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Details

Product Status	Not For New Designs
Core Processor	8051
Core Size	8-Bit
Speed	25MHz
Connectivity	SMBus (2-Wire/I ² C), SPI, UART/USART
Peripherals	Cap Sense, POR, PWM, WDT
Number of I/O	13
Program Memory Size	16KB (16K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	512 x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.6V
Data Converters	-
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	16-SOIC (0.154", 3.90mm Width)
Supplier Device Package	16-SOIC
Purchase URL	https://www.e-xfl.com/product-detail/silicon-labs/c8051f809-gsr

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List of Registers

SFR	Definition 8.1	. ADC0CF: ADC0 Configuration	. 50
SFR	Definition 8.2	. ADC0H: ADC0 Data Word MSB	. 51
SFR	Definition 8.3	. ADC0L: ADC0 Data Word LSB	. 51
SFR	Definition 8.4	. ADC0CN: ADC0 Control	. 52
SFR	Definition 8.5	. ADC0GTH: ADC0 Greater-Than Data High Byte	. 53
SFR	Definition 8.6	. ADC0GTL: ADC0 Greater-Than Data Low Byte	. 53
SFR	Definition 8.7	. ADC0LTH: ADC0 Less-Than Data High Byte	. 54
SFR	Definition 8.8	. ADC0LTL: ADC0 Less-Than Data Low Byte	. 54
		. ADC0MX: AMUX0 Channel Select	
SFR	Definition 10.	1. REF0CN: Voltage Reference Control	. 62
SFR	Definition 11.	1. REG0CN: Voltage Regulator Control	. 64
		1. CPT0CN: Comparator0 Control	
		2. CPT0MD: Comparator0 Mode Selection	
		3. CPT0MX: Comparator0 MUX Selection	
SFR	Definition 13.	1. CS0CN: Capacitive Sense Control	. 75
		2. CS0CF: Capacitive Sense Configuration	
		3. CS0DH: Capacitive Sense Data High Byte	
		4. CS0DL: Capacitive Sense Data Low Byte	
		5. CS0SS: Capacitive Sense Auto-Scan Start Channel	
		6. CS0SE: Capacitive Sense Auto-Scan End Channel	
		7. CS0THH: Capacitive Sense Comparator Threshold High Byte	
		8. CS0THL: Capacitive Sense Comparator Threshold Low Byte	
		9. CS0MX: Capacitive Sense Mux Channel Select	
		1. DPL: Data Pointer Low Byte	
		2. DPH: Data Pointer High Byte	
		3. SP: Stack Pointer	
		4. ACC: Accumulator	
		5. B: B Register	
		6. PSW: Program Status Word	
		1. HWID: Hardware Identification Byte	
		2. DERIVID: Derivative Identification Byte	
		3. REVID: Hardware Revision Identification Byte	
		1. IE: Interrupt Enable	
		2. IP: Interrupt Priority	
		3. EIE1: Extended Interrupt Enable 1	
		4. EIE2: Extended Interrupt Enable 2	
		5. EIP1: Extended Interrupt Priority 1	
		6. EIP2: Extended Interrupt Priority 2	
		7. IT01CF: INT0/INT1 Configuration	
		1. PSCTL: Program Store R/W Control	
		2. FLKEY: Flash Lock and Key	
		1. PCON: Power Control	
SFR	Definition 21.	1. VDM0CN: VDD Monitor Control	126



SFR Definition 21.2. RSTSRC: Reset Source	. 128
SFR Definition 22.1. CLKSEL: Clock Select	. 130
SFR Definition 22.2. OSCICL: Internal H-F Oscillator Calibration	. 131
SFR Definition 22.3. OSCICN: Internal H-F Oscillator Control	. 132
SFR Definition 22.4. OSCXCN: External Oscillator Control	. 134
SFR Definition 23.1. XBR0: Port I/O Crossbar Register 0	. 148
SFR Definition 23.2. XBR1: Port I/O Crossbar Register 1	. 149
SFR Definition 23.3. P0MASK: Port 0 Mask Register	. 151
SFR Definition 23.4. P0MAT: Port 0 Match Register	. 151
SFR Definition 23.5. P1MASK: Port 1 Mask Register	. 152
SFR Definition 23.6. P1MAT: Port 1 Match Register	
SFR Definition 23.7. P0: Port 0	. 153
SFR Definition 23.8. P0MDIN: Port 0 Input Mode	. 154
SFR Definition 23.9. P0MDOUT: Port 0 Output Mode	. 154
SFR Definition 23.10. P0SKIP: Port 0 Skip	. 155
SFR Definition 23.11. P1: Port 1	
SFR Definition 23.12. P1MDIN: Port 1 Input Mode	. 156
SFR Definition 23.13. P1MDOUT: Port 1 Output Mode	. 156
SFR Definition 23.14. P1SKIP: Port 1 Skip	. 157
SFR Definition 23.15. P2: Port 2	
SFR Definition 23.16. P2MDOUT: Port 2 Output Mode	. 158
SFR Definition 24.1. CRC0CN: CRC0 Control	
SFR Definition 24.2. CRC0IN: CRC Data Input	. 164
SFR Definition 24.3. CRC0DATA: CRC Data Output	. 164
SFR Definition 24.4. CRC0AUTO: CRC Automatic Control	. 165
SFR Definition 24.5. CRC0CNT: CRC Automatic Flash Sector Count	. 165
SFR Definition 24.6. CRC0FLIP: CRC Bit Flip	. 166
SFR Definition 25.1. SPI0CFG: SPI0 Configuration	
SFR Definition 25.2. SPI0CN: SPI0 Control	
SFR Definition 25.3. SPI0CKR: SPI0 Clock Rate	. 176
SFR Definition 25.4. SPI0DAT: SPI0 Data	. 176
SFR Definition 26.1. SMB0CF: SMBus Clock/Configuration	. 186
SFR Definition 26.2. SMB0CN: SMBus Control	
SFR Definition 26.3. SMB0ADR: SMBus Slave Address	. 191
SFR Definition 26.4. SMB0ADM: SMBus Slave Address Mask	. 191
SFR Definition 26.5. SMB0DAT: SMBus Data	. 192
SFR Definition 27.1. SCON0: Serial Port 0 Control	. 206
SFR Definition 27.2. SBUF0: Serial (UART0) Port Data Buffer	. 207
SFR Definition 28.1. CKCON: Clock Control	
SFR Definition 28.2. TCON: Timer Control	. 215
SFR Definition 28.3. TMOD: Timer Mode	. 216
SFR Definition 28.4. TL0: Timer 0 Low Byte	
SFR Definition 28.5. TL1: Timer 1 Low Byte	
SFR Definition 28.6. TH0: Timer 0 High Byte	
SFR Definition 28.7. TH1: Timer 1 High Byte	
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Table 7.6. Flash Electrical Characteristics

