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## Details

Product Status	Obsolete
Core Processor	ST7
Core Size	8-Bit
Speed	8MHz
Connectivity	LINbusSCI, SPI
Peripherals	LVD, POR, PWM, WDT
Number of I/O	32
Program Memory Size	60KB (60K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	2K x 8
Voltage - Supply (Vcc/Vdd)	3.8V ~ 5.5V
Data Converters	A/D 11x10b
Oscillator Type	External
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	44-LQFP
Supplier Device Package	-
Purchase URL	https://www.e-xfl.com/product-detail/stmicroelectronics/st72f361j9t3

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# **2 PIN DESCRIPTION**

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## Figure 2. LQFP 64-Pin Package Pinout



Address	Block	Register Label	Register Name	Reset Status	Remarks
000Fh	Port F	PFDR	Port F Data Register	00h <sup>1)</sup>	R/W <sup>2)</sup>
0010h		PFDDR	Port F Data Direction Register	00h	R/W <sup>2)</sup>
0011h		PFOR	Port F Option Register	00h	R/W <sup>2)</sup>
0012h to 0020h			Reserved Area (15 bytes)		
0021h	SPI	SPIDR	SPI Data I/O Register	xxh	R/W
0022h		SPICR	SPI Control Register	0xh	R/W
0023h		SPICSR	SPI Control/Status Register	00h	R/W
0024h	FLASH	FCSR	Flash Control/Status Register	00h	R/W
0025h 0026h 0027h 0028h 0029h 002Ah	ITC	ISPR0 ISPR1 ISPR2 ISPR3 EICR0 EICR1	Interrupt Software Priority Register 0 Interrupt Software Priority Register 1 Interrupt Software Priority Register 2 Interrupt Software Priority Register 3 External Interrupt Control Register 0 External Interrupt Control Register 1	FFh FFh FFh FFh 00h 00h	R/W R/W R/W R/W R/W
002Bh	AWU	AWUCSR	Auto Wake up f. Halt Control/Status Register	00h	R/W
002Ch		AWUPR	Auto Wake Up From Halt Prescaler	FFh	R/W
002Dh	CKCTRL	SICSR	System Integrity Control / Status Register	0xh	R/W
002Eh		MCCSR	Main Clock Control / Status Register	00h	R/W
002Fh	WWDG	WDGCR	Watchdog Control Register	7Fh	R/W
0030h		WDGWR	Watchdog Window Register	7Fh	R/W
0031h 0032h 0033h 0034h 0035h 0035h 0037h 0037h 0038h 0039h 003Ah	PWM ART	PWMDCR3 PWMDCR2 PWMDCR1 PWMDCR0 PWMCR ARTCSR ARTCAR ARTCAR ARTARR ARTICCSR ARTICCSR ARTICR1 ARTICR2	Pulse Width Modulator Duty Cycle Register 3 PWM Duty Cycle Register 2 PWM Duty Cycle Register 1 PWM Duty Cycle Register 0 PWM Control register Auto-Reload Timer Control/Status Register Auto-Reload Timer Counter Access Register Auto-Reload Timer Auto-Reload Register Auto-Reload Timer Auto-Reload Register ART Input Capture Control/Status Register ART Input Capture Register 1 ART Input Capture register 2	00h 00h 00h 00h 00h 00h 00h 00h 00h	R/W R/W R/W R/W R/W R/W R/W R/W Read Only Read Only
003Ch 003Dh 003Eh 003Fh 0040h 0041h 0042h 0043h 0044h	8-BIT TIMER	T8CR2 T8CR1 T8CSR T8IC1R T8OC1R T8CTR T8ACTR T8IC2R T8IC2R T8OC2R	Timer Control Register 2 Timer Control Register 1 Timer Control/Status Register Timer Input Capture 1 Register Timer Output Compare 1 Register Timer Counter Register Timer Alternate Counter Register Timer Input Capture 2 Register Timer Output Compare 2 Register	00h 00h xxh 00h FCh FCh xxh 00h	R/W R/W Read Only Read Only Read Only Read Only Read Only R/W
0045h	ADC	ADCCSR	Control/Status Register	00h	R/W
0046h		ADCDRH	Data High Register	00h	Read Only
0047h		ADCDRL	Data Low Register	00h	Read Only



Address	Block	Register Label	Register Name	Reset Status	Remarks
0048h 0049h 004Ah 004Bh 004Ch 004Dh 004Eh 004Fh	LINSCI1 (LIN Master/ Slave)	SCI1ISR SCI1DR SCI1BRR SCI1CR1 SCI1CR2 SCI1CR3 SCI1ERPR SCI1ETPR	SCI1 Status Register SCI1 Data Register SCI1 Baud Rate Register SCI1 Control Register 1 SCI1 Control Register 2 SCI1Control Register 3 SCI1 Extended Receive Prescaler Register SCI1 Extended Transmit Prescaler Register	C0h xxh 00h xxh 00h 00h 00h 00h	Read Only R/W R/W R/W R/W R/W R/W
0050h			Reserved Area (1 byte)		
0051h 0052h 0053h 0054h 0055h 0056h 0057h 0058h 0059h 005Ah 005Bh 005Ch 005Ch 005Fh	16-BIT TIMER	T16CR2 T16CR1 T16CSR T16IC1HR T16IC1LR T16OC1LR T16OC1LR T16CHR T16CLR T16ACHR T16ACHR T16ACLR T16IC2HR T16IC2LR T16OC2LR	Timer Control Register 2 Timer Control Register 1 Timer Control/Status Register Timer Input Capture 1 High Register Timer Input Capture 1 Low Register Timer Output Compare 1 High Register Timer Output Compare 1 Low Register Timer Counter High Register Timer Counter Low Register Timer Alternate Counter High Register Timer Alternate Counter Low Register Timer Input Capture 2 High Register Timer Input Capture 2 Low Register Timer Output Compare 2 High Register Timer Output Compare 2 Low Register	00h 00h xxh xxh 80h 00h FFh FCh FFh FCh xxh xxh 80h 00h	R/W R/W Read Only Read Only R/W Read Only Read Only
0060h 0061h 0062h 0063h 0064h 0065h 0066h 0067h	LINSCI2 (LIN Master)	SCI2SR SCI2DR SCI2BRR SCI2CR1 SCI2CR2 SCI2CR3 SCI2ERPR SCI2ETPR	SCI2 Status Register SCI2 Data Register SCI2 Baud Rate Register SCI2 Control Register 1 SCI2 Control Register 2 SCI2 Control Register 3 SCI2 Extended Receive Prescaler Register SCI2 Extended Transmit Prescaler Register	C0h xxh 00h xxh 00h 00h 00h 00h	Read Only R/W R/W R/W R/W R/W R/W R/W
to 007Fh	•		Reserved area (24 bytes)		

