

Welcome to [E-XFL.COM](http://E-XFL.COM)

### Understanding [Embedded - Microprocessors](#)

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

### Applications of [Embedded - Microprocessors](#)

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in



#### Details

Product Status	Obsolete
Core Processor	PowerPC e500v2
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	1.0GHz
Co-Processors/DSP	Communications; QUICC Engine, Security; SEC
RAM Controllers	DDR, DDR2, SDRAM
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100/1000Mbps (2)
SATA	-
USB	-
Voltage - I/O	1.8V, 2.5V, 3.3V
Operating Temperature	0°C ~ 105°C (TA)
Security Features	Cryptography, Random Number Generator
Package / Case	1023-BBGA, FCBGA
Supplier Device Package	1023-FCPBGA (33x33)
Purchase URL	<a href="https://www.e-xfl.com/product-detail/nxp-semiconductors/mpc8568epxaqgg">https://www.e-xfl.com/product-detail/nxp-semiconductors/mpc8568epxaqgg</a>

## 1.1 MPC8568E Key Features

- High-performance, Power Architecture® e500v2 core with 36-bit physical addressing
- 512 Kbytes of level-2 cache
- QUICC Engine (QE)
- Integrated security engine with XOR acceleration
- Two integrated 10/100/1Gb enhanced three-speed Ethernet controllers (eTSECs) with TCP/IP acceleration and classification capabilities
- DDR/DDR2 memory controller
- Table lookup unit (TLU) to access application-defined routing topology and control tables
- 32-bit PCI controller
- A 1x/4x Serial RapidIO® and/or x1/x2/x4 PCI Express interface. If x8 PCI Express is needed, then RapidIO is not available due to the limitation of the pin multiplexing.
- Programmable interrupt controller (PIC)
- Four-channel DMA controller, two I<sup>2</sup>C controllers, DUART, and local bus controller (LBC)

### NOTE

The MPC8568E and MPC8567E are also available without a security engine in a configuration known as the MPC8568 and MPC8567. All specifications other than those relating to security apply to the MPC8568 and MPC8567 exactly as described in this document.

## 1.2 MPC8568E Architecture Overview

### 1.2.1 e500 Core and Memory Unit

The MPC8568E contains a high-performance, 32-bit, Book E–enhanced e500v2 Power Architecture core. In addition to 36-bit physical addressing, this version of the e500 core includes the following:

- Double-precision floating-point APU—Provides an instruction set for double-precision (64-bit) floating-point instructions that use the 64-bit GPRs
- Embedded vector and scalar single-precision floating-point APUs—Provide an instruction set for single-precision (32-bit) floating-point instructions

The MPC8568E also contains 512 Kbytes of L2 cache/SRAM, as follows:

- Eight-way set-associative cache organization with 32-byte cache lines
- Flexible configuration (can be configured as part cache, part SRAM)
- External masters can force data to be allocated into the cache through programmed memory ranges or special transaction types (stashing).
- SRAM features include the following:
  - I/O devices access SRAM regions by marking transactions as snooperable (global).
  - Regions can reside at any aligned location in the memory map.

**Table 1. Supported eTSEC1 and eTSEC2 Configurations<sup>1</sup>**

Mode Option	eTSEC1	eTSEC2
Ethernet standard interfaces	TBI, GMII, or MII	TBI, GMII, or MII
Ethernet reduced interfaces	RTBI, RGMII, or RMII	RTBI, RGMII, or RMII
FIFO and mixed interfaces	8-bit FIFO	TBI, GMII, MII, RTBI, RGMII, RMII, or 8-bit FIFO
	TBI, GMII, MII, RTBI, RGMII, RMII, or 8-bit FIFO	8-bit FIFO
	16-bit FIFO	Not used/not available

<sup>1</sup> Both interfaces must use the same voltage (2.5 or 3.3 V).

- TCP/IP acceleration and QoS features:
  - IP v4 and IP v6 header recognition on receive
  - IP v4 header checksum verification and generation
  - TCP and UDP checksum verification and generation
  - Per-packet configurable acceleration
  - Recognition of VLAN, stacked (queue in queue) VLAN, 802.2, PPPoE session, MPLS stacks, and ESP/AH IP-security headers
  - Supported in all FIFO modes
  - Transmission from up to eight physical queues
  - Reception to up to eight physical queues
- Full- and half-duplex Ethernet support (1000 Mbps supports only full duplex):
  - IEEE 802.3 full-duplex flow control (automatic PAUSE frame generation or software-programmed PAUSE frame generation and recognition)
- IEEE Std 802.1™ virtual local area network (VLAN) tags and priority
- VLAN insertion and deletion
  - Per-frame VLAN control word or default VLAN for each eTSEC
  - Extracted VLAN control word passed to software separately
- Programmable Ethernet preamble insertion and extraction of up to 7 bytes
- MAC address recognition
- Ability to force allocation of header information and buffer descriptors into L2 cache

## 1.2.6 DDR SDRAM Controller

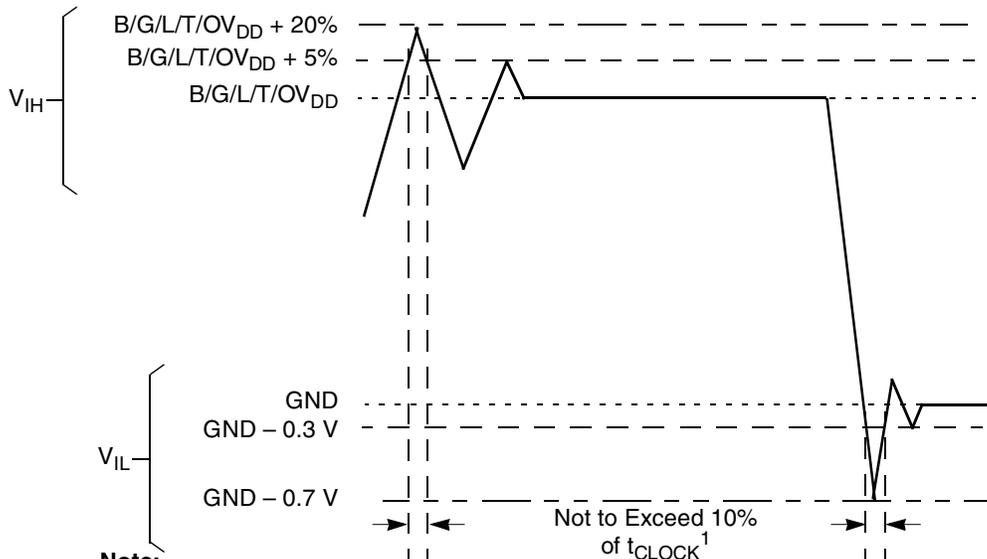
The MPC8568E supports DDR SDRAM and DDR2 SDRAM. The memory interface controls main memory accesses and provides for a maximum of 16 Gbytes of main memory.

