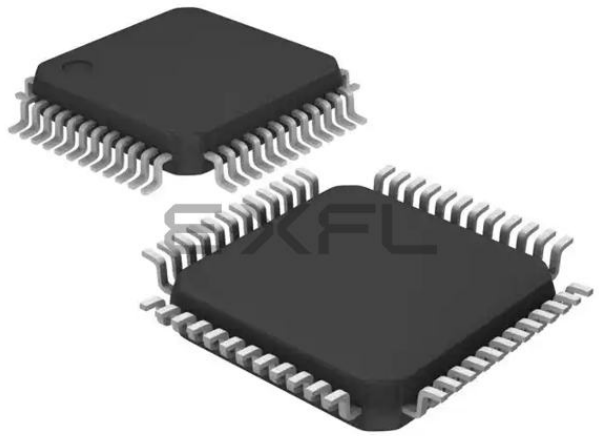


Welcome to E-XFL.COM

What is "Embedded - Microcontrollers"?

"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	S12Z
Core Size	16-Bit
Speed	32MHz
Connectivity	CANbus, I²C, IrDA, LINbus, SCI, SPI, UART/USART
Peripherals	LVD, POR, PWM, WDT
Number of I/O	34
Program Memory Size	128KB (128K x 8)
Program Memory Type	FLASH
EEPROM Size	512 x 8
RAM Size	1K x 8
Voltage - Supply (Vcc/Vdd)	5.5V ~ 18V
Data Converters	A/D 10x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 105°C (TA)
Mounting Type	Surface Mount
Package / Case	48-LQFP
Supplier Device Package	48-LQFP (7x7)
Purchase URL	https://www.e-xfl.com/pro/item?MUrl=&PartUrl=s912zvl12f0vlf

Chapter 8
ECC Generation Module (SRAM_ECCV2)

Chapter 9
S12 Clock, Reset and Power Management Unit (S12CPMU_UHV)

elapse between the consecutive ERASE_FLASH commands then a timeout occurs, which forces a soft reset and initializes the sequence. The ERASE bit is cleared when the mass erase sequence has been completed. No ACK is driven.

During the mass erase operation, which takes many clock cycles, the command status is indicated by the ERASE bit in BDCCSR. Whilst a mass erase operation is ongoing, Always-available commands can be issued. This allows the status of the erase operation to be polled by reading BDCCSR to determine when the operation is finished.

The status of the flash array can be verified by subsequently reading the flash error flags to determine if the erase completed successfully.

ERASE_FLASH can be aborted by a SYNC pulse forcing a soft reset.

NOTE: Device Bus Frequency Considerations

The ERASE_FLASH command requires the default device bus clock frequency after reset. Thus the bus clock frequency must not be changed following reset before issuing an ERASE_FLASH command.

5.4.4.20 STEP1

This command is used to step through application code. In active BDM this command executes the next CPU instruction in application code. If enabled an ACK is driven.

If a STEP1 command is issued and the CPU is not halted, the command is ignored.

Using STEP1 to step through a CPU WAI instruction is explained in Section 5.1.3.3.2, “Wait Mode.

5.4.5 BDC Access Of Internal Resources

Unsuccessful read accesses of internal resources return a value of 0xEE for each data byte. This enables a debugger to recognize a potential error, even if neither the ACK handshaking protocol nor a status command is currently being executed. The value of 0xEE is returned in the following cases.

- Illegal address access, whereby ILLACC is set
- Invalid READ_SAME or DUMP_MEM sequence
- Invalid READ_Rn command (BDM inactive or CRN incorrect)
- Internal resource read with timeout, whereby NORESP is set

The BDC serial interface uses a clocking scheme in which the external host generates a falling edge on the BKGD pin to indicate the start of each bit time. This falling edge is sent for every bit whether data is transmitted or received. Data is transferred most significant bit (MSB) first at 16 target clock cycles per bit. The interface times out if during a command 512 clock cycles occur between falling edges from the host. The timeout forces the current command to be discarded.

The BKGD pin is a pseudo open-drain pin and has a weak on-chip active pull-up that is enabled at all times. It is assumed that there is an external pull-up and that drivers connected to BKGD do not typically drive the high level. Since R-C rise time could be unacceptably long, the target system and host provide brief drive-high (speedup) pulses to drive BKGD to a logic 1. The source of this speedup pulse is the host for transmit cases and the target for receive cases.

The timing for host-to-target is shown in Figure 5-6 and that of target-to-host in Figure 5-7 and Figure 5-8. All cases begin when the host drives the BKGD pin low to generate a falling edge. Since the host and target operate from separate clocks, it can take the target up to one full clock cycle to recognize this edge; this synchronization uncertainty is illustrated in Figure 5-6. The target measures delays from this perceived start of the bit time while the host measures delays from the point it actually drove BKGD low to start the bit up to one target clock cycle earlier. Synchronization between the host and target is established in this manner at the start of every bit time.

Figure 5-6 shows an external host transmitting a logic 1 and transmitting a logic 0 to the BKGD pin of a target system. The host is asynchronous to the target, so there is up to a one clock-cycle delay from the host-generated falling edge to where the target recognizes this edge as the beginning of the bit time. Ten target clock cycles later, the target senses the bit level on the BKGD pin. Internal glitch detect logic requires the pin be driven high no later than eight target clock cycles after the falling edge for a logic 1 transmission.

Since the host drives the high speedup pulses in these two cases, the rising edges look like digitally driven signals.

