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Applications of "[Embedded - Microcontrollers](#)"

Details	
Product Status	Obsolete
Core Processor	HC08
Core Size	8-Bit
Speed	8MHz
Connectivity	-
Peripherals	LVD, POR, PWM
Number of I/O	13
Program Memory Size	1.5KB (1.5K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	128 x 8
Voltage - Supply (Vcc/Vdd)	2.7V ~ 5.5V
Data Converters	-
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	16-TSSOP (0.173", 4.40mm Width)
Supplier Device Package	16-TSSOP
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mc908qy1acdte

MC68HC908QY4A
MC68HC908QT4A
MC68HC908QY2A
MC68HC908QT2A
MC68HC908QY1A
MC68HC908QT1A

Data Sheet

M68HC08
Microcontrollers

MC68HC908QY4A
Rev. 3
03/2010

freescale.com

2.6.6 FLASH Block Protect Register

The FLASH block protect register is implemented as a byte within the FLASH memory, and therefore can only be written during a programming sequence of the FLASH memory. The value in this register determines the starting address of the protected range within the FLASH memory.

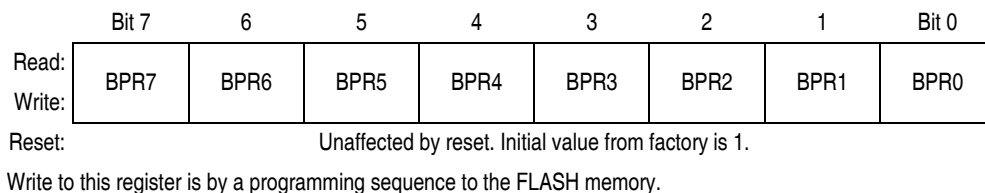


Figure 2-5. FLASH Block Protect Register (FLBPR)

BPR[7:0] — FLASH Protection Register Bits [7:0]

These eight bits in FLBPR represent bits [13:6] of a 16-bit memory address. Bits [15:14] are 1s and bits [5:0] are 0s.

The resultant 16-bit address is used for specifying the start address of the FLASH memory for block protection. The FLASH is protected from this start address to the end of FLASH memory, at \$FFFF. With this mechanism, the protect start address can be XX00, XX40, XX80, or XXC0 within the FLASH memory. See [Figure 2-6](#) and [Table 2-2](#).

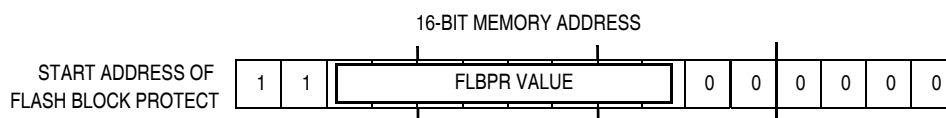


Figure 2-6. FLASH Block Protect Start Address

Table 2-2. Examples of Protect Start Address

BPR[7:0]	Start of Address of Protect Range
\$00–\$B8	The entire FLASH memory is protected.
\$B8 (1011 1000)	\$EE00 (1110 1110 0000 0000)
\$B9 (1011 1001)	\$EE40 (1110 1110 0100 0000)
\$BA (1011 1010)	\$EE80 (1110 1110 1000 0000)
\$BB (1011 1011)	\$EFC0 (1110 1110 1100 0000)
and so on...	
\$DE (1101 1110)	\$F780 (1111 0111 1000 0000)
\$DF (1101 1111)	\$F7C0 (1111 0111 1100 0000)
\$FE (1111 1110)	\$FF80 (1111 1111 1000 0000) FLBPR, internal oscillator trim values, and vectors are protected
\$FF	The entire FLASH memory is not protected.

2.6.7 EEPROM Memory Emulation Using FLASH Memory

In some applications, the user may want to repeatedly store and read a set of data from an area of nonvolatile memory. This is easily implemented in EEPROM memory because single byte erase is allowed in EEPROM.

When using FLASH memory, the minimum erase size is a page. However, the FLASH can be used as EEPROM memory. This technique is called “EEPROM emulation”.

The basic concept of EEPROM emulation using FLASH is that a page is continuously programmed with a new data set without erasing the previously programmed locations. Once the whole page is completely programmed or the page does not have enough bytes to program a new data set, the user software automatically erases the page and then programs a new data set in the erased page.

In EEPROM emulation when data is read from the page, the user software must find the latest data set in the page since the previous data still remains in the same page. There are many ways to monitor the page erase timing and the latest data set. One example is unprogrammed FLASH bytes are detected by checking programmed bytes (non-\$FF value) in a page. In this way, the end of the data set will contain unprogrammed data (\$FF value).

A couple of application notes, describing how to emulate EEPROM using FLASH, are available on our web site. Titles and order numbers for these application notes are given at the end of this subsection.

For EEPROM emulation software to work successfully, the following items must be taken care of in the user software:

1. Each FLASH byte in a page must be programmed only one time until the page is erased.
2. A page must be erased before the FLASH cumulative program HV period (t_{HV}) is beyond the maximum t_{HV} . t_{HV} is defined as the cumulative high-voltage programming time to the same row before the next erase. For more detailed information, refer to [16.15 Memory Characteristics](#).
3. FLASH row erase and program cycles should not exceed 10,000 cycles, respectively.

The above EEPROM emulation software can be easily developed by using the on-chip FLASH routines implemented in the MCU. These routines are located in the ROM memory and support FLASH program and erase operations. Proper utilization of the on-chip FLASH routines guarantee conformance to the FLASH specifications.

In the on-chip FLASH programming routine called PRGRNGE, the high-voltage programming time is enabled for less than 125 μ s when programming a single byte at any operating bus frequency between 1.0 MHz and 8.4 MHz. Therefore, even when a row is programmed by 32 separate single-byte programming operations, t_{HV} is less than the maximum t_{HV} . Hence, item 2 listed above is already taken care of by using this routine.

A page erased operation is provided in the FLASH erase routine called ERARNGE.

Application note AN2635 (*On-Chip FLASH Programming Routines*) describes how to use these routines.

The following application notes, available at www.freescale.com, describe how EEPROM emulation is implemented using FLASH:

AN2183 — *Using FLASH as EEPROM on the MC68HC908GP32*

AN2346 — *EEPROM Emulation Using FLASH in MC68HC908QY/QT MCUs*

AN2690 — *Low Frequency EEPROM Emulation on the MC68HC908QY4*

An EEPROM emulation driver, available at www.freescale.com, has been developed and qualified:

AN3040 — *M68HC08 EEPROM Emulation Driver*

Chapter 7

Central Processor Unit (CPU)

7.1 Introduction

The M68HC08 CPU (central processor unit) is an enhanced and fully object-code-compatible version of the M68HC05 CPU. The *CPU08 Reference Manual* (document order number CPU08RM/AD) contains a description of the CPU instruction set, addressing modes, and architecture.

