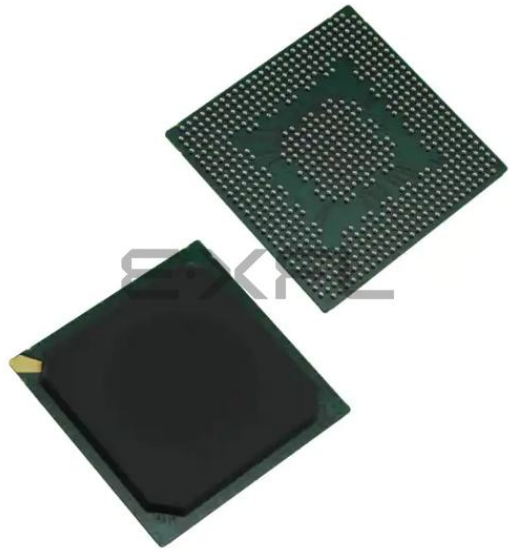

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Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Obsolete
Core Processor	PowerPC e300c3
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	400MHz
Co-Processors/DSP	Security; SEC 3.3
RAM Controllers	DDR, DDR2
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100/1000Mbps (2)
SATA	-
USB	USB 2.0 + PHY (1)
Voltage - I/O	1.8V, 2.5V, 3.3V
Operating Temperature	-40°C ~ 105°C (TA)
Security Features	Cryptography, Random Number Generator
Package / Case	620-BBGA Exposed Pad
Supplier Device Package	620-HBGA (29x29)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/kmpc8314ecvragda

  - Combines a True Random Number Generator (TRNG) and a NIST-approved Pseudo-Random Number Generator (PRNG) (as described in Annex C of FIPS140-2 and ANSI X9.62).
- Cyclical Redundancy Check Hardware Accelerator (CRCA)
  - Implements CRC32C as required for iSCSI header and payload checksums, CRC32 as required for IEEE 802 packets, as well as for programmable 32 bit CRC polynomials

2.4 DDR Memory Controller

The DDR1/DDR2 memory controller includes the following features:

- Single 16- or 32-bit interface supporting both DDR1 and DDR2 SDRAM
- Support for up to 266 MHz data rate
- Support for two physical banks (chip selects), each bank independently addressable
- 64-Mbit to 2-Gbit (for DDR1) and to 4-Gbit (for DDR2) devices with x8/x16 data ports (no direct x4 support)
- Support for one 16-bit device or two 8-bit devices on a 16-bit bus or two 16-bit devices on a 32-bit bus
- Support for up to 16 simultaneous open pages
- Supports auto refresh
- On-the-fly power management using CKE
- 1.8-/2.5-V SSTL2 compatible I/O

2.5 PCI Controller

The PCI controller includes the following features:

- Designed to comply with *PCI Local Bus Specification Revision 2.3*
- Single 32-bit data PCI interface operates at up to 66 MHz
- PCI 3.3-V compatible (not 5-V compatible)
- Support for host and agent modes
- On-chip arbitration, supporting three external masters on PCI
- Selectable hardware-enforced coherency

2.6 TDM Interface

The TDM interface includes the following features:

- Independent receive and transmit with dedicated data, clock and frame sync line
- Separate or shared RCK and TCK whose source can be either internal or external
- Glueless interface to E1/T1 frames and MVIP, SCAS, and H.110 buses
- Up to 128 time slots, where each slot can be programmed to be active or inactive
- 8- or 16-bit word widths
- The TDM Transmitter Sync Signal (TFS), Transmitter Clock Signal (TCK) and Receiver Clock

2.13 DMA Controller, I^{2}C, DUART, Enhanced Local Bus Controller (eLBC), and Timers

The integrated four-channel DMA controller includes the following features:

- Allows chaining (both extended and direct) through local memory-mapped chain descriptors (accessible by local masters)
- Misaligned transfer capability for source/destination address
- Supports external DREQ, DACK and DONE signals

There is one I^{2}C controller. This synchronous, multi-master buses can be connected to additional devices for expansion and system development.

The DUART supports full-duplex operation and is compatible with the PC16450 and PC16550 programming models. 16-byte FIFOs are supported for both the transmitter and the receiver.

The eLBC port allows connections with a wide variety of external DSPs and ASICs. Three separate state machines share the same external pins and can be programmed separately to access different types of devices. The general-purpose chip select machine (GPCM) controls accesses to asynchronous devices using a simple handshake protocol. The three user programmable machines (UPMs) can be programmed to interface to synchronous devices or custom ASIC interfaces. Each chip select can be configured so that the associated chip interface can be controlled by the GPCM or UPM controller. Both may exist in the same system. The local bus can operate at up to 66 MHz.

The system timers include the following features: periodic interrupt timer, real time clock, software watchdog timer, and two general-purpose timer blocks.

3 Electrical Characteristics

This section provides the AC and DC electrical specifications and thermal characteristics for the MPC8314E, which is currently targeted to these specifications. Some of these specifications are independent of the I/O cell, but they are included for complete reference. These are not purely I/O buffer design specifications.

3.1 Overall DC Electrical Characteristics

This section covers the ratings, conditions, and other characteristics.

3.1.1 Absolute Maximum Ratings

This table provides the absolute maximum ratings.

Table 1. Absolute Maximum Ratings [1]

Characteristic	Symbol	Max Value	Unit	Note
Core supply voltage	VDD	–0.3 to 1.26	V	—
PLL supply voltage	AVDD	–0.3 to 1.26	V	—
DDR1 DRAM I/O supply voltage	GVDD	–0.3 to 2.7	V	—

5.1 DC Electrical Characteristics

This table provides the clock input (SYS_CLK_IN/PCI_SYNC_IN) DC timing specifications for the MPC8314E.

Table 6. SYS_CLK_IN DC Electrical Characteristics

Parameter	Condition	Symbol	Min	Max	Unit
Input high voltage	—	V_{IH}	2.4	NVDD + 0.3	V
Input low voltage	—	V_{IL}	-0.3	0.4	V
SYS_CLK_IN input current	0 V ≤V_{IN} ≤ NVDD	I_{IN}	—	±10	μA
SYS_XTAL_IN input current	0 V ≤V_{IN} ≤ NVDD	I_{IN}	—	±40	μA
PCI_SYNC_IN input current	0 V ≤V_{IN} ≤ NVDD	I_{IN}	—	±10	μA
RTC_CLK input current	0 V ≤V_{IN} ≤ NVDD	I_{IN}	—	±10	μA
USB_CLK_IN input current	0 V ≤ V_{IN} ≤ NVDD	I_{IN}	—	±10	μA
USB_XTAL_IN input current	0 V ≤V_{IN} ≤ NVDD	I_{IN}	—	±40	μA

5.2 AC Electrical Characteristics

The primary clock source for the MPC8314E can be one of two inputs, SYS_CLK_IN or PCI_CLK, depending on whether the device is configured in PCI host or PCI agent mode. This table provides the clock input (SYS_CLK_IN/PCI_CLK) AC timing specifications for the MPC8314E.

