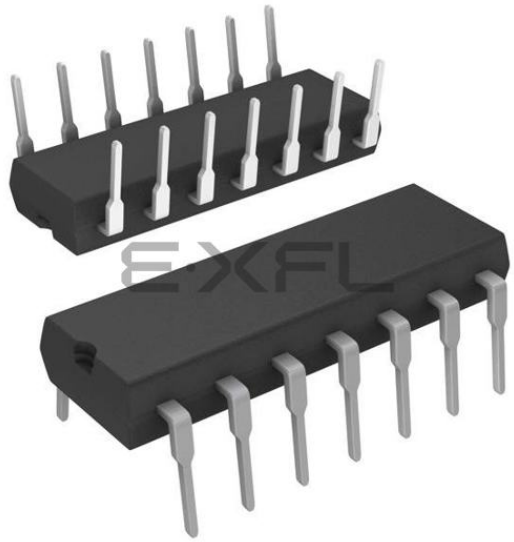

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What is "Embedded - Microcontrollers"?

"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	32MHz
Connectivity	I²C, LINbus, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	12
Program Memory Size	3.5KB (2K x 14)
Program Memory Type	FLASH
EEPROM Size	224 x 8
RAM Size	256 x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.6V
Data Converters	A/D 11x10b; D/A 1x5b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Through Hole
Package / Case	14-DIP (0.300", 7.62mm)
Supplier Device Package	14-PDIP
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic16lf15323-i-p

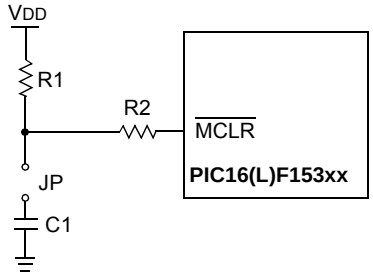
2.3 Master Clear ($\overline{MCLR}$) Pin

The $\overline{MCLR}$ pin provides two specific device functions: Device Reset, and Device Programming and Debugging. If programming and debugging are not required in the end application, a direct connection to VDD may be all that is required. The addition of other components, to help increase the application's resistance to spurious Resets from voltage sags, may be beneficial. A typical configuration is shown in Figure 2-1. Other circuit designs may be implemented, depending on the application's requirements.

During programming and debugging, the resistance and capacitance that can be added to the pin must be considered. Device programmers and debuggers drive the $\overline{MCLR}$ pin. Consequently, specific voltage levels (VIH and VIL) and fast signal transitions must not be adversely affected. Therefore, specific values of R1 and C1 will need to be adjusted based on the application and PCB requirements. For example, it is recommended that the capacitor, C1, be isolated from the $\overline{MCLR}$ pin during programming and debugging operations by using a jumper (Figure 2-2). The jumper is replaced for normal run-time operations.

Any components associated with the $\overline{MCLR}$ pin should be placed within 0.25 inch (6 mm) of the pin.

FIGURE 2-2: EXAMPLE OF $\overline{MCLR}$ PIN CONNECTIONS



Note 1: R1 ≤ 10 kΩ is recommended. A suggested starting value is 10 kΩ. Ensure that the $\overline{MCLR}$ pin VIH and VIL specifications are met.

2: R2 ≤ 470Ω will limit any current flowing into $\overline{MCLR}$ from the external capacitor, C1, in the event of $\overline{MCLR}$ pin breakdown, due to Electrostatic Discharge (ESD) or Electrical Overstress (EOS). Ensure that the $\overline{MCLR}$ pin VIH and VIL specifications are met.

2.4 ICSP™ Pins

The ICSPCLK and ICSPDAT pins are used for In-Circuit Serial Programming™ (ICSP™) and debugging purposes. It is recommended to keep the trace length between the ICSP connector and the ICSP pins on the device as short as possible. If the ICSP connector is expected to experience an ESD event, a series resistor is recommended, with the value in the range of a few tens of ohms, not to exceed 100Ω.

Pull-up resistors, series diodes and capacitors on the ICSPCLK and ICSPDAT pins are not recommended as they will interfere with the programmer/debugger communications to the device. If such discrete components are an application requirement, they should be removed from the circuit during programming and debugging. Alternatively, refer to the AC/DC characteristics and timing requirements information in the respective device Flash programming specification for information on capacitive loading limits, and pin input voltage high (VIH) and input low (VIL) requirements.

For device emulation, ensure that the "Communication Channel Select" (i.e., ICSPCLK/ICSPDAT pins), programmed into the device, matches the physical connections for the ICSP to the Microchip debugger/emulator tool.

For more information on available Microchip development tools connection requirements, refer to **Section 39.0 "Development Support"**.

REGISTER 5-3: CONFIGURATION WORD 3: WINDOWED WATCHDOG

R/P-1	R/P-1	R/P-1	R/P-1	R/P-1	R/P-1
WDTCCS2	WDTCCS1	WDTCCS0	WDTCWS2	WDTCWS1	WDTCWS0
bit 13					bit 8

U-1	R/P-1	R/P-1	R/P-1	R/P-1	R/P-1	R/P-1	R/P-1
—	WDTE1	WDTE0	WDTCPS4	WDTCPS3	WDTCPS2	WDTCPS1	WDTCPS0
bit 7							bit 0

Legend:			
R = Readable bit	P = Programmable bit	x = Bit is unknown	U = Unimplemented bit, read as '1'
'0' = Bit is cleared	'1' = Bit is set	W = Writable bit	n = Value when blank or after Bulk Erase

bit 13-11 **WDTCCS<2:0>**: WDT Input Clock Selector bits
- 111 = Software Control
- 110 = Reserved
- .
- .
- .
- .
- 010 = Reserved
- 001 = WDT reference clock is the 31.0 kHz LFINTOSC
- 000 = WDT reference clock is the 31.25 kHz HFINTOSC (MFINTOSC) output

bit 10-8 **WDTCWS<2:0>**: WDT Window Select bits

WDTCWS	WDTWS at POR			Software control of WDTWS?	Keyed access required?
	Value	Window delay Percent of time	Window opening Percent of time		
111	111	n/a	100	Yes	No
110	111	n/a	100	No	Yes
101	101	25	75		
100	100	37.5	62.5		
011	011	50	50		
010	010	62.5	37.5		
001	001	75	25		
000	000	87.5	12.5		

bit 7 **Unimplemented**: Read as '1'

bit 6-5 **WDTE<1:0>**: WDT Operating mode:
- 11 =WDT enabled regardless of Sleep; SWDTEN is ignored
- 10 =WDT enabled while Sleep = 0, suspended when Sleep = 1; SWDTEN ignored
- 01 =WDT enabled/disabled by SWDTEN bit in WDTCON0
- 00 =WDT disabled, SWDTEN is ignored

8.2.3 BOR CONTROLLED BY SOFTWARE

When the BOREN bits of Configuration Words are programmed to '01', the BOR is controlled by the SBOREN bit of the BORCON register. The device start-up is not delayed by the BOR ready condition or the VDD level.