Parameter	Conditions	Min	Тур	Max	Units
Flash Size (Note 1)	C8051F80x and C8051F810/1		16384		bytes
	C8051F812/3/4/5/6/7/8/9 and C8051F82x		8192		bytes
	C8051F830/1/2/3/4/5		4096		bytes
Endurance (Erase/Write)		10000			cycles
Erase Cycle Time	25 MHz Clock	15	20	26	ms
Write Cycle Time	25 MHz Clock	15	20	26	μs
Clock Speed during Flash Write/Erase Operations		1	—	—	MHz
Note: Includes Security Lock By	rte.	•	•	•	•

Table 7.7. Internal High-Frequency Oscillator Electrical Characteristics

 V_{DD} = 1.8 to 3.6 V; T_A = -40 to +85 °C unless otherwise specified. Use factory-calibrated settings.

Parameter	Conditions	Min	Тур	Max	Units
Oscillator Frequency	IFCN = 11b	24	24.5	25	MHz
Oscillator Supply Current	25 °C, V _{DD} = 3.0 V, OSCICN.7 = 1, OCSICN.5 = 0		350	650	μA

Table 7.8. Capacitive Sense Electrical Characteristics

 V_{DD} = 1.8 to 3.6 V; T_A = -40 to +85 °C unless otherwise specified.

Parameter	Conditions	Min	Тур	Max	Units
Conversion Time	Single Conversion	26	38	50	μs
Capacitance per Code			1		fF
External Capacitive Load		—	_	45	pF
Quantization Noise ¹	RMS		3		fF
	Peak-to-Peak	—	20	—	fF
Supply Current	CS module bias current, 25 °C		40	60	μA
	CS module alone, maximum code output, 25 °C	—	75	105	μA
	Wake-on-CS Threshold ² , 25 °C	—	150	165	μA

Notes:

1. RMS Noise is equivalent to one standard deviation. Peak-to-peak noise encompasses ±3.3 standard deviations.

2. Includes only current from regulator, CS module, and MCU in suspend mode.



Mnemonic	Description	Bytes	Clock Cycles	
XRL direct, #data	Exclusive-OR immediate to direct byte	3	3	
CLR A	Clear A	1	1	
CPL A	Complement A	1	1	
RL A	Rotate A left	1	1	
RLC A	Rotate A left through Carry	1	1	
RR A	Rotate A right	1	1	
RRC A	Rotate A right through Carry	1	1	
SWAP A	Swap nibbles of A	1	1	
Data Transfer	-	L.	1	
MOV A, Rn	Move Register to A	1	1	
MOV A, direct	Move direct byte to A	2	2	
MOV A, @Ri	Move indirect RAM to A	1	2	
MOV A, #data	Move immediate to A	2	2	
MOV Rn, A	Move A to Register	1	1	
MOV Rn, direct	Move direct byte to Register	2	2	
MOV Rn, #data	Move immediate to Register	2	2	
MOV direct, A	Move A to direct byte	2	2	
MOV direct, Rn	Move Register to direct byte	2	2	
MOV direct, direct	Move direct byte to direct byte	3	3	
MOV direct, @Ri	Move indirect RAM to direct byte	2	2	
MOV direct, #data	Move immediate to direct byte	3	3	
MOV @Ri, A	Move A to indirect RAM	1	2	
MOV @Ri, direct	Move direct byte to indirect RAM	2	2	
MOV @Ri, #data	Move immediate to indirect RAM	2	2	
MOV DPTR, #data16	Load DPTR with 16-bit constant	3	3	
MOVC A, @A+DPTR	Move code byte relative DPTR to A	1	3	
MOVC A, @A+PC	Move code byte relative PC to A	1	3	
MOVX A, @Ri	Move external data (8-bit address) to A	1	3	
MOVX @Ri, A	Move A to external data (8-bit address)	1	3	
MOVX A, @DPTR	Move external data (16-bit address) to A	1	3	
MOVX @DPTR, A	Move A to external data (16-bit address)	1	3	
PUSH direct	Push direct byte onto stack	2	2	
POP direct	Pop direct byte from stack	2	2	
XCH A, Rn	Exchange Register with A	1	1	
XCH A, direct	Exchange direct byte with A	2	2	
XCH A, @Ri	Exchange indirect RAM with A	1	2	
XCHD A, @Ri	Exchange low nibble of indirect RAM with A	1	2	
Boolean Manipulation	-	L.	1	
CLR C	Clear Carry	1	1	
CLR bit	Clear direct bit	2	2	
SETB C	Set Carry	1	1	
SETB bit	Set direct bit	2	2	
CPL C	Complement Carry	1	1	
CPL bit	Complement direct bit	2	2	

 Table 14.1. CIP-51 Instruction Set Summary (Continued)



SFR Definition 14.5. B: B Register

-		-											
Bit	7	6	5	4	3	2	1	0					
Nam	e	B[7:0]											
Туре	•	R/W											
Rese	et 0	0	0	0	0	0	0	0					
SFR A	SFR Address = 0xF0; Bit-Addressable												
Bit	Name				Function								
7:0	B[7:0]	B Register.											