Table 7. SYS_CLK_IN AC Timing Specifications

Parameter/Condition	Symbol	Min	Typical	Max	Unit	Note
SYS_CLK_IN/PCI_CLK frequency	$f_{SYS\_CLK\_IN}$	24	—	66.67	MHz	1, 6, 7
SYS_CLK_IN/PCI_CLK cycle time	$t_{SYS\_CLK\_IN}$	15	—	41.6	ns	6
SYS_CLK_IN rise and fall time	t_{KH}, t_{KL}	0.6	—	4	ns	2, 6
PCI_CLK rise and fall time	t_{PCH}, t_{PCL}	0.6	0.8	1.2	ns	2
SYS_CLK_IN/PCI_CLK duty cycle	$t_{KHK}/t_{SYS\_CLK\_IN}$	40	—	60	%	3, 6
SYS_CLK_IN/PCI_CLK jitter	—	—	—	±150	ps	4, 5, 6

Note:

1. **Caution:** The system, core, and security block must not exceed their respective maximum or minimum operating frequencies.
2. Rise and fall times for SYS_CLK_IN/PCI_CLK are specified at 20% to 80% of signal swing.
3. Timing is guaranteed by design and characterization.
4. This represents the total input jitter—short term and long term—and is guaranteed by design.
5. The SYS_CLK_IN/PCI_CLK driver's closed loop jitter bandwidth should be <500 kHz at –20 dB. The bandwidth must be set low to allow cascade-connected PLL-based devices to track SYS_CLK_IN drivers with the specified jitter.
6. The parameter names PCI_CLK and PCI_SYNC_IN are used interchangeably in this document.
7. Spread spectrum is allowed up to 1% down-spread at 33kHz.(max. rate).

Table 9. RESET Initialization Timing Specifications (continued)

Note:
1. $t_{PCI\_SYNC\_IN}$ is the clock period of the input clock applied to PCI_SYNC_IN. When the device is In PCI host mode the primary clock is applied to the SYS_CLK_IN input, and PCI_SYNC_IN period depends on the value of CFG_SYS_CLKIN_DIV.
2. $t_{SYS\_CLK\_IN}$ is the clock period of the input clock applied to SYS_CLK_IN. It is only valid when the device is in PCI host mode.
3. POR configuration signals consists of CFG_RESET_SOURCE[0:3] and CFG_SYS_CLKIN_DIV.
4. The parameter names CFG_SYS_CLKIN_DIV and CFG_CLKIN_DIV are used interchangeably in this document.

This table provides the PLL lock times.

Table 10. PLL Lock Times

Parameter/Condition	Min	Max	Unit	Note
System PLL lock times	—	100	μs	—
e300 core PLL lock times	—	100	μs	—
SerDes (SGMII/PCI Exp Phy) PLL lock times	—	100	μs	—
USB phy PLL lock times	—	100	μs	—

7 DDR and DDR2 SDRAM

This section describes the DC and AC electrical specifications for the DDR SDRAM interface of the MPC8314E. Note that DDR SDRAM is GVDD(typ) = 2.5 V and DDR2 SDRAM is GVDD(typ) = 1.8 V.

7.1 DDR and DDR2 SDRAM DC Electrical Characteristics

This table provides the recommended operating conditions for the DDR2 SDRAM component(s) of the MPC8314E when GVDD(typ) = 1.8 V.

Table 11. DDR2 SDRAM DC Electrical Characteristics for GVDD(typ) = 1.8 V

Parameter/Condition	Symbol	Min	Max	Unit	Note
I/O supply voltage	GVDD	1.7	1.9	V	1
I/O reference voltage	MVREF	0.49 × GVDD	0.51 × GVDD	V	2
I/O termination voltage	V_{TT}	MVREF – 0.04	MVREF + 0.04	V	3
Input high voltage	V_{IH}	MVREF+ 0.125	GVDD + 0.3	V	—
Input low voltage	V_{IL}	–0.3	MVREF – 0.125	V	—
Output leakage current	I_{OZ}	–9.9	9.9	μA	4
Output high current (V_{OUT} = 1.420 V, GVDD= 1.7V)	I_{OH}	–13.4	—	mA	—
Output low current (V_{OUT} = 0.280 V)	I_{OL}	13.4	—	mA	

Note:
1. GVDD is expected to be within 50 mV of the DRAM GVDD at all times.
2. MVREF is expected to be equal to 0.5 × GVDD, and to track GVDD DC variations as measured at the receiver. Peak-to-peak noise on MVREF may not exceed ±2% of the DC value.
3. V_{TT} is not applied directly to the device. It is the supply to which far end signal termination is made and is expected to be equal to MVREF. This rail should track variations in the DC level of MVREF.
4. Output leakage is measured with all outputs disabled, 0 V ≤ V_{OUT} ≤ GVDD.

Table 20. DDR and DDR2 SDRAM Output AC Timing Specifications (continued)

At recommended operating conditions

Parameter	Symbol [1]	Min	Max	Unit	Note
$\overline{\text{MCS}}$[n] output hold with respect to MCK 266 MHz 200 MHz	t_{DDKHCX}	 3.15 4.20	 — —	ns	3
MCK to MDQS Skew	t_{DDKHMH}	–0.6	0.6	ns	4
MDQ//MDM output setup with respect to MDQS 266 MHz 200 MHz	t_{DDKHDS}, t_{DDKLDS}	 900 1000	 — —	ps	5
MDQ//MDM output hold with respect to MDQS 266 MHz 200 MHz	t_{DDKHDX}, t_{DDKLDX}	 1100 1200	 — —	ps	5
MDQS preamble start	t_{DDKHMP}	$-0.5 \times t_{MCK} - 0.6$	$-0.5 \times t_{MCK} + 0.6$	ns	6
MDQS epilogue end	t_{DDKHME}	–0.6	0.6	ns	6

Note:

1. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state) (reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. Output hold time can be read as DDR timing (DD) from the rising or falling edge of the reference clock (KH or KL) until the output went invalid (AX or DX). For example, t_{DDKHAS} symbolizes DDR timing (DD) for the time t_{MCK} memory clock reference (K) goes from the high (H) state until outputs (A) are setup (S) or output valid time. Also, t_{DDKLDX} symbolizes DDR timing (DD) for the time t_{MCK} memory clock reference (K) goes low (L) until data outputs (D) are invalid (X) or data output hold time.
2. All MCK/$\overline{\text{MCK}}$ referenced measurements are made from the crossing of the two signals ±0.1 V.
3. ADDR/CMD includes all DDR SDRAM output signals except MCK/$\overline{\text{MCK}}$, $\overline{\text{MCS}}$, and MDQ//MDM/MDQS.
4. Note that t_{DDKHMH} follows the symbol conventions described in note 1. For example, t_{DDKHMH} describes the DDR timing (DD) from the rising edge of the MCK[n] clock (KH) until the MDQS signal is valid (MH). t_{DDKHMH} can be modified through control of the DQSS override bits in the TIMING_CFG_2 register. This is typically set to the same delay as the clock adjust in the CLK_CNTL register. The timing parameters listed in the table assume that these 2 parameters have been set to the same adjustment value. See the *MPC8315E PowerQUICC II Pro Integrated Host Processor Family Reference Manual* for a description and understanding of the timing modifications enabled by use of these bits.
5. Determined by maximum possible skew between a data strobe (MDQS) and any corresponding bit of data (MDQ), ECC (), or data mask (MDM). The data strobe should be centered inside of the data eye at the pins of the microprocessor.
6. All outputs are referenced to the rising edge of MCK[n] at the pins of the microprocessor. Note that t_{DDKHMP} follows the symbol conventions described in note 1.

Table 29. RGMII and RTBI AC Timing Specifications (continued)

At recommended operating conditions (see Table 2)

Parameter/Condition	Symbol [1]	Min	Typ	Max	Unit
Clock cycle duration [3]	t_{RGT}	7.2	8.0	8.8	ns
Duty cycle for 1000Base-T [4, 5]	t_{RGTH}/t_{RGT}	45	50	55	%
Duty cycle for 10BASE-T and 100BASE-TX [3, 5]	t_{RGTH}/t_{RGT}	40	50	60	%
Rise time (20%–80%)	t_{RGTR}	—	—	0.75	ns
Fall time (20%–80%)	t_{RGTF}	—	—	0.75	ns
GTX_CLK125 reference clock period	t_{G12} [6]	—	8.0	—	ns
GTX_CLK125 reference clock duty cycle	t_{G125H}/t_{G125}	47	—	53	%

Note:

1. Note that, in general, the clock reference symbol representation for this section is based on the symbols RGT to represent RGMII and RTBI timing. For example, the subscript of t_{RGT} represents the RTBI (T) receive (RX) clock. Note also that the notation for rise (R) and fall (F) times follows the clock symbol that is being represented. For symbols representing skews, the subscript is skew (SK) followed by the clock that is being skewed (RGT).
2. This implies that PC board design requires clocks to be routed so that an additional trace delay of greater than 1.5 ns is added to the associated clock signal.
3. For 10 and 100 Mbps, t_{RGT} scales to 400 ns ± 40 ns and 40 ns ± 4 ns, respectively.
4. Duty cycle may be stretched/shrunk during speed changes or while transitioning to a received packet's clock domains as long as the minimum duty cycle is not violated and stretching occurs for no more than three t_{RGT} of the lowest speed transitioned between.
5. Duty cycle reference is LVDD/2.
6. This symbol is used to represent the external GTX_CLK125 and does not follow the original symbol naming convention. GTX_CLK supply voltage is fixed at 3.3V inside the chip. If PHY supplies a 2.5 V Clock signal on this input, set TSCOMOBI bit of System I/O configuration register (SICRH) as 1. See the *MPC8315E PowerQUICC II Pro Integrated Host Processor Family Reference Manual*.
7. The frequency of RX_CLK should not exceed the TX_CLK by more than 300 ppm

Table 30. MII Management DC Electrical Characteristics Powered at 3.3 V (continued)

Parameter	Symbol	Conditions		Min	Max	Unit
Input low current	I_{IL}	NVDD = Max	V_{IN} = 0.5 V	–600	—	μA

Note:
1. The symbol V_{IN}, in this case, represents the NV_{IN} symbol referenced in Table 1 and Table 2.

9.3.2 MII Management AC Electrical Specifications

This table provides the MII management AC timing specifications.

Table 31. MII Management AC Timing Specifications

At recommended operating conditions with NVDD is 3.3 V ± 300 mv

Parameter/Condition	Symbol [1]	Min	Typ	Max	Unit	Note
MDC frequency	f_{MDC}	—	2.5	—	MHz	2
MDC period	t_{MDC}	—	400	—	ns	—
MDC clock pulse width high	t_{MDCH}	32	—	—	ns	—
MDC to MDIO delay	t_{MDKHDX}	10	—	170	ns	3
MDIO to MDC setup time	t_{MDDVKH}	5	—	—	ns	—
MDIO to MDC hold time	t_{MDDXKH}	0	—	—	ns	—
MDC rise time	t_{MDCR}	—	—	10	ns	—
MDC fall time	t_{MDHF}	—	—	10	ns	—

Note:
1. The symbols used for timing specifications herein follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{MDKHDX} symbolizes management data timing (MD) for the time t_{MDC} from clock reference (K) high (H) until data outputs (D) are invalid (X) or data hold time. Also, t_{MDDVKH} symbolizes management data timing (MD) with respect to the time data input signals (D) reach the valid state (V) relative to the t_{MDC} clock reference (K) going to the high (H) state or setup time. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).
2. This parameter is dependent on the csb_clk speed (that is, for a csb_clk of 133 MHz, the maximum frequency is 4.16 MHz and the minimum frequency is 0.593 MHz).
3. This parameter is dependent on the csb_clk speed (that is, for a csb_clk of 133 MHz, the delay is 60 ns).

Table 33. 1588 Timer AC Specifications (continued)

Parameter	Symbol	Min	Max	Unit	Note

Note:
1. The timer can operate on rtc_clock or tmr_clock. These clocks get muxed and any one of them can be selected.
2. Asynchronous signals.
3. Inputs need to be stable at least one TMR clock.

9.5 SGMII Interface Electrical Characteristics

Each SGMII port features a 4-wire AC-Coupled serial link from the dedicated SerDes interface of MPC8315E as shown in Figure 17, where C_{TX} is the external (on board) AC-Coupled capacitor. Each output pin of the SerDes transmitter differential pair features 50-Ω output impedance. Each input of the SerDes receiver differential pair features 50-Ω on-die termination to XCOREVSS. The reference circuit of the SerDes transmitter and receiver is shown in Figure 48.

