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Details

Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	40MHz
Connectivity	I ² C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, LVD, POR, PWM, WDT
Number of I/O	21
Program Memory Size	24KB (12K x 16)
Program Memory Type	FLASH
EEPROM Size	256 x 8
RAM Size	1408 x 8
Voltage - Supply (Vcc/Vdd)	4.2V ~ 5.5V
Data Converters	A/D 5x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	28-SOIC (0.295", 7.50mm Width)
Supplier Device Package	28-SOIC
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic18f2539-i-so

PIC18FXX39

TABLE 2-1: CAPACITOR SELECTION FOR CRYSTAL OSCILLATOR

Ranges Tested:			
Mode	Freq	C1	C2
HS	20.0 MHz	15-33 pF	15-33 pF
These values are for design guidance only. See notes following this table.			
Crystals Used			
20.0 MHz	Epson CA-301 20.000M-C	± 30 PPM	

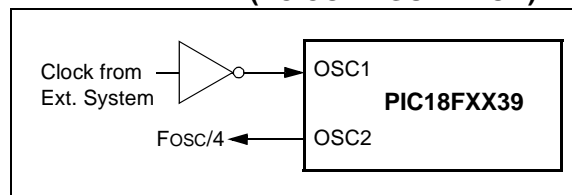
- Note 1:** Higher capacitance increases the stability of the oscillator, but also increases the start-up time.
- 2:** Rs may be required in HS mode to avoid overdriving crystals with low drive level specification.
- 3:** Since each resonator/crystal has its own characteristics, the user should consult the resonator/crystal manufacturer for appropriate values of external components, or verify oscillator performance.

2.3 External Clock Input

The EC and ECIO Oscillator modes require an external clock source to be connected to the OSC1 pin. The feedback device between OSC1 and OSC2 is turned off in these modes to save current. There is no oscillator start-up time required after a Power-on Reset or after a recovery from SLEEP mode.

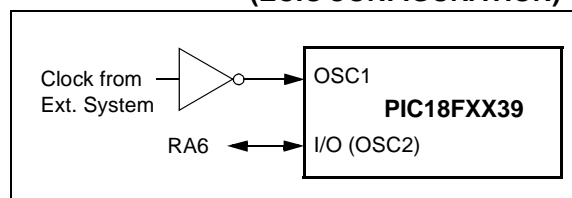
In the EC Oscillator mode, the oscillator frequency divided by 4 is available on the OSC2 pin. This signal may be used for test purposes or to synchronize other logic. Figure 2-3 shows the pin connections for the EC Oscillator mode.

FIGURE 2-3: EXTERNAL CLOCK INPUT OPERATION (EC CONFIGURATION)



The ECIO Oscillator mode functions like the EC mode, except that the OSC2 pin becomes an additional general purpose I/O pin. The I/O pin becomes bit 6 of PORTA (RA6). Figure 2-4 shows the pin connections for the ECIO Oscillator mode.

FIGURE 2-4: EXTERNAL CLOCK INPUT OPERATION (ECIO CONFIGURATION)



2.4 HS/PLL

A Phase Locked Loop circuit is provided as a programmable option for users that want to multiply the frequency of the incoming crystal oscillator signal by 4. For an input clock frequency of 5 MHz, the internal clock frequency will be multiplied to 20 MHz. This is useful for customers who are concerned with EMI due to high frequency crystals.

The PLL can only be enabled when the oscillator configuration bits are programmed for HS mode. If they are programmed for any other mode, the PLL is not enabled and the system clock will come directly from OSC1.

The PLL is one of the modes specified by the FOSC<2:0> configuration bits. The Oscillator mode is specified during device programming.

A PLL lock timer is used to ensure that the PLL has locked before device execution starts. The PLL lock timer has a time-out that is called TPLL.

4.2 Return Address Stack

The return address stack allows any combination of up to 31 program calls and interrupts to occur. The PC (Program Counter) is pushed onto the stack when a `CALL` or `RCALL` instruction is executed, or an interrupt is acknowledged. The PC value is pulled off the stack on a `RETURN`, `RETLW` or a `RETFIE` instruction. `PCLATU` and `PCLATH` are not affected by any of the `RETURN` or `CALL` instructions.

The stack operates as a 31-word by 21-bit RAM and a 5-bit stack pointer, with the stack pointer initialized to 00000b after all RESETS. There is no RAM associated with stack pointer 00000b. This is only a RESET value. During a `CALL` type instruction, causing a push onto the stack, the stack pointer is first incremented and the RAM location pointed to by the stack pointer is written with the contents of the PC. During a `RETURN` type instruction, causing a pop from the stack, the contents of the RAM location pointed to by the `STKPTR` are transferred to the PC and then the stack pointer is decremented.

The stack space is not part of either program or data space. The stack pointer is readable and writable, and the address on the top of the stack is readable and writable through SFR registers. Data can also be pushed to, or popped from the stack using the top-of-stack SFRs. Status bits indicate if the stack pointer is at, or beyond the 31 levels provided.

4.2.1 TOP-OF-STACK ACCESS

The top of the stack is readable and writable. Three register locations, `TOSU`, `TOSH` and `TOSL` hold the contents of the stack location pointed to by the `STKPTR` register. This allows users to implement a software stack if necessary. After a `CALL`, `RCALL` or interrupt, the software can read the pushed value by reading the `TOSU`, `TOSH` and `TOSL` registers. These values can be placed on a user defined software stack. At return time, the software can replace the `TOSU`, `TOSH` and `TOSL` and do a return.

The user must disable the global interrupt enable bits during this time to prevent inadvertent stack operations.

4.2.2 RETURN STACK POINTER (STKPTR)

The `STKPTR` register contains the stack pointer value, the `STKFUL` (stack full) status bit, and the `STKUNF` (stack underflow) status bits. Register 4-1 shows the `STKPTR` register. The value of the stack pointer can be 0 through 31. The stack pointer increments when values are pushed onto the stack and decrements when values are popped off the stack. At RESET, the stack pointer value will be '0'. The user may read and write the stack pointer value. This feature can be used by a Real-Time Operating System for return stack maintenance.

After the PC is pushed onto the stack 31 times (without popping any values off the stack), the `STKFUL` bit is set. The `STKFUL` bit can only be cleared in software or by a POR.

The action that takes place when the stack becomes full depends on the state of the `STVREN` (Stack Overflow Reset Enable) configuration bit. Refer to Section 21.0 for a description of the device configuration bits. If `STVREN` is set (default), the 31st push will push the (PC + 2) value onto the stack, set the `STKFUL` bit, and reset the device. The `STKFUL` bit will remain set and the stack pointer will be set to '0'.

If `STVREN` is cleared, the `STKFUL` bit will be set on the 31st push and the stack pointer will increment to 31. Any additional pushes will not overwrite the 31st push, and `STKPTR` will remain at 31.

When the stack has been popped enough times to unload the stack, the next pop will return a value of zero to the PC and sets the `STKUNF` bit, while the stack pointer remains at '0'. The `STKUNF` bit will remain set until cleared in software or a POR occurs.

Note: Returning a value of zero to the PC on an underflow has the effect of vectoring the program to the RESET vector, where the stack conditions can be verified and appropriate actions can be taken.

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4.14 RCON Register

The Reset Control (RCON) register contains flag bits that allow differentiation between the sources of a device RESET. These flags include the \overline{TO} , \overline{PD} , \overline{POR} , \overline{BOR} and \overline{RI} bits. This register is readable and writable.

