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### Understanding [Embedded - DSP \(Digital Signal Processors\)](#)

[Embedded - DSP \(Digital Signal Processors\)](#) are specialized microprocessors designed to perform complex mathematical computations on digital signals in real-time. Unlike general-purpose processors, DSPs are optimized for high-speed numeric processing tasks, making them ideal for applications that require efficient and precise manipulation of digital data. These processors are fundamental in converting and processing signals in various forms, including audio, video, and communication signals, ensuring that data is accurately interpreted and utilized in embedded systems.

### Applications of [Embedded - DSP \(Digital Signal Processors\)](#)

#### Details

Product Status	Obsolete
Type	SC3850 Dual Core
Interface	Ethernet, I <sup>2</sup> C, PCI, RGMII, Serial RapidIO, SGMII, SPI, UART/USART
Clock Rate	1GHz
Non-Volatile Memory	ROM (96kB)
On-Chip RAM	576kB
Voltage - I/O	2.50V
Voltage - Core	1.00V
Operating Temperature	0°C ~ 105°C (TJ)
Mounting Type	Surface Mount
Package / Case	783-BBGA, FCBGA
Supplier Device Package	783-FCPBGA (29x29)
Purchase URL	<a href="https://www.e-xfl.com/product-detail/nxp-semiconductors/msc8152svt1000b">https://www.e-xfl.com/product-detail/nxp-semiconductors/msc8152svt1000b</a>

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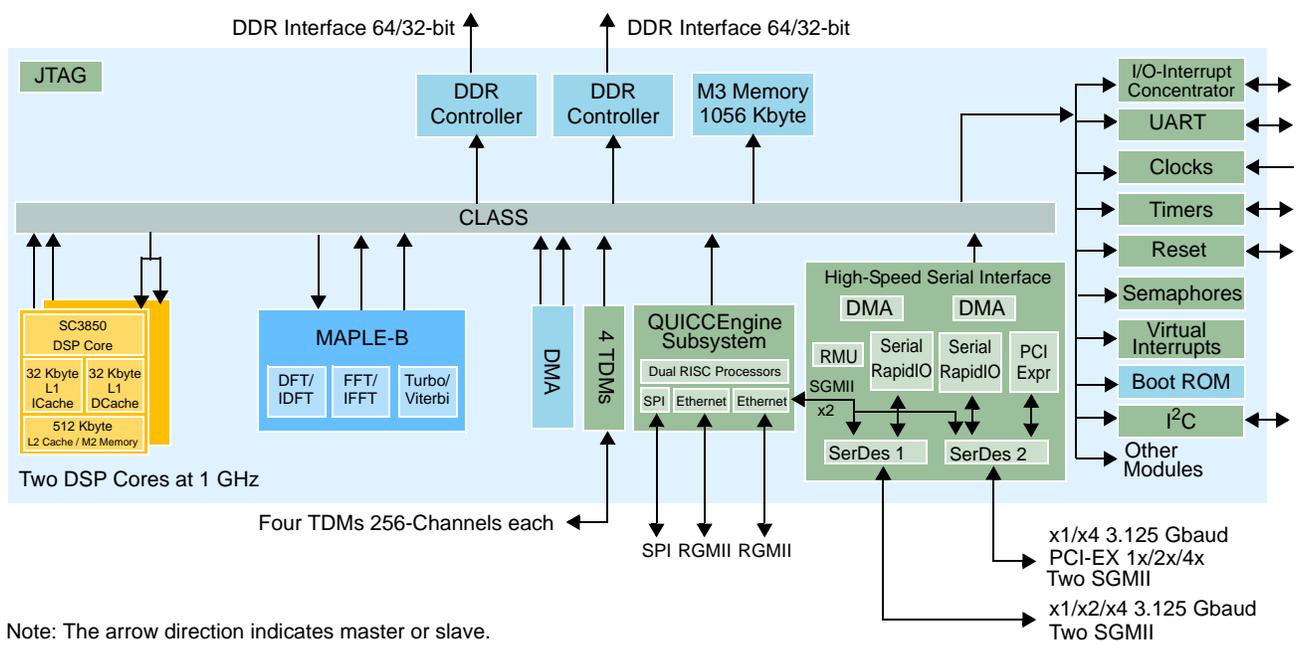


Figure 1. MSC8152 Block Diagram

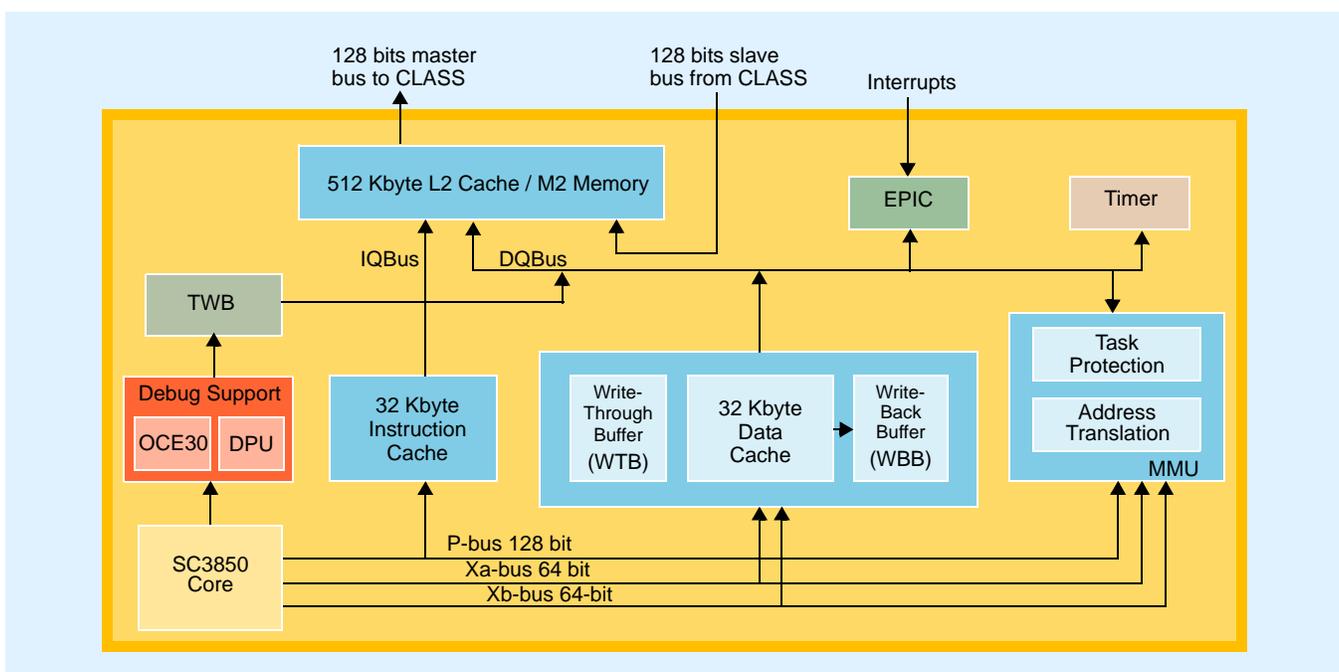


Figure 2. StarCore SC3850 DSP Subsystem Block Diagram

## 1.2 Signal List By Ball Location

Table 1 presents the signal list sorted by ball number. When designing a board, make sure that the power rail for each signal is appropriately considered. The specified power rail must be tied to the voltage level specified in this document if any of the related signal functions are used (active)

**Note:** The information in Table 1 and Table 2 distinguishes among three concepts. First, the power pins are the balls of the device package used to supply specific power levels for different device subsystems (as opposed to signals). Second, the power rails are the electrical lines on the board that transfer power from the voltage regulators to the device. They are indicated here as the reference power rails for signal lines; therefore, the actual power inputs are listed as N/A with regard to the power rails. Third, symbols used in these tables are the names for the voltage levels (absolute, recommended, and so on) and not the power supplies themselves.