When an eTSEC port is configured to operate in SGMII mode, the parallel interface's output signals of this eTSEC port can be left floating. The input signals should be terminated based on the guidelines described in Section 25.4, "Connection Recommendations," as long as such termination does not violate the desired POR configuration requirement on these pins, if applicable.

When operating in SGMII mode, the TSEC_GTX_CLK125 clock is not required for this port. Instead, SerDes reference clock is required on SD_REF_CLK and $\overline{\text{SD\_REF\_CLK}}$ pins.

9.5.1 DC Requirements for SGMII SD_REF_CLK and $\overline{\text{SD\_REF\_CLK}}$

The characteristics and DC requirements of the separate SerDes reference clock are described in Section 15, "High-Speed Serial Interfaces (HSSI)."

9.5.2 AC Requirements for SGMII SD_REF_CLK and $\overline{\text{SD\_REF\_CLK}}$

This table lists the SGMII SerDes reference clock AC requirements. Please note that SD_REF_CLK and $\overline{\text{SD\_REF\_CLK}}$ are not intended to be used with, and should not be clocked by, a spread spectrum clock source.

Table 34. SD_REF_CLK and $\overline{\text{SD\_REF\_CLK}}$ AC Requirements

Symbol	Parameter Description	Min	Typical	Max	Unit	Note
t_{REF}	REFCLK cycle time	—	8	—	ns	—
t_{REFCJ}	REFCLK cycle-to-cycle jitter. Difference in the period of any two adjacent REFCLK cycles	—	—	100	ps	—
t_{REFPJ}	Phase jitter. Deviation in edge location with respect to mean edge location	–50	—	50	ps	—

9.5.3 SGMII Transmitter and Receiver DC Electrical Characteristics

Table 35 and Table 36 describe the SGMII SerDes transmitter and receiver AC-coupled DC electrical characteristics. Transmitter DC characteristics are measured at the transmitter outputs (SD_TX[n] and $\overline{\text{SD\_TX[n]}}$) as depicted in Figure 16.

16.4.3 Differential Receiver (RX) Input Specifications

This table defines the specifications for the differential input at all receivers (RXs). The parameters are specified at the component pins.

Table 55. Differential Receiver (RX) Input Specifications

Parameter	Symbol	Comments	Min	Typical	Max	Unit	Note
Unit interval	UI	Each UI is 400 ps ± 300 ppm. UI does not account for Spread Spectrum Clock dictated variations.	399.88	400	400.12	ps	1
Differential peak-to-peak output voltage	$V_{RX-DIFFp-p}$	$V_{RX-DIFFp-p} = 2*\|V_{RX-D+} - V_{RX-D-}\|$	0.175	—	1.200	V	2
Minimum receiver eye width	T_{RX-EYE}	The maximum interconnect media and Transmitter jitter that can be tolerated by the Receiver can be derived as $T_{RX-MAX-JITTER} = 1 - U_{RX-EYE}$= 0.6 UI.	0.4	—	—	UI	2, 3
Maximum time between the jitter median and maximum deviation from the median.	$T_{RX-EYE-MEDIAN-to-MAX-JITTER}$	Jitter is defined as the measurement variation of the crossing points ($V_{RX-DIFFp-p}$ = 0 V) in relation to a recovered TX UI. A recovered TX UI is calculated over 3500 consecutive unit intervals of sample data. Jitter is measured using all edges of the 250 consecutive UI in the center of the 3500 UI used for calculating the TX UI.	—	—	0.3	UI	2, 3, 7
AC peak common mode input voltage	$V_{RX-CM-ACp}$	$V_{RX-CM-ACp} = \|V_{RXD+} + V_{RXD-}\|/2 - V_{RX-CM-DC}$ $V_{RX-CM-DC} = DC_{(avg)}$ of $\|V_{RX-D+} + V_{RX-D-}\|/2$	—	—	150	mV	2
Differential return loss	$RL_{RX-DIFF}$	Measured over 50 MHz to 1.25 GHz with the D+ and D- lines biased at +300 mV and -300 mV, respectively.	15	—	—	dB	4
Common mode return loss	RL_{RX-CM}	Measured over 50 MHz to 1.25 GHz with the D+ and D- lines biased at 0 V.	6	—	—	dB	4
DC differential input impedance	$Z_{RX-DIFF-DC}$	RX DC differential mode impedance.	80	100	120	Ω	5
DC Input Impedance	Z_{RX-DC}	Required RX D+ as well as D- DC Impedance (50 ± 20% tolerance).	40	50	60	Ω	2, 5

This figure provides the AC test load for the GPIO.

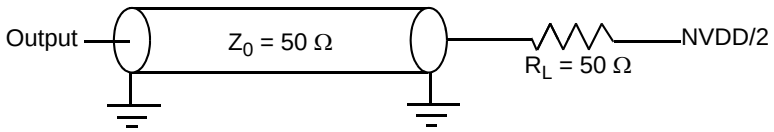


Figure 53. GPIO AC Test Load

19 IPIC

This section describes the DC and AC electrical specifications for the external interrupt pins of the MPC8314E.

19.1 IPIC DC Electrical Characteristics

This table provides the DC electrical characteristics for the external interrupt pins.

Table 60. IPIC DC Electrical Characteristics

Characteristic	Symbol	Condition	Min	Max	Unit
Input high voltage	V_{IH}	—	2.1	NVDD + 0.3	V
Input low voltage	V_{IL}	—	–0.3	0.8	V
Input current	I_{IN}	—	—	±5	μA
Output high voltage	V_{OH}	I_{OH} = –8.0 mA	2.4	—	V
Output low voltage	V_{OL}	I_{OL} = 8.0 mA	—	0.5	V
Output low voltage	V_{OL}	I_{OL} = 3.2 mA	—	0.4	V

19.2 IPIC AC Timing Specifications

This table provides the IPIC input and output AC timing specifications.

Table 61. IPIC Input AC Timing Specifications

Characteristic	Symbol [1]	Min	Unit
IPIC inputs—minimum pulse width	t_{PIWID}	20	ns

Note:

1. IPIC inputs and outputs are asynchronous to any visible clock. IPIC outputs should be synchronized before use by any external synchronous logic. IPIC inputs are required to be valid for at least t_{PIWID} ns to ensure proper operation when working in edge triggered mode.

20 SPI

This section describes the DC and AC electrical specifications for the SPI of the MPC8314E.

21.2 TDM AC Electrical Characteristics

This table provides the TDM AC timing specifications.

