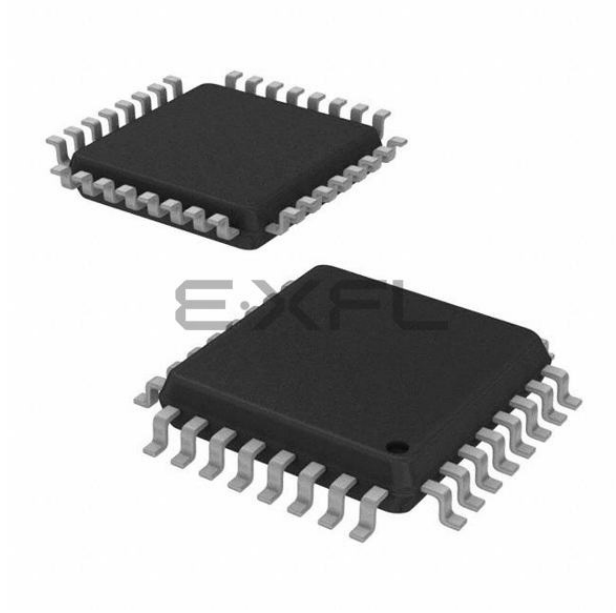

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What is "Embedded - Microcontrollers"?

"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "Embedded - Microcontrollers"

Details	
Product Status	Obsolete
Core Processor	ST7
Core Size	8-Bit
Speed	8MHz
Connectivity	LINbusSCI, SPI
Peripherals	LVD, POR, PWM, WDT
Number of I/O	24
Program Memory Size	32KB (32K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	1.5K x 8
Voltage - Supply (Vcc/Vdd)	3.8V ~ 5.5V
Data Converters	A/D 6x10b
Oscillator Type	External
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	32-LQFP
Supplier Device Package	-
Purchase URL	https://www.e-xfl.com/product-detail/stmicroelectronics/st72f361k6t6

FLASH PROGRAM MEMORY (Cont'd)

4.5 ICP (IN-CIRCUIT PROGRAMMING)

To perform ICP the microcontroller must be switched to ICC (In-Circuit Communication) mode by an external controller or programming tool.

Depending on the ICP code downloaded in RAM, Flash memory programming can be fully customized (number of bytes to program, program locations, or selection serial communication interface for downloading).

When using an STMicroelectronics or third-party programming tool that supports ICP and the specific microcontroller device, the user needs only to implement the ICP hardware interface on the application board (see Figure 7). For more details on the pin locations, refer to the device pinout description.

4.6 IAP (IN-APPLICATION PROGRAMMING)

This mode uses a BootLoader program previously stored in Sector 0 by the user (in ICP mode or by plugging the device in a programming tool).

This mode is fully controlled by user software. This allows it to be adapted to the user application, (user-defined strategy for entering programming mode, choice of communications protocol used to fetch the data to be stored, etc.). For example, it is possible to download code from the SPI, SCI or other type of serial interface and program it in the Flash. IAP mode can be used to program any of the Flash sectors except Sector 0, which is write/erase protected to allow recovery in case errors occur during the programming operation.

4.7 RELATED DOCUMENTATION

For details on Flash programming and ICC protocol, refer to the *ST7 Flash Programming Reference Manual* and to the *ST7 ICC Protocol Reference Manual*.

4.8 REGISTER DESCRIPTION

FLASH CONTROL/STATUS REGISTER (FCSR)

Read/Write

Reset Value: 0000 0000 (00h)

7							0
0	0	0	0	0	0	0	0

This register is reserved for use by Programming Tool software. It controls the Flash programming and erasing operations.

Table 5. Flash Control/Status Register Address and Reset Value

Address (Hex.)	Register Label	7	6	5	4	3	2	1	0
0024h	**FCSR** Reset Value	0	0	0	0	0	0	0	0

5 CENTRAL PROCESSING UNIT

5.1 INTRODUCTION

This CPU has a full 8-bit architecture and contains six internal registers allowing efficient 8-bit data manipulation.

5.2 MAIN FEATURES

- Enable executing 63 basic instructions
- Fast 8-bit by 8-bit multiply
- 17 main addressing modes (with indirect addressing mode)
- Two 8-bit index registers
- 16-bit stack pointer
- Low power HALT and WAIT modes
- Priority maskable hardware interrupts
- Non-maskable software/hardware interrupts

5.3 CPU REGISTERS

The six CPU registers shown in Figure 8 are not present in the memory mapping and are accessed by specific instructions.

Accumulator (A)

The Accumulator is an 8-bit general purpose register used to hold operands and the results of the arithmetic and logic calculations and to manipulate data.

Index Registers (X and Y)

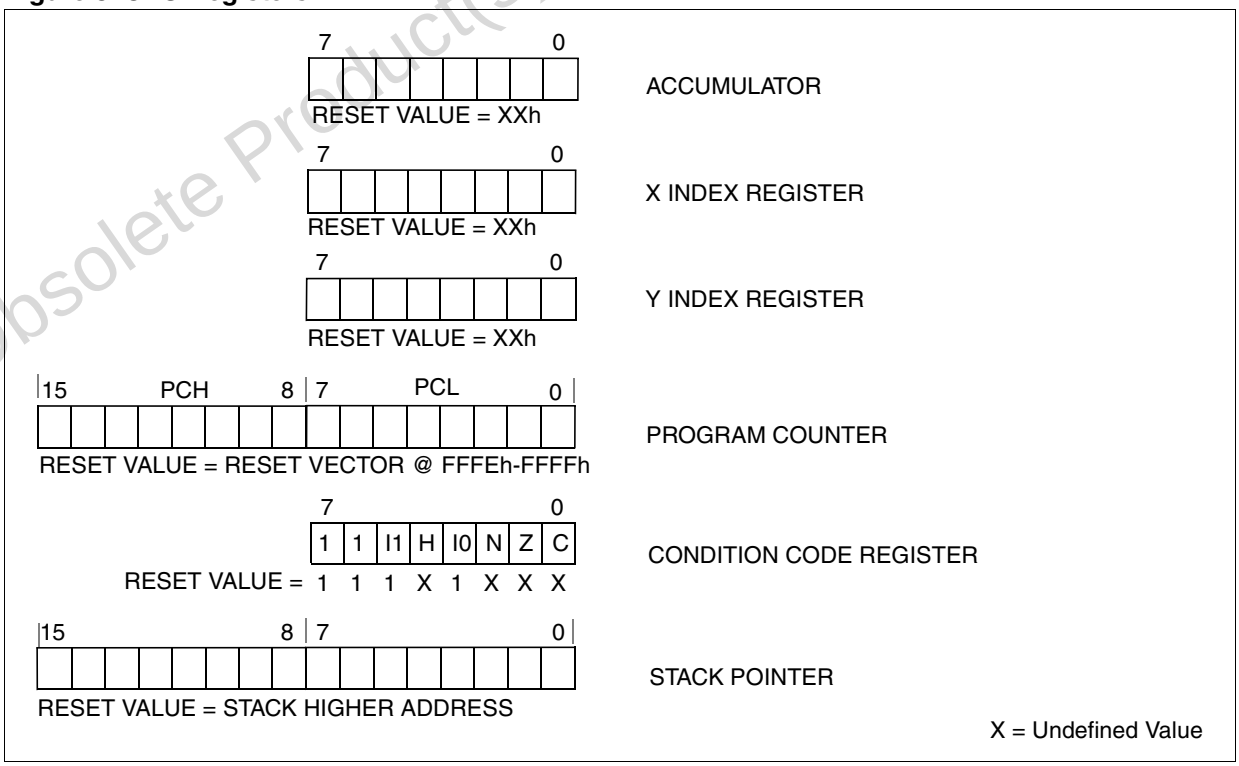
These 8-bit registers are used to create effective addresses or as temporary storage areas for data manipulation. (The Cross-Assembler generates a precede instruction (PRE) to indicate that the following instruction refers to the Y register.)

The Y register is not affected by the interrupt automatic procedures.

Program Counter (PC)

The program counter is a 16-bit register containing the address of the next instruction to be executed by the CPU. It is made of two 8-bit registers PCL (Program Counter Low which is the LSB) and PCH (Program Counter High which is the MSB).

Figure 8. CPU Registers



INTERRUPTS (Cont'd)

7.6 EXTERNAL INTERRUPTS

7.6.1 I/O Port Interrupt Sensitivity

The external interrupt sensitivity is controlled by the ISxx bits in the EICR register (Figure 21). This control allows up to four fully independent external interrupt source sensitivities.