**Table 1. Signal List by Ball Number**

Ball Number	Signal Name <sup>1,2</sup>	Pin Type <sup>10</sup>	Power Rail Name
A2	M2DQS3	I/O	GVDD2
A3	M2DQS3	I/O	GVDD2
A4	M2ECC0	I/O	GVDD2
A5	M2DQS8	I/O	GVDD2
A6	M2DQS8	I/O	GVDD2
A7	M2A5	O	GVDD2
A8	M2CK1	O	GVDD2
A9	M2CK1	O	GVDD2
A10	M2CS0	O	GVDD2
A11	M2BA0	O	GVDD2
A12	M2CAS	O	GVDD2
A13	M2DQ34	I/O	GVDD2
A14	M2DQS4	I/O	GVDD2
A15	M2DQS4	I/O	GVDD2
A16	M2DQ50	I/O	GVDD2
A17	M2DQS6	I/O	GVDD2
A18	M2DQS6	I/O	GVDD2
A19	M2DQ48	I/O	GVDD2
A20	M2DQ49	I/O	GVDD2
A21	VSS	Ground	N/A
A22	Reserved	NC	—
A23	SXPVDD1	Power	N/A
A24	SXPVSS1	Ground	N/A
A25	Reserved	NC	—
A26	Reserved	NC	—
A27	SXCVDD1	Power	N/A
A28	SXCVSS1	Ground	N/A
B1	M2DQ24	I/O	GVDD2
B2	GVDD2	Power	N/A
B3	M2DQ25	I/O	GVDD2
B4	VSS	Ground	N/A
B5	GVDD2	Power	N/A
B6	M2ECC1	I/O	GVDD2
B7	VSS	Ground	N/A
B8	GVDD2	Power	N/A

**Table 1. Signal List by Ball Number (continued)**

Ball Number	Signal Name <sup>1,2</sup>	Pin Type <sup>10</sup>	Power Rail Name
Y11	GVDD1	Power	N/A
Y12	VSS	Ground	N/A
Y13	GVDD1	Power	N/A
Y14	VSS	Ground	N/A
Y15	GVDD1	Power	N/A
Y16	VSS	Ground	N/A
Y17	GVDD1	Power	N/A
Y18	VSS	Ground	N/A
Y19	GVDD1	Power	N/A
Y20	VSS	Ground	N/A
Y21	NVDD	Power	N/A
Y22	GPIO20/SPI_SL <sup>5,8</sup>	I/O	NVDD
Y23	GPIO17/SPI_SCK <sup>5,8</sup>	I/O	NVDD
Y24	GPIO14/DRQ0/IRQ14/RC14 <sup>5,8</sup>	I/O	NVDD
Y25	GPIO12/IRQ12/RC12 <sup>5,8</sup>	I/O	NVDD
Y26	GPIO8/IRQ8/RC8 <sup>5,8</sup>	I/O	NVDD
Y27	NVDD	Power	N/A
Y28	VSS	Ground	N/A
AA1	GVDD1	Power	N/A
AA2	VSS	Ground	N/A
AA3	M1DQ18	I/O	GVDD1
AA4	GVDD1	Power	N/A
AA5	VSS	Ground	N/A
AA6	M1DQ20	I/O	GVDD1
AA7	GVDD1	Power	N/A
AA8	VSS	Ground	N/A
AA9	M1A15	O	GVDD1
AA10	M1CK2	O	GVDD1
AA11	M1MDIC0	I/O	GVDD1
AA12	M1VREF	I	GVDD1
AA13	M1MDIC1	I/O	GVDD1
AA14	M1DQ46	I/O	GVDD1
AA15	M1DQ47	I/O	GVDD1
AA16	M1DQ45	I/O	GVDD1
AA17	M1DQ41	I/O	GVDD1
AA18	M1DQ62	I/O	GVDD1
AA19	M1DQ63	I/O	GVDD1
AA20	M1DQ61	I/O	GVDD1
AA21	VSS	Ground	N/A
AA22	GPIO21 <sup>5,8</sup>	I/O	NVDD
AA23	GPIO18/SPI_MOSI <sup>5,8</sup>	I/O	NVDD
AA24	GPIO16/RC16 <sup>5,8</sup>	I/O	NVDD
AA25	GPIO4/DDN1/IRQ4/RC4 <sup>5,8</sup>	I/O	NVDD
AA26	GPIO9/IRQ9/RC9 <sup>5,8</sup>	I/O	NVDD
AA27	GPIO6/IRQ6/RC6 <sup>5,8</sup>	I/O	NVDD
AA28	GPIO1/IRQ1/RC1 <sup>5,8</sup>	I/O	NVDD

**Table 1. Signal List by Ball Number (continued)**

Ball Number	Signal Name <sup>1,2</sup>	Pin Type <sup>10</sup>	Power Rail Name
AB1	M1DQS2	I/O	GVDD1
AB2	M1DQS2	I/O	GVDD1
AB3	M1DQ19	I/O	GVDD1
AB4	M1DM2	O	GVDD1
AB5	M1DQ21	I/O	GVDD1
AB6	M1DQ22	I/O	GVDD1
AB7	M1CKE0	O	GVDD1
AB8	M1A11	O	GVDD1
AB9	M1A7	O	GVDD1
AB10	M1CK2	O	GVDD1
AB11	M1APAR_OUT	O	GVDD1
AB12	M1ODT1	O	GVDD1
AB13	M1APAR_IN	I	GVDD1
AB14	M1DQ43	I/O	GVDD1
AB15	M1DM5	O	GVDD1
AB16	M1DQ44	I/O	GVDD1
AB17	M1DQ40	I/O	GVDD1
AB18	M1DQ59	I/O	GVDD1
AB19	M1DM7	O	GVDD1
AB20	M1DQ60	I/O	GVDD1
AB21	VSS	Ground	N/A
AB22	GPIO31/I2C_SDA <sup>5,8</sup>	I/O	NVDD
AB23	GPIO27/TMR4/RCW_SRC0 <sup>5,8</sup>	I/O	NVDD
AB24	GPIO25/TMR2/RCW_SRC1 <sup>5,8</sup>	I/O	NVDD
AB25	GPIO24/TMR1/RCW_SRC2 <sup>5,8</sup>	I/O	NVDD
AB26	GPIO10/IRQ10/RC10 <sup>5,8</sup>	I/O	NVDD
AB27	GPIO5/IRQ5/RC5 <sup>5,8</sup>	I/O	NVDD
AB28	GPIO0/IRQ0/RC0 <sup>5,8</sup>	I/O	NVDD
AC1	VSS	Ground	N/A
AC2	GVDD1	Power	N/A
AC3	M1DQ16	I/O	GVDD1
AC4	VSS	Ground	N/A
AC5	GVDD1	Power	N/A
AC6	M1DQ17	I/O	GVDD1
AC7	VSS	Ground	N/A
AC8	GVDD1	Power	N/A
AC9	M1BA2	O	GVDD1
AC10	VSS	Ground	N/A
AC11	GVDD1	Power	N/A
AC12	M1A4	O	GVDD1
AC13	VSS	Ground	N/A
AC14	GVDD1	Power	N/A
AC15	M1DQ42	I/O	GVDD1
AC16	VSS	Ground	N/A
AC17	GVDD1	Power	N/A
AC18	M1DQ58	I/O	GVDD1

