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"[Embedded - Microcontrollers](#)" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

### Applications of "[Embedded - Microcontrollers](#)"

#### Details

Product Status	Obsolete
Core Processor	ST7
Core Size	8-Bit
Speed	8MHz
Connectivity	LINbusSCI, SPI
Peripherals	LVD, POR, PWM, WDT
Number of I/O	32
Program Memory Size	48KB (48K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	2K x 8
Voltage - Supply (Vcc/Vdd)	3.8V ~ 5.5V
Data Converters	A/D 11x10b
Oscillator Type	External
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	44-LQFP
Supplier Device Package	-
Purchase URL	<a href="https://www.e-xfl.com/product-detail/stmicroelectronics/st72f361j7t3">https://www.e-xfl.com/product-detail/stmicroelectronics/st72f361j7t3</a>

## 6.2 MULTI-OSCILLATOR (MO)

The main clock of the ST7 can be generated by two different source types coming from the multi-oscillator block:

- an external source
- a crystal or ceramic resonator oscillator

Each oscillator is optimized for a given frequency range in terms of consumption and is selectable through the option byte. The associated hardware configuration are shown in [Table 6](#). Refer to the electrical characteristics section for more details.

**Caution:** The OSC1 and/or OSC2 pins must not be left unconnected. For the purposes of Failure Mode and Effect Analysis, it should be noted that if the OSC1 and/or OSC2 pins are left unconnected, the ST7 main oscillator may start and, in this configuration, could generate an  $f_{OSC}$  clock frequency in excess of the allowed maximum ( $> 16$  MHz), putting the ST7 in an unsafe/undefined state. The product behavior must therefore be considered undefined when the OSC pins are left unconnected.

### External Clock Source

In external clock mode, a clock signal (square, sinus or triangle) with ~50% duty cycle has to drive the OSC1 pin while the OSC2 pin is tied to ground.

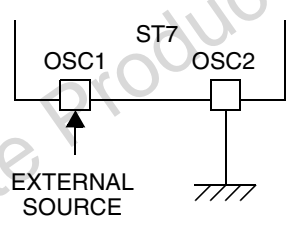
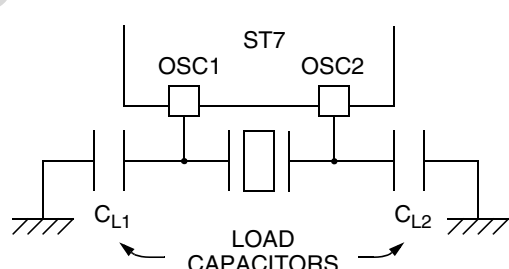
### Crystal/Ceramic Oscillators

This family of oscillators has the advantage of producing a very accurate rate on the main clock of the ST7. The selection within a list of five oscillators with different frequency ranges must be done by option byte in order to reduce consumption (refer to [Section 14.1 on page 210](#) for more details on

the frequency ranges). The resonator and the load capacitors must be placed as close as possible to the oscillator pins in order to minimize output distortion and start-up stabilization time. The loading capacitance values must be adjusted according to the selected oscillator.

These oscillators are not stopped during the RESET phase to avoid losing time in the oscillator start-up phase.

**Table 6. ST7 Clock Sources**

	Hardware Configuration
External Clock	
Crystal/Ceramic Resonators	

## SYSTEM INTEGRITY MANAGEMENT (Cont'd)

## 6.4.4 Register Description

## SYSTEM INTEGRITY (SI) CONTROL/STATUS REGISTER (SICSR)

Read/Write

Reset Value: 000x 000x (00h)

Bits 3:1 = Reserved, must be kept cleared.

7							0
0	AVD IE	AVD F	LVD RF	0	0	0	WDG RF

Bit 7 = Reserved, must be kept cleared.

Bit 6 = **AVDIE** *Voltage Detector interrupt enable*

This bit is set and cleared by software. It enables an interrupt to be generated when the AVDF flag changes (toggles). The pending interrupt information is automatically cleared when software enters the AVD interrupt routine.

0: AVD interrupt disabled

1: AVD interrupt enabled

Bit 5 = **AVDF** *Voltage Detector flag*

This read-only bit is set and cleared by hardware. If the AVDIE bit is set, an interrupt request is generated when the AVDF bit changes value. Refer to [Figure 16](#) and to [Section 6.4.2.1](#) for additional details.

0:  $V_{DD}$  over  $V_{IT+(AVD)}$  threshold1:  $V_{DD}$  under  $V_{IT-(AVD)}$  thresholdBit 4 = **LVDRF** *LVD reset flag*

This bit indicates that the last Reset was generated by the LVD block. It is set by hardware (LVD reset) and cleared by software (writing zero). See WDGRF flag description for more details. When the LVD is disabled by OPTION BYTE, the LVDRF bit value is undefined.

Bit 0 = **WDGRF** *Watchdog reset flag*

This bit indicates that the last Reset was generated by the Watchdog peripheral. It is set by hardware (watchdog reset) and cleared by software (writing zero) or an LVD Reset (to ensure a stable cleared state of the WDGRF flag when CPU starts).

Combined with the LVDRF flag information, the flag description is given by the following table.

RESET Sources	LVDRF	WDGRF
External RESET pin	0	0
Watchdog	0	1
LVD	1	X

## Application notes

The LVDRF flag is not cleared when another RESET type occurs (external or watchdog), the LVDRF flag remains set to keep trace of the original failure.

In this case, a watchdog reset can be detected by software while an external reset can not.

**CAUTION:** When the LVD is not activated with the associated option byte, the WDGRF flag can not be used in the application.

**INTERRUPTS (Cont'd)****7.6.2 Register Description****EXTERNAL INTERRUPT CONTROL****REGISTER 0 (EICR0)**

Read/Write

Reset Value: 0000 0000 (00h)

7							0
IS31	IS30	IS21	IS20	IS11	IS10	IS01	IS00

Bits 7:6 = **IS3[1:0]** *ei3 sensitivity*

The interrupt sensitivity, defined using the IS3[1:0] bits, is applied to the ei3 external interrupts:

IS31	IS30	External Interrupt Sensitivity
0	0	Falling edge & low level
0	1	Rising edge only
1	0	Falling edge only
1	1	Rising and falling edge

These 2 bits can be written only when I1 and I0 of the CC register are both set to 1 (level 3).

Bits 5:4 = **IS2[1:0]** *ei2 sensitivity*

The interrupt sensitivity, defined using the IS2[1:0] bits, is applied to the ei2 external interrupts:

IS21	IS20	External Interrupt Sensitivity
0	0	Falling edge & low level
0	1	Rising edge only
1	0	Falling edge only
1	1	Rising and falling edge

These 2 bits can be written only when I1 and I0 of the CC register are both set to 1 (level 3).

