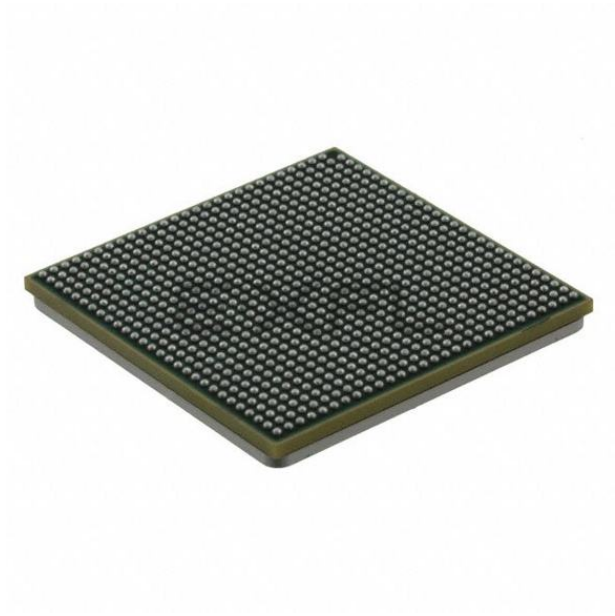

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Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Obsolete
Core Processor	PowerPC e500
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	667MHz
Co-Processors/DSP	Communications; CPM, Security; SEC
RAM Controllers	DDR, SDRAM
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100/1000Mbps (2)
SATA	-
USB	USB 2.0 (1)
Voltage - I/O	2.5V, 3.3V
Operating Temperature	0°C ~ 105°C (TA)
Security Features	Cryptography, Random Number Generator
Package / Case	783-BBGA, FCBGA
Supplier Device Package	783-FCPBGA (29x29)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/kmpc8555epxalf

- Two full-duplex fast communications controllers (FCCs) that support the following protocols:
  - ATM protocol through two UTOPIA level 2 interfaces
  - IEEE Std 802.3™/Fast Ethernet (10/100)
  - HDLC
  - Totally transparent operation
- Three full-duplex serial communications controllers (SCCs) support the following protocols:
  - High level/synchronous data link control (HDLC/SDLC)
  - LocalTalk (HDLC-based local area network protocol)
  - Universal asynchronous receiver transmitter (UART)
  - Synchronous UART (1x clock mode)
  - Binary synchronous communication (BISYNC)
  - Totally transparent operation
  - QMC support, providing 64 channels per SCC using only one physical TDM interface
- Universal serial bus (USB) controller that is full/low-speed compliant (multiplexed on an SCC)
  - USB host mode
  - Supports USB slave mode
- Serial peripheral interface (SPI) support for master or slave
- I^2C bus controller
- Two serial management controllers (SMCs) supporting:
  - UART
  - Transparent
  - General-circuit interfaces (GCI)
- Time-slot assigner supports multiplexing of data from any of the SCCs and FCCs onto eight time-division multiplexed (TDM) interfaces. The time-slot assigner supports the following TDM formats:
  - T1/CEPT lines
  - T3/E3
  - Pulse code modulation (PCM) highway interface
  - ISDN primary rate
  - Freescale interchip digital link (IDL)
  - General circuit interface (GCI)
- User-defined interfaces
- Eight independent baud rate generators (BRGs)
- Four general-purpose 16-bit timers or two 32-bit timers
- General-purpose parallel ports—16 parallel I/O lines with interrupt capability

• 256 Kbytes of on-chip memory
- Can act as a 256-Kbyte level-2 cache
- Can act as a 256-Kbyte or two 128-Kbyte memory-mapped SRAM arrays

6 DDR SDRAM

This section describes the DC and AC electrical specifications for the DDR SDRAM interface of the MPC8555E.

6.1 DDR SDRAM DC Electrical Characteristics

Table 11 provides the recommended operating conditions for the DDR SDRAM component(s) of the MPC8555E.

Table 11. DDR SDRAM DC Electrical Characteristics

Parameter/Condition	Symbol	Min	Max	Unit	Notes
I/O supply voltage	GV_{DD}	2.375	2.625	V	1
I/O reference voltage	MV_{REF}	$0.49 \times GV_{DD}$	$0.51 \times GV_{DD}$	V	2
I/O termination voltage	V_{TT}	$MV_{REF} - 0.04$	$MV_{REF} + 0.04$	V	3
Input high voltage	V_{IH}	$MV_{REF} + 0.18$	$GV_{DD} + 0.3$	V	—
Input low voltage	V_{IL}	–0.3	$MV_{REF} - 0.18$	V	—
Output leakage current	I_{OZ}	–10	10	µA	4
Output high current (V_{OUT} = 1.95 V)	I_{OH}	–15.2	—	mA	—
Output low current (V_{OUT} = 0.35 V)	I_{OL}	15.2	—	mA	—
MV_{REF} input leakage current	I_{VREF}	—	5	µA	—

Notes:
1. GV_{DD} is expected to be within 50 mV of the DRAM GV_{DD} at all times.
2. MV_{REF} is expected to be equal to $0.5 \times GV_{DD}$, and to track GV_{DD} DC variations as measured at the receiver. Peak-to-peak noise on MV_{REF} may not exceed ±2% of the DC value.
3. V_{TT} is not applied directly to the device. It is the supply to which far end signal termination is made and is expected to be equal to MV_{REF}. This rail should track variations in the DC level of MV_{REF}.
4. Output leakage is measured with all outputs disabled, $0 V \le V_{OUT} \le GV_{DD}$.

Table 12 provides the DDR capacitance.

Table 12. DDR SDRAM Capacitance

Parameter/Condition	Symbol	Min	Max	Unit	Notes
Input/output capacitance: DQ, DQS, MSYNC_IN	C_{IO}	6	8	pF	1
Delta input/output capacitance: DQ, DQS	C_{DIO}	—	0.5	pF	1

Note:
1. This parameter is sampled. GV_{DD} = 2.5 V ± 0.125 V, f = 1 MHz, T_A = 25°C, $V_{OUT} = GV_{DD}/2$, V_{OUT} (peak to peak) = 0.2 V.

8.2 GMII, MII, TBI, RGMII, and RTBI AC Timing Specifications

The AC timing specifications for GMII, MII, TBI, RGMII, and RTBI are presented in this section.

8.2.1 GMII AC Timing Specifications

This section describes the GMII transmit and receive AC timing specifications.

8.2.2 GMII Transmit AC Timing Specifications

Table 20 provides the GMII transmit AC timing specifications.

Table 20. GMII Transmit AC Timing Specifications

At recommended operating conditions with LV_{DD} of 3.3 V ± 5%.