**Table 1. Signal List by Ball Number (continued)**

Ball Number	Signal Name <sup>1,2</sup>	Pin Type <sup>10</sup>	Power Rail Name
AE9	M1A8	O	GVDD1
AE10	GVDD1	Power	N/A
AE11	VSS	Ground	N/A
AE12	M1A0	O	GVDD1
AE13	GVDD1	Power	N/A
AE14	VSS	Ground	N/A
AE15	M1DQ39	I/O	GVDD1
AE16	GVDD1	Power	N/A
AE17	VSS	Ground	N/A
AE18	M1DQ54	I/O	GVDD1
AE19	GVDD1	Power	N/A
AE20	VSS	Ground	N/A
AE21	GPIO29/UART_TXD <sup>5,8</sup>	I/O	NVDD
AE22	TDM1TCK/GE2_RX_CLK <sup>3</sup>	I	NVDD
AE23	TDM1RSN/GE2_RX_CTL <sup>3</sup>	I/O	NVDD
AE24	VSS	Ground	N/A
AE25	TDM3RCK/GE1_GTX_CLK <sup>3</sup>	I/O	NVDD
AE26	TDM3TSN/GE1_RX_CLK <sup>3</sup>	I/O	NVDD
AE27	TDM2RSN/GE1_TD2 <sup>3</sup>	I/O	NVDD
AE28	TDM2RDT/GE1_TD1 <sup>3</sup>	I/O	NVDD
AF1	M1DQ28	I/O	GVDD1
AF2	M1DM3	O	GVDD1
AF3	M1DQ26	I/O	GVDD1
AF4	M1ECC4	I/O	GVDD1
AF5	M1DM8	O	GVDD1
AF6	M1ECC2	I/O	GVDD1
AF7	M1CKE1	O	GVDD1
AF8	M1CK0	O	GVDD1
AF9	$\overline{\text{M1CK0}}$	O	GVDD1
AF10	M1BA1	O	GVDD1
AF11	M1A1	O	GVDD1
AF12	M1WE	O	GVDD1
AF13	M1DQ37	I/O	GVDD1
AF14	M1DM4	O	GVDD1
AF15	M1DQ36	I/O	GVDD1
AF16	M1DQ32	I/O	GVDD1
AF17	M1DQ55	I/O	GVDD1
AF18	M1DM6	O	GVDD1
AF19	M1DQ53	I/O	GVDD1
AF20	M1DQ52	I/O	GVDD1
AF21	GPIO28/UART_RXD <sup>5,8</sup>	I/O	NVDD
AF22	TDM0RSN/GE2_TD2 <sup>3</sup>	I/O	NVDD
AF23	TDM0TDT/GE2_TD3 <sup>3</sup>	I/O	NVDD
AF24	NVDD	Power	N/A
AF25	TDM2TSN/GE1_TX_CTL <sup>3</sup>	I/O	NVDD
AF26	GE1_RX_CTL	I	NVDD

## 2 Electrical Characteristics

This document contains detailed information on power considerations, DC/AC electrical characteristics, and AC timing specifications. For additional information, see the *MSC8152 Reference Manual*.

### 2.1 Maximum Ratings

In calculating timing requirements, adding a maximum value of one specification to a minimum value of another specification does not yield a reasonable sum. A maximum specification is calculated using a worst case variation of process parameter values in one direction. The minimum specification is calculated using the worst case for the same parameters in the opposite direction. Therefore, a “maximum” value for a specification never occurs in the same device with a “minimum” value for another specification; adding a maximum to a minimum represents a condition that can never exist.

Table 2 describes the maximum electrical ratings for the MSC8152.

**Table 2. Absolute Maximum Ratings**

Rating	Power Rail Name	Symbol	Value	Unit
Core supply voltage • Cores 0–1	VDD	$V_{DD}$	–0.3 to 1.1	V
PLL supply voltage <sup>3</sup>		$V_{DDPLL0}$	–0.3 to 1.1	V
		$V_{DDPLL1}$	–0.3 to 1.1	V
		$V_{DDPLL2}$	–0.3 to 1.1	V
M3 memory supply voltage	M3VDD	$V_{DDM3}$	–0.3 to 1.1	V
MAPLE-B supply voltage	MVDD	$V_{DDM}$	–0.3 to 1.1	V
DDR memory supply voltage • DDR2 mode • DDR3 mode	GVDD1, GVDD2	$V_{DDDDR}$	–0.3 to 1.98 –0.3 to 1.65	V V
DDR reference voltage	MVREF	$MV_{REF}$	–0.3 to $0.51 \times V_{DDDDR}$	V
Input DDR voltage		$V_{INDDR}$	–0.3 to $V_{DDDDR} + 0.3$	V
I/O voltage excluding DDR and RapidIO lines	NVDD, QVDD	$V_{DDIO}$	–0.3 to 2.625	V
Input I/O voltage		$V_{INIO}$	–0.3 to $V_{DDIO} + 0.3$	V
RapidIO pad voltage	SXPVDD1, SXPVDD2	$V_{DSSXP}$	–0.3 to 1.26	V
Rapid I/O core voltage	SXCVDD1, SXCVDD2	$V_{DSSXC}$	–0.3 to 1.21	V
Rapid I/O PLL voltage <sup>3</sup>		$V_{DDRIOPLL}$	–0.3 to 1.21	V
Input RapidIO I/O voltage		$V_{INRIO}$	–0.3 to $V_{DSSXC} + 0.3$	V
Operating temperature		$T_J$	–40 to 105	°C
Storage temperature range		$T_{STG}$	–55 to +150	°C
<b>Notes:</b>	<ol style="list-style-type: none"> <li>1. Functional operating conditions are given in <a href="#">Table 3</a>.</li> <li>2. Absolute maximum ratings are stress ratings only, and functional operation at the maximum is not guaranteed. Stress beyond the listed limits may affect device reliability or cause permanent damage.</li> <li>3. PLL supply voltage is specified at input of the filter and not at pin of the MSC8152 (see <a href="#">Figure 37</a> and <a href="#">Figure 38</a>)</li> </ol>			

## 2.5 DC Electrical Characteristics

This section describes the DC electrical characteristics for the MSC8152.

### 2.5.1 DDR SDRAM DC Electrical Characteristics

This section describes the DC electrical specifications for the DDR SDRAM interface of the MSC8152.

**Note:** DDR2 SDRAM uses  $V_{DDDDR}(typ) = 1.8\text{ V}$  and DDR3 SDRAM uses  $V_{DDDDR}(typ) = 1.5\text{ V}$ .

#### 2.5.1.1 DDR2 (1.8 V) SDRAM DC Electrical Characteristics

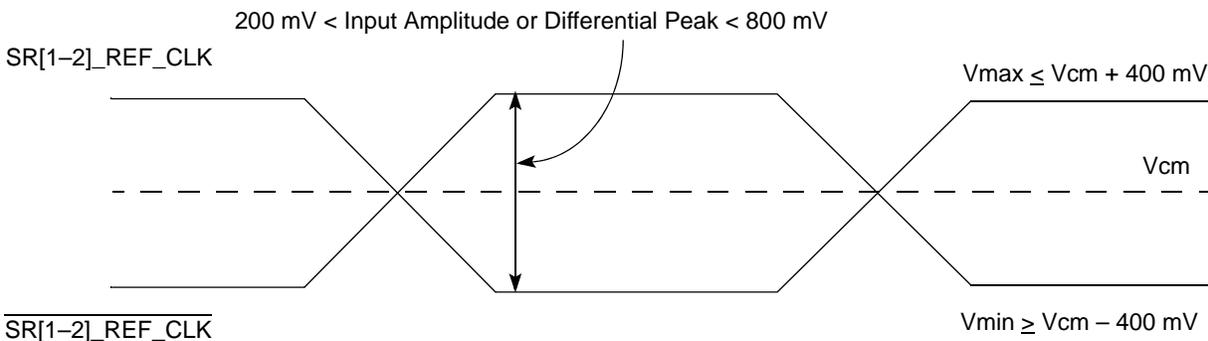
**Table 6** provides the recommended operating conditions for the DDR SDRAM controller when interfacing to DDR2 SDRAM.