Bits 3:2 = **IS1[1:0]** *ei1 sensitivity*

The interrupt sensitivity, defined using the IS1[1:0] bits, is applied to the ei1 external interrupts:

IS11	IS10	External Interrupt Sensitivity
0	0	Falling edge & low level
0	1	Rising edge only
1	0	Falling edge only
1	1	Rising and falling edge

These 2 bits can be written only when I1 and I0 of the CC register are both set to 1 (level 3).

Bits 1:0 = **IS0[1:0]** *ei0 sensitivity*

The interrupt sensitivity, defined using the IS0[1:0] bits, is applied to the ei0 external interrupts:

IS01	IS00	External Interrupt Sensitivity
0	0	Falling edge & low level
0	1	Rising edge only
1	0	Falling edge only
1	1	Rising and falling edge

These 2 bits can be written only when I1 and I0 of the CC register are both set to 1 (level 3).

**EXTERNAL INTERRUPT CONTROL REGISTER 1 (EICR1)**

Read/Write

Reset Value: 0000 0000 (00h)

7							0
0	0	0	0	0	0	TLIS	TLIE

Bits 7:2 = Reserved

Bit 1 = **TLIS** *Top Level Interrupt sensitivity*

This bit configures the TLI edge sensitivity. It can be set and cleared by software only when TLIE bit is cleared.

0: Falling edge

1: Rising edge

Bit 0 = **TLIE** *Top Level Interrupt enable*

This bit allows to enable or disable the TLI capability on the dedicated pin. It is set and cleared by software.

0: TLI disabled

1: TLI enabled

**Notes:**

- A parasitic interrupt can be generated when clearing the TLIE bit.
- In some packages, the TLI pin is not available. In this case, the TLIE bit must be kept low to avoid parasitic TLI interrupts.

## INTERRUPTS (Cont'd)

Table 10. Nested Interrupts Register Map and Reset Values

Address (Hex.)	Register Label	7	6	5	4	3	2	1	0
0025h	<b>ISPR0</b> Reset Value	ei1		ei0		CLKM		TLI	
		I1_3 1	I0_3 1	I1_2 1	I0_2 1	I1_1 1	I0_1 1	1	1
0026h	<b>ISPR1</b> Reset Value					ei3		ei2	
		I1_7 1	I0_7 1	I1_6 1	I0_6 1	I1_5 1	I0_5 1	I1_4 1	I0_4 1
0027h	<b>ISPR2</b> Reset Value	LINSICI 2		TIMER 16		TIMER 8		SPI	
		I1_11 1	I0_11 1	I1_10 1	I0_10 1	I1_9 1	I0_9 1	I1_8 1	I0_8 1
0028h	<b>ISPR3</b> Reset Value					ART		LINSICI 1	
		1	1	1	1	I1_13 1	I0_13 1	I1_12 1	I0_12 1
0029h	<b>EICR0</b> Reset Value	IS31 0	IS30 0	IS21 0	IS20 0	IS11 0	IS10 0	IS01 0	IS00 0
002Ah	<b>EICR1</b> Reset Value	0	0	0	0	0	0	TLIS 0	TLIE 0

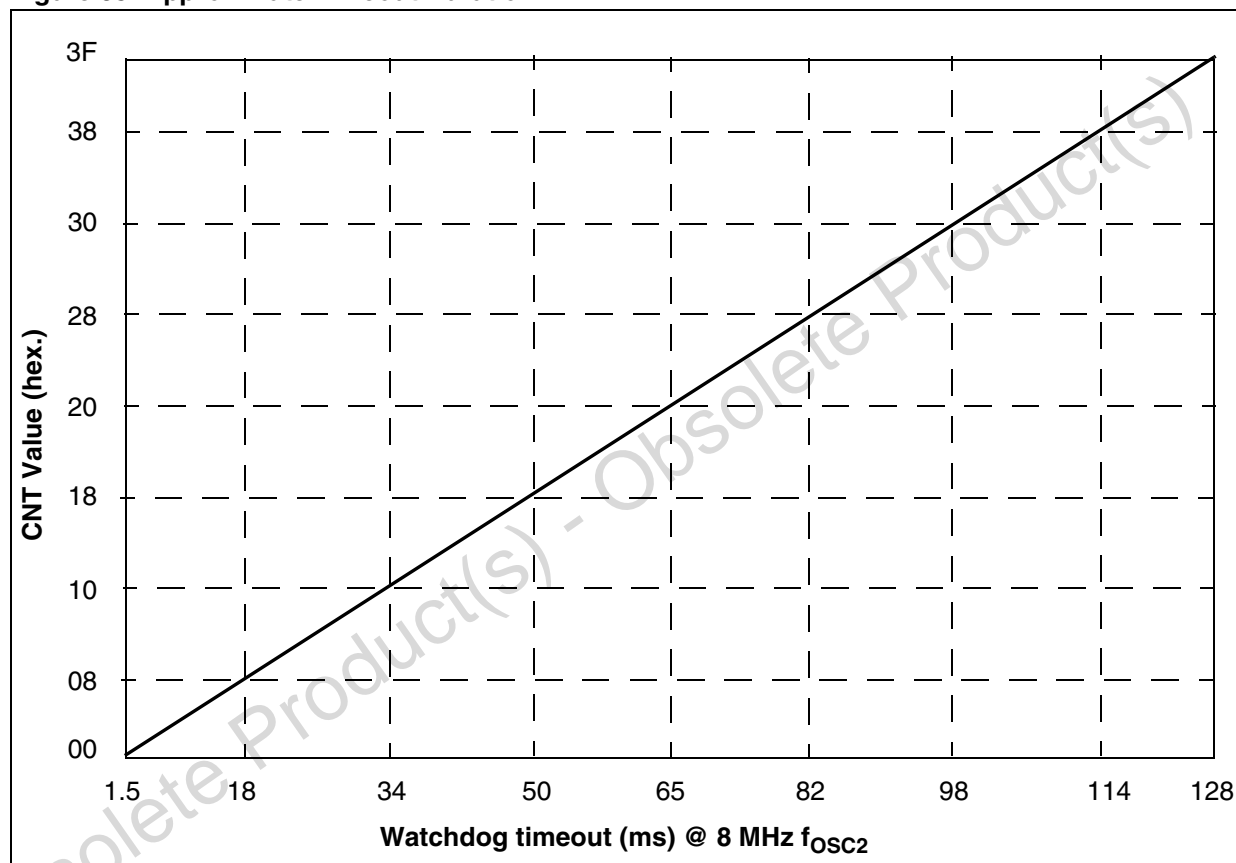
**WINDOW WATCHDOG (Cont'd)****10.1.5 How to Program the Watchdog Timeout**

Figure 2 shows the linear relationship between the 6-bit value to be loaded in the Watchdog Counter (CNT) and the resulting timeout duration in milliseconds. This can be used for a quick calculation without taking the timing variations into account. If

more precision is needed, use the formulae in Figure 3.