7.2 Features

Features of the CPU include:

- Object code fully upward-compatible with M68HC05 Family
- 16-bit stack pointer with stack manipulation instructions
- 16-bit index register with x-register manipulation instructions
- 8-MHz CPU internal bus frequency
- 64-Kbyte program/data memory space
- 16 addressing modes
- Memory-to-memory data moves without using accumulator
- Fast 8-bit by 8-bit multiply and 16-bit by 8-bit divide instructions
- Enhanced binary-coded decimal (BCD) data handling
- Modular architecture with expandable internal bus definition for extension of addressing range beyond 64 Kbytes
- Low-power stop and wait modes

7.3 CPU Registers

Figure 7-1 shows the five CPU registers. CPU registers are not part of the memory map.

Table 7-2. Opcode Map

MSB LSB	Bit Manipulation		Branch	Read-Modify-Write						Control		Register/Memory							
	DIR	DIR	REL	DIR	INH	INH	IX1	SP1	IX	INH	INH	IMM	DIR	EXT	IX2	SP2	IX1	SP1	IX
	0	1	2	3	4	5	6	9E6	7	8	9	A	B	C	D	9ED	E	9EE	F
0	BRSET0 3 DIR	BSET0 2 DIR	BRA 2 REL	NEG 2 DIR	NEGA 1 INH	NEGX 1 INH	NEG 2 IX1	NEG 3 SP1	NEG 1 IX	RTI 1 INH	BGE 2 REL	SUB 2 IMM	SUB 2 DIR	SUB 3 EXT	SUB 3 IX2	SUB 4 SP2	SUB 2 IX1	SUB 3 SP1	SUB 1 IX
1	BRCLR0 3 DIR	BCLR0 2 DIR	BRN 2 REL	CBEQ 3 DIR	CBEQA 3 IMM	CBEQX 3 IMM	CBEQ 3 IX1+	CBEQ 4 SP1	CBEQ 2 IX+	RTS 1 INH	BLT 2 REL	CMP 2 IMM	CMP 2 DIR	CMP 3 EXT	CMP 3 IX2	CMP 4 SP2	CMP 2 IX1	CMP 3 SP1	CMP 1 IX
2	BRSET1 3 DIR	BSET1 2 DIR	BHI 2 REL		MUL 1 INH	DIV 1 INH	NSA 1 INH		DAA 1 INH		BGT 2 REL	SBC 2 IMM	SBC 2 DIR	SBC 3 EXT	SBC 3 IX2	SBC 4 SP2	SBC 2 IX1	SBC 3 SP1	SBC 1 IX
3	BRCLR1 3 DIR	BCLR1 2 DIR	BLS 2 REL	COM 2 DIR	COMA 1 INH	COMX 1 INH	COM 2 IX1	COM 3 SP1	COM 1 IX	SWI 1 INH	BLE 2 REL	CPX 2 IMM	CPX 2 DIR	CPX 3 EXT	CPX 3 IX2	CPX 4 SP2	CPX 2 IX1	CPX 3 SP1	CPX 1 IX
4	BRSET2 3 DIR	BSET2 2 DIR	BCC 2 REL	LSR 2 DIR	LSRA 1 INH	LSRX 1 INH	LSR 2 IX1	LSR 3 SP1	LSR 1 IX	TAP 1 INH	TXS 1 INH	AND 2 IMM	AND 2 DIR	AND 3 EXT	AND 3 IX2	AND 4 SP2	AND 2 IX1	AND 3 SP1	AND 1 IX
5	BRCLR2 3 DIR	BCLR2 2 DIR	BCS 2 REL	STHX 2 DIR	LDHX 3 IMM	LDHX 2 DIR	CPHX 3 IMM		CPHX 2 DIR	TPA 1 INH	TSX 1 INH	BIT 2 IMM	BIT 2 DIR	BIT 3 EXT	BIT 3 IX2	BIT 4 SP2	BIT 2 IX1	BIT 3 SP1	BIT 1 IX
6	BRSET3 3 DIR	BSET3 2 DIR	BNE 2 REL	ROR 2 DIR	RORA 1 INH	RORX 1 INH	ROR 2 IX1	ROR 3 SP1	ROR 1 IX	PULA 1 INH		LDA 2 IMM	LDA 2 DIR	LDA 3 EXT	LDA 3 IX2	LDA 4 SP2	LDA 2 IX1	LDA 3 SP1	LDA 1 IX
7	BRCLR3 3 DIR	BCLR3 2 DIR	BEQ 2 REL	ASR 2 DIR	ASRA 1 INH	ASRX 1 INH	ASR 2 IX1	ASR 3 SP1	ASR 1 IX	PSHA 1 INH	TAX 1 INH	AIS 2 IMM	STA 2 DIR	STA 3 EXT	STA 3 IX2	STA 4 SP2	STA 2 IX1	STA 3 SP1	STA 1 IX
8	BRSET4 3 DIR	BSET4 2 DIR	BHCC 2 REL	LSL 2 DIR	LSLA 1 INH	LSLX 1 INH	LSL 2 IX1	LSL 3 SP1	LSL 1 IX	PULX 1 INH	CLC 1 INH	EOR 2 IMM	EOR 2 DIR	EOR 3 EXT	EOR 3 IX2	EOR 4 SP2	EOR 2 IX1	EOR 3 SP1	EOR 1 IX
9	BRCLR4 3 DIR	BCLR4 2 DIR	BHCS 2 REL	ROL 2 DIR	ROLA 1 INH	ROLX 1 INH	ROL 2 IX1	ROL 3 SP1	ROL 1 IX	PSHX 1 INH	SEC 1 INH	ADC 2 IMM	ADC 2 DIR	ADC 3 EXT	ADC 3 IX2	ADC 4 SP2	ADC 2 IX1	ADC 3 SP1	ADC 1 IX
A	BRSET5 3 DIR	BSET5 2 DIR	BPL 2 REL	DEC 2 DIR	DECA 1 INH	DECX 1 INH	DEC 2 IX1	DEC 3 SP1	DEC 1 IX	PULH 1 INH	CLI 1 INH	ORA 2 IMM	ORA 2 DIR	ORA 3 EXT	ORA 3 IX2	ORA 4 SP2	ORA 2 IX1	ORA 3 SP1	ORA 1 IX
B	BRCLR5 3 DIR	BCLR5 2 DIR	BMI 2 REL	DBNZ 3 DIR	DBNZA 2 INH	DBNZX 2 INH	DBNZ 3 IX1	DBNZ 4 SP1	DBNZ 2 IX	PSHH 1 INH	SEI 1 INH	ADD 2 IMM	ADD 2 DIR	ADD 3 EXT	ADD 3 IX2	ADD 4 SP2	ADD 2 IX1	ADD 3 SP1	ADD 1 IX
C	BRSET6 3 DIR	BSET6 2 DIR	BMC 2 REL	INC 2 DIR	INCA 1 INH	INCX 1 INH	INC 2 IX1	INC 3 SP1	INC 1 IX	CLRH 1 INH	RSP 1 INH		JMP 2 DIR	JMP 3 EXT	JMP 3 IX2		JMP 2 IX1		JMP 1 IX
D	BRCLR6 3 DIR	BCLR6 2 DIR	BMS 2 REL	TST 2 DIR	TSTA 1 INH	TSTX 1 INH	TST 2 IX1	TST 3 SP1	TST 1 IX		NOP 1 INH	BSR 2 REL	JSR 2 DIR	JSR 3 EXT	JSR 3 IX2		JSR 2 IX1		JSR 1 IX
E	BRSET7 3 DIR	BSET7 2 DIR	BIL 2 REL		MOV 3 DD	MOV 2 DIX+	MOV 3 IMD		MOV 2 IX+D	STOP 1 INH	*	LDX 2 IMM	LDX 2 DIR	LDX 3 EXT	LDX 3 IX2	LDX 4 SP2	LDX 2 IX1	LDX 3 SP1	LDX 1 IX
F	BRCLR7 3 DIR	BCLR7 2 DIR	BIH 2 REL	CLR 2 DIR	CLRA 1 INH	CLR 1 INH	CLR 2 IX1	CLR 3 SP1	CLR 1 IX	WAIT 1 INH	TXA 1 INH	AIX 2 IMM	STX 2 DIR	STX 3 EXT	STX 3 IX2	STX 4 SP2	STX 2 IX1	STX 3 SP1	STX 1 IX