The MPC8568E supports a variety of SDRAM configurations. SDRAM banks can be built using DIMMs or directly-attached memory devices. Sixteen multiplexed address signals provide for device densities of 64 Mbits, 128 Mbits, 256 Mbits, 512 Mbits, 1 Gbits, 2 Gbits and 4 Gbits. Four chip select signals support

**Table 3. Recommended Operating Conditions (continued)**

Characteristic		Symbol	Recommended Value	Unit	Notes
Three-speed Ethernet I/O voltage		LV <sub>DD</sub> TV <sub>DD</sub>	3.3 V ± 165 mV 2.5 V ± 125 mV	V	—
PCI, DUART, system control and power management, I <sup>2</sup> C, and JTAG I/O voltage		OV <sub>DD</sub>	3.3 V ± 165 mV	V	—
Local bus I/O voltage		BV <sub>DD</sub>	3.3 V ± 165 mV 2.5 V ± 125 mV	V	—
Input voltage	DDR and DDR2 DRAM signals	MV <sub>IN</sub>	GND to GV <sub>DD</sub>	V	—
	DDR and DDR2 DRAM reference	MV <sub>REF</sub>	GND to GV <sub>DD</sub> /2	V	—
	Three-speed Ethernet signals	LV <sub>IN</sub> TV <sub>IN</sub>	GND to LV <sub>DD</sub> GND to TV <sub>DD</sub>	V	—
	Local bus signals	BV <sub>IN</sub>	GND to BV <sub>DD</sub>	V	—
	PCI, DUART, SYSCLK, system control and power management, I <sup>2</sup> C, and JTAG signals	OV <sub>IN</sub>	GND to OV <sub>DD</sub>	V	—
Junction temperature range		T <sub>j</sub>	0 to 105	°C	—

Figure 2 shows the undershoot and overshoot voltages at the interfaces of the MPC8568E.



**Note:**

- Note that  $t_{\text{CLOCK}}$  refers to the clock period associated with the respective interface. For I<sup>2</sup>C and JTAG,  $t_{\text{CLOCK}}$  references SYSCLK. For DDR,  $t_{\text{CLOCK}}$  references MCLK. For eTSEC,  $t_{\text{CLOCK}}$  references EC\_GTX\_CLK125. For LBIU,  $t_{\text{CLOCK}}$  references LCLK. For PCI,  $t_{\text{CLOCK}}$  references PCI\_CLK or SYSCLK. For SerDes,  $t_{\text{CLOCK}}$  references SD\_REF\_CLK.
- Note that with the PCI overshoot allowed (as specified above), the device does not fully comply with the maximum AC ratings and device protection guideline outlined in the PCI rev. 2.2 standard (section 4.2.2.3)

**Figure 2. Overshoot/Undershoot Voltage for BV<sub>DD</sub>/GV<sub>DD</sub>/LV<sub>DD</sub>/TV<sub>DD</sub>/OV<sub>DD</sub>**

The core voltage must always be provided at nominal 1.1V. (See Table 3 for actual recommended core voltage). Voltage to the processor interface I/Os are provided through separate sets of supply pins and must be provided at the voltages shown in Table 3. The input voltage threshold scales with respect to the associated I/O supply voltage.  $OV_{DD}$  and  $LV_{DD}$  based receivers are simple CMOS I/O circuits and satisfy appropriate LVCMOS type specifications. The DDR SDRAM interface uses a single-ended differential receiver referenced the externally supplied  $MV_{REF}$  signal (nominally set to  $GV_{DD}/2$ ) as is appropriate for the SSTL2 electrical signaling standard.

### 2.1.3 Output Driver Characteristics

Table 4 provides information on the characteristics of the output driver strengths. The values are preliminary estimates.

**Table 4. Output Drive Capability**

Driver Type	Programmable Output Impedance ( $\Omega$ )	Supply Voltage	Notes
Local bus interface utilities signals	25 25	$BV_{DD} = 3.3\text{ V}$ $BV_{DD} = 2.5\text{ V}$	1
	45(default) 45(default)	$BV_{DD} = 3.3\text{ V}$ $BV_{DD} = 2.5\text{ V}$	
PCI signals	25	$OV_{DD} = 3.3\text{ V}$	2
	42 (default)		
DDR signal	20	$GV_{DD} = 2.5\text{ V}$	—
DDR2 signal	16 32 (half strength mode)	$GV_{DD} = 1.8\text{ V}$	—
eTSEC 10/100/1000 signals	42	$L/TV_{DD} = 2.5/3.3\text{ V}$	—
DUART, system control, JTAG	42	$OV_{DD} = 3.3\text{ V}$	—
I2C	150	$OV_{DD} = 3.3\text{ V}$	—

**Notes:**

1. The drive strength of the local bus interface is determined by the configuration of the appropriate bits in PORIMPSR.
2. The drive strength of the PCI interface is determined by the setting of the  $\overline{PCI\_GNT}[1]$  signal at reset.

## 2.2 Power Sequencing

The MPC8568E requires its power rails to be applied in specific sequence in order to ensure proper device operation. These requirements are as follows for power up:

1.  $V_{DD}$ ,  $AV_{DD-n}$ ,  $BV_{DD}$ ,  $SCOREV_{DD}$ ,  $LV_{DD}$ ,  $TV_{DD}$ ,  $XV_{DD}$ ,  $OV_{DD}$
2.  $GV_{DD}$

All supplies must be at their stable values within 50 ms.

Specification Version 1.2 (3/20/1998). The electrical characteristics for MDIO and MDC are specified in Section 8, “Ethernet Interface and MII Management.”

### 8.1.1 Ethernet Interface DC Electrical Characteristics

All GMII, MII, TBI, RGMII, RMII and RTBI drivers and receivers comply with the DC parametric attributes specified in Table 23 and Table 24. The potential applied to the input of a GMII, MII, TBI, RGMII, RMII or RTBI receiver may exceed the potential of the receiver’s power supply (that is, a GMII driver powered from a 3.6-V supply driving  $V_{OH}$  into a GMII receiver powered from a 2.5-V supply). Tolerance for dissimilar GMII driver and receiver supply potentials is implicit in these specifications. The RGMII and RTBI signals are based on a 2.5-V CMOS interface voltage as defined by JEDEC EIA/JESD8-5.

**Table 23. GMII, MII, RMII, and TBI DC Electrical Characteristics**

Parameter	Symbol	Min	Max	Unit	Notes
Supply voltage 3.3 V	$V_{DD}$ $V_{TVDD}$	3.135	3.465	V	1, 2
Output high voltage ( $V_{DD}/V_{TVDD} = \text{Min}$ , $I_{OH} = -4.0 \text{ mA}$ )	$V_{OH}$	2.40	$V_{DD}/V_{TVDD} + 0.3$	V	—
Output low voltage ( $V_{DD}/V_{TVDD} = \text{Min}$ , $I_{OL} = 4.0 \text{ mA}$ )	$V_{OL}$	GND	0.50	V	—
Input high voltage	$V_{IH}$	2.0	$V_{DD}/V_{TVDD} + 0.3$	V	—
Input low voltage	$V_{IL}$	-0.3	0.90	V	—

### 8.2.4.2 TBI Receive AC Timing Specifications

Table 32 provides the TBI receive AC timing specifications.

**Table 32. TBI Receive AC Timing Specifications**

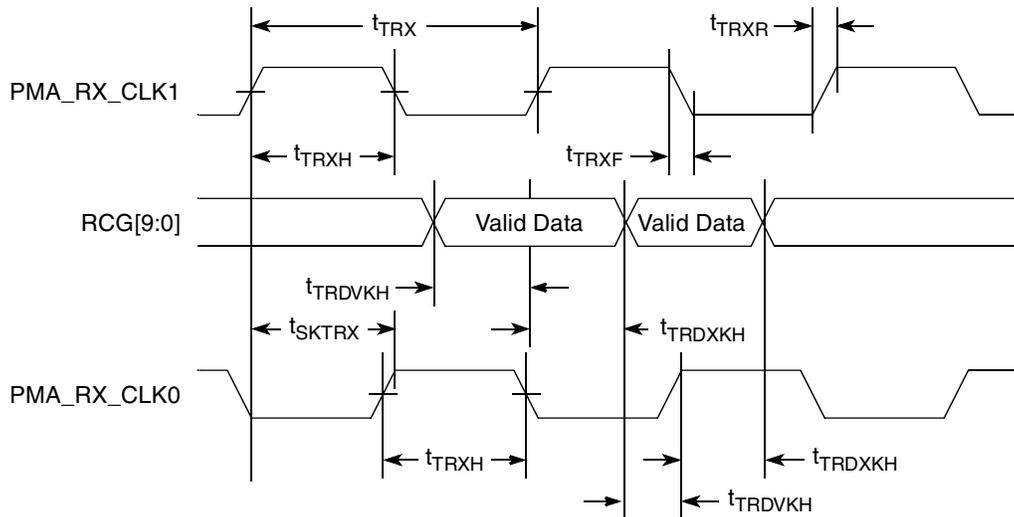
At recommended operating conditions with L/TV<sub>DD</sub> of 3.3 V ± 5%.