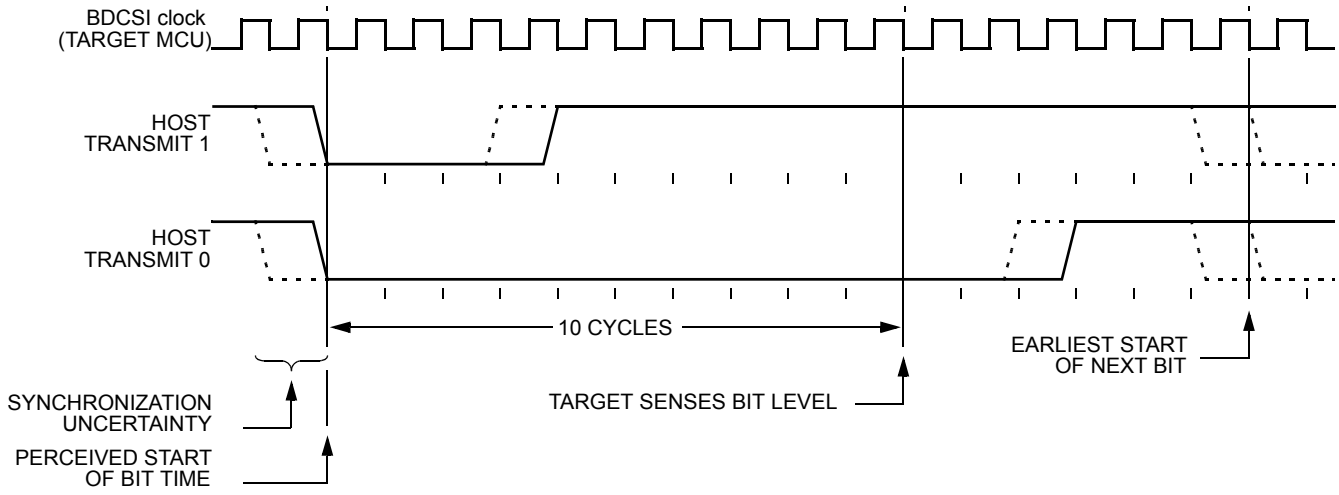


Figure 5-6. BDC Host-to-Target Serial Bit Timing

Figure 5-7 shows the host receiving a logic 1 from the target system. The host holds the BKGD pin low long enough for the target to recognize it (at least two target clock cycles). The host must release the low

10.5.2.12 ADC Conversion Interrupt Flag Register (ADCCONIF)

After being set any of these bits can be cleared by writing a value of 1’b1. All bits are cleared if bit ADC_EN is clear or via ADC soft-reset (bit ADC_SR set). Writing any flag with value 1’b0 does not clear the flag. Writing any flag with value 1’b1 does not set the flag.

Module Base + 0x000C

	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
R W	CON_IF[15:1]															EOL_IF
Reset	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0

= Unimplemented or Reserved

Figure 10-15. ADC Conversion Interrupt Flag Register (ADCCONIF)

Read: Anytime

Write: Anytime

Table 10-17. ADCCONIF Field Descriptions

Field	Description
15-1 CON_IF[15:1]	**Conversion Interrupt Flags** — These bits could be set by the binary coded interrupt select bits INTFLG_SEL[3:0] when the corresponding conversion command has been processed and related data has been stored to RAM. See also notes below.
0 EOL_IF	**End Of List Interrupt Flag** — This bit is set by the binary coded conversion command type select bits CMD_SEL[1:0] for “end of list” type of commands and after such a command has been processed and the related data has been stored RAM. See also second note below

NOTE

These bits can be used to indicate if a certain packet of conversion results is available. Clearing a flag indicates that conversion results have been retrieved by software and the flag can be used again (see also Section 10.9.6, “RVL swapping in RVL double buffer mode and related registers ADCIMDRI and ADCEOLRI.

NOTE

Overrun situation of a flag CON_IF[15:1] and EOL_IF are indicated by flag CONIF_OIF.

10.5.2.16 ADC Command Register 1 (ADCCMD_1)

A command which contains reserved bit settings causes the error flag CMD_EIF being set and ADC cease operation. The CMD_EIF is never set for Internal_x channels, even if the channels are specified as reserved in the Device Overview section of the Reference Manual.

Module Base + 0x0015

	23	22	21	20	19	18	17	16
R W	VRH_SEL[1]	VRL_SEL[1]	CH_SEL[5:0]					
R W	VRH_SEL[1:0][2]		CH_SEL[5:0]					
Reset	0	0	0	0	0	0	0	0

= Unimplemented or Reserved

Figure 10-19. ADC Command Register 1 (ADCCMD_1)

1 Only available on ADC12B_LBA V1 and V2 (see Table 10-2 for details)

2 Only available on ADC12B_LBA V3 (see Table 10-2 for details)

Read: Anytime

Write: Only writable if bit SMOD_ACC is set
(see also Section 10.5.2.2, “ADC Control Register 1 (ADCCTL_1) bit SMOD_ACC description for more details)

