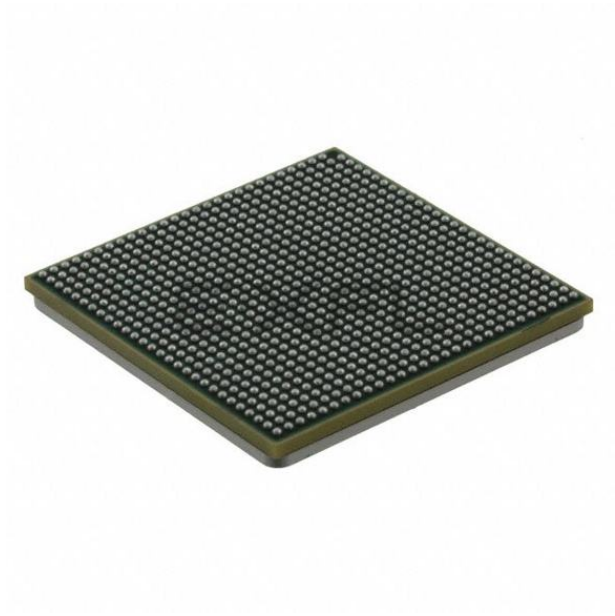

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Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Obsolete
Core Processor	PowerPC e500
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	533MHz
Co-Processors/DSP	Communications; CPM
RAM Controllers	DDR, SDRAM
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100/1000Mbps (2)
SATA	-
USB	USB 2.0 (1)
Voltage - I/O	2.5V, 3.3V
Operating Temperature	0°C ~ 105°C (TA)
Security Features	-
Package / Case	783-BBGA, FCBGA
Supplier Device Package	783-FCPBGA (29x29)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mpc8555vtajd

1 Overview

The following section provides a high-level overview of the MPC8555E features. Figure 1 shows the major functional units within the MPC8555E.

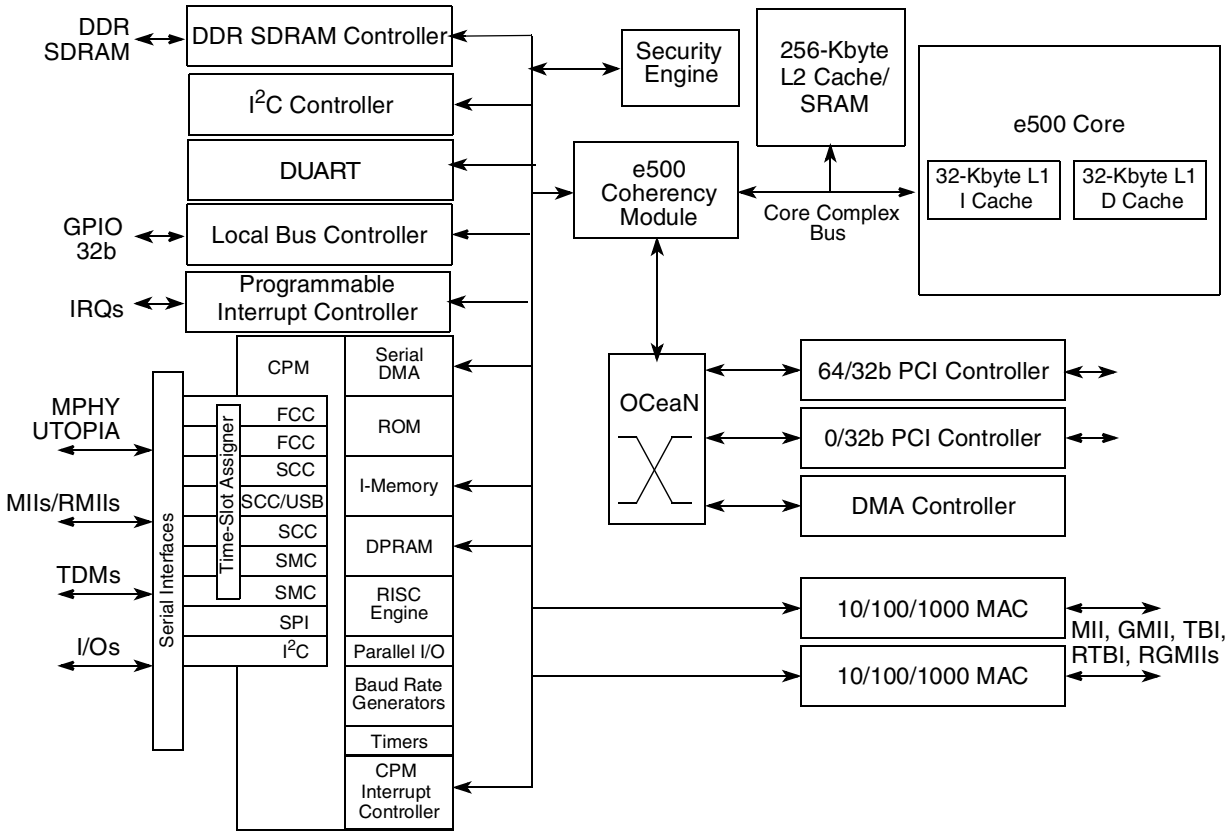

Figure 1. MPC8555E Block Diagram

1.1 Key Features

The following lists an overview of the MPC8555E feature set.

- Embedded e500 Book E-compatible core
  - High-performance, 32-bit Book E-enhanced core that implements the PowerPC architecture
  - Dual-issue superscalar, 7-stage pipeline design
  - 32-Kbyte L1 instruction cache and 32-Kbyte L1 data cache with parity protection
  - Lockable L1 caches—entire cache or on a per-line basis
  - Separate locking for instructions and data
  - Single-precision floating-point operations
  - Memory management unit especially designed for embedded applications
  - Enhanced hardware and software debug support
  - Dynamic power management
  - Performance monitor facility

  - Supports connection of an external interrupt controller such as the 8259 programmable interrupt controller
  - Four global high resolution timers/counters that can generate interrupts
  - Supports additional internal interrupt sources
  - Supports fully nested interrupt delivery
  - Interrupts can be routed to external pin for external processing
  - Interrupts can be routed to the e500 core's standard or critical interrupt inputs
  - Interrupt summary registers allow fast identification of interrupt source
- Two I^2C controllers (one is contained within the CPM, the other is a stand-alone controller which is not part of the CPM)
  - Two-wire interface
  - Multiple master support
  - Master or slave I^2C mode support
  - On-chip digital filtering rejects spikes on the bus
- Boot sequencer
  - Optionally loads configuration data from serial ROM at reset via the stand-alone I^2C interface
  - Can be used to initialize configuration registers and/or memory
  - Supports extended I^2C addressing mode
  - Data integrity checked with preamble signature and CRC
- DUART
  - Two 4-wire interfaces (RXD, TXD, RTS, CTS)
  - Programming model compatible with the original 16450 UART and the PC16550D
- Local bus controller (LBC)
  - Multiplexed 32-bit address and data operating at up to 166 MHz
  - Eight chip selects support eight external slaves
  - Up to eight-beat burst transfers
  - The 32-, 16-, and 8-bit port sizes are controlled by an on-chip memory controller
  - Three protocol engines available on a per chip select basis:
    - General purpose chip select machine (GPCM)
    - Three user programmable machines (UPMs)
    - Dedicated single data rate SDRAM controller
  - Parity support
  - Default boot ROM chip select with configurable bus width (8-, 16-, or 32-bit)
- Two Three-speed (10/100/1000)Ethernet controllers (TSECs)
  - Dual IEEE 802.3, 802.3u, 802.3x, 802.3z AC compliant controllers
  - Support for Ethernet physical interfaces:
    - 10/100/1000 Mbps IEEE 802.3 GMII
    - 10/100 Mbps IEEE 802.3 MII