**Note:** At recommended operating conditions (see **Table 3**) with  $V_{DDDDR} = 1.8\text{ V}$ .

**Table 6. DDR2 SDRAM Interface DC Electrical Characteristics**

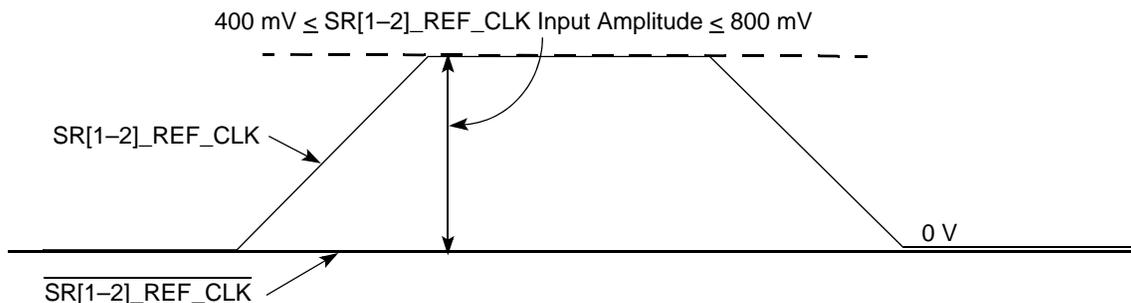
Parameter/Condition	Symbol	Min	Max	Unit	Notes
I/O reference voltage	$MV_{REF}$	$0.49 \times V_{DDDDR}$	$0.51 \times V_{DDDDR}$	V	2, 3, 4
Input high voltage	$V_{IH}$	$MV_{REF} + 0.125$	$V_{DDDDR} + 0.3$	V	5
Input low voltage	$V_{IL}$	-0.3	$MV_{REF} - 0.125$	V	5
I/O leakage current	$I_{OZ}$	-50	50	$\mu\text{A}$	6
Output high current ( $V_{OUT} (VOH) = 1.37\text{ V}$ )	$I_{OH}$	-13.4	—	mA	7
Output low current ( $V_{OUT} (VOL) = 0.33\text{ V}$ )	$I_{OL}$	13.4	—	mA	7
<b>Notes:</b> <ol style="list-style-type: none"> <li><math>V_{DDDDR}</math> is expected to be within 50 mV of the DRAM <math>V_{DD}</math> supply voltage at all times. The DRAM and memory controller can use the same or different sources.</li> <li><math>MV_{REF}</math> is expected to be equal to <math>0.5 \times V_{DDDDR}</math> and to track <math>V_{DDDDR}</math> DC variations as measured at the receiver. Peak-to-peak noise on <math>MV_{REF}</math> may not exceed <math>\pm 2\%</math> of the DC value.</li> <li><math>V_{TT}</math> is not applied directly to the device. It is the supply to which far end signal termination is made and is expected to be equal to <math>MV_{REF}</math> with a minimum value of <math>MV_{REF} - 0.4</math> and a maximum value of <math>MV_{REF} + 0.04\text{ V}</math>. <math>V_{TT}</math> should track variations in the DC-level of <math>MV_{REF}</math>.</li> <li>The voltage regulator for <math>MV_{REF}</math> must be able to supply up to 300 <math>\mu\text{A}</math>.</li> <li>Input capacitance load for DQ, DQS, and DQS signals are available in the IBIS models.</li> <li>Output leakage is measured with all outputs are disabled, <math>0\text{ V} \leq V_{OUT} \leq V_{DDDDR}</math>.</li> <li>Refer to the IBIS model for the complete output IV curve characteristics.</li> </ol>					

- For an external AC-coupled connection, there is no common mode voltage requirement for the clock driver. Because the external AC-coupling capacitor blocks the DC-level, the clock driver and the SerDes reference clock receiver operate in different command mode voltages. The SerDes reference clock receiver in this connection scheme has its common mode voltage set to  $GND_{SXC}$ . Each signal wire of the differential inputs is allowed to swing below and above the command mode voltage  $GND_{SXC}$ . **Figure 8** shows the SerDes reference clock input requirement for AC-coupled connection scheme.



**Figure 8. Differential Reference Clock Input DC Requirements (External AC-Coupled)**

- Single-Ended Mode
  - The reference clock can also be single-ended. The  $SR[1-2]_{REF\_CLK}$  input amplitude (single-ended swing) must be between 400 mV and 800 mV peak-peak (from  $V_{MIN}$  to  $V_{MAX}$ ) with  $\overline{SR[1-2]_{REF\_CLK}}$  either left unconnected or tied to ground.
  - The  $SR[1-2]_{REF\_CLK}$  input average voltage must be between 200 and 400 mV. **Figure 9** shows the SerDes reference clock input requirement for single-ended signalling mode.
  - To meet the input amplitude requirement, the reference clock inputs may need to be DC- or AC-coupled externally. For the best noise performance, the reference of the clock could be DC- or AC-coupled into the unused phase ( $SR[1-2]_{REF\_CLK}$ ) through the same source impedance as the clock input ( $SR[1-2]_{REF\_CLK}$ ) in use.



**Figure 9. Single-Ended Reference Clock Input DC Requirements**

### 2.5.3.2 DC-Level Requirements for PCI Express Configurations

The DC-level requirements for PCI Express implementations have separate requirements for the Tx and Rx lines. The MSC8152 supports a 2.5 Gbps PCI Express interface defined by the *PCI Express Base Specification, Revision 1.0a*. The transmitter specifications are defined in [Table 11](#) and the receiver specifications are defined in [Table 12](#).

**Note:** Specifications are valid at the recommended operating conditions listed in Table 3.

**Table 11. PCI Express (2.5 Gbps) Differential Transmitter (Tx) Output DC Specifications**

Parameter	Symbol	Min	Typical	Max	Units	Notes
Differential peak-to-peak output voltage	$V_{TX-DIFFP-P}$	800	1000	1200	mV	1
De-emphasized differential output voltage (ratio)	$V_{TX-DE-RATIO}$	3.0	3.5	4.0	dB	2
DC differential Tx impedance	$Z_{TX-DIFF-DC}$	80	100	120	$\Omega$	3
Transmitter DC impedance	$Z_{TX-DC}$	40	50	60	$\Omega$	4
<b>Notes:</b> <ol style="list-style-type: none"> <li><math>V_{TX-DIFFP-P} = 2 \times  V_{TX-D+} - V_{TX-D-} </math> Measured at the package pins with a test load of 50 <math>\Omega</math> to GND on each pin.</li> <li>Ratio of the <math>V_{TX-DIFFP-P}</math> of the second and following bits after a transition divided by the <math>V_{TX-DIFFP-P}</math> of the first bit after a transition. Measured at the package pins with a test load of 50 <math>\Omega</math> to GND on each pin.</li> <li>Tx DC differential mode low impedance</li> <li>Required Tx D+ as well as D- DC Impedance during all states</li> </ol>						

**Table 12. PCI Express (2.5 Gbps) Differential Receiver (Rx) Input DC Specifications**

Parameter	Symbol	Min	Typical	Max	Units	Notes
Differential input peak-to-peak voltage	$V_{RX-DIFFP-P}$	120	1000	1200	mV	1
DC differential Input Impedance	$Z_{RX-DIFF-DC}$	80	100	120	$\Omega$	2
DC input impedance	$Z_{RX-DC}$	40	50	60	$\Omega$	3
Powered down DC input impedance	$Z_{RX-HIGH-IMP-DC}$	50	—	—	k $\Omega$	4
Electrical idle detect threshold	$V_{RX-IDLE-DET-DIFFP-P}$	65	—	175	mV	5
<b>Notes:</b> <ol style="list-style-type: none"> <li><math>V_{RX-DIFFP-P} = 2 \times  V_{RX-D+} - V_{RX-D-} </math> Measured at the package pins with a test load of 50 <math>\Omega</math> to GND on each pin.</li> <li>Rx DC differential mode impedance. Impedance during all LTSSM states. When transitioning from a fundamental reset to detect (the initial state of the LTSSM), there is a 5 ms transition time before the receiver termination values must be met on all unconfigured lanes of a port.</li> <li>Required Rx D+ as well as D- DC Impedance (50 <math>\pm</math>20% tolerance). Measured at the package pins with a test load of 50 <math>\Omega</math> to GND on each pin. Impedance during all LTSSM states. When transitioning from a fundamental reset to detect (the initial state of the LTSSM), there is a 5 ms transition time before the receiver termination values must be met on all unconfigured lanes of a port.</li> <li>Required Rx D+ as well as D- DC Impedance when the receiver terminations do not have power. The Rx DC common mode impedance that exists when no power is present or fundamental reset is asserted. This helps ensure that the receiver detect circuit does not falsely assume a receiver is powered on when it is not. This term must be measured at 300 mV above the Rx ground.</li> <li><math>V_{RX-IDLE-DET-DIFFP-P} = 2 \times  V_{RX-D+} - V_{RX-D-} </math>. Measured at the package pins of the receiver</li> </ol>						

### 2.5.3.3 DC-Level Requirements for Serial RapidIO Configurations

This sections provided various DC-level requirements for Serial RapidIO Configurations.

**Note:** Specifications are valid at the recommended operating conditions listed in Table 3.

**Table 13. Serial RapidIO Transmitter DC Specifications**

Parameter	Symbol	Min	Typical	Max	Units	Notes
Output voltage	$V_O$	-0.40	—	2.30	V	1
Long run differential output voltage	$V_{DIFFPP}$	800	—	1600	mVp-p	—
Short run differential output voltage	$V_{DIFFPP}$	500	—	1000	mVp-p	—
<b>Note:</b> Voltage relative to COMMON of either signal comprising a differential pair.						