For PIC18FXX39 devices, the IPEN bit must always be set (= 1) for the ProMPT kernel to function correctly. Refer to Section 8.0 (page 69) for a more detailed discussion.

Note 1: If the BOREN configuration bit is set (Brown-out Reset enabled), the \overline{BOR} bit is '1' on a Power-on Reset. After a Brown-out Reset has occurred, the \overline{BOR} bit will be cleared, and must be set by firmware to indicate the occurrence of the next Brown-out Reset.

2: It is recommended that the \overline{POR} bit be set after a Power-on Reset has been detected, so that subsequent Power-on Resets may be detected.

REGISTER 4-3: RCON REGISTER

R/W-0	U-0	U-0	R/W-1	R-1	R-1	R/W-0	R/W-0
IPEN	—	—	\overline{RI}	\overline{TO}	\overline{PD}	\overline{POR}	\overline{BOR}
bit 7			bit 0				

- bit 7 **IPEN:** Interrupt Priority Enable bit
Always maintain this bit set for proper operation of ProMPT kernel.
- bit 6-5 **Unimplemented:** Read as '0'
- bit 4 **\overline{RI} :** RESET Instruction Flag bit
1 = The RESET instruction was not executed
0 = The RESET instruction was executed causing a device RESET
(must be set in software after a Brown-out Reset occurs)
- bit 3 **\overline{TO} :** Watchdog Time-out Flag bit
1 = After power-up, CLRWD \overline{T} instruction, or SLEEP instruction
0 = A WDT time-out occurred
- bit 2 **\overline{PD} :** Power-down Detection Flag bit
1 = After power-up or by the CLRWD \overline{T} instruction
0 = By execution of the SLEEP instruction
- bit 1 **\overline{POR} :** Power-on Reset Status bit
1 = A Power-on Reset has not occurred
0 = A Power-on Reset occurred
(must be set in software after a Power-on Reset occurs)
- bit 0 **\overline{BOR} :** Brown-out Reset Status bit
1 = A Brown-out Reset has not occurred
0 = A Brown-out Reset occurred
(must be set in software after a Brown-out Reset occurs)

Legend:

R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
- n = Value at POR	'1' = Bit is set	'0' = Bit is cleared x = Bit is unknown

8.6 INT0 Interrupt

External interrupts on the RB0/INT0, RB1/INT1 and RB2/INT2 pins are edge triggered: either rising, if the corresponding INTEDGx bit is set in the INTCON2 register, or falling, if the INTEDGx bit is clear. When a valid edge appears on the RBx/INTx pin, the corresponding flag bit INTxF is set. This interrupt can be disabled by clearing the corresponding enable bit INTxE. Flag bit INTxF must be cleared in software in the Interrupt Service Routine before re-enabling the interrupt. All external interrupts (INT0, INT1 and INT2) can wake-up the processor from SLEEP, if bit INTxE was set prior to going into SLEEP. If the global interrupt enable bit GIE is set, the processor will branch to the interrupt vector following wake-up.

The INT0 interrupt is always configured as a high priority interrupt, and cannot be reconfigured. Interrupt priority for INT1 and INT2 is determined by the value contained in the interrupt priority bits, INT1IP (INTCON3<6>) and INT2IP (INTCON3<7>).

Because it is always configured as a high priority interrupt, INT0 cannot be used in conjunction with the ProMPT kernel; it must always be disabled (INTCON<4> = 0). Failure to do this may result in erratic operation of the motor control.

8.7 TMR0 Interrupt

In 8-bit mode (which is the default), an overflow in the TMR0 register (FFh → 00h) will set flag bit TMR0IF. In 16-bit mode, an overflow in the TMR0H:TMR0L register pair (FFFFh → 0000h) will set flag bit TMR0IF. The interrupt can be enabled or disabled by setting or clearing enable bit TMR0IE (INTCON<5>). Interrupt priority for Timer0 is determined by the value contained in the interrupt priority bit TMR0IP (INTCON2<2>). See Section 10.0 for further details on the Timer0 module.

8.8 PORTB Interrupt-on-Change

An input change on PORTB<7:4> sets flag bit RBIF (INTCON<0>). The interrupt can be enabled or disabled by setting or clearing the enable bit RBIE (INTCON<3>). Interrupt priority for PORTB interrupt-on-change is determined by the value contained in the interrupt priority bit RBIP (INTCON2<0>).

8.9 Context Saving During Interrupts

During an interrupt, the return PC value is saved on the stack. Additionally, the WREG, STATUS and BSR registers are saved on the fast return stack. If a fast return from interrupt is not used (see Section 4.3), the user may need to save the WREG, STATUS and BSR registers in software. Depending on the user's application, other registers may also need to be saved. Example 8-1 saves and restores the WREG, STATUS and BSR registers during an Interrupt Service Routine.

EXAMPLE 8-1: SAVING STATUS, WREG AND BSR REGISTERS IN RAM

```
MOVWF  W_TEMP          ; W_TEMP is in virtual bank
MOVFF  STATUS, STATUS_TEMP ; STATUS_TEMP located anywhere
MOVFF  BSR,    BSR_TEMP   ; BSR located anywhere
;
; USER ISR CODE
;
MOVFF  BSR_TEMP, BSR      ; Restore BSR
MOVF   W_TEMP,   W        ; Restore WREG
MOVFF  STATUS_TEMP, STATUS ; Restore STATUS
```

9.4 PORTD, TRISD and LATD Registers

This section is applicable only to the PIC18F4X39 devices.

PORTD is an 8-bit wide, bi-directional port. The corresponding Data Direction register is TRISD. Setting a TRISD bit (= 1) will make the corresponding PORTD pin an input (i.e., put the corresponding output driver in a High Impedance mode). Clearing a TRISD bit (= 0) will make the corresponding PORTD pin an output (i.e., put the contents of the output latch on the selected pin).

The Data Latch register (LATD) is also memory mapped. Read-modify-write operations on the LATD register reads and writes the latched output value for PORTD.

PORTD is an 8-bit port with Schmitt Trigger input buffers. Each pin is individually configurable as an input or output.

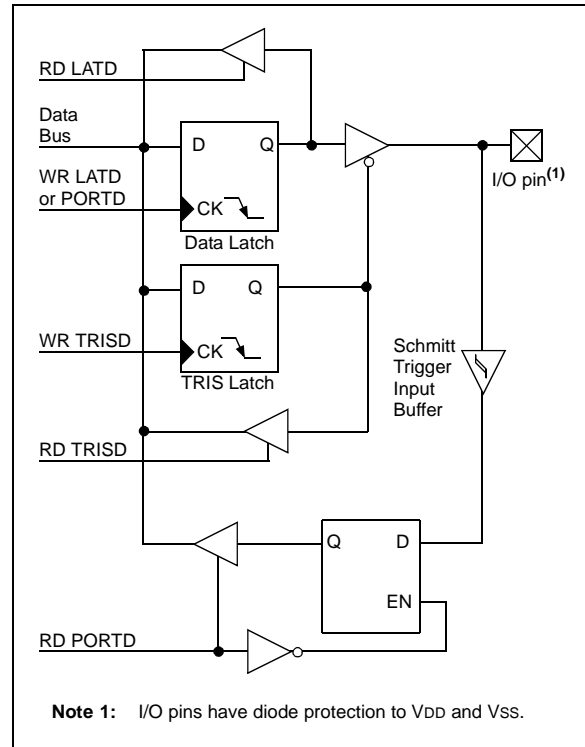
Note: On a Power-on Reset, these pins are configured as digital inputs.