INH Inherent
 IMM Immediate
 DIR Direct
 EXT Extended
 DD Direct-Direct
 IX+D Indexed-Direct
 REL Relative
 IX Indexed, No Offset
 IX1 Indexed, 8-Bit Offset
 IX2 Indexed, 16-Bit Offset
 IMM Direct-Immediate
 DIX+ Direct-Indexed
 SP1 Stack Pointer, 8-Bit Offset
 SP2 Stack Pointer, 16-Bit Offset
 IX+ Indexed, No Offset with Post Increment
 IX1+ Indexed, 1-Byte Offset with Post Increment

*Pre-byte for stack pointer indexed instructions

Low Byte of Opcode in Hexadecimal

MSB	0
LSB	5 BRSET0 3 DIR

High Byte of Opcode in Hexadecimal
 Cycles
 Opcode Mnemonic
 Number of Bytes / Addressing Mode

Keyboard Interrupt Module (KBI)

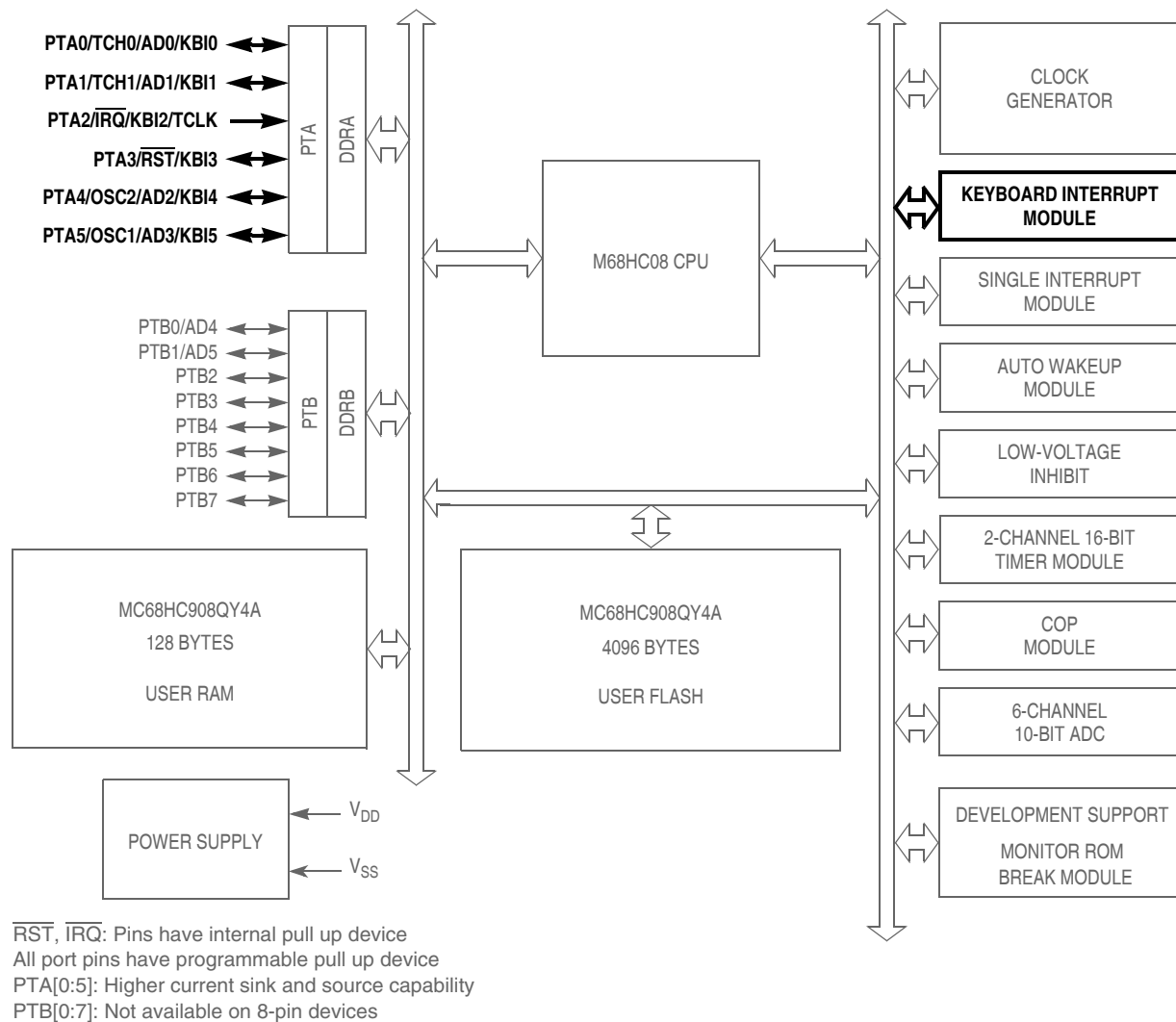


Figure 9-1. Block Diagram Highlighting KBI Block and Pins

9.3.1.1 MODEK = 1

If the MODEK bit is set, the keyboard interrupt inputs are both edge and level sensitive. The KBIPx bit will determine whether a edge sensitive pin detects rising or falling edges and on level sensitive pins whether the pin detects low or high levels. With MODEK set, both of the following actions must occur to clear a keyboard interrupt request:

- Return of all enabled keyboard interrupt inputs to a deasserted level. As long as any enabled keyboard interrupt pin is asserted, the keyboard interrupt remains active.
- Vector fetch or software clear. A KBI vector fetch generates an interrupt acknowledge signal to clear the KBI latch. Software generates the interrupt acknowledge signal by writing a 1 to ACKK in KBSCR. The ACKK bit is useful in applications that poll the keyboard interrupt inputs and require software to clear the KBI latch. Writing to ACKK prior to leaving an interrupt service routine can also prevent spurious interrupts due to noise. Setting ACKK does not affect subsequent transitions on the keyboard interrupt inputs. An edge detect that occurs after writing to ACKK latches another interrupt request. If the keyboard interrupt mask bit, IMASKK, is clear, the CPU loads the program counter with the KBI vector address.