Table 65. TDM AC Timing specifications

Parameter/Condition	Symbol	Min	Max	Unit
TDMxRCK/TDMxTCK	t_{DM}	20.0	—	ns
TDMxRCK/TDMxTCK high pulse width	t_{DM_HIGH}	8.0	—	ns
TDMxRCK/TDMxTCK low pulse width	t_{DM_LOW}	8.0	—	ns
TDMxRCK/TDMxTCK rise time (20% to 80%)	t_{DMKH}	1.0	4.0	ns
TDMxRCK/TDMxTCK fall time (80% to 20%)	t_{DMKL}	1.0	4.0	ns
TDM all input setup time	t_{DMIVKH}	3.0	—	ns
TDMxRD hold time	$t_{DMRDIXKH}$	3.5	—	ns
TDMxTFS/TDMxRFS input hold time	$t_{DMFSIXKH}$	2.0	—	ns
TDMxTCK High to TDMxTD output active	t_{DM_OUTAC}	4.0	—	ns
TDMxTCK High to TDMxTD output valid	$t_{DMTKHOV}$	—	14.0	ns
TDMxTD hold time	$t_{DMTKHOX}$	2.0	—	ns
TDMxTCK High to TDMxTD output high impedance	t_{DM_OUTHI}	—	10.0	ns
TDMxTFS/TDMxRFS output valid	$t_{DMFSKHOV}$	—	13.5	ns
TDMxTFS/TDMxRFS output hold time	$t_{DMFSKHOX}$	2.5	—	ns

Note:

1. The symbols used for timing specifications herein follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state) (reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, $t_{TDMIVKH}$ symbolizes TDM timing (DM) with respect to the time the input signals (I) reach the valid state (V) relative to the TDM Clock, t_{TC}, reference (K) going to the high (H) state or setup time. Also, output signals (O), hold (X).
2. Output values are based on 30 pF capacitive load.
3. Inputs are referenced to the sampling that the TDM is programmed to use. Outputs are referenced to the programming edge they are programmed to use. Use of the rising edge or falling edge as a reference is programmable. TDMxTCK and TDMxRCK are shown using the rising edge.

This figure shows the TDM receive signal timing.

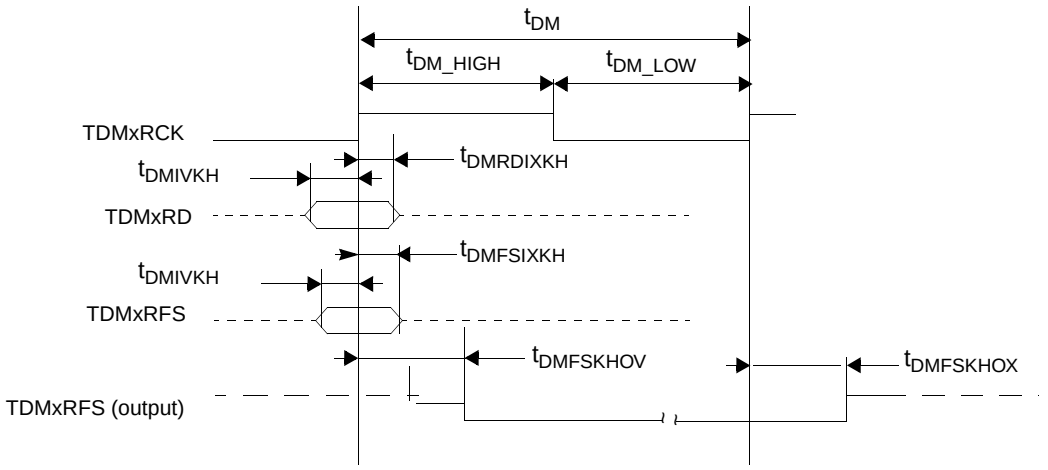
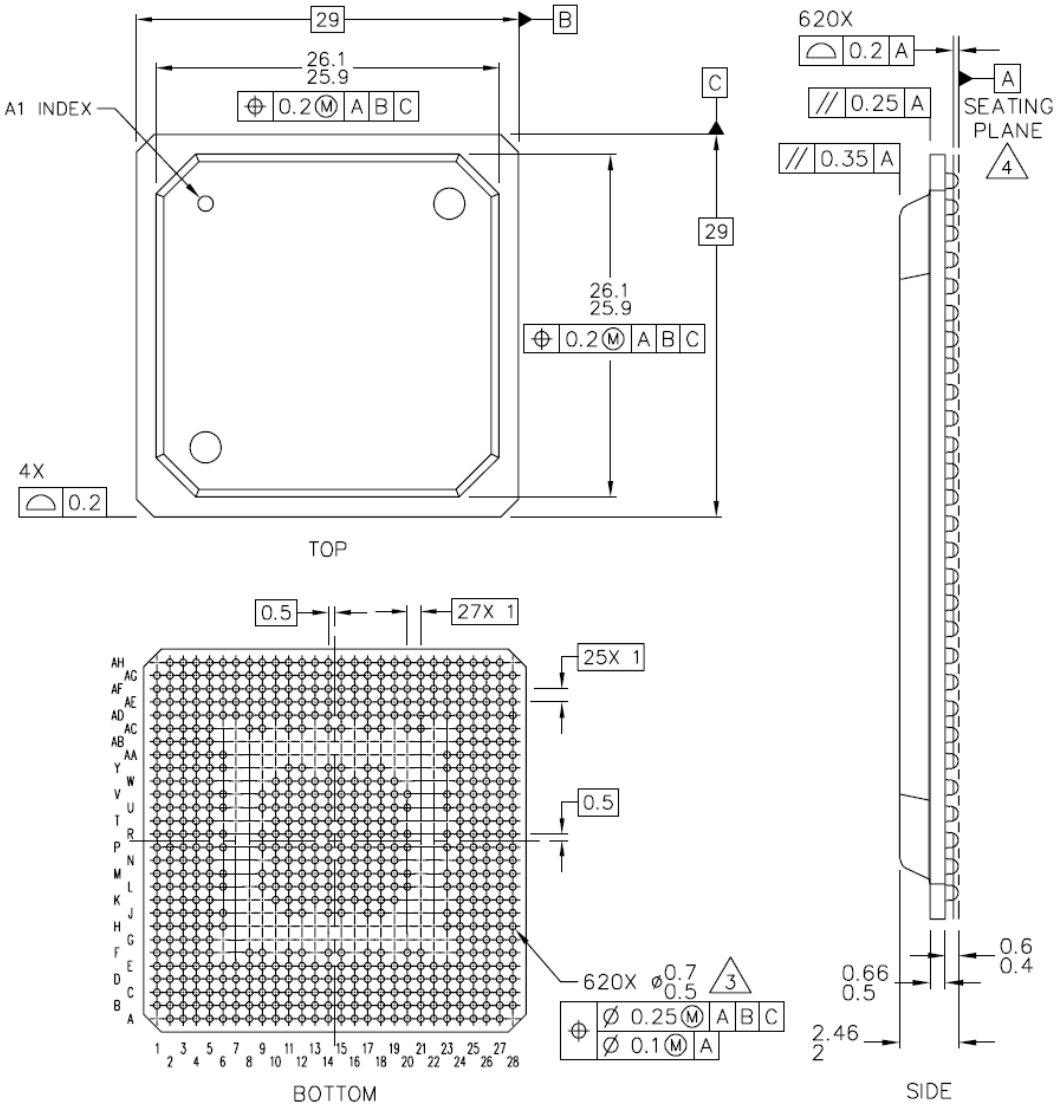