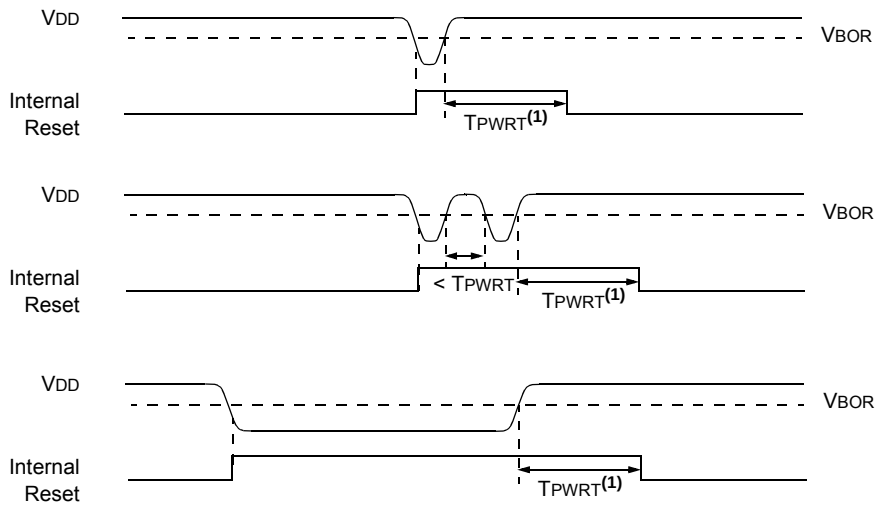
BOR protection begins as soon as the BOR circuit is ready. The status of the BOR circuit is reflected in the BORRDY bit of the BORCON register.

BOR protection is unchanged by Sleep.

8.2.4 BOR IS ALWAYS OFF

When the BOREN bits of the Configuration Words are programmed to '00', the BOR is off at all times. The device start-up is not delayed by the BOR ready condition or the VDD level.

FIGURE 8-2: BROWN-OUT SITUATIONS



Note 1: TPWRT delay only if $\overline{\text{PWRTE}}$ bit is programmed to '0'.

REGISTER 9-4: OSCSTAT: OSCILLATOR STATUS REGISTER 1

R-q/q	R-q/q	R-q/q	R-q/q	U-0	R-q/q	U-0	R-q/q
EXTOR	HFOR	MFOR	LFOR	—	ADOR	—	PLLR
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set	‘0’ = Bit is cleared	

bit 7 **EXTOR:** EXTOSC (external) Oscillator Ready bit
1 = The oscillator is ready to be used
0 = The oscillator is not enabled, or is not yet ready to be used.

bit 6 **HFOR:** HFINTOSC Oscillator Ready bit
1 = The oscillator is ready to be used
0 = The oscillator is not enabled, or is not yet ready to be used.

bit 5 **MFOR:** MFINTOSC Oscillator Ready bit
1 = The oscillator is ready to be used
0 = The oscillator is not enabled, or is not yet ready to be used.

bit 4 **LFOR:** LFINTOSC Oscillator Ready bit
1 = The oscillator is ready to be used
0 = The oscillator is not enabled, or is not yet ready to be used.

bit 3 **Unimplemented:** Read as ‘0’

bit 2 **ADOR:** CRC Oscillator Ready bit
1 = The oscillator is ready to be used
0 = The oscillator is not enabled, or is not yet ready to be used.

bit 1 **Unimplemented:** Read as ‘0’

bit 0 **PLLR:** PLL is Ready bit
1 = The PLL is ready to be used
0 = The PLL is not enabled, the required input source is not ready, or the PLL is not locked.

11.0 POWER-SAVING OPERATION MODES

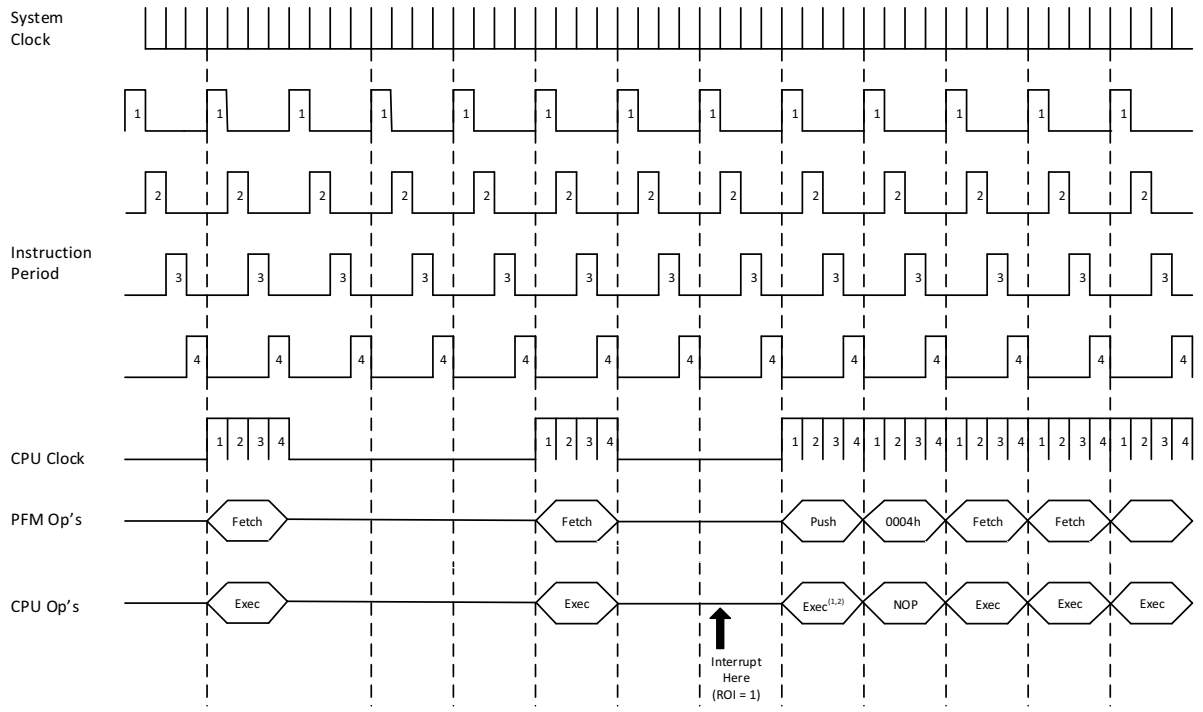
The purpose of the Power-Down modes is to reduce power consumption. There are three Power-Down modes: DOZE mode, IDLE mode, and SLEEP mode.