**Table 14. Serial RapidIO Receiver DC Specifications**

Parameter	Symbol	Min	Typical	Max	Units	Notes
Differential input voltage	$V_{IN}$	200	—	1600	mVp-p	1
<b>Notes:</b> 1. Measured at receiver.						

### 2.5.3.4 DC-Level Requirements for SGMII Configurations

**Note:** Specifications are valid at the recommended operating conditions listed in [Table 3](#)

[Table 15](#) describes the SGMII SerDes transmitter AC-coupled DC electrical characteristics. Transmitter DC characteristics are measured at the transmitter outputs ( $\overline{SR[1-2]_{TX}[n]}$  and  $\overline{SR[1-2]_{TX}[n]}$ ) as shown in [Figure 10](#).

**Table 15. SGMII DC Transmitter Electrical Characteristics**

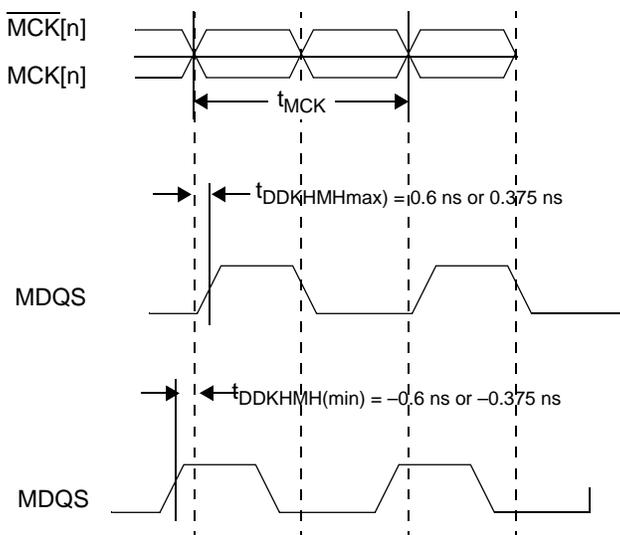
Parameter	Symbol	Min	Typ	Max	Unit	Notes
Output high voltage	$V_{OH}$	—	—	$XV_{DD\_SRDS-Typ}/2 +  V_{OD} _{max}/2$	mV	1
Output low voltage	$V_{OL}$	$XV_{DD\_SRDS-Typ}/2 -  V_{OD} _{max}/2$	—	—	mV	1
Output differential voltage ( $XV_{DD-Typ}$ at 1.0 V)	$ V_{OD} $	323	500	725	mV	2,3,4
		296	459	665		2,3,5
		269	417	604		2,3,6
		243	376	545		2,3,7
		215	333	483		2,3,8
		189	292	424		2,3,9
		162	250	362		2,3,10
Output impedance (single-ended)	$R_O$	40	50	60	$\Omega$	—
<b>Notes:</b> <ol style="list-style-type: none"> <li>This does not align to DC-coupled SGMII. <math>XV_{DD\_SRDS2-Typ} = 1.1</math> V.</li> <li>The <math> V_{OD} </math> value shown in the table assumes full multibyte by setting <code>srd_smit_lvl</code> as 000 and the following transmit equalization setting in the <code>XMITEQAB</code> (for lanes A and B) or <code>XMITEQEF</code> (for lanes E and F) bit field of Control Register: <ul style="list-style-type: none"> <li>The MSB (bit 0) of the above bit field is set to zero (selecting the full <math>V_{DD-DIFF-p-p}</math> amplitude which is power up default);</li> <li>The LSB (bit [1-3]) of the above bit field is set based on the equalization settings listed in notes 4 through 10.</li> </ul> </li> <li>The <math> V_{OD} </math> value shown in the Typ column is based on the condition of <math>XV_{DD\_SRDS2-Typ} = 1.0</math> V, no common mode offset variation (<math>V_{OS} = 500</math> mV), SerDes transmitter is terminated with 100-<math>\Omega</math> differential load between</li> <li>Equalization setting: 1.0x: 0000.</li> <li>Equalization setting: 1.09x: 1000.</li> <li>Equalization setting: 1.2x: 0100.</li> <li>Equalization setting: 1.33x: 1100.</li> <li>Equalization setting: 1.5x: 0010.</li> <li>Equalization setting: 1.71x: 1010.</li> <li>Equalization setting: 2.0x: 0110.</li> <li><math> V_{OD}  =  V_{SR[1-2]_{TX}[n]} - \overline{V_{SR[1-2]_{TX}[n]}} </math>. <math> V_{OD} </math> is also referred to as output differential peak voltage. <math>V_{TX-DIFF-p-p} = 2 *  V_{OD} </math>.</li> </ol>						

**Table 21. DDR SDRAM Output AC Timing Specifications (continued)**

Parameter	Symbol <sup>1</sup>	Min	Max	Unit	Notes
<b>Notes:</b> <ol style="list-style-type: none"> <li>The symbols used for timing specifications follow the pattern of <math>t_{(\text{first two letters of functional block})(\text{signal})(\text{state})(\text{reference})(\text{state})}</math> for inputs and <math>t_{(\text{first two letters of functional block})(\text{reference})(\text{state})(\text{signal})(\text{state})}</math> for outputs. Output hold time can be read as DDR timing (DD) from the rising or falling edge of the reference clock (KH or KL) until the output went invalid (AX or DX). For example, <math>t_{\text{DDKHAS}}</math> symbolizes DDR timing (DD) for the time <math>t_{\text{MCK}}</math> memory clock reference (K) goes from the high (H) state until outputs (A) are setup (S) or output valid time. Also, <math>t_{\text{DDKLDX}}</math> symbolizes DDR timing (DD) for the time <math>t_{\text{MCK}}</math> memory clock reference (K) goes low (L) until data outputs (D) are invalid (X) or data output hold time.</li> <li>All MCK/MCK referenced measurements are made from the crossing of the two signals.</li> <li>ADDR/CMD includes all DDR SDRAM output signals except MCK/MCK, MCS, and MDQ/MECC/MDM/MDQS.</li> <li>Note that <math>t_{\text{DDKHMH}}</math> follows the symbol conventions described in note 1. For example, <math>t_{\text{DDKHMH}}</math> describes the DDR timing (DD) from the rising edge of the MCK(n) clock (KH) until the MDQS signal is valid (MH). <math>t_{\text{DDKHMH}}</math> can be modified through control of the DQSS override bits in the TIMING_CFG_2 register. This will typically be set to the same delay as the clock adjust in the CLK_CNTL register. The timing parameters listed in the table assume that these two parameters have been set to the same adjustment value. See the <i>MSC8152 Reference Manual</i> for a description and understanding of the timing modifications enabled by use of these bits.</li> <li>Determined by maximum possible skew between a data strobe (MDQS) and any corresponding bit of data (MDQ), ECC (MECC), or data mask (MDM). The data strobe should be centered inside of the data eye at the pins of the MSC8152.</li> <li>At recommended operating conditions with <math>V_{\text{DDDDR}}</math> (1.5 V or 1.8 V) <math>\pm</math> 5%.</li> </ol>					

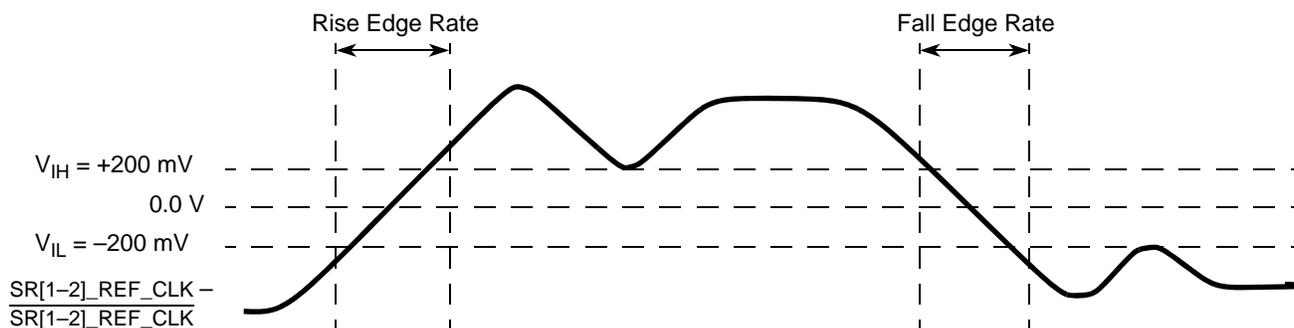
**Note:** For the ADDR/CMD setup and hold specifications in Table 21, it is assumed that the clock control register is set to adjust the memory clocks by  $\frac{1}{2}$  applied cycle.