**Legend**: x = undefined, R/W = read/write

## Notes:

1. The contents of the I/O port DR registers are readable only in output configuration. In input configuration, the values of the I/O pins are returned instead of the DR register contents.

2. The bits associated with unavailable pins must always keep their reset value.

# 6 SUPPLY, RESET AND CLOCK MANAGEMENT

The device includes a range of utility features for securing the application in critical situations (for example, in case of a power brown-out), and reducing the number of external components. An overview is shown in Figure 11.

For more details, refer to dedicated parametric section.

## Main features

- Optional PLL for multiplying the frequency by 2
- Reset Sequence Manager (RSM)
- Multi-Oscillator Clock Management (MO)
  4 Crystal/Ceramic resonator oscillators
- System Integrity Management (SI)
  - Main supply Low voltage detection (LVD)
  - Auxiliary Voltage detector (AVD) with interrupt capability for monitoring the main supply

## 6.1 PHASE LOCKED LOOP

If the clock frequency input to the PLL is in the range 2 to 4 MHz, the PLL can be used to multiply the frequency by two to obtain an  $f_{OSC2}$  of 4 to 8 MHz. The PLL is enabled by option byte. If the PLL is disabled, then  $f_{OSC2} = f_{OSC}/2$ .

**Caution:** The PLL is not recommended for applications where timing accuracy is required. See "PLL Characteristics" on page 187.

## Figure 10. PLL Block Diagram



## Figure 11. Clock, Reset and Supply Block Diagram



## INTERRUPTS (Cont'd)

## **Servicing Pending Interrupts**

As several interrupts can be pending at the same time, the interrupt to be taken into account is determined by the following two-step process:

- the highest software priority interrupt is serviced,
- if several interrupts have the same software priority then the interrupt with the highest hardware priority is serviced first.

Figure 18 describes this decision process.





When an interrupt request is not serviced immediately, it is latched and then processed when its software priority combined with the hardware priority becomes the highest one.

**Note 1**: The hardware priority is exclusive while the software one is not. This allows the previous process to succeed with only one interrupt.

**Note 2**: RESET, TRAP and TLI can be considered as having the highest software priority in the decision process.

## **Different Interrupt Vector Sources**

Two interrupt source types are managed by the ST7 interrupt controller: the non-maskable type (RESET, TRAP) and the maskable type (external or from internal peripherals).

## Non-Maskable Sources

These sources are processed regardless of the state of the 11 and I0 bits of the CC register (see Figure 17). After stacking the PC, X, A and CC registers (except for RESET), the corresponding vector is loaded in the PC register and the I1 and I0 bits of the CC are set to disable interrupts (level 3). These sources allow the processor to exit HALT mode.

TRAP (Non Maskable Software Interrupt)

This software interrupt is serviced when the TRAP instruction is executed. It will be serviced according to the flowchart in Figure 17 as a TLI.

Caution: TRAP can be interrupted by a TLI.

## RESET

The RESET source has the highest priority in the ST7. This means that the first current routine has the highest software priority (level 3) and the highest hardware priority.

See the RESET chapter for more details.

## **Maskable Sources**

Maskable interrupt vector sources can be serviced if the corresponding interrupt is enabled and if its own interrupt software priority (in ISPRx registers) is higher than the one currently being serviced (I1 and I0 in CC register). If any of these two conditions is false, the interrupt is latched and thus remains pending.

TLI (Top Level Hardware Interrupt)

This hardware interrupt occurs when a specific edge is detected on the dedicated TLI pin.

**Caution**: A TRAP instruction must not be used in a TLI service routine.

External Interrupts

External interrupts allow the processor to exit from HALT low power mode.

External interrupt sensitivity is software selectable through the External Interrupt Control register (EICR).

External interrupt triggered on edge will be latched and the interrupt request automatically cleared upon entering the interrupt service routine.

If several input pins of a group connected to the same interrupt line are selected simultaneously, these will be logically ORed.

Peripheral Interrupts

Usually the peripheral interrupts cause the MCU to exit from HALT mode except those mentioned in the "Interrupt Mapping" table.

A peripheral interrupt occurs when a specific flag is set in the peripheral status registers and if the corresponding enable bit is set in the peripheral control register.

The general sequence for clearing an interrupt is based on an access to the status register followed by a read or write to an associated register.

**Note**: The clearing sequence resets the internal latch. A pending interrupt (that is, waiting for being serviced) will therefore be lost if the clear sequence is executed.

## **10.4 16-BIT TIMER**

## 10.4.1 Introduction

The timer consists of a 16-bit free-running counter driven by a programmable prescaler.

It may be used for a variety of purposes, including pulse length measurement of up to two input signals (*input capture*) or generation of up to two output waveforms (*output compare* and *PWM*).

