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Details

Product Status	Last Time Buy
Core Processor	8051
Core Size	8-Bit
Speed	25MHz
Connectivity	SMBus (2-Wire/I ² C), SPI, UART/USART
Peripherals	POR, PWM, WDT
Number of I/O	17
Program Memory Size	16KB (16K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	512 x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.6V
Data Converters	-
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	24-SSOP (0.154", 3.90mm Width)
Supplier Device Package	24-QSOP
Purchase URL	https://www.e-xfl.com/product-detail/silicon-labs/c8051f808-gur

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Table 7.6. Flash Electrical Characteristics

Parameter	Conditions	Min	Typ	Max	Units
Flash Size (Note 1)	C8051F80x and C8051F810/1 C8051F812/3/4/5/6/7/8/9 and C8051F82x C8051F830/1/2/3/4/5		16384 8192 4096		bytes bytes bytes
Endurance (Erase/Write)		10000	—	—	cycles
Erase Cycle Time	25 MHz Clock	15	20	26	ms
Write Cycle Time	25 MHz Clock	15	20	26	μs
Clock Speed during Flash Write/Erase Operations		1	—	—	MHz
Note: Includes Security Lock Byte.					

Table 7.7. Internal High-Frequency Oscillator Electrical Characteristics

$V_{DD} = 1.8$ to 3.6 V; $T_A = -40$ to $+85$ °C unless otherwise specified. Use factory-calibrated settings.

Parameter	Conditions	Min	Typ	Max	Units
Oscillator Frequency	IFCN = 11b	24	24.5	25	MHz
Oscillator Supply Current	25 °C, $V_{DD} = 3.0$ V, OSCICN.7 = 1, OCSICN.5 = 0	—	350	650	μA

Table 7.8. Capacitive Sense Electrical Characteristics

$V_{DD} = 1.8$ to 3.6 V; $T_A = -40$ to $+85$ °C unless otherwise specified.

Parameter	Conditions	Min	Typ	Max	Units
Conversion Time	Single Conversion	26	38	50	μs
Capacitance per Code		—	1	—	fF
External Capacitive Load		—	—	45	pF
Quantization Noise ¹	RMS	—	3	—	fF
	Peak-to-Peak	—	20	—	fF
Supply Current	CS module bias current, 25 °C	—	40	60	μA
	CS module alone, maximum code output, 25 °C	—	75	105	μA
	Wake-on-CS Threshold ² , 25 °C	—	150	165	μA
Notes:					
1. RMS Noise is equivalent to one standard deviation. Peak-to-peak noise encompasses ± 3.3 standard deviations.					
2. Includes only current from regulator, CS module, and MCU in suspend mode.					

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8.3.3. Settling Time Requirements

A minimum tracking time is required before each conversion to ensure that an accurate conversion is performed. This tracking time is determined by any series impedance, including the AMUX0 resistance, the the ADC0 sampling capacitance, and the accuracy required for the conversion. In delayed tracking mode, three SAR clocks are used for tracking at the start of every conversion. For many applications, these three SAR clocks will meet the minimum tracking time requirements.

Figure 8.3 shows the equivalent ADC0 input circuit. The required ADC0 settling time for a given settling accuracy (SA) may be approximated by Equation 8.1. See Table 7.9 for ADC0 minimum settling time requirements as well as the mux impedance and sampling capacitor values.

$$t = \ln\left(\frac{2^n}{SA}\right) \times R_{TOTAL} C_{SAMPLE}$$

Equation 8.1. ADC0 Settling Time Requirements

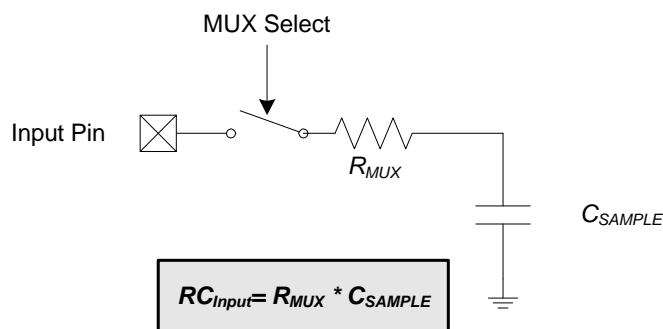
Where:

SA is the settling accuracy, given as a fraction of an LSB (for example, 0.25 to settle within 1/4 LSB)

t is the required settling time in seconds

R_{TOTAL} is the sum of the AMUX0 resistance and any external source resistance.

n is the ADC resolution in bits (10).



Note: See electrical specification tables for R_{MUX} and C_{SAMPLE} parameters.