PORTD can be configured as an 8-bit wide microprocessor port (parallel slave port) by setting control bit PSPMODE (TRISE<4>). In this mode, the input buffers are TTL. See Section 9.6 for additional information on the Parallel Slave Port (PSP).

EXAMPLE 9-4: INITIALIZING PORTD

```
CLRF    PORTD    ; Initialize PORTD by
                  ; clearing output
                  ; data latches
CLRF    LATD     ; Alternate method
                  ; to clear output
                  ; data latches
MOVLW  0xCF     ; Value used to
                  ; initialize data
                  ; direction
MOVWF  TRISD    ; Set RD<3:0> as inputs
                  ; RD<5:4> as outputs
                  ; RD<7:6> as inputs
```

FIGURE 9-8: PORTD BLOCK DIAGRAM IN I/O PORT MODE



16.3.6 SLAVE MODE

In Slave mode, the data is transmitted and received as the external clock pulses appear on SCK. When the last bit is latched, the SSPIF interrupt flag bit is set.

While in Slave mode, the external clock is supplied by the external clock source on the SCK pin. This external clock must meet the minimum high and low times, as specified in the electrical specifications.

While in SLEEP mode, the slave can transmit/receive data. When a byte is received, the device will wake-up from SLEEP.

16.3.7 SLAVE SELECT SYNCHRONIZATION

The \overline{SS} pin allows a Synchronous Slave mode. The SPI must be in Slave mode with \overline{SS} pin control enabled (SSPCON1<3:0> = 04h). The pin must not be driven low for the \overline{SS} pin to function as an input. The Data Latch must be high. When the \overline{SS} pin is low, transmission and reception are enabled and the SDO pin is driven. When the \overline{SS} pin goes high, the SDO pin is no

longer driven, even if in the middle of a transmitted byte, and becomes a floating output. External pull-up/pull-down resistors may be desirable, depending on the application.

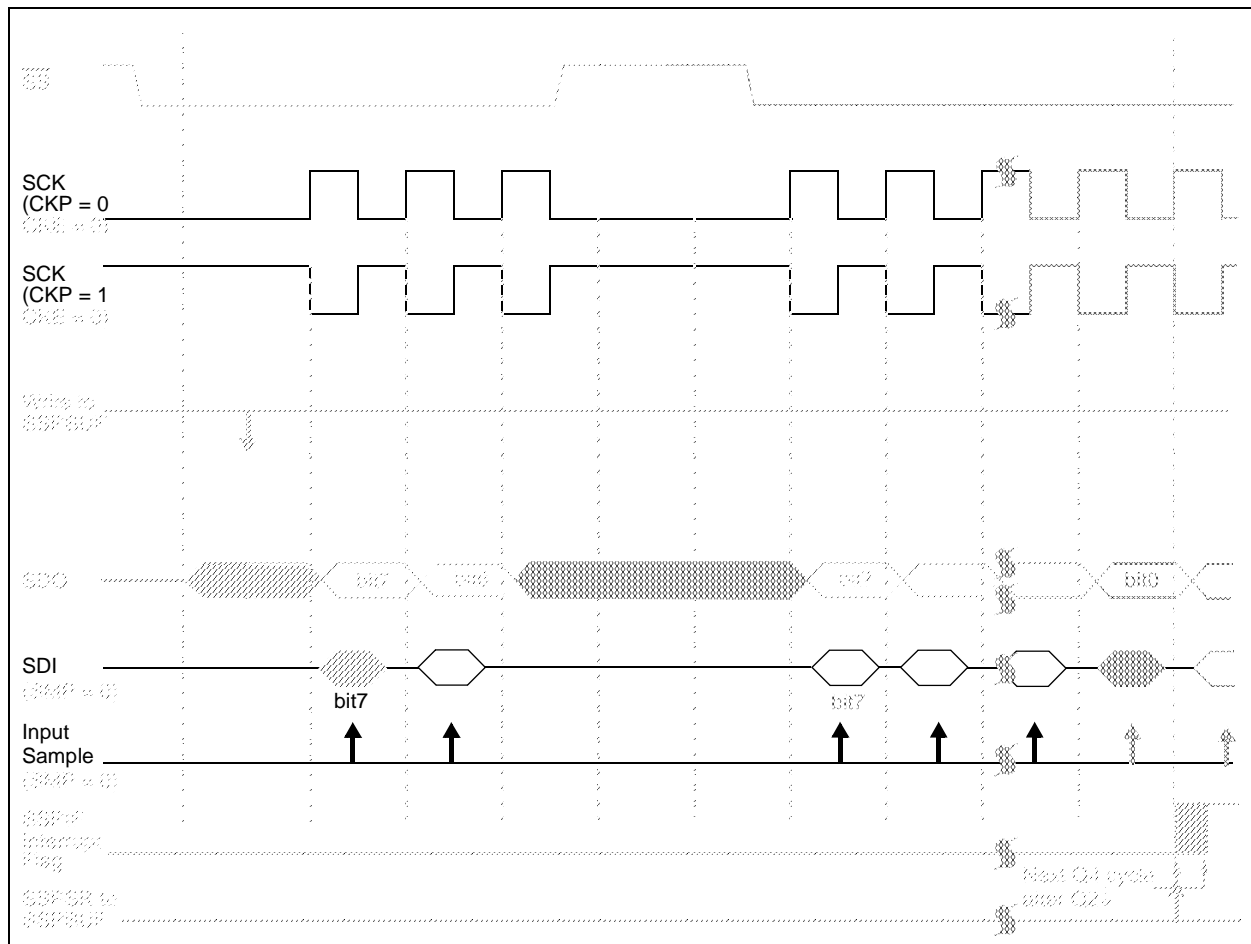
Note 1: When the SPI is in Slave mode with \overline{SS} pin control enabled (SSPCON<3:0> = 0100), the SPI module will reset if the \overline{SS} pin is set to VDD.

2: If the SPI is used in Slave mode with CKE set, then the \overline{SS} pin control must be enabled.

When the SPI module resets, the bit counter is forced to '0'. This can be done by either forcing the \overline{SS} pin to a high level or clearing the SSPEN bit.

To emulate two-wire communication, the SDO pin can be connected to the SDI pin. When the SPI needs to operate as a receiver, the SDO pin can be configured as an input. This disables transmissions from the SDO. The SDI can always be left as an input (SDI function), since it cannot create a bus conflict.

FIGURE 16-4: SLAVE SYNCHRONIZATION WAVEFORM



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16.4.17.1 Bus Collision During a START Condition

During a START condition, a bus collision occurs if:

- SDA or SCL are sampled low at the beginning of the START condition (Figure 16-26).
- SCL is sampled low before SDA is asserted low (Figure 16-27).

During a START condition, both the SDA and the SCL pins are monitored.