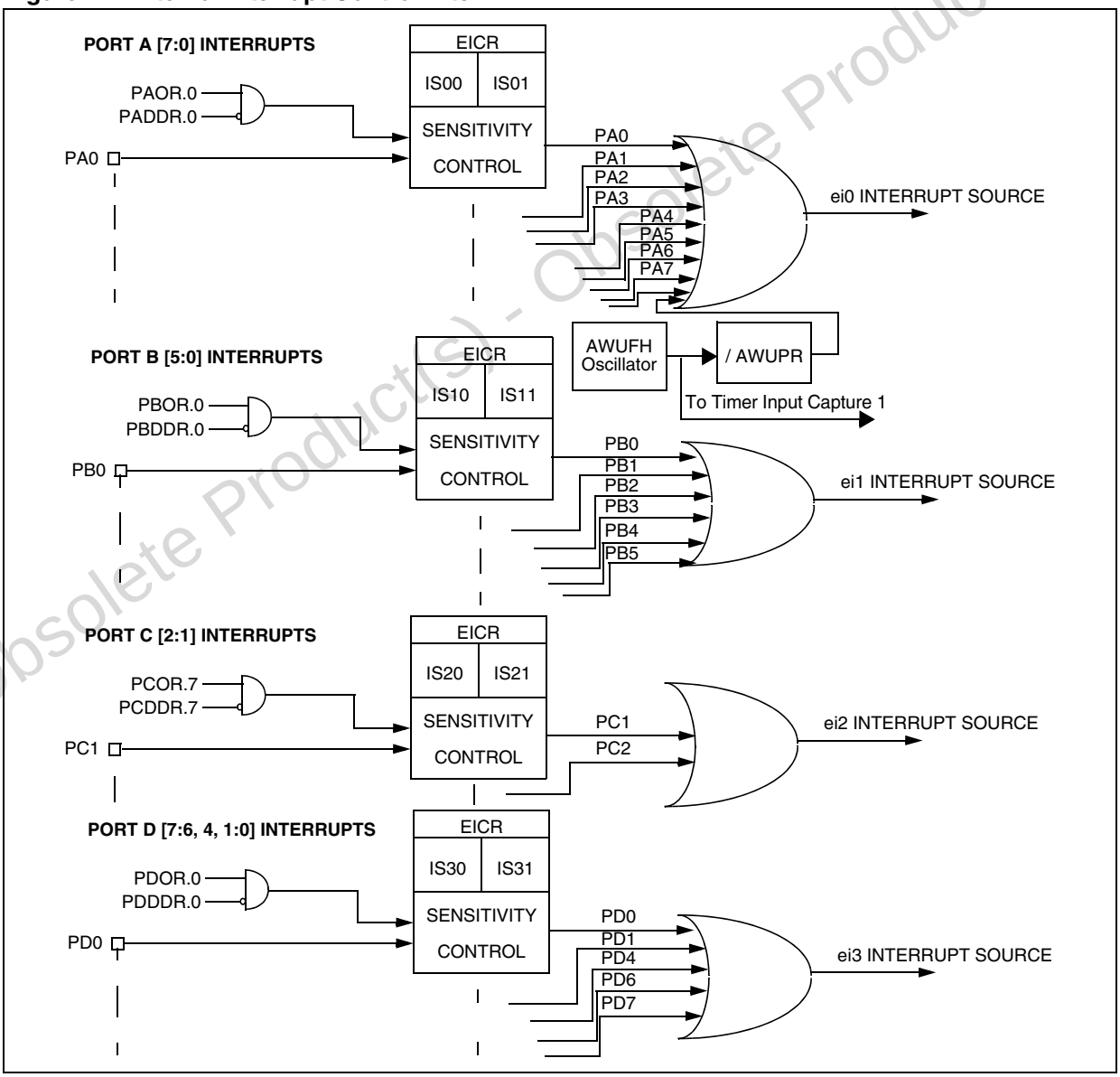
Each external interrupt source can be generated on four (or five) different events on the pin:

- Falling edge
- Rising edge
- Falling and rising edge
- Falling edge and low level

To guarantee correct functionality, the sensitivity bits in the EICR register can be modified only when the I1 and I0 bits of the CC register are both set to 1 (level 3). This means that interrupts must be disabled before changing sensitivity.

The pending interrupts are cleared by writing a different value in the ISx[1:0] of the EICR.

Figure 21. External Interrupt Control Bits



POWER SAVING MODES (Cont'd)

8.4 HALT MODE

The HALT mode is the lowest power consumption mode of the MCU. It is entered by executing the 'HALT' instruction when the OIE bit of the Main Clock Controller Status register (MCCSR) is cleared (see Section 10.2 on page 58 for more details on the MCCSR register) and when the AWUEN bit in the AWUCSR register is cleared.

The MCU can exit HALT mode on reception of either a specific interrupt (see Table 9, "Interrupt Mapping," on page 33) or a RESET. When exiting HALT mode by means of a RESET or an interrupt, the oscillator is immediately turned on and the 256 or 4096 CPU cycle delay is used to stabilize the oscillator. After the start up delay, the CPU resumes operation by servicing the interrupt or by fetching the reset vector which woke it up (see Figure 26).

When entering HALT mode, the I[1:0] bits in the CC register are forced to '10b' to enable interrupts. Therefore, if an interrupt is pending, the MCU wakes up immediately.

In HALT mode, the main oscillator is turned off causing all internal processing to be stopped, including the operation of the on-chip peripherals. All peripherals are not clocked except the ones which get their clock supply from another clock generator (such as an external or auxiliary oscillator).

The compatibility of Watchdog operation with HALT mode is configured by the "WDGHALT" option bit of the option byte. The HALT instruction when executed while the Watchdog system is enabled, can generate a Watchdog RESET (see Section 10.1 on page 52 for more details).

Figure 25. HALT Timing Overview

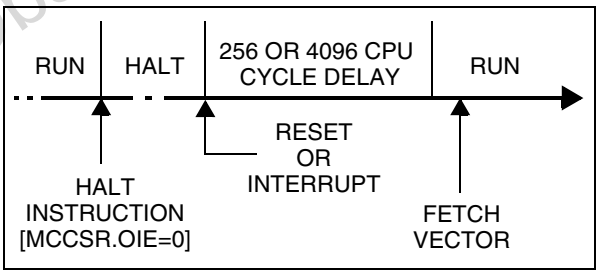
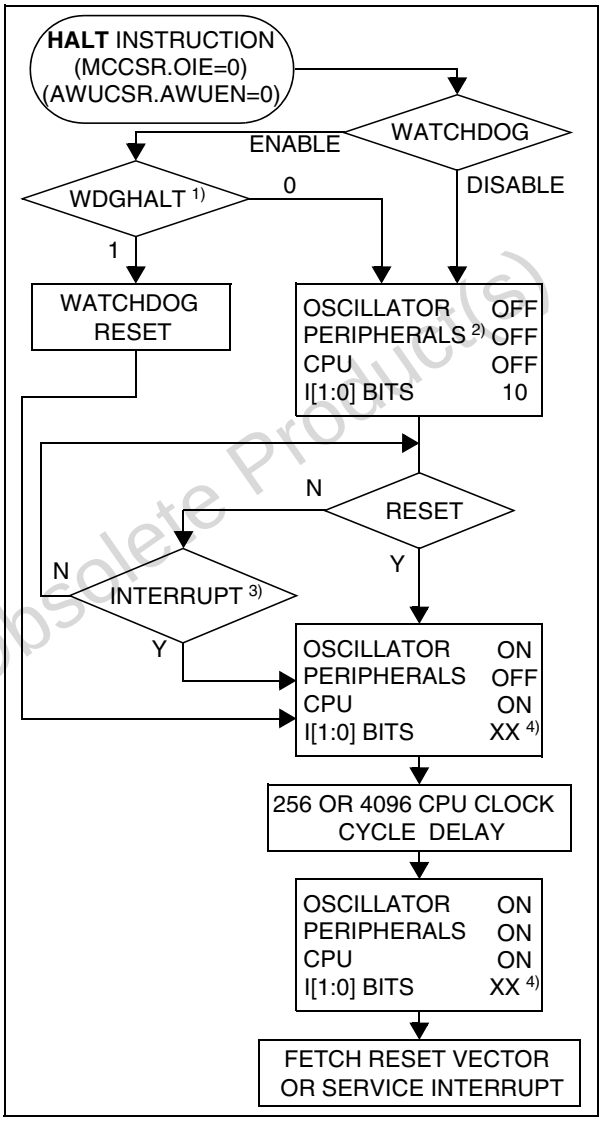