Parameter/Condition	Symbol <sup>1</sup>	Min	Typ	Max	Unit
PMA_RX_CLK[0:1] clock period	$t_{TRX}$	—	16.0	—	ns
PMA_RX_CLK[0:1] skew	$t_{SKTRX}$	7.5	—	8.5	ns
PMA_RX_CLK[0:1] duty cycle	$t_{TRXH}/t_{TRXF}$	40	—	60	%
RCG[9:0] setup time to rising PMA_RX_CLK	$t_{TRDVKH}$	2.5	—	—	ns
RCG[9:0] hold time to rising PMA_RX_CLK	$t_{TRDXKH}$	1.5	—	—	ns
PMA_RX_CLK[0:1] clock rise time (20%-80%)	$t_{TRXR}^2$	0.7	—	2.4	ns
PMA_RX_CLK[0:1] clock fall time (80%-20%)	$t_{TRXF}^2$	0.7	—	2.4	ns

**Note:**

- The symbols used for timing specifications herein follow the pattern of  $t_{(first\ two\ letters\ of\ functional\ block)(signal)(state)\ (reference)(state)}$  for inputs and  $t_{(first\ two\ letters\ of\ functional\ block)(reference)(state)(signal)(state)}$  for outputs. For example,  $t_{TRDVKH}$  symbolizes TBI receive timing (TR) with respect to the time data input signals (D) reach the valid state (V) relative to the  $t_{TRX}$  clock reference (K) going to the high (H) state or setup time. Also,  $t_{TRDXKH}$  symbolizes TBI receive timing (TR) with respect to the time data input signals (D) went invalid (X) relative to the  $t_{TRX}$  clock reference (K) going to the high (H) state. Note that, in general, the clock reference symbol representation is based on three letters representing the clock of a particular functional. For example, the subscript of  $t_{TRX}$  represents the TBI (T) receive (RX) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall). For symbols representing skews, the subscript is skew (SK) followed by the clock that is being skewed (TRX).
- Guaranteed by design.

Figure 16 shows the TBI receive AC timing diagram.



**Figure 16. TBI Receive AC Timing Diagram**

**Table 36. RMII Receive AC Timing Specifications (continued)**

 At recommended operating conditions with  $L/TV_{DD}$  of  $3.3\text{ V} \pm 5\%$ .

Parameter/Condition	Symbol <sup>1</sup>	Min	Typ	Max	Unit
RXD[1:0], CRS_DV, RX_ER hold time to REF_CLK rising edge	$t_{\text{RMRDX}}$	2.0	—	—	ns

**Note:**

- The symbols used for timing specifications herein follow the pattern of  $t_{\text{(first two letters of functional block)(signal)(state) (reference)(state)}}$  for inputs and  $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$  for outputs. For example,  $t_{\text{MRDVKH}}$  symbolizes MII receive timing (MR) with respect to the time data input signals (D) reach the valid state (V) relative to the  $t_{\text{MRX}}$  clock reference (K) going to the high (H) state or setup time. Also,  $t_{\text{MRDXKL}}$  symbolizes MII receive timing (GR) with respect to the time data input signals (D) went invalid (X) relative to the  $t_{\text{MRX}}$  clock reference (K) going to the low (L) state or hold time. Note that, in general, the clock reference symbol representation is based on three letters representing the clock of a particular functional. For example, the subscript of  $t_{\text{MRX}}$  represents the MII (M) receive (RX) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).

Figure 20 provides the AC test load for eTSEC.

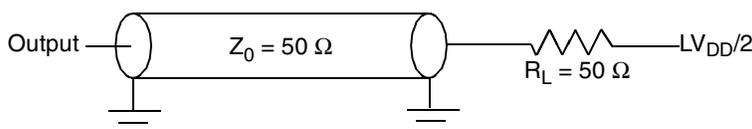
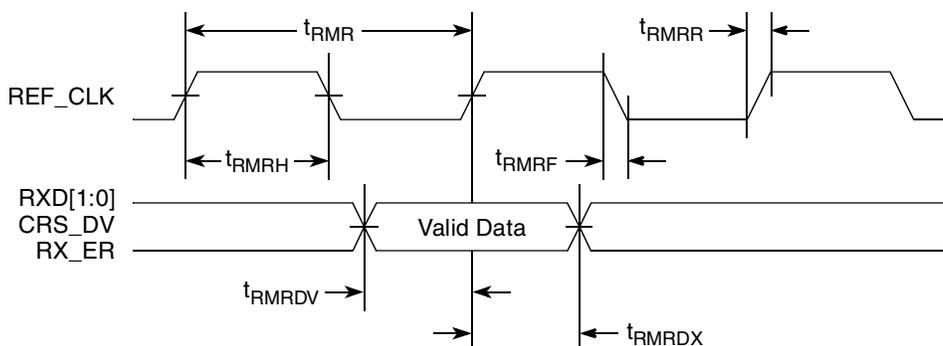

**Figure 20. eTSEC AC Test Load**

Figure 21 shows the RMII receive AC timing diagram.


**Figure 21. RMII Receive AC Timing Diagram**

### 8.3 Ethernet Management Interface Electrical Characteristics

The electrical characteristics specified here apply to MII management interface signals MDIO (management data input/output) and MDC (management data clock).

The Differential Input Voltage (or Swing) of the receiver,  $V_{ID}$ , is defined as the difference of the two complimentary input voltages:  $V_{SD\_RX} - \overline{V_{SD\_RX}}$ . The  $V_{ID}$  value can be either positive or negative.

**4. Differential Peak Voltage,  $V_{DIFFp}$**

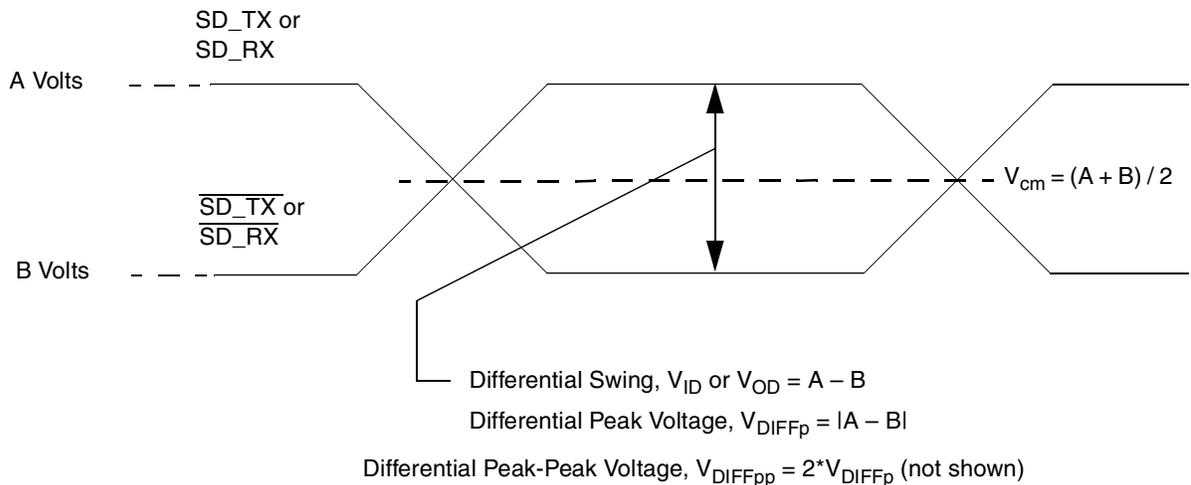
The peak value of the differential transmitter output signal or the differential receiver input signal is defined as Differential Peak Voltage,  $V_{DIFFp} = |A - B|$  Volts.