Pulse lengths and waveform periods can be modulated from a few microseconds to several milliseconds using the timer prescaler and the CPU clock prescaler.

Some ST7 devices have two on-chip 16-bit timers. They are completely independent, and do not share any resources. They are synchronized after a MCU reset as long as the timer clock frequencies are not modified.

This description covers one or two 16-bit timers. In ST7 devices with two timers, register names are prefixed with TA (Timer A) or TB (Timer B).

## 10.4.2 Main Features

- Programmable prescaler: f<sub>CPU</sub> divided by 2, 4 or 8
- Overflow status flag and maskable interrupt
- External clock input (must be at least four times slower than the CPU clock speed) with the choice of active edge
- 1 or 2 Output Compare functions each with:
  - 2 dedicated 16-bit registers
  - 2 dedicated programmable signals
  - 2 dedicated status flags
  - 1 dedicated maskable interrupt
- 1 or 2 Input Capture functions each with:
  - 2 dedicated 16-bit registers
  - 2 dedicated active edge selection signals
  - 2 dedicated status flags
  - 1 dedicated maskable interrupt
- Pulse width modulation mode (PWM)
- One Pulse mode
- Reduced Power Mode
- 5 alternate functions on I/O ports (ICAP1, ICAP2, OCMP1, OCMP2, EXTCLK)\*

The Block Diagram is shown in Figure 48.

\*Note: Some timer pins may not be available (not bonded) in some ST7 devices. Refer to the device pin out description.

When reading an input signal on a non-bonded pin, the value will always be '1'.

## 10.4.3 Functional Description

## 10.4.3.1 Counter

The main block of the Programmable Timer is a 16-bit free running upcounter and its associated 16-bit registers. The 16-bit registers are made up of two 8-bit registers called high and low.

Counter Register (CR):

- Counter High Register (CHR) is the most significant byte (MS Byte).
- Counter Low Register (CLR) is the least significant byte (LS Byte).

Alternate Counter Register (ACR)

- Alternate Counter High Register (ACHR) is the most significant byte (MS Byte).
- Alternate Counter Low Register (ACLR) is the least significant byte (LS Byte).

These two read-only 16-bit registers contain the same value but with the difference that reading the ACLR register does not clear the TOF bit (Timer overflow flag), located in the Status register, (SR), (see note at the end of paragraph titled 16-bit read sequence).

Writing in the CLR register or ACLR register resets the free running counter to the FFFCh value. Both counters have a reset value of FFFCh (this is the only value which is reloaded in the 16-bit timer). The reset value of both counters is also FFFCh in One Pulse mode and PWM mode.

The timer clock depends on the clock control bits of the CR2 register, as illustrated in Table 17 Clock Control Bits. The value in the counter register repeats every 131072, 262144 or 524288 CPU clock cycles depending on the CC[1:0] bits.

The timer frequency can be  $f_{CPU}/2$ ,  $f_{CPU}/4$ ,  $f_{CPU}/8$  or an external frequency.



## 8-BIT TIMER (Cont'd)

## 10.5.3.5 Pulse Width Modulation Mode

Pulse Width Modulation (PWM) mode enables the generation of a signal with a frequency and pulse length determined by the value of the OC1R and OC2R registers.

Pulse Width Modulation mode uses the complete Output Compare 1 function plus the OC2R register, and so this functionality can not be used when PWM mode is activated.

In PWM mode, double buffering is implemented on the output compare registers. Any new values written in the OC1R and OC2R registers are taken into account only at the end of the PWM period (OC2) to avoid spikes on the PWM output pin (OCMP1).

## Procedure

To use pulse width modulation mode:

- 1. Load the OC2R register with the value corresponding to the period of the signal using the formula in the opposite column.
- 2. Load the OC1R register with the value corresponding to the period of the pulse if (OLVL1 = 0 and OLVL2 = 1) using the formula in the opposite column.
- Select the following in the CR1 register:
  - Using the OLVL1 bit, select the level to be applied to the OCMP1 pin after a successful comparison with the OC1R register.
  - Using the OLVL2 bit, select the level to be applied to the OCMP1 pin after a successful comparison with the OC2R register.
- 4. Select the following in the CR2 register:
  - Set OC1E bit: the OCMP1 pin is then dedicated to the output compare 1 function.
  - Set the PWM bit.
  - Select the timer clock (CC[1:0]) (see Table 19 Clock Control Bits).



If OLVL1 = 1 and OLVL2 = 0 the length of the positive pulse is the difference between the OC2R and OC1R registers.

If OLVL1 = OLVL2 a continuous signal will be seen on the OCMP1 pin.

The OC*i*R register value required for a specific timing application can be calculated using the following formula:

$$OC_{i}R Value = \frac{t \cdot f_{CPU}}{PBESC} - 5$$

Where:

t = Signal or pulse period (in seconds)

f<sub>CPU</sub> = PLL output x2 clock frequency in hertz (or f<sub>OSC</sub>/2 if PLL is not enabled)

PRESC = Timer prescaler factor (2, 4, 8 or 8000 depending on CC[1:0] bits, see Table 19 Clock Control Bits)

The Output Compare 2 event causes the counter to be initialized to FCh (See Figure 69)

## Notes:

- 1. The OCF1 and OCF2 bits cannot be set by hardware in PWM mode therefore the Output Compare interrupt is inhibited.
- 2. The ICF1 bit is set by hardware when the counter reaches the OC2R value and can produce a timer interrupt if the ICIE bit is set and the I bit is cleared.
- 3. In PWM mode the ICAP1 pin can not be used to perform input capture because it is disconnected to the timer. The ICAP2 pin can be used to perform input capture (ICF2 can be set and IC2R can be loaded) but the user must take care that the counter is reset each period and ICF1 can also generates interrupt if ICIE is set.
- 4. When the Pulse Width Modulation (PWM) and One Pulse Mode (OPM) bits are both set, the PWM mode is the only active one.