**Caution:** When writing to the WDGCR register, always write 1 in the T6 bit to avoid generating an immediate reset.

**Figure 35. Approximate Timeout Duration**



**16-BIT TIMER** (Cont'd)**10.4.4 Low Power Modes**

Mode	Description
WAIT	No effect on 16-bit Timer. Timer interrupts cause the device to exit from WAIT mode.
HALT	16-bit Timer registers are frozen. In HALT mode, the counter stops counting until Halt mode is exited. Counting resumes from the previous count when the MCU is woken up by an interrupt with "exit from HALT mode" capability or from the counter reset value when the MCU is woken up by a RESET. If an input capture event occurs on the ICAP <sub>i</sub> pin, the input capture detection circuitry is armed. Consequently, when the MCU is woken up by an interrupt with "exit from HALT mode" capability, the ICF <sub>i</sub> bit is set, and the counter value present when exiting from HALT mode is captured into the IC/R register.

**10.4.5 Interrupts**

Interrupt Event	Event Flag	Enable Control Bit	Exit from Wait	Exit from Halt
Input Capture 1 event/Counter reset in PWM mode	ICF1	ICIE	Yes	No
Input Capture 2 event	ICF2			
Output Compare 1 event (not available in PWM mode)	OCF1	OCIE		
Output Compare 2 event (not available in PWM mode)	OCF2			
Timer Overflow event	TOF	TOIE		

**Note:** The 16-bit Timer interrupt events are connected to the same interrupt vector (see Interrupts chapter). These events generate an interrupt if the corresponding Enable Control Bit is set and the interrupt mask in the CC register is reset (RIM instruction).

**10.4.6 Summary of Timer Modes**

MODES	TIMER RESOURCES			
	Input Capture 1	Input Capture 2	Output Compare 1	Output Compare 2
Input Capture (1 and/or 2)	Yes	Yes	Yes	Yes
Output Compare (1 and/or 2)				
One Pulse Mode	No	Not Recommended <sup>1)</sup>	No	Partially <sup>2)</sup>
PWM Mode		Not Recommended <sup>3)</sup>		No

1) See note 4 in [Section 10.4.3.5 "One Pulse Mode"](#)

2) See note 5 in [Section 10.4.3.5 "One Pulse Mode"](#)

3) See note 4 in [Section 10.4.3.6 "Pulse Width Modulation Mode"](#)

**16-BIT TIMER (Cont'd)****INPUT CAPTURE 1 HIGH REGISTER (IC1HR)**

Read Only

Reset Value: Undefined

This is an 8-bit read only register that contains the high part of the counter value (transferred by the input capture 1 event).

7							0
MSB							LSB

**OUTPUT COMPARE 1 HIGH REGISTER (OC1HR)**

Read/Write

Reset Value: 1000 0000 (80h)

This is an 8-bit register that contains the high part of the value to be compared to the CHR register.

7							0
MSB							LSB

**INPUT CAPTURE 1 LOW REGISTER (IC1LR)**

Read Only

Reset Value: Undefined

This is an 8-bit read only register that contains the low part of the counter value (transferred by the input capture 1 event).

7							0
MSB							LSB

**OUTPUT COMPARE 1 LOW REGISTER (OC1LR)**

Read/Write

Reset Value: 0000 0000 (00h)

This is an 8-bit register that contains the low part of the value to be compared to the CLR register.

7							0
MSB							LSB



## 10.5 8-BIT TIMER (TIM8)

### 10.5.1 Introduction

The timer consists of a 8-bit free-running counter driven by a programmable prescaler.

It may be used for a variety of purposes, including pulse length measurement of up to two input signals (*input capture*) or generation of up to two output waveforms (*output compare* and *PWM*).

Pulse lengths and waveform periods can be modulated from a few microseconds to several milliseconds using the timer prescaler and the clock prescaler.

### 10.5.2 Main Features

- Programmable prescaler:  $f_{CPU}$  divided by 2, 4, 8 or  $f_{OSC2}$  divided by 8000.
- Overflow status flag and maskable interrupt
- Output compare functions with
  - 2 dedicated 8-bit registers
  - 2 dedicated programmable signals
  - 2 dedicated status flags
  - 1 dedicated maskable interrupt
- Input capture functions with
  - 2 dedicated 8-bit registers
  - 2 dedicated active edge selection signals
  - 2 dedicated status flags
  - 1 dedicated maskable interrupt
- Pulse width modulation mode (PWM)
- One pulse mode
- Reduced Power Mode
- 4 alternate functions on I/O ports (ICAP1, ICAP2, OCMP1, OCMP2)\*

The Block Diagram is shown in [Figure 59](#).

**\*Note:** Some timer pins may not be available (not bonded) in some ST7 devices. Refer to the device pin out description.

When reading an input signal on a non-bonded pin, the value will always be '1'.

### 10.5.3 Functional Description

#### 10.5.3.1 Counter

The main block of the Programmable Timer is a 8-bit free running upcounter and its associated 8-bit registers.

These two read-only 8-bit registers contain the same value but with the difference that reading the ACTR register does not clear the TOF bit (Timer overflow flag), located in the Status register, (SR).

Writing in the CTR register or ACTR register resets the free running counter to the FCh value. Both counters have a reset value of FCh (this is the only value which is reloaded in the 8-bit timer). The reset value of both counters is also FCh in One Pulse mode and PWM mode.