Figure 12 shows the DDR SDRAM output timing for the MCK to MDQS skew measurement ( $t_{\text{DDKHMH}}$ ).

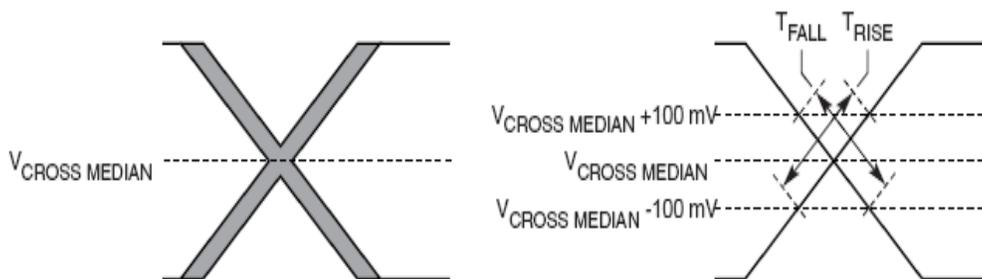

**Figure 12. MCK to MDQS Timing**

**Table 24. SR[1–2]\_REF\_CLK and  $\overline{\text{SR[1–2]_REF\_CLK}}$  Input Clock Requirements (continued)**

Parameter	Symbol	Min	Typical	Max	Units	Notes
<p><b>Notes:</b></p> <ol style="list-style-type: none"> <li>1. Caution: Only 100 and 125 have been tested. Other values will not work correctly with the rest of the system.</li> <li>2. Limits from PCI Express CEM Rev 1.0a</li> <li>3. Measured from <math>-200\text{ mV}</math> to <math>+200\text{ mV}</math> on the differential waveform (derived from <math>\text{SR[1–2]_REF\_CLK}</math> minus <math>\overline{\text{SR[1–2]_REF\_CLK}}</math>). The signal must be monotonic through the measurement region for rise and fall time. The <math>400\text{ mV}</math> measurement window is centered on the differential zero crossing. See <a href="#">Figure 16</a>.</li> <li>4. Measurement taken from differential waveform</li> <li>5. Measurement taken from single-ended waveform</li> <li>6. Matching applies to rising edge for <math>\text{SR[1–2]_REF\_CLK}</math> and falling edge rate for <math>\overline{\text{SR[1–2]_REF\_CLK}}</math>. It is measured using a <math>200\text{ mV}</math> window centered on the median cross point where <math>\text{SR[1–2]_REF\_CLK}</math> rising meets <math>\overline{\text{SR[1–2]_REF\_CLK}}</math> falling. The median cross point is used to calculate the voltage thresholds that the oscilloscope uses for the edge rate calculations. The rise edge rate of <math>\text{SR[1–2]_REF\_CLK}</math> should be compared to the fall edge rate of <math>\overline{\text{SR[1–2]_REF\_CLK}}</math>; the maximum allowed difference should not exceed 20% of the slowest edge rate. See <a href="#">Figure 17</a>.</li> </ol>						



**Figure 16. Differential Measurement Points for Rise and Fall Time**



**Figure 17. Single-Ended Measurement Points for Rise and Fall Time Matching**

### 2.6.2.3 Serial RapidIO AC Timing Specifications

**Note:** Specifications are valid at the recommended operating conditions listed in [Table 3](#).

[Table 27](#) defines the transmitter AC specifications for the Serial RapidIO interface. The AC timing specifications do not include REF\_CLK jitter.

**Table 27. Serial RapidIO Transmitter AC Timing Specifications**

Characteristic	Symbol	Min	Typical	Max	Unit
Deterministic Jitter	$J_D$	—	—	0.17	UI p-p
Total Jitter	$J_T$	—	—	0.35	UI p-p
Unit Interval: 1.25 GBaud	UI	800 – 100ppm	800	800 + 100ppm	ps
Unit Interval: 2.5 GBaud	UI	400 – 100ppm	400	400 + 100ppm	ps
Unit Interval: 3.125 GBaud	UI	320 – 100ppm	320	320 + 100ppm	ps

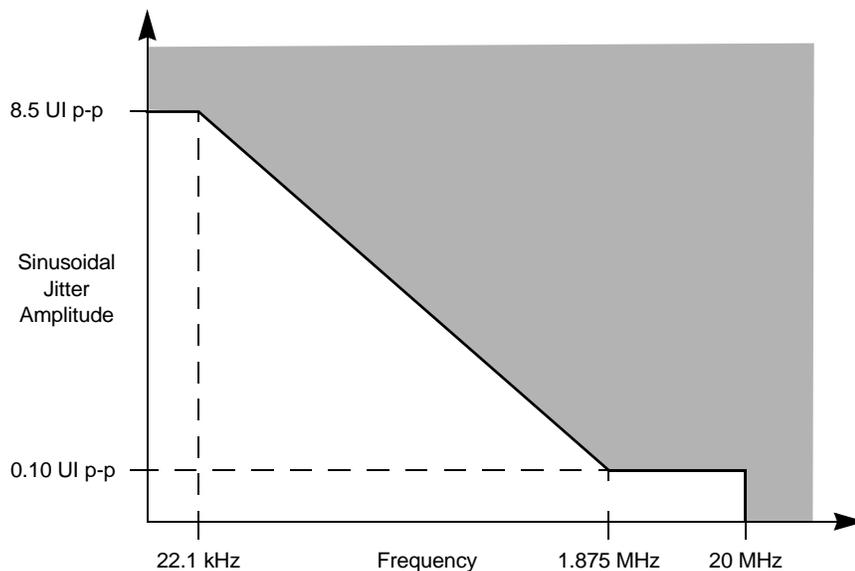
[Table 28](#) defines the Receiver AC specifications for the Serial RapidIO interface. The AC timing specifications do not include REF\_CLK jitter.

**Table 28. Serial RapidIO Receiver AC Timing Specifications**

Characteristic	Symbol	Min	Typical	Max	Unit	Notes
Deterministic Jitter Tolerance	$J_D$	0.37	—	—	UI p-p	1
Combined Deterministic and Random Jitter Tolerance	$J_{DR}$	0.55	—	—	UI p-p	1
Total Jitter Tolerance	$J_T$	0.65	—	—	UI p-p	1, 2
Bit Error Rate	BER	—	—	$10^{-12}$	—	—
Unit Interval: 1.25 GBaud	UI	800 – 100ppm	800	800 + 100ppm	ps	—
Unit Interval: 2.5 GBaud	UI	400 – 100ppm	400	400 + 100ppm	ps	—
Unit Interval: 3.125 GBaud	UI	320 – 100ppm	320	320 + 100ppm	ps	—

**Notes:**

1. Measured at receiver.
2. Total jitter is composed of three components, deterministic jitter, random jitter, and single frequency sinusoidal jitter. The sinusoidal jitter may have any amplitude and frequency in the unshaded region of [Figure 18](#). The sinusoidal jitter component is included to ensure margin for low frequency jitter, wander, noise, crosstalk, and other variable system effects.



**Figure 18. Single Frequency Sinusoidal Jitter Limits**

## 2.6.7 Asynchronous Signal Timing

Table 35 lists the asynchronous signal timing specifications.

**Table 37. Signal Timing**

Characteristics	Symbol	Type	Min
Input	$t_{IN}$	Asynchronous	One CLKIN cycle
Output	$t_{OUT}$	Asynchronous	Application dependent

**Note:** Input value relevant for  $\overline{EE0}$ ,  $\overline{IRQ}[15-0]$ , and  $\overline{NMI}$  only.

The following interfaces use the specified asynchronous signals:

- *GPIO*. Signals GPIO[31–0], when used as GPIO signals, that is, when the alternate multiplexed special functions are not selected.

