register | | | | | | | | 0000 0000 | 0000 0000 |
| 8Ch | PIE1 | EEIE | CMIE | RCIE | TXIE | — | CCP1IE | TMR2IE | TMR1IE | 0000 -000 | 0000 -000 |
| 98h | TXSTA | CSRC | TX9 | TXEN | SYNC | — | BRGH | TRMT | TX9D | 0000 -010 | 0000 -010 |
| 99h | SPBRG | Baud Rate Generator Register | | | | | | | | 0000 0000 | 0000 0000 |

Legend: x = unknown, - = unimplemented, read as '0'. Shaded cells are not used for Synchronous Master Transmission.

FIGURE 12-12: SYNCHRONOUS TRANSMISSION

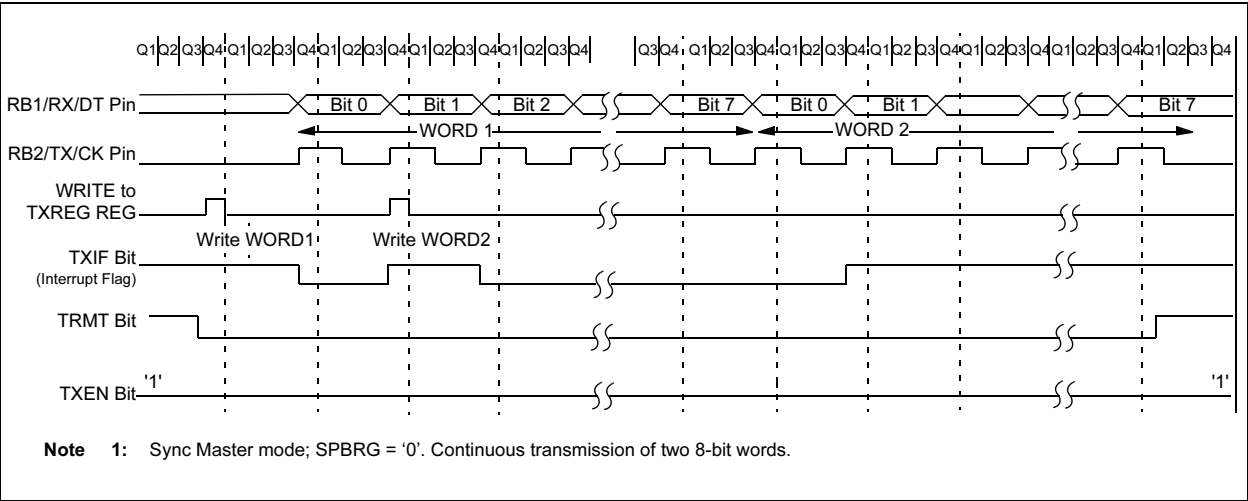
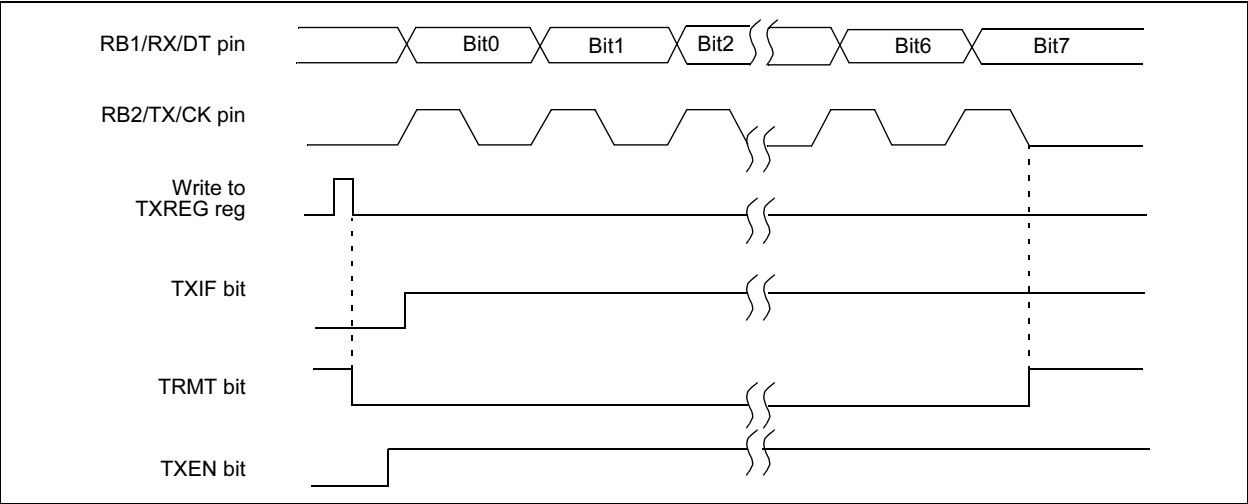