Parameter/Condition	Symbol [1]	Min	Typ	Max	Unit
GTX_CLK clock period	t_{GTX}	—	8.0	—	ns
GTX_CLK duty cycle	t_{GTXH}/t_{GTX}	40	—	60	%
GMII data TXD[7:0], TX_ER, TX_EN setup time	t_{GTKHDV}	2.5	—	—	ns
GTX_CLK to GMII data TXD[7:0], TX_ER, TX_EN delay	t_{GTKHDX}	0.5	—	5.0	ns
GTX_CLK data clock rise and fall times	t_{GTXR}[3], t_{GTXR}[2,4]	—	—	1.0	ns

Notes:

1. The symbols used for timing specifications herein follow the pattern $t_{\text{(first two letters of functional block)(signal)(state) (reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{GTKHDV} symbolizes GMII transmit timing (GT) with respect to the t_{GTX} clock reference (K) going to the high state (H) relative to the time date input signals (D) reaching the valid state (V) to state or setup time. Also, t_{GTKHDX} symbolizes GMII transmit timing (GT) with respect to the t_{GTX} clock reference (K) going to the high state (H) relative to the time date input signals (D) going invalid (X) or hold time. Note that, in general, the clock reference symbol representation is based on three letters representing the clock of a particular functional. For example, the subscript of t_{GTX} represents the GMII(G) transmit (TX) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).
2. Signal timings are measured at 0.7 V and 1.9 V voltage levels.
3. Guaranteed by characterization.
4. Guaranteed by design.

Figure 7 shows the GMII transmit AC timing diagram.

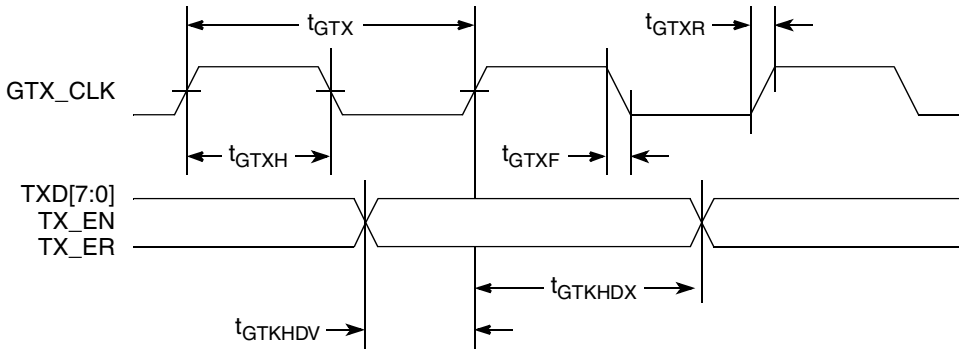


Figure 7. GMII Transmit AC Timing Diagram

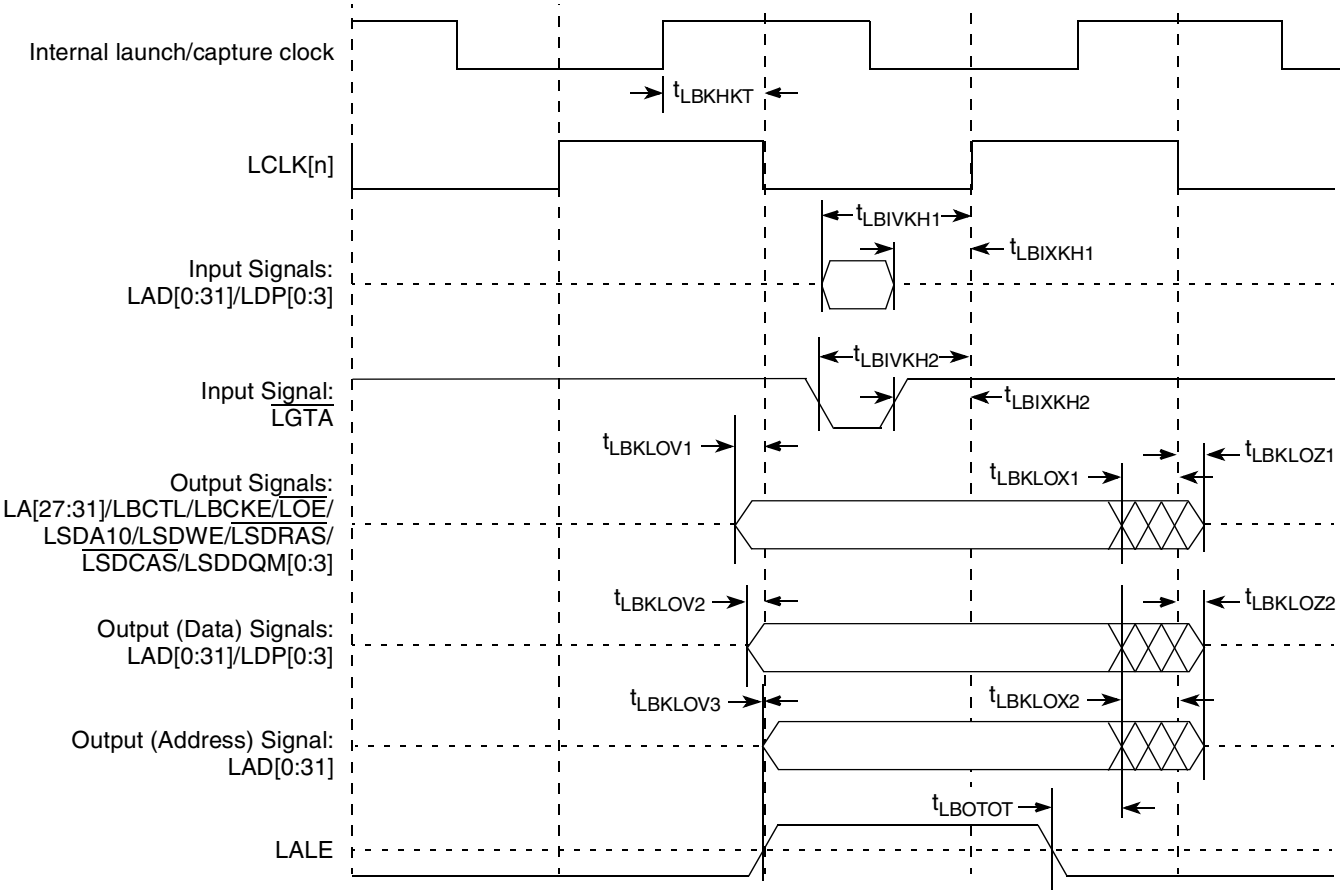


Figure 18. Local Bus Signals, Nonspecial Signals Only (DLL Bypass Mode)