If the SDA pin is already low, or the SCL pin is already low, then all of the following occur:

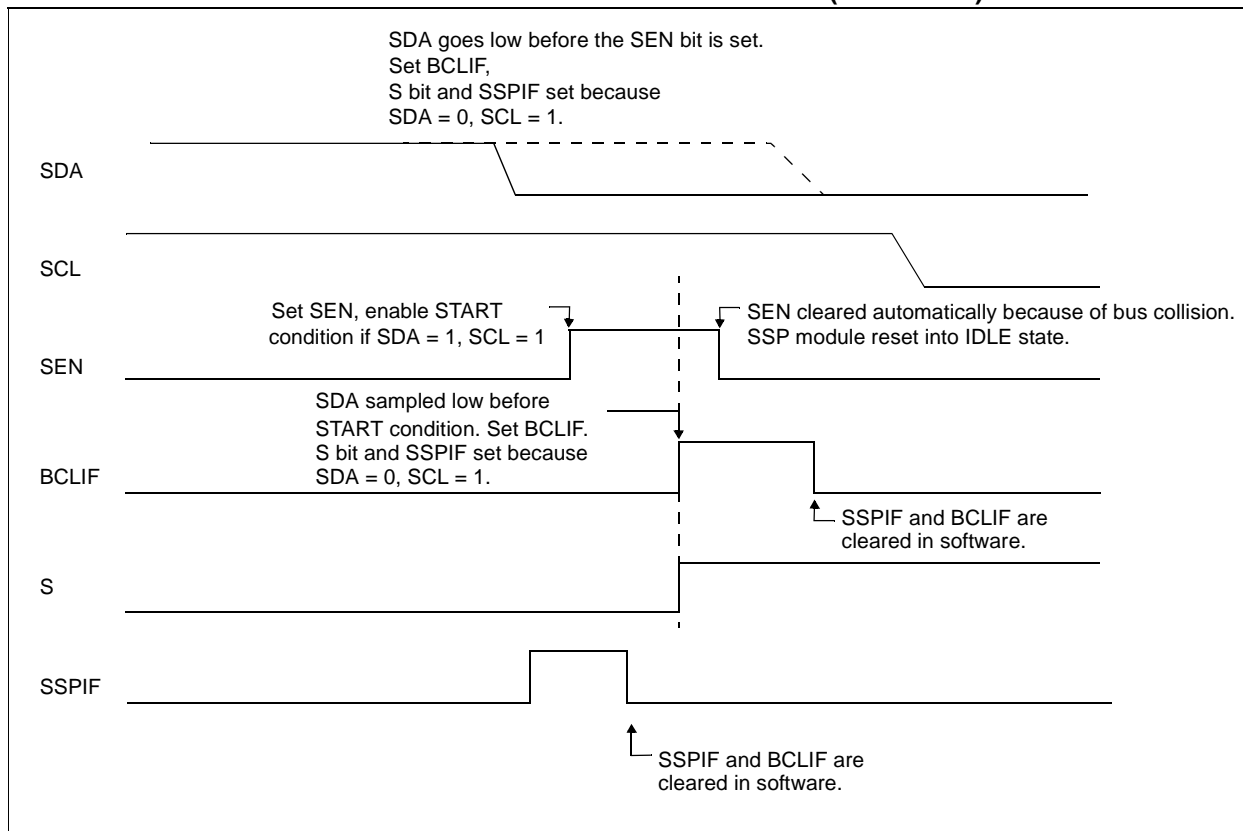
- the START condition is aborted,
- the BCLIF flag is set, and
- the MSSP module is reset to its IDLE state (Figure 16-26).

The START condition begins with the SDA and SCL pins de-asserted. When the SDA pin is sampled high, the baud rate generator is loaded from SSPADD<6:0> and counts down to '0'. If the SCL pin is sampled low while SDA is high, a bus collision occurs, because it is assumed that another master is attempting to drive a data '1' during the START condition.

If the SDA pin is sampled low during this count, the BRG is reset and the SDA line is asserted early (Figure 16-28). If, however, a '1' is sampled on the SDA pin, the SDA pin is asserted low at the end of the BRG count. The baud rate generator is then reloaded and counts down to '0', and during this time, if the SCL pins are sampled as '0', a bus collision does not occur. At the end of the BRG count, the SCL pin is asserted low.

Note: The reason that bus collision is not a factor during a START condition, is that no two bus masters can assert a START condition at the exact same time. Therefore, one master will always assert SDA before the other. This condition does not cause a bus collision, because the two masters must be allowed to arbitrate the first address following the START condition. If the address is the same, arbitration must be allowed to continue into the data portion, Repeated START or STOP conditions.

FIGURE 16-26: BUS COLLISION DURING START CONDITION (SDA ONLY)



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17.3 USART Synchronous Master Mode

In Synchronous Master mode, the data is transmitted in a half-duplex manner (i.e., transmission and reception do not occur at the same time). When transmitting data, the reception is inhibited and vice versa. Synchronous mode is entered by setting bit SYNC (TXSTA<4>). In addition, enable bit SPEN (RCSTA<7>) is set in order to configure the RC6/TX/CK and RC7/RX/DT I/O pins to CK (clock) and DT (data) lines, respectively. The Master mode indicates that the processor transmits the master clock on the CK line. The Master mode is entered by setting bit CSRC (TXSTA<7>).

17.3.1 USART SYNCHRONOUS MASTER TRANSMISSION

The USART transmitter block diagram is shown in Figure 17-1. The heart of the transmitter is the Transmit (serial) Shift Register (TSR). The shift register obtains its data from the read/write transmit buffer register TXREG. The TXREG register is loaded with data in software. The TSR register is not loaded until the last bit has been transmitted from the previous load. As soon as the last bit is transmitted, the TSR is loaded with new data from the TXREG (if available). Once the TXREG register transfers the data to the TSR register (occurs in one TCYCLE), the TXREG is empty and interrupt bit TXIF (PIR1<4>) is set. The interrupt can be enabled/disabled by setting/clearing enable bit TXIE

(PIE1<4>). Flag bit TXIF will be set, regardless of the state of enable bit TXIE, and cannot be cleared in software. It will reset only when new data is loaded into the TXREG register. While flag bit TXIF indicates the status of the TXREG register, another bit TRMT (TXSTA<1>) shows the status of the TSR register. TRMT is a read only bit, which is set when the TSR is empty. No interrupt logic is tied to this bit, so the user has to poll this bit in order to determine if the TSR register is empty. The TSR is not mapped in data memory, so it is not available to the user.

To set up a Synchronous Master Transmission:

1. Initialize the SPBRG register for the appropriate baud rate (Section 17.1).
2. Enable the synchronous master serial port by setting bits SYNC, SPEN, and CSRC.
3. If interrupts are desired, set enable bit TXIE.
4. If 9-bit transmission is desired, set bit TX9.
5. Enable the transmission by setting bit TXEN.
6. If 9-bit transmission is selected, the ninth bit should be loaded in bit TX9D.
7. Start transmission by loading data to the TXREG register.

Note: TXIF is not cleared immediately upon loading data into the transmit buffer TXREG. The flag bit becomes valid in the second instruction cycle following the load instruction.

TABLE 17-8: REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER TRANSMISSION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on All Other RESETS
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	0000 000x	0000 000u
PIR1	PSPIF ⁽¹⁾	ADIF	RCIF	TXIF	SSPIF	—	TMR2IF	TMR1IF	0000 0000	0000 0000
PIE1	PSPIE ⁽¹⁾	ADIE	RCIE	TXIE	SSPIE	—	TMR2IE	TMR1IE	0000 0000	0000 0000
IPR1	PSPIP ⁽¹⁾	ADIP	RCIP	TXIP	SSPIP	—	TMR2IP	TMR1IP	0000 0000	0000 0000
RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	0000 -00x	0000 -00x
TXREG	USART Transmit Register								0000 0000	0000 0000
TXSTA	CSRC	TX9	TXEN	SYNC	—	BRGH	TRMT	TX9D	0000 -010	0000 -010
SPBRG	Baud Rate Generator Register								0000 0000	0000 0000

Legend: x = unknown, - = unimplemented, read as '0'.