11.1 DOZE Mode

DOZE mode allows for power saving by reducing CPU operation and program memory (PFM) access, without affecting peripheral operation. DOZE mode differs from Sleep mode because the system oscillators continue to operate, while only the CPU and PFM are affected. The reduced execution saves power by eliminating unnecessary operations within the CPU and memory.

When the Doze Enable (DOZEN) bit is set (DOZEN = 1), the CPU executes only one instruction cycle out of every N cycles as defined by the DOZE<2:0> bits of the CPUDOZE register. For example, if DOZE<2:0> = 100, the instruction cycle ratio is 1:32. The CPU and memory execute for one instruction cycle and then lay idle for 31 instruction cycles. During the unused cycles, the peripherals continue to operate at the system clock speed.

FIGURE 11-1: DOZE MODE OPERATION EXAMPLE



Note 1: Multi-cycle instructions are executed to completion before fetching 0004h.
2: If the pre-fetched instruction clears GIE, the ISR will not occur, but DOZEN is still cleared and the CPU will resume execution at full speed.

11.1.1 DOZE OPERATION

The Doze operation is illustrated in Figure 11-1. For this example:

- Doze enable (DOZEN) bit set (DOZEN = 1)
- DOZE<2:0> = 001 (1:4) ratio
- Recover-on-Interrupt (ROI) bit set (ROI = 1)

As with normal operation, the PFM fetches for the next instruction cycle. The Q-clocks to the peripherals continue throughout.

11.4 Register Definitions: Voltage Regulator and DOZE Control

REGISTER 11-1: VREGCON: VOLTAGE REGULATOR CONTROL REGISTER(1)

U-0	U-0	U-0	U-0	U-0	U-0	R/W-0/0	U-1
—	—	—	—	—	—	VREGPM	—
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-2 **Unimplemented:** Read as '0'

bit 1 **VREGPM**: Voltage Regulator Power Mode Selection bit
- 1 = Low-Power Sleep mode enabled in Sleep(2)
  Draws lowest current in Sleep, slower wake-up
- 0 = Normal Power mode enabled in Sleep(2)
  Draws higher current in Sleep, faster wake-up

bit 0 **Unimplemented:** Read as '1'. Maintain this bit set

Note 1: PIC16F15313/23 only.

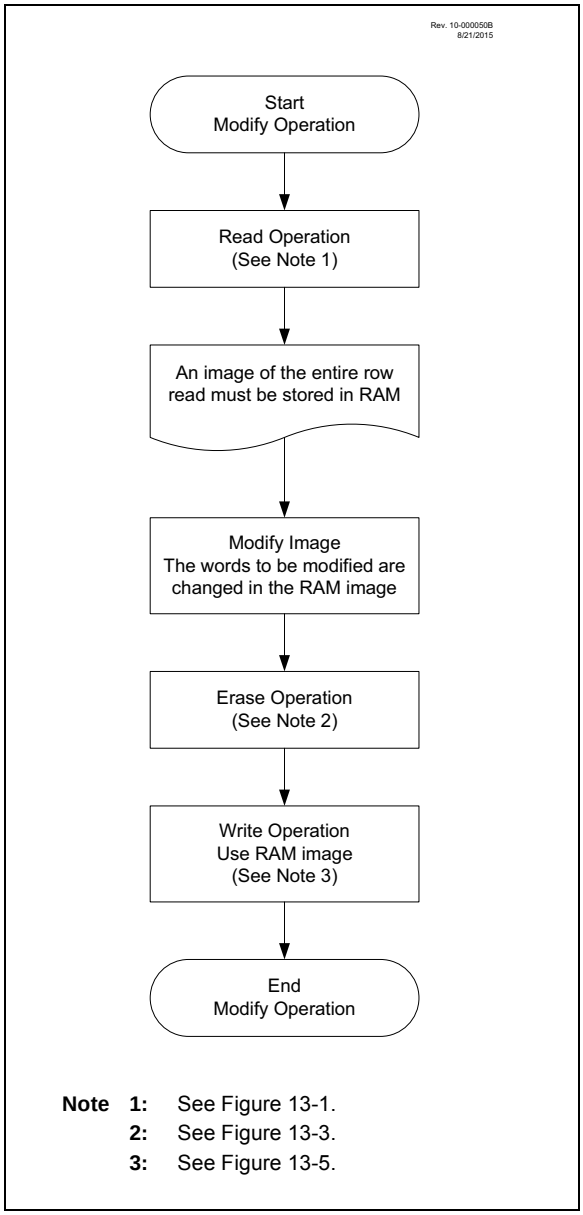
2: See **Section 37.0 "Electrical Specifications"**.

13.3.5 MODIFYING FLASH PROGRAM MEMORY

When modifying existing data in a program memory row, and data within that row must be preserved, it must first be read and saved in a RAM image. Program memory is modified using the following steps:

1. Load the starting address of the row to be modified.
2. Read the existing data from the row into a RAM image.
3. Modify the RAM image to contain the new data to be written into program memory.
4. Load the starting address of the row to be rewritten.
5. Erase the program memory row.
6. Load the write latches with data from the RAM image.
7. Initiate a programming operation.

FIGURE 13-6: FLASH PROGRAM MEMORY MODIFY FLOWCHART



Note 1: See Figure 13-1.
2: See Figure 13-3.
3: See Figure 13-5.

TABLE 14-2: SUMMARY OF REGISTERS ASSOCIATED WITH PORTA

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
PORTA	—	—	RA5	RA4	RA3	RA2	RA1	RA0	175
TRISA	—	—	TRISA5	TRISA4	—	TRISA2	TRISA1	TRISA0	175
LATA	—	—	LATA5	LATA4	—	LATA2	LATA1	LATA0	176
ANSELA	—	—	ANSA5	ANSA4	—	ANSA2	ANSA1	ANSA0	176
WPUA	—	—	WPUA5	WPUA4	WPUA3	WPUA2	WPUA1	WPUA0	177
ODCONA	—	—	ODCA5	ODCA4	—	ODCA2	ODCA1	ODCA0	177
SLRCONA	—	—	SLRA5	SLRA4	—	SLRA2	SLRA1	SLRA0	178
INLVLA	—	—	INLVLA5	INLVLA4	INLVLA3	INLVLA2	INLVLA1	INLVLA0	178

Legend: x = unknown, u = unchanged, – = unimplemented locations read as '0'. Shaded cells are not used by PORTA.