## SERIAL PERIPHERAL INTERFACE (cont'd)

## 10.6.3.3 Master Mode Operation

In master mode, the serial clock is output on the SCK pin. The clock frequency, polarity and phase are configured by software (refer to the description of the SPICSR register).

Note: The idle state of SCK must correspond to the polarity selected in the SPICSR register (by pulling up SCK if CPOL = 1 or pulling down SCK if CPOL = 0).

## How to operate the SPI in master mode

To operate the SPI in master mode, perform the following steps in order:

- Write to the SPICR register:
  - Select the clock frequency by configuring the SPR[2:0] bits.
  - Select the clock polarity and clock phase by configuring the CPOL and CPHA bits. Figure 74 shows the four possible configurations. **Note:** The slave must have the same CPOL and CPHA settings as the master.
- 2. Write to the SPICSR register:
  - Either set the SSM bit and set the SSI bit or clear the SSM bit and tie the SS pin high for the complete byte transmit sequence.
- 3. Write to the SPICR register:

  - Set the MSTR and SPE bits
    Note: MSTR and SPE bits remain set only if SS is high).

Important note: if the SPICSR register is not written first, the SPICR register setting (MSTR bit) may be not taken into account.

The transmit sequence begins when software writes a byte in the SPIDR register.

#### 10.6.3.4 Master Mode Transmit Sequence

When software writes to the SPIDR register, the data byte is loaded into the 8-bit shift register and then shifted out serially to the MOSI pin most significant bit first.

When data transfer is complete:

- The SPIF bit is set by hardware.
- An interrupt request is generated if the SPIE bit is set and the interrupt mask in the CCR register is cleared.

Clearing the SPIF bit is performed by the following software sequence:

- 1. An access to the SPICSR register while the SPIF bit is set
- 2. A read to the SPIDR register

Note: While the SPIF bit is set, all writes to the SPIDR register are inhibited until the SPICSR register is read.

## 10.6.3.5 Slave Mode Operation

In slave mode, the serial clock is received on the SCK pin from the master device.

To operate the SPI in slave mode:

- 1. Write to the SPICSR register to perform the following actions:
  - Select the clock polarity and clock phase by configuring the CPOL and CPHA bits (see Figure 74). Note: The slave must have the same CPOL and CPHA settings as the master.
  - Manage the  $\overline{SS}$  pin as described in Section 10.6.3.2 and Figure 72. If CPHA = 1 SS must be held low continuously. If CPHA = 0 SS must be held low during byte transmission and pulled up between each byte to let the slave write in the shift register.
- 2. Write to the SPICR register to clear the MSTR bit and set the SPE bit to enable the SPI I/O functions.

## 10.6.3.6 Slave Mode Transmit Sequence

When software writes to the SPIDR register, the data byte is loaded into the 8-bit shift register and then shifted out serially to the MISO pin most significant bit first.

The transmit sequence begins when the slave device receives the clock signal and the most significant bit of the data on its MOSI pin.

When data transfer is complete:

- The SPIF bit is set by hardware.
- An interrupt request is generated if SPIE bit is set and interrupt mask in the CCR register is cleared.

Clearing the SPIF bit is performed by the following software sequence:

- 1. An access to the SPICSR register while the SPIF bit is set
- A write or a read to the SPIDR register

Notes: While the SPIF bit is set, all writes to the SPIDR register are inhibited until the SPICSR register is read.

The SPIF bit can be cleared during a second transmission; however, it must be cleared before the second SPIF bit in order to prevent an Overrun condition (see Section 10.6.5.2).



# SERIAL PERIPHERAL INTERFACE (cont'd) SPI CONTROL/STATUS REGISTER (SPICSR)

Read/Write (some bits Read Only) Reset Value: 0000 0000 (00h)

7							0
SPIF	WCOL	OVR	MODF	-	SOD	SSM	SSI

Bit 7 = **SPIF** Serial Peripheral Data Transfer Flag (Read only)

This bit is set by hardware when a transfer has been completed. An interrupt is generated if SPIE = 1 in the SPICR register. It is cleared by a software sequence (an access to the SPICSR register followed by a write or a read to the SPIDR register).

- 0: Data transfer is in progress or the flag has been cleared.
- 1: Data transfer between the device and an external device has been completed.

**Note:** While the SPIF bit is set, all writes to the SPIDR register are inhibited until the SPICSR register is read.

Bit 6 = **WCOL** *Write Collision status (Read only)* This bit is set by hardware when a write to the SPIDR register is done during a transmit sequence. It is cleared by a software sequence (see Figure 75).

0: No write collision occurred

1: A write collision has been detected

Bit 5 = **OVR** SPI Overrun error (Read only) This bit is set by hardware when the byte currently being received in the shift register is ready to be transferred into the SPIDR register while SPIF = 1 (See Section 10.6.5.2). An interrupt is generated if SPIE = 1 in the SPICR register. The OVR bit is cleared by software reading the SPICSR register.

- 0: No overrun error
- 1: Overrun error detected

## Bit 4 = MODF Mode Fault flag (Read only)

This bit is set by hardware when the  $\overline{SS}$  pin is pulled low in master mode (see Section 10.6.5.1 "Master Mode Fault (MODF)"). An SPI interrupt can be generated if SPIE = 1 in the SPICR register. This bit is cleared by a software sequence (An access to the SPICSR register while MODF = 1 followed by a write to the SPICR register). 0: No master mode fault detected

1: A fault in master mode has been detected

Bit 3 = Reserved, must be kept cleared.