This register serves as a second accumulator for certain arithmetic operations.
I This redister serves as a second accumulator for certain animmetic operations



whether the CPU accesses the upper 128 bytes of data memory space or the SFRs. Instructions that use direct addressing will access the SFR space. Instructions using indirect addressing above 0x7F access the upper 128 bytes of data memory. Figure 15.1 illustrates the data memory organization of the C8051F80x-83x.

15.2.1.1. General Purpose Registers

The lower 32 bytes of data memory, locations 0x00 through 0x1F, may be addressed as four banks of general-purpose registers. Each bank consists of eight byte-wide registers designated R0 through R7. Only one of these banks may be enabled at a time. Two bits in the program status word, RS0 (PSW.3) and RS1 (PSW.4), select the active register bank (see description of the PSW in SFR Definition 14.6). This allows fast context switching when entering subroutines and interrupt service routines. Indirect addressing modes use registers R0 and R1 as index registers.

15.2.1.2. Bit Addressable Locations

In addition to direct access to data memory organized as bytes, the sixteen data memory locations at 0x20 through 0x2F are also accessible as 128 individually addressable bits. Each bit has a bit address from 0x00 to 0x7F. Bit 0 of the byte at 0x20 has bit address 0x00 while bit7 of the byte at 0x20 has bit address 0x07. Bit 7 of the byte at 0x2F has bit address 0x7F. A bit access is distinguished from a full byte access by the type of instruction used (bit source or destination operands as opposed to a byte source or destination).

The MCS-51[™] assembly language allows an alternate notation for bit addressing of the form XX.B where XX is the byte address and B is the bit position within the byte. For example, the instruction:

MOV C, 22.3h

moves the Boolean value at 0x13 (bit 3 of the byte at location 0x22) into the Carry flag.

15.2.1.3. Stack

A programmer's stack can be located anywhere in the 256-byte data memory. The stack area is designated using the Stack Pointer (SP) SFR. The SP will point to the last location used. The next value pushed on the stack is placed at SP+1 and then SP is incremented. A reset initializes the stack pointer to location 0x07. Therefore, the first value pushed on the stack is placed at location 0x08, which is also the first register (R0) of register bank 1. Thus, if more than one register bank is to be used, the SP should be initialized to a location in the data memory not being used for data storage. The stack depth can extend up to 256 bytes.



SFR Definition 18.1. IE: Interrupt Enable

Bit	7	6	5	4	3	2	1	0
Name	EA	ESPI0	ET2	ES0	ET1	EX1	ET0	EX0
Туре	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xA8; Bit-Addressable

Bit	Name	Function
7	EA	 Enable All Interrupts. Globally enables/disables all interrupts. It overrides individual interrupt mask settings. 0: Disable all interrupt sources. 1: Enable each interrupt according to its individual mask setting.
6	ESPI0	Enable Serial Peripheral Interface (SPI0) Interrupt. This bit sets the masking of the SPI0 interrupts. 0: Disable all SPI0 interrupts. 1: Enable interrupt requests generated by SPI0.
5	ET2	 Enable Timer 2 Interrupt. This bit sets the masking of the Timer 2 interrupt. 0: Disable Timer 2 interrupt. 1: Enable interrupt requests generated by the TF2L or TF2H flags.
4	ES0	Enable UART0 Interrupt. This bit sets the masking of the UART0 interrupt. 0: Disable UART0 interrupt. 1: Enable UART0 interrupt.
3	ET1	Enable Timer 1 Interrupt. This bit sets the masking of the Timer 1 interrupt. 0: Disable all Timer 1 interrupt. 1: Enable interrupt requests generated by the TF1 flag.
2	EX1	Enable External Interrupt 1. This bit sets the masking of External Interrupt 1. 0: Disable external interrupt 1. 1: Enable interrupt requests generated by the INT1 input.
1	ET0	Enable Timer 0 Interrupt. This bit sets the masking of the Timer 0 interrupt. 0: Disable all Timer 0 interrupt. 1: Enable interrupt requests generated by the TF0 flag.
0	EX0	Enable External Interrupt 0. This bit sets the masking of External Interrupt 0. 0: Disable external interrupt 0. 1: Enable interrupt requests generated by the INTO input.



21.1. Power-On Reset

During power-up, the device is held in a reset state and the \overline{RST} pin is driven low until V_{DD} settles above V_{RST}. A delay occurs before the device is released from reset; the delay decreases as the V_{DD} ramp time increases (V_{DD} ramp time is defined as how fast V_{DD} ramps from 0 V to V_{RST}). Figure 21.2. plots the power-on and V_{DD} monitor reset timing. The maximum V_{DD} ramp time is 1 ms; slower ramp times may cause the device to be released from reset before V_{DD} reaches the V_{RST} level. For ramp times less than 1 ms, the power-on reset delay (T_{PORDelay}) is typically less than 10 ms.

On exit from a power-on reset, the PORSF flag (RSTSRC.1) is set by hardware to logic 1. When PORSF is set, all of the other reset flags in the RSTSRC Register are indeterminate (PORSF is cleared by all other resets). Since all resets cause program execution to begin at the same location (0x0000) software can read the PORSF flag to determine if a power-up was the cause of reset. The content of internal data memory should be assumed to be undefined after a power-on reset. The V_{DD} monitor is enabled and selected as a reset source following a power-on reset.