**Note:** When used as a general purpose input (GPI), the input signal should be driven until it is acknowledged by the MSC8152 device, that is, when the expected input value is read from the GPIO data register.

- *EE port*. Signals EE0, EE1.
- *Boot function*. Signal STOP\_BS.
- *I<sup>2</sup>C interface*. Signals I2C\_SCL and I2C\_SDA.
- *Interrupt inputs*. Signals  $\overline{IRQ}[15-0]$  and  $\overline{NMI}$ .
- *Interrupt outputs*. Signals  $\overline{INT\_OUT}$  and  $\overline{NMI\_OUT}$  (minimum pulse width is 32 ns).

## 2.6.8 JTAG Signals

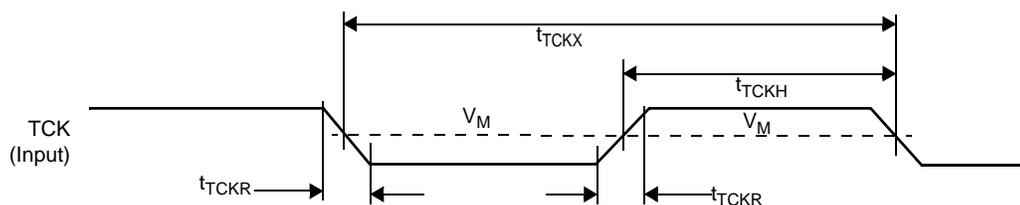
Table 38 lists the JTAG timing specifications shown in Figure 29 through Figure 32.

**Table 38. JTAG Timing**

Characteristics	Symbol	All frequencies		Unit
		Min	Max	
TCK cycle time	$t_{TCKX}$	36.0	—	ns
TCK clock high phase measured at $V_M = V_{DDIO}/2$	$t_{TCKH}$	15.0	—	ns
Boundary scan input data setup time	$t_{BSVKH}$	0.0	—	ns
Boundary scan input data hold time	$t_{BSXKH}$	15.0	—	ns
TCK fall to output data valid	$t_{TCKHOV}$	—	20.0	ns
TCK fall to output high impedance	$t_{TCKHOZ}$	—	24.0	ns
TMS, TDI data setup time	$t_{TDIVKH}$	0.0	—	ns
TMS, TDI data hold time	$t_{TDIXKH}$	5.0	—	ns
TCK fall to TDO data valid	$t_{TDOHOV}$	—	10.0	ns
TCK fall to TDO high impedance	$t_{TDOHOZ}$	—	12.0	ns
TRST assert time	$t_{TRST}$	100.0	—	ns

**Note:** All timings apply to OnCE module data transfers as well as any other transfers via the JTAG port.

Figure 29 shows the test clock input timing diagram



**Figure 29. Test Clock Input Timing**

### 3 Hardware Design Considerations

The following sections discuss areas to consider when the MSC8152 device is designed into a system.

#### 3.1 Power Supply Ramp-Up Sequence

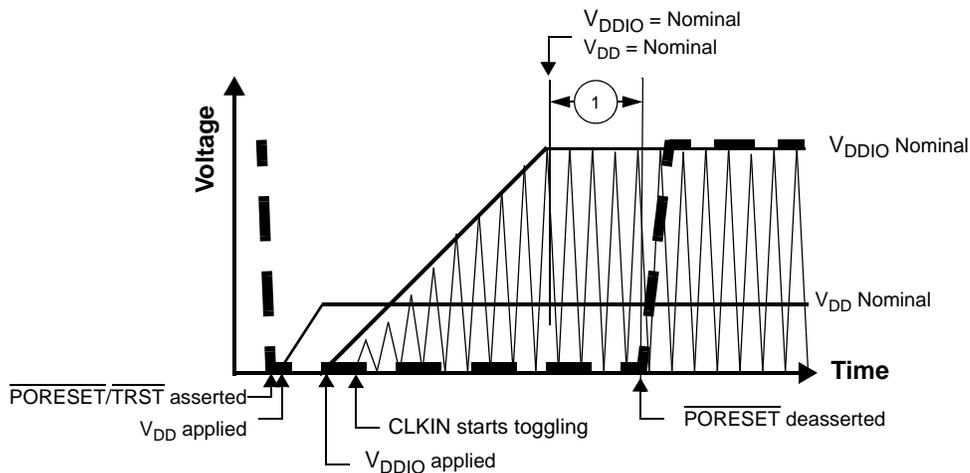
The following subsections describe the required device initialization sequence.

##### 3.1.1 Clock, Reset, and Supply Coordination

Starting the device requires coordination between several inputs including: clock, reset, and power supplies. Follow this guidelines when starting up an MSC8152 device:

- $\overline{\text{PORESET}}$  and  $\overline{\text{TRST}}$  must be asserted externally for the duration of the supply ramp-up, using the  $V_{\text{DDIO}}$  supply.  $\overline{\text{TRST}}$  deassertion does not have to be synchronized with  $\overline{\text{PORESET}}$  deassertion. However,  $\overline{\text{TRST}}$  must be deasserted before normal operation begins to ensure correct functionality of the device.
- CLKIN should toggle at least 32 cycles before  $\overline{\text{PORESET}}$  deassertion to guarantee correct device operation. The 32 cycles should only be counted from the time after  $V_{\text{DDIO}}$  reaches its nominal value (see timing 1 in Figure 33).
- CLKIN should either be stable low during ramp-up of  $V_{\text{DDIO}}$  supply (and start its swings after ramp-up) or should swing within  $V_{\text{DDIO}}$  range during  $V_{\text{DDIO}}$  ramp-up, so its amplitude grows as  $V_{\text{DDIO}}$  grows during ramp-up.

Figure 33 shows a sequence in which  $V_{\text{DDIO}}$  ramps-up after  $V_{\text{DD}}$  and CLKIN begins to toggle with the raise of  $V_{\text{DDIO}}$  supply.



**Figure 33. Supply Ramp-Up Sequence with  $V_{\text{DD}}$  Ramping Before  $V_{\text{DDIO}}$  and CLKIN Starting With  $V_{\text{DDIO}}$**

**Note:** For details on power-on reset flow and duration, see the *Reset* chapter in the *MSC8152 Reference Manual*.

## 3.2 PLL Power Supply Design Considerations

Each global PLL power supply must have an external RC filter for the PLL<sub>n</sub>\_AVDD input (see Figure 37) in which the following components are defined as listed:

- $R = 5\ \Omega \pm 5\%$
- $C1 = 10\ \mu\text{F} \pm 10\%$ , 0603, X5R, with  $\text{ESL} \leq 0.5\ \text{nH}$ , low ESL Surface Mount Capacitor.
- $C2 = 1.0\ \mu\text{F} \pm 10\%$ , 0402, X5R, with  $\text{ESL} \leq 0.5\ \text{nH}$ , low ESL Surface Mount Capacitor.

**Note:** A higher capacitance value for C2 may be used to improve the filter as long as the other C2 parameters do not change.

All three PLLs can connect to a single supply voltage source (such as a voltage regulator) as long as the external RC filter is applied to each PLL separately. For optimal noise filtering, place the circuit as close as possible to its PLL<sub>n</sub>\_AVDD inputs.