13.6 Exception Control

Normal sequential program execution can be changed in three different ways:

1. Interrupts
 - a. Maskable hardware CPU interrupts
 - b. Non-maskable software interrupt instruction (SWI)
2. Reset
3. Break interrupts

13.6.1 Interrupts

An interrupt temporarily changes the sequence of program execution to respond to a particular event. [Figure 13-7](#) flow charts the handling of system interrupts.

Interrupts are latched, and arbitration is performed in the SIM at the start of interrupt processing. The arbitration result is a constant that the CPU uses to determine which vector to fetch. Once an interrupt is latched by the SIM, no other interrupt can take precedence, regardless of priority, until the latched interrupt is serviced (or the I bit is cleared).

At the beginning of an interrupt, the CPU saves the CPU register contents on the stack and sets the interrupt mask (I bit) to prevent additional interrupts. At the end of an interrupt, the RTI instruction recovers the CPU register contents from the stack so that normal processing can resume. [Figure 13-8](#) shows interrupt entry timing. [Figure 13-9](#) shows interrupt recovery timing.

13.6.1.1 Hardware Interrupts

A hardware interrupt does not stop the current instruction. Processing of a hardware interrupt begins after completion of the current instruction. When the current instruction is complete, the SIM checks all pending hardware interrupts. If interrupts are not masked (I bit clear in the condition code register), and if the corresponding interrupt enable bit is set, the SIM proceeds with interrupt processing; otherwise, the next instruction is fetched and executed.

If more than one interrupt is pending at the end of an instruction execution, the highest priority interrupt is serviced first. [Figure 13-10](#) demonstrates what happens when two interrupts are pending. If an interrupt is pending upon exit from the original interrupt service routine, the pending interrupt is serviced before the LDA instruction is executed.

The LDA opcode is prefetched by both the INT1 and INT2 return-from-interrupt (RTI) instructions. However, in the case of the INT1 RTI prefetch, this is a redundant operation.

NOTE

To maintain compatibility with the M6805 Family, the H register is not pushed on the stack during interrupt entry. If the interrupt service routine modifies the H register or uses the indexed addressing mode, software should save the H register and then restore it prior to exiting the routine.

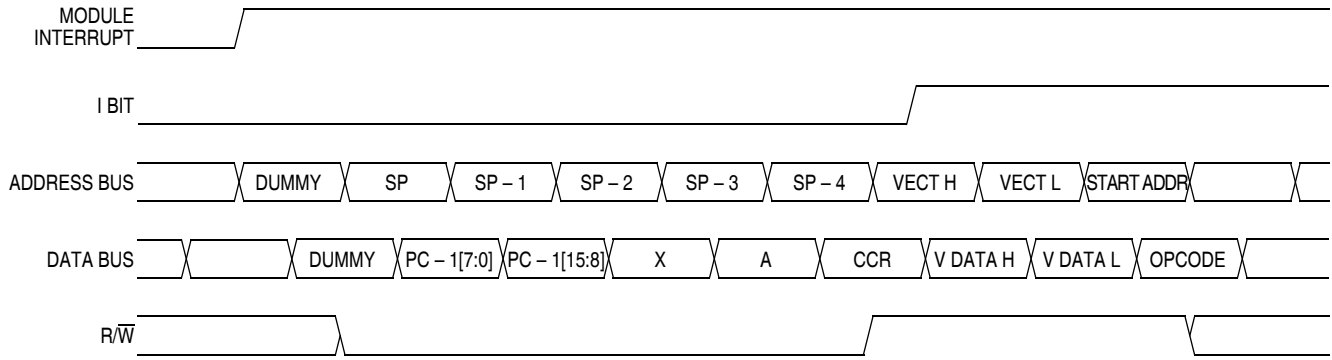


Figure 13-8. Interrupt Entry

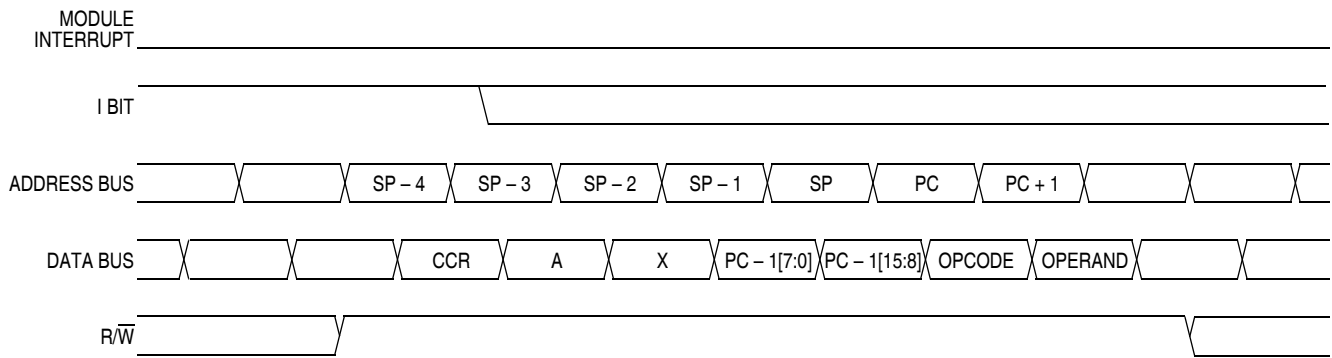


Figure 13-9. Interrupt Recovery

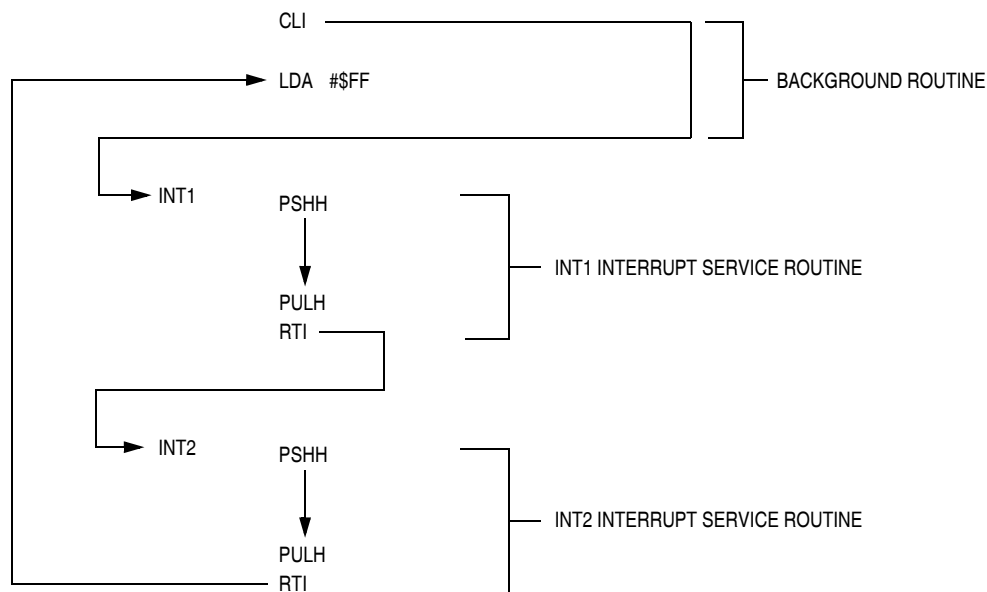


Figure 13-10. Interrupt Recognition Example

13.6.1.2 SWI Instruction

The SWI instruction is a non-maskable instruction that causes an interrupt regardless of the state of the interrupt mask (I bit) in the condition code register.