14.5 Register Definitions: PORTC

REGISTER 14-9: PORTC: PORTC REGISTER

U-0	U-0	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
—	—	RC5	RC4	RC3	RC2	RC1	RC0
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set	‘0’ = Bit is cleared	

bit 7-6 **Unimplemented**: Read as ‘0’

bit 5-0 **RC<5:0>**: PORTC General Purpose I/O Pin bits(1)
1 = Port pin is ≥ VIH
0 = Port pin is ≤ VIL

Note 1: Writes to PORTC are actually written to corresponding LATC register. The actual I/O pin values are read from the PORTC register.

REGISTER 14-10: TRISC: PORTC TRI-STATE REGISTER

U-0	U-0	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1
—	—	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set	‘0’ = Bit is cleared	

bit 7-6 **Unimplemented**: Read as ‘0’

bit 5-0 **TRISC<5:0>:** PORTC Tri-State Control bits
1 = PORTC pin configured as an input (tri-stated)
0 = PORTC pin configured as an output

REGISTER 14-11: LATC: PORTC DATA LATCH REGISTER

U-0	U-0	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
—	—	LATC5	LATC4	LATC3	LATC2	LATC1	LATC0
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set	‘0’ = Bit is cleared	

bit 7-6 **Unimplemented**: Read as ‘0’

bit 5-0 **LATC<5:0>**: PORTC Output Latch Value bits((1)

Note 1: Writes to PORTC are actually written to corresponding LATC register. Reads from PORTC register returns actual I/O pin values.

TABLE 15-1: PPS INPUT SIGNAL ROUTING OPTIONS (PIC16(L)F15313)

INPUT SIGNAL NAME	Input Register Name	Default Location at POR	Reset Value (xxxPPS<4:0>)	Remappable to Pins of PORTx
				PIC16(L)F15313
				PORTA
INT	INTPPS	RA2	00010	•
T0CKI	T0CKIPPS	RA2	00010	•
T1CKI	T1CKIPSS	RA5	00101	•
T1G	T1GPPS	RA4	00100	•
T2IN	T2INPPS	RA5	00101	•
CCP1	CCP1PPS	RA5	00101	•
CCP2	CCP2PPS	RA5	00101	•
CWG1IN	CWG1INPPS	RA2	00010	•
CLCIN0	CLCIN0PPS	RA3	00011	•
CLCIN1	CLCIN1PPS	RA5	00101	•
CLCIN2	CLCIN2PPS	RA1	00001	•
CLCIN3	CLCIN3PPS	RA0	00000	•
ADACT	ADACTPPS	RA5	00101	•
SCK1/SCL1	SSP1CLKPPS	RA1	00001	•
SDI1/SDA1	SSP1DATPPS	RA2	00010	•
SS1	SSP1SS1PPS	RA3	00011	•
RX1/DT1	RX1PPS	RA1	00001	•
CK1	TX1PPS	RA0	00000	•

20.3 ADC Acquisition Requirements

For the ADC to meet its specified accuracy, the charge holding capacitor (CHOLD) must be allowed to fully charge to the input channel voltage level. The Analog Input model is shown in Figure 20-4. The source impedance (RS) and the internal sampling switch (RSS) impedance directly affect the time required to charge the capacitor CHOLD. The sampling switch (RSS) impedance varies over the device voltage (VDD), refer to Figure 20-4. **The maximum recommended impedance for analog sources is 10 kΩ.** As the source impedance is decreased, the acquisition time may be decreased. After the analog input channel is selected (or changed), an ADC acquisition must be done before the conversion can be started. To calculate the minimum acquisition time, Equation 20-1 may be used. This equation assumes that 1/2 LSb error is used (1024 steps for the ADC). The 1/2 LSb error is the maximum error allowed for the ADC to meet its specified resolution.

EQUATION 20-1: ACQUISITION TIME EXAMPLE

Assumptions: $Temperature = 50°C$ *and external impedance of* $10k\Omega$ *5.0V* V_{DD}

$$T_{ACQ} = \text{Amplifier Settling Time} + \text{Hold Capacitor Charging Time} + \text{Temperature Coefficient}$$
$$= T_{AMP} + T_C + T_{COFF}$$
$$= 2\mu s + T_C + [(\text{Temperature} - 25°C)(0.05\mu s/°C)]$$

The value for T_C *can be approximated with the following equations:*

$$V_{APPLIED}\left(1 - \frac{1}{(2^{n+1}) - 1}\right) = V_{CHOLD} \quad ;[1]\ V_{CHOLD}\ \text{charged to within 1/2 lsb}$$

$$V_{APPLIED}\left(1 - e^{\frac{-T_C}{RC}}\right) = V_{CHOLD} \quad ;[2]\ V_{CHOLD}\ \text{charge response to}\ V_{APPLIED}$$

$$V_{APPLIED}\left(1 - e^{\frac{-T_C}{RC}}\right) = V_{APPLIED}\left(1 - \frac{1}{(2^{n+1}) - 1}\right) \quad ;\text{combining [1] and [2]}$$

Note: Where n = number of bits of the ADC.

Solving for T_C*:*

$$T_C = -C_{HOLD}(R_{IC} + R_{SS} + R_S)\ ln(1/2047)$$
$$= -10pF(1k\Omega + 7k\Omega + 10k\Omega)\ ln(0.0004885)$$
$$= 1.37\mu s$$

Therefore:

$$T_{ACQ} = 2\mu s + 1.37 + [(50°C - 25°C)(0.05\mu s/°C)]$$
$$= 4.62\mu s$$

Note 1: The VAPPLIED has no effect on the equation, since it cancels itself out.

2: The charge holding capacitor (CHOLD) is not discharged after each conversion.

3: The maximum recommended impedance for analog sources is 10 kΩ. This is required to meet the pin leakage specification.

24.0 ZERO-CROSS DETECTION (ZCD) MODULE

The ZCD module detects when an A/C signal crosses through the ground potential. The actual zero crossing threshold is the zero crossing reference voltage, VCPINV, which is typically 0.75V above ground.

The connection to the signal to be detected is through a series current limiting resistor. The module applies a current source or sink to the ZCD pin to maintain a constant voltage on the pin, thereby preventing the pin voltage from forward biasing the ESD protection diodes. When the applied voltage is greater than the reference voltage, the module sinks current. When the applied voltage is less than the reference voltage, the module sources current. The current source and sink action keeps the pin voltage constant over the full range of the applied voltage. The ZCD module is shown in the simplified block diagram Figure 24-2.