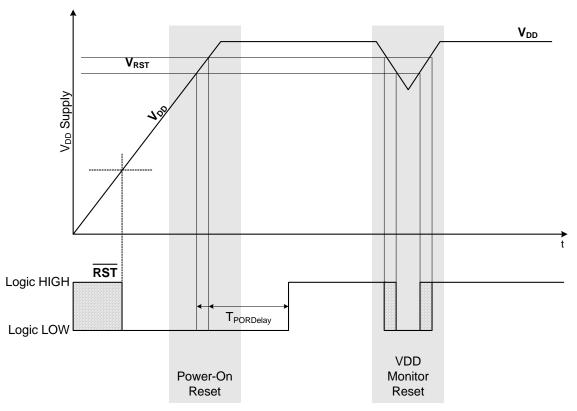


Figure 21.2. Power-On and V_{DD} Monitor Reset Timing



Special Function Signals (Control Signals) (Control SCK) (Control SCK)	P0.0 Skipped	AGND AGND	P0.2 Skipped XTAL1 N	P0.3 Skipped		5	CNVSTR 9	7			2	3	4 ¹	5 ¹	6 ¹	7 ¹	Signal Unavailable to Crossbar	
Function SignalsTX0RX0RX0SCKMISOMOSINSS2SDASCLCP0ACP0ASYSCLKCEX0CEX1CEX2ECIT1Pin Skip SettingsIn this example RX0 signals, th signals are ass P0.3 are config	Skipped	AGND		Skipped			CNVSTR										Crossbar	
RX0 SCK MISO MOSI NSS ² SDA SCL CP0 CP0A SYSCLK CEX0 CEX1 CEX1 CEX2 ECI CEX2 ECI T0 T1 Pin Skip Settings In this example RX0 signals, th signals are ass P0.3 are config	P0.0 Skipped		P0.2 Skipped	P0.3 Skipped													Crossbar	
SCK MISO MOSI NSS ² SDA SCL CP0 CP0A SYSCLK CEX0 CEX1 CEX2 ECI CEX2 ECI T0 T1 Pin Skip Settings In this example RX0 signals, th signals are ass P0.3 are config	P0.0 Skipped		P0.2 Skipped	P0.3 Skipped													Crossbar	
MISO MOSI NSS ² SDA SCL CP00 CP0A SYSCLK CEX0 CEX1 CEX1 CEX2 ECI CEX1 CEX2 ECI T0 T1 Pin Skip Settings In this example RX0 signals, th signals are ass P0.3 are config	P0.0 Skipped		P0.2 Skipped	P0.3 Skipped													Crossbar	
MOSI NSS ² SDA SCL CP0 CP0A SYSCLK CEX0 CEX1 CEX2 ECI CEX2 ECI T0 T1 Pin Skip Settings In this example RX0 signals, th signals are ass P0.3 are config	P0.0 Skipped		P0.2 Skipped	P0.3 Skipped													Crossbar	
NSS ² SDA SCL CP00 CP0A SYSCLK CEX0 CEX1 CEX2 ECI CEX2 ECI T0 T1 Pin Skip Settings In this example RX0 signals, th signals are ass P0.3 are config	P0.0 Skipped		P0.2 Skipped	P0.3 Skipped													Crossl	
SDA SCL CP00 SYSCLK CEX0 CEX1 CEX1 CEX2 ECI ECI T0 T1 Pin Skip Settings In this example RX0 signals, th signals are ass P0.3 are config	P0.0 Skipped		P0.2 Skipped	P0.3 Skipped													ō	
SCL CP0A SYSCLK CEX0 CEX1 CEX2 ECI T0 T1 Pin Skip Settings In this example RX0 signals, th signals are ass P0.3 are config	P0.0 Skipped		P0.2 Skipped	P0.3 Skipped														
CP0A SYSCLK CEX0 CEX1 CEX2 ECI T0 T1 Pin Skip Settings In this example RX0 signals, th signals are ass P0.3 are config	P0.0 Skippe		P0.2 Skippe	P0.3 Skippe													e to	
CP0A SYSCLK CEX0 CEX1 CEX2 ECI T0 T1 Pin Skip Settings In this example RX0 signals, th signals are ass P0.3 are config	P0.0 Ski		P0.2 Ski	P0.3 Ski													able	
CEX0 CEX1 CEX2 ECI T0 T1 Pin Skip Settings In this example RX0 signals, th signals are ass P0.3 are config	P0.0		P0.2	P0.3													vail	
CEX0 CEX1 CEX2 ECI T0 T1 Pin Skip Settings In this example RX0 signals, th signals are ass P0.3 are config	ط 		₫.	۹.													Jna	
CEX1 CEX2 ECI T0 T1 Pin Skip Settings In this example RX0 signals, th signals are ass P0.3 are config																	al	
CEX2 ECI T0 T1 Pin Skip Settings In this example RX0 signals, th signals are ass P0.3 are config																	sign	
ECI T0 T1 Pin Skip Settings In this example RX0 signals, th signals are ass P0.3 are config																	0	
T0 T1 Pin Skip Settings In this example RX0 signals, th signals are ass P0.3 are config	-																	
T1 Pin Skip Settings In this example RX0 signals, th signals are ass P0.3 are config																		
Pin Skip Settings In this example RX0 signals, th signals are ass P0.3 are config																		
Settings In this example RX0 signals, th signals are ass P0.3 are config																		
In this example RX0 signals, th signals are ass P0.3 are config	1	0	1	1	0	0	0	0	0	0	0	0	0	0	0	0		
RX0 signals, th signals are ass P0.3 are config				P0S								P1S						
In this example, the crossbar is configured to assign the UART TX0 and RX0 signals, the SPI signals, and the PCA signals. Note that the SPI signals are assigned as multiple signals. Additionally, pins P0.0, P0.2, and P0.3 are configured to be skipped using the POSKIP register. These boxes represent the port pins which are used by the peripherals in this configuration. 1 st TX0 is assigned to P0.4 2 nd RX0 is assigned to P0.5 3 rd SCK, MISO, MOSI, and NSS are assigned to P0.1, P0.6, P0.7, and P1.0, respectively.																		
4 th CEX0, CEX respectively. All unassigned GPIO or for oth Notes: 1. P1.4-P1.7 ar	l pin her r	ıs, i nor	nclu n-cre	udin ossł	g th bar i	iose func	ski tion	ppe ıs.	d by	/ XE	3R0					s		