NOTE

*A software interrupt pushes PC onto the stack. A software interrupt does **not** push PC – 1, as a hardware interrupt does.*

13.6.2 Interrupt Status Registers

The flags in the interrupt status registers identify maskable interrupt sources. [Table 13-3](#) summarizes the interrupt sources and the interrupt status register flags that they set. The interrupt status registers can be useful for debugging.

Table 13-3. Interrupt Sources

Priority	Source	Flag	Mask ⁽¹⁾	INT Register Flag	Vector Address
Highest ↓ Lowest	Reset	—	—	—	\$FFFE–\$FFFF
	SWI instruction	—	—	—	\$FFFC–\$FFFD
	IRQ pin	IRQF	IMASK	IF1	\$FFFA–\$FFFB
	Timer channel 0 interrupt	CH0F	CH0IE	IF3	\$FFF6–\$FFF7
	Timer channel 1 interrupt	CH1F	CH1IE	IF4	\$FFF4–\$FFF5
	Timer overflow interrupt	TOF	TOIE	IF5	\$FFF2–\$FFF3
	Keyboard interrupt	KEYF	IMASKK	IF14	\$FFE0–\$FFE1
	ADC conversion complete interrupt	COCO	AIEN	IF15	\$FFDE–\$FFDF

1. The I bit in the condition code register is a global mask for all interrupt sources except the SWI instruction.

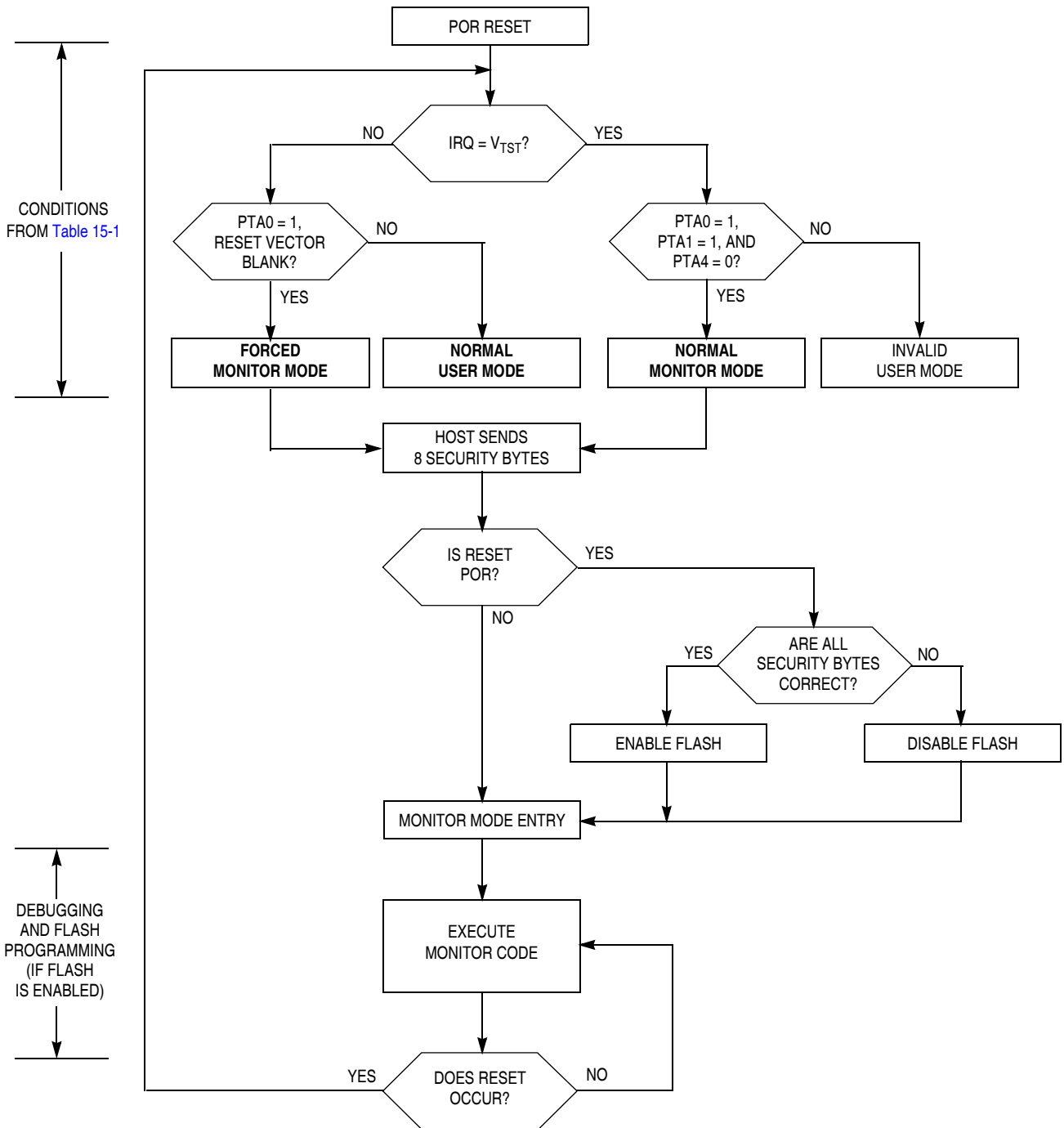


Figure 15-9. Simplified Monitor Mode Entry Flowchart

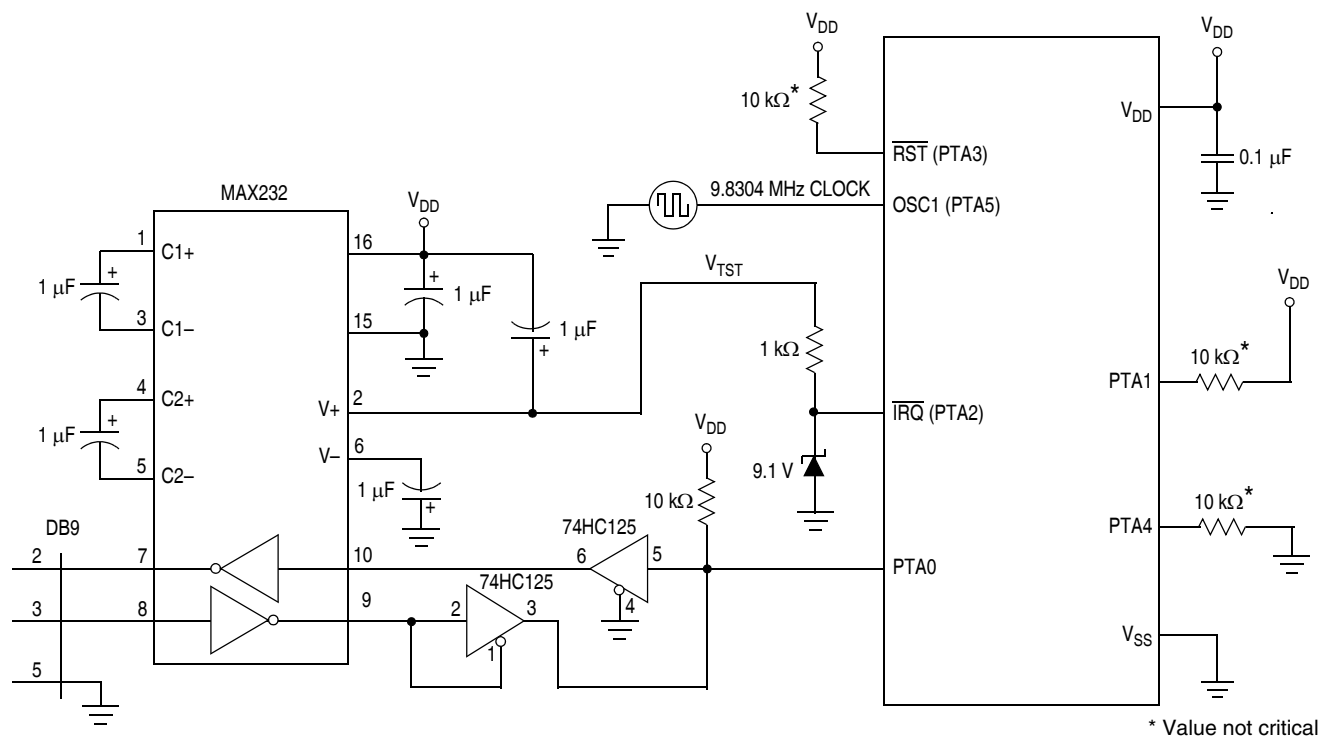