Table 5. Typical I/O Power Dissipation

Interface	Parameters	GV_{DD} (2.5 V)	OV_{DD} (3.3 V)	LV_{DD} (3.3 V)	LV_{DD} (2.5 V)	Unit	Comments
DDR I/O	CCB = 200 MHz	0.46	—	—	—	W	—
	CCB = 266 MHz	0.59	—	—	—	W	—
	CCB = 300 MHz	0.66	—	—	—	W	—
	CCB = 333 MHz	0.73	—	—	—	W	—
PCI I/O	64b, 66 MHz	—	0.14	—	—	W	—
	64b, 33 MHz	—	0.08	—	—	W	—
	32b, 66 MHz	—	0.07	—	—	W	Multiply by 2 if using two 32b ports
	32b, 33 MHz	—	0.04	—	—	W	
Local Bus I/O	32b, 167 MHz	—	0.30	—	—	W	—
	32b, 133 MHz	—	0.24	—	—	W	—
	32b, 83 MHz	—	0.16	—	—	W	—
	32b, 66 MHz	—	0.13	—	—	W	—
	32b, 33 MHz	—	0.07	—	—	W	—
TSEC I/O	MII	—	—	0.01	—	W	Multiply by number of interfaces used.
	GMII or TBI	—	—	0.07	—	W	
	RGMII or RTBI	—	—	—	0.04	W	
CPM - FCC	MII	—	0.015	—	—	W	—
	RMII	—	0.013	—	—	W	—
	HDLC 16 Mbps	—	0.009	—	—	W	—
	UTOPIA-8 SPHY	—	0.06	—	—	W	—
	UTOPIA-8 MPHY	—	0.1	—	—	W	—
	UTOPIA-16 SPHY	—	0.094	—	—	W	—
	UTOPIA-16 MPHY	—	0.135	—	—	W	—
CPM - SCC	HDLC 16 Mbps	—	0.004	—	—	W	—
TDMA or TDMB	Nibble Mode	—	0.01	—	—	W	—
TDMA or TDMB	Per Channel	—	0.005	—	—	W	Up to 4 TDM channels, multiply by number of TDM channels.

4.3 Real Time Clock Timing

Table 8 provides the real time clock (RTC) AC timing specifications.

Table 8. RTC AC Timing Specifications

Parameter/Condition	Symbol	Min	Typical	Max	Unit	Notes
RTC clock high time	t_{RTCH}	2 x t_{CCB_CLK}	—	—	ns	—
RTC clock low time	t_{RTCL}	2 x t_{CCB_CLK}	—	—	ns	—

5 RESET Initialization

This section describes the AC electrical specifications for the RESET initialization timing requirements of the MPC8555E. Table 9 provides the RESET initialization AC timing specifications.

Table 9. RESET Initialization Timing Specifications

Parameter/Condition	Min	Max	Unit	Notes
Required assertion time of $\overline{\text{HRESET}}$	100	—	µs	—
Minimum assertion time for $\overline{\text{SRESET}}$	512	—	SYSCLKs	1
PLL input setup time with stable SYSCLK before HRESET negation	100	—	µs	—
Input setup time for POR configs (other than PLL config) with respect to negation of $\overline{\text{HRESET}}$	4	—	SYSCLKs	1
Input hold time for POR configs (including PLL config) with respect to negation of $\overline{\text{HRESET}}$	2	—	SYSCLKs	1
Maximum valid-to-high impedance time for actively driven POR configs with respect to negation of $\overline{\text{HRESET}}$	—	5	SYSCLKs	1

Notes:
1. SYSCLK is identical to the PCI_CLK signal and is the primary clock input for the MPC8555E. See the *MPC8555E PowerQUICC™ III Integrated Communications Processor Reference Manual* for more details.

Table 10 provides the PLL and DLL lock times.

Table 10. PLL and DLL Lock Times

Parameter/Condition	Min	Max	Unit	Notes
PLL lock times	—	100	µs	—
DLL lock times	7680	122,880	CCB Clocks	1, 2

Notes:
1. DLL lock times are a function of the ratio between the output clock and the platform (or CCB) clock. A 2:1 ratio results in the minimum and an 8:1 ratio results in the maximum.
2. The CCB clock is determined by the SYSCLK × platform PLL ratio.

Table 14. DDR SDRAM Output AC Timing Specifications for Source Synchronous Mode (continued)

At recommended operating conditions with GV_{DD} of 2.5 V ± 5%.

Parameter	Symbol [1]	Min	Max	Unit	Notes
$\overline{MCS}$(n) output hold with respect to MCK 333 MHz 266 MHz 200 MHz	t_{DDKHCX}	 2.0 2.65 3.8	—	ns	4
MCK to MDQS 333 MHz 266 MHz 200 MHz	t_{DDKHMH}	 –0.9 –1.1 –1.2	 0.3 0.5 0.6	ns	5
MDQ/MECC/MDM output setup with respect to MDQS 333 MHz 266 MHz 200 MHz	t_{DDKHDS}, t_{DDKLDS}	 900 900 1200	—	ps	6
MDQ/MECC/MDM output hold with respect to MDQS 333 MHz 266 MHz 200 MHz	t_{DDKHDX}, t_{DDKLDX}	 900 900 1200	—	ps	6
MDQS preamble start	t_{DDKHMP}	$-0.5 \times t_{MCK} - 0.9$	$-0.5 \times t_{MCK} + 0.3$	ns	7
MDQS epilogue end	t_{DDKLME}	–0.9	0.3	ns	7

Notes:

1. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state) (reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. Output hold time can be read as DDR timing (DD) from the rising or falling edge of the reference clock (KH or KL) until the output went invalid (AX or DX). For example, t_{DDKHAS} symbolizes DDR timing (DD) for the time t_{MCK} memory clock reference (K) goes from the high (H) state until outputs (A) are setup (S) or output valid time. Also, t_{DDKLDX} symbolizes DDR timing (DD) for the time t_{MCK} memory clock reference (K) goes low (L) until data outputs (D) are invalid (X) or data output hold time.
2. All MCK/$\overline{MCK}$ referenced measurements are made from the crossing of the two signals ±0.1 V.
3. In the source synchronous mode, MCK/$\overline{MCK}$ can be shifted in 1/4 applied cycle increments through the Clock Control Register. For the skew measurements referenced for t_{AOSKEW} it is assumed that the clock adjustment is set to align the address/command valid with the rising edge of MCK.
4. ADDR/CMD includes all DDR SDRAM output signals except MCK/$\overline{MCK}$, $\overline{MCS}$, and MDQ/MECC/MDM/MDQS. For the ADDR/CMD setup and hold specifications, it is assumed that the Clock Control register is set to adjust the memory clocks by 1/2 applied cycle. The MCSx pins are separated from the ADDR/CMD (address and command) bus in the HW spec. This was separated because the MCSx pins typically have different loadings than the rest of the address and command bus, even though they have the same timings.
5. Note that t_{DDKHMH} follows the symbol conventions described in note 1. For example, t_{DDKHMH} describes the DDR timing (DD) from the rising edge of the MCK(n) clock (KH) until the MDQS signal is valid (MH). In the source synchronous mode, MDQS can launch later than MCK by 0.3 ns at the maximum. However, MCK may launch later than MDQS by as much as 0.9 ns. t_{DDKHMH} can be modified through control of the DQSS override bits in the TIMING_CFG_2 register. In source synchronous mode, this typically is set to the same delay as the clock adjust in the CLK_CNTL register. The timing parameters listed in the table assume that these two parameters have been set to the same adjustment value. See the *MPC8555E PowerQUICC™ III Integrated Communications Processor Reference Manual* for a description and understanding of the timing modifications enabled by use of these bits.
6. Determined by maximum possible skew between a data strobe (MDQS) and any corresponding bit of data (MDQ), ECC (MECC), or data mask (MDM). The data strobe should be centered inside of the data eye at the pins of the MPC8555E.
7. All outputs are referenced to the rising edge of MCK(n) at the pins of the MPC8555E. Note that t_{DDKHMP} follows the symbol conventions described in note 1.