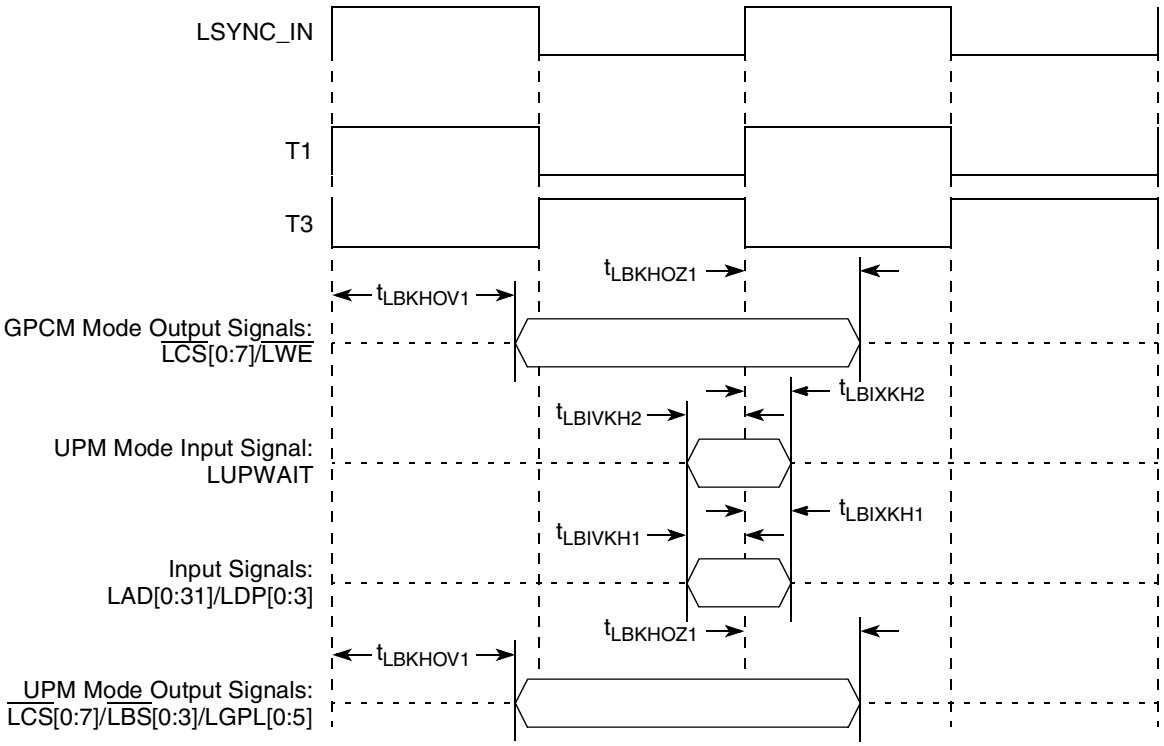


Figure 19. Local Bus Signals, GPCM/UPM Signals for LCCR[CLKDIV] = 2 (DLL Enabled)

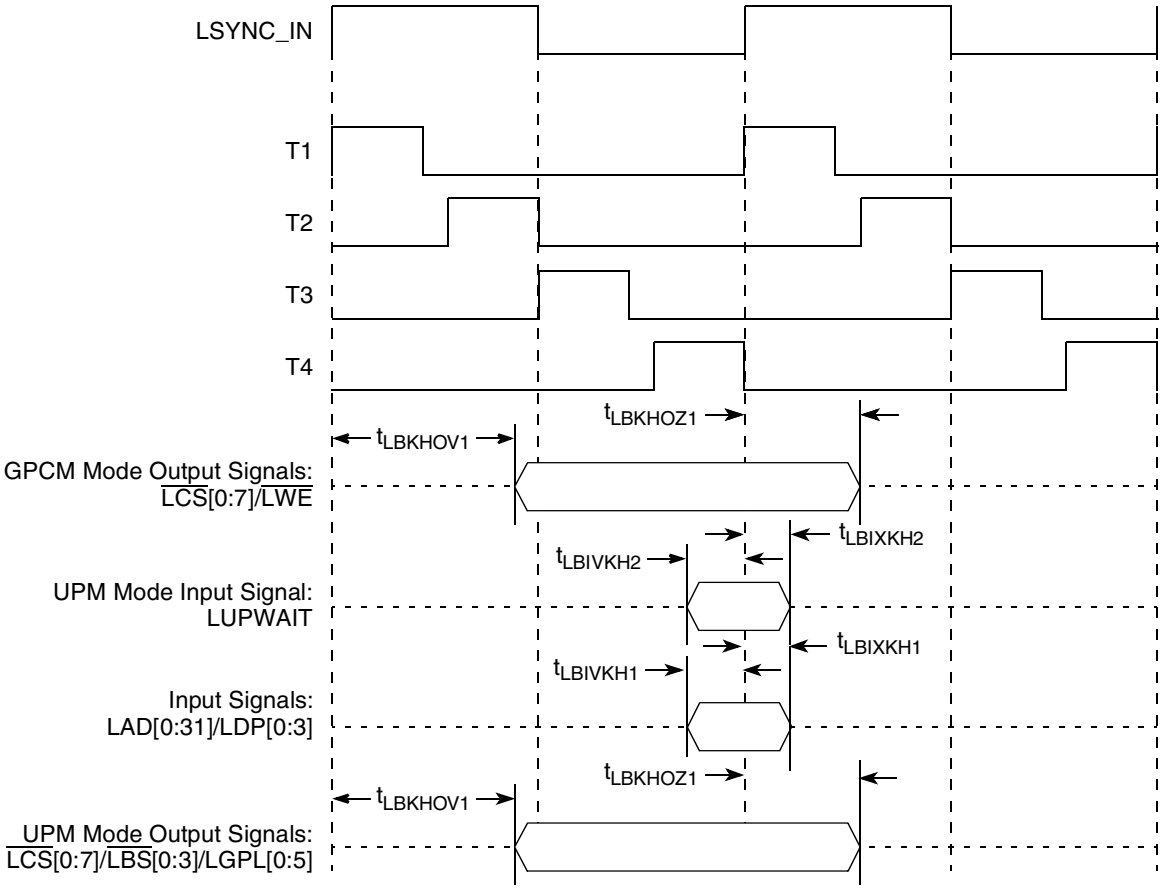


Figure 21. Local Bus Signals, GPCM/UPM Signals for LCCR[CLKDIV] = 4 or 8 (DLL Enabled)

Table 33. CPM Input AC Timing Specifications [1] (continued)

Characteristic	Symbol [2]	Min[3]	Unit
TDM inputs/SI—hold time	t_{TDIXKH}	3	ns
PIO inputs—input setup time	t_{PIIVKH}	8	ns
PIO inputs—input hold time	t_{PIIXKH}	1	ns
COL width high (FCC)	t_{FCCH}	1.5	CLK

Notes:
1. Input specifications are measured from the 50% level of the signal to the 50% level of the rising edge of CLKIN. Timings are measured at the pin.
2. The symbols used for timing specifications herein follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{FIIVKH} symbolizes the FCC inputs internal timing (FI) with respect to the time the input signals (I) reaching the valid state (V) relative to the reference clock t_{FCC} (K) going to the high (H) state or setup time. And t_{TDIXKH} symbolizes the TDM timing (TD) with respect to the time the input signals (I) reach the invalid state (X) relative to the reference clock t_{FCC} (K) going to the high (H) state or hold time.
3. PIO and TIMER inputs and outputs are asynchronous to SYSCLK or any other externally visible clock. PIO/TIMER inputs are internally synchronized to the CPM internal clock. PIO/TIMER outputs should be treated as asynchronous.

Table 34. CPM Output AC Timing Specifications [1]

Characteristic	Symbol [2]	Min	Max	Unit
FCC outputs—internal clock (NMSI) delay	t_{FIKHOX}	1	5.5	ns
FCC outputs—external clock (NMSI) delay	t_{FEKHOX}	2	8	ns
SCC/SMC/SPI outputs—internal clock (NMSI) delay	t_{NIKHOX}	0.5	10	ns
SCC/SMC/SPI outputs—external clock (NMSI) delay	t_{NEKHOX}	2	8	ns
TDM outputs/SI delay	t_{TDKHOX}	2.5	11	ns
PIO outputs delay	t_{PIKHOX}	1	11	ns

Notes:
1. Output specifications are measured from the 50% level of the rising edge of CLKIN to the 50% level of the signal. Timings are measured at the pin.
2. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{FIKHOX} symbolizes the FCC inputs internal timing (FI) for the time t_{FCC} memory clock reference (K) goes from the high state (H) until outputs (O) are invalid (X).

Figure 23 provides the AC test load for the CPM.

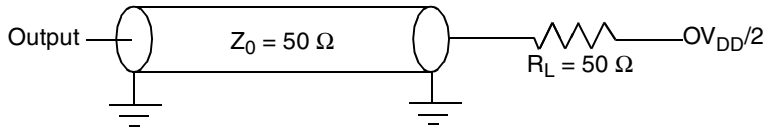
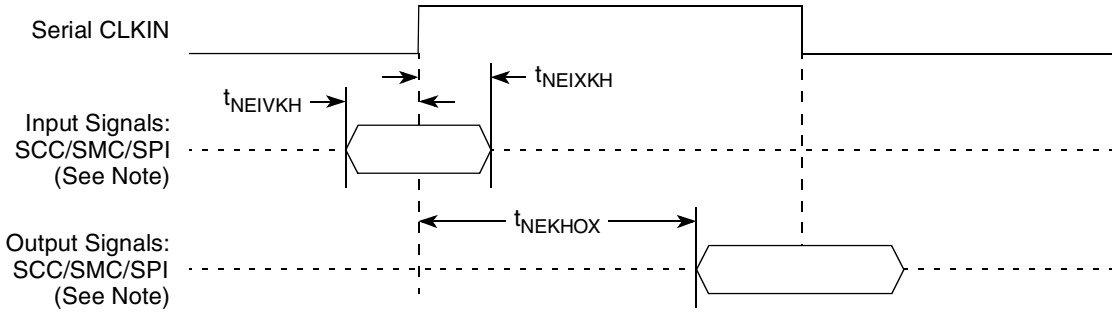


Figure 23. CPM AC Test Load

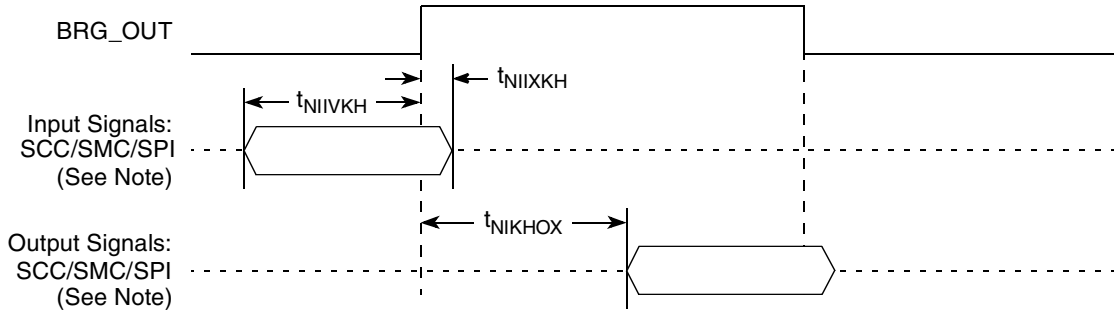
Figure 27 shows the SCC/SMC/SPI external clock.



Note: The clock edge is selectable on SCC and SPI.

Figure 27. SCC/SMC/SPI AC Timing External Clock Diagram

Figure 28 shows the SCC/SMC/SPI internal clock.



Note: The clock edge is selectable on SCC and SPI.

Figure 28. SCC/SMC/SPI AC Timing Internal Clock Diagram

NOTE

[1] SPI AC timings are internal mode when it is master because SPICLK is an output, and external mode when it is slave.

[2] SPI AC timings refer always to SPICLK.

The following two tables are examples of I2C AC parameters at I2C clock value of 100k and 400k respectively.

Table 36. CPM I2C Timing (f_{SCL}=100 kHz)

Characteristic	Expression	Frequency = 100 kHz		Unit
		Min	Max	
SCL clock frequency (slave)	f_{SCL}	—	100	kHz
SCL clock frequency (master)	f_{SCL}	—	100	kHz
Bus free time between transmissions	t_{SDHDL}	4.7	—	µs
Low period of SCL	t_{SCLCH}	4.7	—	µs
High period of SCL	t_{SCHCL}	4	—	µs
Start condition setup time	t_{SCHDL}	2	—	µs
Start condition hold time	t_{SDLCL}	3	—	µs
Data hold time	t_{SCLDX}	2	—	µs
Data setup time	t_{SDVCH}	3	—	µs
SDA/SCL rise time	t_{SRISE}	—	1	µs
SDA/SCL fall time (master)	t_{SFALL}	—	303	ns
Stop condition setup time	t_{SCHDH}	2	—	µs

Table 37. CPM I2C Timing (f_{SCL}=400 kHz)

Characteristic	Expression	Frequency = 400 kHz		Unit
		Min	Max	
SCL clock frequency (slave)	f_{SCL}	—	400	kHz
SCL clock frequency (master)	f_{SCL}	—	400	kHz
Bus free time between transmissions	t_{SDHDL}	1.2	—	µs
Low period of SCL	t_{SCLCH}	1.2	—	µs
High period of SCL	t_{SCHCL}	1	—	µs
Start condition setup time	t_{SCHDL}	420	—	ns
Start condition hold time	t_{SDLCL}	630	—	ns
Data hold time	t_{SCLDX}	420	—	ns
Data setup time	t_{SDVCH}	630	—	ns
SDA/SCL rise time	t_{SRISE}	—	250	ns
SDA/SCL fall time	t_{SFALL}	—	75	ns
Stop condition setup time	t_{SCHDH}	420	—	ns

Figure 40 shows the PCI input AC timing conditions.

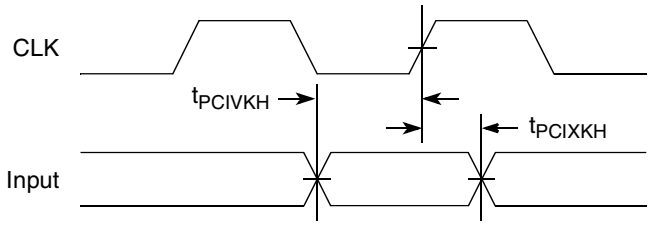


Figure 40. PCI Input AC Timing Measurement Conditions

Figure 41 shows the PCI output AC timing conditions.

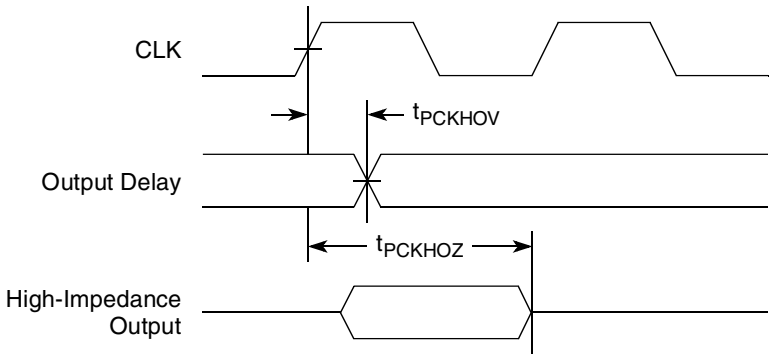


Figure 41. PCI Output AC Timing Measurement Condition

14 Package and Pin Listings

This section details package parameters, pin assignments, and dimensions.

14.1 Package Parameters for the MPC8555E FC-PBGA

The package parameters are as provided in the following list. The package type is 29 mm × 29 mm, 783 flip chip plastic ball grid array (FC-PBGA).

Die size	8.7 mm × 9.3 mm × 0.75 mm
Package outline	29 mm × 29 mm
Interconnects	783
Pitch	1 mm
Minimum module height	3.07 mm
Maximum module height	3.75 mm
Solder Balls	62 Sn/36 Pb/2 Ag
Ball diameter (typical)	0.5 mm

Table 43. MPC8555E Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
PCI2_GNT[1:4]	AD18, AE18, AE19, AD19	O	OV_{DD}	5, 9
PCI2_IDSEL	AC22	I	OV_{DD}	—
PCI2_IRDY	AD20	I/O	OV_{DD}	2
PCI2_PERR	AC20	I/O	OV_{DD}	2
PCI2_REQ[0]	AD21	I/O	OV_{DD}	—
PCI2_REQ[1:4]	AE21, AD22, AE22, AC23	I	OV_{DD}	—
PCI2_SERR	AE20	I/O	OV_{DD}	2,4
PCI2_STOP	AC21	I/O	OV_{DD}	2
PCI2_TRDY	AC19	I/O	OV_{DD}	2
DDR SDRAM Memory Interface				
MDQ[0:63]	M26, L27, L22, K24, M24, M23, K27, K26, K22, J28, F26, E27, J26, J23, H26, G26, C26, E25, C24, E23, D26, C25, A24, D23, B23, F22, J21, G21, G22, D22, H21, E21, N18, J18, D18, L17, M18, L18, C18, A18, K17, K16, C16, B16, G17, L16, A16, L15, G15, E15, C14, K13, C15, D15, E14, D14, D13, E13, D12, A11, F13, H13, A13, B12	I/O	GV_{DD}	—
MECC[0:7]	N20, M20, L19, E19, C21, A21, G19, A19	I/O	GV_{DD}	—
MDM[0:8]	L24, H28, F24, L21, E18, E16, G14, B13, M19	O	GV_{DD}	—
MDQS[0:8]	L26, J25, D25, A22, H18, F16, F14, C13, C20	I/O	GV_{DD}	—
MBA[0:1]	B18, B19	O	GV_{DD}	—
MA[0:14]	N19, B21, F21, K21, M21, C23, A23, B24, H23, G24, K19, B25, D27, J14, J13	O	GV_{DD}	—
MWE	D17	O	GV_{DD}	—
MRAS	F17	O	GV_{DD}	—
MCAS	J16	O	GV_{DD}	—
MCS[0:3]	H16, G16, J15, H15	O	GV_{DD}	—
MCKE[0:1]	E26, E28	O	GV_{DD}	11
MCK[0:5]	J20, H25, A15, D20, F28, K14	O	GV_{DD}	—
MCK[0:5]	F20, G27, B15, E20, F27, L14	O	GV_{DD}	—
MSYNC_IN	M28	I	GV_{DD}	22
MSYNC_OUT	N28	O	GV_{DD}	22
Local Bus Controller Interface				
LA[27]	U18	O	OV_{DD}	5, 9

Table 43. MPC8555E Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
PB[18:31]	P7, P6, P5, P4, P3, P2, P1, R1, R2, R3, R4, R5, R6, R7	I/O	OV_{DD}	—
PC[0, 1, 4–29]	R8, R9, T9, T6, T5, T4, T1, U1, U2, U3, U4, U7, U8, U9, U10, V9, V6, V5, V4, V3, V2, V1, W1, W2, W3, W6, W7, W8	I/O	OV_{DD}	—
PD[7, 14–25, 29–31]	Y4, AA2, AA1, AB1, AB2, AB3, AB5, AB6, AC7, AC4, AC3, AC2, AC1, AD6, AE3, AE2	I/O	OV_{DD}	—

Notes:
1. All multiplexed signals are listed only once and do not re-occur. For example, $\overline{\text{LCS5}}$/$\overline{\text{DMA\_REQ2}}$ is listed only once in the Local Bus Controller Interface section, and is not mentioned in the DMA section even though the pin also functions as $\overline{\text{DMA\_REQ2}}$.
2. Recommend a weak pull-up resistor (2–10 kΩ) be placed on this pin to OV_{DD}.
3. TEST_SEL0 must be pulled-high, TEST_SEL1 must be tied to ground.
4. This pin is an open drain signal.
5. This pin is a reset configuration pin. It has a weak internal pull-up P-FET which is enabled only when the MPC8555E is in the reset state. This pull-up is designed such that it can be overpowered by an external 4.7-kΩ pull-down resistor. If an external device connected to this pin might pull it down during reset, then a pull-up or active driver is needed if the signal is intended to be high during reset.
6. Treat these pins as no connects (NC) unless using debug address functionality.
7. The value of LA[28:31] during reset sets the CCB clock to SYSCLK PLL ratio. These pins require 4.7-kΩ pull-up or pull-down resistors. See Section 15.2, “Platform/System PLL Ratio.”
8. The value of LALE and LGPL2 at reset set the e500 core clock to CCB Clock PLL ratio. These pins require 4.7-kΩ pull-up or pull-down resistors. See the Section 15.3, “e500 Core PLL Ratio.”
9. Functionally, this pin is an output, but structurally it is an I/O because it either samples configuration input during reset or because it has other manufacturing test functions. This pin therefore is described as an I/O for boundary scan.
10. This pin functionally requires a pull-up resistor, but during reset it is a configuration input that controls 32- vs. 64-bit PCI operation. Therefore, it must be actively driven low during reset by reset logic if the device is to be configured to be a 64-bit PCI device. Refer to the *PCI Specification*.
11. This output is actively driven during reset rather than being three-stated during reset.
12. These JTAG pins have weak internal pull-up P-FETs that are always enabled.
13. These pins are connected to the V_{DD}/GND planes internally and may be used by the core power supply to improve tracking and regulation.
14. Internal thermally sensitive resistor.
15. No connections should be made to these pins.
16. These pins are not connected for any functional use.
17. PCI specifications recommend that a weak pull-up resistor (2–10 kΩ) be placed on the higher order pins to OV_{DD} when using 64-bit buffer mode (pins PCI_AD[63:32] and $\overline{\text{PCI2\_C\_BE}}$[7:4]).
18. If this pin is connected to a device that pulls down during reset, an external pull-up is required to that is strong enough to pull this signal to a logic 1 during reset.
19. Recommend a pull-up resistor (~1 kΩ) be placed on this pin to OV_{DD}.
20. These are test signals for factory use only and must be pulled up (100Ω to $1\text{k}\Omega$) to OV_{DD} for normal machine operation.
21. If this signal is used as both an input and an output, a weak pull-up ($\sim 10\text{k}\Omega$) is required on this pin.
22. MSYNC_IN and MSYNC_OUT should be connected together for proper operation.

15.2 Platform/System PLL Ratio

The platform clock is the clock that drives the L2 cache, the DDR SDRAM data rate, and the e500 core complex bus (CCB), and is also called the CCB clock. The values are determined by the binary value on LA[28:31] at power up, as shown in Table 46.

There is no default for this PLL ratio; these signals must be pulled to the desired values.

For specifications on the PCI_CLK, refer to the PCI 2.2 Specification.

Table 46. CCB Clock Ratio

Binary Value of LA[28:31] Signals	Ratio Description
0000	16:1 ratio CCB clock: SYSCLK (PCI bus)
0001	Reserved
0010	2:1 ratio CCB clock: SYSCLK (PCI bus)
0011	3:1 ratio CCB clock: SYSCLK (PCI bus)
0100	4:1 ratio CCB clock: SYSCLK (PCI bus)
0101	5:1 ratio CCB clock: SYSCLK (PCI bus)
0110	6:1 ratio CCB clock: SYSCLK (PCI bus)
0111	Reserved
1000	8:1 ratio CCB clock: SYSCLK (PCI bus)
1001	9:1 ratio CCB clock: SYSCLK (PCI bus)
1010	10:1 ratio CCB clock: SYSCLK (PCI bus)
1011	Reserved
1100	12:1 ratio CCB clock: SYSCLK (PCI bus)
1101	Reserved
1110	Reserved
1111	Reserved

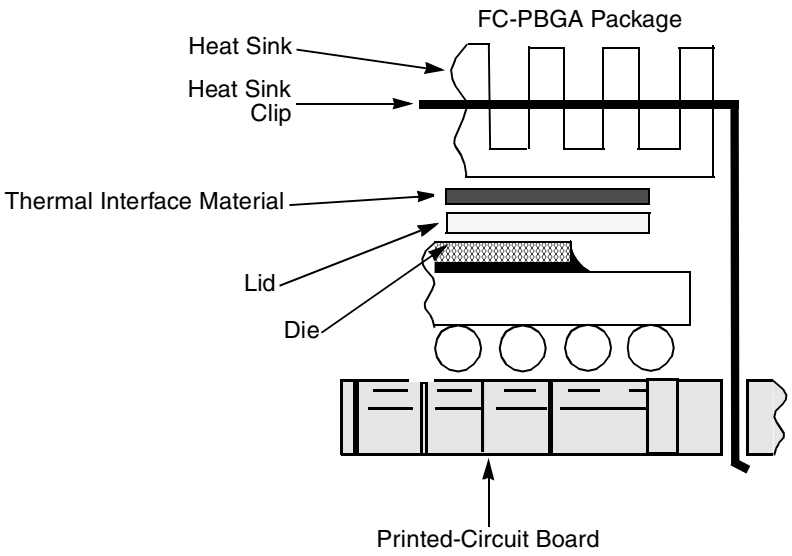


Figure 43. Package Exploded Cross-Sectional View with Several Heat Sink Options

The system board designer can choose between several types of heat sinks to place on the MPC8555E. There are several commercially-available heat sinks from the following vendors:

Aavid Thermalloy 80 Commercial St. Concord, NH 03301 Internet: www.aavidthermalloy.com	603-224-9988
Alpha Novatech 473 Sapena Ct. #15 Santa Clara, CA 95054 Internet: www.alphanovatech.com	408-749-7601
International Electronic Research Corporation (IERC) 413 North Moss St. Burbank, CA 91502 Internet: www.ctscorp.com	818-842-7277
Millennium Electronics (MEI) Loroco Sites 671 East Brokaw Road San Jose, CA 95112 Internet: www.mei-millennium.com	408-436-8770
Tyco Electronics Chip Coolers™ P.O. Box 3668 Harrisburg, PA 17105-3668 Internet: www.chipcoolers.com	800-522-6752
Wakefield Engineering 33 Bridge St. Pelham, NH 03076 Internet: www.wakefield.com	603-635-5102

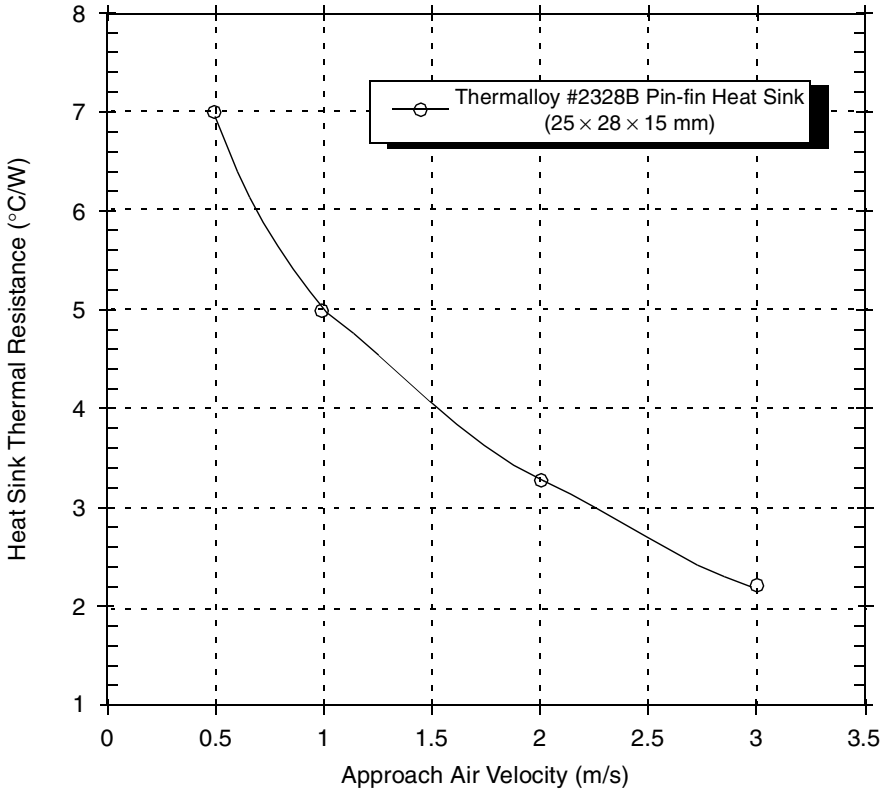


Figure 47. Thermalloy #2328B Heat Sink-to-Ambient Thermal Resistance Versus Airflow Velocity