Figure 15-10. Monitor Mode Circuit (External Clock, with High Voltage)

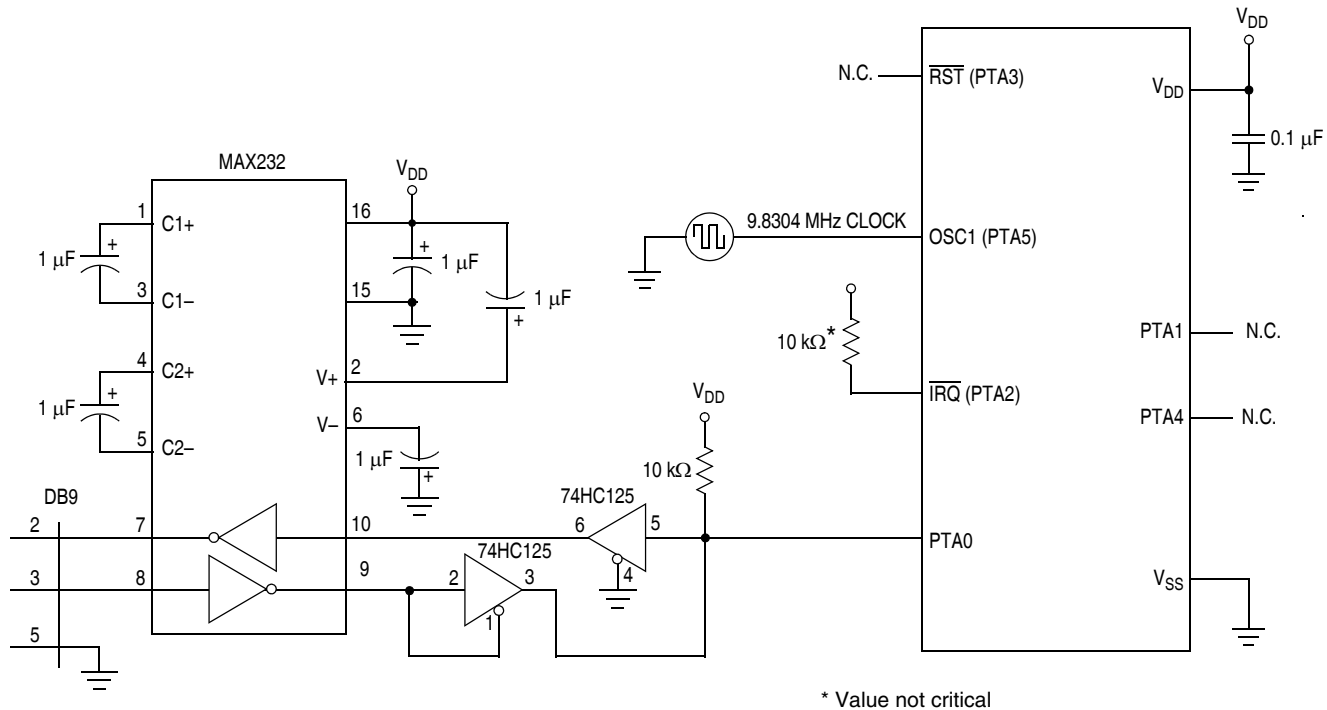


Figure 15-11. Monitor Mode Circuit (External Clock, No High Voltage)

16.3 Functional Operating Range

Characteristic	Symbol	Value	Unit	Temperature Code
Operating temperature range	T_A (T_L to T_H)	-40 to +125 -40 to +105 -40 to +85	°C	M V C
Operating voltage range	V_{DD}	2.7 to 5.5	V	—

16.4 Thermal Characteristics

Characteristic	Symbol	Value	Unit
Thermal resistance 8-pin PDIP 8-pin SOIC 16-pin PDIP 16-pin SOIC 16-pin TSSOP	θ_{JA}	105 142 76 90 133	°C/W
I/O pin power dissipation	$P_{I/O}$	User determined	W
Power dissipation ⁽¹⁾	P_D	$P_D = (I_{DD} \times V_{DD})$ $+ P_{I/O} = K/(T_J + 273^\circ\text{C})$	W
Constant ⁽²⁾	K	$P_D \times (T_A + 273^\circ\text{C})$ $+ P_D^2 \times \theta_{JA}$	W/°C
Average junction temperature	T_J	$T_A + (P_D \times \theta_{JA})$	°C
Maximum junction temperature	T_{JM}	150	°C

1. Power dissipation is a function of temperature.

2. K constant unique to the device. K can be determined for a known T_A and measured P_D . With this value of K, P_D and T_J can be determined for any value of T_A .