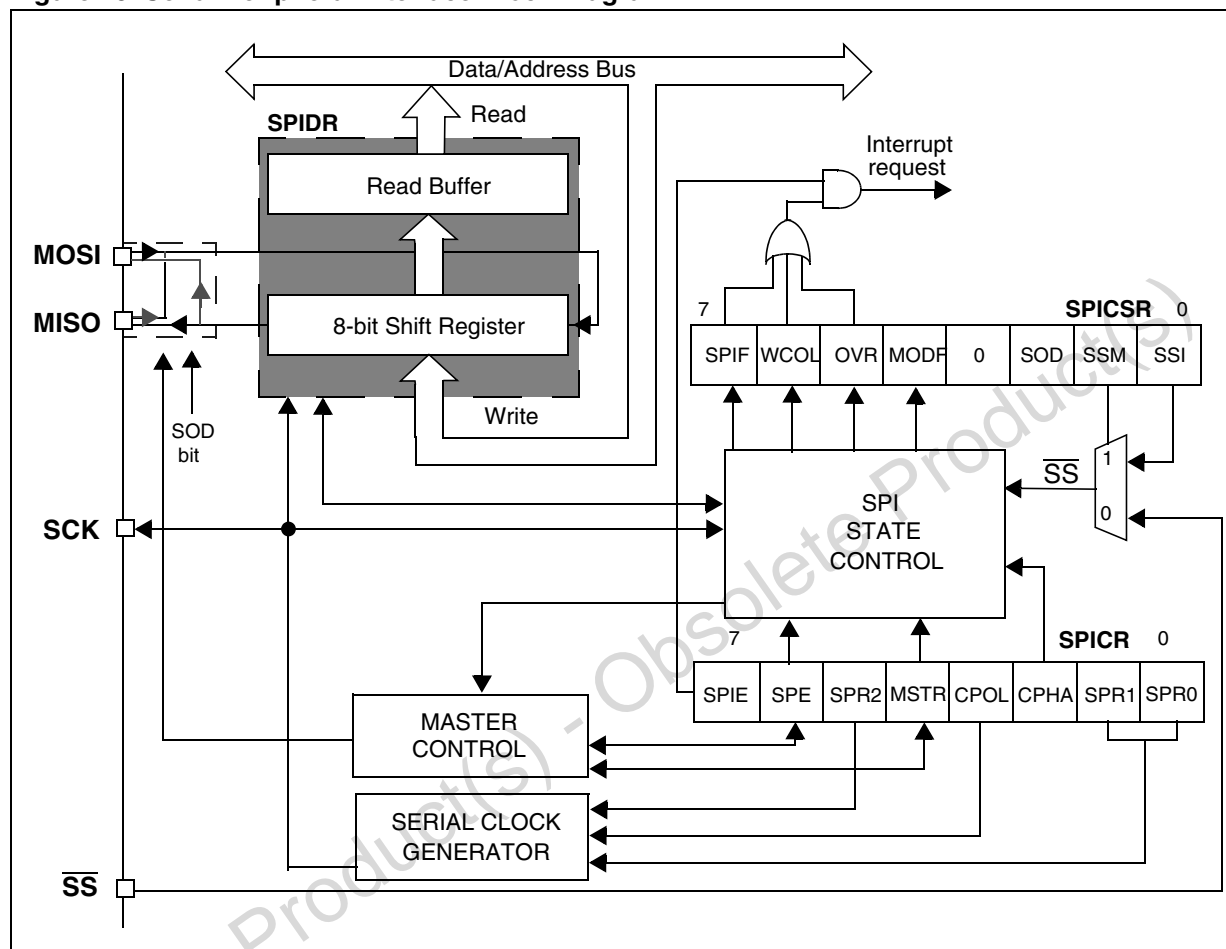
The timer clock depends on the clock control bits of the CR2 register, as shown in [Table 19 Clock Control Bits](#). The value in the counter register repeats every 512, 1024, 2048 or 20480000  $f_{CPU}$  clock cycles depending on the CC[1:0] bits.

The timer frequency can be  $f_{CPU}/2$ ,  $f_{CPU}/4$ ,  $f_{CPU}/8$  or  $f_{OSC2}/8000$ .

For example, if  $f_{OSC2}/8000$  is selected, and  $f_{OSC2} = 8$  MHz, the timer frequency will be 1 ms. Refer to [Table 19 on page 105](#).

## SERIAL PERIPHERAL INTERFACE (SPI) (cont'd)

Figure 70. Serial Peripheral Interface Block Diagram



## LINSI™ SERIAL COMMUNICATION INTERFACE (LIN Mode)

### 10.7.9 LIN Mode - Functional Description.

The block diagram of the Serial Control Interface, in LIN slave mode is shown in [Figure 5](#).

It uses six registers:

- 3 control registers: SCICR1, SCICR2 and SCICR3
- 2 status registers: the SCISR register and the LHLR register mapped at the SCIERPR address
- A baud rate register: LPR mapped at the SCIBRR address and an associated fraction register LPRF mapped at the SCIETPR address

The bits dedicated to LIN are located in the SCICR3. Refer to the register descriptions in [Section 0.1.10](#) for the definitions of each bit.

#### 10.7.9.1 Entering LIN Mode

To use the LINSI in LIN mode the following configuration must be set in SCICR3 register:

- Clear the M bit to configure 8-bit word length.
- Set the LINE bit.

#### Master

To enter master mode the LSLV bit must be reset. In this case, setting the SBK bit will send 13 low bits.

Then the baud rate can be programmed using the SCIBRR, SCIERPR and SCIETPR registers.

In LIN master mode, the Conventional and / or Extended Prescaler define the baud rate (as in standard SCI mode)

#### Slave

Set the LSLV bit in the SCICR3 register to enter LIN slave mode. In this case, setting the SBK bit will have no effect.

In LIN Slave mode the LIN baud rate generator is selected instead of the Conventional or Extended Prescaler. The LIN baud rate generator is common to the transmitter and the receiver.

Then the baud rate can be programmed using LPR and LPRF registers.

**Note:** It is mandatory to set the LIN configuration first before programming LPR and LPRF, because the LIN configuration uses a different baud rate generator from the standard one.

#### 10.7.9.2 LIN Transmission

In LIN mode the same procedure as in SCI mode has to be applied for a LIN transmission.

To transmit the LIN Header the procedure is as follows:

- First set the SBK bit in the SCICR2 register to start transmitting a 13-bit LIN Synch Break
- reset the SBK bit
- Load the LIN Synch Field (0x55) in the SCIDR register to request Synch Field transmission
- Wait until the SCIDR is empty (TDRE bit set in the SCISR register)
- Load the LIN message Identifier in the SCIDR register to request Identifier transmission.

**LINSPI™ SERIAL COMMUNICATION INTERFACE (LIN Mode) (cont'd)****10.7.9.9 Error due to LIN Synch measurement**

The LIN Synch Field is measured over eight bit times.

This measurement is performed using a counter clocked by the CPU clock. The edge detections are performed using the CPU clock cycle.

This leads to a precision of 2 CPU clock cycles for the measurement which lasts  $16 \cdot 8 \cdot \text{LDIV}$  clock cycles.

Consequently, this error ( $D_{\text{MEAS}}$ ) is equal to:

$$2 / (128 \cdot \text{LDIV}_{\text{MIN}}).$$

$\text{LDIV}_{\text{MIN}}$  corresponds to the minimum LIN prescaler content, leading to the maximum baud rate, taking into account the maximum deviation of +/-15%.

**10.7.9.10 Error due to Baud Rate Quantization**

The baud rate can be adjusted in steps of  $1 / (16 \cdot \text{LDIV})$ . The worst case occurs when the "real" baud rate is in the middle of the step.

This leads to a quantization error ( $D_{\text{QUANT}}$ ) equal to  $1 / (2 \cdot 16 \cdot \text{LDIV}_{\text{MIN}})$ .

**10.7.9.11 Impact of Clock Deviation on Maximum Baud Rate**

The choice of the nominal baud rate ( $\text{LDIV}_{\text{NOM}}$ ) will influence both the quantization error ( $D_{\text{QUANT}}$ ) and the measurement error ( $D_{\text{MEAS}}$ ). The worst case occurs for  $\text{LDIV}_{\text{MIN}}$ .