FIGURE 12-13: SYNCHRONOUS TRANSMISSION (THROUGH TXEN)



12.4.2 USART SYNCHRONOUS MASTER RECEPTION

Once Synchronous mode is selected, reception is enabled by setting either enable bit SREN (RCSTA<5>) or enable bit CREN (RCSTA<4>). Data is sampled on the RB1/RX/DT pin on the falling edge of the clock. If enable bit SREN is set, then only a single word is received. If enable bit CREN is set, the reception is continuous until CREN is cleared. If both bits are set, then CREN takes precedence. After clocking the last bit, the received data in the Receive Shift Register (RSR) is transferred to the RCREG register (if it is empty). When the transfer is complete, interrupt flag bit RCIF (PIR1<5>) is set. The actual interrupt can be enabled/disabled by setting/clearing enable bit RCIE (PIE1<5>). Flag bit RCIF is a read only bit which is RESET by the hardware. In this case, it is RESET when the RCREG register has been read and is empty. The RCREG is a double buffered register (i.e., it is a two-deep FIFO). It is possible for two bytes of data to be received and transferred to the RCREG FIFO and a third byte to begin shifting into the RSR register. On the clocking of the last bit of the third byte, if the RCREG register is still full then overrun error bit OERR (RCSTA<1>) is set. The word in the RSR will be lost. The RCREG register can be read twice to retrieve the two bytes in the FIFO. Bit OERR has to be cleared in software (by clearing bit CREN). If bit OERR is set, transfers from the RSR to the RCREG are inhibited, so it is essential to clear bit OERR if it is set. The 9th

receive bit is buffered the same way as the receive data. Reading the RCREG register, will load bit RX9D with a new value, therefore it is essential for the user to read the RCSTA register before reading RCREG in order not to lose the old RX9D information.

Steps to follow when setting up a Synchronous Master Reception:

1. Initialize the SPBRG register for the appropriate baud rate. (Section 12.1)
2. Enable the synchronous master serial port by setting bits SYNC, SPEN, and CSRC.
3. Ensure bits CREN and SREN are clear.
4. If interrupts are desired, then set enable bit RCIE.
5. If 9-bit reception is desired, then set bit RX9.
6. If a single reception is required, set bit SREN. For continuous reception set bit CREN.
7. Interrupt flag bit RCIF will be set when reception is complete and an interrupt will be generated if enable bit RCIE was set.
8. Read the RCSTA register to get the ninth bit (if enabled) and determine if any error occurred during reception.
9. Read the 8-bit received data by reading the RCREG register.
10. If any error occurred, clear the error by clearing bit CREN.

TABLE 12-10: REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER RECEPTION

| Address | Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Value on: POR | Value on all other RESETS |
|---------|-------|------------------------------|-------|-------|-------|-------|--------|--------|--------|------------------|---------------------------------|
| 0Ch | PIR1 | EEIF | CMIF | RCIF | TXIF | — | CCP1IF | TMR2IF | TMR1IF | 0000 -000 | 0000 -000 |
| 18h | RCSTA | SPEN | RX9 | SREN | CREN | ADEN | FERR | OERR | RX9D | 0000 -00x | 0000 -00x |
| 1Ah | RCREG | USART Receive Register | | | | | | | | 0000 0000 | 0000 0000 |
| 8Ch | PIE1 | EEPIE | CMIE | RCIE | TXIE | — | CCP1IE | TMR2IE | TMR1IE | -000 0000 | -000 -000 |
| 98h | TXSTA | CSRC | TX9 | TXEN | SYNC | — | BRGH | TRMT | TX9D | 0000 -010 | 0000 -010 |
| 99h | SPBRG | Baud Rate Generator Register | | | | | | | | 0000 0000 | 0000 0000 |

Legend: x = unknown, - = unimplemented read as '0'. Shaded cells are not used for Synchronous Master Reception.

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BCF Bit Clear f

Syntax: [*label*] BCF f,b

Operands: $0 \leq f \leq 127$
 $0 \leq b \leq 7$

Operation: $0 \rightarrow (f)$

Status Affected: None

Encoding:

| | | | |
|----|------|------|------|
| 01 | 00bb | bfff | ffff |
|----|------|------|------|

Description: Bit 'b' in register 'f' is cleared.