**5. Differential Peak-to-Peak,  $V_{DIFFp-p}$**

Since the differential output signal of the transmitter and the differential input signal of the receiver each range from  $A - B$  to  $-(A - B)$  Volts, the peak-to-peak value of the differential transmitter output signal or the differential receiver input signal is defined as Differential Peak-to-Peak Voltage,  $V_{DIFFp-p} = 2 * V_{DIFFp} = 2 * |A - B|$  Volts, which is twice of differential swing in amplitude, or twice of the differential peak. For example, the output differential peak-peak voltage can also be calculated as  $V_{TX-DIFFp-p} = 2 * |V_{OD}|$ .

**6. Common Mode Voltage,  $V_{cm}$**

The Common Mode Voltage is equal to one half of the sum of the voltages between each conductor of a balanced interchange circuit and ground. In this example, for SerDes output,  $V_{cm\_out} = \overline{V_{SD\_TX}} + V_{SD\_TX} = (A + B) / 2$ , which is the arithmetic mean of the two complimentary output voltages within a differential pair. In a system, the common mode voltage may often differ from one component's output to the other's input. Sometimes, it may be even different between the receiver input and driver output circuits within the same component. It is also referred as the DC offset in some occasion.



**Figure 39. Differential Voltage Definitions for Transmitter or Receiver**

To illustrate these definitions using real values, consider the case of a CML (Current Mode Logic) transmitter that has a common mode voltage of 2.25 V and each of its outputs, TD and  $\overline{TD}$ , has a swing that goes between 2.5 V and 2.0 V. Using these values, the peak-to-peak voltage swing of each signal (TD or  $\overline{TD}$ ) is 500 mV p-p, which is referred as the single-ended swing for each signal. In this example, since the differential signaling environment is fully symmetrical, the transmitter output's differential swing ( $V_{OD}$ ) has the same amplitude as each signal's single-ended swing. The differential output signal ranges

### 13.3 SerDes Transmitter and Receiver Reference Circuits

Figure 48 shows the reference circuits for SerDes data lane’s transmitter and receiver.

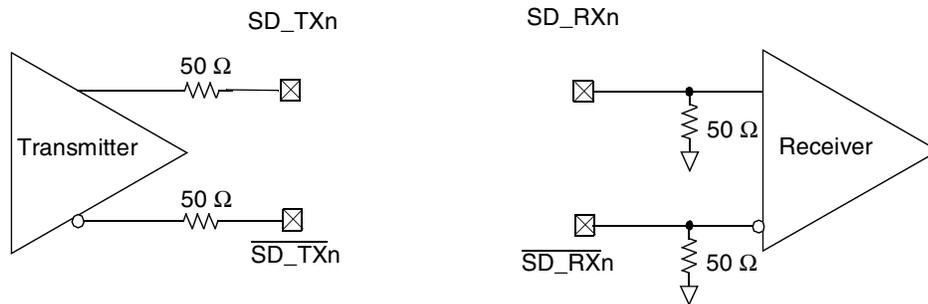


Figure 48. SerDes Transmitter and Receiver Reference Circuits

The DC and AC specification of SerDes data lanes are defined in each interface protocol section below (PCI Express, Serial Rapid IO or SGMII) in this document based on the application usage:

- Section 14, “PCI Express”
- Section 15, “Serial RapidIO”

Note that external AC Coupling capacitor is required for the above three serial transmission protocols with the capacitor value defined in specification of each protocol section.

## 14 PCI Express

This section describes the DC and AC electrical specifications for the PCI Express bus of the MPC8568E.

### 14.1 DC Requirements for PCI Express SD\_REF\_CLK and SD\_REF\_CLK

For more information, see Section 13, “High-Speed Serial Interfaces (HSSI).”

### 14.2 AC Requirements for PCI Express SerDes Clocks

Table 50 lists AC requirements.

Table 50. SD\_REF\_CLK and SD\_REF\_CLK AC Requirements

Symbol	Parameter Description	Min	Typical	Max	Units	Notes
t <sub>REF</sub>	REFCLK cycle time	—	10	—	ns	1
t <sub>REFCJ</sub>	REFCLK cycle-to-cycle jitter. Difference in the period of any two adjacent REFCLK cycles	—	—	100	ps	—

**Table 50. SD\_REF\_CLK and SD\_REF\_CLK AC Requirements**

Symbol	Parameter Description	Min	Typical	Max	Units	Notes
$t_{REFPJ}$	Phase jitter. Deviation in edge location with respect to mean edge location	-50	—	50	ps	—

**Notes:**

1. Typical based on PCI Express Specification 2.0.

## 14.3 Clocking Dependencies

The ports on the two ends of a link must transmit data at a rate that is within 600 parts per million (ppm) of each other at all times. This is specified to allow bit rate clock sources with a +/- 300 ppm tolerance.

## 14.4 Physical Layer Specifications

The following is a summary of the specifications for the physical layer of PCI Express on this device. For further details as well as the specifications of the Transport and Data Link layer please use the PCI EXPRESS Base Specification. REV. 1.0a document.

### 14.4.1 Differential Transmitter (TX) Output

Table 51 defines the specifications for the differential output at all transmitters (TXs). The parameters are specified at the component pins.

**Table 51. Differential Transmitter (TX) Output Specifications**

Symbol	Parameter	Min	Nom	Max	Units	Comments
UI	Unit Interval	399.88	400	400.12	ps	Each UI is 400 ps $\pm$ 300 ppm. UI does not account for Spread Spectrum Clock dictated variations. See Note 1.
$V_{TX-DIFFp-p}$	Differential Peak-to-Peak Output Voltage	0.8	—	1.2	V	$V_{TX-DIFFp-p} = 2 *  V_{TX-D+} - V_{TX-D-} $ See Note 2.
$V_{TX-DE-RATIO}$	De-Emphasized Differential Output Voltage (Ratio)	-3.0	-3.5	-4.0	dB	Ratio of the $V_{TX-DIFFp-p}$ of the second and following bits after a transition divided by the $V_{TX-DIFFp-p}$ of the first bit after a transition. See Note 2.
$T_{TX-EYE}$	Minimum TX Eye Width	0.70	—	—	UI	The maximum Transmitter jitter can be derived as $T_{TX-MAX-JITTER} = 1 - T_{TX-EYE} = 0.3$ UI. See Notes 2 and 3.
$T_{TX-EYE-MEDIAN-to-MAX-JITTER}$	Maximum time between the jitter median and maximum deviation from the median.	—	—	0.15	UI	Jitter is defined as the measurement variation of the crossing points ( $V_{TX-DIFFp-p} = 0$ V) in relation to a recovered TX UI. A recovered TX UI is calculated over 3500 consecutive unit intervals of sample data. Jitter is measured using all edges of the 250 consecutive UI in the center of the 3500 UI used for calculating the TX UI. See Notes 2 and 3.
$T_{TX-RISE}, T_{TX-FALL}$	D+/D- TX Output Rise/Fall Time	0.125	—	—	UI	See Notes 2 and 5

**Table 60. Transmitter Differential Output Eye Diagram Parameters**

Transmitter Type	V <sub>DIFFmin</sub> (mV)	V <sub>DIFFmax</sub> (mV)	A (UI)	B (UI)
1.25 GBaud short range	250	500	0.175	0.39
1.25 GBaud long range	400	800	0.175	0.39
2.5 GBaud short range	250	500	0.175	0.39
2.5 GBaud long range	400	800	0.175	0.39
3.125 GBaud short range	250	500	0.175	0.39
3.125 GBaud long range	400	800	0.175	0.39

## 15.7 Receiver Specifications

LP-Serial receiver electrical and timing specifications are stated in the text and tables of this section.