Table 10-23. ADCCMD_1 Field Descriptions

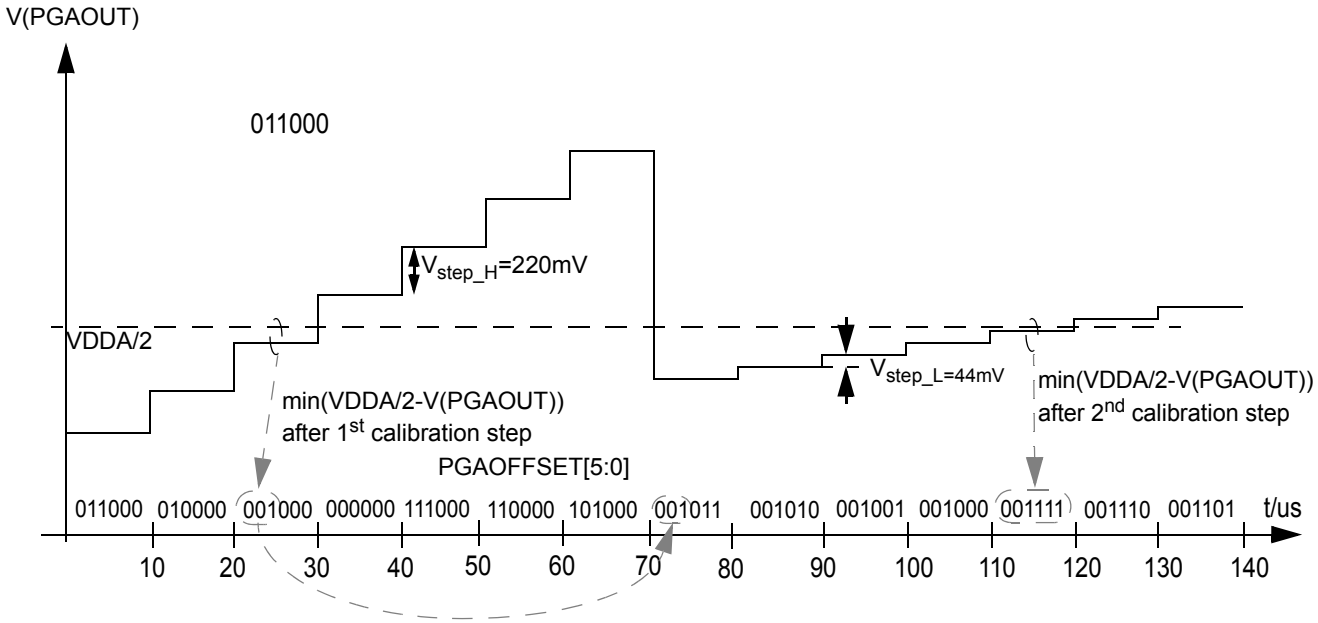
Field	Description
	ADC12B_LBA V1 and V2 (includes VRH_SEL/VRL_SEL)
23 VRH_SEL	**Reference High Voltage Select Bit** — This bit selects the high voltage reference for current conversion. 0 VRH_0 input selected as high voltage reference. 1 VRH_1 input selected as high voltage reference.
22 VRL_SEL	**Reference Low Voltage Select Bit** — This bit selects the low voltage reference for current conversion. 0 VRL_0 input selected as low voltage reference. 1 VRL_1 input selected as low voltage reference.
	ADC12B_LBA V3 (includes VRH_SEL[1:0])
23-22 VRH_SEL	**Reference High Voltage Select Bit** — These bits select the high voltage reference for current conversion. 00 VRH_0 input selected as high voltage reference 01 VRH_1 input selected as high voltage reference 10 VRH_2 input selected as high voltage reference 11 Reserved
21-16 CH_SEL[5:0]	**ADC Input Channel Select Bits** — These bits select the input channel for the current conversion. See Table 10-24 for channel coding information.

NOTE

If bit SMOD_ACC is set modifying this register must be done carefully - only when no conversion and conversion sequence is ongoing.

2. With the optimal PGAOFFSET[5:3] setting step through the offset compensation values PGAOFFSET[2:0]= {0x011, 0x010, 0x001, 0x000, 0x111, 0x110, 0x101} and measure the PGA_OUT value with the ADC. Select as optimal offset compensation value for the lower three bits the PGAOFFSET[2:0] which is closest to the expected ADC reading of VDDA/2.

Figure 12-8. Offset compensation timing diagram



12.4.3 Application Example for differential voltage measurement

For sensor applications it is often required to measure a small differential voltage V_{diff}. The PGA is not capable of amplifying a differential voltage, but an algorithm to calculate the differential voltage can be implemented. The PGA contains two input pins PGA_IN0 and PGA_IN1 which can be multiplexed by the ADC command list. By subtracting the ADC readings of the two pins the amplified differential voltage can be calculated.

For this algorithm two requirements must be met:

1. The minimum time for the input signal multiplexing is given by PGA to ADC settling time $t_{PGA\_settling}$. The rate of signal change within $t_{PGA\_settling}$ must be small.
2. The common mode input voltage range of the differential input signals must be limited that for a given gain A_{PGA} a reference voltage V_{ref} can be selected so that both amplified signals do not saturate.

If both requirements are met the algorithm can be implemented. The error calculation is the following:

$$V_{diff} = (ADC\_reading(PGA\_IN1) - ADC\_reading(PGA\_IN0))/A_{PGA} \qquad \textit{Eqn. 12-1}$$

13.1.3 Features

The basic features of the MSCAN are as follows:

- Implementation of the CAN protocol — Version 2.0A/B
  - — Standard and extended data frames
  - — Zero to eight bytes data length
  - — Programmable bit rate up to 1 Mbps[1]
  - — Support for remote frames
- Five receive buffers with FIFO storage scheme
- Three transmit buffers with internal prioritization using a “local priority” concept
- Flexible maskable identifier filter supports two full-size (32-bit) extended identifier filters, or four 16-bit filters, or eight 8-bit filters
- Programmable wake-up functionality with integrated low-pass filter
- Programmable loopback mode supports self-test operation
- Programmable listen-only mode for monitoring of CAN bus
- Separate signalling and interrupt capabilities for all CAN receiver and transmitter error states (warning, error passive, bus-off)
- Programmable MSCAN clock source either bus clock or oscillator clock
- Internal timer for time-stamping of received and transmitted messages
- Three low-power modes: sleep, power down, and MSCAN enable
- Global initialization of configuration registers

13.1.4 Modes of Operation

For a description of the specific MSCAN modes and the module operation related to the system operating modes refer to Section 13.4.4, “Modes of Operation”.

1. Depending on the actual bit timing and the clock jitter of the PLL.

NOTE

The CANCTL0 register, except WUPE, INITRQ, and SLPRQ, is held in the reset state when the initialization mode is active (INITRQ = 1 and INITAK = 1). This register is writable again as soon as the initialization mode is exited (INITRQ = 0 and INITAK = 0).