Words: 1

Cycles: 1

Example

```
BCF    REG1, 7

Before Instruction
    REG1 = 0xC7
After Instruction
    REG1 = 0x47
```

BSF Bit Set f

Syntax: [*label*] BSF f,b

Operands: $0 \leq f \leq 127$
 $0 \leq b \leq 7$

Operation: $1 \rightarrow (f)$

Status Affected: None

Encoding:

| | | | |
|----|------|------|------|
| 01 | 01bb | bfff | ffff |
|----|------|------|------|

Description: Bit 'b' in register 'f' is set.

Words: 1

Cycles: 1

Example

```
BSF    REG1, 7

Before Instruction
    REG1 = 0x0A
After Instruction
    REG1 = 0x8A
```

BTFSC Bit Test f, Skip if Clear

Syntax: [*label*] BTFSC f,b

Operands: $0 \leq f \leq 127$
 $0 \leq b \leq 7$

Operation: skip if $(f) = 0$

Status Affected: None

Encoding:

| | | | |
|----|------|------|------|
| 01 | 10bb | bfff | ffff |
|----|------|------|------|

Description: If bit 'b' in register 'f' is '0' then the next instruction is skipped.
 If bit 'b' is '0' then the next instruction fetched during the current instruction execution is discarded, and a NOP is executed instead, making this a two-cycle instruction.

Words: 1

Cycles: 1(2)

Example

```
HERE    BTFSC    REG1
FALSE   GOTO     PROCESS_CODE
TRUE    •
        •
        •
```

Before Instruction
 PC = address HERE

After Instruction
 if $REG<1> = 0$,
 PC = address TRUE
 if $REG<1> \neq 0$,
 PC = address FALSE

| DECFSZ | | Decrement f, Skip if 0 | | | |
|------------------|---|------------------------|----------|----------|--|
| Syntax: | [<i>label</i>] DECFSZ f,d | | | | |
| Operands: | $0 \leq f \leq 127$ $d \in [0,1]$ | | | | |
| Operation: | $(f) - 1 \rightarrow (dest);$ skip if result = 0 | | | | |
| Status Affected: | None | | | | |
| Encoding: | 00 | 1011 | dfff | ffff | |
| Description: | <p>The contents of register 'f' are decremented. If 'd' is 0 the result is placed in the W register. If 'd' is 1 the result is placed back in register 'f'.</p> <p>If the result is 0, the next instruction, which is already fetched, is discarded. A NOP is executed instead making it a two-cycle instruction.</p> | | | | |
| Words: | 1 | | | | |
| Cycles: | 1(2) | | | | |
| Example | HERE | DECFSZ | REG1, 1 | | |
| | | GOTO | LOOP | | |
| | CONTINUE | • | | | |
| | | • | | | |
| | | • | | | |
| | Before Instruction | | | | |
| | PC | = | address | HERE | |
| | After Instruction | | | | |
| | REG1 | = | REG1 - 1 | | |
| | if REG1 | = | 0, | | |
| | PC | = | address | CONTINUE | |
| | if REG1 | ≠ | 0, | | |
| | PC | = | address | HERE+1 | |

| GOTO | Unconditional Branch | | | |
|------------------|---|------|------|------|
| Syntax: | [<i>label</i>] GOTO k | | | |
| Operands: | 0 ≤ k ≤ 2047 | | | |
| Operation: | k → PC<10:0> PCLATH<4:3> → PC<12:11> | | | |
| Status Affected: | None | | | |
| Encoding: | 10 | 1kkk | kkkk | kkkk |
| Description: | GOTO is an unconditional branch. The eleven bit immediate value is loaded into PC bits <10:0>. The upper bits of PC are loaded from PCLATH<4:3>. GOTO is a two-cycle instruction. | | | |
| Words: | 1 | | | |
| Cycles: | 2 | | | |
| Example | GOTO THERE | | | |
| | After Instruction | | | |
| | PC = Address THERE | | | |