Shaded cells are not used for Synchronous Master Transmission.

Note 1: The PSPIF, PSPIE and PSPIP bits are reserved on the PIC18F2X39 devices; always maintain these bits clear.

19.2.1 REFERENCE VOLTAGE SET POINT

The Internal Reference Voltage of the LVD module may be used by other internal circuitry (the Programmable Brown-out Reset). If these circuits are disabled (lower current consumption), the reference voltage circuit requires a time to become stable before a low voltage condition can be reliably detected. This time is invariant of system clock speed. This start-up time is specified in electrical specification parameter 36. The low voltage interrupt flag will not be enabled until a stable reference voltage is reached. Refer to the waveform in Figure 19-4.

19.2.2 CURRENT CONSUMPTION

When the module is enabled, the LVD comparator and voltage divider are enabled and will consume static current. The voltage divider can be tapped from multiple places in the resistor array. Total current consumption, when enabled, is specified in electrical specification parameter #D022B.

19.3 Operation During SLEEP

When enabled, the LVD circuitry continues to operate during SLEEP. If the device voltage crosses the trip point, the LVDIF bit will be set and the device will wake-up from SLEEP. Device execution will continue from the interrupt vector address if interrupts have been globally enabled.

19.4 Effects of a RESET

A device RESET forces all registers to their RESET state. This forces the LVD module to be turned off.

20.0 SPECIAL FEATURES OF THE CPU

There are several features intended to maximize system reliability, minimize cost through elimination of external components, provide power saving Operating modes and offer code protection. These are:

- OSC Selection
- RESET
 - Power-on Reset (POR)
 - Power-up Timer (PWRT)
 - Oscillator Start-up Timer (OST)
 - Brown-out Reset (BOR)
- Interrupts
- Watchdog Timer (WDT)
- SLEEP
- Code Protection
- ID Locations
- In-Circuit Serial Programming

All PIC18FXX39 devices have a Watchdog Timer, which is permanently enabled via the configuration bits, or software controlled. It runs off its own RC oscillator for added reliability. There are two timers that offer necessary delays on power-up. One is the Oscillator Start-up Timer (OST), intended to keep the chip in RESET until the crystal oscillator is stable. The other is the Power-up Timer (PWRT), which provides a fixed delay on power-up only, designed to keep the part in RESET while the power supply stabilizes. With these two timers on-chip, most applications need no external RESET circuitry.

SLEEP mode is designed to offer a very low current Power-down mode. The user can wake-up from SLEEP through external RESET, Watchdog Timer Wake-up or through an interrupt. Several oscillator options are also made available to allow the part to fit the application. The RC oscillator option saves system cost, while the LP crystal option saves power. A set of configuration bits are used to select various options.

20.1 Configuration Bits

The configuration bits can be programmed (read as '0'), or left unprogrammed (read as '1'), to select various device configurations. These bits are mapped starting at program memory location 300000h.

The user will note that address 300000h is beyond the user program memory space. In fact, it belongs to the configuration memory space (300000h - 3FFFFh), which can only be accessed using Table Reads and Table Writes.

Programming the configuration registers is done in a manner similar to programming the FLASH memory (see Section 5.5.1). The only difference is the configuration registers are written a byte at a time. The sequence of events for programming configuration registers is:

1. Load table pointer with address of configuration register being written.
2. Write a single byte using the `TBLWT` instruction.
3. Set `EEPGD` to point to program memory, set the `CFGS` bit to access configuration registers, and set `WREN` to enable byte writes.
4. Disable interrupts.
5. Write 55h to `EECON2`.
6. Write AAh to `EECON2`.
7. Set the `WR` bit. This will begin the write cycle.
8. CPU will stall for duration of write (approximately 2 ms using internal timer).
9. Execute a `NOP`.
10. Re-enable interrupts.

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TABLE 21-1: OPCODE FIELD DESCRIPTIONS

Field	Description
a	RAM access bit a = 0: RAM location in Access RAM (BSR register is ignored) a = 1: RAM bank is specified by BSR register
bbb	Bit address within an 8-bit file register (0 to 7).
BSR	Bank Select Register. Used to select the current RAM bank.
d	Destination select bit d = 0: store result in WREG, d = 1: store result in file register f.
dest	Destination, either the WREG register or the specified register file location.
f	8-bit Register file address (0x00 to 0xFF).
fs	12-bit Register file address (0x000 to 0xFFF). This is the source address.
fd	12-bit Register file address (0x000 to 0xFFF). This is the destination address.
k	Literal field, constant data or label (may be either an 8-bit, 12-bit or a 20-bit value).
label	Label name.
mm	The mode of the TBLPTR register for the Table Read and Table Write instructions. Only used with Table Read and Table Write instructions:
*	No Change to register (such as TBLPTR with Table Reads and Writes).
*+	Post-Increment register (such as TBLPTR with Table Reads and Writes).
*-	Post-Decrement register (such as TBLPTR with Table Reads and Writes).
+*	Pre-Increment register (such as TBLPTR with Table Reads and Writes).
n	The relative address (2's complement number) for relative branch instructions, or the direct address for Call/Branch and Return instructions.
PRODH	Product of Multiply high byte.
PRODL	Product of Multiply low byte.
s	Fast Call/Return mode select bit s = 0: do not update into/from shadow registers s = 1: certain registers loaded into/from shadow registers (Fast mode)
u	Unused or Unchanged.
WREG	Working register (accumulator).
x	Don't care (0 or 1). The assembler will generate code with x = 0. It is the recommended form of use for compatibility with all Microchip software tools.
TBLPTR	21-bit Table Pointer (points to a Program Memory location).
TABLAT	8-bit Table Latch.
TOS	Top-of-Stack.
PC	Program Counter.
PCL	Program Counter Low Byte.
PCH	Program Counter High Byte.
PCLATH	Program Counter High Byte Latch.
PCLATU	Program Counter Upper Byte Latch.
GIE	Global Interrupt Enable bit.
WDT	Watchdog Timer.
T \overline{O}	Time-out bit.
P \overline{D}	Power-down bit.
C, DC, Z, OV, N	ALU status bits Carry, Digit Carry, Zero, Overflow, Negative.
[]	Optional.
()	Contents.
→	Assigned to.
< >	Register bit field.
∈	In the set of.
<i>italics</i>	User defined term (font is courier).

TABLE 21-2: PIC18FXXX INSTRUCTION SET (CONTINUED)