Receiver input impedance shall result in a differential return loss better than 10 dB and a common mode return loss better than 6 dB from 100 MHz to  $(0.8) \times (\text{Baud Frequency})$ . This includes contributions from on-chip circuitry, the chip package and any off-chip components related to the receiver. AC coupling components are included in this requirement. The reference impedance for return loss measurements is 100  $\Omega$  resistive for differential return loss and 25  $\Omega$  resistive for common mode.

**Table 61. Receiver AC Timing Specifications—1.25 GBaud**

Characteristic	Symbol	Range		Unit	Notes
		Min	Max		
Differential Input Voltage	V <sub>IN</sub>	200	1600	mV p-p	Measured at receiver
Deterministic Jitter Tolerance	J <sub>D</sub>	0.37	—	UI p-p	Measured at receiver
Combined Deterministic and Random Jitter Tolerance	J <sub>DR</sub>	0.55	—	UI p-p	Measured at receiver
Total Jitter Tolerance <sup>1</sup>	J <sub>T</sub>	0.65	—	UI p-p	Measured at receiver
Multiple Input Skew	S <sub>MI</sub>	—	24	ns	Skew at the receiver input between lanes of a multilane link
Bit Error Rate	BER	—	10 <sup>-12</sup>	—	—
Unit Interval	UI	800	800	ps	+/- 100 ppm

**Note:**

- Total jitter is composed of three components, deterministic jitter, random jitter and single frequency sinusoidal jitter. The sinusoidal jitter may have any amplitude and frequency in the unshaded region of [Figure 54](#). The sinusoidal jitter component is included to ensure margin for low frequency jitter, wander, noise, crosstalk and other variable system effects.

**Table 76. HDLC, BiSync, Transparent AC Timing Specifications <sup>1</sup> (continued)**

Characteristic	Symbol <sup>2</sup>	Min	Max	Unit
Outputs—Internal clock High Impedance	$t_{HIKHOX}$	0	5.5	ns
Outputs—External clock High Impedance	$t_{HEKHOX}$	1	8	ns
Inputs—Internal clock input setup time	$t_{HIIVKH}$	6	—	ns
Inputs—External clock input setup time	$t_{HEIVKH}$	4	—	ns
Inputs—Internal clock input Hold time	$t_{HIIXKH}$	0	—	ns
Inputs—External clock input hold time	$t_{HEIXKH}$	1	—	ns

**Notes:**

- Output specifications are measured from the 50% level of the rising edge of CLKIN to the 50% level of the signal. Timings are measured at the pin.
- The symbols used for timing specifications follow the pattern of  $t_{(first\ two\ letters\ of\ functional\ block)(signal)(state)}$  (reference)(state) for inputs and  $t_{(first\ two\ letters\ of\ functional\ block)(reference)(state)(signal)(state)}$  for outputs. For example,  $t_{HIKHOX}$  symbolizes the outputs internal timing (HI) for the time  $t_{serial}$  memory clock reference (K) goes from the high state (H) until outputs (O) are invalid (X).

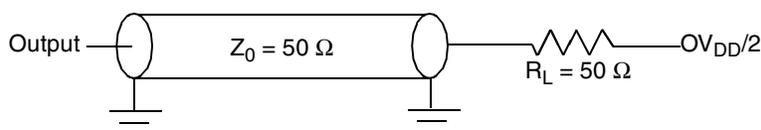
**Table 77. Synchronous UART AC Timing Specifications**

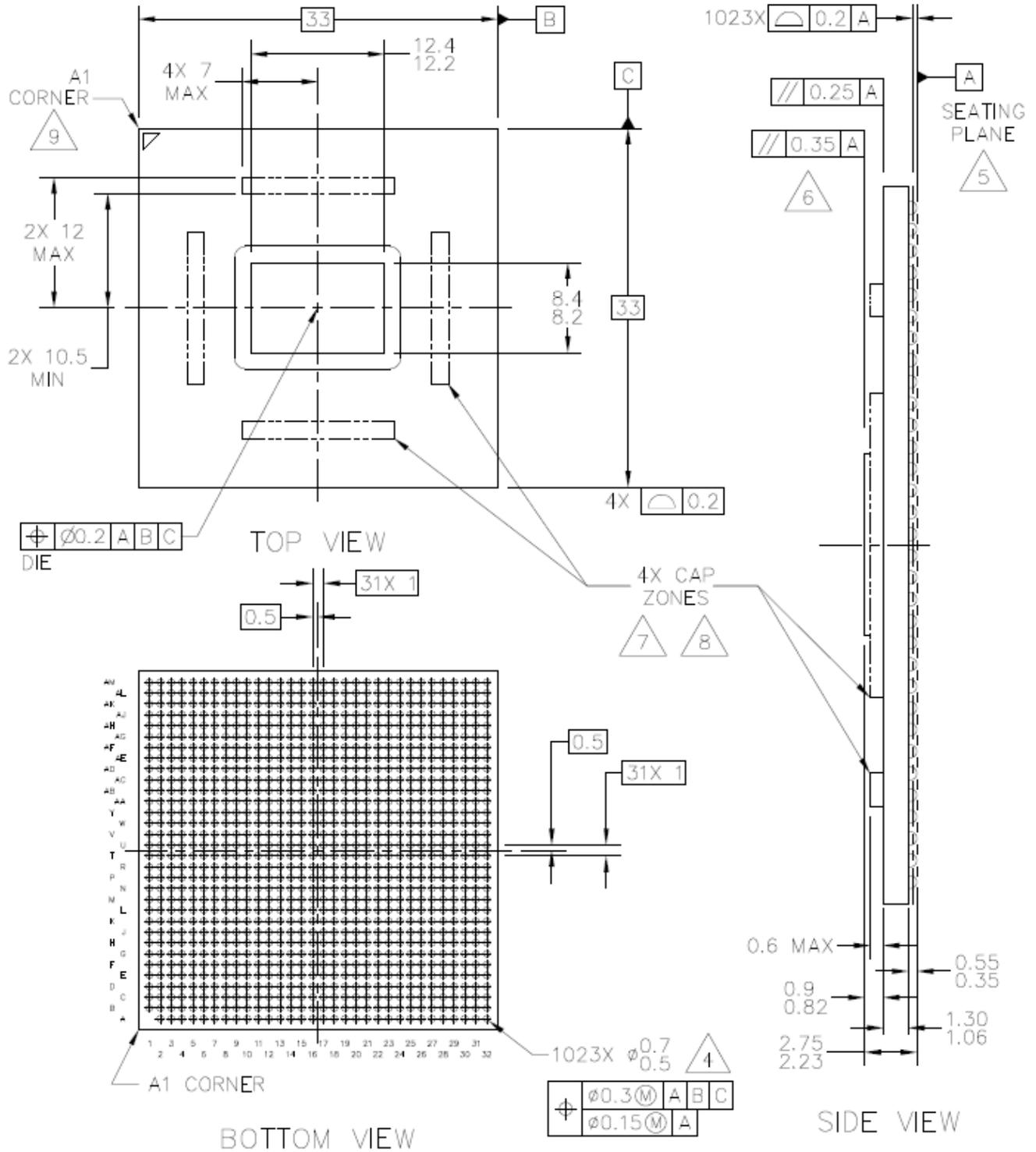
Characteristic	Symbol <sup>2</sup>	Min	Max	Unit
Outputs—Internal clock delay	$t_{HIKHOV}$	0	11	ns
Outputs—External clock delay	$t_{HEKHOV}$	1	14	ns
Outputs—Internal clock High Impedance	$t_{HIKHOX}$	0	11	ns
Outputs—External clock High Impedance	$t_{HEKHOX}$	1	14	ns
Inputs—Internal clock input setup time	$t_{HIIVKH}$	6	—	ns
Inputs—External clock input setup time	$t_{HEIVKH}$	8	—	ns
Inputs—Internal clock input Hold time	$t_{HIIXKH}$	0	—	ns
Inputs—External clock input hold time	$t_{HEIXKH}$	1	—	ns