Figure 8.3. ADC0 Equivalent Input Circuits

SFR Definition 8.4. ADC0CN: ADC0 Control

Bit	7	6	5	4	3	2	1	0
Name	AD0EN	AD0TM	AD0INT	AD0BUSY	AD0WINT	AD0CM[2:0]		
Type	R/W	R/W	R/W	R/W	R/W	R/W		
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xE8; Bit-Addressable

Bit	Name	Function						
7	AD0EN	<p>ADC0 Enable Bit.</p> <p>0: ADC0 Disabled. ADC0 is in low-power shutdown. 1: ADC0 Enabled. ADC0 is active and ready for data conversions.</p>						
6	AD0TM	<p>ADC0 Track Mode Bit.</p> <p>0: Normal Track Mode: When ADC0 is enabled, tracking is continuous unless a conversion is in progress. Conversion begins immediately on start-of-conversion event, as defined by AD0CM[2:0]. 1: Delayed Track Mode: When ADC0 is enabled, input is tracked when a conversion is not in progress. A start-of-conversion signal initiates three SAR clocks of additional tracking, and then begins the conversion.</p>						
5	AD0INT	<p>ADC0 Conversion Complete Interrupt Flag.</p> <p>0: ADC0 has not completed a data conversion since AD0INT was last cleared. 1: ADC0 has completed a data conversion.</p>						
4	AD0BUSY	<table style="width: 100%; border: none;"> <tr> <td style="width: 40%; border: none;">ADC0 Busy Bit.</td> <td style="width: 20%; border: none;">Read:</td> <td style="width: 40%; border: none;">Write:</td> </tr> <tr> <td style="border: none;"></td> <td style="border: none;">0: ADC0 conversion is not in progress. 1: ADC0 conversion is in progress.</td> <td style="border: none;">0: No Effect. 1: Initiates ADC0 Conversion if AD0CM[2:0] = 000b</td> </tr> </table>	ADC0 Busy Bit.	Read:	Write:		0: ADC0 conversion is not in progress. 1: ADC0 conversion is in progress.	0: No Effect. 1: Initiates ADC0 Conversion if AD0CM[2:0] = 000b
ADC0 Busy Bit.	Read:	Write:						
	0: ADC0 conversion is not in progress. 1: ADC0 conversion is in progress.	0: No Effect. 1: Initiates ADC0 Conversion if AD0CM[2:0] = 000b						
3	AD0WINT	<p>ADC0 Window Compare Interrupt Flag.</p> <p>0: ADC0 Window Comparison Data match has not occurred since this flag was last cleared. 1: ADC0 Window Comparison Data match has occurred.</p>						
2:0	AD0CM[2:0]	<p>ADC0 Start of Conversion Mode Select.</p> <p>000: ADC0 start-of-conversion source is write of 1 to AD0BUSY. 001: ADC0 start-of-conversion source is overflow of Timer 0. 010: ADC0 start-of-conversion source is overflow of Timer 2. 011: ADC0 start-of-conversion source is overflow of Timer 1. 100: ADC0 start-of-conversion source is rising edge of external CNVSTR. 101–111: Reserved.</p>						

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13.5. CS0 Conversion Accumulator

CS0 can be configured to accumulate multiple conversions on an input channel. The number of samples to be accumulated is configured using the CS0ACU2:0 bits (CS0CF2:0). The accumulator can accumulate 1, 4, 8, 16, 32, or 64 samples. After the defined number of samples have been accumulated, the result is converted to a 16-bit value by dividing the 22-bit accumulator by either 1, 4, 8, 16, 32, or 64 (depending on the CS0ACU[2:0] setting) and copied to the CS0DH:CS0DL SFRs.

Table 13.1. Operation with Auto-scan and Accumulate

Auto-Scan Enabled	Accumulator Enabled	CS0 Conversion Complete Interrupt Behavior	CS0 Greater Than Interrupt Behavior	CS0MX Behavior
N	N	CS0INT Interrupt serviced after 1 conversion completes	Interrupt serviced after 1 conversion completes if value in CS0DH:CS0DL is greater than CS0THH:CS0THL	CS0MX unchanged.
N	Y	CS0INT Interrupt serviced after <i>M</i> conversions complete	Interrupt serviced after <i>M</i> conversions complete if value in 16-bit accumulator is greater than CS0THH:CS0THL	CS0MX unchanged.
Y	N	CS0INT Interrupt serviced after 1 conversion completes	Interrupt serviced after conversion completes if value in CS0DH:CS0DL is greater than CS0THH:CS0THL; Auto-Scan stopped	If greater-than comparator detects conversion value is greater than CS0THH:CS0THL, CMUX0 is left unchanged; otherwise, CMUX0 updates to the next channel (CS0MX + 1) and wraps back to CS0SS after passing CS0SE
Y	Y	CS0INT Interrupt serviced after <i>M</i> conversions complete	Interrupt serviced after <i>M</i> conversions complete if value in 16-bit accumulator is greater than CS0THH:CS0THL; Auto-Scan stopped	If greater-than comparator detects conversion value is greater than CS0THH:CS0THL, CS0MX is left unchanged; otherwise, CS0MX updates to the next channel (CS0MX + 1) and wraps back to CS0SS after passing CS0SE
M = Accumulator setting (1x, 4x, 8x, 16x, 32x, 64x)				

SFR Definition 13.1. CS0CN: Capacitive Sense Control

Bit	7	6	5	4	3	2	1	0
Name	CS0EN		CS0INT	CS0BUSY	CS0CMPEN			CS0CMPF
Type	R/W	R	R/W	R/W	R/W	R	R	R
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xB0; Bit-Addressable

