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Understanding Embedded - DSP (Digital Signal Processors)

Embedded - DSP (Digital Signal Processors) are specialized microprocessors designed to perform complex mathematical computations on digital signals in real-time. Unlike general-purpose processors, DSPs are optimized for high-speed numeric processing tasks, making them ideal for applications that require efficient and precise manipulation of digital data. These processors are fundamental in converting and processing signals in various forms, including audio, video, and communication signals, ensuring that data is accurately interpreted and utilized in embedded systems.

Applications of Embedded - DSP (Digital Signal Processors)

Details	
Product Status	Active
Type	SC3850 Single Core
Interface	Ethernet, I²C, PCI, RGMII, Serial RapidIO, SGMII, SPI, UART/USART
Clock Rate	1GHz
Non-Volatile Memory	ROM (96kB)
On-Chip RAM	576kB
Voltage - I/O	2.50V
Voltage - Core	1.00V
Operating Temperature	-40°C ~ 105°C
Mounting Type	Surface Mount
Package / Case	783-BBGA, FCBGA
Supplier Device Package	783-FCPBGA (29x29)
Purchase URL	https://www.e-xfl.com/pro/item?MUrl=&PartUrl=msc8151tag1000b

1 Pin Assignment

This section includes diagrams of the MSC8151 package ball grid array layouts and tables showing how the pinouts are allocated for the package.

1.1 FC-PBGA Ball Layout Diagram

The top view of the FC-PBGA package is shown in Figure 3 with the ball location index numbers.