## Bit 2 = SOD SPI Output Disable

This bit is set and cleared by software. When set, it disables the alternate function of the SPI output (MOSI in master mode / MISO in slave mode) 0: SPI output enabled (if SPE = 1) 1: SPI output disabled

Bit 1 = **SSM** SS Management

This bit is set and cleared by software. When set, it disables the alternate function of the SPI SS pin and uses the SSI bit value instead. See Section 10.6.3.2 "Slave Select Management".

- 0: Hardware management (SS managed by external pin)
- 1: Software management (internal SS signal controlled by SSI bit. External SS pin free for general-purpose I/O)

## Bit 0 = SSI SS Internal Mode

This bit is set and cleared by software. It acts as a 'chip select' by controlling the level of the  $\overline{SS}$  slave select signal when the SSM bit is set.

0: Slave selected

1: Slave deselected

## SPI DATA I/O REGISTER (SPIDR)

Read/Write Reset Value: Undefined

7							0
D7	D6	D5	D4	D3	D2	D1	D0

The SPIDR register is used to transmit and receive data on the serial bus. In a master device, a write to this register will initiate transmission/reception of another byte.

**Notes:** During the last clock cycle the SPIF bit is set, a copy of the received data byte in the shift register is moved to a buffer. When the user reads the serial peripheral data I/O register, the buffer is actually being read.

While the SPIF bit is set, all writes to the SPIDR register are inhibited until the SPICSR register is read.

**Warning:** A write to the SPIDR register places data directly into the shift register for transmission.

A read to the SPIDR register returns the value located in the buffer and not the content of the shift register (see Figure 70).



## LINSCI™ SERIAL COMMUNICATION INTERFACE (SCI Mode) (cont'd)

## 10.7.5.2 Transmitter

The transmitter can send data words of either 8 or 9 bits depending on the M bit status. When the M bit is set, word length is 9 bits and the 9th bit (the MSB) has to be stored in the T8 bit in the SCICR1 register.

## Character Transmission

During an SCI transmission, data shifts out least significant bit first on the TDO pin. In this mode, the SCIDR register consists of a buffer (TDR) between the internal bus and the transmit shift register (see Figure 1).

## Procedure

- Select the M bit to define the word length.
- Select the desired baud rate using the SCIBRR and the SCIETPR registers.
- Set the TE bit to send a preamble of 10 (M = 0) or 11 (M = 1) consecutive ones (Idle Line) as first transmission.
- Access the SCISR register and write the data to send in the SCIDR register (this sequence clears the TDRE bit). Repeat this sequence for each data to be transmitted.

Clearing the TDRE bit is always performed by the following software sequence:

- 1. An access to the SCISR register
- 2. A write to the SCIDR register
- The TDRE bit is set by hardware and it indicates:
- The TDR register is empty.
- The data transfer is beginning.
- The next data can be written in the SCIDR register without overwriting the previous data.

This flag generates an interrupt if the TIE bit is set and the I[1:0] bits are cleared in the CCR register.

When a transmission is taking place, a write instruction to the SCIDR register stores the data in the TDR register and which is copied in the shift register at the end of the current transmission. When no transmission is taking place, a write instruction to the SCIDR register places the data directly in the shift register, the data transmission starts, and the TDRE bit is immediately set.

When a character transmission is complete (after the stop bit) the TC bit is set and an interrupt is generated if the TCIE is set and the I[1:0] bits are cleared in the CCR register.

Clearing the TC bit is performed by the following software sequence:

1. An access to the SCISR register

2. A write to the SCIDR register

**Note:** The TDRE and TC bits are cleared by the same software sequence.

## Break Characters

Setting the SBK bit loads the shift register with a break character. The break character length depends on the M bit (see Figure 2).

As long as the SBK bit is set, the SCI sends break characters to the TDO pin. After clearing this bit by software, the SCI inserts a logic 1 bit at the end of the last break character to guarantee the recognition of the start bit of the next character.

## Idle Line

Setting the TE bit drives the SCI to send a preamble of 10 (M = 0) or 11 (M = 1) consecutive '1's (idle line) before the first character.

In this case, clearing and then setting the TE bit during a transmission sends a preamble (idle line) after the current word. Note that the preamble duration (10 or 11 consecutive '1's depending on the M bit) does not take into account the stop bit of the previous character.

**Note:** Resetting and setting the TE bit causes the data in the TDR register to be lost. Therefore the best time to toggle the TE bit is when the TDRE bit is set, that is, before writing the next byte in the SCIDR.

## LINSCI™ SERIAL COMMUNICATION INTERFACE (SCI Mode) (cont'd)

1<sub>CPU</sub>

(16\*PR)\*RR

#### 10.7.5.4 Conventional Baud Rate Generation

The baud rates for the receiver and transmitter (Rx and Tx) are set independently and calculated as follows:

 $Tx = \frac{f_{CPU}}{(16 \cdot PR) \cdot TR} \qquad Rx = \frac{1}{(16 \cdot PR) \cdot TR}$ 

with:

PR = 1, 3, 4 or 13 (see SCP[1:0] bits)

TR = 1, 2, 4, 8, 16, 32, 64,128

(see SCT[2:0] bits)

RR = 1, 2, 4, 8, 16, 32, 64,128

(see SCR[2:0] bits)

All these bits are in the SCIBRR register.

**Example:** If  $f_{CPU}$  is 8 MHz (normal mode) and if PR = 13 and TR = RR = 1, the transmit and receive baud rates are 38400 baud.

Note: The baud rate registers MUST NOT be changed while the transmitter or the receiver is enabled.

#### 10.7.5.5 Extended Baud Rate Generation

The extended prescaler option gives a very fine tuning on the baud rate, using a 255 value prescaler, whereas the conventional Baud Rate Generator retains industry standard software compatibility.

The extended baud rate generator block diagram is described in Figure 3.