Bit	Name	Description
7	CS0EN	CS0 Enable. 0: CS0 disabled and in low-power mode. 1: CS0 enabled and ready to convert.
6	Unused	Read = 0b; Write = Don't care
5	CS0INT	CS0 Interrupt Flag. 0: CS0 has not completed a data conversion since the last time CS0INT was cleared. 1: CS0 has completed a data conversion. This bit is not automatically cleared by hardware.
4	CS0BUSY	CS0 Busy. Read: 0: CS0 conversion is complete or a conversion is not currently in progress. 1: CS0 conversion is in progress. Write: 0: No effect. 1: Initiates CS0 conversion if CS0CM[2:0] = 000b, 110b, or 111b.
3	CS0CMPEN	CS0 Digital Comparator Enable Bit. Enables the digital comparator, which compares accumulated CS0 conversion output to the value stored in CS0THH:CS0THL. 0: CS0 digital comparator disabled. 1: CS0 digital comparator enabled.
2:1	Unused	Read = 00b; Write = Don't care
0	CS0CMPF	CS0 Digital Comparator Interrupt Flag. 0: CS0 result is smaller than the value set by CS0THH and CS0THL since the last time CS0CMPF was cleared. 1: CS0 result is greater than the value set by CS0THH and CS0THL since the last time CS0CMPF was cleared.
Note: On waking from suspend mode due to a CS0 greater-than comparator event, the CS0CN register should be accessed only after at least two system clock cycles have elapsed.		

14.2. CIP-51 Register Descriptions

Following are descriptions of SFRs related to the operation of the CIP-51 System Controller. Reserved bits should always be written to the value indicated in the SFR description. Future product versions may use these bits to implement new features in which case the reset value of the bit will be the indicated value, selecting the feature's default state. Detailed descriptions of the remaining SFRs are included in the sections of the data sheet associated with their corresponding system function.

SFR Definition 14.1. DPL: Data Pointer Low Byte

Bit	7	6	5	4	3	2	1	0
Name	DPL[7:0]							
Type	R/W							
Reset	0	0	0	0	0	0	0	0

SFR Address = 0x82

Bit	Name	Function
7:0	DPL[7:0]	Data Pointer Low. The DPL register is the low byte of the 16-bit DPTR.

SFR Definition 14.2. DPH: Data Pointer High Byte

Bit	7	6	5	4	3	2	1	0
Name	DPH[7:0]							
Type	R/W							
Reset	0	0	0	0	0	0	0	0

SFR Address = 0x83

Bit	Name	Function
7:0	DPH[7:0]	Data Pointer High. The DPH register is the high byte of the 16-bit DPTR.

24. Cyclic Redundancy Check Unit (CRC0)

C8051F80x-83x devices include a cyclic redundancy check unit (CRC0) that can perform a CRC using a 16-bit or 32-bit polynomial. CRC0 accepts a stream of 8-bit data written to the CRC0IN register. CRC0 posts the 16-bit or 32-bit result to an internal register. The internal result register may be accessed indirectly using the CRC0PNT bits and CRC0DAT register, as shown in Figure 24.1. CRC0 also has a bit reverse register for quick data manipulation.

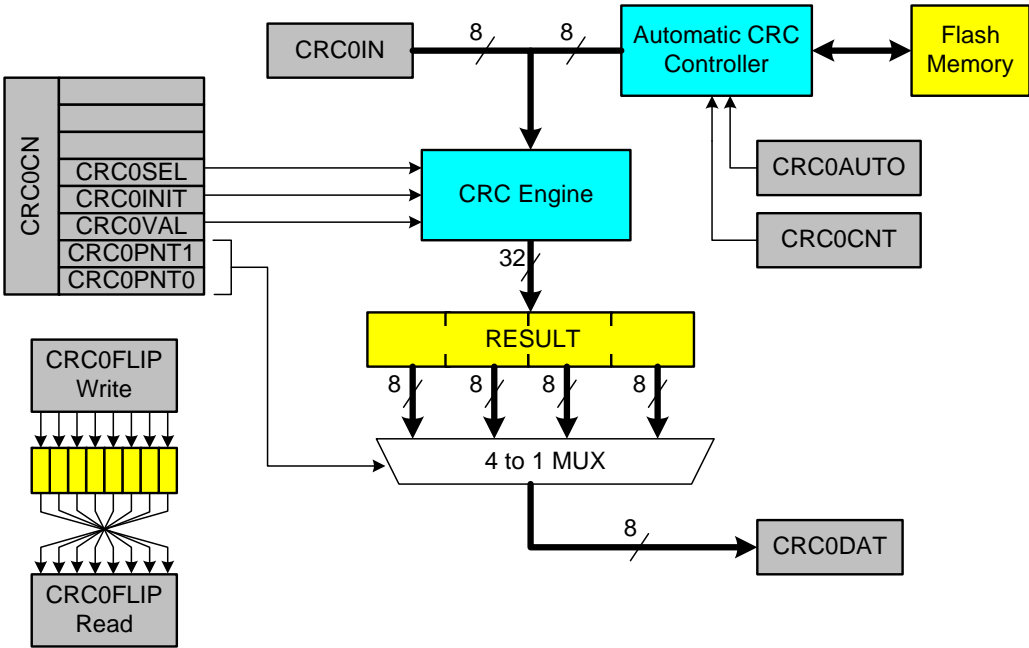


Figure 24.1. CRC0 Block Diagram

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24.1. 16-bit CRC Algorithm

The C8051F80x-83x CRC unit calculates the 16-bit CRC MSB-first, using a poly of 0x1021. The following describes the 16-bit CRC algorithm performed by the hardware:

1. XOR the most-significant byte of the current CRC result with the input byte. If this is the first iteration of the CRC unit, the current CRC result will be the set initial value (0x0000 or 0xFFFF).
2. If the MSB of the CRC result is set, left-shift the CRC result, and then XOR the CRC result with the polynomial (0x1021).
3. If the MSB of the CRC result is not set, left-shift the CRC result.
4. Repeat at Step 2 for the number of input bits (8).