Figure 26. HALT Mode Flow-chart



Notes:

1. WDGHALT is an option bit. See option byte section for more details.

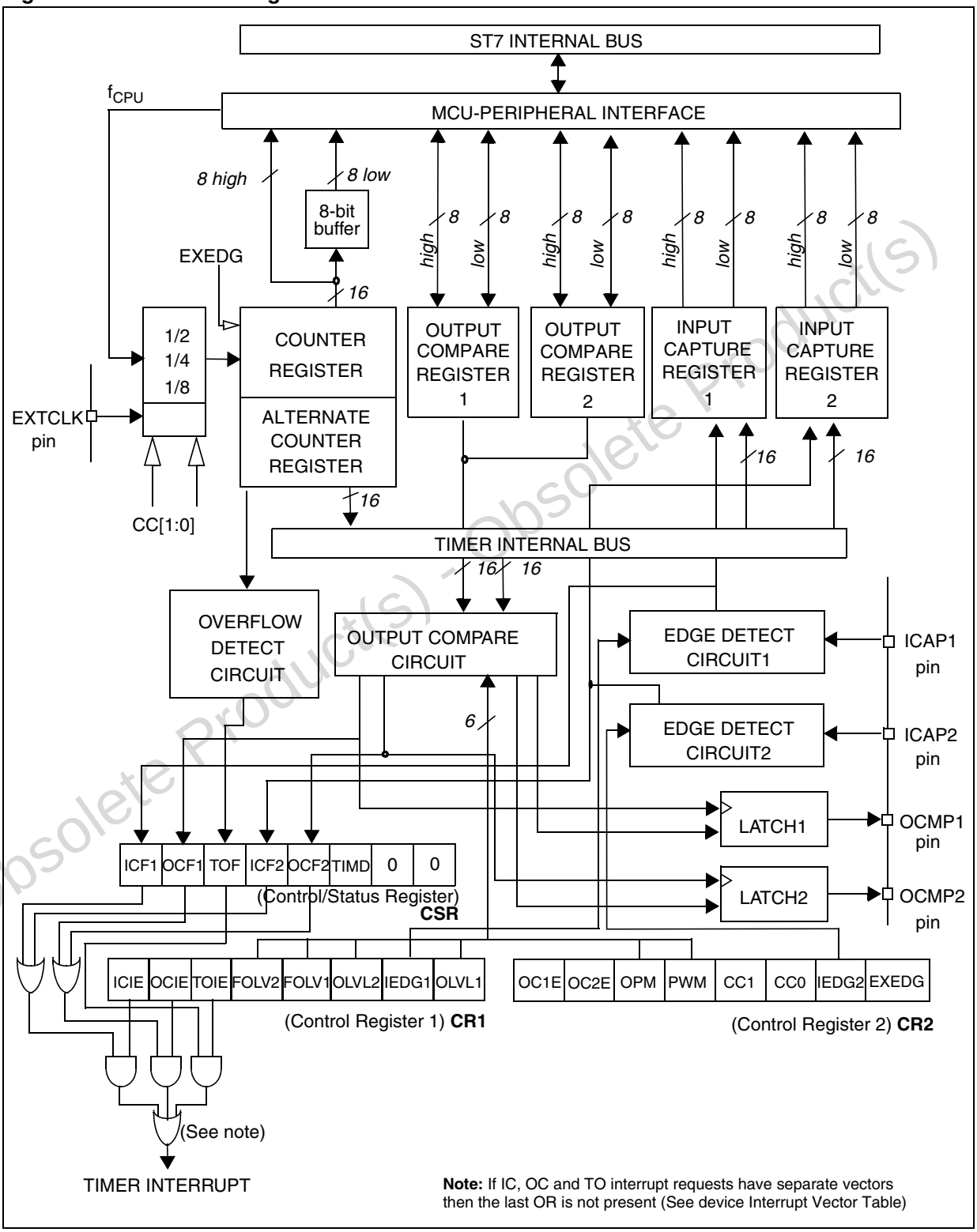
2. Peripheral clocked with an external clock source can still be active.

3. Only some specific interrupts can exit the MCU from HALT mode (such as external interrupt). Refer to Table 9, "Interrupt Mapping," on page 33 for more details.

4. Before servicing an interrupt, the CC register is pushed on the stack. The I[1:0] bits of the CC register are set to the current software priority level of the interrupt routine and recovered when the CC register is popped.

16-BIT TIMER (Cont'd)

Figure 48. Timer Block Diagram



16-BIT TIMER (Cont'd)

Table 19. 16-Bit Timer Register Map

Address (Hex.)	Register Name	7	6	5	4	3	2	1	0
51	**CR2**	OC1E	OC2E	OPM	PWM	CC1	CC0	IEDG2	EXEDG
52	**CR1**	ICIE	OCIE	TOIE	FOLV2	FOLV1	OLVL2	IEDG1	OLVL1
53	**CSR**	ICF1	OCF1	TOF	ICF2	OCF2	TIMD		
54	**IC1HR**	MSB							LSB
55	**IC1LR**	MSB							LSB
56	**OC1HR**	MSB							LSB
57	**OC1LR**	MSB							LSB
58	**CHR**	MSB							LSB
59	**CLR**	MSB							LSB
5A	**ACHR**	MSB							LSB
5B	**ACLR**	MSB							LSB
5C	**IC2HR**	MSB							LSB
5D	**IC2LR**	MSB							LSB
5E	**OC2HR**	MSB							LSB
5F	**OC2LR**	MSB							LSB

8-BIT TIMER (Cont'd)

10.5.3.4 One Pulse Mode

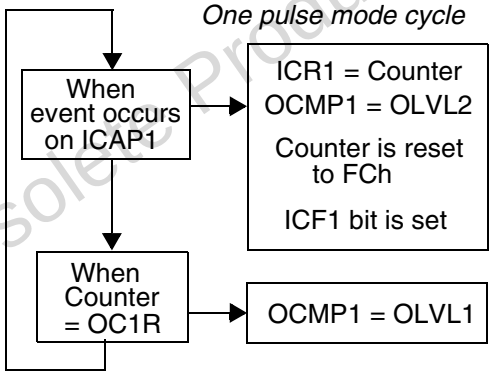
One Pulse mode enables the generation of a pulse when an external event occurs. This mode is selected via the OPM bit in the CR2 register.

The one pulse mode uses the Input Capture1 function and the Output Compare1 function.

Procedure:

To use one pulse mode:

1. Load the OC1R register with the value corresponding to the length of the pulse (see the formula in the opposite column).
2. Select the following in the CR1 register:
   - Using the OLVL1 bit, select the level to be applied to the OCMP1 pin after the pulse.
   - Using the OLVL2 bit, select the level to be applied to the OCMP1 pin during the pulse.
   - Select the edge of the active transition on the ICAP1 pin with the IEDG1 bit (the ICAP1 pin must be configured as floating input).
3. Select the following in the CR2 register:
   - Set the OC1E bit, the OCMP1 pin is then dedicated to the Output Compare 1 function.
   - Set the OPM bit.
   - Select the timer clock CC[1:0] (see Table 19 Clock Control Bits).



Then, on a valid event on the ICAP1 pin, the counter is initialized to FCh and OLVL2 bit is loaded on the OCMP1 pin, the ICF1 bit is set and the value FFFDh is loaded in the IC1R register.

Because the ICF1 bit is set when an active edge occurs, an interrupt can be generated if the ICIE bit is set.

Clearing the Input Capture interrupt request (that is, clearing the ICF*i* bit) is done in two steps:

1. Reading the SR register while the ICF*i* bit is set.
2. An access (read or write) to the IC*i*LR register.

The OC1R register value required for a specific timing application can be calculated using the following formula:

$$OCiR\ Value = \frac{t * f_{CPU}}{PRESC} - 5$$

Where:

t = Pulse period (in seconds)

f_{CPU} = PLL output x2 clock frequency in hertz (or $f_{OSC}/2$ if PLL is not enabled)

PRESC = Timer prescaler factor (2, 4, 8 or 8000 depending on the CC[1:0] bits, see Table 19 Clock Control Bits)

When the value of the counter is equal to the value of the contents of the OC1R register, the OLVL1 bit is output on the OCMP1 pin, (See Figure 68).