Figure 6 provides the AC test load for the DDR bus.

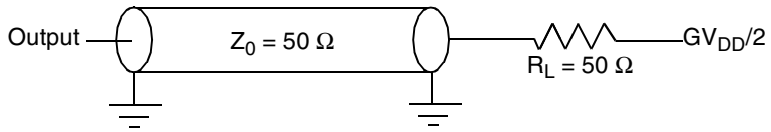

Figure 6. DDR AC Test Load

Table 15. DDR SDRAM Measurement Conditions

Symbol	DDR	Unit	Notes
V_{TH}	MV_{REF} ± 0.31 V	V	1
V_{OUT}	0.5 × GV_{DD}	V	2

Notes:
1. Data input threshold measurement point.
2. Data output measurement point.

7 DUART

This section describes the DC and AC electrical specifications for the DUART interface of the MPC8555E.

7.1 DUART DC Electrical Characteristics

Table 16 provides the DC electrical characteristics for the DUART interface of the MPC8555E.

Table 16. DUART DC Electrical Characteristics

Parameter	Symbol	Test Condition	Min	Max	Unit
High-level input voltage	V_{IH}	$V_{OUT} \geq V_{OH}$ (min) or	2	OV_{DD} + 0.3	V
Low-level input voltage	V_{IL}	$V_{OUT} \leq V_{OL}$ (max)	–0.3	0.8	V
Input current	I_{IN}	V_{IN} [1] = 0 V or $V_{IN} = V_{DD}$	—	±5	μA
High-level output voltage	V_{OH}	OV_{DD} = min, I_{OH} = –100 μA	OV_{DD} – 0.2	—	V
Low-level output voltage	V_{OL}	OV_{DD} = min, I_{OL} = 100 μA	—	0.2	V

Note:
1. Note that the symbol V_{IN}, in this case, represents the OV_{IN} symbol referenced in Table 1 and Table 2.

8.2.2.1 GMII Receive AC Timing Specifications

Table 21 provides the GMII receive AC timing specifications.

Table 21. GMII Receive AC Timing Specifications

At recommended operating conditions with LV_{DD} of 3.3 V ± 5%.

Parameter/Condition	Symbol [1]	Min	Typ	Max	Unit
RX_CLK clock period	t_{GRX}	—	8.0	—	ns
RX_CLK duty cycle	t_{GRXH}/t_{GRX}	40	—	60	%
RXD[7:0], RX_DV, RX_ER setup time to RX_CLK	t_{GRDVKH}	2.0	—	—	ns
RXD[7:0], RX_DV, RX_ER hold time to RX_CLK	t_{GRDXKH}	0.5	—	—	ns
RX_CLK clock rise and fall time	t_{GRXR}, t_{GRXF} [2,3]	—	—	1.0	ns

Note:

1. The symbols used for timing specifications herein follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{GRDVKH} symbolizes GMII receive timing (GR) with respect to the time data input signals (D) reaching the valid state (V) relative to the t_{RX} clock reference (K) going to the high state (H) or setup time. Also, t_{GRDXKL} symbolizes GMII receive timing (GR) with respect to the time data input signals (D) went invalid (X) relative to the t_{GRX} clock reference (K) going to the low (L) state or hold time. Note that, in general, the clock reference symbol representation is based on three letters representing the clock of a particular functional. For example, the subscript of t_{GRX} represents the GMII (G) receive (RX) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).
2. Signal timings are measured at 0.7 V and 1.9 V voltage levels.
3. Guaranteed by design.

Figure 8 provides the AC test load for TSEC.

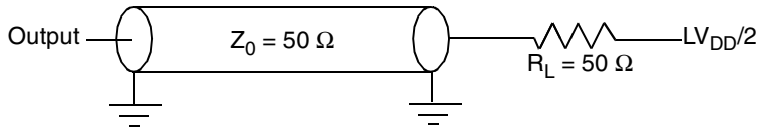

Figure 8. TSEC AC Test Load

Figure 9 shows the GMII receive AC timing diagram.

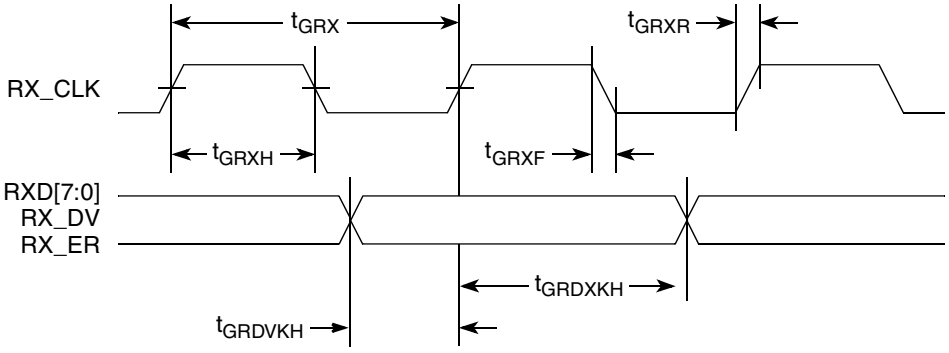

Figure 9. GMII Receive AC Timing Diagram

Table 27. MII Management DC Electrical Characteristics (continued)

Parameter	Symbol	Conditions		Min	Max	Unit
Input high current	I_{IH}	LV_{DD} = Max	V_{IN} [1] = 2.1 V	—	40	μA
Input low current	I_{IL}	LV_{DD} = Max	V_{IN} = 0.5 V	–600	—	μA

Note:
1. Note that the symbol V_{IN}, in this case, represents the OV_{IN} symbol referenced in Table 1 and Table 2.

8.3.2 MII Management AC Electrical Specifications

Table 28 provides the MII management AC timing specifications.

Table 28. MII Management AC Timing Specifications

At recommended operating conditions with LV_{DD} is 3.3 V ± 5%.