For example, the 16-bit C8051F80x-83x CRC algorithm can be described by the following code:

```
unsigned short UpdateCRC (unsigned short CRC_acc, unsigned char CRC_input){
    unsigned char i;                // loop counter
    #define POLY 0x1021
    // Create the CRC "dividend" for polynomial arithmetic (binary arithmetic
    // with no carries)
    CRC_acc = CRC_acc ^ (CRC_input << 8);
    // "Divide" the poly into the dividend using CRC XOR subtraction
    // CRC_acc holds the "remainder" of each divide
    // Only complete this division for 8 bits since input is 1 byte
    for (i = 0; i < 8; i++)
    {
        // Check if the MSB is set (if MSB is 1, then the POLY can "divide"
        // into the "dividend")
        if ((CRC_acc & 0x8000) == 0x8000)
        {
            // if so, shift the CRC value, and XOR "subtract" the poly
            CRC_acc = CRC_acc << 1;
            CRC_acc ^= POLY;
        }
        else
        {
            // if not, just shift the CRC value
            CRC_acc = CRC_acc << 1;
        }
    }
    return CRC_acc; // Return the final remainder (CRC value)
}
```

Table 24.1 lists example input values and the associated outputs using the 16-bit C8051F80x-83x CRC algorithm (an initial value of 0xFFFF is used):

Table 24.1. Example 16-bit CRC Outputs

Input	Output
0x63	0xBD35
0xAA, 0xBB, 0xCC	0x6CF6
0x00, 0x00, 0xAA, 0xBB, 0xCC	0xB166

24.2. 32-bit CRC Algorithm

The C8051F80x-83x CRC unit calculates the 32-bit CRC using a poly of 0x04C11DB7. The CRC-32 algorithm is "reflected", meaning that all of the input bytes and the final 32-bit output are bit-reversed in the processing engine. The following is a description of a simplified CRC algorithm that produces results identical to the hardware:

1. XOR the least-significant byte of the current CRC result with the input byte. If this is the first iteration of the CRC unit, the current CRC result will be the set initial value (0x00000000 or 0xFFFFFFFF).
2. Right-shift the CRC result.
3. If the LSB of the CRC result is set, XOR the CRC result with the reflected polynomial (0xEDB88320).
4. Repeat at Step 2 for the number of input bits (8).

For example, the 32-bit C8051F80x-83x CRC algorithm can be described by the following code:

```
unsigned long UpdateCRC (unsigned long CRC_acc, unsigned char CRC_input){
    unsigned char i; // loop counter
    #define POLY 0xEDB88320 // bit-reversed version of the poly 0x04C11DB7
    // Create the CRC "dividend" for polynomial arithmetic (binary arithmetic
    // with no carries)
    CRC_acc = CRC_acc ^ CRC_input;
    // "Divide" the poly into the dividend using CRC XOR subtraction
    // CRC_acc holds the "remainder" of each divide
    // Only complete this division for 8 bits since input is 1 byte
    for (i = 0; i < 8; i++)
    {
        // Check if the MSB is set (if MSB is 1, then the POLY can "divide"
        // into the "dividend")
        if ((CRC_acc & 0x00000001) == 0x00000001)
        {
            // if so, shift the CRC value, and XOR "subtract" the poly
            CRC_acc = CRC_acc >> 1;
            CRC_acc ^= POLY;
        }
        else
        {
            // if not, just shift the CRC value
            CRC_acc = CRC_acc >> 1;
        }
    }
    return CRC_acc; // Return the final remainder (CRC value)
}
```

Table 24.2 lists example input values and the associated outputs using the 32-bit C8051F80x-83x CRC algorithm (an initial value of 0xFFFFFFFF is used):

Table 24.2. Example 32-bit CRC Outputs

Input	Output
0x63	0xF9462090
0xAA, 0xBB, 0xCC	0x41B207B3
0x00, 0x00, 0xAA, 0xBB, 0xCC	0x78D129BC

26. SMBus

The SMBus I/O interface is a two-wire, bi-directional serial bus. The SMBus is compliant with the System Management Bus Specification, version 1.1, and compatible with the I²C serial bus. Reads and writes to the interface by the system controller are byte oriented with the SMBus interface autonomously controlling the serial transfer of the data. Data can be transferred at up to 1/20th of the system clock as a master or slave (this can be faster than allowed by the SMBus specification, depending on the system clock used). A method of extending the clock-low duration is available to accommodate devices with different speed capabilities on the same bus.

The SMBus interface may operate as a master and/or slave, and may function on a bus with multiple masters. The SMBus provides control of SDA (serial data), SCL (serial clock) generation and synchronization, arbitration logic, and START/STOP control and generation. The SMBus peripheral can be fully driven by software (i.e., software accepts/rejects slave addresses, and generates ACKs), or hardware slave address recognition and automatic ACK generation can be enabled to minimize software overhead. A block diagram of the SMBus peripheral and the associated SFRs is shown in Figure 26.1.