RETLW Return with Literal in W

| | | | | | |
|------------------|--|------|------|------|------|
| Syntax: | [<i>label</i>] RETLW k | | | | |
| Operands: | 0 ≤ k ≤ 255 | | | | |
| Operation: | k → (W); TOS → PC | | | | |
| Status Affected: | None | | | | |
| Encoding: | <table><tr><td>11</td><td>01xx</td><td>kkkk</td><td>kkkk</td></tr></table> | 11 | 01xx | kkkk | kkkk |
| 11 | 01xx | kkkk | kkkk | | |
| Description: | The W register is loaded with the eight bit literal 'k'. The program counter is loaded from the top of the stack (the return address). This is a two-cycle instruction. | | | | |
| Words: | 1 | | | | |
| Cycles: | 2 | | | | |
| Example | <pre>CALL TABLE;W contains table ;offset value • ;W now has table value TABLE • • ADDWF PC ;W = offset RETLW k1 ;Begin table RETLW k2 ; • • • RETLW kn ; End of table Before Instruction W = 0x07 After Instruction W = value of k8</pre> | | | | |

RETURN Return from Subroutine

| | | | | | |
|------------------|--|------|------|------|------|
| Syntax: | [<i>label</i>] RETURN | | | | |
| Operands: | None | | | | |
| Operation: | TOS → PC | | | | |
| Status Affected: | None | | | | |
| Encoding: | <table border="1"><tr><td>00</td><td>0000</td><td>0000</td><td>1000</td></tr></table> | 00 | 0000 | 0000 | 1000 |
| 00 | 0000 | 0000 | 1000 | | |
| Description: | Return from subroutine. The stack is POPed and the top of the stack (TOS) is loaded into the program counter. This is a two-cycle instruction. | | | | |
| Words: | 1 | | | | |
| Cycles: | 2 | | | | |
| Example | RETURN After Interrupt PC = TOS | | | | |

RLF Rotate Left f through Carry

| | | | | | |
|------------------|--|------|------|------|------|
| Syntax: | [<i>label</i>] RLF <i>f</i> , <i>d</i> | | | | |
| Operands: | $0 \leq f \leq 127$ $d \in [0,1]$ | | | | |
| Operation: | See description below | | | | |
| Status Affected: | C | | | | |
| Encoding: | <table><tr><td>00</td><td>1101</td><td>dfff</td><td>ffff</td></tr></table> | 00 | 1101 | dfff | ffff |
| 00 | 1101 | dfff | ffff | | |
| Description: | <p>The contents of register 'f' are rotated one bit to the left through the Carry Flag. If 'd' is 0 the result is placed in the W register. If 'd' is 1 the result is stored back in register 'f'.</p>  | | | | |
| Words: | 1 | | | | |
| Cycles: | 1 | | | | |
| Example | <pre>RLF REG1, 0</pre> <p>Before Instruction</p> <pre> REG1 = 1110 0110 C = 0</pre> <p>After Instruction</p> <pre> REG1 = 1110 0110 W = 1100 1100 C = 1</pre> | | | | |

16.0 DEVELOPMENT SUPPORT

The PICmicro® microcontrollers are supported with a full range of hardware and software development tools:

- Integrated Development Environment
 - MPLAB® IDE Software
- Assemblers/Compilers/Linkers
 - MPASM™ Assembler
 - MPLAB C17 and MPLAB C18 C Compilers
 - MPLINK™ Object Linker/
MPLIB™ Object Librarian
 - MPLAB C30 C Compiler
 - MPLAB ASM30 Assembler/Linker/Library
- Simulators
 - MPLAB SIM Software Simulator
 - MPLAB dsPIC30 Software Simulator
- Emulators
 - MPLAB ICE 2000 In-Circuit Emulator
 - MPLAB ICE 4000 In-Circuit Emulator
- In-Circuit Debugger
 - MPLAB ICD 2
- Device Programmers
 - PRO MATE® II Universal Device Programmer
 - PICSTART® Plus Development Programmer
- Low Cost Demonstration Boards
 - PICDEM™ 1 Demonstration Board
 - PICDEM.net™ Demonstration Board
 - PICDEM 2 Plus Demonstration Board
 - PICDEM 3 Demonstration Board
 - PICDEM 17 Demonstration Board
 - PICDEM 18R Demonstration Board
 - PICDEM LIN Demonstration Board
 - PICDEM USB Demonstration Board
- Evaluation Kits
 - KEELOQ®
 - PICDEM MSC
 - microID®
 - CAN
 - PowerSmart®
 - Analog

16.1 MPLAB Integrated Development Environment Software

The MPLAB IDE software brings an ease of software development previously unseen in the 8/16-bit microcontroller market. The MPLAB IDE is a Windows® based application that contains:

- An interface to debugging tools
 - simulator
 - programmer (sold separately)
 - emulator (sold separately)
 - in-circuit debugger (sold separately)
- A full-featured editor with color coded context
- A multiple project manager
- Customizable data windows with direct edit of contents
- High level source code debugging
- Mouse over variable inspection
- Extensive on-line help

The MPLAB IDE allows you to:

- Edit your source files (either assembly or C)
- One touch assemble (or compile) and download to PICmicro emulator and simulator tools (automatically updates all project information)
- Debug using:
 - source files (assembly or C)
 - absolute listing file (mixed assembly and C)
 - machine code

MPLAB IDE supports multiple debugging tools in a single development paradigm, from the cost effective simulators, through low cost in-circuit debuggers, to full-featured emulators. This eliminates the learning curve when upgrading to tools with increasing flexibility and power.