Figure 23.6. Priority Crossbar Decoder Example 2—Skipping Pins



SFR Definition 23.2. XBR1: Port I/O Crossbar Register 1

Bit	7	6	5	4	3	2	1	0
Name	WEAKPUD	XBARE	T1E	T0E	ECIE		PCA0ME[1:0]	
Туре	R/W	R/W	R/W	R/W	R/W	R	R/W	R/W
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xE2

Bit	Name	Function
7	WEAKPUD	Port I/O Weak Pullup Disable.
		0: Weak Pullups enabled (except for Ports whose I/O are configured for analog
		mode).
		1: Weak Pullups disabled.
6	XBARE	Crossbar Enable.
		0: Crossbar disabled.
		1: Crossbar enabled.
5	T1E	T1 Enable.
		0: T1 unavailable at Port pin.
		1: T1 routed to Port pin.
4	T0E	T0 Enable.
		0: T0 unavailable at Port pin.
		1: T0 routed to Port pin.
3	ECIE	PCA0 External Counter Input Enable.
		0: ECI unavailable at Port pin.
		1: ECI routed to Port pin.
2	Unused	Read = 0b; Write = Don't Care.
1:0	PCA0ME[1:0]	PCA Module I/O Enable Bits.
		00: All PCA I/O unavailable at Port pins.
		01: CEX0 routed to Port pin.
		10: CEX0, CEX1 routed to Port pins.
		11: CEX0, CEX1, CEX2 routed to Port pins.



24.1. 16-bit CRC Algorithm

The C8051F80x-83x CRC unit calculates the 16-bit CRC MSB-first, using a poly of 0x1021. The following describes the 16-bit CRC algorithm performed by the hardware:

- 1. XOR the most-significant byte of the current CRC result with the input byte. If this is the first iteration of the CRC unit, the current CRC result will be the set initial value (0x0000 or 0xFFFF).
- 2. If the MSB of the CRC result is set, left-shift the CRC result, and then XOR the CRC result with the polynomial (0x1021).
- 3. If the MSB of the CRC result is not set, left-shift the CRC result.
- 4. Repeat at Step 2 for the number of input bits (8).

For example, the 16-bit C8051F80x-83x CRC algorithm can be described by the following code:

```
unsigned short UpdateCRC (unsigned short CRC_acc, unsigned char CRC_input) {
   unsigned char i;
                                         // loop counter
   #define POLY 0x1021
   // Create the CRC "dividend" for polynomial arithmetic (binary arithmetic
   // with no carries)
   CRC_acc = CRC_acc ^ (CRC_input << 8);</pre>
   // "Divide" the poly into the dividend using CRC XOR subtraction
   // CRC_acc holds the "remainder" of each divide
   // Only complete this division for 8 bits since input is 1 byte
   for (i = 0; i < 8; i++)
   {
      // Check if the MSB is set (if MSB is 1, then the POLY can "divide"
      // into the "dividend")
      if ((CRC_acc & 0x8000) == 0x8000)
       {
          // if so, shift the CRC value, and XOR "subtract" the poly
          CRC_acc = CRC_acc << 1;</pre>
          CRC_acc ^= POLY;
       }
      else
       {
          // if not, just shift the CRC value
          CRC_acc = CRC_acc << 1;</pre>
       }
   }
   return CRC_acc; // Return the final remainder (CRC value)
}
```

Table 24.1 lists example input values and the associated outputs using the 16-bit C8051F80x-83x CRC algorithm (an initial value of 0xFFFF is used):

Input	Output
0x63	0xBD35
0xAA, 0xBB, 0xCC	0x6CF6
0x00, 0x00, 0xAA, 0xBB, 0xCC	0xB166

Table 24.1. Example 16-bit CRC Outputs



SFR Definition 24.2. CRC0IN: CRC Data Input

Bit	7	6	5	4	3	2	1	0
Name	CRC0IN[7:0]							
Туре	R/W							
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xDD

Bit	Name	Function
7:0	CRC0IN[7:0]	CRC0 Data Input.
		Each write to CRC0IN results in the written data being computed into the existing CRC result according to the CRC algorithm described in Section 24.1

SFR Definition 24.3. CRC0DATA: CRC Data Output

Bit	7	6	5	4	3	2	1	0	
Name	CRC0DAT[7:0]								
Туре		R/W							
Reset	0	0	0	0	0	0	0	0	

SFR Address = 0xDE

Bit	Name	Function
7:0	CRC0DAT[7:0]	CRC0 Data Output.
		Each read or write performed on CRC0DAT targets the CRC result bits pointed to by the CRC0 Result Pointer (CRC0PNT bits in CRC0CN).