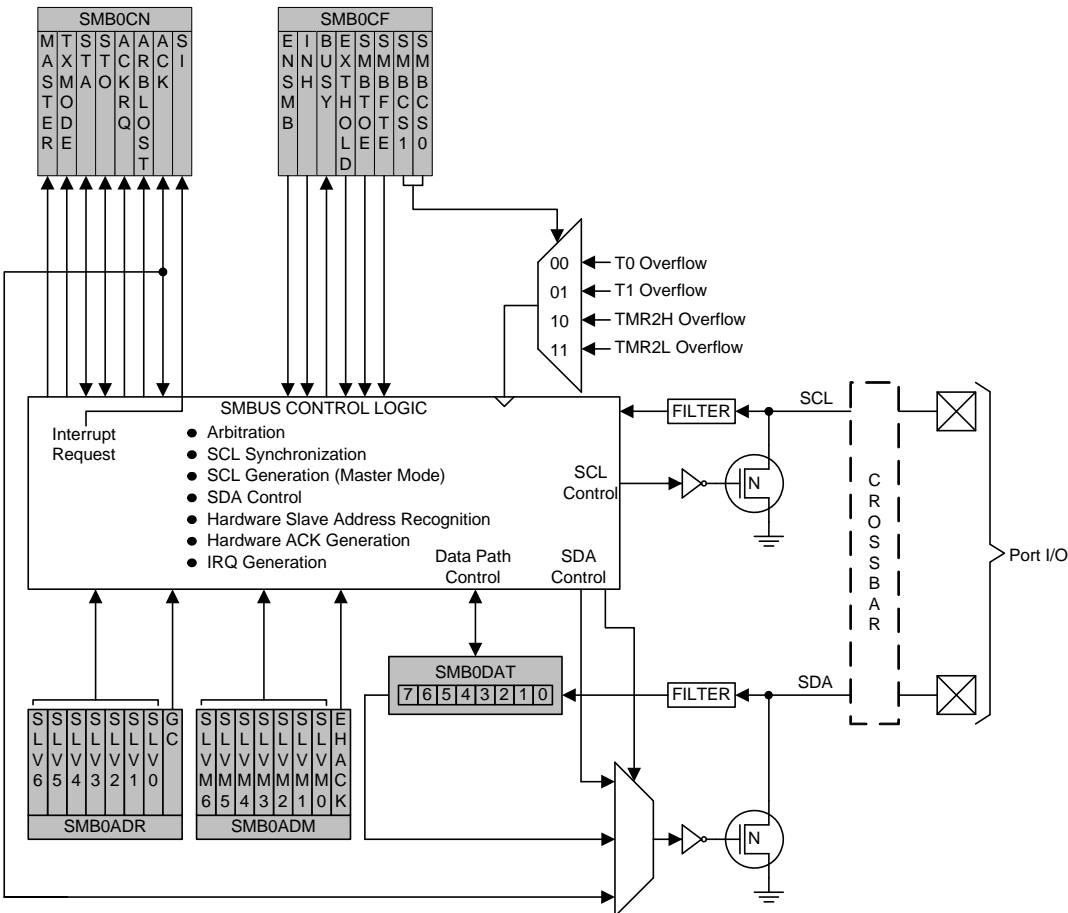


Figure 26.1. SMBus Block Diagram

C8051F80x-83x

26.1. Supporting Documents

It is assumed the reader is familiar with or has access to the following supporting documents:

1. The I²C-Bus and How to Use It (including specifications), Philips Semiconductor.
2. The I²C-Bus Specification—Version 2.0, Philips Semiconductor.
3. System Management Bus Specification—Version 1.1, SBS Implementers Forum.

26.2. SMBus Configuration

Figure 26.2 shows a typical SMBus configuration. The SMBus specification allows any recessive voltage between 3.0 V and 5.0 V; different devices on the bus may operate at different voltage levels. The bi-directional SCL (serial clock) and SDA (serial data) lines must be connected to a positive power supply voltage through a pullup resistor or similar circuit. Every device connected to the bus must have an open-drain or open-collector output for both the SCL and SDA lines, so that both are pulled high (recessive state) when the bus is free. The maximum number of devices on the bus is limited only by the requirement that the rise and fall times on the bus not exceed 300 ns and 1000 ns, respectively.