16.2 MPASM Assembler

The MPASM assembler is a full-featured, universal macro assembler for all PICmicro MCUs.

The MPASM assembler generates relocatable object files for the MPLINK object linker, Intel® standard HEX files, MAP files to detail memory usage and symbol reference, absolute LST files that contains source lines and generated machine code and COFF files for debugging.

The MPASM assembler features include:

- Integration into MPLAB IDE projects
- User defined macros to streamline assembly code
- Conditional assembly for multi-purpose source files
- Directives that allow complete control over the assembly process

16.19 PICDEM 18R PIC18C601/801 Demonstration Board

The PICDEM 18R demonstration board serves to assist development of the PIC18C601/801 family of Microchip microcontrollers. It provides hardware implementation of both 8-bit Multiplexed/De-multiplexed and 16-bit Memory modes. The board includes 2 Mb external FLASH memory and 128 Kb SRAM memory, as well as serial EEPROM, allowing access to the wide range of memory types supported by the PIC18C601/801.

16.20 PICDEM LIN PIC16C43X Demonstration Board

The powerful LIN hardware and software kit includes a series of boards and three PICmicro microcontrollers. The small footprint PIC16C432 and PIC16C433 are used as slaves in the LIN communication and feature on-board LIN transceivers. A PIC16F874 FLASH microcontroller serves as the master. All three microcontrollers are programmed with firmware to provide LIN bus communication.

16.21 PICDEM USB PIC16C7X5 Demonstration Board

The PICDEM USB Demonstration Board shows off the capabilities of the PIC16C745 and PIC16C765 USB microcontrollers. This board provides the basis for future USB products.

16.22 Evaluation and Programming Tools

In addition to the PICDEM series of circuits, Microchip has a line of evaluation kits and demonstration software for these products.

- KEELOQ evaluation and programming tools for Microchip's HCS Secure Data Products
- CAN developers kit for automotive network applications
- Analog design boards and filter design software
- PowerSmart battery charging evaluation/calibration kits
- IrDA® development kit
- microID development and RFLab™ development software
- SEEVAL® designer kit for memory evaluation and endurance calculations
- PICDEM MSC demo boards for Switching mode power supply, high power IR driver, delta sigma ADC, and flow rate sensor

Check the Microchip web page and the latest Product Line Card for the complete list of demonstration and evaluation kits.

FIGURE 17-3: PIC16LF62X VOLTAGE-FREQUENCY GRAPH, $0^{\circ}\text{C} \leq T_A \leq +70^{\circ}\text{C}$