Consequently, at a given CPU frequency, the maximum possible nominal baud rate ( $\text{LPR}_{\text{MIN}}$ ) should be chosen with respect to the maximum tolerated deviation given by the equation:

$$D_{\text{TRA}} + 2 / (128 \cdot \text{LDIV}_{\text{MIN}}) + 1 / (2 \cdot 16 \cdot \text{LDIV}_{\text{MIN}}) + D_{\text{REC}} + D_{\text{TCL}} < 3.75\%$$

Example:

A nominal baud rate of 20Kbits/s at  $T_{\text{CPU}} = 125\text{ns}$  (8 MHz) leads to  $\text{LDIV}_{\text{NOM}} = 25\text{d}$ .

$$\text{LDIV}_{\text{MIN}} = 25 - 0.15 \cdot 25 = 21.25$$

$$D_{\text{MEAS}} = 2 / (128 \cdot \text{LDIV}_{\text{MIN}}) \cdot 100 = 0.00073\%$$

$$D_{\text{QUANT}} = 1 / (2 \cdot 16 \cdot \text{LDIV}_{\text{MIN}}) \cdot 100 = 0.0015\%$$

**LIN Slave systems**

For LIN Slave systems (the LINE and LSLV bits are set), receivers wake up by LIN Synch Break or LIN Identifier detection (depending on the LHDM bit).

**Hot Plugging Feature for LIN Slave Nodes**

In LIN Slave Mute Mode (the LINE, LSLV and RWU bits are set) it is possible to hot plug to a network during an ongoing communication flow. In this case the SCI monitors the bus on the RDI line until 11 consecutive dominant bits have been detected and discards all the other bits received.

**10-BIT A/D CONVERTER (ADC) (Cont'd)****10.9.6 Register Description****CONTROL/STATUS REGISTER (ADCCSR)**

Read/Write (Except bit 7 read only)

Reset Value: 0000 0000 (00h)

7							0
EOC	SPEED	ADON	SLOW	CH3	CH2	CH1	CH0

Bit 7 = **EOC** *End of Conversion*

This bit is set by hardware. It is cleared by software reading the ADCDRH register or writing to any bit of the ADCCSR register.

0: Conversion is not complete

1: Conversion complete

Bit 6 = **SPEED** *A/D clock selection*

This bit is set and cleared by software.

**Table 29. A/D Clock Selection**

$f_{ADC}$	SLOW	SPEED
$f_{CPU}/2$	0	0
$f_{CPU}$ (where $f_{CPU} \leq 4$ MHz)	0	1
$f_{CPU}/4$	1	0
$f_{CPU}/2$ (same frequency as SLOW=0, SPEED=0)	1	1

Bit 5 = **ADON** *A/D Converter on*

This bit is set and cleared by software.

0: Disable ADC and stop conversion

1: Enable ADC and start conversion

Bit 4 = **SLOW** *A/D Clock Selection*This bit is set and cleared by software. It works together with the SPEED bit. Refer to [Table 29](#).Bits 3:0 = **CH[3:0]** *Channel Selection*

These bits are set and cleared by software. They select the analog input to convert.

\*The number of channels is device dependent. Refer to the device pinout description.

Channel Pin*	CH3	CH2	CH1	CH0
AIN0	0	0	0	0
AIN1	0	0	0	1
AIN2	0	0	1	0
AIN3	0	0	1	1
AIN4	0	1	0	0
AIN5	0	1	0	1
AIN6	0	1	1	0
AIN7	0	1	1	1
AIN8	1	0	0	0
AIN9	1	0	0	1
AIN10	1	0	1	0
AIN11	1	0	1	1
AIN12	1	1	0	0
AIN13	1	1	0	1
AIN14	1	1	1	0
AIN15	1	1	1	1

**DATA REGISTER (ADCDRH)**

Read Only

Reset Value: 0000 0000 (00h)

7							0
D9	D8	D7	D6	D5	D4	D3	D2

Bits 7:0 = **D[9:2]** *MSB of Analog Converted Value***DATA REGISTER (ADCDRL)**

Read Only

Reset Value: 0000 0000 (00h)

7							0
0	0	0	0	0	0	D1	D0

Bits 7:2 = Reserved. Forced by hardware to 0.

Bits 1:0 = **D[1:0]** *LSB of Analog Converted Value*

## 12.4 SUPPLY CURRENT CHARACTERISTICS

The following current consumption specified for the ST7 functional operating modes over temperature range does not take into account the clock source current consumption. To get the total de-

vice consumption, the two current values must be added (except for HALT mode for which the clock is stopped).

Symbol	Parameter	Conditions	Flash Devices		ROM Devices		Unit
			Typ <sup>1)</sup>	Max <sup>2)</sup>	Typ <sup>1)</sup>	Max <sup>2)</sup>	
I <sub>DD</sub>	Supply current in RUN mode <sup>3)</sup>	f <sub>OSC</sub> = 2 MHz, f <sub>CPU</sub> = 1 MHz f <sub>OSC</sub> = 4 MHz, f <sub>CPU</sub> = 2 MHz f <sub>OSC</sub> = 8 MHz, f <sub>CPU</sub> = 4 MHz f <sub>OSC</sub> = 16 MHz, f <sub>CPU</sub> = 8 MHz	1.8 3.2 6 10	3 5 8 15	1.1 2.2 4.4 8.9	2 3.5 6 12	mA
	Supply current in SLOW mode <sup>3)</sup>	f <sub>OSC</sub> = 2 MHz, f <sub>CPU</sub> = 62.5 kHz f <sub>OSC</sub> = 4 MHz, f <sub>CPU</sub> = 125 kHz f <sub>OSC</sub> = 8 MHz, f <sub>CPU</sub> = 250 kHz f <sub>OSC</sub> = 16 MHz, f <sub>CPU</sub> = 500 kHz	0.5 0.6 0.85 1.25	2.7 3 3.6 4	0.1 0.2 0.4 0.8	0.2 0.4 0.8 1.5	
	Supply current in WAIT mode <sup>3)</sup>	f <sub>OSC</sub> = 2 MHz, f <sub>CPU</sub> = 1 MHz f <sub>OSC</sub> = 4 MHz, f <sub>CPU</sub> = 2 MHz f <sub>OSC</sub> = 8 MHz, f <sub>CPU</sub> = 4 MHz f <sub>OSC</sub> = 16 MHz, f <sub>CPU</sub> = 8 MHz	1 1.8 3.4 6.4	3 4 5 7	0.7 1.4 2.9 5.7	3 4 5 7	
	Supply current in SLOW WAIT mode <sup>2)</sup>	f <sub>OSC</sub> = 2 MHz, f <sub>CPU</sub> = 62.5 kHz f <sub>OSC</sub> = 4 MHz, f <sub>CPU</sub> = 125 kHz f <sub>OSC</sub> = 8 MHz, f <sub>CPU</sub> = 250 kHz f <sub>OSC</sub> = 16 MHz, f <sub>CPU</sub> = 500 kHz	0.4 0.5 0.6 0.8	1.2 1.3 1.8 2	0.07 0.14 0.28 0.56	0.12 0.25 0.5 1	
	Supply current in HALT mode <sup>4)</sup>	V <sub>DD</sub> = 5.5V	<1	10	<1	10	μA
		-40°C ≤ T <sub>A</sub> ≤ +85°C -40°C ≤ T <sub>A</sub> ≤ +125°C		50		50	
	Supply current in ACTIVE HALT mode <sup>4)5)</sup>		0.5	1.2	0.18	0.25	mA
	Supply current in AWUFH mode <sup>4)5)</sup>	V <sub>DD</sub> = 5.5V	25	30	25	30	μA
		-40°C ≤ T <sub>A</sub> ≤ +85°C -40°C ≤ T <sub>A</sub> ≤ +125°C		70		70	