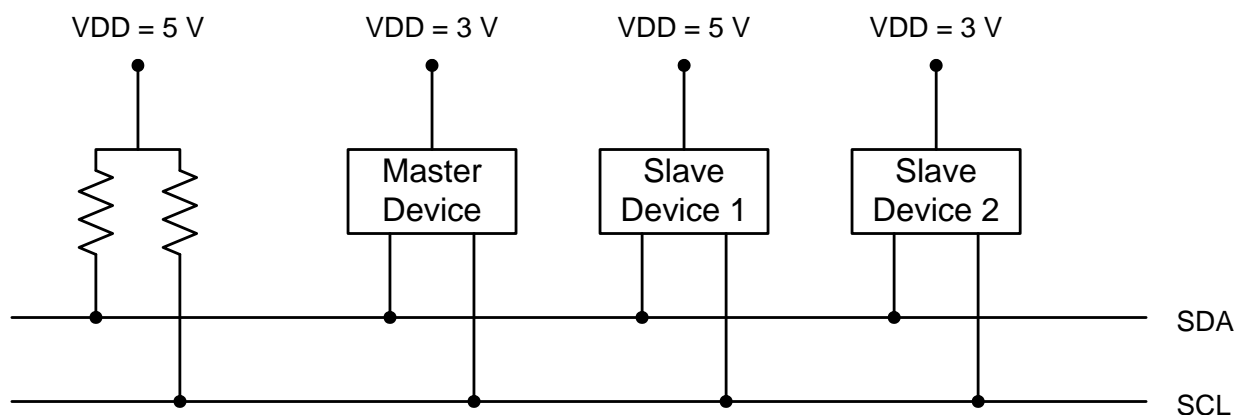


Figure 26.2. Typical SMBus Configuration

26.3. SMBus Operation

Two types of data transfers are possible: data transfers from a master transmitter to an addressed slave receiver (WRITE), and data transfers from an addressed slave transmitter to a master receiver (READ). The master device initiates both types of data transfers and provides the serial clock pulses on SCL. The SMBus interface may operate as a master or a slave, and multiple master devices on the same bus are supported. If two or more masters attempt to initiate a data transfer simultaneously, an arbitration scheme is employed with a single master always winning the arbitration. Note that it is not necessary to specify one device as the Master in a system; any device who transmits a START and a slave address becomes the master for the duration of that transfer.

A typical SMBus transaction consists of a START condition followed by an address byte (Bits7–1: 7-bit slave address; Bit0: R/W direction bit), one or more bytes of data, and a STOP condition. Bytes that are received (by a master or slave) are acknowledged (ACK) with a low SDA during a high SCL (see Figure 26.3). If the receiving device does not ACK, the transmitting device will read a NACK (not acknowledge), which is a high SDA during a high SCL.

The direction bit (R/W) occupies the least-significant bit position of the address byte. The direction bit is set to logic 1 to indicate a "READ" operation and cleared to logic 0 to indicate a "WRITE" operation.

All transactions are initiated by a master, with one or more addressed slave devices as the target. The master generates the START condition and then transmits the slave address and direction bit. If the transaction is a WRITE operation from the master to the slave, the master transmits the data a byte at a time waiting for an ACK from the slave at the end of each byte. For READ operations, the slave transmits the data waiting for an ACK from the master at the end of each byte. At the end of the data transfer, the master generates a STOP condition to terminate the transaction and free the bus. Figure 26.3 illustrates a typical SMBus transaction.

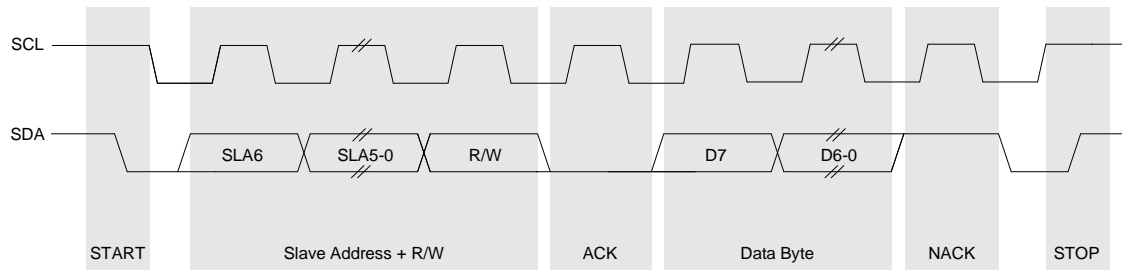


Figure 26.3. SMBus Transaction

26.3.1. Transmitter Vs. Receiver

On the SMBus communications interface, a device is the “transmitter” when it is sending an address or data byte to another device on the bus. A device is a “receiver” when an address or data byte is being sent to it from another device on the bus. The transmitter controls the SDA line during the address or data byte. After each byte of address or data information is sent by the transmitter, the receiver sends an ACK or NACK bit during the ACK phase of the transfer, during which time the receiver controls the SDA line.

26.3.2. Arbitration

A master may start a transfer only if the bus is free. The bus is free after a STOP condition or after the SCL and SDA lines remain high for a specified time (see Section “26.3.5. SCL High (SMBus Free) Timeout” on page 183). In the event that two or more devices attempt to begin a transfer at the same time, an arbitration scheme is employed to force one master to give up the bus. The master devices continue transmitting until one attempts a HIGH while the other transmits a LOW. Since the bus is open-drain, the bus will be pulled LOW. The master attempting the HIGH will detect a LOW SDA and lose the arbitration. The winning master continues its transmission without interruption; the losing master becomes a slave and receives the rest of the transfer if addressed. This arbitration scheme is non-destructive: one device always wins, and no data is lost.

26.3.3. Clock Low Extension

SMBus provides a clock synchronization mechanism, similar to I2C, which allows devices with different speed capabilities to coexist on the bus. A clock-low extension is used during a transfer in order to allow slower slave devices to communicate with faster masters. The slave may temporarily hold the SCL line LOW to extend the clock low period, effectively decreasing the serial clock frequency.

26.3.4. SCL Low Timeout

If the SCL line is held low by a slave device on the bus, no further communication is possible. Furthermore, the master cannot force the SCL line high to correct the error condition. To solve this problem, the SMBus protocol specifies that devices participating in a transfer must detect any clock cycle held low longer than 25 ms as a “timeout” condition. Devices that have detected the timeout condition must reset the communication no later than 10 ms after detecting the timeout condition.

When the SMBTOE bit in SMB0CF is set, Timer 3 is used to detect SCL low timeouts. Timer 3 is forced to reload when SCL is high, and allowed to count when SCL is low. With Timer 3 enabled and configured to

case, either a 1 or a 0 value are acceptable on the incoming slave address. Additionally, if the GC bit in register SMB0ADR is set to 1, hardware will recognize the General Call Address (0x00). Table 26.4 shows some example parameter settings and the slave addresses that will be recognized by hardware under those conditions.