**Notes:**

- Output specifications are measured from the 50% level of the rising edge of CLKIN to the 50% level of the signal. Timings are measured at the pin.
- The symbols used for timing specifications follow the pattern of  $t_{(first\ two\ letters\ of\ functional\ block)(signal)(state)}$  (reference)(state) for inputs and  $t_{(first\ two\ letters\ of\ functional\ block)(reference)(state)(signal)(state)}$  for outputs. For example,  $t_{HIKHOX}$  symbolizes the outputs internal timing (HI) for the time  $t_{serial}$  memory clock reference (K) goes from the high state (H) until outputs (O) are invalid (X).

Figure 65 provides the AC test load.


**Figure 65. AC Test Load**



**Figure 68. Top, Bottom, Side Views**

1. All dimensions are in millimeters.
2. Dimensions and tolerances per ASME Y14.5M-1994.
3. All dimensions are symmetric across the package center lines, unless dimensioned otherwise.

4. Maximum solder ball diameter measured parallel to datum A.
5. Datum A, the seating plane, is defined by the spherical crowns of the solder balls.
6. ParalleUsm measurement shall exclude any effect of mark on top surface of package
7. Capacitors may not be present on all devices.
8. Caution must be taken not to short capacitors or exposed metal capacitor pads on top of package.

## 22.3 Pinout Listings

Some of the non-QE signals are multiplexed with QE port pins, as follows:

PC[0]     UART\_SOUT[1]  
 PC[1]      $\overline{\text{UART\_RTS}}[1]$   
 PC[2]      $\overline{\text{UART\_CTS}}[1]$   
 PC[3]     UART\_SIN[1]

PC[11]    IRQ[8]  
 PC[12]    IRQ[9]/ $\overline{\text{DMA\_DREQ}}[3]$   
 PC[13]    IRQ[10]/ $\overline{\text{DMA\_DACK}}[3]$   
 PC[14]    IRQ[11]/ $\overline{\text{DMA\_DDONE}}[3]$   
 PC[15]     $\overline{\text{DMA\_DREQ}}[1]$   
 PC[16]     $\overline{\text{DMA\_DACK}}[1]$   
 PC[17]     $\overline{\text{DMA\_DDONE}}[1]$   
 PC[18]    IIC2\_SCL  
 PC[19]    IIC2\_SDA

PD[28]    UART\_SOUT[1]  
 PD[29]     $\overline{\text{UART\_RTS}}[1]$   
 PD[30]     $\overline{\text{UART\_CTS}}[1]$   
 PD[31]    UART\_SIN[1]

This applies to both MPC8568E and MPC8568E. Note that for DUART1, there are two options. DUART0 is multiplexed with PCI Req/Grant pins.

$\overline{\text{PCI\_REQ}}[3]$       $\overline{\text{UART\_CTS}}[0]$   
 $\overline{\text{PCI\_REQ}}[4]$      UART\_SIN[0]  
 $\overline{\text{PCI\_GNT}}[3]$      $\overline{\text{UART\_RTS}}[0]$   
 $\overline{\text{PCI\_GNT}}[4]$     UART\_SOUT[0]

For MPC8568E, GPIO is multiplexed with the TSEC2 interface:

TSEC2\_TXD[7:0]    GPOUT[0:7]

Table 78. MPC8568E Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
<b>Debug</b>				
TRIG_IN	AL29	I	OV <sub>DD</sub>	—
TRIG_OUT	AM29	O	OV <sub>DD</sub>	6,9,19,29
MSRCID[0:1]	AK29, AJ29	O	OV <sub>DD</sub>	5,6,9
MSRCID[2:4]	AM28, AL28, AK27	O	OV <sub>DD</sub>	6,19,29
MDVAL	AJ28	O	OV <sub>DD</sub>	6
CLK_OUT	AF18	O	OV <sub>DD</sub>	11
<b>Clock</b>				
RTC	AH20	I	OV <sub>DD</sub>	—
SYSCLK	AK22	I	OV <sub>DD</sub>	—
<b>JTAG</b>				
TCK	AH18	I	OV <sub>DD</sub>	—
TDI	AH19	I	OV <sub>DD</sub>	12
TDO	AJ18	O	OV <sub>DD</sub>	11
TMS	AK19	I	OV <sub>DD</sub>	12
$\overline{\text{TRST}}$	AK20	I	OV <sub>DD</sub>	12
<b>DFT</b>				
L1_TSTCLK	AJ20	I	OV <sub>DD</sub>	25
L2_TSTCLK	AJ19	I	OV <sub>DD</sub>	25
$\overline{\text{LSSD\_MODE}}$	AH31	I	OV <sub>DD</sub>	25
$\overline{\text{TEST\_SEL}}$	AJ31	I	OV <sub>DD</sub>	25
<b>Thermal Management</b>				
THERM0	AB30	—	—	14
THERM1	AB31	—	—	14
<b>Power Management</b>				
ASLEEP	AK21	O	OV <sub>DD</sub>	9,19,29

**Table 78. MPC8568E Pinout Listing (continued)**

Signal	Package Pin Number	Pin Type	Power Supply	Notes
30. This pin requires an external 4.7-kΩ pull-down resistor to prevent PHY from seeing a valid Transmit Enable before it is actively driven.				
33. PF[21:22] are multiplexed as cfg_dram_type[0:1]. THEY MUST BE VALID AT POWER-UP, EVEN BEFORE $\overline{\text{HRESET}}$ ASSERTION.				
35. When a PCI block is disabled, either the POR config pin that selects between internal and external arbiter must be pulled down to select external arbiter if there is any other PCI device connected on the PCI bus, or leave the PCIn_AD pins as "No Connect" or terminated through 2–10 KΩ pull-up resistors with the default of internal arbiter if the PCIn_AD pins are not connected to any other PCI device. The PCI block will drive the PCIn_AD pins if it is configured to be the PCI arbiter—through POR config pins—irrespective of whether it is disabled via the DEVDISR register or not. It may cause contention if there is any other PCI device connected on the bus.				
36. MDIC[0] is grounded through an 18.2-Ω precision 1% resistor and MDIC[1] is connected to GV <sub>DD</sub> through an 18.2-Ω precision 1% resistor. These pins are used for automatic calibration of the DDR IOs.				
39. If PCI is configured as PCI asynchronous mode, a valid clock must be provided on pin PCI_CLK . Otherwise the processor will not boot up.				
41. These pins should be tied to SCOREGND through a 300 ohm resistor if the high speed interface is used.				
43. It is highly recommended that unused SD_RX/SD_RX lanes should be powered down with lane_x_pd. Otherwise the receivers will burn extra power and the internal circuitry may develop long term reliability problems.				
44. See <a href="#">Section 25.9, "Guidelines for High-Speed Interface Termination."</a>				
46. Must be high during $\overline{\text{HRESET}}$ . It is recommended to leave the pin open during $\overline{\text{HRESET}}$ since it has internal pullup resistor.				
47. Must be pulled down with 4.7-kΩ resistor.				
48. This pin must be left no connect.				
49. A pull-up on LGPL4 is required for systems that boot from local bus (GPCM)-controlled NOR Flash.				