Table 13-3. CANCTL0 Register Field Descriptions

Field	Description
7 RXFRM	**Received Frame Flag** — This bit is read and clear only. It is set when a receiver has received a valid message correctly, independently of the filter configuration. After it is set, it remains set until cleared by software or reset. Clearing is done by writing a 1. Writing a 0 is ignored. This bit is not valid in loopback mode. 0 No valid message was received since last clearing this flag 1 A valid message was received since last clearing of this flag
6 RXACT	**Receiver Active Status** — This read-only flag indicates the MSCAN is receiving a message[1]. The flag is controlled by the receiver front end. This bit is not valid in loopback mode. 0 MSCAN is transmitting or idle 1 MSCAN is receiving a message (including when arbitration is lost)
5 CSWAI[2]	**CAN Stops in Wait Mode** — Enabling this bit allows for lower power consumption in wait mode by disabling all the clocks at the CPU bus interface to the MSCAN module. 0 The module is not affected during wait mode 1 The module ceases to be clocked during wait mode
4 SYNCH	**Synchronized Status** — This read-only flag indicates whether the MSCAN is synchronized to the CAN bus and able to participate in the communication process. It is set and cleared by the MSCAN. 0 MSCAN is not synchronized to the CAN bus 1 MSCAN is synchronized to the CAN bus
3 TIME	**Timer Enable** — This bit activates an internal 16-bit wide free running timer which is clocked by the bit clock rate. If the timer is enabled, a 16-bit time stamp will be assigned to each transmitted/received message within the active TX/RX buffer. Right after the EOF of a valid message on the CAN bus, the time stamp is written to the highest bytes (0x000E, 0x000F) in the appropriate buffer (see Section 13.3.3, “Programmer’s Model of Message Storage”). In loopback mode no receive timestamp is generated. The internal timer is reset (all bits set to 0) when disabled. This bit is held low in initialization mode. 0 Disable internal MSCAN timer 1 Enable internal MSCAN timer
2 WUPE[3]	**Wake-Up Enable** — This configuration bit allows the MSCAN to restart from sleep mode or from power down mode (entered from sleep) when traffic on CAN is detected (see Section 13.4.5.5, “MSCAN Sleep Mode”). This bit must be configured before sleep mode entry for the selected function to take effect. 0 Wake-up disabled — The MSCAN ignores traffic on CAN 1 Wake-up enabled — The MSCAN is able to restart

eight identifier acceptance filters.
Figure 13-39 shows how the first 32-bit filter bank (CANIDAR0–CANIDAR3, CANIDMR0–CANIDMR3) produces a filter 0 hit. Similarly, the second filter bank (CANIDAR4–CANIDAR7, CANIDMR4–CANIDMR7) produces a filter 1 hit.

- Four identifier acceptance filters, each to be applied to:
  - The 14 most significant bits of the extended identifier plus the SRR and IDE bits of CAN 2.0B messages.
  - The 11 bits of the standard identifier, the RTR and IDE bits of CAN 2.0A/B messages.
    Figure 13-40 shows how the first 32-bit filter bank (CANIDAR0–CANIDAR3, CANIDMR0–CANIDMR3) produces filter 0 and 1 hits. Similarly, the second filter bank (CANIDAR4–CANIDAR7, CANIDMR4–CANIDMR7) produces filter 2 and 3 hits.
- Eight identifier acceptance filters, each to be applied to the first 8 bits of the identifier. This mode implements eight independent filters for the first 8 bits of a CAN 2.0A/B compliant standard identifier or a CAN 2.0B compliant extended identifier.
  Figure 13-41 shows how the first 32-bit filter bank (CANIDAR0–CANIDAR3, CANIDMR0–CANIDMR3) produces filter 0 to 3 hits. Similarly, the second filter bank (CANIDAR4–CANIDAR7, CANIDMR4–CANIDMR7) produces filter 4 to 7 hits.
- Closed filter. No CAN message is copied into the foreground buffer RxFG, and the RXF flag is never set.

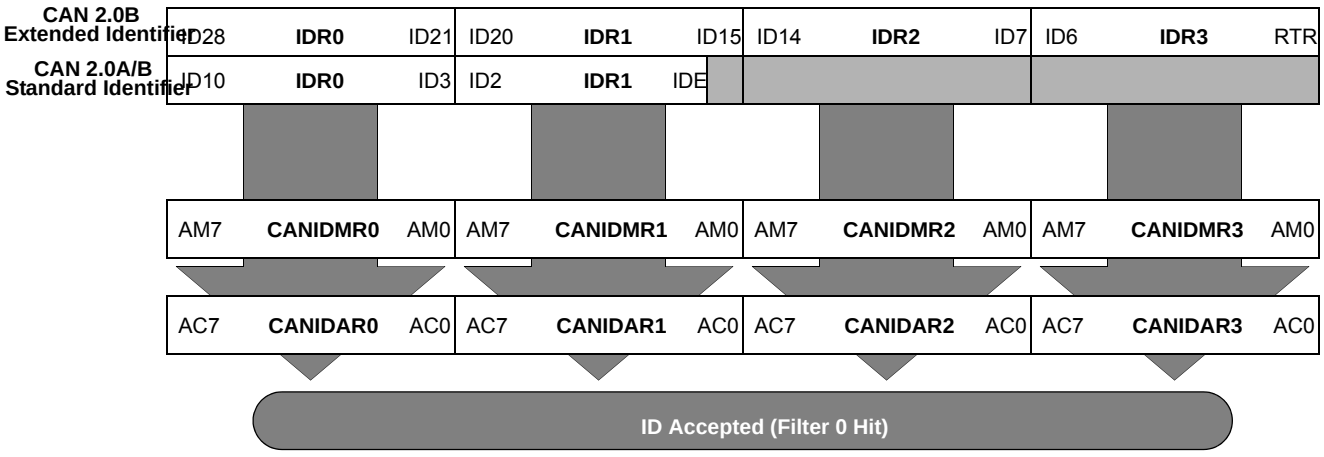


Figure 13-39. 32-bit Maskable Identifier Acceptance Filter

13.4.7.3 Receive Interrupt

A message is successfully received and shifted into the foreground buffer (RxFG) of the receiver FIFO. This interrupt is generated immediately after receiving the EOF symbol. The RXF flag is set. If there are multiple messages in the receiver FIFO, the RXF flag is set as soon as the next message is shifted to the foreground buffer.