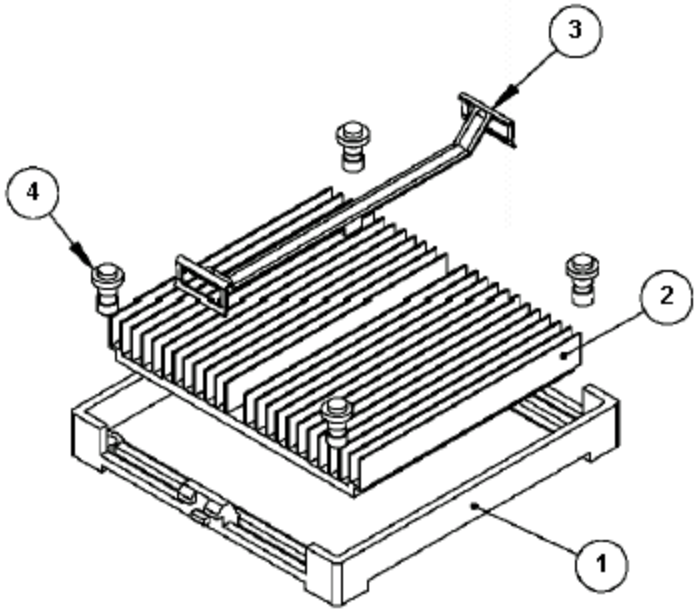
16.2.4.2 Case 2

Every system application has different conditions that the thermal management solution must solve. As an alternate example, assume that the air reaching the component is 85 °C with an approach velocity of 1 m/sec. For a maximum junction temperature of 105 °C at 8 W, the total thermal resistance of junction to case thermal resistance plus thermal interface material plus heat sink thermal resistance must be less than 2.5 °C/W. The value of the junction to case thermal resistance in Table 49 includes the thermal interface resistance of a thin layer of thermal grease as documented in footnote 4 of the table. Assuming that the heat sink is flat enough to allow a thin layer of grease or phase change material, then the heat sink must be less than 1.5 °C/W.

Millennium Electronics (MEI) has tooled a heat sink MTHERM-1051 for this requirement assuming a compactPCI environment at 1 m/sec and a heat sink height of 12 mm. The MEI solution is illustrated in Figure 48 and Figure 49. This design has several significant advantages:

- The heat sink is clipped to a plastic frame attached to the application board with screws or plastic inserts at the corners away from the primary signal routing areas.
- The heat sink clip is designed to apply the force holding the heat sink in place directly above the die at a maximum force of less than 10 lbs.
- For applications with significant vibration requirements, silicone damping material can be applied between the heat sink and plastic frame.

Item No	QTY	MEI PN	Description
1	1	MFRAME-2000	HEATSINK FRAME
2	1	MSNK-1120	EXTRUDED HEATSINK
3	1	MCLIP-1013	CLIP
4	4	MPPINS-1000	FRAME ATTACHMENT PINS



Illustrative source provided by
Millennium Electronics (MEI)

Figure 49. Exploded Views (2) of a Heat Sink Attachment using a Plastic Force

The die junction-to-ambient and the heat sink-to-ambient thermal resistances are common figure-of-merits used for comparing the thermal performance of various microelectronic packaging technologies, one should exercise caution when only using this metric in determining thermal management because no single parameter can adequately describe three-dimensional heat flow. The final die-junction operating temperature is not only a function of the component-level thermal resistance, but the system level design and its operating conditions. In addition to the component's power consumption, a number of factors affect the final operating die-junction temperature: airflow, board population (local heat flux of adjacent components), system air temperature rise, altitude, etc.

Due to the complexity and the many variations of system-level boundary conditions for today's microelectronic equipment, the combined effects of the heat transfer mechanisms (radiation convection and conduction) may vary widely. For these reasons, we recommend using conjugate heat transfer models for the boards, as well as, system-level designs.

When data is held high, SW1 is closed (SW2 is open) and R_P is trimmed until the voltage at the pad equals $OV_{DD}/2$. R_P then becomes the resistance of the pull-up devices. R_P and R_N are designed to be close to each other in value. Then, $Z_0 = (R_P + R_N)/2$.

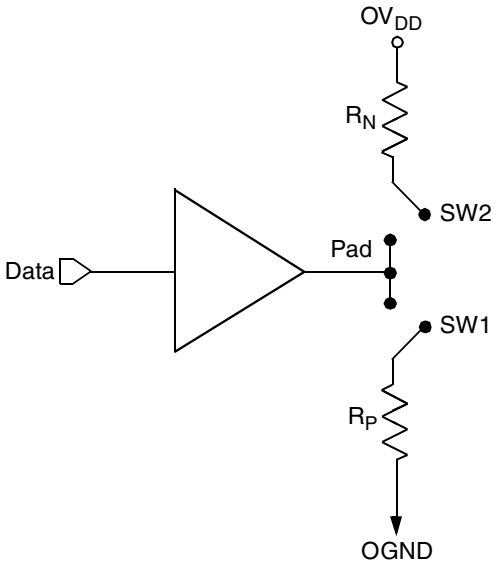


Figure 51. Driver Impedance Measurement

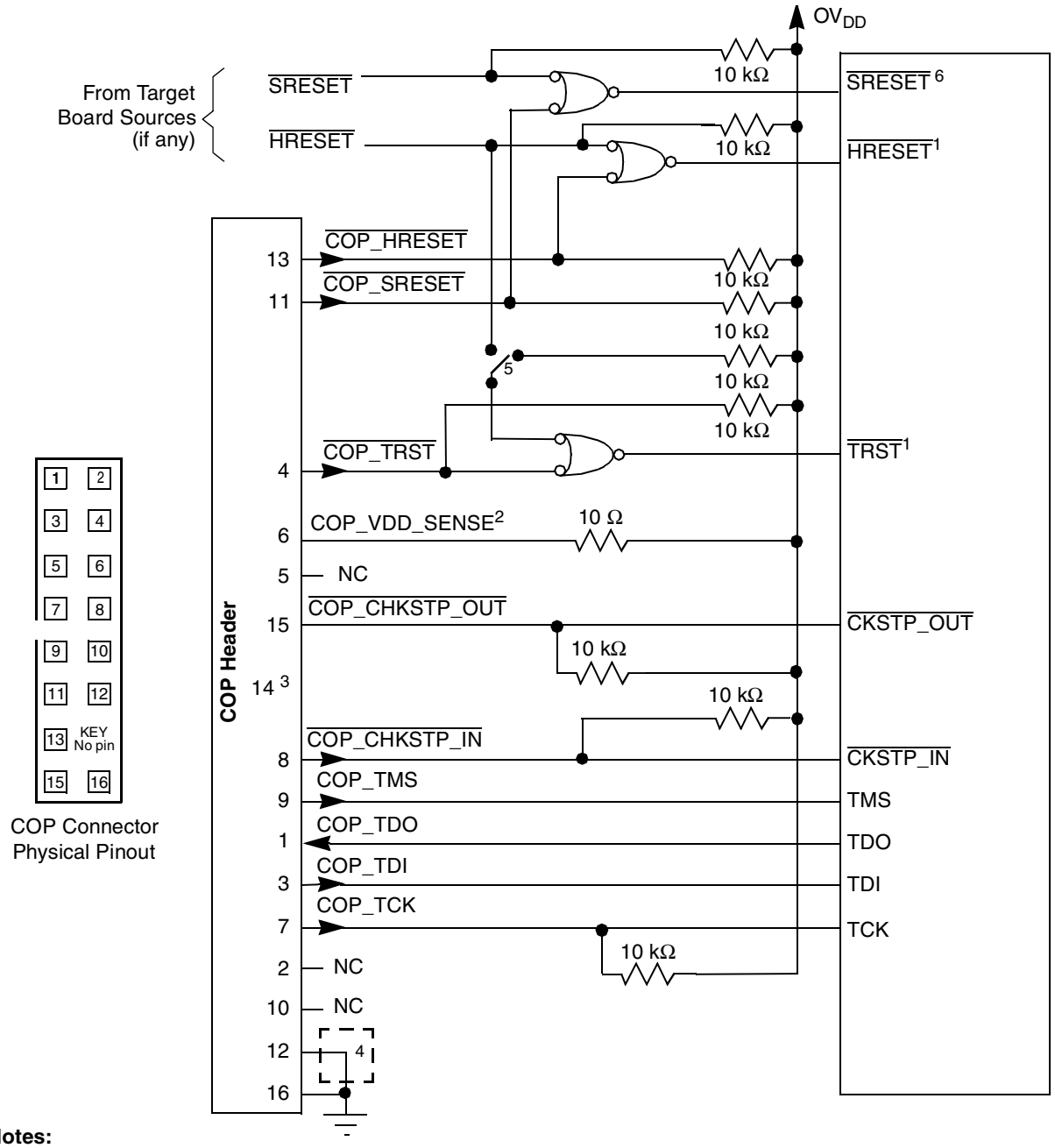
The value of this resistance and the strength of the driver's current source can be found by making two measurements. First, the output voltage is measured while driving logic 1 without an external differential termination resistor. The measured voltage is $V_1 = R_{source} \times I_{source}$. Second, the output voltage is measured while driving logic 1 with an external precision differential termination resistor of value R_{term}. The measured voltage is $V_2 = 1/(1/R_1 + 1/R_2)) \times I_{source}$. Solving for the output impedance gives $R_{source} = R_{term} \times (V_1/V_2 - 1)$. The drive current is then $I_{source} = V_1/R_{source}$.

Table 50 summarizes the signal impedance targets. The driver impedance are targeted at minimum V_{DD}, nominal OV_{DD}, 105°C.

Table 50. Impedance Characteristics

Impedance	Local Bus, Ethernet, DUART, Control, Configuration, Power Management	PCI	DDR DRAM	Symbol	Unit
R_N	43 Target	25 Target	20 Target	Z_0	Ω
R_P	43 Target	25 Target	20 Target	Z_0	Ω
Differential	NA	NA	NA	Z_{DIFF}	Ω

Note: Nominal supply voltages. See Table 1, T_j = 105°C.



Notes:

1. The COP port and target board should be able to independently assert $\overline{\text{HRESET}}$ and $\overline{\text{TRST}}$ to the processor in order to fully control the processor as shown here.
2. Populate this with a 10 Ω resistor for short-circuit/current-limiting protection.
3. The KEY location (pin 14) is not physically present on the COP header.
4. Although pin 12 is defined as a No-Connect, some debug tools may use pin 12 as an additional GND pin for improved signal integrity.
5. This switch is included as a precaution for BSDL testing. The switch should be open during BSDL testing to avoid accidentally asserting the $\overline{\text{TRST}}$ line. If BSDL testing is not being performed, this switch should be closed or removed.
6. Asserting $\overline{\text{SRESET}}$ causes a machine check interrupt to the e500 core.

Figure 53. JTAG Interface Connection

18 Document Revision History

Table 51 provides a revision history for this hardware specification.

Table 51. Document Revision History

Rev. No.	Date	Substantive Change(s)
4.2	1/2008	Added “Note: Rise/Fall Time on CPM Input Pins” and following note text to Section 10.2, “CPM AC Timing Specifications.”
4.1	7/2007	Inserted Figure 3, ““Maximum AC Waveforms on PCI interface for 3.3-V Signaling.”
4	12/2006	Updated Section 2.1.2, “Power Sequencing.” Updated back page information.
3.2	11/2006	Updated Section 2.1.2, “Power Sequencing.” Replaced Section 17.8, “JTAG Configuration Signals.”
3.1	10/2005	Added footnote 2 about junction temperature in Table 4. Added max. power values for 1000 MHz core frequency in Table 4. Removed Figure 3, “Maximum AC Waveforms on PCI Interface for 3.3-V Signaling.” Modified note to $t_{LBKSKEW}$ from 8 to 9 in Table 30. Changed $t_{LBKHOZ1}$ and $t_{LBKHOV2}$ values inTable 30. Added note 3 to $t_{LBKHOV1}$ in Table 30. Modified note 3 in Table 30 and Table 31. Added note 3 to $t_{LBKLOV1}$ in Table 31. Modified values for t_{LBKHKT}, $t_{LBKLOV1}$, $t_{LBKLOV2}$, $t_{LBKLOV3}$, $t_{LBKLOZ1}$, and $t_{LBKLOZ2}$ in Table 31. Changed Input Signals: LAD[0:31]/LDP[0:3] in Figure 21. Modified note for signal CLK_OUT in Table 43. PCI1_CLK and PCI2_CLK changed from I/O to I in Table 43. Added column for Encryption Acceleration in Table 52.
3	8/2005	Modified max. power values in Table 4. Modified notes for signals TSEC1_TXD[3:0], TSEC2_TXD[3:0], TRIG_OUT/READY, MSRCID4, CLK_OUT, and MDVAL in Table 43.
2	8/2005	Previous revision’s history listed incorrect cross references. Table 2 is now correctly listed as Table 27 and Table 38 is now listed as Table 31. Added note 2 in Table 7. Modified min and max values for t_{DDKHMP} in Table 14.
1	6/2005	Changed LV_{dd} to OV_{dd} for the supply voltage Ethernet management interface in Table 27. Modified footnote 4 and changed typical power for the 1000 MHz core frequency inTable 4. Corrected symbols for body rows 9–15, effectively changing them from a high state to a low state in Table 31.
0	6/2005	Initial release.

19 Device Nomenclature

Ordering information for the parts fully covered by this specification document is provided in Section 19.1, “Nomenclature of Parts Fully Addressed by this Document.”

19.1 Nomenclature of Parts Fully Addressed by this Document

Table 52 provides the Freescale part numbering nomenclature for the MPC8555E. Note that the individual part numbers correspond to a maximum processor core frequency. For available frequencies, contact your local Freescale sales office. In addition to the processor frequency, the part numbering scheme also includes an application modifier which may specify special application conditions. Each part number also contains a revision code which refers to the die mask revision number.

Table 52. Part Numbering Nomenclature

MPC	*nnnn*		*t*	*pp*	*aa*	*a*	*r*
Product Code	**Part Identifier**	**Encryption Acceleration**	**Temperature Range[1]**	**Package [2]**	**Processor Frequency [3]**	**Platform Frequency**	**Revision Level[4]**
MPC	8555	Blank = not included E = included	Blank = 0 to 105°C C = –40 to 105°C	PX = FC-PBGA VT = FC-PBGA (lead free)	AJ = 533 MHz AK = 600 MHz AL = 667 MHz AP = 833 MHz AQ = 1000 MHZ	D = 266 MHz E = 300 MHz F = 333 MHz	

Notes:

1. For Temperature Range=C, Processor Frequency is limited to 667 MHz with a Platform Frequency selector of 333 MHz, Processor Frequency is limited to 533 MHz with a Platform Frequency selector of 266 MHz.
2. See Section 14, “Package and Pin Listings,” for more information on available package types.
3. Processor core frequencies supported by parts addressed by this specification only. Not all parts described in this specification support all core frequencies. Additionally, parts addressed by Part Number Specifications may support other maximum core frequencies.
4. Contact you local Freescale field applications engineer (FAE).