Notes:

1. The OCF1 bit cannot be set by hardware in one pulse mode but the OCF2 bit can generate an Output Compare interrupt.
2. When the Pulse Width Modulation (PWM) and One Pulse Mode (OPM) bits are both set, the PWM mode is the only active one.
3. If OLVL1=OLVL2 a continuous signal will be seen on the OCMP1 pin.
4. The ICAP1 pin can not be used to perform input capture. The ICAP2 pin can be used to perform input capture (ICF2 can be set and IC2R can be loaded) but the user must take care that the counter is reset each time a valid edge occurs on the ICAP1 pin and ICF1 can also generates interrupt if ICIE is set.
5. When one pulse mode is used OC1R is dedicated to this mode. Nevertheless OC2R and OCF2 can be used to indicate a period of time has been elapsed but cannot generate an output waveform because the level OLVL2 is dedicated to the one pulse mode.

SERIAL PERIPHERAL INTERFACE (cont'd)

10.6.3.2 Slave Select Management

As an alternative to using the $\overline{SS}$ pin to control the Slave Select signal, the application can choose to manage the Slave Select signal by software. This is configured by the SSM bit in the SPICSR register (see Figure 73).

In software management, the external $\overline{SS}$ pin is free for other application uses and the internal $\overline{SS}$ signal level is driven by writing to the SSI bit in the SPICSR register.

In Master mode:

– $\overline{SS}$ internal must be held high continuously

In Slave Mode:

There are two cases depending on the data/clock timing relationship (see Figure 72):

If CPHA = 1 (data latched on second clock edge):

– $\overline{SS}$ internal must be held low during the entire transmission. This implies that in single slave applications the $\overline{SS}$ pin either can be tied to V_{SS}, or made free for standard I/O by managing the $\overline{SS}$ function by software (SSM = 1 and SSI = 0 in the in the SPICSR register)

If CPHA = 0 (data latched on first clock edge):

– $\overline{SS}$ internal must be held low during byte transmission and pulled high between each byte to allow the slave to write to the shift register. If $\overline{SS}$ is not pulled high, a Write Collision error will occur when the slave writes to the shift register (see Section 10.6.5.3).

Figure 72. Generic $\overline{SS}$ Timing Diagram

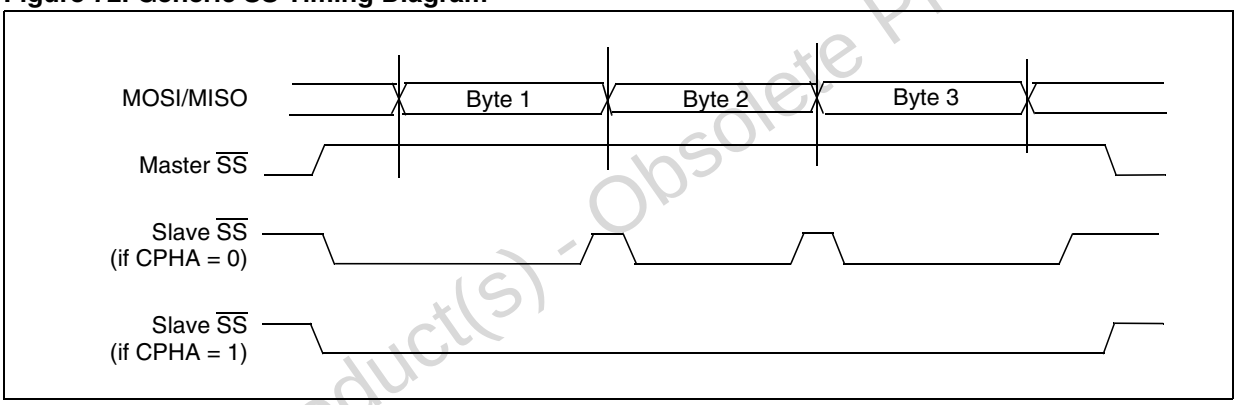
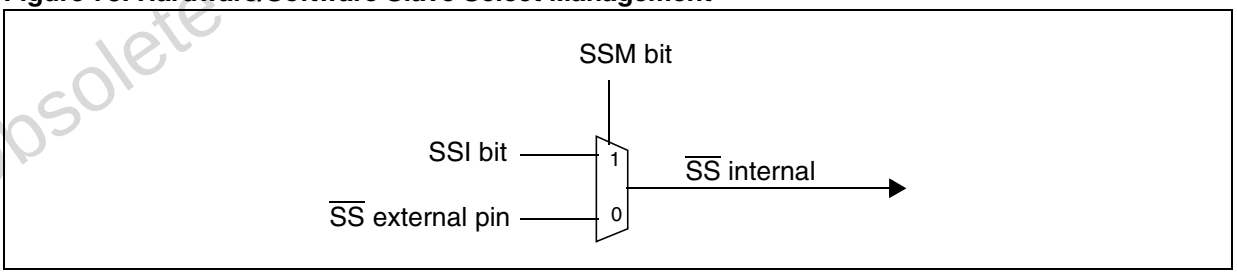


Figure 73. Hardware/Software Slave Select Management



10.7 LINSCI SERIAL COMMUNICATION INTERFACE (LIN MASTER/SLAVE)

10.7.1 Introduction

The Serial Communications Interface (SCI) offers a flexible means of full-duplex data exchange with external equipment requiring an industry standard NRZ asynchronous serial data format. The SCI offers a very wide range of baud rates using two baud rate generator systems.

The LIN-dedicated features support the LIN (Local Interconnect Network) protocol for both master and slave nodes.

This chapter is divided into SCI Mode and LIN mode sections. For information on general SCI communications, refer to the SCI mode section. For LIN applications, refer to both the SCI mode and LIN mode sections.

10.7.2 SCI Features

- Full duplex, asynchronous communications
- NRZ standard format (Mark/Space)
- Independently programmable transmit and receive baud rates up to 500K baud
- Programmable data word length (8 or 9 bits)
- Receive buffer full, Transmit buffer empty and End of Transmission flags
- 2 receiver wake-up modes:
  - Address bit (MSB)
  - Idle line
- Muting function for multiprocessor configurations
- Separate enable bits for Transmitter and Receiver
- Overrun, Noise and Frame error detection
- 6 interrupt sources
  - Transmit data register empty
  - Transmission complete
  - Receive data register full
  - Idle line received
  - Overrun error
  - Parity interrupt
- Parity control:
  - Transmits parity bit
  - Checks parity of received data byte
- Reduced power consumption mode