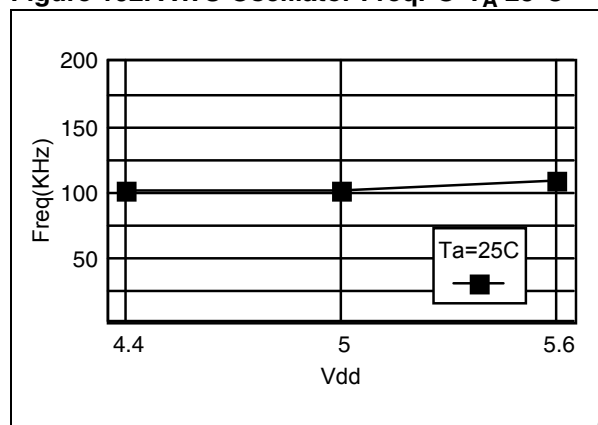
### Notes:

- Typical data are based on T<sub>A</sub> = 25°C, V<sub>DD</sub> = 5V (4.5V ≤ V<sub>DD</sub> ≤ 5.5V range).
- Data based on characterization results, tested in production at V<sub>DD</sub> max., f<sub>CPU</sub> max. and T<sub>A</sub> max.
- Measurements are done in the following conditions:
  - Program executed from Flash, CPU running with Flash (for flash devices).
  - All I/O pins in input mode with a static value at V<sub>DD</sub> or V<sub>SS</sub> (no load)
  - All peripherals in reset state.
  - Clock input (OSC1) driven by external square wave.
  - In SLOW and SLOW WAIT mode, f<sub>CPU</sub> is based on f<sub>OSC</sub> divided by 32.
- To obtain the total current consumption of the device, add the clock source (Section 12.5.3) and the peripheral power consumption (Section 12.4.2).
- All I/O pins in input mode with a static value at V<sub>DD</sub> or V<sub>SS</sub> (no load). Data based on characterization results, tested in production at V<sub>DD</sub> max., f<sub>CPU</sub> max. and T<sub>A</sub> max.
- This consumption refers to the Halt period only and not the associated run period which is software dependent.

**CLOCK CHARACTERISTICS** (Cont'd)**12.6 Auto Wakeup from Halt Oscillator (AWU)**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$f_{AWU}$	AWU oscillator frequency <sup>1)</sup>		50	100	250	kHz
$t_{RCSRT}$	AWU oscillator startup time			10		$\mu s$

1. Data based on characterization results, not tested in production.

**Figure 102. AWU Oscillator Freq. @  $T_A$  25°C**

**EMC CHARACTERISTICS (Cont'd)****12.8.3 Absolute Maximum Ratings (Electrical Sensitivity)**

Based on two different tests (ESD and LU) using specific measurement methods, the product is stressed in order to determine its performance in terms of electrical sensitivity (see [Table 33](#) and [Table 34](#) below). For more details, refer to application note AN1181.

**12.8.3.1 Electro-Static Discharge (ESD)**

Electro-Static Discharges (a positive then a negative pulse separated by 1 second) are applied to the pins of each sample according to each pin combination. The sample size depends on the

number of supply pins in the device (3 parts\*(n+1) supply pin). Two models can be simulated: Human Body Model and Machine Model. This test conforms to the JESD22-A114A/A115A standard.

**12.8.3.2 Static Latch-Up**

- **LU:** Two complementary static tests are required on 6 parts to assess the latch-up performance. A supply overvoltage (applied to each power supply pin) and a current injection (applied to each input, output and configurable I/O pin) are performed on each sample. This test conforms to the EIA/JESD 78 IC latch-up standard.

**Table 33. Absolute Maximum Ratings**

Symbol	Ratings	Conditions	Maximum value <sup>1)</sup>	Unit
V <sub>ESD(HBM)</sub>	Electro-static discharge voltage (Human Body Model)	T <sub>A</sub> = +25°C	2000	V
V <sub>ESD(MM)</sub>	Electro-static discharge voltage (Machine Model)		200	
V <sub>ESD(CDM)</sub>	Electro-static discharge voltage (Charge Device Model)		750 on corner pins, 500 on others	

**Notes:**

1. Data based on characterization results, not tested in production.

**Table 34. Electrical Sensitivities**

Symbol	Parameter	Conditions	Class
LU	Static latch-up class	T <sub>A</sub> =+125°C conforming to JESD 78	II level A

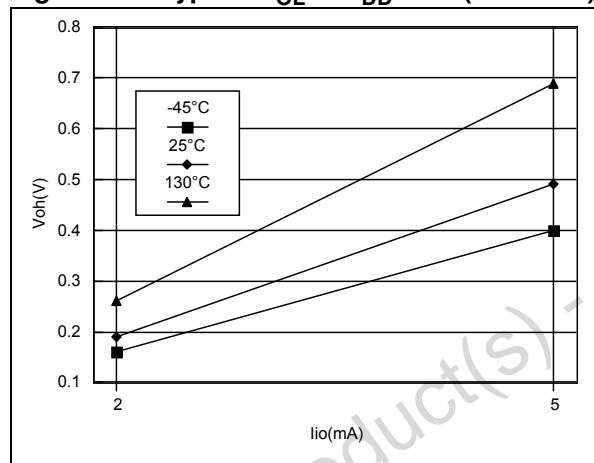
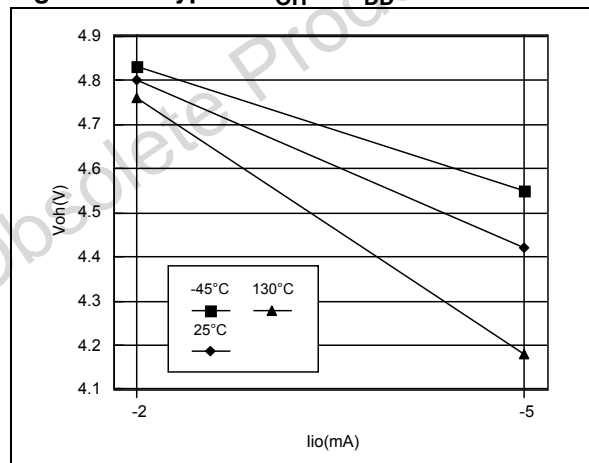
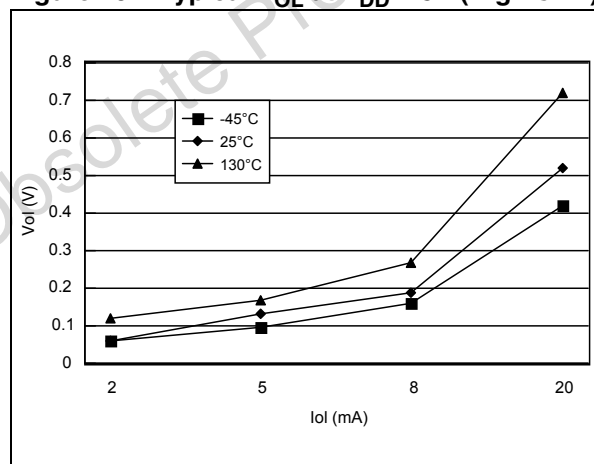