SFR Definition 25.3. SPI0CKR: SPI0 Clock Rate

Bit	7	6	5	4	3	2	1	0
Name	SCR[7:0]							
Туре	R/W							
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xA2

Bit	Name	Function
7:0	SCR[7:0]	SPI0 Clock Rate.
		These bits determine the frequency of the SCK output when the SPI0 module is configured for master mode operation. The SCK clock frequency is a divided version of the system clock, and is given in the following equation, where <i>SYSCLK</i> is the system clock frequency and <i>SPI0CKR</i> is the 8-bit value held in the SPI0CKR register.
		$f_{SCK} = \frac{SYSCLK}{2 \times (SPIOCKR[7:0] + 1)}$
		for 0 <= SPI0CKR <= 255
		Example: If SYSCLK = 2 MHz and SPI0CKR = 0x04,
		$f_{SCK} = \frac{2000000}{2 \times (4+1)}$ $f_{SCK} = 200kHz$
		$f_{SCK} = 200kHz$

SFR Definition 25.4. SPI0DAT: SPI0 Data

Bit	7	6	5	4	3	2	1	0	
Name	SPI0DAT[7:0]								
Туре		R/W							
Reset	0	0	0	0	0	0	0	0	

SFR Address = 0xA3

Bit	Name	Function
7:0	SPI0DAT[7:0]	SPI0 Transmit and Receive Data.
		The SPI0DAT register is used to transmit and receive SPI0 data. Writing data to SPI0DAT places the data into the transmit buffer and initiates a transfer when in Master Mode. A read of SPI0DAT returns the contents of the receive buffer.



Parameter	Description	Min	Max	Units
Master Mode	Timing (See Figure 25.8 and Figure 25.9)			I
Т _{МСКН}	SCK High Time	1 x T _{SYSCLK}	—	ns
T _{MCKL}	SCK Low Time	1 x T _{SYSCLK}	—	ns
T _{MIS}	MISO Valid to SCK Shift Edge	1 x T _{SYSCLK} + 20		ns
Т _{МІН}	SCK Shift Edge to MISO Change	0		ns
Slave Mode	Timing (See Figure 25.10 and Figure 25.11)	•		
T _{SE}	NSS Falling to First SCK Edge	2 x T _{SYSCLK}	—	ns
T _{SD}	Last SCK Edge to NSS Rising	2 x T _{SYSCLK}		ns
T _{SEZ}	NSS Falling to MISO Valid	—	4 x T _{SYSCLK}	ns
T _{SDZ}	NSS Rising to MISO High-Z	—	4 x T _{SYSCLK}	ns
Т _{СКН}	SCK High Time	5 x T _{SYSCLK}		ns
T _{CKL}	SCK Low Time	5 x T _{SYSCLK}	—	ns
T _{SIS}	MOSI Valid to SCK Sample Edge	2 x T _{SYSCLK}		ns
T _{SIH}	SCK Sample Edge to MOSI Change	2 x T _{SYSCLK}		ns
Т _{SOH}	SCK Shift Edge to MISO Change	—	4 x T _{SYSCLK}	ns
T _{SLH}	Last SCK Edge to MISO Change (CKPHA = 1 ONLY)	6 x T _{SYSCLK}	8 x T _{SYSCLK}	ns
Note: T _{SYSCL}	K is equal to one period of the device system clock (S	YSCLK).		I

Table 25.1. SPI Slave Timing Parameters



26. SMBus

The SMBus I/O interface is a two-wire, bi-directional serial bus. The SMBus is compliant with the System Management Bus Specification, version 1.1, and compatible with the I²C serial bus. Reads and writes to the interface by the system controller are byte oriented with the SMBus interface autonomously controlling the serial transfer of the data. Data can be transferred at up to 1/20th of the system clock as a master or slave (this can be faster than allowed by the SMBus specification, depending on the system clock used). A method of extending the clock-low duration is available to accommodate devices with different speed capabilities on the same bus.

The SMBus interface may operate as a master and/or slave, and may function on a bus with multiple masters. The SMBus provides control of SDA (serial data), SCL (serial clock) generation and synchronization, arbitration logic, and START/STOP control and generation. The SMBus peripheral can be fully driven by software (i.e., software accepts/rejects slave addresses, and generates ACKs), or hardware slave address recognition and automatic ACK generation can be enabled to minimize software overhead. A block diagram of the SMBus peripheral and the associated SFRs is shown in Figure 26.1.

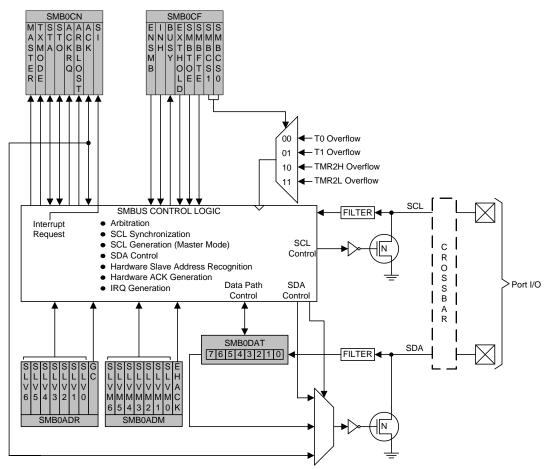


Figure 26.1. SMBus Block Diagram



overflow after 25 ms (and SMBTOE set), the Timer 3 interrupt service routine can be used to reset (disable and re-enable) the SMBus in the event of an SCL low timeout.

26.3.5. SCL High (SMBus Free) Timeout

The SMBus specification stipulates that if the SCL and SDA lines remain high for more that 50 μ s, the bus is designated as free. When the SMBFTE bit in SMB0CF is set, the bus will be considered free if SCL and SDA remain high for more than 10 SMBus clock source periods (as defined by the timer configured for the SMBus clock source). If the SMBus is waiting to generate a Master START, the START will be generated following this timeout. A clock source is required for free timeout detection, even in a slave-only implementation.