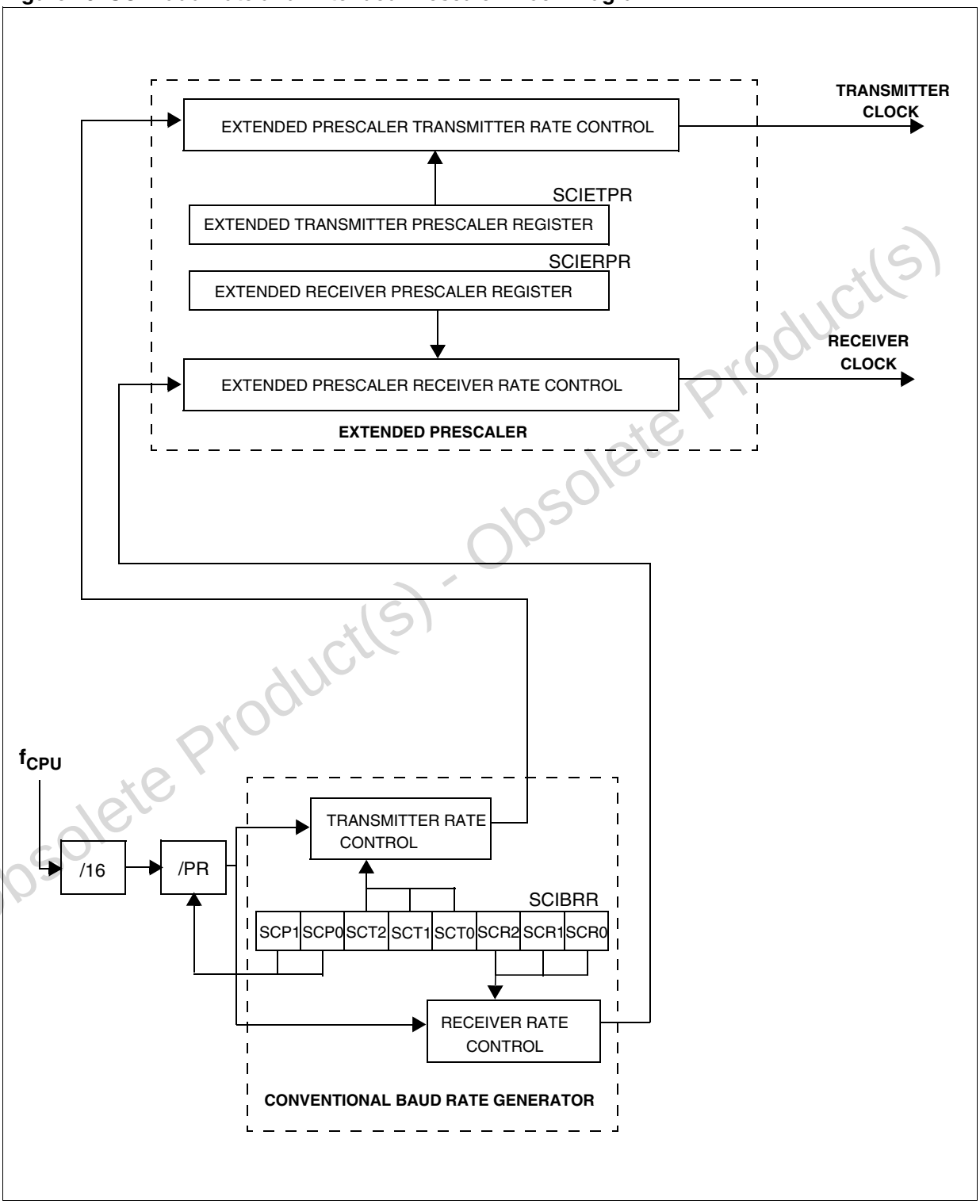
10.7.3 LIN Features

– LIN Master
  – 13-bit LIN Synch Break generation

– LIN Slave
  – Automatic Header Handling
  – Automatic baud rate resynchronization based on recognition and measurement of the LIN Synch Field (for LIN slave nodes)
  – Automatic baud rate adjustment (at CPU frequency precision)
  – 11-bit LIN Synch Break detection capability
  – LIN Parity check on the LIN Identifier Field (only in reception)
  – LIN Error management
  – LIN Header Timeout
  – Hot plugging support

LINSCI™ SERIAL COMMUNICATION INTERFACE (SCI Mode) (cont'd)

Figure 79. SCI Baud Rate and Extended Prescaler Block Diagram



LINSCI™ SERIAL COMMUNICATION INTERFACE (SCI Mode) (cont'd)

CONTROL REGISTER 2 (SCICR2)

Read/Write
Reset Value: 0000 0000 (00h)

7							0
TIE	TCIE	RIE	ILIE	TE	RE	RWU[1)]	SBK[1)]

[1)]This bit has a different function in LIN mode, please refer to the LIN mode register description.

Bit 7 = **TIE** *Transmitter interrupt enable*
This bit is set and cleared by software.
0: Interrupt is inhibited
1: In SCI interrupt is generated whenever TDRE = 1 in the SCISR register

Bit 6 = **TCIE** *Transmission complete interrupt enable*
This bit is set and cleared by software.
0: Interrupt is inhibited
1: An SCI interrupt is generated whenever TC = 1 in the SCISR register

Bit 5 = **RIE** *Receiver interrupt enable*
This bit is set and cleared by software.
0: Interrupt is inhibited
1: An SCI interrupt is generated whenever OR = 1 or RDRF = 1 in the SCISR register

Bit 4 = **ILIE** *Idle line interrupt enable*
This bit is set and cleared by software.
0: Interrupt is inhibited
1: An SCI interrupt is generated whenever IDLE = 1 in the SCISR register.

Bit 3 = **TE** *Transmitter enable*
This bit enables the transmitter. It is set and cleared by software.
0: Transmitter is disabled
1: Transmitter is enabled

Notes:

– During transmission, a "0" pulse on the TE bit ("0" followed by "1") sends a preamble (idle line) after the current word.

– When TE is set there is a 1 bit-time delay before the transmission starts.

Bit 2 = **RE** *Receiver enable*
This bit enables the receiver. It is set and cleared by software.
0: Receiver is disabled in the SCISR register
1: Receiver is enabled and begins searching for a start bit

Bit 1 = **RWU** *Receiver wake-up*
This bit determines if the SCI is in mute mode or not. It is set and cleared by software and can be cleared by hardware when a wake-up sequence is recognized.
0: Receiver in active mode
1: Receiver in mute mode

Notes:

– Before selecting Mute mode (by setting the RWU bit) the SCI must first receive a data byte, otherwise it cannot function in Mute mode with wake-up by Idle line detection.

– In Address Mark Detection Wake-Up configuration (WAKE bit = 1) the RWU bit cannot be modified by software while the RDRF bit is set.

Bit 0 = **SBK** *Send break*
This bit set is used to send break characters. It is set and cleared by software.
0: No break character is transmitted
1: Break characters are transmitted

Note: If the SBK bit is set to "1" and then to "0", the transmitter will send a BREAK word at the end of the current word.

DATA REGISTER (SCIDR)

Read/Write
Reset Value: Undefined

Contains the Received or Transmitted data character, depending on whether it is read from or written to.

7							0
DR7	DR6	DR5	DR4	DR3	DR2	DR1	DR0

The Data register performs a double function (read and write) since it is composed of two registers, one for transmission (TDR) and one for reception (RDR).
The TDR register provides the parallel interface between the internal bus and the output shift register (see Figure 1).
The RDR register provides the parallel interface between the input shift register and the internal bus (see Figure 1).

LINSCI™ SERIAL COMMUNICATION INTERFACE (LIN Mode) (cont'd)

Figure 84. LDIV Read / Write Operations When LDUM = 0

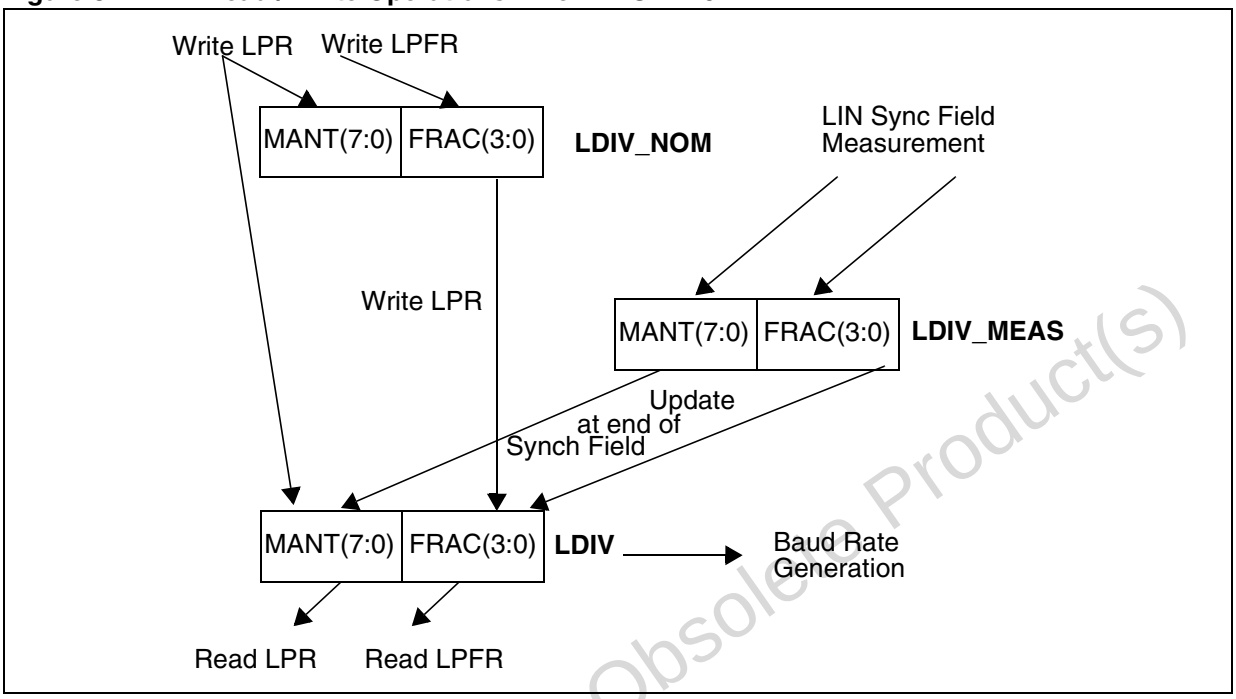
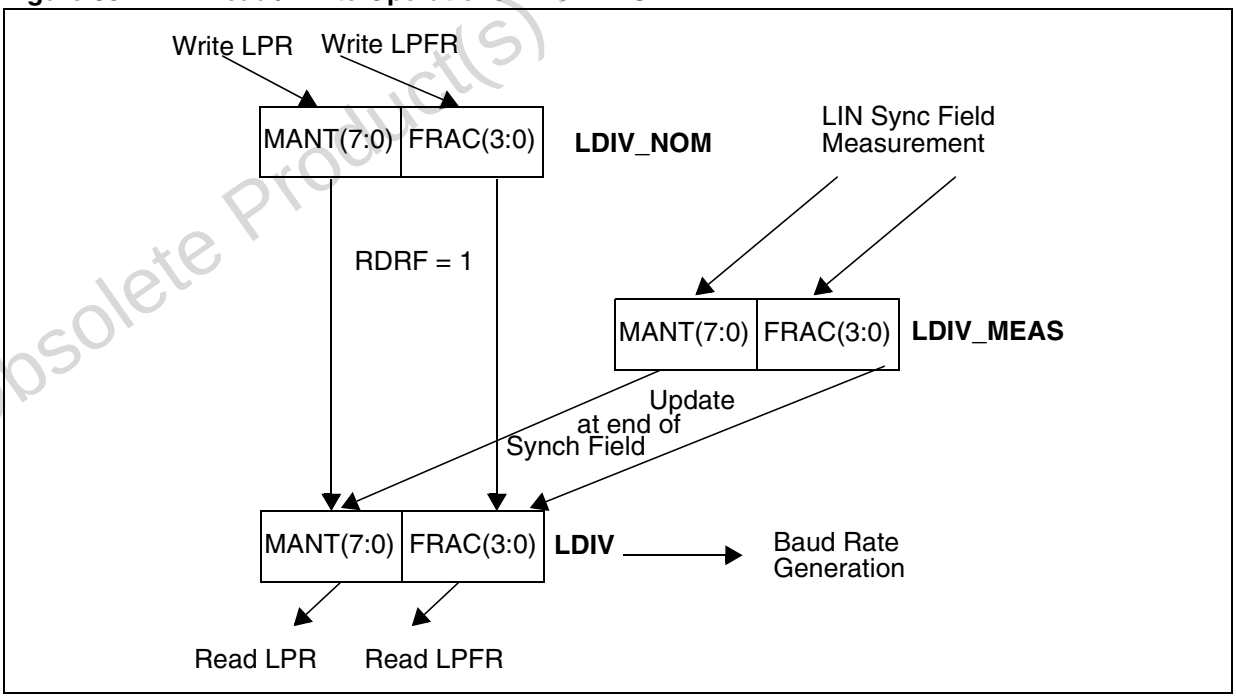