13.4.7.4 Wake-Up Interrupt

A wake-up interrupt is generated if activity on the CAN bus occurs during MSCAN sleep or power-down mode.

NOTE

This interrupt can only occur if the MSCAN was in sleep mode (SLPRQ = 1 and SLPAK = 1) before entering power down mode, the wake-up option is enabled (WUPE = 1), and the wake-up interrupt is enabled (WUPIE = 1).

13.4.7.5 Error Interrupt

An error interrupt is generated if an overrun of the receiver FIFO, error, warning, or bus-off condition occurs. MSCAN Receiver Flag Register (CANRFLG) indicates one of the following conditions:

- **Overrun** — An overrun condition of the receiver FIFO as described in Section 13.4.2.3, “Receive Structures,” occurred.
- **CAN Status Change** — The actual value of the transmit and receive error counters control the CAN bus state of the MSCAN. As soon as the error counters skip into a critical range (Tx/Rx-warning, Tx/Rx-error, bus-off) the MSCAN flags an error condition. The status change, which caused the error condition, is indicated by the TSTAT and RSTAT flags (see Section 13.3.2.5, “MSCAN Receiver Flag Register (CANRFLG)” and Section 13.3.2.6, “MSCAN Receiver Interrupt Enable Register (CANRIER)”).

13.4.7.6 Interrupt Acknowledge

Interrupts are directly associated with one or more status flags in either the MSCAN Receiver Flag Register (CANRFLG) or the MSCAN Transmitter Flag Register (CANTFLG). Interrupts are pending as long as one of the corresponding flags is set. The flags in CANRFLG and CANTFLG must be reset within the interrupt handler to handshake the interrupt. The flags are reset by writing a 1 to the corresponding bit position. A flag cannot be cleared if the respective condition prevails.

NOTE

It must be guaranteed that the CPU clears only the bit causing the current interrupt. For this reason, bit manipulation instructions (BSET) must not be used to clear interrupt flags. These instructions may cause accidental clearing of interrupt flags which are set after entering the current interrupt service routine.

Table 15-12. Timer Clock Selection

PR2	PR1	PR0	Timer Clock
0	0	0	Bus Clock / 1
0	0	1	Bus Clock / 2
0	1	0	Bus Clock / 4
0	1	1	Bus Clock / 8
1	0	0	Bus Clock / 16
1	0	1	Bus Clock / 32
1	1	0	Bus Clock / 64
1	1	1	Bus Clock / 128

NOTE

The newly selected prescale factor will not take effect until the next synchronized edge where all prescale counter stages equal zero.

15.3.2.10 Main Timer Interrupt Flag 1 (TFLG1)

Module Base + 0x000E

	7	6	5	4	3	2	1	0
R W	RESERVED	RESERVED	C5F	C4F	C3F	C2F	C1F	C0F
Reset	0	0	0	0	0	0	0	0

Figure 15-16. Main Timer Interrupt Flag 1 (TFLG1)

Read: Anytime

Write: Used in the clearing mechanism (set bits cause corresponding bits to be cleared). Writing a zero will not affect current status of the bit.

Table 15-13. TRLG1 Field Descriptions

Note: Writing to unavailable bits has no effect. Reading from unavailable bits return a zero.

Field	Description
5:0 C[5:0]F	**Input Capture/Output Compare Channel “x” Flag** — These flags are set when an input capture or output compare event occurs. Clearing requires writing a one to the corresponding flag bit while TEN is set to one. **Note:** When TFFCA bit in TSCR register is set, a read from an input capture or a write into an output compare channel (0x0010–0x001F) will cause the corresponding channel flag CxF to be cleared.

17.2.1 PWM7 - PWM0 — PWM Channel 7 - 0

Those pins serve as waveform output of PWM channel 7 - 0.

17.3 Memory Map and Register Definition

17.3.1 Module Memory Map

This section describes the content of the registers in the scalable PWM module. The base address of the scalable PWM module is determined at the MCU level when the MCU is defined. The register decode map is fixed and begins at the first address of the module address offset. The figure below shows the registers associated with the scalable PWM and their relative offset from the base address. The register detail description follows the order they appear in the register map.

Reserved bits within a register will always read as 0 and the write will be unimplemented. Unimplemented functions are indicated by shading the bit.

NOTE

Register Address = Base Address + Address Offset, where the Base Address is defined at the MCU level and the Address Offset is defined at the module level.

17.3.2 Register Descriptions

This section describes in detail all the registers and register bits in the scalable PWM module.

Register Name		Bit 7	6	5	4	3	2	1	Bit 0
0x0000 PWME[1]	R	PWME7	PWME6	PWME5	PWME4	PWME3	PWME2	PWME1	PWME0
	W								
0x0001 PWMPOL[1]	R	PPOL7	PPOL6	PPOL5	PPOL4	PPOL3	PPOL2	PPOL1	PPOL0
	W								
0x0002 PWMCLK[1]	R	PCLK7	PCLKL6	PCLK5	PCLK4	PCLK3	PCLK2	PCLK1	PCLK0
	W								
0x0003 PWMPRCLK	R	0	PCKB2	PCKB1	PCKB0	0	PCKA2	PCKA1	PCKA0
	W								
0x0004 PWMCAE[1]	R	CAE7	CAE6	CAE5	CAE4	CAE3	CAE2	CAE1	CAE0
	W								
0x0005 PWMCTL[1]	R	CON67	CON45	CON23	CON01	PSWAI	PFRZ	0	0
	W								