Figure 57. TDM Receive Signals

22.2 Mechanical Dimensions of the TEPBGA II

This figure shows the mechanical dimensions and bottom surface nomenclature of the 620-pin TEPBGA II package.



Notes:
1. All dimensions are in millimeters.
2. Dimensions and tolerances per ASME Y14.5M-1994.
3. Maximum solder ball diameter measured parallel to datum A.
4. Datum A, the seating plane, is determined by the spherical crowns of the solder balls.

Figure 59. Mechanical Dimensions and Bottom Surface Nomenclature of the TEPBGA II

22.3 Pinout Listings

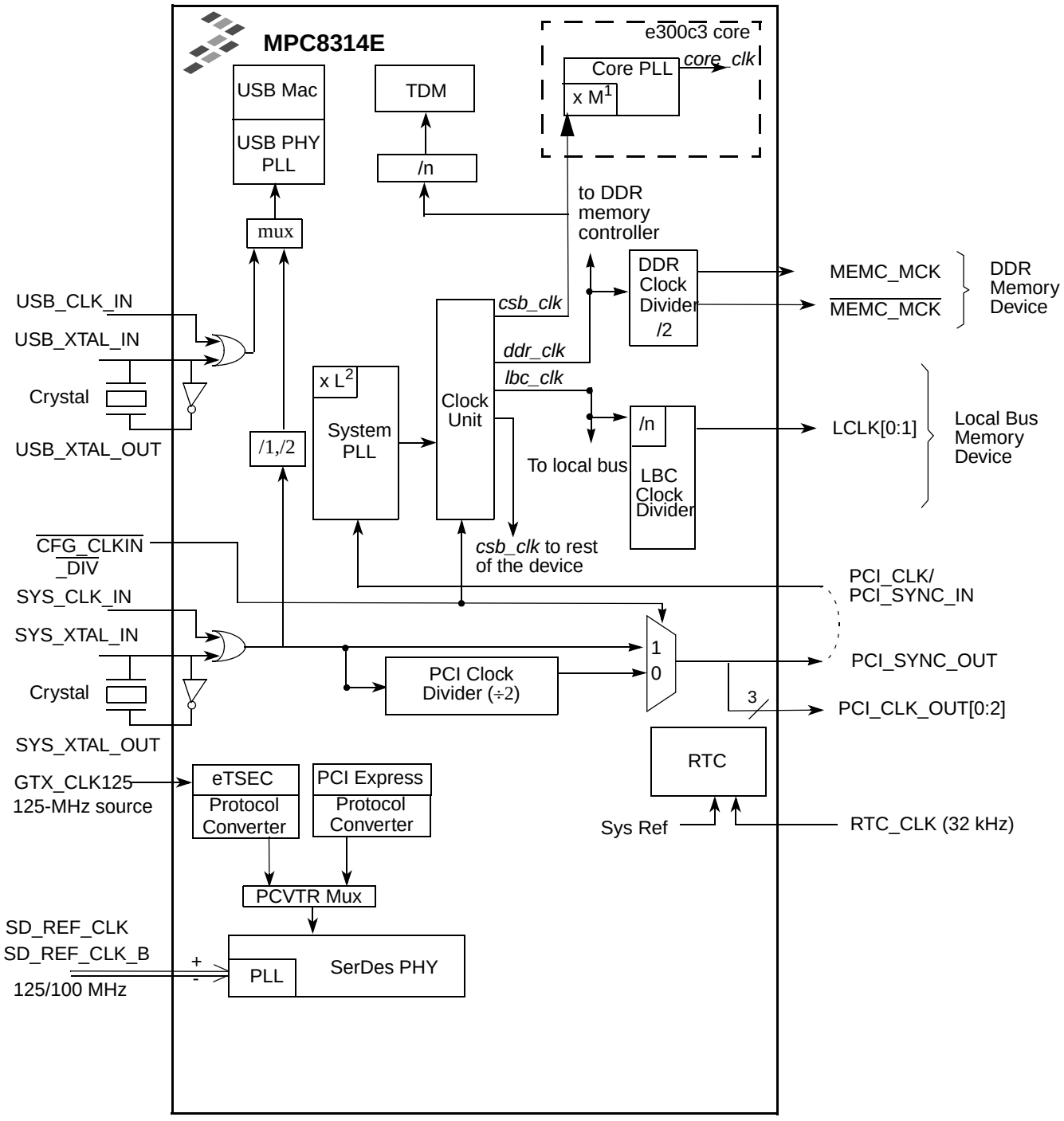
This table provides the pin-out listing for the TEPBGA II package.