Figure 85. LDIV Read / Write Operations When LDUM = 1



LINSCI™ SERIAL COMMUNICATION INTERFACE (LIN Mode) (cont'd)

10.7.9.7 LINSCI Clock Tolerance

LINSCI Clock Tolerance when unsynchronized

When LIN slaves are unsynchronized (meaning no characters have been transmitted for a relatively long time), the maximum tolerated deviation of the LINSCI clock is +/-15%.

If the deviation is within this range then the LIN Synch Break is detected properly when a new reception occurs.

This is made possible by the fact that masters send 13 low bits for the LIN Synch Break, which can be interpreted as 11 low bits (13 bits -15% = 11.05) by a "fast" slave and then considered as a LIN Synch Break. According to the LIN specification, a LIN Synch Break is valid when its duration is greater than t_{SBRKTS} = 10. This means that the LIN Synch Break must last at least 11 low bits.

Note: If the period desynchronization of the slave is +15% (slave too slow), the character "00h" which represents a sequence of 9 low bits must not be interpreted as a break character (9 bits + 15% = 10.35). Consequently, a valid LIN Synch break must last at least 11 low bits.

LINSCI Clock Tolerance when Synchronized

When synchronization has been performed, following reception of a LIN Synch Break, the LINSCI, in LIN mode, has the same clock deviation tolerance as in SCI mode, which is explained below:

During reception, each bit is oversampled 16 times. The mean of the 8th, 9th and 10th samples is considered as the bit value.

Consequently, the clock frequency should not vary more than 6/16 (37.5%) within one bit.

The sampling clock is resynchronized at each start bit, so that when receiving 10 bits (one start bit, 1 data byte, 1 stop bit), the clock deviation should not exceed 3.75%.

10.7.9.8 Clock Deviation Causes

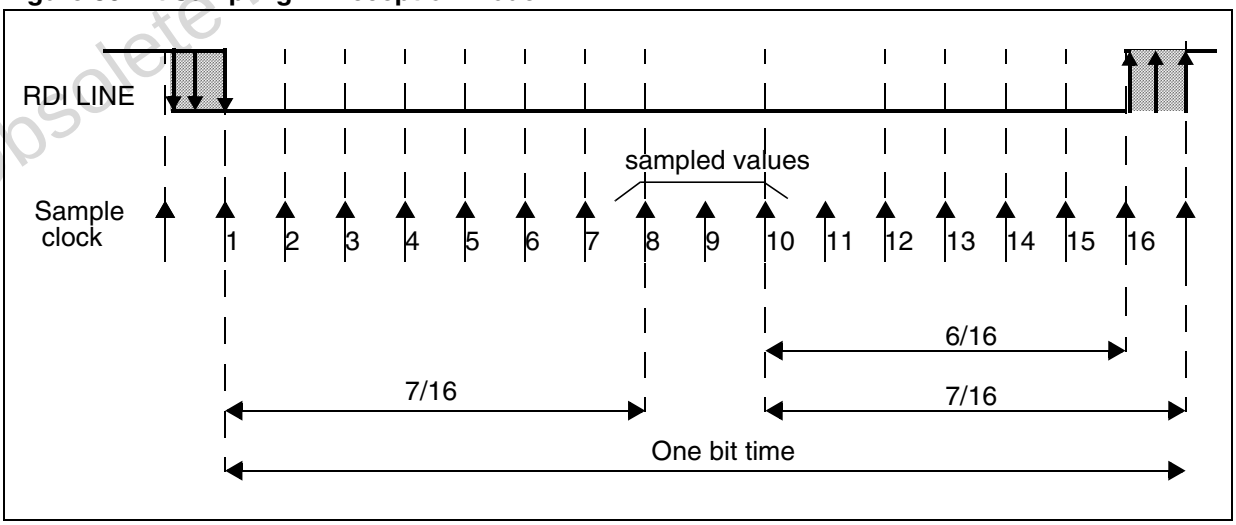
The causes which contribute to the total deviation are:

- D_{TRA}: Deviation due to transmitter error.
  Note: The transmitter can be either a master or a slave (in case of a slave listening to the response of another slave).
- D_{MEAS}: Error due to the LIN Synch measurement performed by the receiver.
- D_{QUANT}: Error due to the baud rate quantization of the receiver.
- D_{REC}: Deviation of the local oscillator of the receiver: This deviation can occur during the reception of one complete LIN message assuming that the deviation has been compensated at the beginning of the message.
- D_{TCL}: Deviation due to the transmission line (generally due to the transceivers)

All the deviations of the system should be added and compared to the LINSCI clock tolerance:

$D_{TRA} + D_{MEAS} + D_{QUANT} + D_{REC} + D_{TCL} < 3.75\%$

Figure 86.Bit Sampling in Reception Mode



LINSCI™ SERIAL COMMUNICATION INTERFACE (LIN Mode) (cont'd)

LIN PRESCALER FRACTION REGISTER (LPFR)

Read/Write

Reset Value: 0000 0000 (00h)

7							0
0	0	0	0	LPFR 3	LPFR 2	LPFR 1	LPFR 0

Bits 7:4 = Reserved.

Bits 3:0 = **LPFR[3:0]** *Fraction of LDIV*

These 4 bits define the fraction of the LIN Divider (LDIV):

LPFR[3:0]	Fraction (LDIV)
0h	0
1h	1/16
...	...
Eh	14/16
Fh	15/16

1. When initializing LDIV, the LPFR register must be written first. Then, the write to the LPR register will effectively update LDIV and so the clock generation.
2. In LIN Slave mode, if the LPR[7:0] register is equal to 00h, the transceiver and receiver input clocks are switched off.

Examples of LDIV coding:

Example 1: LPR = 27d and LPFR = 12d

This leads to:

Mantissa (LDIV) = 27d

Fraction (LDIV) = 12/16 = 0.75d

Therefore LDIV = 27.75d

Example 2: LDIV = 25.62d

This leads to:

LPFR = rounded(16*0.62d)

= rounded(9.92d) = 10d = Ah

LPR = mantissa (25.620d) = 25d = 1Bh

Example 3: LDIV = 25.99d

This leads to:

LPFR = rounded(16*0.99d)

= rounded(15.84d) = 16d

LINSCI™ SERIAL COMMUNICATION INTERFACE (LIN Mode) (cont'd)

LIN HEADER LENGTH REGISTER (LHLR)
Read Only
Reset Value: 0000 0000 (00h).