The ZCD module is useful when monitoring an A/C waveform for, but not limited to, the following purposes:

- A/C period measurement
- Accurate long term time measurement
- Dimmer phase delayed drive
- Low EMI cycle switching

24.1 External Resistor Selection

The ZCD module requires a current limiting resistor in series with the external voltage source. The impedance and rating of this resistor depends on the external source peak voltage. Select a resistor value that will drop all of the peak voltage when the current through the resistor is nominally 300 μA. Refer to Equation 24-1 and Figure 24-1. Make sure that the ZCD I/O pin internal weak pull-up is disabled so it does not interfere with the current source and sink.

EQUATION 24-1: EXTERNAL RESISTOR

$$R_{SERIES} = \frac{V_{PEAK}}{3 \times 10^{-4}}$$

FIGURE 24-1: EXTERNAL VOLTAGE

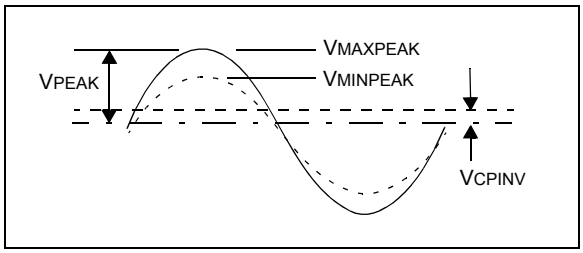
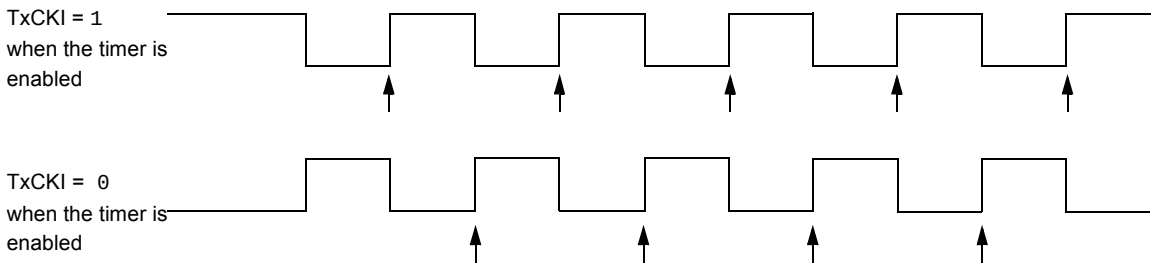


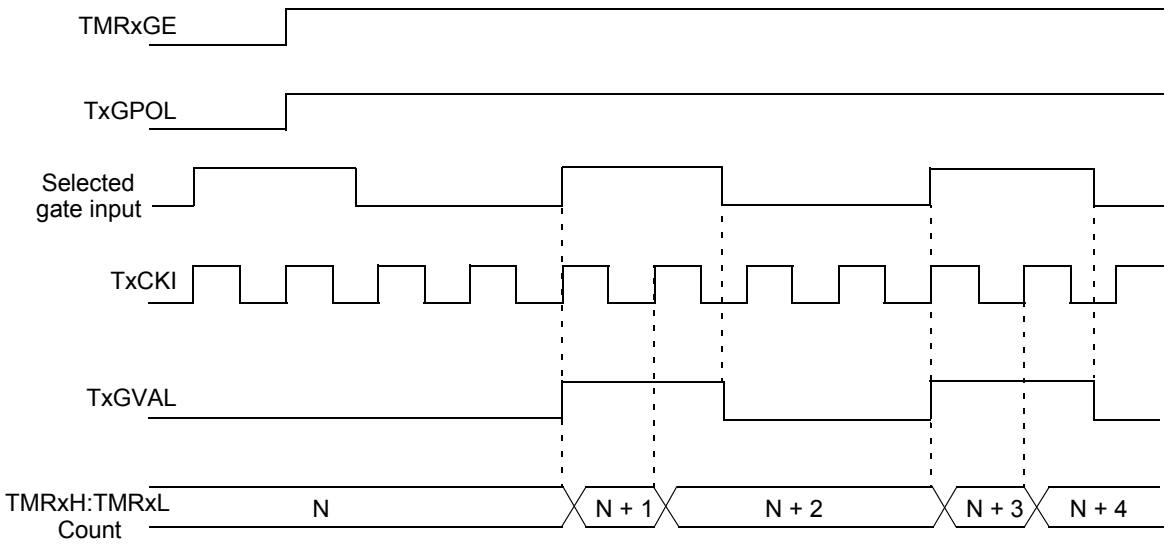
FIGURE 26-2: TIMER1 INCREMENTING EDGE



Note 1: Arrows indicate counter increments.

2: In Counter mode, a falling edge must be registered by the counter prior to the first incrementing rising edge of the clock.

FIGURE 26-3: TIMER1 GATE ENABLE MODE



REGISTER 26-4: T1GATE TIMER1 GATE SELECT REGISTER

U-0	U-0	U-0	R/W-0/u	R/W-0/u	R/W-0/u	R/W-0/u	R/W-0/u
—	—	—	GSS<4:0>				
bit 7							bit 0

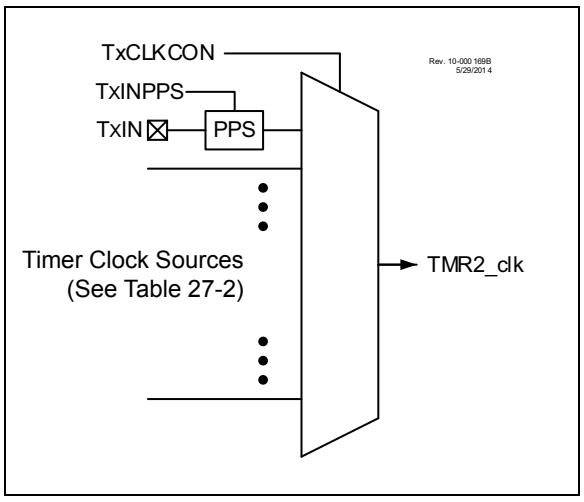
Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set	‘0’ = Bit is cleared	HC = Bit is cleared by hardware

bit 7-5 **Unimplemented:** Read as ‘0’

bit 4-0 **GSS<4:0>:** Timer1 Gate Select bits

11111-10001 = Reserved
10000 = LC4_out
01111 = LC3_out
01110 = LC2_out
01101 = LC1_out
00100 = ZCD1_output
01011 = C2OUT_sync
01010 = C1OUT_sync
01001 = NCO1_out
01000 = PWM6_out
00111 = PWM5_out
00110 = PWM4_out
00101 = PWM3_out
00100 = CCP2_out
00011 = CCP1_out
00010 = TMR2_postscaled
00001 = Timer0 overflow output
00000 = T1GPPS

FIGURE 27-2: TIMER2 CLOCK SOURCE BLOCK DIAGRAM



27.1 Timer2 Operation

Timer2 operates in three major modes:

- Free Running Period
- One-shot
- Monostable

Within each mode there are several options for starting, stopping, and reset. Table 27-1 lists the options.