Parameter/Condition	Symbol [1]	Min	Typ	Max	Unit	Notes
MDC frequency	f_{MDC}	0.893	—	10.4	MHz	2
MDC period	t_{MDC}	96	—	1120	ns	
MDC clock pulse width high	t_{MDCH}	32	—	—	ns	
MDC to MDIO valid	t_{MDKHDV}			$2*[1/(f_{ccb\_clk}/8)]$	ns	3
MDC to MDIO delay	t_{MDKHDX}	10	—	$2*[1/(f_{ccb\_clk}/8)]$	ns	3
MDIO to MDC setup time	t_{MDDVKH}	5	—	—	ns	
MDIO to MDC hold time	t_{MDDXKH}	0	—	—	ns	
MDC rise time	t_{MDCR}	—	—	10	ns	
MDC fall time	t_{MDHF}	—	—	10	ns	

Notes:
1. The symbols used for timing specifications herein follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{MDKHDX} symbolizes management data timing (MD) for the time t_{MDC} from clock reference (K) high (H) until data outputs (D) are invalid (X) or data hold time. Also, t_{MDDVKH} symbolizes management data timing (MD) with respect to the time data input signals (D) reach the valid state (V) relative to the t_{MDC} clock reference (K) going to the high (H) state or setup time. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).
2. This parameter is dependent on the system clock speed (that is, for a system clock of 267 MHz, the delay is 70 ns and for a system clock of 333 MHz, the delay is 58 ns).
3. This parameter is dependent on the CCB clock speed (that is, for a CCB clock of 267 MHz, the delay is 60 ns and for a CCB clock of 333 MHz, the delay is 48 ns).
4. Guaranteed by design.

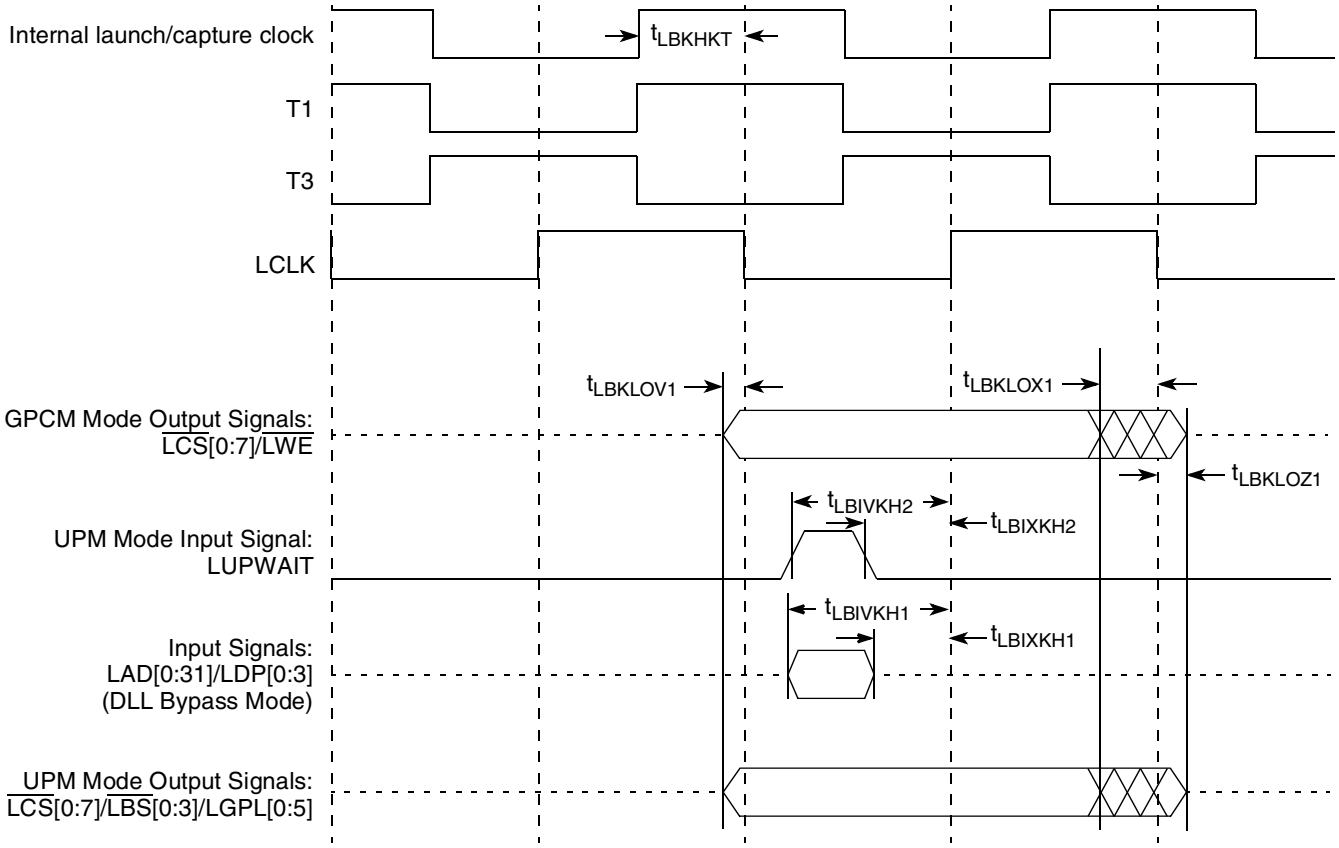

Figure 20. Local Bus Signals, GPCM/UPM Signals for LCCR[CLKDIV] = 2 (DLL Bypass Mode)