7							0
LHL7	LHL6	LHL5	LHL4	LHL3	LHL2	LHL1	LHL0

Note: In LIN Slave mode when LASE = 1 or LHDM = 1, the LHLR register is accessible at the address of the SCIERPR register.

Otherwise this register is always read as 00h.

Bits 7:0 = **LHL[7:0]** *LIN Header Length.*

This is a read-only register, which is updated by hardware if one of the following conditions occurs:
- After each break detection, it is loaded with "FFh".
- If a timeout occurs on T_{HEADER}, it is loaded with 00h.
- After every successful LIN Header reception (at the same time than the setting of LHDF bit), it is loaded with a value (LHL) which gives access to the number of bit times of the LIN header length (T_{HEADER}). The coding of this value is explained below:

LHL Coding:

T_{HEADER_MAX} = 57
LHL(7:2) represents the mantissa of (57 - T_{HEADER})
LHL(1:0) represents the fraction (57 - T_{HEADER})

LHL[7:2]	Mantissa (57 - T_{HEADER})	Mantissa (T_{HEADER})
0h	0	57
1h	1	56
...	...	...
39h	56	1
3Ah	57	0
3Bh	58	Never Occurs
...	...	...
3Eh	62	Never Occurs
3Fh	63	Initial value

LHL[1:0]	Fraction (57 - T_{HEADER})
0h	0
1h	1/4
2h	1/2
3h	3/4

Example of LHL coding:

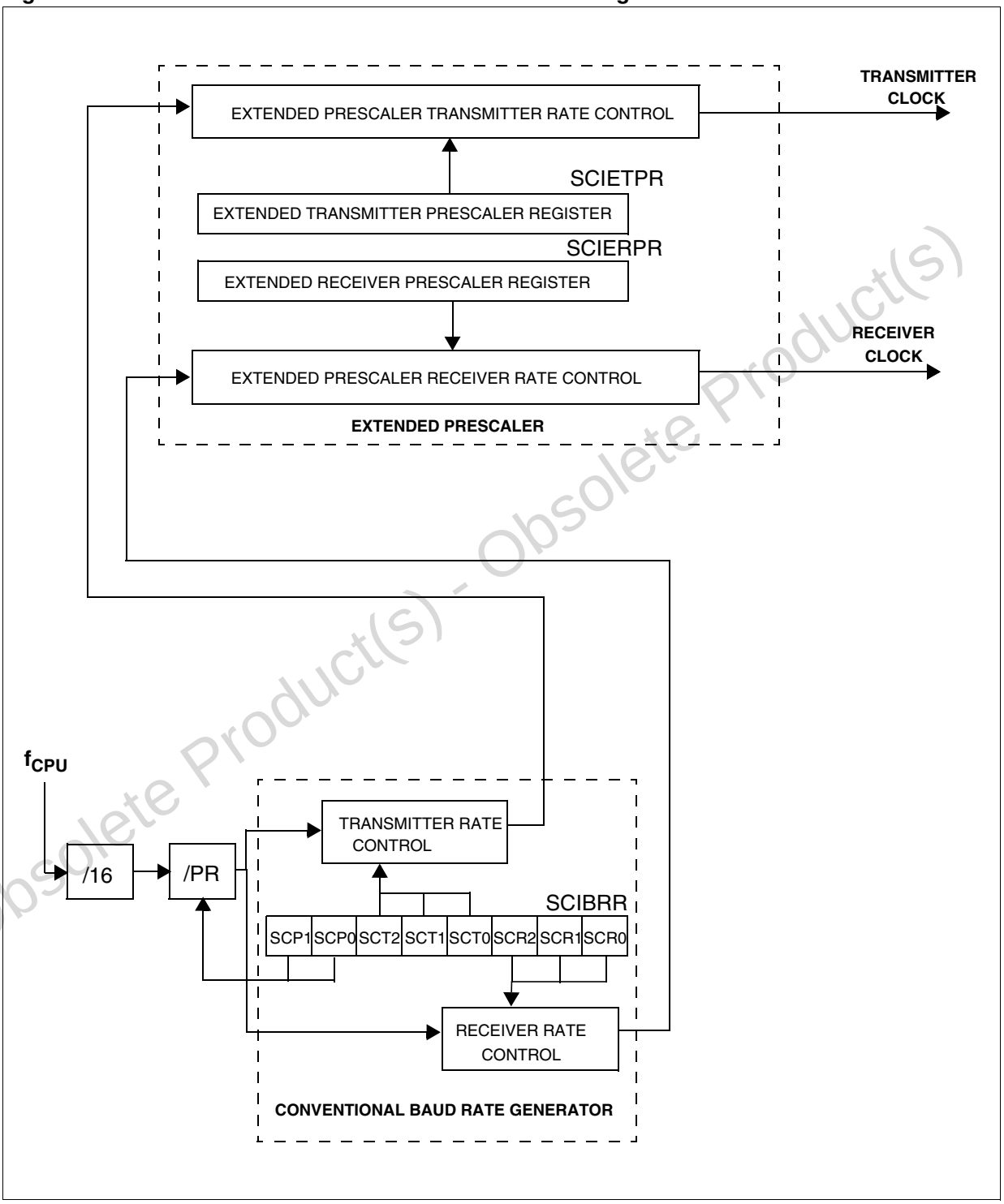
Example 1: LHL = 33h = 001100 11b
LHL(7:3) = 1100b = 12d
LHL(1:0) = 11b = 3d
This leads to:
Mantissa (57 - T_{HEADER}) = 12d
Fraction (57 - T_{HEADER}) = 3/4 = 0.75
Therefore:
(57 - T_{HEADER}) = 12.75d
and T_{HEADER} = 44.25d

Example 2:
57 - T_{HEADER} = 36.21d
LHL(1:0) = rounded(4*0.21d) = 1d
LHL(7:2) = Mantissa (36.21d) = 36d = 24h
Therefore LHL(7:0) = 10010001 = 91h

Example 3:
57 - T_{HEADER} = 36.90d
LHL(1:0) = rounded(4*0.90d) = 4d
The carry must be propagated to the matissa:
LHL(7:2) = Mantissa (36.90d) + 1 = 37d =
Therefore LHL(7:0) = 10110000 = A0h

LINSCI™ SERIAL COMMUNICATION INTERFACE (LIN Master Only) (Cont'd)

Figure 90. SCI Baud Rate and Extended Prescaler Block Diagram



INSTRUCTION SET OVERVIEW (Cont'd)

11.1.6 Indirect Indexed (Short, Long)

This is a combination of indirect and short indexed addressing modes. The operand is referenced by its memory address, which is defined by the unsigned addition of an index register value (X or Y) with a pointer value located in memory. The pointer address follows the opcode.

The indirect indexed addressing mode consists of two submodes:

Indirect Indexed (Short)

The pointer address is a byte, the pointer size is a byte, thus allowing 00 - 1FE addressing space, and requires 1 byte after the opcode.

Indirect Indexed (Long)

The pointer address is a byte, the pointer size is a word, thus allowing 64 Kbyte addressing space, and requires 1 byte after the opcode.

Table 32. Instructions Supporting Direct, Indexed, Indirect and Indirect Indexed Addressing Modes

Long and Short Instructions	Function
LD	Load
CP	Compare
AND, OR, XOR	Logical Operations
ADC, ADD, SUB, SBC	Arithmetic Additions/Substractions operations
BCP	Bit Compare

Short Instructions Only	Function
CLR	Clear
INC, DEC	Increment/Decrement
TNZ	Test Negative or Zero
CPL, NEG	1 or 2 Complement
BSET, BRES	Bit Operations
BTJT, BTJF	Bit Test and Jump Operations
SLL, SRL, SRA, RLC, RRC	Shift and Rotate Operations
SWAP	Swap Nibbles
CALL, JP	Call or Jump subroutine

11.1.7 Relative mode (Direct, Indirect)

This addressing mode is used to modify the PC register value, by adding an 8-bit signed offset to it.

Available Relative Direct/Indirect Instructions	Function
JRxx	Conditional Jump
CALLR	Call Relative

The relative addressing mode consists of two submodes:

Relative (Direct)

The offset is following the opcode.

Relative (Indirect)

The offset is defined in memory, which address follows the opcode.