Mnemonic, Operands		Description	Cycles	16-Bit Instruction Word				Status Affected	Notes
				MSb		LSb			
CONTROL OPERATIONS									
BC	n	Branch if Carry	1 (2)	1110	0010	nnnn	nnnn	None	4
BN	n	Branch if Negative	1 (2)	1110	0110	nnnn	nnnn	None	
BNC	n	Branch if Not Carry	1 (2)	1110	0011	nnnn	nnnn	None	
BNN	n	Branch if Not Negative	1 (2)	1110	0111	nnnn	nnnn	None	
BN OV	n	Branch if Not Overflow	1 (2)	1110	0101	nnnn	nnnn	None	
BN Z	n	Branch if Not Zero	2	1110	0001	nnnn	nnnn	None	
BO V	n	Branch if Overflow	1 (2)	1110	0100	nnnn	nnnn	None	
BRA	n	Branch Unconditionally	1 (2)	1101	0nnn	nnnn	nnnn	None	
BZ	n	Branch if Zero	1 (2)	1110	0000	nnnn	nnnn	None	
CALL	n, s	Call subroutine 1st word	2	1110	110s	kkkk	kkkk	None	
		2nd word		1111	kkkk	kkkk	kkkk		
CLRWDT	—	Clear Watchdog Timer	1	0000	0000	0000	0100	\overline{TO} , \overline{PD}	
DAW	—	Decimal Adjust WREG	1	0000	0000	0000	0111	C	
GOTO	n	Go to address 1st word	2	1110	1111	kkkk	kkkk	None	
		2nd word		1111	kkkk	kkkk	kkkk		
NOP	—	No Operation	1	0000	0000	0000	0000	None	
NOP	—	No Operation	1	1111	xxxx	xxxx	xxxx	None	
POP	—	Pop top of return stack (TOS)	1	0000	0000	0000	0110	None	
PUSH	—	Push top of return stack (TOS)	1	0000	0000	0000	0101	None	
RCALL	n	Relative Call	2	1101	1nnn	nnnn	nnnn	None	
RESET		Software device RESET	1	0000	0000	1111	1111	All	
RETFIE	s	Return from interrupt enable	2	0000	0000	0001	000s	GIE/GIEH, PEIE/GIEL	
RETLW	k	Return with literal in WREG	2	0000	1100	kkkk	kkkk	None	
RETURN	s	Return from Subroutine	2	0000	0000	0001	001s	None	
SLEEP	—	Go into Standby mode	1	0000	0000	0000	0011	\overline{TO} , \overline{PD}	

- Note 1:** When a PORT register is modified as a function of itself (e.g., `MOVF PORTB, 1, 0`), the value used will be that value present on the pins themselves. For example, if the data latch is '1' for a pin configured as input and is driven low by an external device, the data will be written back with a '0'.
- 2:** If this instruction is executed on the TMR0 register (and, where applicable, $d = 1$), the prescaler will be cleared if assigned.
- 3:** If Program Counter (PC) is modified or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOP.
- 4:** Some instructions are 2-word instructions. The second word of these instructions will be executed as a NOP, unless the first word of the instruction retrieves the information embedded in these 16-bits. This ensures that all program memory locations have a valid instruction.
- 5:** If the Table Write starts the write cycle to internal memory, the write will continue until terminated.

BRA Unconditional Branch

Syntax: [*label*] BRA n

Operands: $-1024 \leq n \leq 1023$

Operation: $(PC) + 2 + 2n \rightarrow PC$

Status Affected: None

Encoding:

1101	0nnn	nnnn	nnnn
------	------	------	------

Description: Add the 2's complement number '2n' to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be $PC+2+2n$. This instruction is a two-cycle instruction.

Words: 1

Cycles: 2

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read literal 'n'	Process Data	Write to PC
No operation	No operation	No operation	No operation

Example: HERE BRA Jump

Before Instruction
PC = address (HERE)

After Instruction
PC = address (Jump)

BSF Bit Set f

Syntax: [*label*] BSF f,b[,a]

Operands: $0 \leq f \leq 255$
 $0 \leq b \leq 7$
 $a \in [0,1]$

Operation: $1 \rightarrow f < b >$

Status Affected: None

Encoding:

1000	bbba	ffff	ffff
------	------	------	------

Description: Bit 'b' in register 'f' is set. If 'a' is 0 Access Bank will be selected, overriding the BSR value. If 'a' = 1, then the bank will be selected as per the BSR value.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write register 'f'

Example: BSF FLAG_REG, 7, 1

Before Instruction
FLAG_REG = 0x0A

After Instruction
FLAG_REG = 0x8A

DECFSZ		Decrement f, skip if 0							
Syntax:	[<i>label</i>] DECFSZ f [,d [,a]]								
Operands:	0 ≤ f ≤ 255 d ∈ [0,1] a ∈ [0,1]								
Operation:	(f) − 1 → dest, skip if result = 0								
Status Affected:	None								
Encoding:	<table border="1"><tr><td>0010</td><td>11da</td><td>ffff</td><td>ffff</td></tr></table>					0010	11da	ffff	ffff
0010	11da	ffff	ffff						
Description:	<p>The contents of register 'f' are decremented. If 'd' is 0, the result is placed in W. If 'd' is 1, the result is placed back in register 'f' (default). If the result is 0, the next instruction, which is already fetched, is discarded, and a NOP is executed instead, making it a two-cycle instruction. If 'a' is 0, the Access Bank will be selected, overriding the BSR value. If 'a' = 1, then the bank will be selected as per the BSR value (default).</p>								
Words:	1								
Cycles:	1(2)								
	Note: 3 cycles if skip and followed by a 2-word instruction.								

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write to destination

If skip:

Q1	Q2	Q3	Q4
No operation	No operation	No operation	No operation

If skip and followed by 2-word instruction:

Q1	Q2	Q3	Q4
No operation	No operation	No operation	No operation
No operation	No operation	No operation	No operation

Example:

```

HERE      DECFSZ  CNT, 1, 1
          GOTO    LOOP
          CONTINUE
  
```

Before Instruction

PC = Address (HERE)

After Instruction

CNT = CNT - 1

If CNT = 0;

PC = Address (CONTINUE)

If CNT \neq 0;

PC = Address (HERE+2)

DCFSNZ		Decrement f, skip if not 0							
Syntax:	[<i>label</i>] DCFSNZ f [,d [,a]]								
Operands:	$0 \leq f \leq 255$ $d \in [0,1]$ $a \in [0,1]$								
Operation:	$(f) - 1 \rightarrow \text{dest}$, skip if result $\neq 0$								
Status Affected:	None								
Encoding:	<table border="1"><tr><td>0100</td><td>11da</td><td>ffff</td><td>ffff</td></tr></table>					0100	11da	ffff	ffff
0100	11da	ffff	ffff						
Description:	<p>The contents of register 'f' are decremented. If 'd' is 0, the result is placed in W. If 'd' is 1, the result is placed back in register 'f' (default). If the result is not 0, the next instruction, which is already fetched, is discarded, and a NOP is executed instead, making it a two-cycle instruction. If 'a' is 0, the Access Bank will be selected, overriding the BSR value. If 'a' = 1, then the bank will be selected as per the BSR value (default).</p>								
Words:	1								
Cycles:	1(2)								
	Note: 3 cycles if skip and followed by a 2-word instruction.								

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write to destination

If skip:

Q1	Q2	Q3	Q4
No operation	No operation	No operation	No operation

If skip and followed by 2-word instruction:

Q1	Q2	Q3	Q4
No operation	No operation	No operation	No operation
No operation	No operation	No operation	No operation

Example:

```

HERE      DCFSNZ  TEMP, 1, 0
          ZERO    :
          NZERO   :
  
```

Before Instruction

TEMP = ?

After Instruction

TEMP = TEMP - 1,

If TEMP = 0;

PC = Address (ZERO)

If TEMP \neq 0;

PC = Address (NZERO)

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MULLW Multiply Literal with W

Syntax: [*label*] MULLW k

Operands: $0 \leq k \leq 255$

Operation: $(W) \times k \rightarrow \text{PRODH:PRODL}$

Status Affected: None

Encoding:

0000	1101	kkkk	kkkk
------	------	------	------

Description: An unsigned multiplication is carried out between the contents of W and the 8-bit literal 'k'. The 16-bit result is placed in PRODH:PRODL register pair. PRODH contains the high byte. W is unchanged. None of the status flags are affected. Note that neither overflow nor carry is possible in this operation. A zero result is possible but not detected.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read literal 'k'	Process Data	Write registers PRODH: PRODL

Example: MULLW 0xC4

Before Instruction

W = 0xE2
PRODH = ?
PRODL = ?