## I/O PORT PIN CHARACTERISTICS (Cont'd)

## 12.9.2 Output Driving Current

Subject to general operating conditions for  $V_{DD}$ ,  $f_{OSC}$ , and  $T_A$  unless otherwise specified.

Symbol	Parameter	Conditions	Min	Max	Unit
$V_{OL}^{1)}$	Output low level voltage for a standard I/O pin when eight pins are sunk at same time (see Figure 106)	$I_{IO}=+5\text{mA}$		1.2	V
		$I_{IO}=+2\text{mA}$		0.5	
	Output low level voltage for a high sink I/O pin when four pins are sunk at same time (see Figure 107 and Figure 110)	$I_{IO}=+20\text{mA}, T_A \leq 85^\circ\text{C}$		1.3	
		$T_A \geq 85^\circ\text{C}$		1.5	
$V_{OH}^{2)}$	Output high level voltage for an I/O pin when four pins are sourced at same time (see Figure 108 and Figure 111)	$I_{IO}=+8\text{mA}$		0.6	V
		$I_{IO}=-5\text{mA}, T_A \leq 85^\circ\text{C}$	$V_{DD}-1.4$		
		$T_A \geq 85^\circ\text{C}$	$V_{DD}-1.6$		
		$I_{IO}=-2\text{mA}$	$V_{DD}-0.7$		

Figure 106. Typical  $V_{OL}$  at  $V_{DD} = 5\text{V}$  (Standard)Figure 108. Typical  $V_{OH}$  at  $V_{DD} = 5\text{V}$ Figure 107. Typical  $V_{OL}$  at  $V_{DD} = 5\text{V}$  (High-sink)

## Notes:

1. The  $I_{IO}$  current sunk must always respect the absolute maximum rating specified in Section 12.2.2 and the sum of  $I_{IO}$  (I/O ports and control pins) must not exceed  $I_{VSS}$ .
2. The  $I_{IO}$  current sourced must always respect the absolute maximum rating specified in Section 12.2.2 and the sum of  $I_{IO}$  (I/O ports and control pins) must not exceed  $I_{VDD}$ . True open drain I/O pins does not have  $V_{OH}$ .

## 14 DEVICE CONFIGURATION AND ORDERING INFORMATION

Each device is available for production in user programmable versions (FLASH) as well as in factory coded versions (ROM/FASTROM).

ST72361 devices are ROM versions. ST72P361 devices are Factory Advanced Service Technique ROM (FASTROM) versions: They are factory-programmed HDFSFlash devices.

ST72F361 FLASH devices are shipped to customers with a default content (FFh), while ROM factory coded parts contain the code supplied by the customer. This implies that FLASH devices have to be configured by the customer using the Option Bytes while the ROM devices are factory-configured.

### 14.1 FLASH OPTION BYTES

The option bytes allows the hardware configuration of the microcontroller to be selected. They have no address in the memory map and can be accessed only in programming mode (for example using a standard ST7 programming tool). The default content of the FLASH is fixed to FFh. To program directly the FLASH devices using ICP, FLASH devices are shipped to customers with a reserved internal clock source enabled. In masked ROM devices, the option bytes are fixed in hardware by the ROM code (see option list).

#### OPTION BYTE 0

**OPT7 = WDGHALT** Watchdog reset on HALT  
This option bit determines if a RESET is generated when entering HALT mode while the Watchdog is

active.

0: No Reset generation when entering Halt mode

1: Reset generation when entering Halt mode

**OPT6 = WDGSW** Hardware or software watchdog

This option bit selects the watchdog type.

0: Hardware (watchdog always enabled)

1: Software (watchdog to be enabled by software)

**OPT5** = Reserved, must be kept at default value.

**OPT4 = LVD** Voltage detection

This option bit enables the voltage detection block (LVD).

Selected Low Voltage Detector	VD
LVD Off	1
LVD On	0

**OPT3 = PLL OFF** PLL activation

This option bit activates the PLL which allows multiplication by two of the main input clock frequency. The PLL is guaranteed only with an input frequency between 2 and 4 MHz.

0: PLL x2 enabled

1: PLL x2 disabled

**Caution:** The PLL can be enabled only if the "OSC RANGE" (OPT11:10) bits are configured to "MP - 2~4 MHz". Otherwise, the device functionality is not guaranteed.

	STATIC OPTION BYTE 0								STATIC OPTION BYTE 1							
	7		Reserved	LVD	PLLOFF	PKG		FMP_R	7		OSCTYPE		OSCRANGE		Reserved	RSTC
	HALT	SW				1	0		1	0	1	0	1	0		
De-fault(*)	1	1	1	1	1	1	1	1	1	1	1	0	1	1	1	1

(\*): Option bit values programmed by ST

**FLASH OPTION BYTES (Cont'd)****OPT2:1 = PKG[1:0] Package selection**

These option bits select the device package.

Selected Package	PKG	
	1	0
LQFP 64	1	x
LQFP 44	0	1
LQFP 32	0	0

**Note:** Pads that are not bonded to external pins are in input pull-up configuration when the package selection option bits have been properly programmed. The configuration of these pads must be kept in reset state to avoid added current consumption.

**OPT0 = FMP\_R Flash memory read-out protection**

Read-out protection, when selected provides a protection against program memory content extraction and against write access to Flash memory. Erasing the option bytes when the FMP\_R option is selected causes the whole user memory to be erased first, and the device can be reprogrammed. Refer to [Section 4.3.1](#) and the *ST7 Flash Programming Reference Manual* for more details.