= Unimplemented or Reserved

Figure 17-2. The scalable PWM Register Summary (Sheet 1 of 4)

Table 18-3. SCICR1 Field Descriptions (continued)

Field	Description
5 RSRC	**Receiver Source Bit** — When LOOPS = 1, the RSRC bit determines the source for the receiver shift register input. See Table 18-4. 0 Receiver input internally connected to transmitter output 1 Receiver input connected externally to transmitter
4 M	**Data Format Mode Bit** — MODE determines whether data characters are eight or nine bits long. 0 One start bit, eight data bits, one stop bit 1 One start bit, nine data bits, one stop bit
3 WAKE	**Wakeup Condition Bit** — WAKE determines which condition wakes up the SCI: a logic 1 (address mark) in the most significant bit position of a received data character or an idle condition on the RXD pin. 0 Idle line wakeup 1 Address mark wakeup
2 ILT	**Idle Line Type Bit** — ILT determines when the receiver starts counting logic 1s as idle character bits. The counting begins either after the start bit or after the stop bit. If the count begins after the start bit, then a string of logic 1s preceding the stop bit may cause false recognition of an idle character. Beginning the count after the stop bit avoids false idle character recognition, but requires properly synchronized transmissions. 0 Idle character bit count begins after start bit 1 Idle character bit count begins after stop bit
1 PE	**Parity Enable Bit** — PE enables the parity function. When enabled, the parity function inserts a parity bit in the most significant bit position. 0 Parity function disabled 1 Parity function enabled
0 PT	**Parity Type Bit** — PT determines whether the SCI generates and checks for even parity or odd parity. With even parity, an even number of 1s clears the parity bit and an odd number of 1s sets the parity bit. With odd parity, an odd number of 1s clears the parity bit and an even number of 1s sets the parity bit. 0 Even parity 1 Odd parity

Table 18-4. Loop Functions

LOOPS	RSRC	Function
0	x	Normal operation
1	0	Loop mode with transmitter output internally connected to receiver input
1	1	Single-wire mode with TXD pin connected to receiver input

18.3.2.3 SCI Alternative Status Register 1 (SCIASR1)

Module Base + 0x0000

	7	6	5	4	3	2	1	0
R	RXEDGIF	0	0	0	0	BERRV	BERRIF	BKDIF
W								
Reset	0	0	0	0	0	0	0	0

= Unimplemented or Reserved

Figure 18-6. SCI Alternative Status Register 1 (SCIASR1)

Read: Anytime, if AMAP = 1

Table 18-10. SCICR2 Field Descriptions (continued)

Field	Description
3 TE	**Transmitter Enable Bit** — TE enables the SCI transmitter and configures the TXD pin as being controlled by the SCI. The TE bit can be used to queue an idle preamble. 0 Transmitter disabled 1 Transmitter enabled
2 RE	**Receiver Enable Bit** — RE enables the SCI receiver. 0 Receiver disabled 1 Receiver enabled
1 RWU	**Receiver Wakeup Bit** — Standby state 0 Normal operation. 1 RWU enables the wakeup function and inhibits further receiver interrupt requests. Normally, hardware wakes the receiver by automatically clearing RWU.
0 SBK	**Send Break Bit** — Toggling SBK sends one break character (10 or 11 logic 0s, respectively 13 or 14 logics 0s if BRK13 is set). Toggling implies clearing the SBK bit before the break character has finished transmitting. As long as SBK is set, the transmitter continues to send complete break characters (10 or 11 bits, respectively 13 or 14 bits). 0 No break characters 1 Transmit break characters

18.3.2.7 SCI Status Register 1 (SCISR1)

The SCISR1 and SCISR2 registers provides inputs to the MCU for generation of SCI interrupts. Also, these registers can be polled by the MCU to check the status of these bits. The flag-clearing procedures require that the status register be read followed by a read or write to the SCI data register.It is permissible to execute other instructions between the two steps as long as it does not compromise the handling of I/O, but the order of operations is important for flag clearing.

Module Base + 0x0004

	7	6	5	4	3	2	1	0
R	TDRE	TC	RDRF	IDLE	OR	NF	FE	PF
W								
Reset	1	1	0	0	0	0	0	0

= Unimplemented or Reserved

Figure 18-10. SCI Status Register 1 (SCISR1)