Table 26.4. Hardware Address Recognition Examples (EHACK = 1)

Hardware Slave Address SLV[6:0]	Slave Address Mask SLVM[6:0]	GC bit	Slave Addresses Recognized by Hardware
0x34	0x7F	0	0x34
0x34	0x7F	1	0x34, 0x00 (General Call)
0x34	0x7E	0	0x34, 0x35
0x34	0x7E	1	0x34, 0x35, 0x00 (General Call)
0x70	0x73	0	0x70, 0x74, 0x78, 0x7C

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Table 26.6. SMBus Status Decoding With Hardware ACK Generation Enabled (EHACK = 1)

Mode	Values Read			Current SMBus State	Typical Response Options	Values to Write			Next Status Vector Expected	
	Status Vector	ACKRQ	ARBLOST			ACK	STA	STO		ACK
Master Transmitter	1110	0	0	X	A master START was generated.	Load slave address + R/W into SMB0DAT.	0	0	X	1100
	1100	0	0	0	A master data or address byte was transmitted; NACK received.	Set STA to restart transfer.	1	0	X	1110
						Abort transfer.	0	1	X	—
		0	0	1	A master data or address byte was transmitted; ACK received.	Load next data byte into SMB0DAT.	0	0	X	1100
						End transfer with STOP.	0	1	X	—
						End transfer with STOP and start another transfer.	1	1	X	—
						Send repeated START.	1	0	X	1110
				Switch to Master Receiver Mode (clear SI without writing new data to SMB0DAT). Set ACK for initial data byte.	0	0	1	1000		
Master Receiver	1000	0	0	1	A master data byte was received; ACK sent.	Set ACK for next data byte; Read SMB0DAT.	0	0	1	1000
						Set NACK to indicate next data byte as the last data byte; Read SMB0DAT.	0	0	0	1000
						Initiate repeated START.	1	0	0	1110
						Switch to Master Transmitter Mode (write to SMB0DAT before clearing SI).	0	0	X	1100
	0	0	0	0	A master data byte was received; NACK sent (last byte).	Read SMB0DAT; send STOP.	0	1	0	—
						Read SMB0DAT; Send STOP followed by START.	1	1	0	1110
						Initiate repeated START.	1	0	0	1110
						Switch to Master Transmitter Mode (write to SMB0DAT before clearing SI).	0	0	X	1100

28. Timers

Each MCU includes three counter/timers: two are 16-bit counter/timers compatible with those found in the standard 8051, and one is a 16-bit auto-reload timer for use with the ADC, SMBus, or for general purpose use. These timers can be used to measure time intervals, count external events and generate periodic interrupt requests. Timer 0 and Timer 1 are nearly identical and have four primary modes of operation. Timer 2 offers 16-bit and split 8-bit timer functionality with auto-reload. Additionally, Timer 2 offers the ability to be clocked from the external oscillator while the device is in Suspend mode, and can be used as a wake-up source. This allows for implementation of a very low-power system, including RTC capability.

Timer 0 and Timer 1 Modes	Timer 2 Modes
13-bit counter/timer	16-bit timer with auto-reload
16-bit counter/timer	
8-bit counter/timer with auto-reload	Two 8-bit timers with auto-reload
Two 8-bit counter/timers (Timer 0 only)	

Timers 0 and 1 may be clocked by one of five sources, determined by the Timer Mode Select bits (T1M–T0M) and the Clock Scale bits (SCA1–SCA0). The Clock Scale bits define a pre-scaled clock from which Timer 0 and/or Timer 1 may be clocked (See SFR Definition 28.1 for pre-scaled clock selection).

Timer 0/1 may then be configured to use this pre-scaled clock signal or the system clock. Timer 2 may be clocked by the system clock, the system clock divided by 12, or the external oscillator clock source divided by 8.

Timer 0 and Timer 1 may also be operated as counters. When functioning as a counter, a counter/timer register is incremented on each high-to-low transition at the selected input pin (T0 or T1). Events with a frequency of up to one-fourth the system clock frequency can be counted. The input signal need not be periodic, but it should be held at a given level for at least two full system clock cycles to ensure the level is properly sampled.

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SFR Definition 28.3. TMOD: Timer Mode

Bit	7	6	5	4	3	2	1	0
Name	GATE1	C/T1	T1M[1:0]		GATE0	C/T0	T0M[1:0]	
Type	R/W	R/W	R/W		R/W	R/W	R/W	
Reset	0	0	0	0	0	0	0	0