The output clock rate sent to the transmitter or to the receiver will be the output from the 16 divider divided by a factor ranging from 1 to 255 set in the SCIERPR or the SCIETPR register.

**Note:** The extended prescaler is activated by setting the SCIETPR or SCIERPR register to a value other than zero. The baud rates are calculated as follows:

$$Tx = \frac{f_{CPU}}{16 \cdot ETPR^{*}(PR^{*}TR)} Rx = \frac{f_{CPU}}{16 \cdot ERPR^{*}(PR^{*}RR)}$$

with:



# LINSCI<sup>TM</sup> SERIAL COMMUNICATION INTERFACE (SCI Mode) (cont'd)

## 10.7.8 SCI Mode Register Description

## **STATUS REGISTER (SCISR)**

Read Only

Reset Value: 1100 0000 (C0h)

7							0	
TDRE	тс	RDRF	IDLE	OR <sup>1)</sup>	NF <sup>1)</sup>	FE <sup>1)</sup>	PE <sup>1)</sup>	

Bit 7 = **TDRE** *Transmit data register empty* 

This bit is set by hardware when the content of the TDR register has been transferred into the shift register. An interrupt is generated if the TIE = 1 in the SCICR2 register. It is cleared by a software sequence (an access to the SCISR register followed by a write to the SCIDR register).

0: Data is not transferred to the shift register

1: Data is transferred to the shift register

## Bit 6 = **TC** Transmission complete

This bit is set by hardware when transmission of a character containing Data is complete. An interrupt is generated if TCIE = 1 in the SCICR2 register. It is cleared by a software sequence (an access to the SCISR register followed by a write to the SCIDR register).

0: Transmission is not complete

1: Transmission is complete

**Note:** TC is not set after the transmission of a Preamble or a Break.

#### Bit 5 = **RDRF** *Received data ready flag*

This bit is set by hardware when the content of the RDR register has been transferred to the SCIDR register. An interrupt is generated if RIE = 1 in the SCICR2 register. It is cleared by a software sequence (an access to the SCISR register followed by a read to the SCIDR register).

0: Data is not received

1: Received data is ready to be read

### Bit 4 = **IDLE** *Idle line detected*

This bit is set by hardware when an Idle Line is detected. An interrupt is generated if the ILIE = 1 in the SCICR2 register. It is cleared by a software sequence (an access to the SCISR register followed by a read to the SCIDR register).

0: No Idle Line is detected

1: Idle Line is detected

**Note:** The IDLE bit will not be set again until the RDRF bit has been set itself (that is, a new idle line occurs).

## Bit 3 = **OR** Overrun error

The OR bit is set by hardware when the word currently being received in the shift register is ready to be transferred into the RDR register whereas RDRF is still set. An interrupt is generated if RIE = 1 in the SCICR2 register. It is cleared by a software sequence (an access to the SCISR register followed by a read to the SCIDR register).

0: No Overrun error

1: Overrun error detected

**Note:** When this bit is set, RDR register contents will not be lost but the shift register will be overwritten.

## Bit 2 = NF Character Noise flag

This bit is set by hardware when noise is detected on a received character. It is cleared by a software sequence (an access to the SCISR register followed by a read to the SCIDR register).

0: No noise 1: Noise is detected

**Note:** This bit does not generate interrupt as it appears at the same time as the RDRF bit which itself generates an interrupt.

## Bit 1 = FE Framing error

This bit is set by hardware when a desynchronization, excessive noise or a break character is detected. It is cleared by a software sequence (an access to the SCISR register followed by a read to the SCIDR register).

0: No Framing error

1: Framing error or break character detected

**Note:** This bit does not generate an interrupt as it appears at the same time as the RDRF bit which itself generates an interrupt. If the word currently being transferred causes both a frame error and an overrun error, it will be transferred and only the OR bit will be set.

## Bit 0 = **PE** Parity error

This bit is set by hardware when a byte parity error occurs (if the PCE bit is set) in receiver mode. It is cleared by a software sequence (a read to the status register followed by an access to the SCIDR data register). An interrupt is generated if PIE = 1 in the SCICR1 register.

0: No parity error

1: Parity error detected



#### LINSCI™ SERIAL COMMUNICATION INTERFACE (SCI Mode) (cont'd) EXTENDED RECEIVE PRESCALER DIVISION REGISTER (SCIERPR) EXTENDED TRAN REGISTER (SCIERPR)

Read/Write

Reset Value: 0000 0000 (00h)

7							0
ERPR							
7	6	5	4	3	2	1	0

# Bits 7:0 = **ERPR[7:0]** 8-bit Extended Receive Prescaler Register

The extended Baud Rate Generator is activated when a value other than 00h is stored in this register. The clock frequency from the 16 divider (see Figure 3) is divided by the binary factor set in the SCIERPR register (in the range 1 to 255).

The extended baud rate generator is not active after a reset.

# EXTENDED TRANSMIT PRESCALER DIVISION REGISTER (SCIETPR)

Read/Write

Reset Value:0000 0000 (00h)

7							0
ETPR							
7	6	5	4	3	2	1	0

## Bits 7:0 = **ETPR[7:0]** 8-bit Extended Transmit Prescaler Register

The extended Baud Rate Generator is activated when a value other than 00h is stored in this register. The clock frequency from the 16 divider (see Figure 3) is divided by the binary factor set in the SCIETPR register (in the range 1 to 255).

The extended baud rate generator is not active after a reset.

Note: In LIN slave mode, the Conventional and Extended Baud Rate Generators are disabled.



## LINSCI™ SERIAL COMMUNICATION INTERFACE (LIN Master Only) (Cont'd)

## 10.8.4.3 Receiver

The SCI can receive data words of either 8 or 9 bits. When the M bit is set, word length is 9 bits and the MSB is stored in the R8 bit in the SCICR1 register.