Read: Anytime

Write: Has no meaning or effect

18.4.5 Transmitter

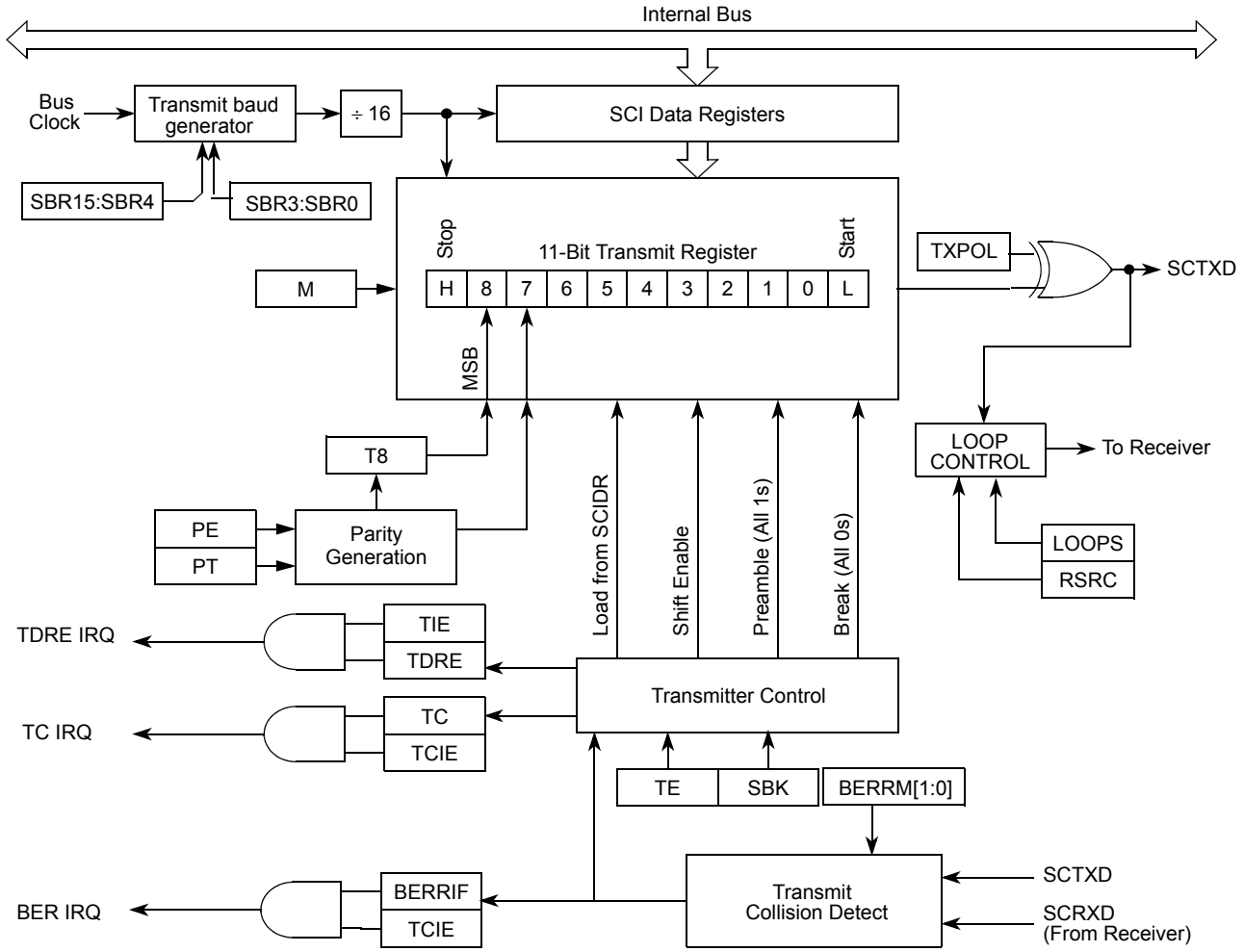


Figure 18-16. Transmitter Block Diagram

18.4.5.1 Transmitter Character Length

The SCI transmitter can accommodate either 8-bit or 9-bit data characters. The state of the M bit in SCI control register 1 (SCICR1) determines the length of data characters. When transmitting 9-bit data, bit T8 in SCI data register high (SCIDRH) is the ninth bit (bit 8).

18.4.5.2 Character Transmission

To transmit data, the MCU writes the data bits to the SCI data registers (SCIDRH/SCIDRL), which in turn are transferred to the transmitter shift register. The transmit shift register then shifts a frame out through the TXD pin, after it has prefaced them with a start bit and appended them with a stop bit. The SCI data registers (SCIDRH and SCIDRL) are the write-only buffers between the internal data bus and the transmit shift register.