In all modes, the TMR2 count register is incremented on the rising edge of the clock signal from the programmable prescaler. When TMR2 equals T2PR, a high level is output to the postscaler counter. TMR2 is cleared on the next clock input.

An external signal from hardware can also be configured to gate the timer operation or force a TMR2 count Reset. In Gate modes the counter stops when the gate is disabled and resumes when the gate is enabled. In Reset modes the TMR2 count is reset on either the level or edge from the external source.

The TMR2 and T2PR registers are both directly readable and writable. The TMR2 register is cleared and the T2PR register initializes to FFh on any device Reset. Both the prescaler and postscaler counters are cleared on the following events:

- a write to the TMR2 register
- a write to the T2CON register
- any device Reset
- External Reset Source event that resets the timer.

Note:	TMR2 is not cleared when T2CON is written.

27.1.1 FREE RUNNING PERIOD MODE

The value of TMR2 is compared to that of the Period register, T2PR, on each TMR2_clk cycle. When the two values match, the comparator resets the value of TMR2 to 00h on the next rising TMR2_clk edge and increments the output postscaler counter. When the postscaler count equals the value in the OUTPS<4:0> bits of the TMRxCON1 register, a one TMR2_clk period wide pulse occurs on the TMR2_postscaled output, and the postscaler count is cleared.

27.1.2 ONE-SHOT MODE

The One-Shot mode is identical to the Free Running Period mode except that the ON bit is cleared and the timer is stopped when TMR2 matches T2PR and will not restart until the T2ON bit is cycled off and on. Postscaler OUTPS<4:0> values other than 0 are meaningless in this mode because the timer is stopped at the first period event and the postscaler is reset when the timer is restarted.

27.1.3 MONOSTABLE MODE

Monostable modes are similar to One-Shot modes except that the ON bit is not cleared and the timer can be restarted by an external Reset event.

27.2 Timer2 Output

The Timer2 module's primary output is TMR2_postscaled, which pulses for a single TMR2_clk period when the postscaler counter matches the value in the OUTPS bits of the TMR2CON register. The T2PR postscaler is incremented each time the TMR2 value matches the T2PR value. This signal can be selected as an input to several other input modules:

- The ADC module, as an Auto-conversion Trigger
- COG, as an auto-shutdown source

In addition, the Timer2 is also used by the CCP module for pulse generation in PWM mode. Both the actual TMR2 value as well as other internal signals are sent to the CCP module to properly clock both the period and pulse width of the PWM signal. See **Section 28.0 "Capture/Compare/PWM Modules"** for more details on setting up Timer2 for use with the CCP, as well as the timing diagrams in **Section 27.5 "Operation Examples"** for examples of how the varying Timer2 modes affect CCP PWM output.

27.3 External Reset Sources

In addition to the clock source, the Timer2 also takes in an external Reset source. This external Reset source is selected for Timer2 with the T2RST register. This source can control starting and stopping of the timer, as well as resetting the timer, depending on which mode the timer is in. The mode of the timer is controlled by the MODE<4:0> bits of the TMRxHLT register. Edge-Triggered modes require six Timer clock periods between external triggers. Level-Triggered modes require the triggering level to be at least three Timer clock periods long. External triggers are ignored while in Debug Freeze mode.

REGISTER 28-2: CCPxCAP: CAPTURE INPUT SELECTION REGISTER

U-0	U-0	U-0	U-0	U-0	R/W-0/x	R/W-0/x	R/W-0/x
—	—	—	—	—	CTS<2:0>		
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Reset
‘1’ = Bit is set	‘0’ = Bit is cleared	

bit 7-3 **Unimplemented:** Read as ‘0’

bit 2-0 **CTS<2:0>:** Capture Trigger Input Selection bits

CTS	CCP1.capture	CCP2.capture
111	LC4_out	
110	LC3_out	
101	LC2_out	
100	LC1_out	
011	IOC_interrupt	
010	C2OUT	
001	C1OUT	
000	CCP1PPS	CCP2PPS

REGISTER 28-3: CCPRxL REGISTER: CCPx REGISTER LOW BYTE

R/W-x/x	R/W-x/x	R/W-x/x	R/W-x/x	R/W-x/x	R/W-x/x	R/W-x/x	R/W-x/x
CCPRx<7:0>							
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Reset
‘1’ = Bit is set	‘0’ = Bit is cleared	

bit 7-0 CCPxMODE = Capture mode
CCPRxL<7:0>: Capture value of TMR1L
CCPxMODE = Compare mode
CCPRxL<7:0>: LS Byte compared to TMR1L
CCPxMODE = PWM modes when CCPxFMT = 0:
CCPRxL<7:0>: Pulse-width Least Significant eight bits
CCPxMODE = PWM modes when CCPxFMT = 1:
CCPRxL<7:6>: Pulse-width Least Significant two bits
CCPRxL<5:0>: Not used.

32.2.3 SPI MASTER MODE

The master can initiate the data transfer at any time because it controls the SCK line. The master determines when the slave (Processor 2, Figure 32-5) is to broadcast data by the software protocol.

In Master mode, the data is transmitted/received as soon as the SSP1BUF register is written to. If the SPI is only going to receive, the SDO output could be disabled (programmed as an input). The SSP1SR register will continue to shift in the signal present on the SDI pin at the programmed clock rate. As each byte is received, it will be loaded into the SSP1BUF register as if a normal received byte (interrupts and Status bits appropriately set).