Table 66. MPC8314E TEPBGA II Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Note
GPIO_1/DMA_DACK1/GTM1_TIN2/GTM2_TIN1	A4	I/O	$NVDD1_{ON}$	—
GPIO_2/DMA_DONE1/GTM1_TGATE2/GTM2_TGATE1	K3	I/O	$NVDD4_{OFF}$	—
GPIO_3/GTM1_TIN3/GTM2_TIN4	K1	I/O	$NVDD4_{OFF}$	—
GPIO_4/GTM1_TGATE3/GTM2_TGATE4	K2	I/O	$NVDD4_{OFF}$	—
GPIO_5/GTM1_TOUT3/GTM2_TOUT1	L5	I/O	$NVDD4_{OFF}$	—
GPIO_6/GTM1_TIN4/GTM2_TIN3	L3	I/O	$NVDD4_{OFF}$	—
GPIO_7/GTM1_TGATE4/GTM2_TGATE3	L1	I/O	$NVDD4_{OFF}$	—
GPIO_8/USBDR_DRIVE_VBUS/GTM1_TIN1/GTM2_TIN2	M1	I/O	$NVDD4_{OFF}$	—
GPIO_9/USBDR_PWRFAULT/GTM1_TGATE1/GTM2_TGATE2	M2	I/O	$NVDD4_{OFF}$	—
GPIO_10/USBDR_PCTL0/GTM1_TOUT2/GTM2_TOUT1	M5	I/O	$NVDD4_{OFF}$	—
GPIO_11/USBDR_PCTL1/GTM1_TOUT4/GTM2_TOUT3	M4	I/O	$NVDD4_{OFF}$	—
SPI				
SPIMOSI/GPIO_15	W3	I/O	$NVDD1_{OFF}$	—
SPIMISO/GPIO_16	W4	I/O	$NVDD1_{OFF}$	—
SPICLK	Y1	I/O	$NVDD1_{OFF}$	—
SPISEL/GPIO_17	W2	I/O	$NVDD1_{OFF}$	—
Power and Ground Supplies				
GVDD	Y11, Y12, Y14, Y15, Y17, AC8, AC11, AC14, AC17, AD6, AD9, AD17, AE8, AE13, AE19, AF10, AF15, AF21, AG2, AG3, AG8, AG13, AG19, AH2	I	—	—
$LVDD1_{OFF}$	H6, J3, L6, L9, M9	I	—	—
$LVDD2_{ON}$	C11, D9, E10, F11, J12	I	—	—
$NVDD1_{OFF}$	U9, V9, W10, Y4, Y6, AA3, AB4	I	—	—
$NVDD1_{ON}$	B1, B2, C1, D5, E7, F5, F9, J11, K10	I	—	—
$NVDD2_{OFF}$	B22, B27, C19, E16, F15, F18, F21, F25, H25, J17, J18, J23, L20, M20	I	—	—
$NVDD2_{ON}$	L26, N19	I	—	—
$NVDD3_{OFF}$	U20, V20, V23, V26, W19, Y18, Y26, AA23, AA25, AC20, AC25, AD23, AE25, AG25, AG27, T27, U27	I	—	—
$NVDD4_{OFF}$	K4, L2, M6, N10	I	—	—

23 Clocking

This figure shows the internal distribution of clocks within the MPC8314E



1 Multiplication factor M = 1, 1.5, 2, 2.5, and 3. Value is decided by RCWLR[COREPLL].

2 Multiplication factor L = 2, 3, 4 and 5. Value is decided by RCWLR[SPMF].

Figure 60. MPC8314E Clock Subsystem

Millennium Electronics (MEI) Loroco Sites 671 East Brokaw Road San Jose, CA 95112 Internet: www.mei-thermal.com	408-436-8770
Tyco Electronics Chip Coolers™ P.O. Box 3668 Harrisburg, PA 17105 Internet: www.tycoelectronics.com	800-522-6752
Wakefield Engineering 33 Bridge St. Pelham, NH 03076 Internet: www.wakefield.com	603-635-2800

Interface material vendors include the following:

Chomerics, Inc. 77 Dragon Ct. Woburn, MA 01801 Internet: www.chomerics.com	781-935-4850
Dow-Corning Corporation Corporate Center PO BOX 994 Midland, MI 48686-0994 Internet: www.dowcorning.com	800-248-2481
Shin-Etsu MicroSi, Inc. 10028 S. 51st St. Phoenix, AZ 85044 Internet: www.microsi.com	888-642-7674
The Bergquist Company 18930 West 78th St. Chanhassen, MN 55317 Internet: www.bergquistcompany.com	800-347-4572

24.3 Heat Sink Attachment

When attaching heat sinks to these devices, an interface material is required. The best method is to use thermal grease and a spring clip. The spring clip should connect to the printed circuit board, either to the board itself, to hooks soldered to the board, or to a plastic stiffener. Avoid attachment forces which would lift the edge of the package or peel the package from the board. Such peeling forces reduce the solder joint lifetime of the package. Recommended maximum force on the top of the package is 10 lb force (45 Newtons). If an adhesive attachment is planned, the adhesive should be intended for attachment to painted or plastic surfaces and its performance verified under the application requirements.

This circuit is intended to filter noise in the PLLs resonant frequency range from a 500 kHz to 10 MHz range. It should be built with surface mount capacitors with minimum Effective Series Inductance (ESL). Consistent with the recommendations of Dr. Howard Johnson in *High Speed Digital Design: A Handbook of Black Magic* (Prentice Hall, 1993), multiple small capacitors of equal value are recommended over a single large value capacitor.

Each circuit should be placed as close as possible to the specific AV_{DD} pin being supplied to minimize noise coupled from nearby circuits. It should be possible to route directly from the capacitors to the AV_{DD} pin, which is on the periphery of package, without the inductance of vias. Note that the RC filter results in lower voltage level on AVDD. This does not imply that the DC specification can be relaxed.

This figure shows the PLL power supply filter circuit.

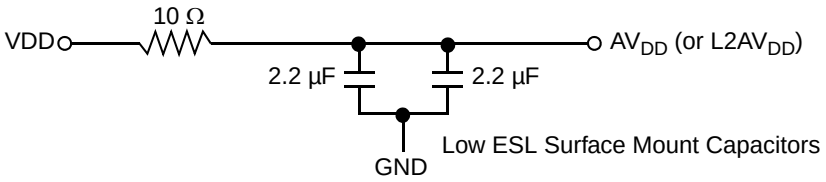


Figure 61. PLL Power Supply Filter Circuit

25.3 Decoupling Recommendations

Due to large address and data buses, and high operating frequencies, the device can generate transient power surges and high frequency noise in its power supply, especially while driving large capacitive loads. This noise must be prevented from reaching other components in the MPC8314E system, and the MPC8314E itself requires a clean, tightly regulated source of power. Therefore, it is recommended that the system designer place at least one decoupling capacitor at each VDD, NVDD, GVDD, and LVDD pins of the device. These decoupling capacitors should receive their power from separate VDD, NVDD, GVDD, LVDD, and GND power planes in the PCB, utilizing thick and short traces to minimize inductance. Capacitors may be placed directly under the device using a standard escape pattern. Others may surround the part.

These capacitors should have a value of 0.01 or 0.1 µF. Only ceramic SMT (surface mount technology) capacitors should be used to minimize lead inductance, preferably 0402 or 0603 sizes.