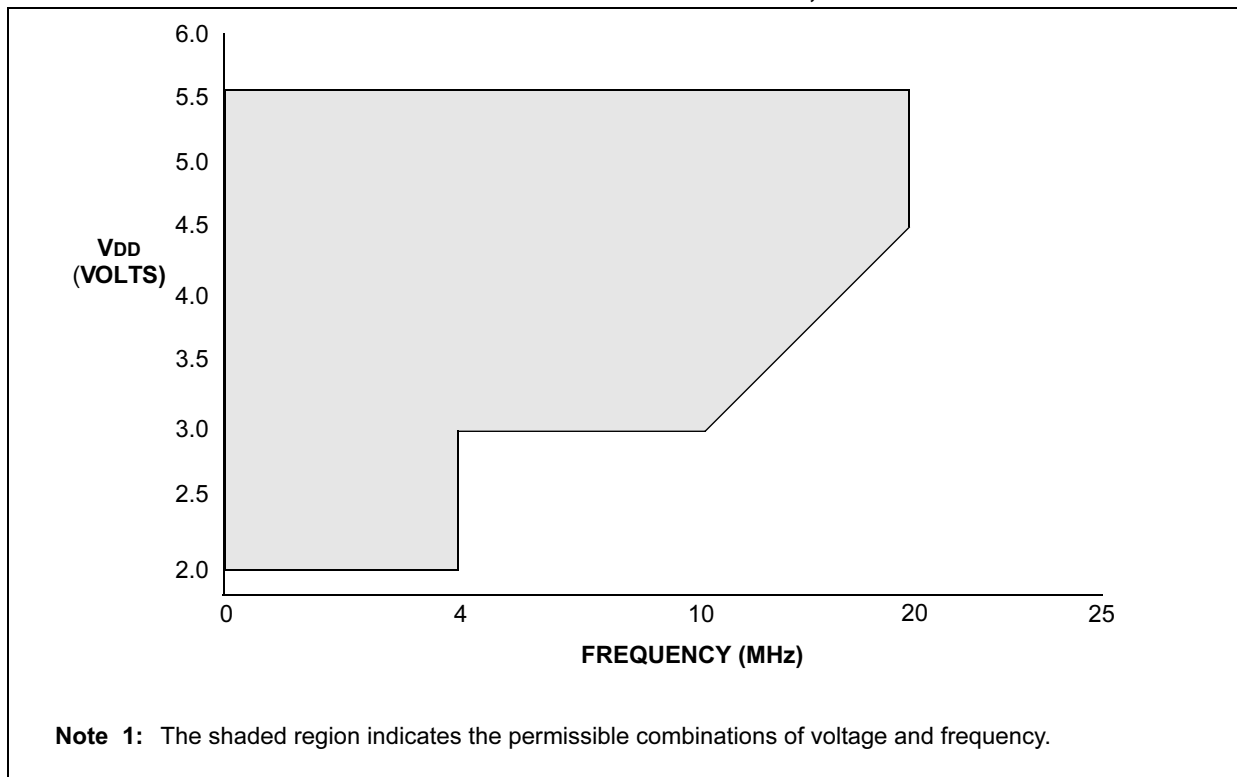


FIGURE 17-4: PIC16LF62X VOLTAGE-FREQUENCY GRAPH, $-40^{\circ}\text{C} \leq T_A < 0^{\circ}\text{C}$, $+70^{\circ}\text{C} < T_A \leq 85^{\circ}\text{C}$