[Table 79](#) provides the pin-out listing for the MPC8567E 1023 FC-PBGA package.

**Table 79. MPC8567E Pinout Listing**

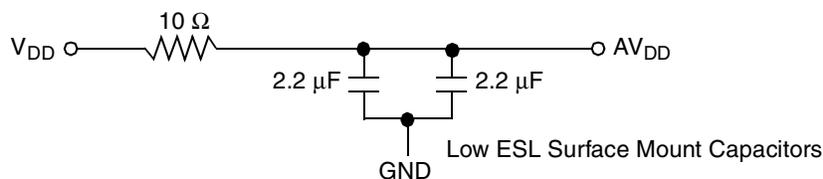
Signal	Package Pin Number	Pin Type	Power Supply	Notes
<b>PCI</b>				
PCI_AD[31:0]	AE19, AG20, AF19, AB20, AC20, AG21, AG22, AB21, AF22, AH22, AE22, AF20, AB22, AE20, AE23, AJ23, AJ24, AF27, AJ26, AE29, AH24, AD24, AE25, AE26, AH27, AG27, AJ25, AE30, AF26, AG26, AF28, AH26	I/O	OV <sub>DD</sub>	—
PCI_C_BE[3:0]	AC22, AD20, AE28, AH25	I/O	OV <sub>DD</sub>	—
PCI_GNT[4:1]	AF29, AB18, AC18, AD18	O	OV <sub>DD</sub>	5,9,35
PCI_GNT0	AE18	I/O	OV <sub>DD</sub>	—
PCI_IRDY	AF23	I/O	OV <sub>DD</sub>	2
PCI_PAR	AJ22	I/O	OV <sub>DD</sub>	—
PCI_PERR	AF24	I/O	OV <sub>DD</sub>	2
PCI_SERR	AD22	I/O	OV <sub>DD</sub>	2,4
PCI_STOP	AE24	I/O	OV <sub>DD</sub>	2
PCI_TRDY	AK24	I/O	OV <sub>DD</sub>	2

Table 79. MPC8567E Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
PE[20]	AH6	I/O	OV <sub>DD</sub>	—
PE[21:23]	AM1, AE10, AG5	I/O	OV <sub>DD</sub>	5
PE[24]	AJ1	I/O	OV <sub>DD</sub>	5
PE[25:31]	AH2, AM2, AE9, AH5, AL1, AD9, AL4	I/O	OV <sub>DD</sub>	—
PF[7]	AG9	I/O	TV <sub>DD</sub>	—
PF[8:10]	AF10, AK7, AJ6	I/O	TV <sub>DD</sub>	5
PF[11:19]	AH7, AF9, AJ7, AJ5, AF7, AG8, AG7, AM5, AK5	I/O	TV <sub>DD</sub>	—
PF[20]	AK1	I/O	OV <sub>DD</sub>	—
PF[21:22]	AH3, AL3	I/O	OV <sub>DD</sub>	5,33
PF[23:31]	AB11, AE7, AJ3, AC11, AG6, AG3, AH4, AM3, AD11	I/O	OV <sub>DD</sub>	—
<b>System Control</b>				
HRESET	AL21	I	OV <sub>DD</sub>	—
HRESET_REQ	AL23	O	OV <sub>DD</sub>	29
SRESET	AK18	I	OV <sub>DD</sub>	—
CKSTP_IN	AL17	I	OV <sub>DD</sub>	—
CKSTP_OUT	AM17	O	OV <sub>DD</sub>	2,4
<b>Debug</b>				
TRIG_IN	AL29	I	OV <sub>DD</sub>	—
TRIG_OUT	AM29	O	OV <sub>DD</sub>	6,9,19,29
MSRCID[0:1]	AK29, AJ29	O	OV <sub>DD</sub>	5,6,9
MSRCID[2:4]	AM28, AL28, AK27	O	OV <sub>DD</sub>	6,19,29
MDVAL	AJ28	O	OV <sub>DD</sub>	6
CLK_OUT	AF18	O	OV <sub>DD</sub>	11
<b>Clock</b>				
RTC	AH20	I	OV <sub>DD</sub>	—
SYCLK	AK22	I	OV <sub>DD</sub>	—
<b>JTAG</b>				
TCK	AH18	I	OV <sub>DD</sub>	—
TDI	AH19	I	OV <sub>DD</sub>	12
TDO	AJ18	O	OV <sub>DD</sub>	11
TMS	AK19	I	OV <sub>DD</sub>	12
TRST	AK20	I	OV <sub>DD</sub>	12

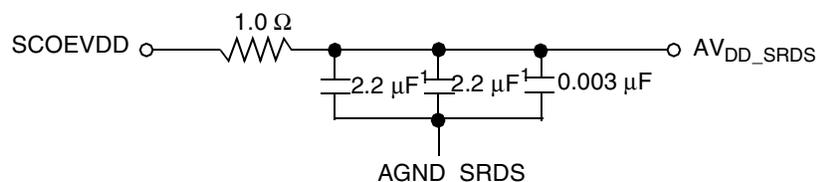
Each circuit should be placed as close as possible to the specific  $AV_{DD}$  type pin being supplied to minimize noise coupled from nearby circuits. It should be possible to route directly from the capacitors to the  $AV_{DD}$  type pin, which is on the periphery of 1023FC-PBGA the footprint, without the inductance of vias.

Figure 73 shows the PLL power supply filter circuits for all PLLs except SerDes PLL.



**Figure 73. MPC8568E PLL Power Supply Filter Circuit**

The  $AV_{DD\_SRDS}$  signal provides power for the analog portions of the SerDes PLL. To ensure stability of the internal clock, the power supplied to the PLL is filtered using a circuit similar to the one shown in following figure. For maximum effectiveness, the filter circuit is placed as closely as possible to the  $AV_{DD\_SRDS}$  and  $AGND\_SRDS$  ball to ensure it filters out as much noise as possible. The  $0.003\text{-}\mu\text{F}$  capacitor is closest to the ball, followed by the  $2.2\text{-}\mu\text{F}$  capacitors, and finally the 1 ohm resistor to the board supply plane. Use ceramic chip capacitors with the highest possible self-resonant frequency. All traces should be kept short, wide and direct.



1. An 0805 sized capacitor is recommended for system initial bring-up.

**Figure 74. SerDes PLL Power Supply Filter**

Note the following:

- $AV_{DD\_SRDS}$  should be a filtered version of SCOREVDD.
- The transmitter output signals on the SerDes interface are fed from the  $XV_{DD}$  power plan.
- Power:  $XV_{DD}$  consumes less than 300mW.  $SCOREVDD + AV_{DD\_SRDS}$  consumes less than 750mW.