0: Read-out protection enabled

1: Read-out protection disabled

**OPTION BYTE 1****OPT7:6 = AFI\_MAP[1:0] AFI Mapping**

These option bits allow the mapping of some of the Alternate Functions to be changed.

AFI Mapping 1	AFI_MAP(1)
T16_OCMP1 on PD3 T16_OCMP2 on PD5 T16_ICAP1 on PD4 LINSI2_SCK not available LINSI2_TDO not available LINSI2_RDI not available	0
T16_OCMP1 on PB6 T16_OCMP2 on PB7 T16_ICAP1 on PC0 LINSI2_SCK on PD3 LINSI2_TDO on PD5 LINSI2_RDI on PD4	1

AFI Mapping 0	AFI_MAP(0)
T16_ICAP2 is mapped on PD1	0
T16_ICAP2 is mapped on PC1	1

**OPT5:4 = OSCTYPE[1:0] Oscillator Type**

These option bits select the ST7 main clock source type.

Clock Source	OSCTYPE	
	1	0
Resonator Oscillator	0	0
Reserved	0	1
Reserved internal clock source (used only in ICC mode)	1	0
External Source	1	1

**OPT3:2 = OSCRANGE[1:0] Oscillator range**

If the resonator oscillator type is selected, these option bits select the resonator oscillator. This selection corresponds to the frequency range of the resonator used. If external source is selected with the OSCTYPE option, then the OSCRANGE option must be selected with the corresponding range.

Typ. Freq. Range		OSCRANGE	
		1	0
LP	1~2 MHz	0	0
MP	2~4 MHz	0	1
MS	4~8 MHz	1	0
HS	8~16 MHz	1	1

OPT1 = Reserved

**OPT0 = RSTC RESET clock cycle selection**

This option bit selects the number of CPU cycles inserted during the RESET phase and when exiting HALT mode. For resonator oscillators, it is advised to select 4096 due to the long crystal stabilization time.

0: Reset phase with 4096 CPU cycles

1: Reset phase with 256 CPU cycles

## 16 IMPORTANT NOTES

### 16.1 ALL DEVICES

#### 16.1.1 RESET Pin Protection with LVD Enabled

As mentioned in note 2 below [Figure 112 on page 199](#), when the LVD is enabled, it is recommended not to connect a pull-up resistor or capacitor. A 10nF pull-down capacitor is required to filter noise on the reset line.

#### 16.1.2 Clearing Active Interrupts Outside Interrupt Routine

When an active interrupt request occurs at the same time as the related flag or interrupt mask is being cleared, the CC register may be corrupted.

##### Concurrent interrupt context

The symptom does not occur when the interrupts are handled normally, that is, when:

- The interrupt request is cleared (flag reset or interrupt mask) within its own interrupt routine
- The interrupt request is cleared (flag reset or interrupt mask) within any interrupt routine
- The interrupt request is cleared (flag reset or interrupt mask) in any part of the code while this interrupt is disabled

If these conditions are not met, the symptom can be avoided by implementing the following sequence:

Perform SIM and RIM operation before and after resetting an active interrupt request

Example:

SIM

reset flag or interrupt mask

RIM

##### Nested interrupt context

The symptom does not occur when the interrupts are handled normally, that is, when:

- The interrupt request is cleared (flag reset or interrupt mask) within its own interrupt routine
- The interrupt request is cleared (flag reset or interrupt mask) within any interrupt routine with higher or identical priority level
- The interrupt request is cleared (flag reset or interrupt mask) in any part of the code while this interrupt is disabled

If these conditions are not met, the symptom can be avoided by implementing the following sequence:

PUSH CC

SIM

reset flag or interrupt mask

POP CC

#### 16.1.3 External Interrupt Missed

To avoid any risk of generating a parasitic interrupt, the edge detector is automatically disabled for one clock cycle during an access to either DDR and OR. Any input signal edge during this period will not be detected and will not generate an interrupt.

This case can typically occur if the application refreshes the port configuration registers at intervals during runtime.

##### Workaround

The workaround is based on software checking the level on the interrupt pin before and after writing to the PxOR or PxDDR registers. If there is a level change (depending on the sensitivity programmed for this pin) the interrupt routine is invoked using the call instruction with three extra PUSH instructions before executing the interrupt routine (this is to make the call compatible with the IRET instruction at the end of the interrupt service routine).

But detection of the level change does ensure that edge occurs during the critical 1 cycle duration and the interrupt has been missed. This may lead to occurrence of same interrupt twice (one hardware and another with software call).

To avoid this, a semaphore is set to '1' before checking the level change. The semaphore is changed to level '0' inside the interrupt routine. When a level change is detected, the semaphore status is checked and if it is '1' this means that the last interrupt has been missed. In this case, the interrupt routine is invoked with the call instruction.

There is another possible case, that is, if writing to PxOR or PxDDR is done with global interrupts disabled (interrupt mask bit set). In this case, the semaphore is changed to '1' when the level change is detected. Detecting a missed interrupt is done after the global interrupts are enabled (interrupt mask bit reset) and by checking the status of the semaphore. If it is '1' this means that the last interrupt was missed and the interrupt routine is invoked with the call instruction.

To implement the workaround, the following software sequence is to be followed for writing into the PxOR/PxDDR registers. The example is for Port PF1 with falling edge interrupt sensitivity. The

**IMPORTANT NOTES (Cont'd)****16.1.5 Header Time-out Does Not Prevent Wake-up from Mute Mode**

Normally, when LINSPI is configured in LIN slave mode, if a header time-out occurs during a LIN header reception (that is, header length > 57 bits), the LIN Header Error bit (LHE) is set, an interrupt occurs to inform the application but the LINSPI should stay in mute mode, waiting for the next header reception.

**Problem Description**

The LINSPI sampling period is  $T_{bit} / 16$ . If a LIN Header time-out occurs between the 9th and the 15th sample of the Identifier Field Stop Bit (refer to [Figure 130](#)), the LINSPI wakes up from mute mode. Nevertheless, LHE is set and LIN Header Detection Flag (LHDF) is kept cleared.

In addition, if LHE is reset by software before this 15th sample (by accessing the SCISR register and

reading the SCIDR register in the LINSPI interrupt routine), the LINSPI will generate another LINSPI interrupt (due to the RDRF flag setting).

**Impact on application**

Software may execute the interrupt routine twice after header reception.

Moreover, in reception mode, as the receiver is no longer in mute mode, an interrupt will be generated on each data byte reception.

**Workaround**

The problem can be detected in the LINSPI interrupt routine. In case of timeout error (LHE is set and LHLR is loaded with 00h), the software can check the RWU bit in the SCICR2 register. If RWU is cleared, it can be set by software. Refer to [Figure 131 on page 221](#). Workaround is shown in bold characters.

**Figure 130. Header Reception Event Sequence**