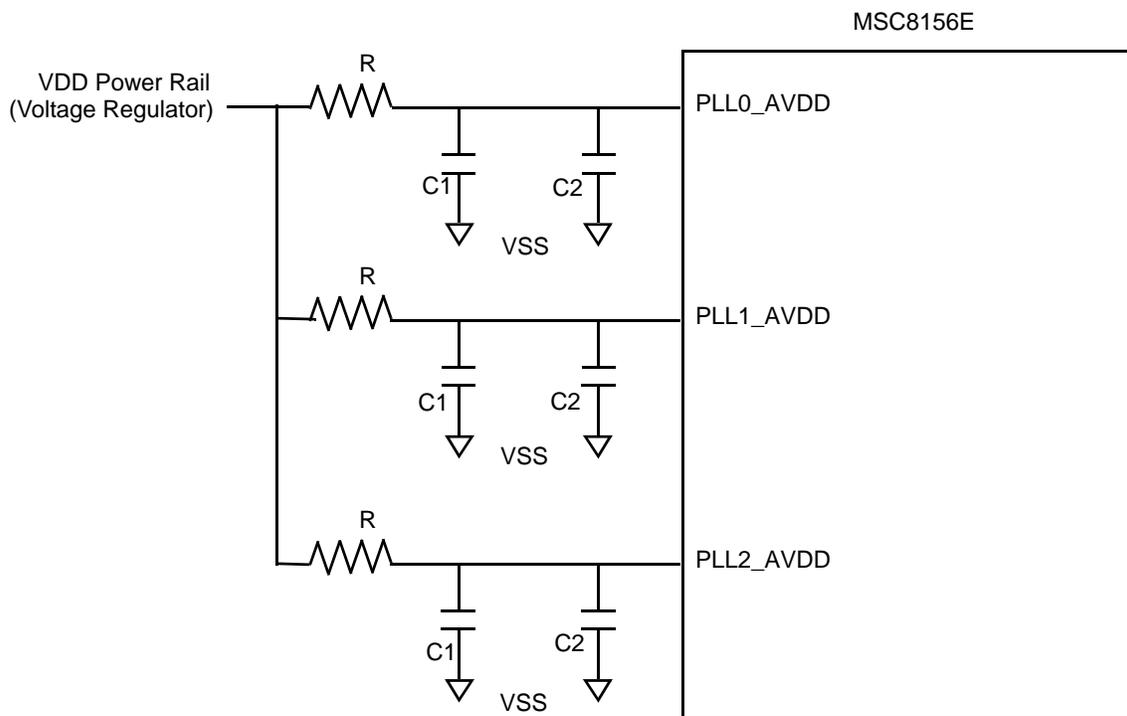


Figure 37. PLL Supplies

Each SerDes PLL power supply must be filtered using a circuit similar to the one shown in Figure 38, to ensure stability of the internal clock. For maximum effectiveness, the filter circuit should be placed as closely as possible to the SR<sub>n</sub>\_PLL\_AVDD ball to ensure it filters out as much noise as possible. The ground connection should be near the SR<sub>n</sub>\_PLL\_AVDD ball. The 0.003  $\mu\text{F}$  capacitor is closest to the ball, followed by the two 2.2  $\mu\text{F}$  capacitors, and finally the 1  $\Omega$  resistor to the board supply plane. The capacitors are connected from SR<sub>n</sub>\_PLL\_AVDD to the ground plane. Use ceramic chip capacitors with the highest possible self-resonant frequency. All trances should be kept short, wide, and direct.

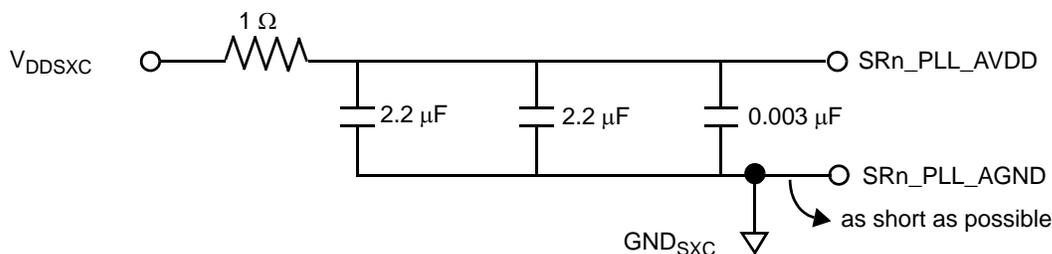


Figure 38. SerDes PLL Supplies

## 3.5 Connectivity Guidelines

**Note:** Although the package actually uses a ball grid array, the more conventional term pin is used to denote signal connections in this discussion.

First, select the pin multiplexing mode to allocate the required I/O signals. Then use the guidelines presented in the following subsections for board design and connections. The following conventions are used in describing the connectivity requirements:

1. GND indicates using a 10 k $\Omega$  pull-down resistor (recommended) or a direct connection to the ground plane. Direct connections to the ground plane may yield DC current up to 50 mA through the I/O supply that adds to overall power consumption.
2. V<sub>DD</sub> indicates using a 10 k $\Omega$  pull-up resistor (recommended) or a direct connection to the appropriate power supply. Direct connections to the supply may yield DC current up to 50 mA through the I/O supply that adds to overall power consumption.
3. Mandatory use of a pull-up or pull-down resistor is clearly indicated as “pull-up/pull-down.” For buses, each pin on the bus should have its own resistor.
4. NC indicates “not connected” and means do not connect anything to the pin.
5. The phrase “in use” indicates a typical pin connection for the required function.

**Note:** Please see recommendations #1 and #2 as mandatory pull-down or pull-up connection for unused pins in case of subset interface connection.

**Table 48. Connectivity of TDM Related Pins When TDM Interface Is Not Used**

Signal Name	Pin Connection
TDM $n$ TCLK	GND
TDM $n$ DAT	GND
TDM $n$ TSYN	GND
V <sub>DDIO</sub>	2.5 V
<b>Notes:</b> <ol style="list-style-type: none"> <li>1. <math>n = \{0, 1, 2, 3\}</math></li> <li>2. In case of subset of TDM interface usage please make sure to disable unused TDM modules. See <i>TDM</i> chapter in the <i>MSC8152 Reference Manual</i> for details.</li> </ol>	

### 3.5.5 Miscellaneous Pins

Table 49 lists the board connections for the pins not required by the system design. Table 49 assumes that the alternate function of the specified pin is not used. If the alternate function is used, connect that pin as required to support the selected function.

**Table 49. Connectivity of Individual Pins When They Are Not Required**

Signal Name	Pin Connection
CLKOUT	NC
EE0	GND
EE1	NC
GPIO[31–0]	NC
SCL	See the GPIO connectivity guidelines in this table.
SDA	See the GPIO connectivity guidelines in this table.
$\overline{\text{INT\_OUT}}$	NC
$\overline{\text{IRQ}}[15–0]$	See the GPIO connectivity guidelines in this table.
$\overline{\text{NMI}}$	V <sub>DDIO</sub>
$\overline{\text{NMI\_OUT}}$	NC
RC[21–0]	GND
STOP_BS	GND
TCK	GND
TDI	GND
TDO	NC
TMR[4–0]	See the GPIO connectivity guidelines in this table.
TMS	GND
$\overline{\text{TRST}}$	See <b>Section 3.1</b> for guidelines.
URXD	See the GPIO connectivity guidelines in this table.
UTXD	See the GPIO connectivity guidelines in this table.
DDN[1–0]	See the GPIO connectivity guidelines in this table.
DRQ[1–0]	See the GPIO connectivity guidelines in this table.
RCW_LSEL_0	GND
RCW_LSEL_1	GND
RCW_LSEL_2	GND
RCW_LSEL_3	GND
V <sub>DDIO</sub>	2.5 V

**Note:** For details on configuration, see the *MSC8152 Reference Manual*. For additional information, refer to the *MSC815x and MSC825x DSP Family Design Checklist*.

## 3.6 Guide to Selecting Connections for Remote Power Supply Sensing

To assure consistency of input power levels, some applications use a practice of connecting the remote sense signal input of an on-board power supply to one of power supply pins of the IC device. The advantage of using this connection is the ability to compensate for the slow components of the IR drop caused by resistive supply current path from on-board power supply to the pins layer on the package. However, because of specific device requirements, not every ball connection can be selected as the remote sense pin. Some of these pins must be connected to the appropriate power supply or ground to ensure correct device functionality. Some connections supply critical power to a specific high usage area of the IC die; using such a connection as a non-supply pin could impact necessary supply current during high current events. The following balls can be used as the board supply remote sense output without degrading the power and ground supply quality:

- *VDD*: W10, T19
- *VSS*: J18, Y10
- *M3VDD*: None

Do not use any other connections for remote sensing. Use of any other connections for this purpose can result in application and device failure.

## 4 Ordering Information

Consult a Freescale Semiconductor sales office or authorized distributor to determine product availability and place an order.

Qual Status	Cores	Encryption	Temperature Range	Package Type	Core Frequency	Die Revision
PC = Prototype MSC = Production	8152 = 2 Core	[blank] = Non-encrypted	S = 0° to 105°C T = -40°C to 105°C	VT = FC-PBGA Lead Free AG = FC-PBGA C4/C5 Lead Free	1000 = 1Ghz	B = Rev 2.1