## 25.3 Decoupling Recommendations

Due to large address and data buses, and high operating frequencies, the device can generate transient power surges and high frequency noise in its power supply, especially while driving large capacitive loads. MPC8568E system, and the device itself requires a clean, tightly regulated source of power. Therefore, it is recommended that the system designer place at least one decoupling capacitor at each  $V_{DD}$ ,  $TV_{DD}$ ,  $BV_{DD}$ ,  $OV_{DD}$ ,  $GV_{DD}$ , and  $LV_{DD}$  pin of the device. These decoupling capacitors should receive their power from separate  $V_{DD}$ ,  $TV_{DD}$ ,  $BV_{DD}$ ,  $OV_{DD}$ ,  $GV_{DD}$ , and  $LV_{DD}$  and GND power planes in the PCB, utilizing short traces to minimize inductance. Capacitors may be placed directly under the device using a standard escape pattern. Others may surround the part.

These capacitors should have a value of 0.01 or 0.1  $\mu\text{F}$ . Only ceramic SMT (surface mount technology) capacitors should be used to minimize lead inductance, preferably 0402 or 0603 sizes.

In addition, it is recommended that there be several bulk storage capacitors distributed around the PCB, feeding the  $V_{\text{DD}}$ ,  $TV_{\text{DD}}$ ,  $BV_{\text{DD}}$ ,  $OV_{\text{DD}}$ ,  $GV_{\text{DD}}$ , and  $LV_{\text{DD}}$  planes, to enable quick recharging of the smaller chip capacitors. These bulk capacitors should have a low ESR (equivalent series resistance) rating to ensure the quick response time necessary. They should also be connected to the power and ground planes through two vias to minimize inductance. Suggested bulk capacitors—100–330  $\mu\text{F}$  (AVX TPS tantalum or Sanyo OSCON).

## 25.4 SerDes Block Power Supply Decoupling Recommendations

The SerDes block requires a clean, tightly regulated source of power ( $\text{SCOREVDD}$  and  $XV_{\text{DD}}$ ) to ensure low jitter on transmit and reliable recovery of data in the receiver. An appropriate decoupling scheme is outlined below.

Only surface mount technology (SMT) capacitors should be used to minimize inductance. Connections from all capacitors to power and ground should be done with multiple vias to further reduce inductance.

- First, the board should have at least 10 x 10-nF SMT ceramic chip capacitors as close as possible to the supply balls of the device. Where the board has blind vias, these capacitors should be placed directly below the chip supply and ground connections. Where the board does not have blind vias, these capacitors should be placed in a ring around the device as close to the supply and ground connections as possible.
- Second, there should be a 1- $\mu\text{F}$  ceramic chip capacitor on each side of the device. This should be done for all SerDes supplies.
- Third, between the device and any SerDes voltage regulator there should be a 10- $\mu\text{F}$ , low equivalent series resistance (ESR) SMT tantalum chip capacitor and a 100- $\mu\text{F}$ , low ESR SMT tantalum chip capacitor. This should be done for all SerDes supplies.

## 25.5 Connection Recommendations

To ensure reliable operation, it is highly recommended to connect unused inputs to an appropriate signal level. All unused active low inputs should be tied to  $V_{\text{DD}}$ ,  $TV_{\text{DD}}$ ,  $BV_{\text{DD}}$ ,  $OV_{\text{DD}}$ ,  $GV_{\text{DD}}$  and  $LV_{\text{DD}}$  as required. All unused active high inputs should be connected to GND. All NC (no-connect) signals must remain unconnected. Power and ground connections must be made to all external  $V_{\text{DD}}$ ,  $LV_{\text{DD}}$ ,  $TV_{\text{DD}}$ ,  $BV_{\text{DD}}$ ,  $OV_{\text{DD}}$ ,  $GV_{\text{DD}}$  and GND pins of the device.

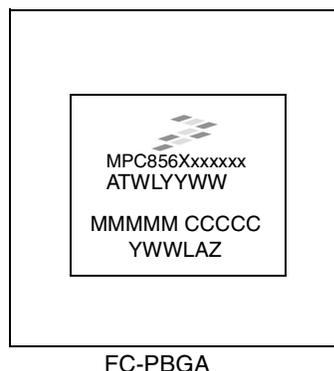
## 25.6 Pull-Up and Pull-Down Resistor Requirements

The MPC8568E requires weak pull-up resistors (2–10  $\text{k}\Omega$  is recommended) on open drain type pins including I<sup>2</sup>C pins and MPIC interrupt pins.

Correct operation of the JTAG interface requires configuration of a group of system control pins as demonstrated in [Figure 75](#). Care must be taken to ensure that these pins are maintained at a valid deasserted state under normal operating conditions as most have asynchronous behavior and spurious assertion will give unpredictable results.

## 26.1 Part Marking

Parts are marked as the example shown in [Figure 76](#).



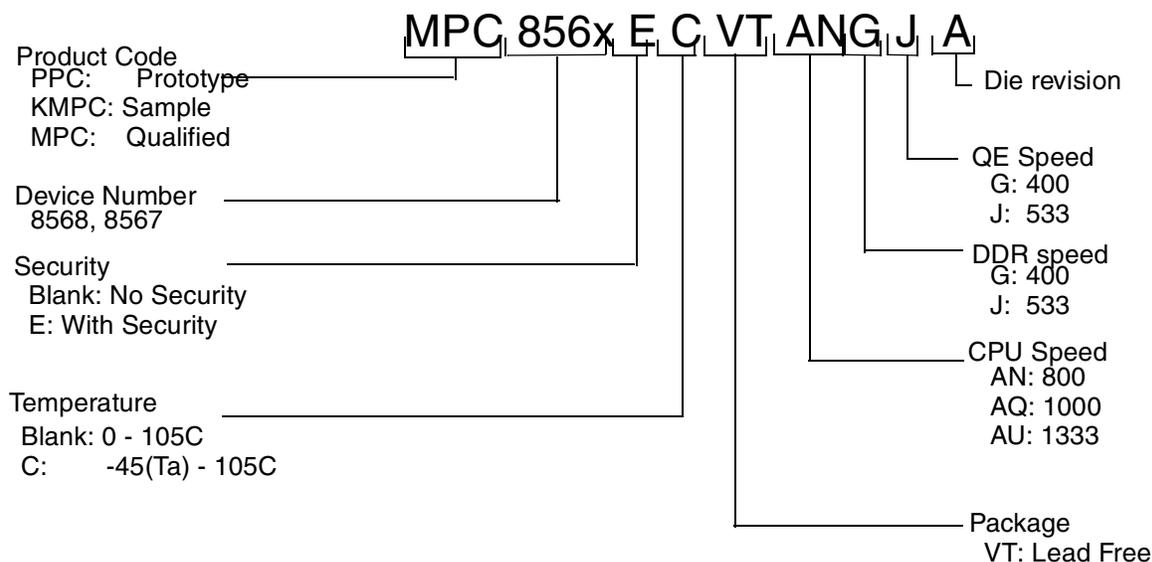
**Notes:**

MPC856Xxxxxxx is the orderable part number  
 ATWLYYWW is the freescale assembly, year and workweek code  
 MMMMM is the mask code  
 CCCC is the contry code for assembly.  
 YWWLAZ is the trace code for assembly.

**Figure 76. Part Marking for FC-PBGA Device**

## 26.2 Part Number Decoder

[Figure 77](#) shows the MPC8568E/MPC8567E number decoder.



**Figure 77. MPC8568E Part Number Decoder**