After Instruction

W = 0xE2
PRODH = 0xAD
PRODL = 0x08

MULWF Multiply W with f

Syntax: [*label*] MULWF f [,a]

Operands: $0 \leq f \leq 255$

$a \in [0,1]$

Operation: $(W) \times (f) \rightarrow \text{PRODH:PRODL}$

Status Affected: None

Encoding:

0000	001a	ffff	ffff
------	------	------	------

Description: An unsigned multiplication is carried out between the contents of W and the register file location 'f'. The 16-bit result is stored in the PRODH:PRODL register pair. PRODH contains the high byte. Both W and 'f' are unchanged. None of the status flags are affected. Note that neither overflow nor carry is possible in this operation. A zero result is possible but not detected. If 'a' is 0, the Access Bank will be selected, overriding the BSR value. If 'a' = 1, then the bank will be selected as per the BSR value (default).

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write registers PRODH: PRODL

Example: MULWF REG, 1

Before Instruction

W = 0xC4
REG = 0xB5
PRODH = ?
PRODL = ?

After Instruction

W = 0xC4
REG = 0xB5
PRODH = 0x8A
PRODL = 0x94

23.2 DC Characteristics: PIC18FXX39 (Industrial, Extended) PIC18LFXX39 (Industrial) (Continued)

DC CHARACTERISTICS			Standard Operating Conditions (unless otherwise stated) Operating temperature $-40^{\circ}\text{C} \leq T_A \leq +85^{\circ}\text{C}$ for industrial $-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$ for extended			
Param No.	Symbol	Characteristic	Min	Max	Units	Conditions
D080	VOL	Output Low Voltage I/O ports	—	0.6	V	$I_{OL} = 8.5\text{ mA}$, $V_{DD} = 4.5\text{V}$, -40°C to $+85^{\circ}\text{C}$
D080A			—	0.6	V	$I_{OL} = 7.0\text{ mA}$, $V_{DD} = 4.5\text{V}$, -40°C to $+125^{\circ}\text{C}$
D090	VOH	Output High Voltage⁽²⁾ I/O ports	$V_{DD} - 0.7$	—	V	$I_{OH} = -3.0\text{ mA}$, $V_{DD} = 4.5\text{V}$, -40°C to $+85^{\circ}\text{C}$
D090A			$V_{DD} - 0.7$	—	V	$I_{OH} = -2.5\text{ mA}$, $V_{DD} = 4.5\text{V}$, -40°C to $+125^{\circ}\text{C}$
D150	VOD	Open Drain High Voltage	—	8.5	V	RA4 pin
Capacitive Loading Specs on Output Pins						
D100 ⁽³⁾	COSC2	OSC2 pin	—	15	pF	In HS mode when external clock is used to drive OSC1
D101	CIO	All I/O pins	—	50	pF	To meet the AC Timing Specifications
D102	CB	SCL, SDA	—	400	pF	In I ² C mode

Note 1: The leakage current on the MCLR pin is strongly dependent on the applied voltage level. The specified levels represent normal operating conditions. Higher leakage current may be measured at different input voltages.

2: Negative current is defined as current sourced by the pin.

3: Parameter is characterized but not tested.

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23.3.3 TIMING DIAGRAMS AND SPECIFICATIONS

FIGURE 23-5: EXTERNAL CLOCK TIMING (ALL MODES EXCEPT PLL)

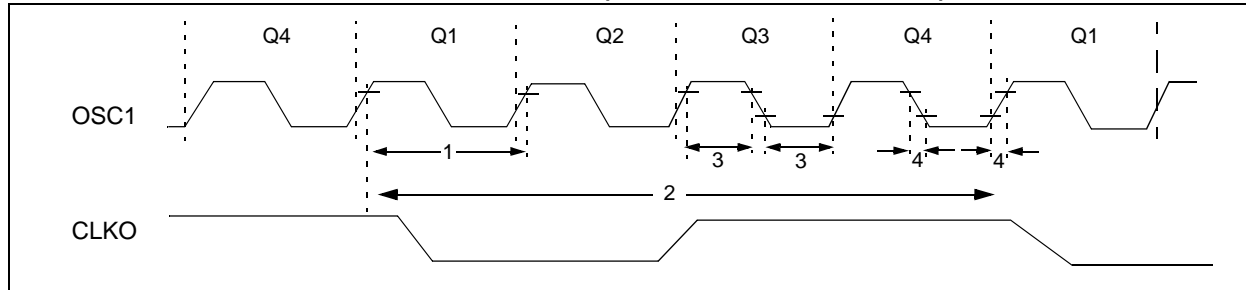


TABLE 23-4: EXTERNAL CLOCK TIMING REQUIREMENTS

Param. No.	Symbol	Characteristic	Min	Max	Units	Conditions
1A	FOSC	External CLKI Frequency ⁽¹⁾ Oscillator Frequency ⁽¹⁾	DC	40	MHz	EC, ECIO, -40°C to +85°C
			DC	25	MHz	EC, ECIO, +85°C to +125°C
			4	25	MHz	HS osc
			4	10	MHz	HS + PLL osc, -40°C to +85°C
			4	6.25	MHz	HS + PLL osc, +85°C to +125°C
1	TOSC	External CLKI Period ⁽¹⁾ Oscillator Period ⁽¹⁾	25	—	ns	EC, ECIO, -40°C to +85°C
			40	—	ns	EC, ECIO, +85°C to +125°C
			40	250	ns	HS osc
			100	250	ns	HS + PLL osc, -40°C to +85°C
			160	250	ns	HS + PLL osc, +85°C to +125°C
2	Tcy	Instruction Cycle Time ⁽¹⁾	100	—	ns	Tcy = 4/FOSC, -40°C to +85°C
			160	—	ns	Tcy = 4/FOSC, +85°C to +125°C
3	TosL, TosH	External Clock in (OSC1) High or Low Time	10	—	ns	HS osc
4	TosR, TosF	External Clock in (OSC1) Rise or Fall Time	—	7.5	ns	HS osc

Note 1: Instruction cycle period (Tcy) equals four times the input oscillator time-base period for all configurations except PLL. All specified values are based on characterization data for that particular oscillator type under standard operating conditions with the device executing code. Exceeding these specified limits may result in an unstable oscillator operation and/or higher than expected current consumption. All devices are tested to operate at “min.” values with an external clock applied to the OSC1/CLKI pin. When an external clock input is used, the “max.” cycle time limit is “DC” (no clock) for all devices.

TABLE 23-5: PLL CLOCK TIMING SPECIFICATIONS (VDD = 4.2 TO 5.5V)

Param No.	Sym	Characteristic	Min	Typ†	Max	Units	Conditions
—	FOSC	Oscillator Frequency Range	4	—	10	MHz	HS mode only
—	FSYS	On-Chip VCO System Frequency	16	—	40	MHz	HS mode only
—	t _{rc}	PLL Start-up Time (Lock Time)	—	—	2	ms	
—	ΔCLK	CLKO Stability (Jitter)	-2	—	+2	%	

† Data in “Typ” column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

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TABLE 23-22: A/D CONVERSION REQUIREMENTS

Param No.	Symbol	Characteristic		Min	Max	Units	Conditions
130	TAD	A/D clock period	PIC18FXXXX	1.6	20 ⁽⁴⁾	μs	TOSC based
			PIC18LFXXXX	2.0	6.0	μs	A/D RC mode
131	T _{CONV}	Conversion time (not including acquisition time) (Note 1)		11	12	TAD	
132	TACQ	Acquisition time (Note 2)		5	—	μs	VREF = VDD = 5.0V
				10	—	μs	VREF = VDD = 2.5V
135	T _{SWC}	Switching Time from convert → sample		—	(Note 3)		

Note 1: ADRES register may be read on the following T_{cy} cycle.