## Character reception

During a SCI reception, data shifts in least significant bit first through the RDI pin. In this mode, the SCIDR register consists or a buffer (RDR) between the internal bus and the received shift register (see Figure 88 on page 153).

#### Procedure

- Select the M bit to define the word length.
- Select the desired baud rate using the SCIBRR and the SCIERPR registers.
- Set the RE bit, this enables the receiver which begins searching for a start bit.

When a character is received:

- The RDRF bit is set. It indicates that the content of the shift register is transferred to the RDR.
- An interrupt is generated if the RIE bit is set and the I bit is cleared in the CCR register.
- The error flags can be set if a frame error, noise or an overrun error has been detected during reception.

Clearing the RDRF bit is performed by the following software sequence done by:

- 1. An access to the SCISR register
- 2. A read to the SCIDR register.

The RDRF bit must be cleared before the end of the reception of the next character to avoid an overrun error.

## **Break Character**

When a break character is received, the SCI handles it as a framing error.

## Idle Character

When an idle frame is detected, there is the same procedure as a data received character plus an interrupt if the ILIE bit is set and the I bit is cleared in the CCR register.

### **Overrun Error**

An overrun error occurs when a character is received when RDRF has not been reset. Data cannot be transferred from the shift register to the RDR register until the RDRF bit is cleared.

When a overrun error occurs:

- The OR bit is set.
- The RDR content is not lost.
- The shift register is overwritten.
- An interrupt is generated if the RIE bit is set and the I bit is cleared in the CCR register.

The OR bit is reset by an access to the SCISR register followed by a SCIDR register read operation.

#### Noise Error

Oversampling techniques are used for data recovery by discriminating between valid incoming data and noise.

When noise is detected in a frame:

- The NF is set at the rising edge of the RDRF bit.
- Data is transferred from the Shift register to the SCIDR register.
- No interrupt is generated. However this bit rises at the same time as the RDRF bit which itself generates an interrupt.

The NF bit is reset by a SCISR register read operation followed by a SCIDR register read operation.

## Framing Error

A framing error is detected when:

- The stop bit is not recognized on reception at the expected time, following either a de-synchronization or excessive noise.
- A break is received.
- When the framing error is detected:
- the FE bit is set by hardware
- Data is transferred from the Shift register to the SCIDR register.
- No interrupt is generated. However this bit rises at the same time as the RDRF bit which itself generates an interrupt.

The FE bit is reset by a SCISR register read operation followed by a SCIDR register read operation.

# **11 INSTRUCTION SET**

## **11.1 CPU ADDRESSING MODES**

The CPU features 17 different addressing modes which can be classified in seven main groups:

Addressing Mode	Example
Inherent	nop
Immediate	ld A,#\$55
Direct	ld A,\$55
Indexed	ld A,(\$55,X)
Indirect	ld A,([\$55],X)
Relative	jrne loop
Bit operation	bset byte,#5

The CPU Instruction set is designed to minimize the number of bytes required per instruction: To do so, most of the addressing modes may be subdivided in two submodes called long and short:

- Long addressing mode is more powerful because it can use the full 64 Kbyte address space, however it uses more bytes and more CPU cycles.
- Short addressing mode is less powerful because it can generally only access page zero (0000h -00FFh range), but the instruction size is more compact, and faster. All memory to memory instructions use short addressing modes only (CLR, CPL, NEG, BSET, BRES, BTJT, BTJF INC, DEC, RLC, RRC, SLL, SRL, SRA, SWAP)

The ST7 Assembler optimizes the use of long and short addressing modes.

Table 31. CP	U Addres	ssing Mod	le Overview		te '		
	Mode		Syntax	Destination	Pointer Address (Hex.)	Pointer Size (Hex.)	Length (Bytes)
Inherent			nop	$\mathcal{O}^{P}$			+ 0
Immediate			ld A,#\$55				+ 1
Short	Direct		ld A,\$10	00FF			+ 1
Long	Direct		ld A,\$1000	0000FFFF			+ 2
No Offset	Direct	Indexed	ld A,(X)	00FF			+ 0
Short	Direct	Indexed	ld A,(\$10,X)	001FE			+ 1
Long	Direct	Indexed	ld A,(\$1000,X)	0000FFFF			+ 2
Short	Indirect		ld A,[\$10]	00FF	00FF	byte	+ 2
Long	Indirect		ld A,[\$10.w]	0000FFFF	00FF	word	+ 2
Short	Indirect	Indexed	ld A,([\$10],X)	001FE	00FF	byte	+ 2
Long	Indirect	Indexed	ld A,([\$10.w],X)	0000FFFF	00FF	word	+ 2
Relative	Direct		jrne loop	PC+/-127			+ 1
Relative	Indirect		jrne [\$10]	PC+/-127	00FF	byte	+ 2
Bit	Direct		bset \$10,#7	00FF			+ 1
Bit	Indirect		bset [\$10],#7	00FF	00FF	byte	+ 2
Bit	Direct	Relative	btjt \$10,#7,skip	00FF			+ 2
Bit	Indirect	Relative	btjt [\$10],#7,skip	00FF	00FF	byte	+ 3

## CLOCK AND TIMING CHARACTERISTICS (Cont'd)

## 12.5.3 Crystal and Ceramic Resonator Oscillators

The ST7 internal clock can be supplied with four different Crystal/Ceramic resonator oscillators. All the information given in this paragraph is based on characterization results with specified typical external components. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and start-up stabilization time. Refer to the crystal/ceramic resonator manufacturer for more details (frequency, package, accuracy...).  $^{1/2}$ 