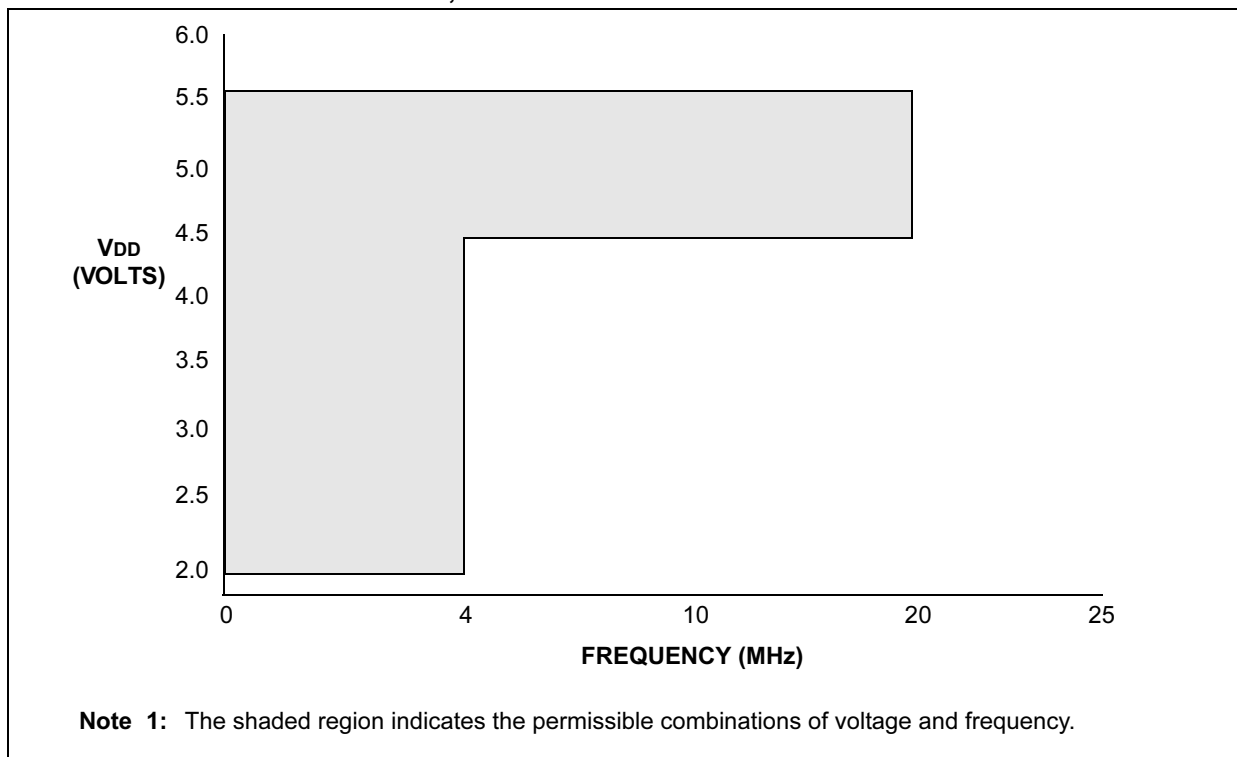


TABLE 17-1: COMPARATOR SPECIFICATIONS

| Operating Conditions: 3.0V < VDD < 5.5V, -40°C < TA < +125°C, unless otherwise stated. | | | | | | | |
|--|---|--------|-----|------|------------|----------|-------------------|
| Param No. | Characteristics | Sym | Min | Typ | Max | Units | Comments |
| D300 | Input offset voltage | VIOFF | — | ±5.0 | ±10 | mV | |
| D301* | Input Common mode voltage | VICM | 0 | — | VDD - 1.5 | V | |
| D302* | Common Mode Rejection Ratio | CMRR | 55 | — | — | db | |
| 300* 300A | Response Time ⁽¹⁾ | TRESP | — | 150 | 400 600 | ns ns | 16F62X 16LF62X |
| 301 | Comparator Mode Change to Output Valid* | TMC2OV | — | — | 10 | µs | |

* These parameters are characterized but not tested.

Note 1: Response time measured with one comparator input at (VDD - 1.5)/2 while the other input transitions from VSS to VDD.

TABLE 17-2: VOLTAGE REFERENCE SPECIFICATIONS

| Operating Conditions: 3.0V < VDD < 5.5V, -40°C < TA < +125°C, unless otherwise stated. | | | | | | | |
|--|------------------------------|------|--------|--------|------------|------------|---|
| Spec No. | Characteristics | Sym | Min | Typ | Max | Units | Comments |
| D310 | Resolution | VRES | VDD/24 | — | VDD/32 | LSb | |
| D311 | Absolute Accuracy | VRaa | — — | — — | 1/4 1/2 | LSb LSb | Low Range (VRR = 1) High Range (VRR = 0) |
| D312* | Unit Resistor Value (R) | VRur | — | 2k | — | Ω | |
| 310* | Settling Time ⁽¹⁾ | Tset | — | — | 10 | µs | |

* These parameters are characterized but not tested.

Note 1: Settling time measured while VRR = 1 and VR<3:0> transitions from 0000 to 1111.

PIC16F62X

TABLE 17-4: EXTERNAL CLOCK TIMING REQUIREMENTS

| Param No. | Sym | Characteristic | Min | Typ† | Max | Units | Conditions |
|-----------|---------------|--|--------|------|--------|-------|-------------------------------------|
| | Fosc | External CLKIN Frequency ⁽¹⁾ | DC | — | 4 | MHz | XT and ER Osc mode, VDD = 5.0V |
| | | | DC | — | 20 | MHz | HS Osc mode |
| | | | DC | — | 200 | kHz | LP Osc mode |
| | | Oscillator Frequency ⁽¹⁾ | | — | 4 | MHz | ER Osc mode, VDD = 5.0V |
| | | | 0.1 | — | 4 | MHz | XT Osc mode |
| | | | 1 | — | 20 | MHz | HS Osc mode |
| | | | | — | 200 | kHz | LP Osc mode |
| | | | 3.65 | 4 | 4.28 | MHz | INTRC mode (fast), VDD = 5.0V |
| | | | | 37 | | kHz | INTRC mode (slow) |
| 4 | INTRC | Internal Calibrated RC | 3.65 | 4.00 | 4.28 | MHz | VDD = 5.0V |
| 5 | ER | External Biased ER Frequency | 10 kHz | | 8 MHz | | VDD = 5.0V |
| 1 | Tosc | External CLKIN Period ⁽¹⁾ | 250 | — | — | ns | XT and ER Osc mode |
| | | | 50 | — | — | ns | HS Osc mode |
| | | | 5 | — | — | μs | LP Osc mode |
| | | Oscillator Period ⁽¹⁾ | 250 | — | — | ns | ER Osc mode |
| | | | 250 | — | 10,000 | ns | XT Osc mode |
| | | | 50 | — | 1,000 | ns | HS Osc mode |
| | | | 5 | | | μs | LP Osc mode |
| | | | | 250 | | ns | INTRC mode (fast) |
| | | | | 27 | | μs | INTRC mode (slow) |
| 2 | Tcy | Instruction Cycle Time | 1.0 | Tcy | DC | ns | Tcy = 4/Fosc |
| 3 | TosL, TosH | External CLKIN (OSC1) High External CLKIN Low | 100 * | — | — | ns | XT oscillator, TOSC L/H duty cycle* |