SFR Address = 0x89

Bit	Name	Function
7	GATE1	Timer 1 Gate Control. 0: Timer 1 enabled when TR1 = 1 irrespective of $\overline{INT1}$ logic level. 1: Timer 1 enabled only when TR1 = 1 AND $\overline{INT1}$ is active as defined by bit IN1PL in register IT01CF (see SFR Definition 18.7).
6	C/T1	Counter/Timer 1 Select. 0: Timer: Timer 1 incremented by clock defined by T1M bit in register CKCON. 1: Counter: Timer 1 incremented by high-to-low transitions on external pin (T1).
5:4	T1M[1:0]	Timer 1 Mode Select. These bits select the Timer 1 operation mode. 00: Mode 0, 13-bit Counter/Timer 01: Mode 1, 16-bit Counter/Timer 10: Mode 2, 8-bit Counter/Timer with Auto-Reload 11: Mode 3, Timer 1 Inactive
3	GATE0	Timer 0 Gate Control. 0: Timer 0 enabled when TR0 = 1 irrespective of $\overline{INT0}$ logic level. 1: Timer 0 enabled only when TR0 = 1 AND $\overline{INT0}$ is active as defined by bit IN0PL in register IT01CF (see SFR Definition 18.7).
2	C/T0	Counter/Timer 0 Select. 0: Timer: Timer 0 incremented by clock defined by T0M bit in register CKCON. 1: Counter: Timer 0 incremented by high-to-low transitions on external pin (T0).
1:0	T0M[1:0]	Timer 0 Mode Select. These bits select the Timer 0 operation mode. 00: Mode 0, 13-bit Counter/Timer 01: Mode 1, 16-bit Counter/Timer 10: Mode 2, 8-bit Counter/Timer with Auto-Reload 11: Mode 3, Two 8-bit Counter/Timers

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29.1. PCA Counter/Timer

The 16-bit PCA counter/timer consists of two 8-bit SFRs: PCA0L and PCA0H. PCA0H is the high byte (MSB) of the 16-bit counter/timer and PCA0L is the low byte (LSB). Reading PCA0L automatically latches the value of PCA0H into a “snapshot” register; the following PCA0H read accesses this “snapshot” register. **Reading the PCA0L Register first guarantees an accurate reading of the entire 16-bit PCA0 counter.** Reading PCA0H or PCA0L does not disturb the counter operation. The CPS2–CPS0 bits in the PCA0MD register select the timebase for the counter/timer as shown in Table 29.1.

When the counter/timer overflows from 0xFFFF to 0x0000, the Counter Overflow Flag (CF) in PCA0MD is set to logic 1 and an interrupt request is generated if CF interrupts are enabled. Setting the ECF bit in PCA0MD to logic 1 enables the CF flag to generate an interrupt request. The CF bit is not automatically cleared by hardware when the CPU vectors to the interrupt service routine, and must be cleared by software. Clearing the CIDL bit in the PCA0MD register allows the PCA to continue normal operation while the CPU is in Idle mode.

Table 29.1. PCA Timebase Input Options

CPS2	CPS1	CPS0	Timebase
0	0	0	System clock divided by 12
0	0	1	System clock divided by 4
0	1	0	Timer 0 overflow
0	1	1	High-to-low transitions on ECI (max rate = system clock divided by 4)
1	0	0	System clock
1	0	1	External oscillator source divided by 8 (Note)
1	1	x	Reserved

Note: External oscillator source divided by 8 is synchronized with the system clock.

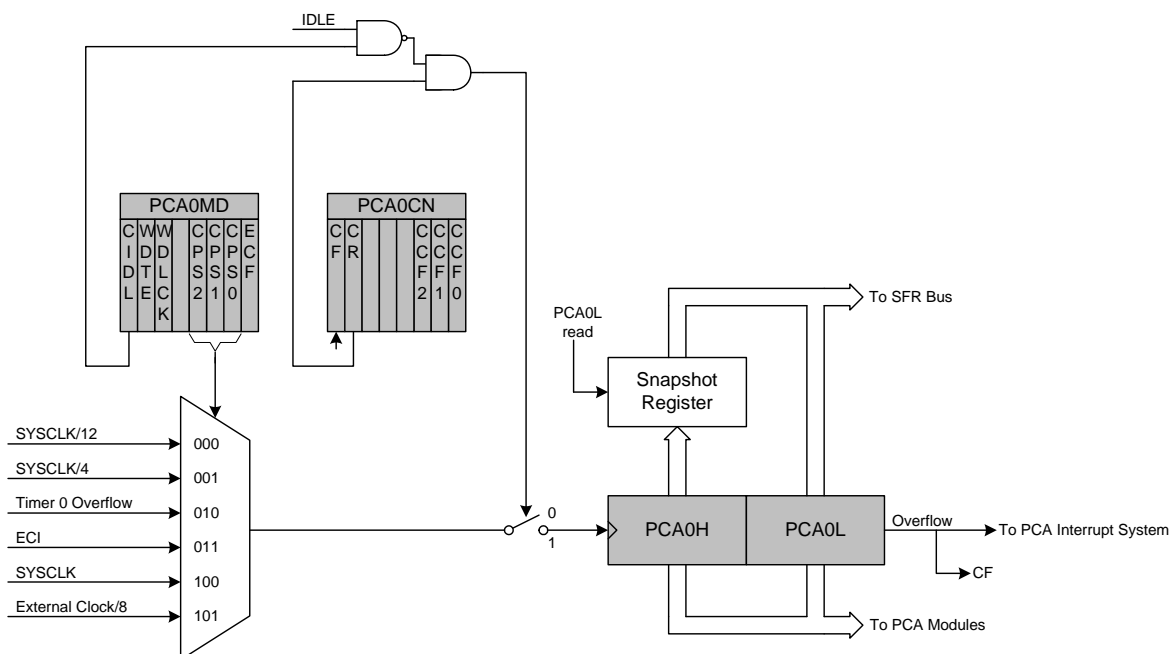


Figure 29.2. PCA Counter/Timer Block Diagram

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NOTES: