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What is "[Embedded - Microcontrollers](#)"?

"[Embedded - Microcontrollers](#)" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "[Embedded - Microcontrollers](#)"

Details

Product Status	Obsolete
Core Processor	HC11
Core Size	8-Bit
Speed	2MHz
Connectivity	SCI, SPI
Peripherals	POR, WDT
Number of I/O	38
Program Memory Size	12KB (12K x 8)
Program Memory Type	OTP
EEPROM Size	512 x 8
RAM Size	512 x 8
Voltage - Supply (Vcc/Vdd)	4.5V ~ 5.5V
Data Converters	A/D 8x8b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 105°C (TA)
Mounting Type	Surface Mount
Package / Case	52-LCC (J-Lead)
Supplier Device Package	52-PLCC (19.1x19.1)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mc68hc711e9vfne2

Operating Modes and On-Chip Memory

Addr.	Register Name		Bit 7	6	5	4	3	2	1	Bit 0	
\$1031	Analog-to-Digital Results Register 1 (ADR1) See page 64.	Read:	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
		Write:									
		Reset:	Indeterminate after reset								
\$1032	Analog-to-Digital Results Register 2 (ADR2) See page 64.	Read:	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
		Write:									
		Reset:	Indeterminate after reset								
\$1033	Analog-to-Digital Results Register 3 (ADR3) See page 64.	Read:	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
		Write:									
		Reset:	Indeterminate after reset								
\$1034	Analog-to-Digital Results Register 4 (ADR4) See page 64.	Read:	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
		Write:									
		Reset:	Indeterminate after reset								
\$1035	Block Protect Register (BPROT) See page 52.	Read:				PTCON	BPRT3	BPRT2	BPRT1	BPRT0	
		Write:									
		Reset:	0	0	0	1	1	1	1	1	
\$1036	EPROM Programming Control Register (EPROG) ⁽¹⁾ See page 53.	Read:	MBE		ELAT	EXCOL	EXROW	T1	T0	PGM	
		Write:									
		Reset:	0	0	0	0	0	0	0	0	
\$1037	Reserved		R	R	R	R	R	R	R	R	
1. MC68HC711E20 only											
\$1038	Reserved		R	R	R	R	R	R	R	R	
\$1039	System Configuration Options Register (OPTION) See page 46.	Read:	ADPU	CSEL	IRQE ⁽¹⁾	DLY ⁽¹⁾	CME		CR1 ⁽¹⁾	CR0 ⁽¹⁾	
		Write:									
		Reset:	0	0	0	1	0	0	0	0	
\$103A	Arm/Reset COP Timer Circuitry Register (COPRST) See page 81.	Read:	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
		Write:									
		Reset:	0	0	0	0	0	0	0	0	
\$103B	EPROM and EEPROM Programming Control Register (PPROG) See page 49.	Read:	ODD	EVEN	ELAT ⁽²⁾	BYTE	ROW	ERASE	EELAT	EPGM	
		Write:									
		Reset:	0	0	0	0	0	0	0	0	
\$103C	Highest Priority I Bit Interrupt and Miscellaneous Register (HPRIO) See page 41.	Read:	RBOOT	SMOD	MDA	IRV(NE)	PSEL3	PSEL2	PSEL1	PSEL0	
		Write:									
		Reset:	0	0	0	0	0	1	1	0	
\$103D	RAM and I/O Mapping Register (INIT) See page 45.	Read:	RAM3	RAM2	RAM1	RAM0	REG3	REG2	REG1	REG0	
		Write:									
		Reset:	0	0	0	0	0	0	0	1	

= Unimplemented
R = Reserved
 U = Unaffected
 I = Indeterminate after reset

Figure 2-7. Register and Control Bit Assignments (Sheet 5 of 6)

Addr.	Register Name	Bit 7	6	5	4	3	2	1	Bit 0
\$103E	Reserved	R	R	R	R	R	R	R	R
\$103F	System Configuration Register (CONFIG) See page 43.	Read:				NOSEC	NOCOP	ROMON	EEON
		Write:							
		Reset:	0	0	0	0	U	U	1
\$103F	System Configuration Register (CONFIG) ⁽³⁾ See page 43.	Read:	EE3	EE2	EE1	EE0	NOSEC	NOCOP	EEON
		Write:							
		Reset:	1	1	1	1	U	U	1

1. Can be written only once in first 64 cycles out of reset in normal modes or at any time during special modes.

2. MC68HC711E9 only

3. MC68HC811E2 only

= Unimplemented
R = Reserved U = Unaffected
 I = Indeterminate after reset

Figure 2-7. Register and Control Bit Assignments (Sheet 6 of 6)

2.3.1 RAM and Input/Output Mapping

Hardware priority is built into RAM and I/O mapping. Registers have priority over RAM and RAM has priority over ROM. When a lower priority resource is mapped at the same location as a higher priority resource, a read/write of a location results in a read/write of the higher priority resource only. For example, if both the register block and the RAM are mapped to the same location, only the register block will be accessed. If RAM and ROM are located at the same position, RAM has priority.

The fully static RAM can be used to store instructions, variables, and temporary data. The direct addressing mode can access RAM locations using a 1-byte address operand, saving program memory space and execution time, depending on the application.

RAM contents can be preserved during periods of processor inactivity by two methods, both of which reduce power consumption. They are:

1. In the software-based stop mode, the clocks are stopped while V_{DD} powers the MCU. Because power supply current is directly related to operating frequency in CMOS integrated circuits, only a very small amount of leakage exists when the clocks are stopped.
2. In the second method, the MODB/ V_{STBY} pin can supply RAM power from a battery backup or from a second power supply. [Figure 2-8](#) shows a typical standby voltage circuit for a standard 5-volt device. Adjustments to the circuit must be made for devices that operate at lower voltages. Using the MODB/ V_{STBY} pin may require external hardware, but can be justified when a significant amount of external circuitry is operating from V_{DD} . If V_{STBY} is used to maintain RAM contents, reset must be held low whenever V_{DD} is below normal operating level. Refer to [Chapter 5 Resets and Interrupts](#).

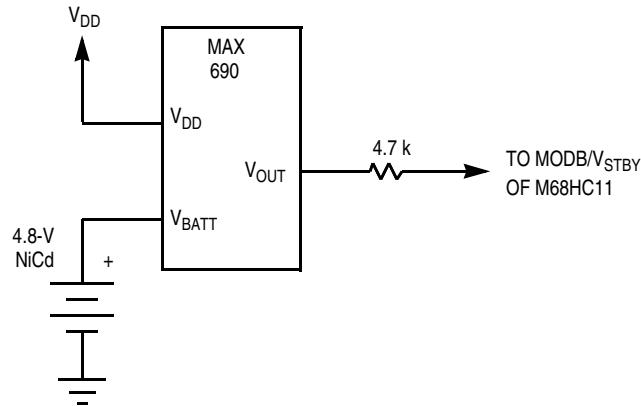


Figure 2-8. RAM Standby MODB/V_{STBY} Connections

The bootloader program is contained in the internal bootstrap ROM. This ROM, which appears as internal memory space at locations \$BF00–\$BFFF, is enabled only if the MCU is reset in special bootstrap mode.

In expanded modes, the ROM/EPROM/OTPROM (if present) is enabled out of reset and located at the top of the memory map if the ROMON bit in the CONFIG register is set. ROM or EPROM is enabled out of reset in single-chip and bootstrap modes, regardless of the state of ROMON.

For devices with 512 bytes of EEPROM, the EEPROM is located at \$B600–\$B7FF and has the same read cycle time as the internal ROM. The 512 bytes of EEPROM cannot be remapped to other locations.

For the MC68HC811E2, EEPROM is located at \$F800–\$FFFF and can be remapped to any 4-Kbyte boundary. EEPROM mapping control bits (EE[3:0] in CONFIG) determine the location of the 2048 bytes of EEPROM and are present only on the MC68HC811E2. Refer to [2.3.3.1 System Configuration Register](#) for a description of the MC68HC811E2 CONFIG register.

EEPROM can be programmed or erased by software and an on-chip charge pump, allowing EEPROM changes using the single V_{DD} supply.

2.3.2 Mode Selection

The four mode variations are selected by the logic states of the MODA and MODB pins during reset. The MODA and MODB logic levels determine the logic state of SMOD and the MDA control bits in the highest priority I-bit interrupt and miscellaneous (HPRIO) register.

After reset is released, the mode select pins no longer influence the MCU operating mode. In single-chip operating mode, the MODA pin is connected to a logic level 0. In expanded mode, MODA is normally connected to V_{DD} through a pullup resistor of 4.7 kΩ. The MODA pin also functions as the load instruction register $\overline{\text{LIR}}$ pin when the MCU is not in reset. The open-drain active low $\overline{\text{LIR}}$ output pin drives low during the first E cycle of each instruction. The MODB pin also functions as standby power input (V_{STBY}), which allows RAM contents to be maintained in absence of V_{DD}.

Refer to [Table 2-1](#), which is a summary of mode pin operation, the mode control bits, and the four operating modes.

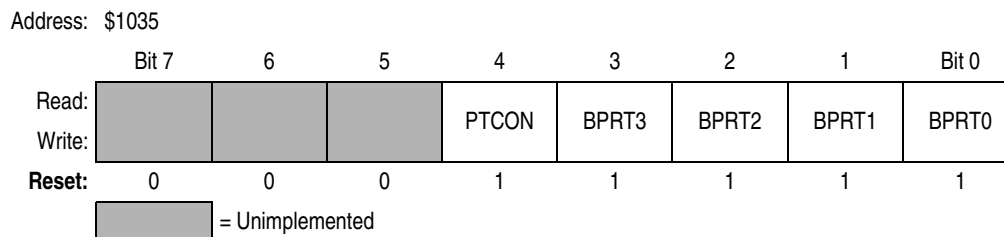


Figure 2-16. Block Protect Register (BPROT)

Bits [7:5] — Unimplemented

Always read 0

PTCON — Protect CONFIG Register Bit

0 = CONFIG register can be programmed or erased normally.

1 = CONFIG register cannot be programmed or erased.

BPRT[3:0] — Block Protect Bits for EEPROM

When set, these bits protect a block of EEPROM from being programmed or electronically erased.

Ultraviolet light, however, can erase the entire EEPROM contents regardless of BPRT[3:0] (windowed packages only). Refer to [Table 2-6](#) and [Table 2-7](#).

When cleared, BPRT[3:0] allow programming and erasure of the associated block.

Table 2-6. EEPROM Block Protect

Bit Name	Block Protected	Block Size
BPRT0	\$B600–\$B61F	32 bytes
BPRT1	\$B620–\$B65F	64 bytes
BPRT2	\$B660–\$B6DF	128 bytes
BPRT3	\$B6E0–\$B7FF	288 bytes

Table 2-7. EEPROM Block Protect in MC68HC811E2 MCUs

Bit Name	Block Protected	Block Size
BPRT0	\$x800–\$x9FF ⁽¹⁾	512 bytes
BPRT1	\$xA00–\$xBFF ⁽¹⁾	512 bytes
BPRT2	\$xC00–\$xDFF ⁽¹⁾	512 bytes
BPRT3	\$xE00–\$xFF ⁽¹⁾	512 bytes

1. x is determined by the value of EE[3:0] in CONFIG register. Refer to [Figure 2-13](#).

3.4 Conversion Process

The A/D conversion sequence begins one E-clock cycle after a write to the A/D control/status register, ADCTL. The bits in ADCTL select the channel and the mode of conversion.

An input voltage equal to V_{RL} converts to \$00 and an input voltage equal to V_{RH} converts to \$FF (full scale), with no overflow indication. For ratiometric conversions of this type, the source of each analog input should use V_{RH} as the supply voltage and be referenced to V_{RL} .

3.5 Channel Assignments

The multiplexer allows the A/D converter to select one of 16 analog signals. Eight of these channels correspond to port E input lines to the MCU, four of the channels are internal reference points or test functions, and four channels are reserved. Refer to [Table 3-1](#).

Table 3-1. Converter Channel Assignments

Channel Number	Channel Signal	Result in ADRx if MULT = 1
1	AN0	ADR1
2	AN1	ADR2
3	AN2	ADR3
4	AN3	ADR4
5	AN4	ADR1
6	AN5	ADR2
7	AN6	ADR3
8	AN7	ADR4
9 – 12	Reserved	—
13	$V_{RH}^{(1)}$	ADR1
14	$V_{RL}^{(1)}$	ADR2
15	$(V_{RH})/2^{(1)}$	ADR3
16	Reserved ⁽¹⁾	ADR4

1. Used for factory testing

3.6 Single-Channel Operation

The two types of single-channel operation are:

- When SCAN = 0, the single selected channel is converted four consecutive times. The first result is stored in A/D result register 1 (ADR1), and the fourth result is stored in ADR4. After the fourth conversion is complete, all conversion activity is halted until a new conversion command is written to the ADCTL register.
- When SCAN = 1, conversions continue to be performed on the selected channel with the fifth conversion being stored in register ADR1 (overwriting the first conversion result), the sixth conversion overwriting ADR2, and so on.

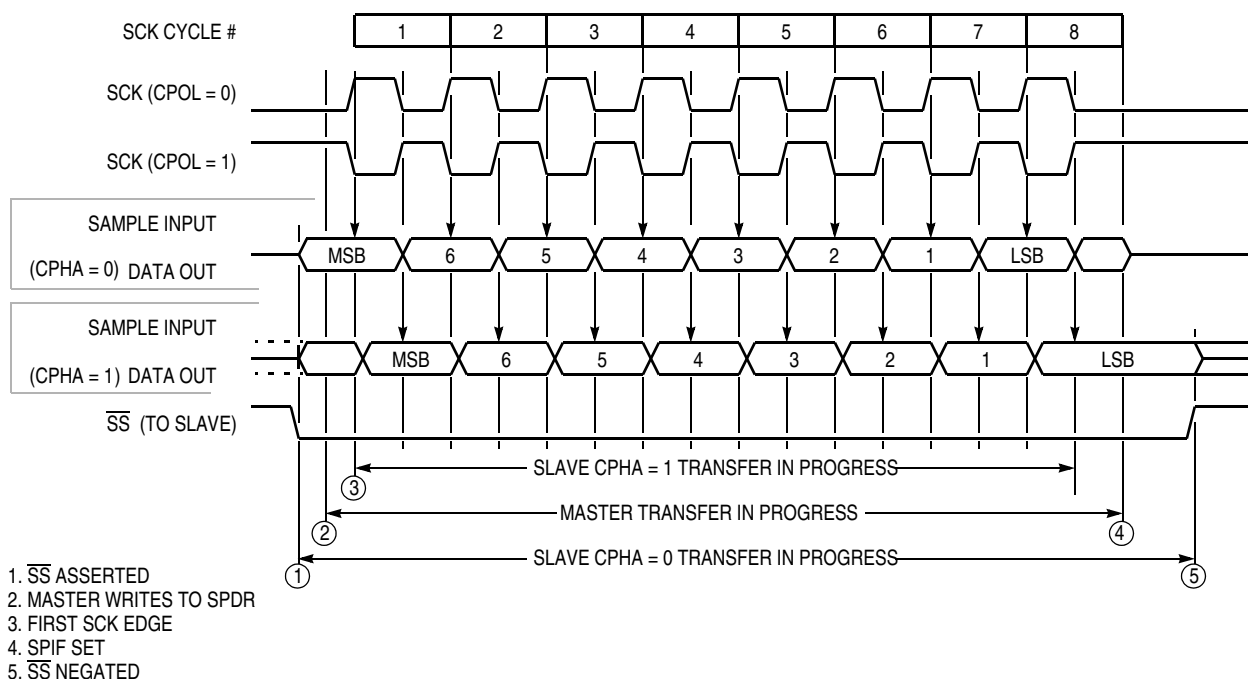


Figure 8-2. SPI Transfer Format

8.5 SPI Signals

This subsection contains descriptions of the four SPI signals:

- Master in/slave out (MISO)
- Master out/slave in (MOSI)
- Serial clock (SCK)
- Slave select (\overline{SS})

Any SPI output line must have its corresponding data direction bit in DDRD register set. If the DDR bit is clear, that line is disconnected from the SPI logic and becomes a general-purpose input. All SPI input lines are forced to act as inputs regardless of the state of the corresponding DDR bits in DDRD register.

8.5.1 Master In/Slave Out

MISO is one of two unidirectional serial data signals. It is an input to a master device and an output from a slave device. The MISO line of a slave device is placed in the high-impedance state if the slave device is not selected.

8.5.2 Master Out/Slave In

The MOSI line is the second of the two unidirectional serial data signals. It is an output from a master device and an input to a slave device. The master device places data on the MOSI line a half-cycle before the clock edge that the slave device uses to latch the data.

9.5.2 Timer Interrupt Flag Register 2

Bits of this register indicate the occurrence of timer system events. Coupled with the four high-order bits of TMSK2, the bits of TFLG2 allow the timer subsystem to operate in either a polled or interrupt driven system. Each bit of TFLG2 corresponds to a bit in TMSK2 in the same position.

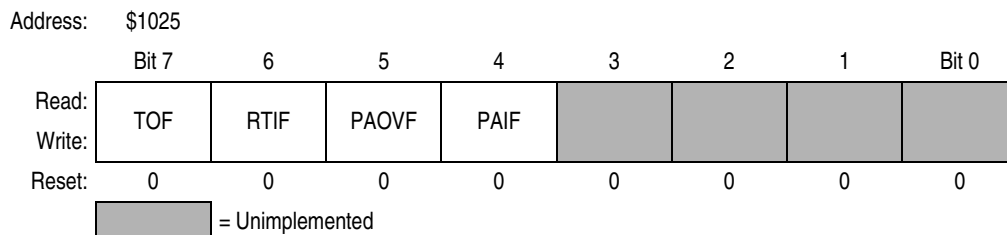


Figure 9-22. Timer Interrupt Flag 2 Register (TFLG2)

Clear flags by writing a 1 to the corresponding bit position(s).

TOF — Timer Overflow Interrupt Flag

Set when TCNT changes from \$FFFF to \$0000

RTIF — Real-Time Interrupt Flag

The RTIF status bit is automatically set to 1 at the end of every RTI period. To clear RTIF, write a byte to TFLG2 with bit 6 set.

PAOVF — Pulse Accumulator Overflow Interrupt Flag

Refer to [9.7 Pulse Accumulator](#).

PAIF — Pulse Accumulator Input Edge Interrupt Flag

Refer to [9.7 Pulse Accumulator](#).

Bits [3:0] — Unimplemented

Always read 0

9.5.3 Pulse Accumulator Control Register

Bits RTR[1:0] of this register select the rate for the RTI system. The remaining bits control the pulse accumulator and IC4/OC5 functions.

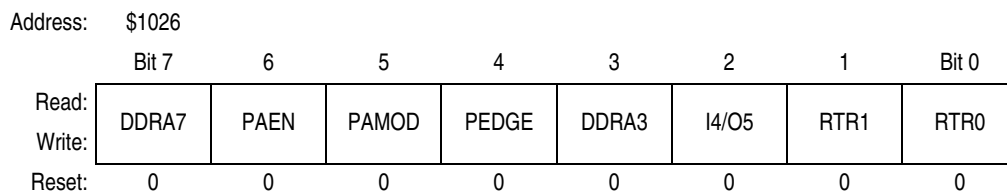


Figure 9-23. Pulse Accumulator Control Register (PACTL)

DDRA7 — Data Direction for Port A Bit 7

Refer to [Chapter 6 Parallel Input/Output \(I/O\) Ports](#).

PAEN — Pulse Accumulator System Enable Bit

Refer to [9.7 Pulse Accumulator](#).

PAMOD — Pulse Accumulator Mode Bit

Refer to [9.7 Pulse Accumulator](#).

10.11 Peripheral Port Timing

Characteristic ^{(1) (2)}	Symbol	1.0 MHz		2.0 MHz		3.0 MHz		Unit
		Min	Max	Min	Max	Min	Max	
Frequency of operation E-clock frequency	f_o	dc	1.0	dc	2.0	dc	3.0	MHz
E-clock period	t_{CYC}	1000	—	500	—	333	—	ns
Peripheral data setup time MCU read of ports A, C, D, and E	t_{PDSU}	100	—	100	—	100	—	ns
Peripheral data hold time MCU read of ports A, C, D, and E	t_{PDH}	50	—	50	—	50	—	ns
Delay time, peripheral data write $t_{PWD} = 1/4 t_{CYC} + 100$ ns MCU writes to port A MCU writes to ports B, C, and D	t_{PWD}	— —	200 350	— —	200 225	— —	200 183	ns
Port C input data setup time	t_{IS}	60	—	60	—	60	—	ns
Port C input data hold time	t_{IH}	100	—	100	—	100	—	ns
Delay time, E fall to STRB $t_{DEB} = 1/4 t_{CYC} + 100$ ns	t_{DEB}	—	350	—	225	—	183	ns
Setup time, STRA asserted to E fall ⁽³⁾	t_{AES}	0	—	0	—	0	—	ns
Delay time, STRA asserted to port C data output valid	t_{PCD}	—	100	—	100	—	100	ns
Hold time, STRA negated to port C data	t_{PCH}	10	—	10	—	10	—	ns
3-state hold time	t_{PCZ}	—	150	—	150	—	150	ns

- $V_{DD} = 5.0$ Vdc \pm 10%, $V_{SS} = 0$ Vdc, $T_A = T_L$ to T_H , all timing is shown with respect to 20% V_{DD} and 70% V_{DD} , unless otherwise noted
- Ports C and D timing is valid for active drive. (CWOM and DWOM bits are not set in PIOC and SPCR registers, respectively.)
- If this setup time is met, STRB acknowledges in the next cycle. If it is not met, the response may be delayed one more cycle.

10.13 Analog-to-Digital Converter Characteristics

Characteristic ⁽¹⁾	Parameter ⁽²⁾	Min	Absolute	2.0 MHz	3.0 MHz	Unit
				Max	Max	
Resolution	Number of bits resolved by A/D converter	—	8	—	—	Bits
Non-linearity	Maximum deviation from the ideal A/D transfer characteristics	—	—	±1/2	±1	LSB
Zero error	Difference between the output of an ideal and an actual for 0 input voltage	—	—	±1/2	±1	LSB
Full scale error	Difference between the output of an ideal and an actual A/D for full-scale input voltage	—	—	±1/2	±1	LSB
Total unadjusted error	Maximum sum of non-linearity, zero error, and full-scale error	—	—	±1/2	±1/2	LSB
Quantization error	Uncertainty because of converter resolution	—	—	±1/2	±1/2	LSB
Absolute accuracy	Difference between the actual input voltage and the full-scale weighted equivalent of the binary output code, all error sources included	—	—	±1	±2	LSB
Conversion range	Analog input voltage range	V _{RL}	—	V _{RH}	V _{RH}	V
V _{RH}	Maximum analog reference voltage ⁽³⁾	V _{RL}	—	V _{DD} +0.1	V _{DD} +0.1	V
V _{RL}	Minimum analog reference voltage ⁽²⁾	V _{SS} -0.1	—	V _{RH}	V _{RH}	V
ΔV _R	Minimum difference between V _{RH} and V _{RL} ⁽²⁾	3	—	—	—	V
Conversion time	Total time to perform a single A/D conversion: E clock Internal RC oscillator	— —	32 —	— t _{CYC} +32	— t _{CYC} +32	t _{CYC} μs
Monotonicity	Conversion result never decreases with an increase in input voltage; has no missing codes	—	Guaranteed	—	—	—
Zero input reading	Conversion result when V _{In} = V _{RL}	00	—	—	—	Hex
Full scale reading	Conversion result when V _{In} = V _{RH}	—	—	FF	FF	Hex
Sample acquisition time	Analog input acquisition sampling time: E clock Internal RC oscillator	— —	12 —	— 12	— 12	t _{CYC} μs
Sample/hold capacitance	Input capacitance during sample PE[7:0]	—	20 typical	—	—	pF
Input leakage	Input leakage on A/D pins PE[7:0] V _{RL} , V _{RH}	— —	— —	400 1.0	400 1.0	nA μA

1. V_{DD} = 5.0 Vdc ±10%, V_{SS} = 0 Vdc, T_A = T_L to T_H, 750 kHz ≤ E ≤ 3.0 MHz, unless otherwise noted
2. Source impedances greater than 10 kΩ affect accuracy adversely because of input leakage.
3. Performance verified down to 2.5 V ΔV_R, but accuracy is tested and guaranteed at ΔV_R = 5 V ±10%.

10.15 Expansion Bus Timing Characteristics

Num	Characteristic ⁽¹⁾	Symbol	1.0 MHz		2.0 MHz		3.0 MHz		Unit
			Min	Max	Min	Max	Min	Max	
	Frequency of operation (E-clock frequency)	f_o	dc	1.0	dc	2.0	dc	3.0	MHz
1	Cycle time	t_{CYC}	1000	—	500	—	333	—	ns
2	Pulse width, E low ⁽²⁾ , $PW_{EL} = 1/2 t_{CYC} - 23$ ns	PW_{EL}	477	—	227	—	146	—	ns
3	Pulse width, E high ⁽²⁾ , $PW_{EH} = 1/2 t_{CYC} - 28$ ns	PW_{EH}	472	—	222	—	141	—	ns
4a	E and AS rise time	t_r	—	20	—	20	—	20	ns
4b	E and AS fall time	t_f	—	20	—	20	—	15	ns
9	Address hold time ⁽²⁾ ^{(3)a} , $t_{AH} = 1/8 t_{CYC} - 29.5$ ns	t_{AH}	95.5	—	33	—	26	—	ns
12	Non-multiplexed address valid time to E rise $t_{AV} = PW_{EL} - (t_{ASD} + 80 \text{ ns})$ ⁽²⁾ ^{(3)a}	t_{AV}	281.5	—	94	—	54	—	ns
17	Read data setup time	t_{DSR}	30	—	30	—	30	—	ns
18	Read data hold time, max = t_{MAD}	t_{DHR}	0	145.5	0	83	0	51	ns
19	Write data delay time, $t_{DDW} = 1/8 t_{CYC} + 65.5$ ns ⁽²⁾ ^{(3)a}	t_{DDW}	—	190.5	—	128	—	71	ns
21	Write data hold time, $t_{DHW} = 1/8 t_{CYC} - 29.5$ ns ⁽²⁾ ^{(3)a}	t_{DHW}	95.5	—	33	—	26	—	ns
22	Multiplexed address valid time to E rise $t_{AVM} = PW_{EL} - (t_{ASD} + 90 \text{ ns})$ ⁽²⁾ ^{(3)a}	t_{AVM}	271.5	—	84	—	54	—	ns
24	Multiplexed address valid time to AS fall $t_{ASL} = PW_{ASH} - 70$ ns ⁽²⁾	t_{ASL}	151	—	26	—	13	—	ns
25	Multiplexed address hold time $t_{AHL} = 1/8 t_{CYC} - 29.5$ ns ⁽²⁾ ^{(3)b}	t_{AHL}	95.5	—	33	—	31	—	ns
26	Delay time, E to AS rise, $t_{ASD} = 1/8 t_{CYC} - 9.5$ ns ⁽²⁾ ^{(3)a}	t_{ASD}	115.5	—	53	—	31	—	ns
27	Pulse width, AS high, $PW_{ASH} = 1/4 t_{CYC} - 29$ ns ⁽²⁾	PW_{ASH}	221	—	96	—	63	—	ns
28	Delay time, AS to E rise, $t_{ASED} = 1/8 t_{CYC} - 9.5$ ns ⁽²⁾ ^{(3)b}	t_{ASED}	115.5	—	53	—	31	—	ns
29	MPU address access time ^{(3)a} $t_{ACCA} = t_{CYC} - (PW_{EL} - t_{AVM}) - t_{DSR} - t_f$	t_{ACCA}	744.5	—	307	—	196	—	ns
35	MPU access time, $t_{ACCE} = PW_{EH} - t_{DSR}$	t_{ACCE}	—	442	—	192	—	111	ns
36	Multiplexed address delay (Previous cycle MPU read) $t_{MAD} = t_{ASD} + 30$ ns ⁽²⁾ ^{(3)a}	t_{MAD}	145.5	—	83	—	51	—	ns

1. $V_{DD} = 5.0 \text{ Vdc} \pm 10\%$, $V_{SS} = 0 \text{ Vdc}$, $T_A = T_L$ to T_H , all timing is shown with respect to 20% V_{DD} and 70% V_{DD} , unless otherwise noted

2. Formula only for dc to 2 MHz

3. Input clocks with duty cycles other than 50% affect bus performance. Timing parameters affected by input clock duty cycle are identified by (a) and (b). To recalculate the approximate bus timing values, substitute the following expressions in place of $1/8 t_{CYC}$ in the above formulas, where applicable:

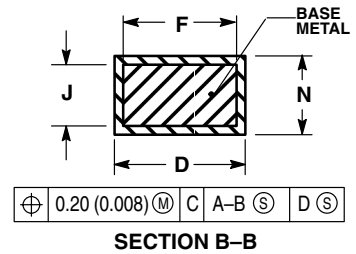
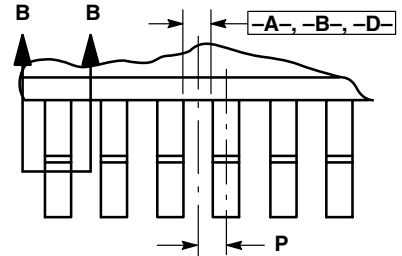
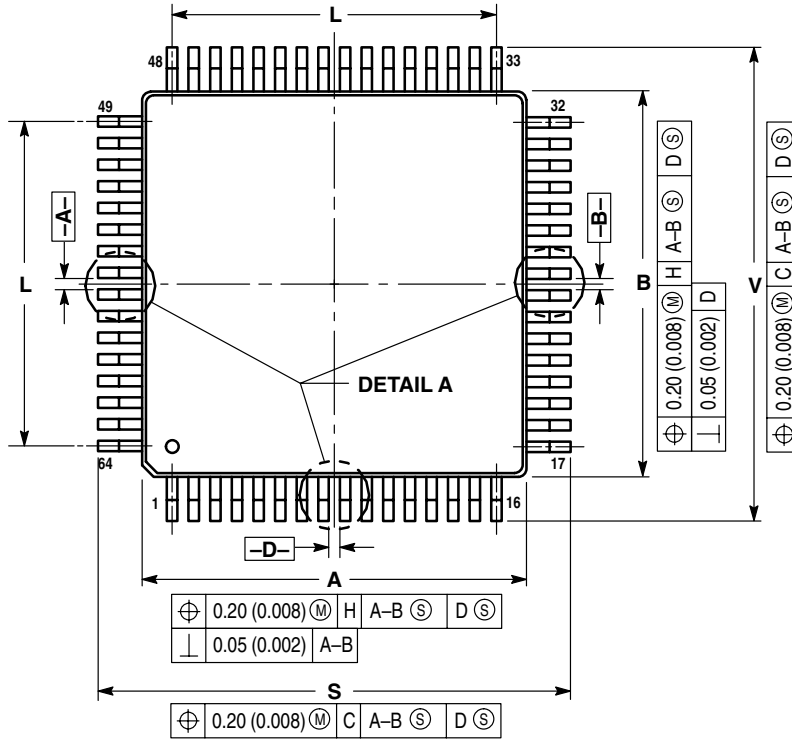
(a) $(1 - dc) \times 1/4 t_{CYC}$

(b) $dc \times 1/4 t_{CYC}$

Where:

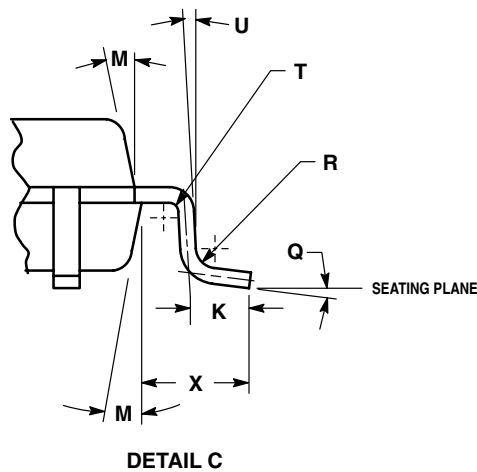
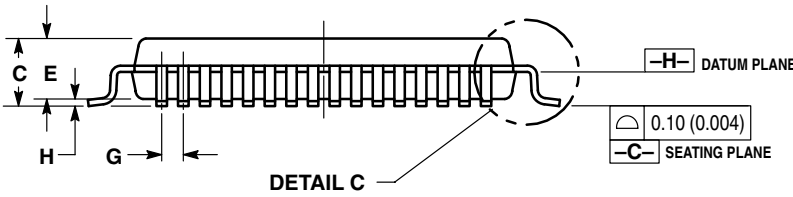
dc is the decimal value of duty cycle percentage (high time)

11.7 64-Pin Quad Flat Pack (Case 840C)



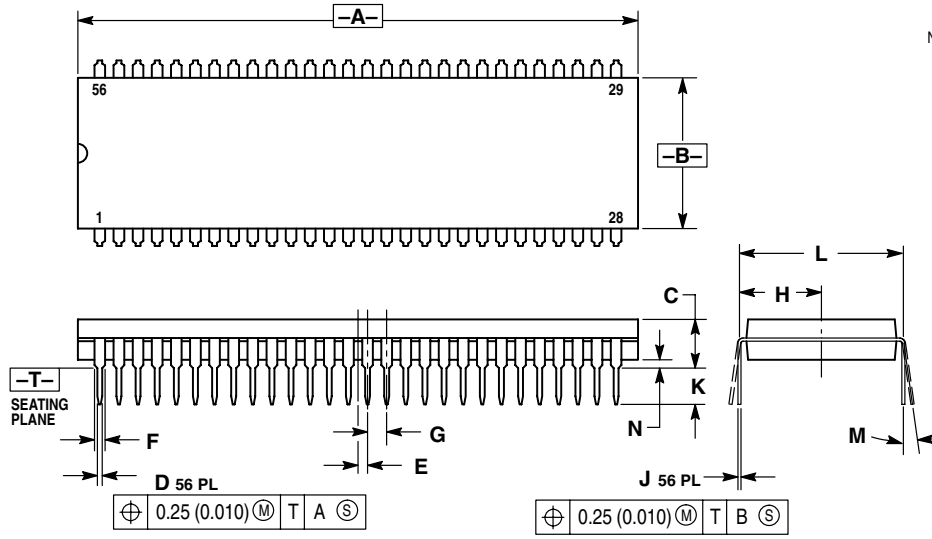
NOTES:

1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
2. CONTROLLING DIMENSION: MILLIMETER.
3. DATUM PLANE -H- IS LOCATED AT BOTTOM OF LEAD AND IS COINCIDENT WITH THE LEAD WHERE THE LEAD EXITS THE PLASTIC BODY AT THE BOTTOM OF THE PARTING LINE.
4. DATUMS A-B AND -D- TO BE DETERMINED AT DATUM PLANE -H-.
5. DIMENSIONS S AND V TO BE DETERMINED AT SEATING PLANE -C-.
6. DIMENSIONS A AND B DO NOT INCLUDE MOLD PROTRUSION. ALLOWABLE PROTRUSION IS 0.25 (0.010) PER SIDE. DIMENSIONS A AND B DO INCLUDE MOLD MISMATCH AND ARE DETERMINED AT DATUM PLANE -H-.
7. DIMENSION D DOES NOT INCLUDE DAMBAR PROTRUSION. DAMBAR PROTRUSION SHALL NOT CAUSE THE D DIMENSION TO EXCEED 0.53 (0.021). DAMBAR CANNOT BE LOCATED ON THE LOWER RADIUS OR THE FOOT.
8. DIMENSION K IS TO BE MEASURED FROM THE THEORETICAL INTERSECTION OF LEAD FOOT AND LEG CENTERLINES.



DIM	MILLIMETERS		INCHES	
	MIN	MAX	MIN	MAX
A	13.90	14.10	0.547	0.555
B	13.90	14.10	0.547	0.555
C	2.07	2.46	0.081	0.097
D	0.30	0.45	0.012	0.018
E	2.00	2.40	0.079	0.094
F	0.30	—	0.012	—
G	0.80 BSC	—	0.031 BSC	—
H	0.067	0.250	0.003	0.010
J	0.130	0.230	0.005	0.090
K	0.50	0.66	0.020	0.026
L	12.00 REF	—	0.472 REF	—
M	5°	10°	5°	10°
N	0.130	0.170	0.005	0.007
P	0.40 BSC	—	0.016 BSC	—
Q	2°	8°	2°	8°
R	0.13	0.30	0.005	0.012
S	16.20	16.60	0.638	0.654
T	0.20 REF	—	0.008 REF	—
U	0°	—	0°	—
V	16.20	16.60	0.638	0.654
X	1.10	1.30	0.043	0.051

11.9 56-Pin Dual in-Line Package (Case 859)



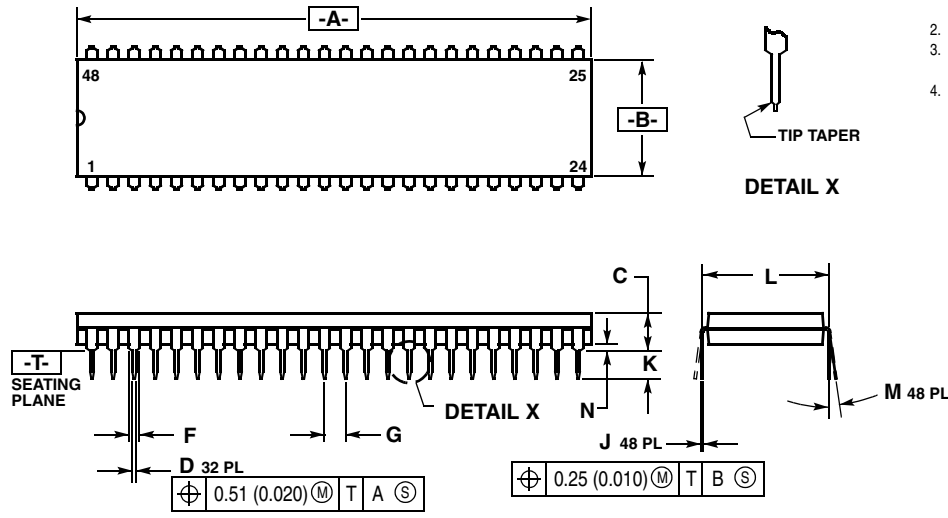
- NOTES:
1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
 2. CONTROLLING DIMENSION: INCH.
 3. DIMENSION L TO CENTER OF LEAD WHEN FORMED PARALLEL.
 4. DIMENSIONS A AND B DO NOT INCLUDE MOLD FLASH. MAXIMUM MOLD FLASH 0.25 (0.010)

DIM	INCHES		MILLIMETERS	
	MIN	MAX	MIN	MAX
A	2.035	2.065	51.69	52.45
B	0.540	0.560	13.72	14.22
C	0.155	0.200	3.94	5.08
D	0.014	0.022	0.36	0.56
E	0.035 BSC		0.89 BSC	
F	0.032	0.046	0.81	1.17
G	0.070 BSC		1.778 BSC	
H	0.300 BSC		7.62 BSC	
J	0.008	0.015	0.20	0.38
K	0.115	0.135	2.92	3.43
L	0.600 BSC		15.24 BSC	
M	0°	15°	0°	15°
N	0.020	0.040	0.51	1.02

11.10 48-Pin Plastic DIP (Case 767)

NOTE

The MC68HC811E2 is the only member of the E series that is offered in a 48-pin plastic dual in-line package.



- NOTES:
1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
 2. CONTROLLING DIMENSION: INCH.
 3. DIMENSION L TO CENTER OF LEAD WHEN FORMED PARALLEL.
 4. DIMENSIONS A AND B DO NOT INCLUDE MOLD FLASH. MAXIMUM MOLD FLASH 0.25 (0.010).

DIM	INCHES		MILLIMETERS	
	MIN	MAX	MIN	MAX
A	2.415	2.445	61.34	62.10
B	0.540	0.560	13.72	14.22
C	0.155	0.200	3.94	5.08
D	0.014	0.022	0.36	0.55
F	0.040	0.060	1.02	1.52
G	0.100 BSC		2.54 BSC	
H	0.070 BSC		1.79 BSC	
J	0.008	0.015	0.20	0.38
K	0.115	0.150	2.92	3.81
L	0.600 BSC		15.24 BSC	
M	0°	15°	0°	15°
N	0.020	0.040	0.51	1.01

Appendix A

Development Support

A.1 Introduction

This section provides information on the development support offered for the E-series devices.

A.2 M68HC11 E-Series Development Tools

Device	Package	Emulation Module ^{(1) (2)}	Flex Cable ^{(1) (2)}	MMDS11 Target Head ^{(1) (2)}	SPGMR Programming Adapter ⁽³⁾
MC68HC11E9 MC68HC711E9	52 FN	M68EM11E20	M68CBL11C	M68TC11E20FN52	M68PA11E20FN52
	52 PB	M68EM11E20	M68CBL11C	M68TC11E20PB52	M68PA11E20PB52
	56 B	M68EM11E20	M68CBL11B	M68TC11E20B56	M68PA11E20B56
	64 FU	M68EM11E20	M68CBL11C	M68TC11E20FU64	M68PA11E20FU64
MC68HC11E20 MC68HC711E20	52 FN	M68EM11E20	M68CBL11C	M68TC11E20FN52	M68PA11E20FN52
	64 FU	M68EM11E20	M68CBL11C	M68TC11E20FU64	M68PA11E20FU64
MC68HC811E2	48 P	M68EM11E20	M68CBL11B	M68TB11E20P48	M68PA11A8P48
	52 FN	M68EM11E20	M68CBL11C	M68TC11E20FN52	M68PA11E20FN52

1. Each MMDS11 system consists of a system console (M68MMDS11), an emulation module, a flex cable, and a target head.
2. A complete EVS consists of a platform board (M68HC11PFB), an emulation module, a flex cable, and a target head.
3. Each SPGMR system consists of a universal serial programmer (M68SPGMR11) and a programming adapter. It can be used alone or in conjunction with the MMDS11.

A.3 EVS — Evaluation System

The EVS is an economical tool for designing, debugging, and evaluating target systems based on the M68HC11. EVS features include:

- Monitor/debugger firmware
- One-line assembler/disassembler
- Host computer download capability
- Dual memory maps:
 - 64-Kbyte monitor map that includes 16 Kbytes of monitor EPROM
 - M68HC11 E-series user map that includes 64 Kbytes of emulation RAM
- MCU extension input/output (I/O) port for single-chip, expanded, and special-test operation modes
- RS-232C terminal and host I/O ports
- Logic analyzer connector

After the MCU sends \$FF [8], it enters the WAIT1 loop [9] and waits for the first data character from the host. When this character is received [10], the MCU programs it into the address pointed to by the Y index register. When the programming time delay is over, the MCU reads the programmed data, transmits it to the host for verification [11], and returns to the top of the WAIT1 loop to wait for the next data character [12]. Because the host previously sent the second data character, it is already waiting in the SCI receiver of the MCU. Steps [13], [14], and [15] correspond to the second pass through the WAIT1 loop.

Back in the host, the first verify character is received, and the third data character is sent [6]. The host then waits for the second verify character [7] to come back from the MCU. The sequence continues as long as the host continues to send data to the MCU. Since the WAIT1 loop in the PROGRAM utility is an indefinite loop, reset is used to end the process in the MCU after the host has finished sending data to be programmed.

Allowing for Bootstrap Mode

Since bootstrap mode requires few connections to the MCU, it is easy to design systems that accommodate bootstrap mode.

Bootstrap mode is useful for diagnosing or repairing systems that have failed due to changes in the CONFIG register or failures of the expansion address/data buses, (rendering programs in external memory useless). Bootstrap mode can also be used to load information into the EPROM or EEPROM of an M68HC11 after final assembly of a module. Bootstrap mode is also useful for performing system checks and calibration routines. The following paragraphs explain system requirements for use of bootstrap mode in a product.

Mode Select Pins

It must be possible to force the MODA and MODB pins to logic 0, which implies that these two pins should be pulled up to V_{DD} through resistors rather than being tied directly to V_{DD} . If mode pins are connected directly to V_{DD} , it is not possible to force a mode other than the one the MCU is hard wired for. It is also good practice to use pulldown resistors to V_{SS} rather than connecting mode pins directly to V_{SS} because it is sometimes a useful debug aid to attempt reset in modes other than the one the system was primarily designed for. Physically, this requirement sometimes calls for the addition of a test point or a wire connected to one or both mode pins. Mode selection only uses the mode pins while $\overline{\text{RESET}}$ is active.

$\overline{\text{RESET}}$

It must be possible to initiate a reset while the mode select pins are held low. In systems where there is no provision for manual reset, it is usually possible to generate a reset by turning power off and back on.

RxD Pin

It must be possible to drive the PD0/RxD pin with serial data from a host computer (or another MCU). In many systems, this pin is already used for SCI communications; thus no changes are required.

A problem arose with the BASIC programming technique used. The draft versions of this program tried saving the object code bytes directly as binary in a string array. This caused "Out of Memory" or "Out of String Space" errors on both a 2-Mbyte Macintosh and a 640-Kbyte PC. The solution was to make the array an integer array and perform the integer-to-binary conversion on each byte as it is sent to the target part.

The one compromise made to accommodate both Macintosh and PC versions of BASIC is in lines 1500 and 1505. Use line 1500 and comment out line 1505 if the program is to be run on a Macintosh, and, conversely, use line 1505 and comment out line 1500 if a PC is used.

After the COM port is opened, the code to be bootloaded is modified by adding the \$FF to the start of the string. \$FF synchronizes the bootloader in the MC68HC711E9 to 1200 baud. The entire string is simply sent to the COM port by PRINTing the string. This is possible since the string is actually queued in BASIC's COM buffer, and the operating system takes care of sending the bytes out one at a time. The M68HC11 echoes the data received for verification. No automatic verification is provided, though the data is printed to the screen for manual verification.

Once the MCU has received this bootloaded code, the bootloader automatically jumps to it. The small bootloaded program in turn includes a jump to the EPROM programming routine in the boot ROM.

Refer to the previous explanation of the [EPROM Programming Utility](#) for the following discussion. The host system sends the first byte to be programmed through the COM port to the SCI of the MCU. The SCI port on the MCU buffers one byte while receiving another byte, increasing the throughput of the EPROM programming operation by sending the second byte while the first is being programmed.

When the first byte has been programmed, the MCU reads the EPROM location and sends the result back to the host system. The host then compares what was actually programmed to what was originally sent. A message indicating which byte is being verified is displayed in the lower half of the screen. If there is an error, it is displayed at the top of the screen.

As soon as the first byte is verified, the third byte is sent. In the meantime, the MCU has already started programming the second byte. This process of verifying and queueing a byte continues until the host finishes sending data. If the programming is completely successful, no error messages will have been displayed at the top of the screen. Subroutines follow the end of the program to handle some of the repetitive tasks. These routines are short, and the commenting in the source code should be sufficient explanation.

Modifications

This example programmed version 3.4 of the BUFFALO monitor into the EPROM of an MC68HC711E9; the changes to the BASIC program to download some other program are minor.

The necessary changes are:

1. In line 30, the length of the program to be downloaded must be assigned to the variable CODESIZE%.
2. Also in line 30, the starting address of the program is assigned to the variable ADRSTART.
3. In line 9570, the start address of the program is stored in the third and fourth items in that DATA statement in hexadecimal.
4. If any changes are made to the number of bytes in the boot code in the DATA statements in lines 9500–9580, then the new count must be set in the variable "BOOTCOUNT" in line 25.

Operation

Configure the EVBU for boot mode operation by putting a jumper at J3. Ensure that the trace command jumper at J7 is not installed because this would connect the 12-V programming voltage to the OC5 output of the MCU.

Connect the EVBU to its dc power supply. When it is time to program the MCU EPROM, turn on the 12-volt programming power supply to the new circuitry in the wire-wrap area.

Connect the EVBU serial port to the appropriate serial port on the host system. For the Macintosh, this is the modem port with a modem cable. For the MS-DOS[®] computer, it is connected to COM1 with a straight through or modem cable. Power up the host system and start the BASIC program. If the program has not been compiled, this is accomplished from within the appropriate BASIC compiler or interpreter. Power up the EVBU.

Answer the prompt for filename with either a [RETURN] to accept the default shown or by typing in a new filename and pressing [RETURN].

The program will inform the user that it is working on converting the file from S records to binary. This process will take from 30 seconds to a few minutes, depending on the computer.

A prompt reading, "Comm port open?" will appear at the end of the file conversion. This is the last chance to ensure that everything is properly configured on the EVBU. Pressing [RETURN] will send the bootcode to the target MC68HC711E9. The program then informs the user that the bootload code is being sent to the target, and the results of the echoing of this code are displayed on the screen.

Another prompt reading "Programming is ready to begin. Are you?" will appear. Turn on the 12-volt programming power supply and press [RETURN] to start the actual programming of the target EPROM.

A count of the byte being verified will be updated continually on the screen as the programming progresses. Any failures will be flagged as they occur.

When programming is complete, a message will be displayed as well as a prompt requesting the user to press [RETURN] to quit.

Turn off the 12-volt programming power supply before turning off 5 volts to the EVBU.

[®] MS-DOS is a registered trademark of Microsoft Corporation in the United States and other countries.

Step 6

After the programming operation is complete, PCbug11 will display this message

Total bytes loaded: \$xxxx

Total bytes programmed: \$yyyy

- You should now remove the programming voltage from P4 connector pin 18, the XIRQ* pin.
- Each ORG directive in your assembly language source will cause a pair of these lines to be generated. For this operation, \$yyyy will be incremented by the size of each block of code programmed into the EPROM of the MC68HC711E9.
- PCbug11 will display the above message whether or not the programming operation was successful. As a precaution, you should have PCbug11 verify your code.
- At the PCbug11 command prompt type: VERF C:\MYPROG\ISHERE.S19

Substitute the name of your program into the command above. Use a full path name if your program is not located in the same directory as PCbug11.

If the verify operation fails, a list of addresses which did not program correctly is displayed. Should this occur, you probably need to erase your part more completely. To do so, allow the MC68HC711E9 to sit for at least 45 minutes under an ultraviolet light source. Attempt the programming operation again. If you have purchased devices in plastic packages (one-time programmable parts), you will need to try again with a new, unprogrammed device.

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