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: Instruction cycle period (Tcy) equals four times the input oscillator time-based period. All specified values are based on characterization data for that particular oscillator type under standard operating conditions with the device executing code. Exceeding these specified limits may result in an unstable oscillator operation and/or higher than expected current consumption. All devices are tested to operate at "Min." values with an external clock applied to the OSC1 pin. When an external clock input is used, the "Max" cycle time limit is "DC" (no clock) for all devices.

Note: The graphs and tables provided in this section are for design guidance and are not tested.

FIGURE 18-10: I_{PD} vs V_{DD} SLEEP MODE, ALL PERIPHERALS DISABLED

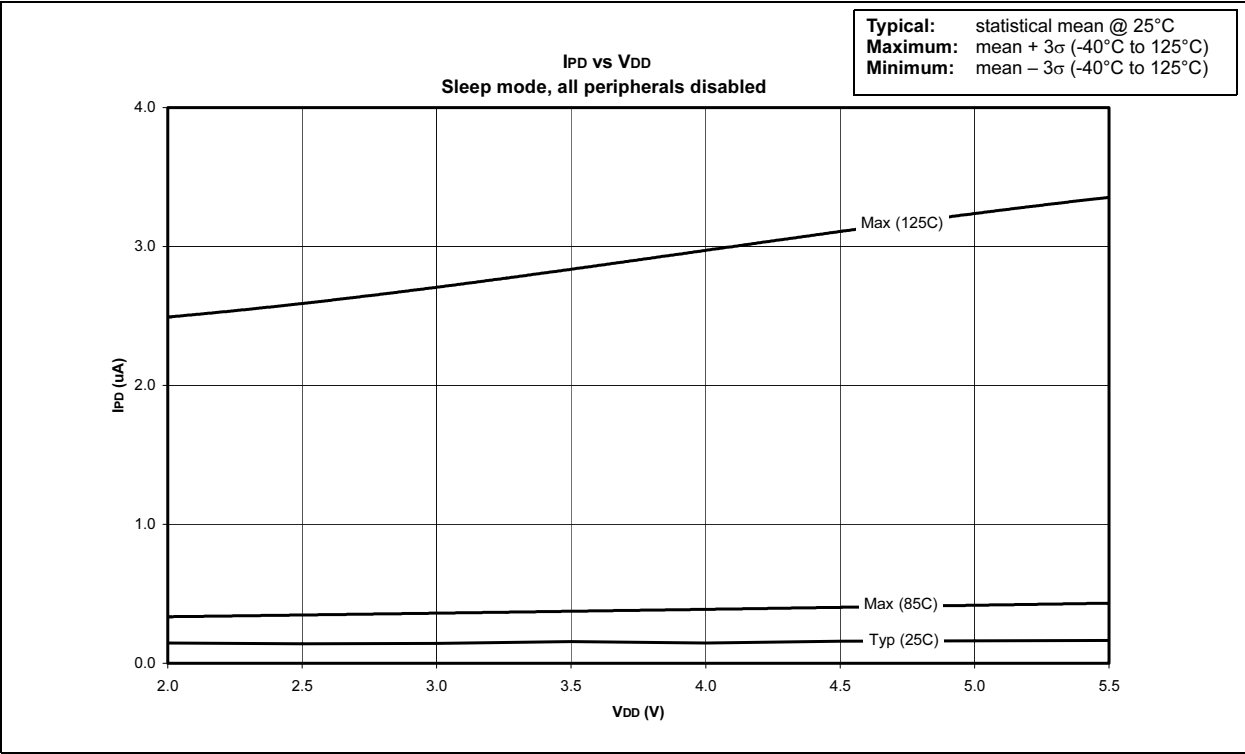


FIGURE 18-11: ΔI_{BOD} vs V_{OH} OVER TEMPERATURE (-40 to 125°C)