26.4. Using the SMBus

The SMBus can operate in both Master and Slave modes. The interface provides timing and shifting control for serial transfers; higher level protocol is determined by user software. The SMBus interface provides the following application-independent features:

- Byte-wise serial data transfers
- Clock signal generation on SCL (Master Mode only) and SDA data synchronization
- Timeout/bus error recognition, as defined by the SMB0CF configuration register
- START/STOP timing, detection, and generation
- Bus arbitration
- Interrupt generation
- Status information
- Optional hardware recognition of slave address and automatic acknowledgement of address/data

SMBus interrupts are generated for each data byte or slave address that is transferred. When hardware acknowledgement is disabled, the point at which the interrupt is generated depends on whether the hardware is acting as a data transmitter or receiver. When a transmitter (i.e., sending address/data, receiving an ACK), this interrupt is generated after the ACK cycle so that software may read the received ACK value; when receiving data (i.e., receiving address/data, sending an ACK), this interrupt is generated before the ACK cycle so that software may define the outgoing ACK value. If hardware acknowledgement is enabled, these interrupts are always generated after the ACK cycle. See Section 26.5 for more details on transmission sequences.

Interrupts are also generated to indicate the beginning of a transfer when a master (START generated), or the end of a transfer when a slave (STOP detected). Software should read the SMB0CN (SMBus Control register) to find the cause of the SMBus interrupt. The SMB0CN register is described in Section 26.4.2; Table 26.5 provides a quick SMB0CN decoding reference.

26.4.1. SMBus Configuration Register

The SMBus Configuration register (SMB0CF) is used to enable the SMBus Master and/or Slave modes, select the SMBus clock source, and select the SMBus timing and timeout options. When the ENSMB bit is set, the SMBus is enabled for all master and slave events. Slave events may be disabled by setting the INH bit. With slave events inhibited, the SMBus interface will still monitor the SCL and SDA pins; however, the interface will NACK all received addresses and will not generate any slave interrupts. When the INH bit is set, all slave events will be inhibited following the next START (interrupts will continue for the duration of the current transfer).

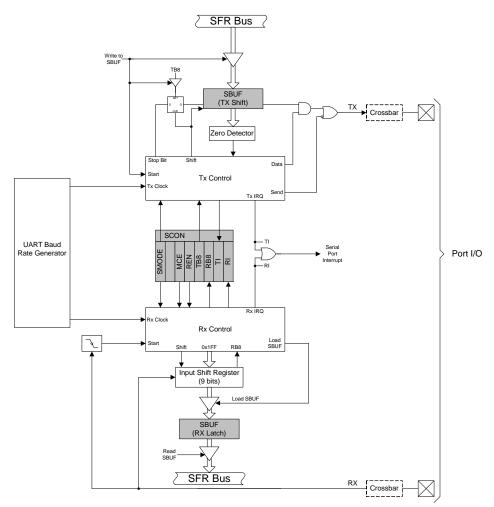


27. UART0

UART0 is an asynchronous, full duplex serial port offering modes 1 and 3 of the standard 8051 UART. Enhanced baud rate support allows a wide range of clock sources to generate standard baud rates (details in Section "27.1. Enhanced Baud Rate Generation" on page 202). Received data buffering allows UART0 to start reception of a second incoming data byte before software has finished reading the previous data byte.

UART0 has two associated SFRs: Serial Control Register 0 (SCON0) and Serial Data Buffer 0 (SBUF0). The single SBUF0 location provides access to both transmit and receive registers. Writes to SBUF0 always access the Transmit register. Reads of SBUF0 always access the buffered Receive register; it is not possible to read data from the Transmit register.

With UART0 interrupts enabled, an interrupt is generated each time a transmit is completed (TI0 is set in SCON0), or a data byte has been received (RI0 is set in SCON0). The UART0 interrupt flags are not cleared by hardware when the CPU vectors to the interrupt service routine. They must be cleared manually by software, allowing software to determine the cause of the UART0 interrupt (transmit complete or receive complete).







SFR Definition 28.8. TMR2CN: Timer 2 Control

Bit	7	6	5	4	3	2	1	0
Name	TF2H	TF2L	TF2LEN	TF2CEN	T2SPLIT	TR2		T2XCLK
Туре	R/W	R/W	R/W	R/W	R/W	R/W	R	R/W
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xC8; Bit-Addressable

Bit	Name	Function
7	TF2H	Timer 2 High Byte Overflow Flag.
		Set by hardware when the Timer 2 high byte overflows from 0xFF to 0x00. In 16 bit mode, this will occur when Timer 2 overflows from 0xFFFF to 0x0000. When the Timer 2 interrupt is enabled, setting this bit causes the CPU to vector to the Timer 2 interrupt service routine. This bit is not automatically cleared by hardware.
6	TF2L	Timer 2 Low Byte Overflow Flag.
		Set by hardware when the Timer 2 low byte overflows from 0xFF to 0x00. TF2L will be set when the low byte overflows regardless of the Timer 2 mode. This bit is not automatically cleared by hardware.
5	TF2LEN	Timer 2 Low Byte Interrupt Enable.
		When set to 1, this bit enables Timer 2 Low Byte interrupts. If Timer 2 interrupts are also enabled, an interrupt will be generated when the low byte of Timer 2 overflows.
4	TF2CEN	Timer 2 Comparator Capture Enable.
		When set to 1, this bit enables Timer 2 Comparator Capture Mode. If TF2CEN is set, on a rising edge of the Comparator0 output the current 16-bit timer value in TMR2H:TMR2L will be copied to TMR2RLH:TMR2RLL. If Timer 2 interrupts are also enabled, an interrupt will be generated on this event.
3	T2SPLIT	Timer 2 Split Mode Enable.
		When this bit is set, Timer 2 operates as two 8-bit timers with auto-reload.0: Timer 2 operates in 16-bit auto-reload mode.1: Timer 2 operates as two 8-bit auto-reload timers.
2	TR2	Timer 2 Run Control.
		Timer 2 is enabled by setting this bit to 1. In 8-bit mode, this bit enables/disables TMR2H only; TMR2L is always enabled in split mode.
1	Unused	Read = 0b; Write = Don't Care.
0	T2XCLK	Timer 2 External Clock Select.
		 This bit selects the external clock source for Timer 2. If Timer 2 is in 8-bit mode, this bit selects the external oscillator clock source for both timer bytes. However, the Timer 2 Clock Select bits (T2MH and T2ML in register CKCON) may still be used to select between the external clock and the system clock for either timer. 0: System clock divided by 12. 1: External clock divided by 8 (synchronized with SYSCLK when not in suspend).