16.6 Typical 5-V Output Drive Characteristics

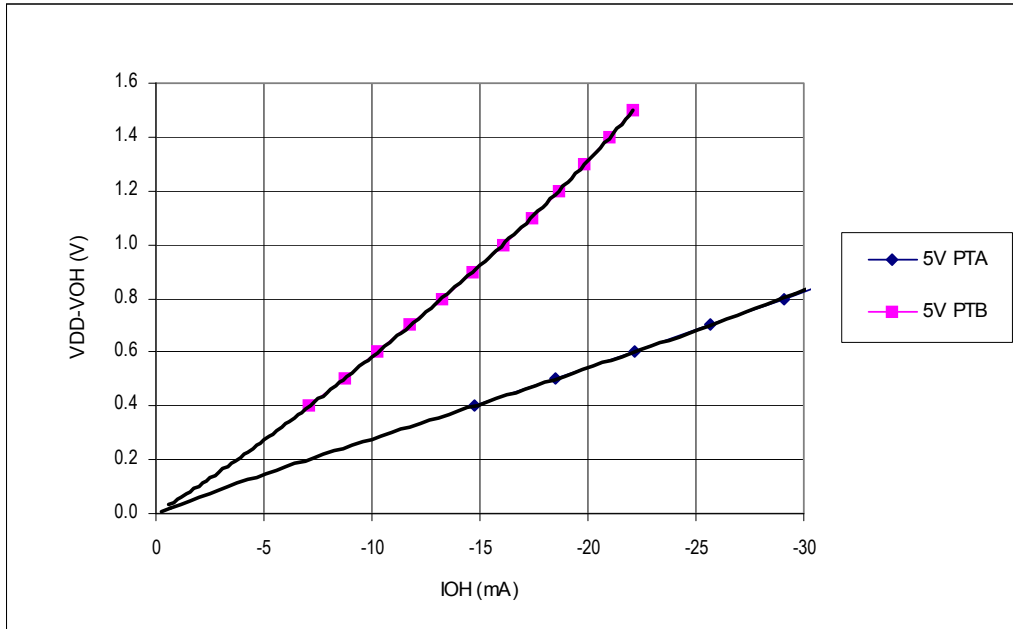


Figure 16-1. Typical 5-Volt Output High Voltage versus Output High Current (25°C)

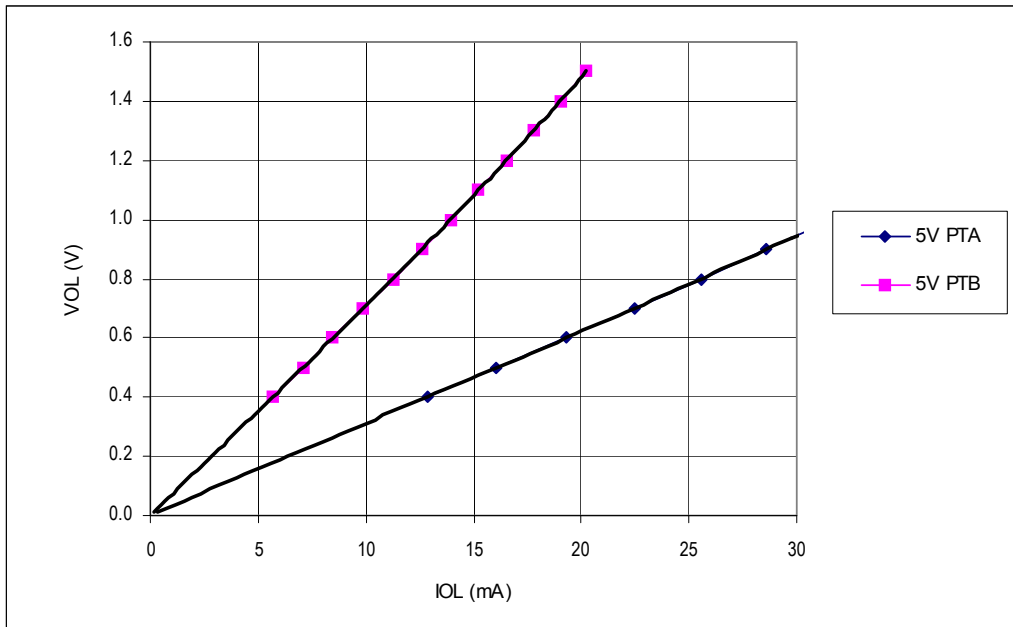


Figure 16-2. Typical 5-Volt Output Low Voltage versus Output Low Current (25°C)

16.13 ADC10 Characteristics

Characteristic	Conditions	Symbol	Min	Typ ⁽¹⁾	Max	Unit	Comment
Supply voltage	Absolute	V_{DD}	2.7	—	5.5	V	
Supply Current ADLPC = 1 ADLSMP = 1 ADCO = 1	$V_{DD} \leq 3.3$ V (3.0 V Typ)	$I_{DD}^{(2)}$	—	55	—	μ A	
	$V_{DD} \leq 5.5$ V (5.0 V Typ)		—	75	—		
Supply current ADLPC = 1 ADLSMP = 0 ADCO = 1	$V_{DD} \leq 3.3$ V (3.0 V Typ)	$I_{DD}^{(2)}$	—	120	—	μ A	
	$V_{DD} \leq 5.5$ V (5.0 V Typ)		—	175	—		
Supply current ADLPC = 0 ADLSMP = 1 ADCO = 1	$V_{DD} \leq 3.3$ V (3.0 V Typ)	$I_{DD}^{(2)}$	—	140	—	μ A	
	$V_{DD} \leq 5.5$ V (5.0 V Typ)		—	180	—		
Supply current ADLPC = 0 ADLSMP = 0 ADCO = 1	$V_{DD} \leq 3.3$ V (3.0 V Typ)	$I_{DD}^{(2)}$	—	340	—	μ A	
	$V_{DD} \leq 5.5$ V (5.0 V Typ)		—	440	615		
ADC internal clock	High speed (ADLPC = 0)	f_{ADCK}	0.40 ⁽³⁾	—	2.00	MHz	$t_{ADCK} = 1/f_{ADCK}$
	Low power (ADLPC = 1)		0.40 ⁽³⁾	—	1.00		
Conversion time ⁽⁴⁾ 10-bit Mode	Short sample (ADLSMP = 0)	t_{ADC}	19	19	21	t_{ADCK} cycles	
	Long sample (ADLSMP = 1)		39	39	41		
Conversion time ⁽⁴⁾ 8-bit Mode	Short sample (ADLSMP = 0)	t_{ADC}	16	16	18	t_{ADCK} cycles	
	Long sample (ADLSMP = 1)		36	36	38		
Sample time	Short sample (ADLSMP = 0)	t_{ADS}	4	4	4	t_{ADCK} cycles	
	Long sample (ADLSMP = 1)		24	24	24		
Input voltage		V_{ADIN}	V_{SS}	—	V_{DD}	V	
Input capacitance		C_{ADIN}	—	7	10	pF	Not tested
Input impedance		R_{ADIN}	—	5	15	k Ω	Not tested
Analog source impedance		R_{AS}	—	—	10	k Ω	External to MCU
Ideal resolution (1 LSB)	10-bit mode	RES	1.758	5	5.371	mV	$V_{REFH}/2^N$
	8-bit mode		7.031	20	21.48		
Total unadjusted error	10-bit mode	E_{TUE}	0	± 1.5	± 2.5	LSB	Includes quantization
	8-bit mode		0	± 0.7	± 1.0		
Differential non-linearity	10-bit mode	DNL	0	± 0.5	—	LSB	
	8-bit mode		0	± 0.3	—		
Monotonicity and no-missing-codes guaranteed							