2: The time for the holding capacitor to acquire the “New” input voltage, when the new input value has not changed by more than 1 LSB from the last sampled voltage. The source impedance (*R_s*) on the input channels is 50Ω. See Section 18.0 for more information on acquisition time consideration.

3: On the next Q4 cycle of the device clock.

4: The time of the A/D clock period is dependent on the device frequency and the TAD clock divider.

FIGURE 24-19: A/D NON-LINEARITY vs. VREFH (VDD = VREFH, -40°C TO +125°C)

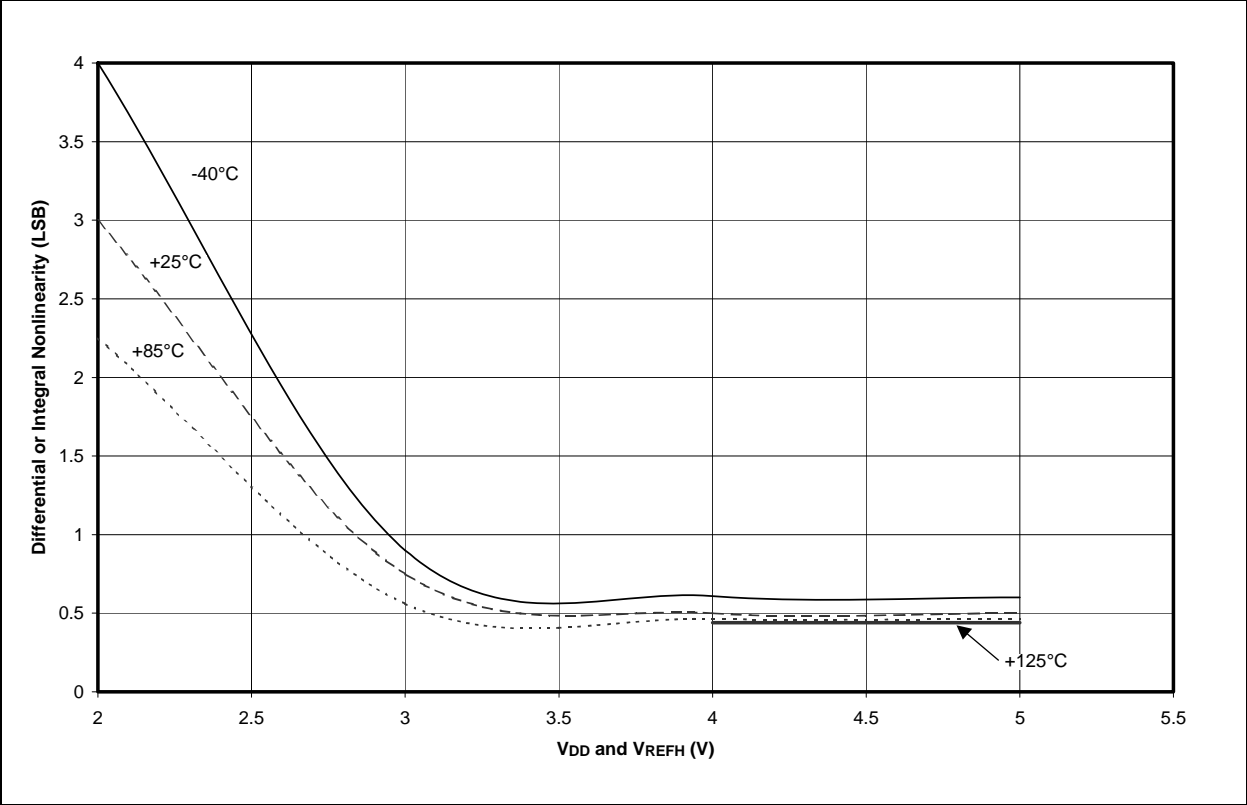
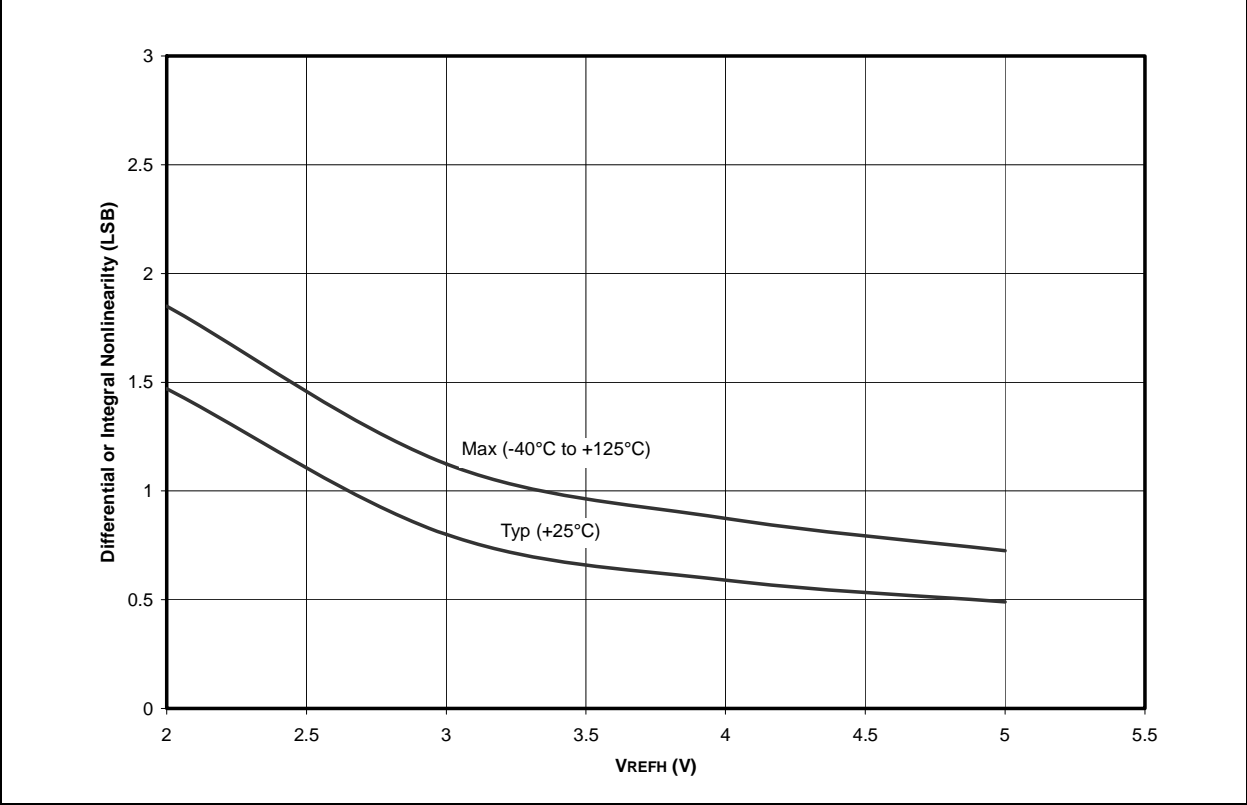


FIGURE 24-20: A/D NON-LINEARITY vs. VREFH (VDD = 5V, -40°C TO +125°C)



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