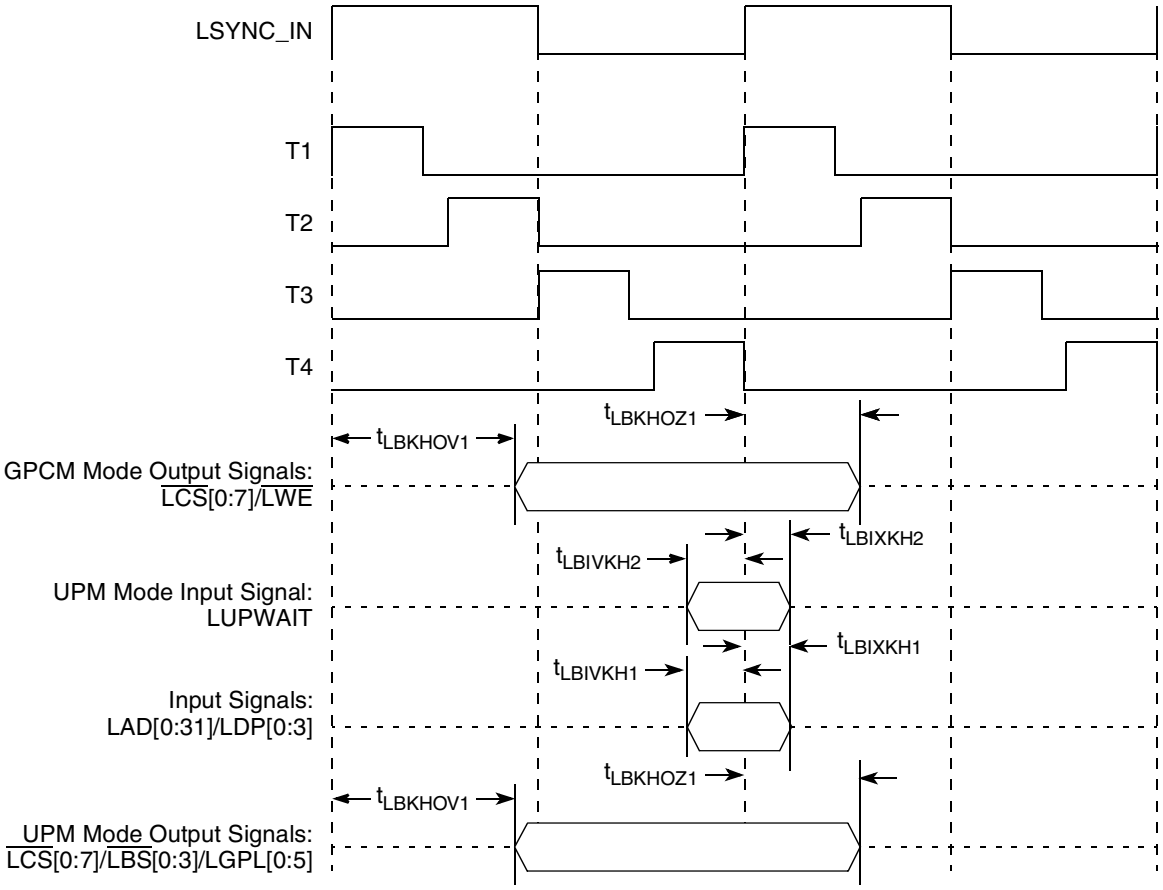

Figure 21. Local Bus Signals, GPCM/UPM Signals for LCCR[CLKDIV] = 4 or 8 (DLL Enabled)

Table 33. CPM Input AC Timing Specifications [1] (continued)

Characteristic	Symbol [2]	Min[3]	Unit
TDM inputs/SI—hold time	t_{TDIXKH}	3	ns
PIO inputs—input setup time	t_{PIIVKH}	8	ns
PIO inputs—input hold time	t_{PIIXKH}	1	ns
COL width high (FCC)	t_{FCCH}	1.5	CLK

Notes:
1. Input specifications are measured from the 50% level of the signal to the 50% level of the rising edge of CLKIN. Timings are measured at the pin.
2. The symbols used for timing specifications herein follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{FIIVKH} symbolizes the FCC inputs internal timing (FI) with respect to the time the input signals (I) reaching the valid state (V) relative to the reference clock t_{FCC} (K) going to the high (H) state or setup time. And t_{TDIXKH} symbolizes the TDM timing (TD) with respect to the time the input signals (I) reach the invalid state (X) relative to the reference clock t_{FCC} (K) going to the high (H) state or hold time.
3. PIO and TIMER inputs and outputs are asynchronous to SYSCLK or any other externally visible clock. PIO/TIMER inputs are internally synchronized to the CPM internal clock. PIO/TIMER outputs should be treated as asynchronous.

Table 34. CPM Output AC Timing Specifications [1]

Characteristic	Symbol [2]	Min	Max	Unit
FCC outputs—internal clock (NMSI) delay	t_{FIKHOX}	1	5.5	ns
FCC outputs—external clock (NMSI) delay	t_{FEKHOX}	2	8	ns
SCC/SMC/SPI outputs—internal clock (NMSI) delay	t_{NIKHOX}	0.5	10	ns
SCC/SMC/SPI outputs—external clock (NMSI) delay	t_{NEKHOX}	2	8	ns
TDM outputs/SI delay	t_{TDKHOX}	2.5	11	ns
PIO outputs delay	t_{PIKHOX}	1	11	ns

Notes:
1. Output specifications are measured from the 50% level of the rising edge of CLKIN to the 50% level of the signal. Timings are measured at the pin.
2. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{FIKHOX} symbolizes the FCC inputs internal timing (FI) for the time t_{FCC} memory clock reference (K) goes from the high state (H) until outputs (O) are invalid (X).

Figure 23 provides the AC test load for the CPM.

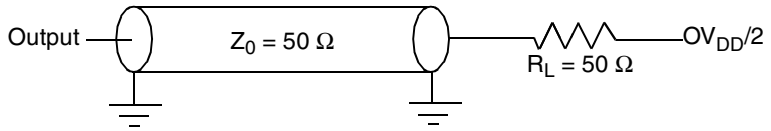

Figure 23. CPM AC Test Load

Table 35. I2C Timing (continued)

Characteristic	Expression	All Frequencies		Unit
		Min	Max	
SDA/SCL fall time	t_{SFALL}	-	$1/(33 * f_{SCL})$	s
Stop condition setup time	t_{SCHDH}	$2/(\text{divider} * f_{SCL})$	-	s

Notes:
1. F_{MAX} = BRGCLK/(min_divider*prescale. Where prescaler=25-I2MODE[PDIV]; and min_divider=12 if digital filter disabled and 18 if enabled.
   Example #1: if I2MODE[PDIV]=11 (prescaler=4) and I2MODE[FLT]=0 (digital filter disabled) then FMAX=BRGCLK/48
   Example #2: if I2MODE[PDIV]=00 (prescaler=32) and I2MODE[FLT]=1 (digital filter enabled) then FMAX=BRGCLK/576
2. divider = f_{SCL}/prescaler.
   In master mode: divider=BRGCLK/(f_{SCL}*prescaler)=2*(I2BRG[DIV]+3)
   In slave mode: divider=BRGCLK/(f_{SCL}*prescaler)

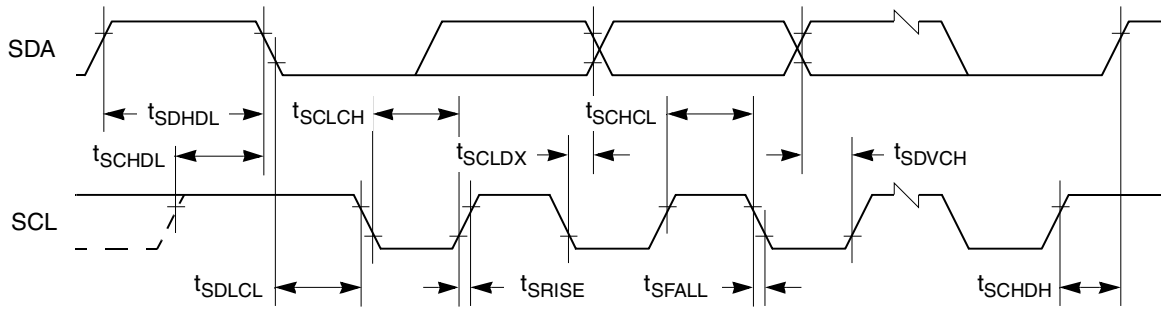

Figure 31. CPM I2C Bus Timing Diagram

Figure 32 provides the AC test load for TDO and the boundary-scan outputs of the MPC8555E.

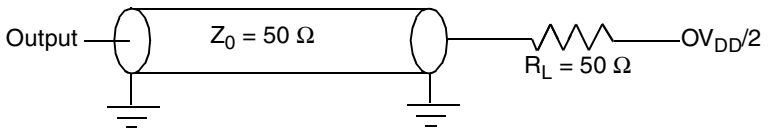

Figure 32. AC Test Load for the JTAG Interface

Figure 33 provides the JTAG clock input timing diagram.

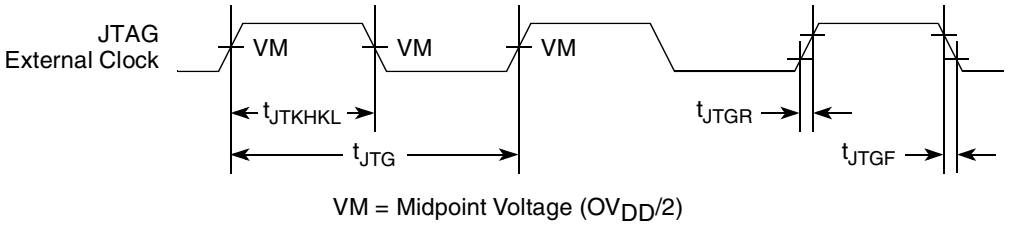

Figure 33. JTAG Clock Input Timing Diagram

Figure 34 provides the $\overline{\text{TRST}}$ timing diagram.

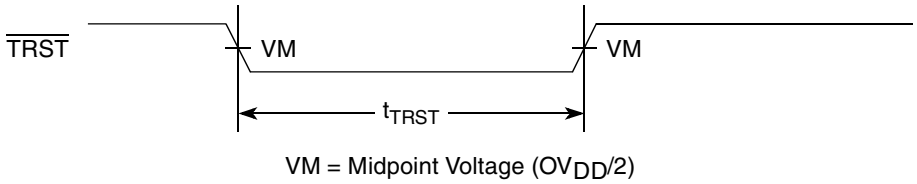

Figure 34. $\overline{\text{TRST}}$ Timing Diagram

Figure 35 provides the boundary-scan timing diagram.

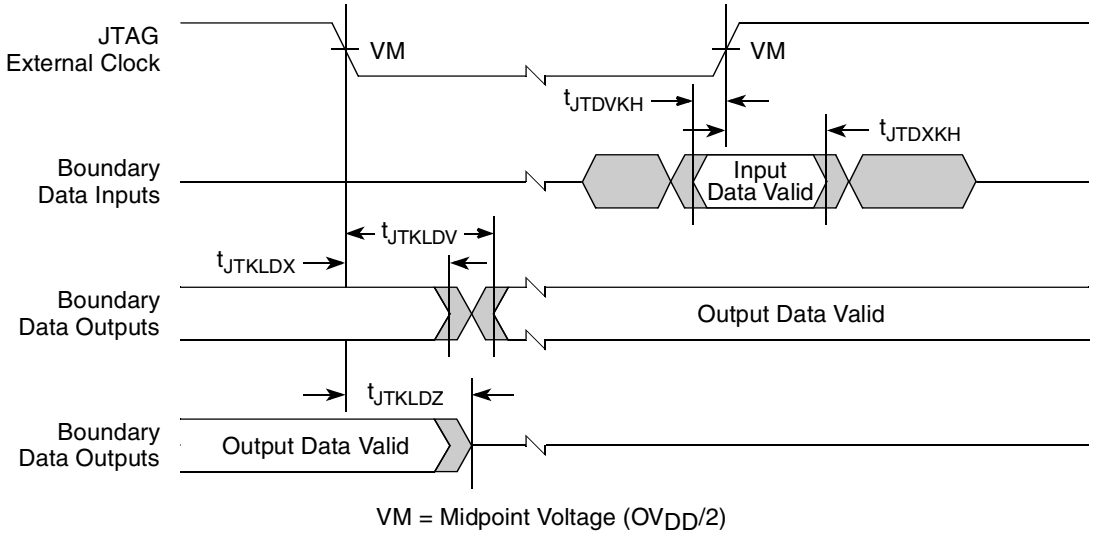

Figure 35. Boundary-Scan Timing Diagram

14.2 Mechanical Dimensions of the FC-PBGA

Figure 42 the mechanical dimensions and bottom surface nomenclature of the MPC8555E 783 FC-PBGA package.

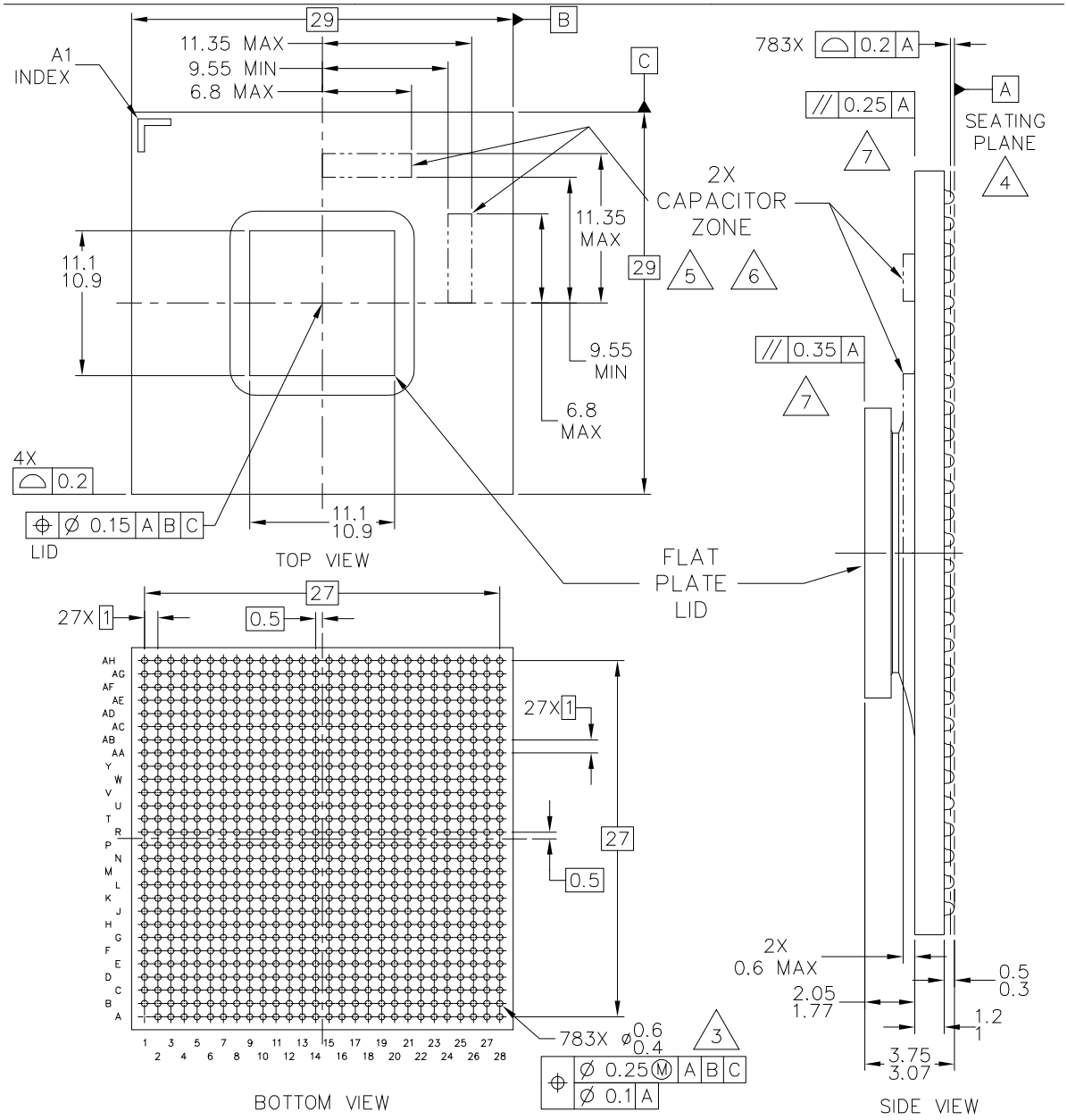

Figure 42. Mechanical Dimensions and Bottom Surface Nomenclature of the FC-PBGA

Notes:

1. All dimensions are in millimeters.
2. Dimensions and tolerances per ASME Y14.5M-1994.
3. Maximum solder ball diameter measured parallel to datum A.
4. Datum A, the seating plane, is defined by the spherical crowns of the solder balls.
5. Capacitors may not be present on all devices.
6. Caution must be taken not to short capacitors or exposed metal capacitor pads on package top.
7. The socket lid must always be oriented to A1.

15.3 e500 Core PLL Ratio

Table 47 describes the clock ratio between the e500 core complex bus (CCB) and the e500 core clock. This ratio is determined by the binary value of LALE and LGPL2 at power up, as shown in Table 47.

Table 47. e500 Core to CCB Ratio

Binary Value of LALE, LGPL2 Signals	Ratio Description
00	2:1 e500 core:CCB
01	5:2 e500 core:CCB
10	3:1 e500 core:CCB
11	7:2 e500 core:CCB

15.4 Frequency Options

Table 48 shows the expected frequency values for the platform frequency when using a CCB to SYSCLK ratio in comparison to the memory bus speed.

Table 48. Frequency Options with Respect to Memory Bus Speeds

CCB to SYSCLK Ratio	SYSCLK (MHz)								
	17	25	33	42	67	83	100	111	133
	Platform/CCB Frequency (MHz)								
2							200	222	267
3					200	250	300	333	
4					267	333			
5				208	333				
6			200	250					
8		200	267	333					
9		225	300						
10		250	333						
12	200	300							
16	267								

the heat sink should be slowly removed. Heating the heat sink to 40–50°C with an air gun can soften the interface material and make the removal easier. The use of an adhesive for heat sink attach is not recommended.

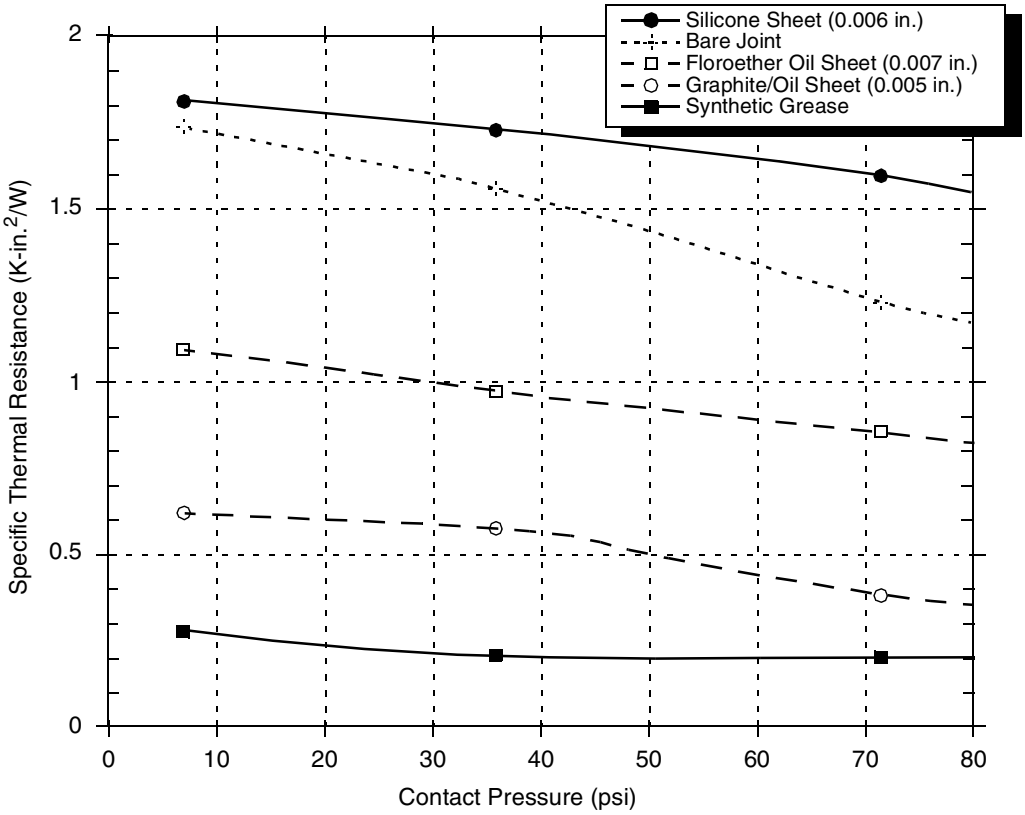

Figure 46. Thermal Performance of Select Thermal Interface Materials

The system board designer can choose between several types of thermal interface. There are several commercially-available thermal interfaces provided by the following vendors:

Chomerics, Inc. 77 Dragon Ct. Woburn, MA 01888-4014 Internet: www.chomerics.com	781-935-4850
Dow-Corning Corporation Dow-Corning Electronic Materials 2200 W. Salzburg Rd. Midland, MI 48686-0997 Internet: www.dowcorning.com	800-248-2481
Shin-Etsu MicroSi, Inc. 10028 S. 51st St. Phoenix, AZ 85044 Internet: www.microsi.com	888-642-7674
The Bergquist Company 18930 West 78th St.	800-347-4572

Chanhassen, MN 55317
Internet: www.bergquistcompany.com

Thermagon Inc. 888-246-9050
4707 Detroit Ave.
Cleveland, OH 44102
Internet: www.thermagon.com

16.2.4 Heat Sink Selection Examples

The following section provides a heat sink selection example using one of the commercially available heat sinks.

16.2.4.1 Case 1

For preliminary heat sink sizing, the die-junction temperature can be expressed as follows:

$$T_J = T_I + T_R + (\theta_{JC} + \theta_{INT} + \theta_{SA}) \times P_D$$

where

T_J is the die-junction temperature

T_I is the inlet cabinet ambient temperature

T_R is the air temperature rise within the computer cabinet

θ_{JC} is the junction-to-case thermal resistance

θ_{INT} is the adhesive or interface material thermal resistance

θ_{SA} is the heat sink base-to-ambient thermal resistance

P_D is the power dissipated by the device. See Table 4 and Table 5.

During operation the die-junction temperatures (T_J) should be maintained within the range specified in Table 2. The temperature of air cooling the component greatly depends on the ambient inlet air temperature and the air temperature rise within the electronic cabinet. An electronic cabinet inlet-air temperature (T_A) may range from 30° to 40°C. The air temperature rise within a cabinet (T_R) may be in the range of 5° to 10°C. The thermal resistance of some thermal interface material (θ_{INT}) may be about 1°C/W. For the purposes of this example, the θ_{JC} value given in Table 49 that includes the thermal grease interface and is documented in note 4 is used. If a thermal pad is used, θ_{INT} must be added.

Assuming a T_I of 30°C, a T_R of 5°C, a FC-PBGA package θ_{JC} = 0.96, and a power consumption (P_D) of 8.0 W, the following expression for T_J is obtained:

$$\text{Die-junction temperature: } T_J = 30°C + 5°C + (0.96°C/W + \theta_{SA}) \times 8.0\ W$$

The heat sink-to-ambient thermal resistance (θ_{SA}) versus airflow velocity for a Thermalloy heat sink #2328B is shown in Figure 47.