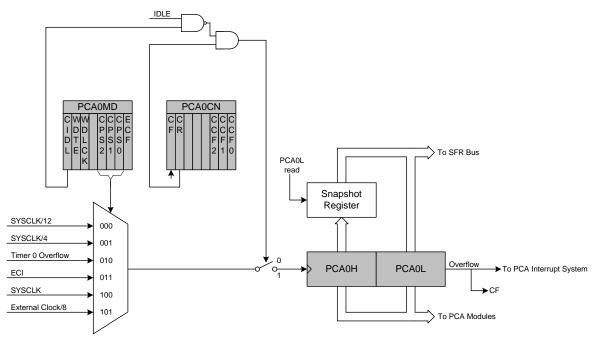
29.1. PCA Counter/Timer

The 16-bit PCA counter/timer consists of two 8-bit SFRs: PCA0L and PCA0H. PCA0H is the high byte (MSB) of the 16-bit counter/timer and PCA0L is the low byte (LSB). Reading PCA0L automatically latches the value of PCA0H into a "snapshot" register; the following PCA0H read accesses this "snapshot" register. **Reading the PCA0L Register first guarantees an accurate reading of the entire 16-bit PCA0 counter.** Reading PCA0H or PCA0L does not disturb the counter operation. The CPS2–CPS0 bits in the PCA0MD register select the timebase for the counter/timer as shown in Table 29.1.

When the counter/timer overflows from 0xFFFF to 0x0000, the Counter Overflow Flag (CF) in PCA0MD is set to logic 1 and an interrupt request is generated if CF interrupts are enabled. Setting the ECF bit in PCA0MD to logic 1 enables the CF flag to generate an interrupt request. The CF bit is not automatically cleared by hardware when the CPU vectors to the interrupt service routine, and must be cleared by software. Clearing the CIDL bit in the PCA0MD register allows the PCA to continue normal operation while the CPU is in Idle mode.

CPS2	CPS1	CPS0	Timebase		
0	0	0	System clock divided by 12		
0	0	1	System clock divided by 4		
0	1	0	Timer 0 overflow		
0	1	1	High-to-low transitions on ECI (max rate = system clock divided by 4)		
1	0	0	System clock		
1	0	1	External oscillator source divided by 8 (Note)		
1	1	Х	Reserved		
Note: Ext	lote: External oscillator source divided by 8 is synchronized with the system clock.				

Table 29.1. PCA Timebase Input Options







29.4. Watchdog Timer Mode

A programmable watchdog timer (WDT) function is available through the PCA Module 2. The WDT is used to generate a reset if the time between writes to the WDT update register (PCA0CPH2) exceed a specified limit. The WDT can be configured and enabled/disabled as needed by software.

With the WDTE bit set in the PCA0MD register, Module 2 operates as a watchdog timer (WDT). The Module 2 high byte is compared to the PCA counter high byte; the Module 2 low byte holds the offset to be used when WDT updates are performed. The Watchdog Timer is enabled on reset. Writes to some PCA registers are restricted while the Watchdog Timer is enabled. The WDT will generate a reset shortly after code begins execution. To avoid this reset, the WDT should be explicitly disabled (and optionally re-configured and re-enabled if it is used in the system).

29.4.1. Watchdog Timer Operation

While the WDT is enabled:

- PCA counter is forced on.
- Writes to PCA0L and PCA0H are not allowed.
- PCA clock source bits (CPS2–CPS0) are frozen.
- PCA Idle control bit (CIDL) is frozen.
- Module 2 is forced into software timer mode.
- Writes to the Module 2 mode register (PCA0CPM2) are disabled.

While the WDT is enabled, writes to the CR bit will not change the PCA counter state; the counter will run until the WDT is disabled. The PCA counter run control bit (CR) will read zero if the WDT is enabled but user software has not enabled the PCA counter. If a match occurs between PCA0CPH2 and PCA0H while the WDT is enabled, a reset will be generated. To prevent a WDT reset, the WDT may be updated with a write of any value to PCA0CPH2. Upon a PCA0CPH2 write, PCA0H plus the offset held in PCA0CPL2 is loaded into PCA0CPH2 (See Figure 29.11).

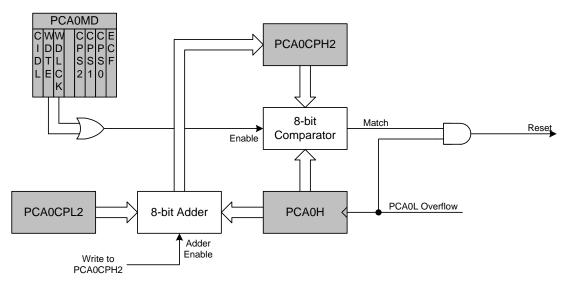


Figure 29.11. PCA Module 2 with Watchdog Timer Enabled

The 8-bit offset held in PCA0CPH2 is compared to the upper byte of the 16-bit PCA counter. This offset value is the number of PCA0L overflows before a reset. Up to 256 PCA clocks may pass before the first PCA0L overflow occurs, depending on the value of the PCA0L when the update is performed. The total off-