INSTRUCTION SET OVERVIEW (Cont'd)

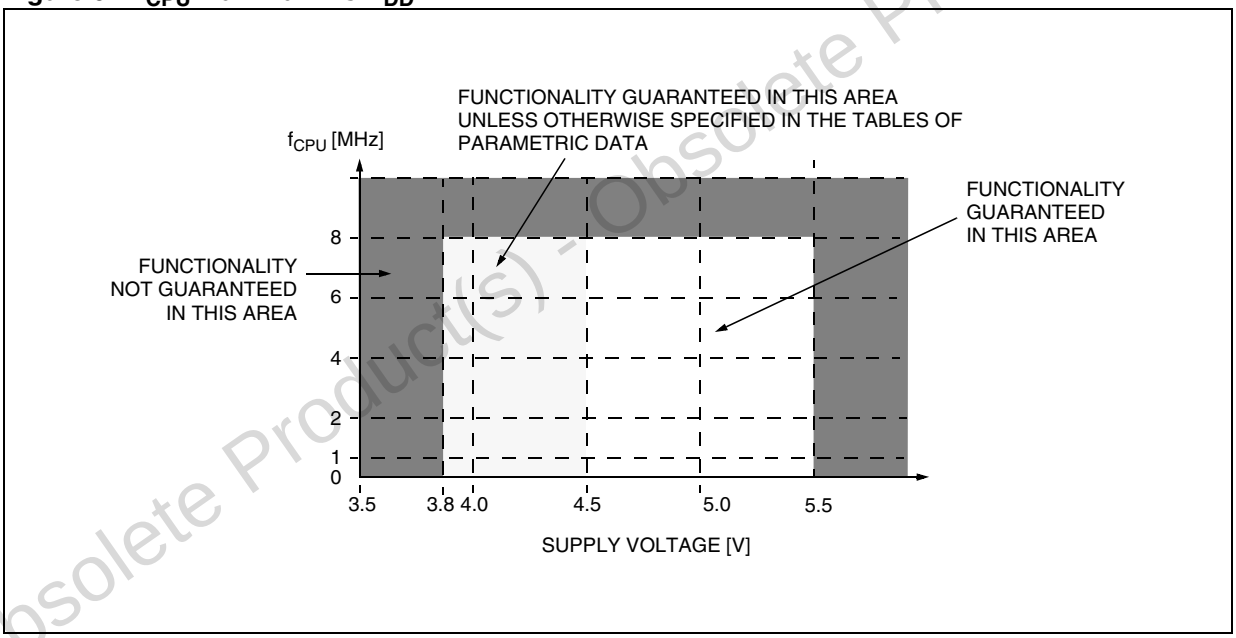
Mnemo	Description	Function/Example	Dst	Src	I1	H	I0	N	Z	C
JRULE	Jump if (C + Z = 1)	Unsigned <=								
LD	Load	dst <= src	reg, M	M, reg				N	Z	
MUL	Multiply	X,A = X * A	A, X, Y	X, Y, A		0				0
NEG	Negate (2's compl)	neg $10	reg, M					N	Z	C
NOP	No Operation									
OR	OR operation	A = A + M	A	M				N	Z	
POP	Pop from the Stack	pop reg	reg	M						
		pop CC	CC	M	I1	H	I0	N	Z	C
PUSH	Push onto the Stack	push Y	M	reg, CC						
RCF	Reset carry flag	C = 0								0
RET	Subroutine Return									
RIM	Enable Interrupts	I1:0 = 10 (level 0)			1		0			
RLC	Rotate left true C	C <= A <= C	reg, M					N	Z	C
RRC	Rotate right true C	C => A => C	reg, M					N	Z	C
RSP	Reset Stack Pointer	S = Max allowed								
SBC	Substract with Carry	A = A - M - C	A	M				N	Z	C
SCF	Set carry flag	C = 1								1
SIM	Disable Interrupts	I1:0 = 11 (level 3)			1		1			
SLA	Shift left Arithmetic	C <= A <= 0	reg, M					N	Z	C
SLL	Shift left Logic	C <= A <= 0	reg, M					N	Z	C
SRL	Shift right Logic	0 => A => C	reg, M					0	Z	C
SRA	Shift right Arithmetic	A7 => A => C	reg, M					N	Z	C
SUB	Substraction	A = A - M	A	M				N	Z	C
SWAP	SWAP nibbles	A7-A4 <=> A3-A0	reg, M					N	Z	
TNZ	Test for Neg & Zero	tnz lbl1						N	Z	
TRAP	S/W trap	S/W interrupt			1		1			
WFI	Wait for Interrupt				1		0			
XOR	Exclusive OR	A = A XOR M	A	M				N	Z	

12.3 OPERATING CONDITIONS

12.3.1 General Operating Conditions

Symbol	Parameter	Conditions	Min	Max	Unit
f_{CPU}	Internal clock frequency		0	8	MHz
V_{DD}	Extended Operating voltage	No Flash Write/Erase. Analog parameters not guaranteed.	3.8	4.5	V
	Standard Operating Voltage		4.5	5.5	
	Operating Voltage for Flash Write/Erase	V_{PP} = 11.4 to 12.6V	4.5	5.5	
T_A	Ambient temperature range	1 Suffix Version	0	70	°C
		5 Suffix Version	-10	85	
		6 Suffix Version	-40	85	
		7 Suffix Version		105	
		3 Suffix Version		125	

Figure 97. f_{CPU} Maximum vs V_{DD}



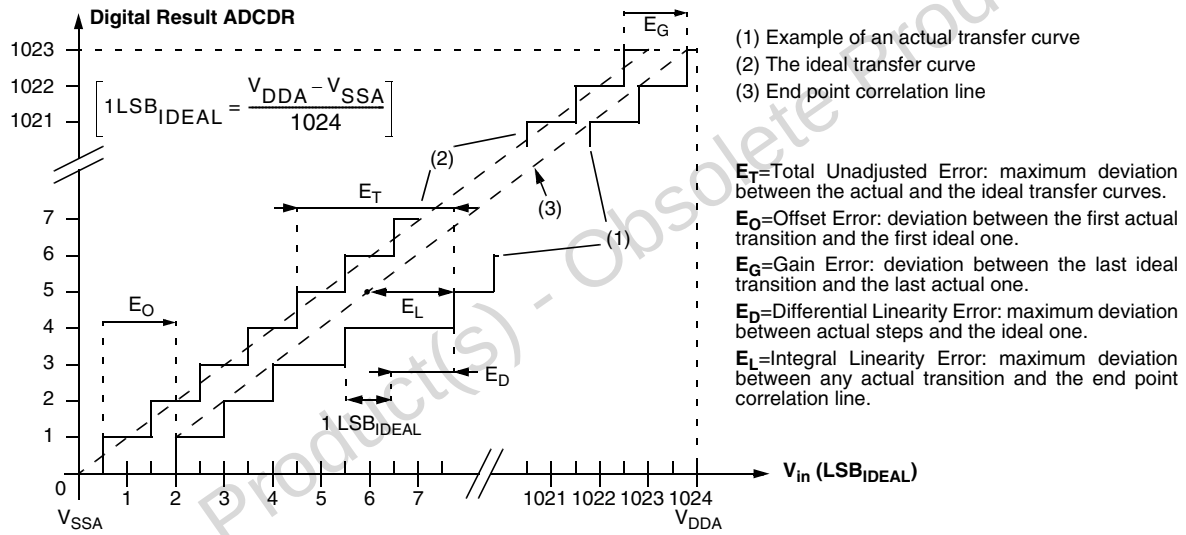
Note: It is mandatory to connect all available V_{DD} and V_{DDA} pins to the supply voltage and all V_{SS} and V_{SSA} pins to ground.

ADC CHARACTERISTICS (Cont'd)

ADC Accuracy with f_{CPU} = 8 MHz, f_{ADC} = 4 MHz R_{AIN} < 10kΩ, V_{DD} = 5V

Symbol	Parameter	Conditions	Typ	Max	Unit
$\|E_T\|$	Total unadjusted error[1)]		3.2	5	LSB
$\|E_O\|$	Offset error[1)]		1	4	
$\|E_G\|$	Gain Error[1)]		0.7		
$\|E_D\|$	Differential linearity error[1)]		1.5	2.3	
$\|E_L\|$	Integral linearity error[1)]		1.2	3.6	

Figure 123. ADC Accuracy Characteristics



Notes:

1. Data based on characterization results, not tested in production. ADC Accuracy vs. Negative Injection Current: Injecting negative current on any of the standard (non-robust) analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to standard analog pins which may potentially inject negative current. The effect of negative injection current on robust pins is specified in Section 12.9.

Any positive injection current within the limits specified for $I_{INJ(PIN)}$ and $\Sigma I_{INJ(PIN)}$ in Section 12.9 does not affect the ADC accuracy.