Assuming an air velocity of 2 m/s, we have an effective θ_{SA+} of about 3.3°C/W, thus

$$T_J = 30°C + 5°C + (0.96°C/W + 3.3°C/W) \times 8.0\ W,$$

resulting in a die-junction temperature of approximately 69°C which is well within the maximum operating temperature of the component.

17 System Design Information

This section provides electrical and thermal design recommendations for successful application of the MPC8555E.

17.1 System Clocking

The MPC8555E includes five PLLs.

1. The platform PLL ($AV_{DD}1$) generates the platform clock from the externally supplied SYSCLK input. The frequency ratio between the platform and SYSCLK is selected using the platform PLL ratio configuration bits as described in Section 15.2, "Platform/System PLL Ratio."
2. The e500 Core PLL ($AV_{DD}2$) generates the core clock as a slave to the platform clock. The frequency ratio between the e500 core clock and the platform clock is selected using the e500 PLL ratio configuration bits as described in Section 15.3, "e500 Core PLL Ratio."
3. The CPM PLL ($AV_{DD}3$) is slaved to the platform clock and is used to generate clocks used internally by the CPM block. The ratio between the CPM PLL and the platform clock is fixed and not under user control.
4. The PCI1 PLL ($AV_{DD}4$) generates the clocking for the first PCI bus.
5. The PCI2 PLL ($AV_{DD}5$) generates the clock for the second PCI bus.

17.2 PLL Power Supply Filtering

Each of the PLLs listed above is provided with power through independent power supply pins ($AV_{DD}1$, $AV_{DD}2$, $AV_{DD}3$, $AV_{DD}4$, and $AV_{DD}5$ respectively). The AV_{DD} level should always be equivalent to V_{DD}, and preferably these voltages are derived directly from V_{DD} through a low frequency filter scheme such as the following.

There are a number of ways to reliably provide power to the PLLs, but the recommended solution is to provide five independent filter circuits as illustrated in Figure 50, one to each of the five AV_{DD} pins. By providing independent filters to each PLL the opportunity to cause noise injection from one PLL to the other is reduced.

This circuit is intended to filter noise in the PLLs resonant frequency range from a 500 kHz to 10 MHz range. It should be built with surface mount capacitors with minimum Effective Series Inductance (ESL). Consistent with the recommendations of Dr. Howard Johnson in *High Speed Digital Design: A Handbook of Black Magic* (Prentice Hall, 1993), multiple small capacitors of equal value are recommended over a single large value capacitor.

Each circuit should be placed as close as possible to the specific AV_{DD} pin being supplied to minimize noise coupled from nearby circuits. It should be possible to route directly from the capacitors to the AV_{DD} pin, which is on the periphery of the 783 FC-PBGA footprint, without the inductance of vias.

Figure 50 shows the PLL power supply filter circuit.

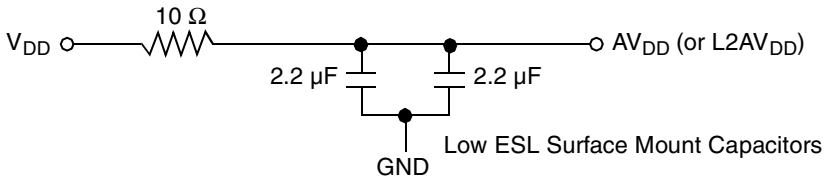

Figure 50. PLL Power Supply Filter Circuit

17.3 Decoupling Recommendations

Due to large address and data buses, and high operating frequencies, the MPC8555E can generate transient power surges and high frequency noise in its power supply, especially while driving large capacitive loads. This noise must be prevented from reaching other components in the MPC8555E system, and the MPC8555E itself requires a clean, tightly regulated source of power. Therefore, it is recommended that the system designer place at least one decoupling capacitor at each V_{DD}, OV_{DD}, GV_{DD}, and LV_{DD} pins of the MPC8555E. These decoupling capacitors should receive their power from separate V_{DD}, OV_{DD}, GV_{DD}, LV_{DD}, and GND power planes in the PCB, utilizing short traces to minimize inductance. Capacitors may be placed directly under the device using a standard escape pattern. Others may surround the part.

These capacitors should have a value of 0.01 or 0.1 µF. Only ceramic SMT (surface mount technology) capacitors should be used to minimize lead inductance, preferably 0402 or 0603 sizes.

In addition, it is recommended that there be several bulk storage capacitors distributed around the PCB, feeding the V_{DD}, OV_{DD}, GV_{DD}, and LV_{DD} planes, to enable quick recharging of the smaller chip capacitors. These bulk capacitors should have a low ESR (equivalent series resistance) rating to ensure the quick response time necessary. They should also be connected to the power and ground planes through two vias to minimize inductance. Suggested bulk capacitors—100–330 µF (AVX TPS tantalum or Sanyo OSCON).

17.4 Connection Recommendations

To ensure reliable operation, it is highly recommended to connect unused inputs to an appropriate signal level. Unused active low inputs should be tied to OV_{DD}, GV_{DD}, or LV_{DD} as required. Unused active high inputs should be connected to GND. All NC (no-connect) signals must remain unconnected.

Power and ground connections must be made to all external V_{DD}, GV_{DD}, LV_{DD}, OV_{DD}, and GND pins of the MPC8555E.

17.5 Output Buffer DC Impedance

The MPC8555E drivers are characterized over process, voltage, and temperature. For all buses, the driver is a push-pull single-ended driver type (open drain for I^2C).

To measure Z_0 for the single-ended drivers, an external resistor is connected from the chip pad to OV_{DD} or GND. Then, the value of each resistor is varied until the pad voltage is $OV_{DD}/2$ (see Figure 51). The output impedance is the average of two components, the resistances of the pull-up and pull-down devices.

19 Device Nomenclature

Ordering information for the parts fully covered by this specification document is provided in Section 19.1, “Nomenclature of Parts Fully Addressed by this Document.”

19.1 Nomenclature of Parts Fully Addressed by this Document

Table 52 provides the Freescale part numbering nomenclature for the MPC8555E. Note that the individual part numbers correspond to a maximum processor core frequency. For available frequencies, contact your local Freescale sales office. In addition to the processor frequency, the part numbering scheme also includes an application modifier which may specify special application conditions. Each part number also contains a revision code which refers to the die mask revision number.

Table 52. Part Numbering Nomenclature

MPC	*nnnn*		*t*	*pp*	*aa*	*a*	*r*
Product Code	**Part Identifier**	**Encryption Acceleration**	**Temperature Range[1]**	**Package [2]**	**Processor Frequency [3]**	**Platform Frequency**	**Revision Level[4]**
MPC	8555	Blank = not included E = included	Blank = 0 to 105°C C = –40 to 105°C	PX = FC-PBGA VT = FC-PBGA (lead free)	AJ = 533 MHz AK = 600 MHz AL = 667 MHz AP = 833 MHz AQ = 1000 MHZ	D = 266 MHz E = 300 MHz F = 333 MHz	

Notes:

1. For Temperature Range=C, Processor Frequency is limited to 667 MHz with a Platform Frequency selector of 333 MHz, Processor Frequency is limited to 533 MHz with a Platform Frequency selector of 266 MHz.
2. See Section 14, “Package and Pin Listings,” for more information on available package types.
3. Processor core frequencies supported by parts addressed by this specification only. Not all parts described in this specification support all core frequencies. Additionally, parts addressed by Part Number Specifications may support other maximum core frequencies.
4. Contact you local Freescale field applications engineer (FAE).