— Continued on next page

Case 626 page 3 of 3



Ordering Information and Mechanical Specifications

Case 648 page 2 of 3

- The ADC that is on the QYxA can operate while the MCU is in stop mode allowing lower power operation. This also adds a lower noise environment for precise ADC results.
- Enabling an ADC channel no longer overrides the digital I/O function of the associated pin. To prevent the digital I/O from interfering with the ADC read of the pin, the data direction bit associated with the port pin must be set as input.
- Finally, the new ADC can be configured to select two different reference clock sources:
 - The internal bus x 4
 - An internal asynchronous source
 The internal asynchronous clock source allows the ADC to be clocked for operation in stop mode.

A.2.1.1 Registers Affected

	Bit 7	6	5	4	3	2	1	Bit 0
Read:	COCO	AIEN	ADCO	ADCH4	ADCH3	ADCH2	ADCH1	ADCH0
Write:								
Reset:	0	0	0	1	1	1	1	1

= Unimplemented

Figure A-1. ADC10 Status and Control Register (ADSCR)

The ADCHx bits can be used to select additional ADC channels or bandgap measurement.

	Bit 7	6	5	4	3	2	1	Bit 0
Read:	0	0	0	0	0	0	AD9	AD8
Write:								
Reset:	0	0	0	0	0	0	0	0

= Unimplemented

Figure A-2. ADC10 Data Register High (ADRH), 10-Bit Mode

10-bit ADC uses the new ADRH register for the upper 2 bits.

	Bit 7	6	5	4	3	2	1	Bit 0
Read:	ADLPC	ADIV1	ADIV0	ADICLK	MODE1	MODE0	ADLSMP	ACLKEN
Write:								
Reset:	0	0	0	0	0	0	0	0

Figure A-3. ADC10 Clock Register (ADCLK)

A long sample time option has been added to conserve power at the expense of longer conversion times. This option is selected using the new ADLSMP bit in the ADCLK register. (The bit location was previously reserved.)

The ADC will now run in stop mode if the ACLKEN bit is set to enable the asynchronous clock inside the ADC module. Utilizing stop mode for an ADC conversion gives the quietest operating mode to get extremely accurate ADC readings. (This bit location now used by ACLKEN was reserved — it always read as a 0 and writes to that location had no affect.)

A.2.5 Keyboard Interface Module (KBI) Functionality

The KBI module for the QYxA has the added capability of:

- Triggering a KBI interrupt on the rising or falling edge of an input while the QYx Classic has the capability of triggering on falling edges only.
 - A new register (Keyboard Interrupt Polarity Register) determines the polarity of KBI and the default state of this register configures the QYxA for triggering on falling edges to be compatible with QYx Classic.
 - The QYxA now has pull down resistors for the input pins that are configured for rising edge operation.

A.2.5.1 Registers Affected

	Bit 7	6	5	4	3	2	1	Bit 0
Read:	0	0	KBIP5	KBIP4	KBIP3	KBIP2	KBIP1	KBIP0
Write:								
Reset:	0	0	0	0	0	0	0	0


 = Unimplemented

Figure A-6. Keyboard Interrupt Polarity Register (KBIPR)

The KBIPR allows the selection of polarity, if any of these bits are set the corresponding interrupt pin will be configured for rising edge and a pulldown resistor will be added to the pin.

A.2.6 On-Chip Routine Enhancements

Enhancements have been made to the on-chip routines that are used for FLASH as EEPROM. Refer to AN2346 for information about using FLASH as EEPROM.

- A new mass erase routine requires a valid FLASH address loaded into the H:X register to perform an erase. This added step helps ensure that the erase routine is not inadvertently used to cause an unwanted erase. Also, on-chip FLASH programming routine ERARNGE variable CTRLBYT requires \$00 for page erase and \$40 for mass erase. The entire control byte must be set for proper operation.
- Separate routines will allow easy access to perform software SCI (Serial Communications Interface). For information on how to use on-chip FLASH programming routines refer to AN2635.
- Finally, there is improved security and robustness. The latest Monitor ROM implements updated security checks to make the program memory more secure.

A.3 Conversion Considerations

Enhancements lead to slight differences in operation from QYx Classic to the QYxA. There are a few points that should be considered in the conversion process.

- The Monitor ROM changed from 2 K to 1 K in size. This has led to the limitation that programming across page boundaries is no longer supported by the on-chip program range routine. Also, in very rare cases, ROM code improvements could cause customers to have to modify a few instructions in their application code. For example, when performing a mass erase, a valid address is required instead of an unspecified address.
- The QYxA contains new modules like the 10-bit ADC and OSC. In rare cases, new modules could cause customers to have to modify a few instructions in their application code. For example, if ADC code was written so that entire registers are configured without respect to reserve bits, then the ADC code will need to be revised to work correctly on the QYxA.
- The Reference Clock for ADC conversions has changed from the bus clock to the system clock (Bus Clock * 4). A change to the divide register may be necessary to set the reference clock to a specified value.

A.4 Code Changes Checklist

Below is a checklist that should be reviewed in the conversion process. This checklist will point out all the issues that should be addressed as your code is ported.

1. Does the original software use Auxiliary ROM routines (for example, Getbyte, Putbyte, delnus)?
If so, the software will have to be changed to handle new Auxiliary ROM routines, addresses of these routines have changed in QYxA. Code will have to be changed to use the proper addresses.
2. Does the software use FLASH as EEPROM?
If so, there are several possible issues for the page erase and mass erase routine. Software will have to be checked to ensure that proper procedure is used and the CTRLBYT is set with a MOV instruction not a BSET. Also, on-chip FLASH programming routines can no longer program across row boundaries
3. Does the code use the auto wake up timer and does the application depend on the typical auto wake time out?
Since the timeout has been improved for QYxA it may be necessary to modify software to compensate for the change in timeout.
4. Bits changed in the OSCSC, CONFIG2, and ADC registers?
Any code that writes to these registers should be reviewed to ensure that the writes are not affecting the changed bits
5. Does the code use external OSC, crystal, or RC?
If so, since the OSCOPT bits have changed locations code will have to be updated to update these bits in their proper locations.
6. Does the code use the ADC?
If so, because on QYxA the ADC clock is driven from 4XBUSCLK instead of BUSCLK changes to the ADC clock divider bits may be needed to maintain proper operation.