Symbol	Parameter	Cond	itions	Min	Max	Unit
fosc		LP: Low power of	oscillator	1	2	SMI I−
	O coillator Fraguenov <sup>3)</sup>	MP: Medium pov	wer oscillator	>2	4	
	Oscillator Frequency	MS: Medium spe	eed oscillator	>4	8	
		HS: High speed	>8	16	51	
R <sub>F</sub>	Feedback resistor			20	40	kΩ
	Recommended load consoitance ver	R <sub>S</sub> = 200Ω	LP oscillator	22	56	
C <sub>L1</sub> R	Recommended load capacitance ver-	$R_{S} = 200\Omega$	MP oscillator	22	46	nΕ
C <sub>12</sub>	sus equivalent serial resistance of the	$R_{S} = 200\Omega$	MS oscillator	18	33	рг
LZ	crystal of ceramic resonator (n <sub>S</sub> )	R <sub>S</sub> = 100Ω	HS oscillator	15	33	

Symbol	Parameter	Conditions	Тур	Max	Unit
		V <sub>DD</sub> = 5V LP oscillator	80	150	
	OSC2 driving ourrent	$V_{IN} = V_{SS}$ MP oscillator	160	250	^
<sup>1</sup> 2	USC2 driving current	MS oscillator	310	460	μΑ
		HS oscillator	610	910	

## Figure 100. Typical Application with a Crystal or Ceramic Resonator



#### Notes:

1. Resonator characteristics given by the crystal/ceramic resonator manufacturer.

2.  $t_{SU(OSC)}$  is the typical oscillator start-up time measured between  $V_{DD}$  = 2.8V and the fetch of the first instruction (with a quick  $V_{DD}$  ramp-up from 0 to 5V (< 50µs).

3. The oscillator selection can be optimized in terms of supply current using an high quality resonator with small R<sub>S</sub> value. Refer to crystal/ceramic resonator manufacturer for more details.



## **12.9 I/O PORT PIN CHARACTERISTICS**

## 12.9.1 General Characteristics

Subject to general operating conditions for  $V_{DD}$ ,  $f_{OSC}$ , and  $T_A$  unless otherwise specified.

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
V <sub>IL</sub>	Input low level voltage <sup>1)</sup>	CMOS ports				$0.3 \times V_{DD}$	
V <sub>IH</sub>	Input high level voltage <sup>1)</sup>			$0.7  ext{ x V}_{\text{DD}}$			
V <sub>hys</sub>	Schmitt trigger voltage hysteresis <sup>2)</sup>				1		V
V <sub>IL</sub>	Input low level voltage <sup>1)</sup>	TTL ports				0.8	
V <sub>IH</sub>	Input high level voltage <sup>1)</sup>			2			
V <sub>hys</sub>	Schmitt trigger voltage hysteresis <sup>2)</sup>				400		mV
I <sub>INJ(PIN)</sub>	Injected Current on PB3	V <sub>DD</sub> = 5V	Flash devices	0		+4	5
			ROM devices			±4	
	Injected Current on any other I/O pin				2	±4	mA
$\Sigma I_{\rm INJ(PIN)}^{3)}$	Total injected current (sum of all I/O and control pins) <sup>7)</sup>				,00	±25	
1	Input leakage current on robust pins	See "10-BIT ADC CHARACTERISTICS" on page 204					
Ikg	Input leakage current <sup>4)</sup>	$V_{SS} \le V_{IN} \le V_{DD}$		XC		±1	
۱ <sub>S</sub>	Static current consumption <sup>5)</sup>	Floating input mode		5	200		μΑ
R <sub>PU</sub>	Weak pull-up equivalent resistor <sup>6)</sup>	$V_{IN} = V_{SS}$	$V_{DD} = 5V$	50	90	250	kΩ
C <sub>IO</sub>	I/O pin capacitance	202			5		pF
t <sub>f(IO)out</sub>	Output high to low level fall time	C <sub>L</sub> = 50pF Between 10% and 90%			25		ne
t <sub>r(IO)out</sub>	Output low to high level rise time				25		113
t <sub>w(IT)in</sub>	External interrupt pulse time <sup>7)</sup>			1			t <sub>CPU</sub>

#### Notes:

1. Data based on characterization results, not tested in production.

2. Hysteresis voltage between Schmitt trigger switching levels. Based on characterization results, not tested.

3. When the current limitation is not possible, the V<sub>IN</sub> absolute maximum rating must be respected, otherwise refer to  $I_{INJ(PIN)}$  specification. A positive injection is induced by  $V_{IN} > V_{DD}$  while a negative injection is induced by  $V_{IN} < V_{SS}$ . Refer to Section 12.2 on page 179 for more details. 4. Leakage could be higher than max. if negative current is injected on adjacent pins.

5. Configuration not recommended, all unused pins must be kept at a fixed voltage: Using the output mode of the I/O, for example, or an external pull-up or pull-down resistor (see Figure 103). Data based on design simulation and/or technology characteristics, not tested in production.

6. The R<sub>PU</sub> pull-up equivalent resistor is based on a resistive transistor (corresponding I<sub>PU</sub> current characteristics described in Figure 104).

7. To generate an external interrupt, a minimum pulse width must be applied on an I/O port pin configured as an external interrupt source.

# I/O PORT PIN CHARACTERISTICS (Cont'd)

# Figure 111. Typical V<sub>OH</sub> vs V<sub>DD</sub>

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duration is always correct assuming the application is not doing anything between the idle and the break. This can be ensured by temporarily disabling interrupts.

The exact sequence is:

- Disable interrupts
- Reset and Set TE (IDLE request)
- Set and Reset SBK (Break Request)

- Re-enable interrupts

## LIN mode

If the LINE bit in the SCICR3 is set and the M bit in the SCICR1 register is reset, the LINSCI is in LIN master mode. A single break character is sent by setting and resetting the SBK bit in the SCICR2 register. In some cases, the break character may have a longer duration than expected:

- 24 bits instead of 13 bits



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