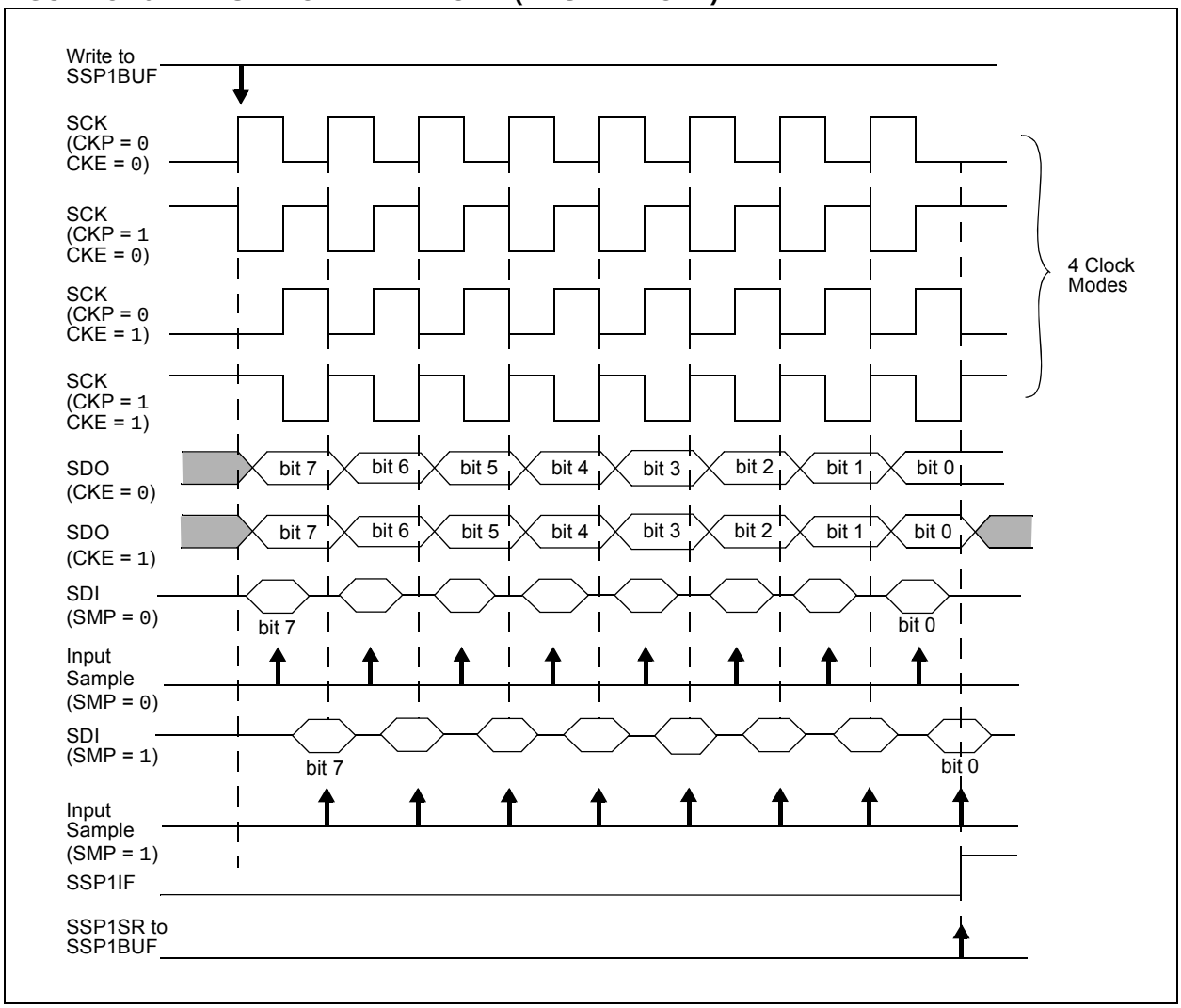
The clock polarity is selected by appropriately programming the CKP bit of the SSP1CON1 register and the CKE bit of the SSP1STAT register. This then, would give waveforms for SPI communication as shown in Figure 32-6, Figure 32-8, Figure 32-9 and Figure 32-10, where the MSB is transmitted first. In Master mode, the SPI clock rate (bit rate) is user programmable to be one of the following:

- FOSC/4 (or TCY)
- FOSC/16 (or 4 * TCY)
- FOSC/64 (or 16 * TCY)
- Timer2 output/2
- FOSC/(4 * (SSP1ADD + 1))

Figure 32-6 shows the waveforms for Master mode.

When the CKE bit is set, the SDO data is valid before there is a clock edge on SCK. The change of the input sample is shown based on the state of the SMP bit. The time when the SSP1BUF is loaded with the received data is shown.

FIGURE 32-6: SPI MODE WAVEFORM (MASTER MODE)



32.4.4 SDA HOLD TIME

The hold time of the SDA pin is selected by the SDAHT bit of the SSP1CON3 register. Hold time is the time SDA is held valid after the falling edge of SCL. Setting the SDAHT bit selects a longer 300 ns minimum hold time and may help on buses with large capacitance.

TABLE 32-1: I²C BUS TERMS

TERM	Description
Transmitter	The device which shifts data out onto the bus.
Receiver	The device which shifts data in from the bus.
Master	The device that initiates a transfer, generates clock signals and terminates a transfer.
Slave	The device addressed by the master.
Multi-master	A bus with more than one device that can initiate data transfers.
Arbitration	Procedure to ensure that only one master at a time controls the bus. Winning arbitration ensures that the message is not corrupted.
Synchronization	Procedure to synchronize the clocks of two or more devices on the bus.
Idle	No master is controlling the bus, and both SDA and SCL lines are high.
Active	Any time one or more master devices are controlling the bus.
Addressed Slave	Slave device that has received a matching address and is actively being clocked by a master.
Matching Address	Address byte that is clocked into a slave that matches the value stored in SSP1ADD.
Write Request	Slave receives a matching address with R/$\overline{W}$ bit clear, and is ready to clock in data.
Read Request	Master sends an address byte with the R/$\overline{W}$ bit set, indicating that it wishes to clock data out of the Slave. This data is the next and all following bytes until a Restart or Stop.
Clock Stretching	When a device on the bus hold SCL low to stall communication.
Bus Collision	Any time the SDA line is sampled low by the module while it is outputting and expected high state.

32.4.5 START CONDITION

The I²C specification defines a Start condition as a transition of SDA from a high to a low state while SCL line is high. A Start condition is always generated by the master and signifies the transition of the bus from an Idle to an active state. Figure 32-12 shows wave forms for Start and Stop conditions.

32.4.6 STOP CONDITION

A Stop condition is a transition of the SDA line from low-to-high state while the SCL line is high.

Note: At least one SCL low time must appear before a Stop is valid, therefore, if the SDA line goes low then high again while the SCL line stays high, only the Start condition is detected.

32.4.7 RESTART CONDITION

A Restart is valid any time that a Stop would be valid. A master can issue a Restart if it wishes to hold the bus after terminating the current transfer. A Restart has the same effect on the slave that a Start would, resetting all slave logic and preparing it to clock in an address. The master may want to address the same or another slave. Figure 32-13 shows the wave form for a Restart condition.

In 10-bit Addressing Slave mode a Restart is required for the master to clock data out of the addressed slave. Once a slave has been fully addressed, matching both high and low address bytes, the master can issue a Restart and the high address byte with the R/$\overline{W}$ bit set. The slave logic will then hold the clock and prepare to clock out data.

32.4.8 START/STOP CONDITION INTERRUPT MASKING

The SCIE and PCIE bits of the SSP1CON3 register can enable the generation of an interrupt in Slave modes that do not typically support this function. Slave modes where interrupt on Start and Stop detect are already enabled, these bits will have no effect.