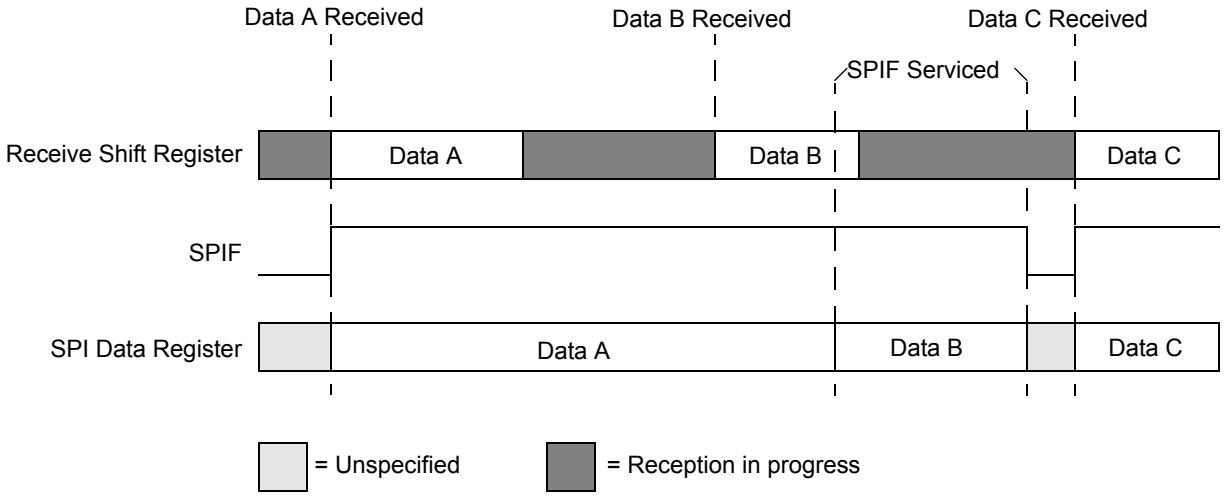


Figure 19-9. Reception with SPIF serviced in Time

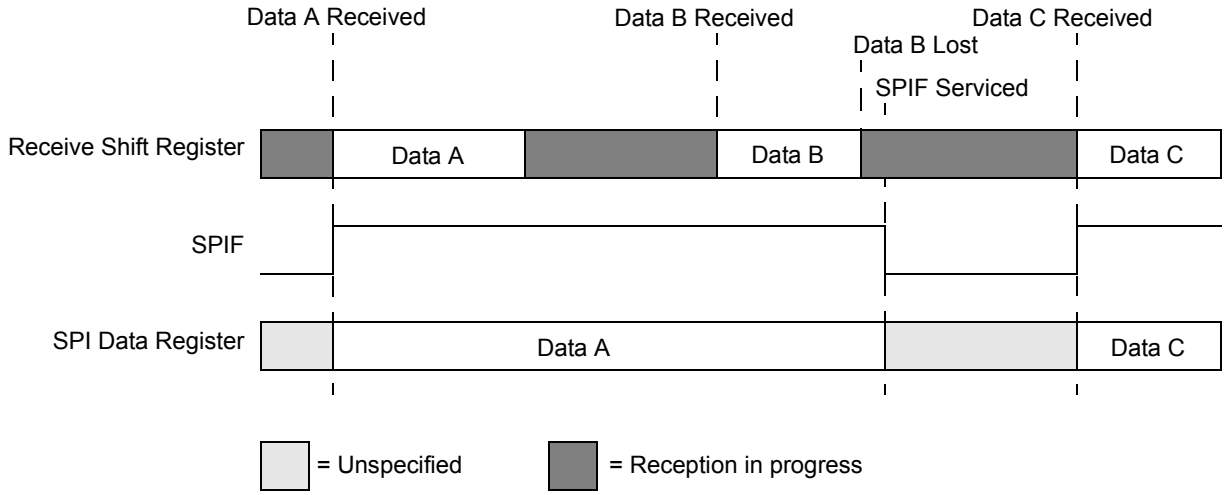


Figure 19-10. Reception with SPIF serviced too late

19.4 Functional Description

The SPI module allows a duplex, synchronous, serial communication between the MCU and peripheral devices. Software can poll the SPI status flags or SPI operation can be interrupt driven.

The SPI system is enabled by setting the SPI enable (SPE) bit in SPI control register 1. While SPE is set, the four associated SPI port pins are dedicated to the SPI function as:

- Slave select ($\overline{SS}$)
- Serial clock (SCK)
- Master out/slave in (MOSI)
- Master in/slave out (MISO)

Table 20-4. I-Bus Tap and Prescale Values

IBC2-0 (bin)	SCL Tap (clocks)	SDA Tap (clocks)
000	5	1
001	6	1
010	7	2
011	8	2
100	9	3
101	10	3
110	12	4
111	15	4

Table 20-5. Prescale Divider Encoding

IBC5-3 (bin)	scl2start (clocks)	scl2stop (clocks)	scl2tap (clocks)	tap2tap (clocks)
000	2	7	4	1
001	2	7	4	2
010	2	9	6	4
011	6	9	6	8
100	14	17	14	16
101	30	33	30	32
110	62	65	62	64
111	126	129	126	128

Table 20-6. Multiplier Factor

IBC7-6	MUL
00	01
01	02
10	04
11	RESERVED

The number of clocks from the falling edge of SCL to the first tap (Tap[1]) is defined by the values shown in the scl2tap column of Table 20-4, all subsequent tap points are separated by 2^{IBC5-3} as shown in the tap2tap column in Table 20-5. The SCL Tap is used to generated the SCL period and the SDA Tap is used to determine the delay from the falling edge of SCL to SDA changing, the SDA hold time.

IBC7–6 defines the multiplier factor MUL. The values of MUL are shown in the Table 20-6.

G.1 Static Electrical Characteristics

Table G-1. Static Electrical Characteristics - dac_8b5v_analog_ll18 @5V VDDA

Characteristics noted under conditions 4.85V ≤ V_{DDA} ≤ 5.15V, -40°C ≤ T_J ≤ 175°C, V_{RH} = V_{DDA}, V_{RL} = V_{SSA} unless otherwise noted. Typical values noted reflect the approximate parameter mean at T_A = 25°C under nominal conditions unless otherwise noted.

Num	Ratings	Symbol	Min	Typ	Max	Unit
1	Supply Current of dac_8b5v_analog_ll18 buffer disabled buffer enabled FVR=0 DRIVE=1 buffer enabled FVR=1 DRIVE=0	I_{buf}	- - -	- 365 215	5 800 800	µA
2	Reference current reference disabled reference enabled	I_{ref}	-	- 50	1 150	µA
3	Resolution		8			bit
4	Relative Accuracy measured at AMP -40°C < T_J ≤ 150°C 150°C < T_J < 175°C	INL	-0.5 -0.75		+0.5 +0.75	LSB
5	Differential Nonlinearity measure at AMP -40°C < T_J ≤ 150°C 150°C < T_J < 175°C	DNL	-0.5 -0.75		+0.5 +0.75	LSB
6	DAC Range A (FVR bit = 1)	V_{out}	0...255/256(V_{RH}-V_{RL})+V_{RL}			V
7	DAC Range B (FVR bit = 0	V_{out}	32...287/320(V_{RH}-V_{RL})+V_{RL}			V
8	Output Voltage unbuffered range A or B (load >= 50MΩ)	V_{out}	full DAC Range A or B			V
9	Output Voltage (DRIVE bit = 0) *) buffered range A (load >= 100KΩ to VSSA) or buffered range A (load >= 100KΩ to VDDA) buffered range B (load >= 100KΩ to VSSA) buffered range B (load >= 100KΩ to VDDA)	V_{out}	0 0.15 full DAC Range B	- -	V_{DDA}-0.15 V_{DDA}	V
10	Output Voltage (DRIVE bit = 1) **) buffered range B with 6.4KΩ load into resistor divider of 800Ω /6.56KΩ between VDDA and VSSA. (equivalent load is >= 65KΩ to VSSA) or (equivalent load is >= 7.5KΩ to VDDA)	V_{out}	full DAC Range B			V
11	Buffer Output Capacitive load	C_{load}	0	-	100	pF
12	Buffer Output Offset	V_{offset}	-30	-	+30	mV
13	Settling time	t_{delay}	-	3	5	µs
14	Reverence voltage high	V_{refh}	V_{DDA}-0.1V	V_{DDA}	V_{DDA}+0.1V	V

*) DRIVE bit = 1 is not recomended in this case.