In addition, it is recommended that there be several bulk storage capacitors distributed around the PCB, feeding the VDD, NVDD, GVDD, and LVDD planes, to enable quick recharging of the smaller chip capacitors. These bulk capacitors should have a low ESR (equivalent series resistance) rating to ensure the quick response time necessary. They should also be connected to the power and ground planes through two vias to minimize inductance. Suggested bulk capacitors—100–330 µF (AVX TPS tantalum or Sanyo OSCON).

25.4 Connection Recommendations

To ensure reliable operation, it is highly recommended to connect unused inputs to an appropriate signal level. Unused active low inputs should be tied to NVDD, GVDD, or LVDD as required. Unused active high inputs should be connected to GND. All NC (no-connect) signals must remain unconnected.

Power and ground connections must be made to all external VDD, GVDD, LVDD, NVDD, and GND pins of the device.

25.5 Output Buffer DC Impedance

The MPC8314E drivers are characterized over process, voltage, and temperature. For all buses, the driver is a push-pull single-ended driver type (open drain for I^2C).

To measure Z_0 for the single-ended drivers, an external resistor is connected from the chip pad to NVDD or GND. Then, the value of each resistor is varied until the pad voltage is NVDD/2 (see Figure 62). The output impedance is the average of two components, the resistances of the pull-up and pull-down devices. When data is held high, SW1 is closed (SW2 is open) and R_P is trimmed until the voltage at the pad equals NVDD/2. R_P then becomes the resistance of the pull-up devices. R_P and R_N are designed to be close to each other in value. Then, $Z_0 = (R_P + R_N)/2$.

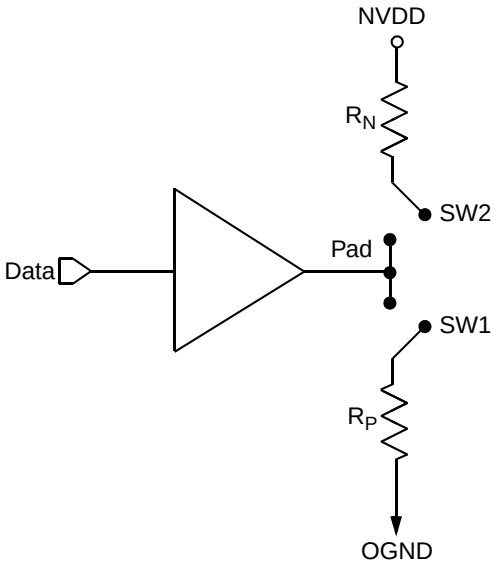


Figure 62. Driver Impedance Measurement

The value of this resistance and the strength of the driver's current source can be found by making two measurements. First, the output voltage is measured while driving logic 1 without an external differential termination resistor. The measured voltage is $V_1 = R_{source} \times I_{source}$. Second, the output voltage is measured while driving logic 1 with an external precision differential termination resistor of value R_{term}. The measured voltage is $V_2 = (1/(1/R_1 + 1/R_2)) \times I_{source}$. Solving for the output impedance gives $R_{source} = R_{term} \times (V_1/V_2 - 1)$. The drive current is then $I_{source} = V_1/R_{source}$.

This table summarizes the signal impedance targets. The driver impedance are targeted at minimum VDD, nominal NVDD, 105°C.

Table 76. Impedance Characteristics

Impedance	Local Bus, Ethernet, DUART, Control, Configuration, Power Management	PCI Signals (not including PCI Output Clocks)	PCI Output Clocks (including PCI_SYNC_OUT)	DDR DRAM	Symbol	Unit
R_N	42 Target	25 Target	42 Target	20 Target	Z_0	Ω
R_P	42 Target	25 Target	42 Target	20 Target	Z_0	Ω
Differential	NA	NA	NA	NA	Z_{DIFF}	Ω

Note: Nominal supply voltages. See Table 1, T_j = 105°C.

25.6 Configuration Pin Multiplexing

The MPC8314E provides the user with power-on configuration options that can be set through the use of external pull-up or pull-down resistors of 4.7 kΩ on certain output pins (see customer visible configuration pins). These pins are generally used as output only pins in normal operation.

While $\overline{\text{HRESET}}$ is asserted however, these pins are treated as inputs. The value presented on these pins while $\overline{\text{HRESET}}$ is asserted, is latched when $\overline{\text{PORESET}}$ deasserts, at which time the input receiver is disabled and the I/O circuit takes on its normal function. Careful board layout with stubless connections to these pull-up/pull-down resistors coupled with the large value of the pull-up/pull-down resistor should minimize the disruption of signal quality or speed for output pins thus configured.

25.7 Pull-Up Resistor Requirements

The MPC8314E requires high resistance pull-up resistors (10 kΩ is recommended) on open drain type pins including I^2C pins and EPIC interrupt pins.

For more information on required pull up resistors and the connections required for JTAG interface, see AN3438, MPC8315 Design Checklist

26 Ordering Information

Ordering information for the parts fully covered by this specification document is provided in Section 26.1, “Part Numbers Fully Addressed by this Document.”

26.1 Part Numbers Fully Addressed by this Document

This table provides the Freescale part numbering nomenclature for the MPC8314E. Note that the individual part numbers correspond to a maximum processor core frequency. For available frequencies, contact your local Freescale sales office. In addition to the processor frequency, the part numbering scheme

also includes an application modifier which may specify special application conditions. Each part number also contains a revision code which refers to the die mask revision number.

Table 77. Part Numbering Nomenclature

MPC	8314	E	C	VR	AG	D	A
Product Code	**Part Identifier**	**Encryption Acceleration**	**Temperature Range [3]**	**Package [1]**	**e300 Core Frequency [2]**	**DDR Frequency**	**Revision Level**
MPC	8314	Blank = Not included E = included	Blank = 0 to 105°C C = –40 to 105°C	VR= Pb Free TEPBGA II	AD = 266 MHz AF = 333 MHz AG = 400 MHz	D = 266 MHz	Contact local Freescale sales office

Note:
1. See Section 22, “Package and Pin Listings,” for more information on available package types.
2. Processor core frequencies supported by parts addressed by this specification only. Not all parts described in this specification support all core frequencies. Additionally, parts addressed by electric may support other maximum core frequencies.
3. Contact your local Freescale field applications engineer (FAE).

This table shows the SVR settings by device and package type.

Table 78. SVR Settings

Device	Package	SVR (Rev 1.0)	SVR (Rev 1.1)	SVR (Rev 1.2)
MPC8314E	TEPBGA II	0x80B6_0010	0x80B6_0011	0x80B6_0012
MPC8314	TEPBGA II	0x80B7_0010	0x80B7_0011	0x80B7_0012

Note:
1. PVR = 8085_0020 for all devices and revisions in this table.

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