32.6.4 I²C MASTER MODE START CONDITION TIMING

To initiate a Start condition (Figure 32-26), the user sets the Start Enable bit, SEN bit of the SSP1CON2 register. If the SDA and SCL pins are sampled high, the Baud Rate Generator is reloaded with the contents of SSP1ADD<7:0> and starts its count. If SCL and SDA are both sampled high when the Baud Rate Generator times out (TBRG), the SDA pin is driven low. The action of the SDA being driven low while SCL is high is the Start condition and causes the S bit of the SSP1STAT1 register to be set. Following this, the Baud Rate Generator is reloaded with the contents of SSP1ADD<7:0> and resumes its count. When the Baud Rate Generator times out (TBRG), the SEN bit of the SSP1CON2 register will be automatically cleared by hardware; the Baud Rate Generator is suspended, leaving the SDA line held low and the Start condition is complete.

Note 1: If at the beginning of the Start condition, the SDA and SCL pins are already sampled low, or if during the Start condition, the SCL line is sampled low before the SDA line is driven low, a bus collision occurs, the Bus Collision Interrupt Flag, BCLIF, is set, the Start condition is aborted and the I²C module is reset into its Idle state.

2: The Philips I²C specification states that a bus collision cannot occur on a Start.

FIGURE 32-26: FIRST START BIT TIMING

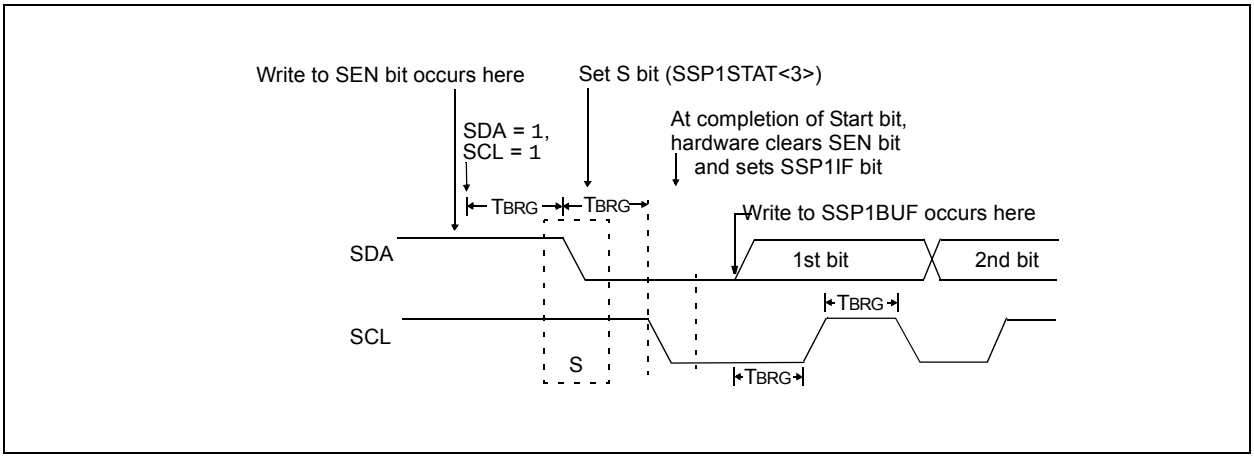
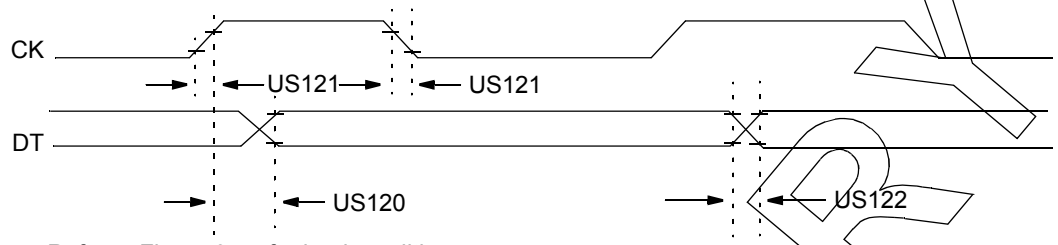


FIGURE 37-15: EUSART SYNCHRONOUS TRANSMISSION (MASTER/SLAVE) TIMING

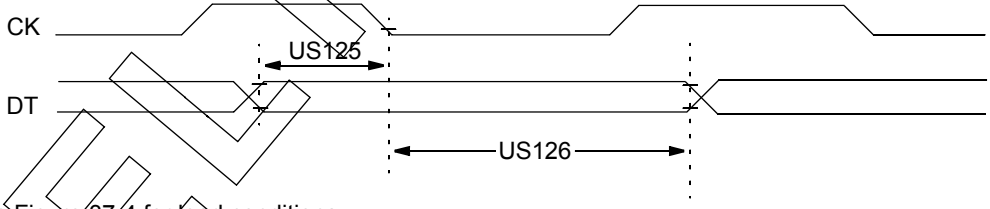


Note: Refer to Figure 37-4 for load conditions.

TABLE 37-21: EUSART SYNCHRONOUS TRANSMISSION CHARACTERISTICS

Standard Operating Conditions (unless otherwise stated)						
Param. No.	**Symbol**	**Characteristic**	**Min.**	**Max.**	**Units**	**Conditions**
US120	TCKH2DTV	SYNC XMIT (Master and Slave) Clock high to data-out valid	—	80	ns	3.0V ≤ VDD ≤ 5.5V
			—	100	ns	1.8V ≤ VDD ≤ 5.5V
US121	TCKRF	Clock out rise time and fall time (Master mode)	—	45	ns	3.0V ≤ VDD ≤ 5.5V
			—	50	ns	1.8V ≤ VDD ≤ 5.5V
US122	TDTRF	Data-out rise time and fall time	—	45	ns	3.0V ≤ VDD ≤ 5.5V
			—	50	ns	1.8V ≤ VDD ≤ 5.5V

FIGURE 37-16: EUSART SYNCHRONOUS RECEIVE (MASTER/SLAVE) TIMING



Note: Refer to Figure 37-4 for load conditions.

TABLE 37-22: EUSART SYNCHRONOUS RECEIVE REQUIREMENTS

Standard Operating Conditions (unless otherwise stated)						
Param. No.	**Symbol**	**Characteristic**	**Min.**	**Max.**	**Units**	**Conditions**
US125	TDTV2CKL	SYNC RCV (Master and Slave) Data-setup before CK ↓ (DT hold time)	10	—	ns	
US126	TCKL2DTL	Data-hold after CK ↓ (DT hold time)	15	—	ns	