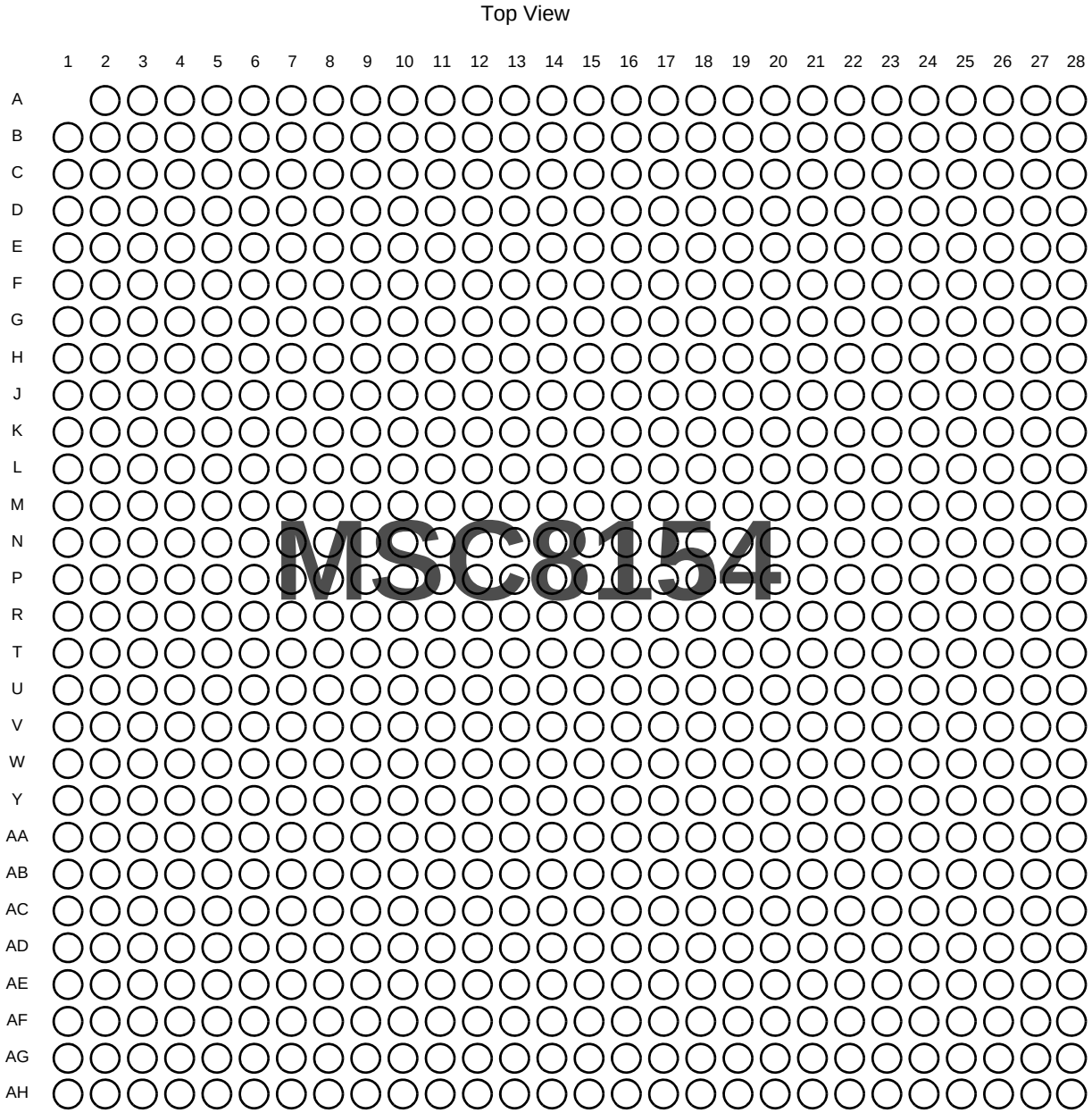


Figure 3. MSC8151 FC-PBGA Package, Top View

Table 1. Signal List by Ball Number (continued)

Ball Number	Signal Name[1,2]	Pin Type[10]	Power Rail Name
G7	M2CKE0	O	GVDD2
G8	M2A11	O	GVDD2
G9	M2A7	O	GVDD2
G10	$\overline{\text{M2CK2}}$	O	GVDD2
G11	M2APAR_OUT	O	GVDD2
G12	M2ODT1	O	GVDD2
G13	$\overline{\text{M2APAR\_IN}}$	I	GVDD2
G14	M2DQ43	I/O	GVDD2
G15	M2DM5	O	GVDD2
G16	M2DQ44	I/O	GVDD2
G17	M2DQ40	I/O	GVDD2
G18	M2DQ59	I/O	GVDD2
G19	M2DM7	O	GVDD2
G20	M2DQ60	I/O	GVDD2
G21	Reserved	NC	—
G22	Reserved	NC	—
G23	SXPVSS1	Ground	N/A
G24	SXPVDD1	Power	N/A
G25	SR1_IMP_CAL_TX	I	SXCVDD1
G26	SXCVSS1	Ground	N/A
G27	Reserved	NC	—
G28	Reserved	NC	—
H1	GVDD2	Power	N/A
H2	VSS	Ground	N/A
H3	M2DQ18	I/O	GVDD2
H4	GVDD2	Power	N/A
H5	VSS	Ground	N/A
H6	M2DQ20	I/O	GVDD2
H7	GVDD2	Power	N/A
H8	VSS	Ground	N/A
H9	M2A15	O	GVDD2
H10	M2CK2	O	GVDD2
H11	M2MDIC0	I/O	GVDD2
H12	M2VREF	I	GVDD2
H13	M2MDIC1	I/O	GVDD2
H14	M2DQ46	I/O	GVDD2
H15	M2DQ47	I/O	GVDD2
H16	M2DQ45	I/O	GVDD2
H17	M2DQ41	I/O	GVDD2
H18	M2DQ62	I/O	GVDD2
H19	M2DQ63	I/O	GVDD2
H20	M2DQ61	I/O	GVDD2
H21	Reserved	NC	—
H22	Reserved	NC	—
H23	SR1_TXD3/SG2_TX[4]	O	SXPVDD1
H24	$\overline{\text{SR1\_TXD3}}$/$\overline{\text{SG2\_TX}}$[4]	O	SXPVDD1

Table 1. Signal List by Ball Number (continued)

Ball Number	Signal Name[1,2]	Pin Type[10]	Power Rail Name
K15	VDD	Power	N/A
K16	VSS	Ground	N/A
K17	VSS	Ground	N/A
K18	VSS	Ground	N/A
K19	VDD	Power	N/A
K20	Reserved	NC	—
K21	Reserved	NC	—
K22	Reserved	NC	—
K23	SXPVDD2	Power	N/A
K24	SXPVSS2	Ground	N/A
K25	SXCVDD2	Power	N/A
K26	SXCVSS2	Ground	N/A
K27	SXCVDD2	Power	N/A
K28	SXCVSS2	Ground	N/A
L1	M2DQ9	I/O	GVDD2
L2	M2DQ12	I/O	GVDD2
L3	M2DQ13	I/O	GVDD2
L4	$\overline{\text{M2DQS0}}$	I/O	GVDD2
L5	M2DQS0	I/O	GVDD2
L6	M2DM0	O	GVDD2
L7	M2DQ3	I/O	GVDD2
L8	M2DQ2	I/O	GVDD2
L9	M2DQ4	I/O	GVDD2
L10	VDD	Power	N/A
L11	VSS	Ground	N/A
L12	M3VDD	Power	N/A
L13	VSS	Ground	N/A
L14	VSS	Ground	N/A
L15	VSS	Ground	N/A
L16	VSS	Ground	N/A
L17	VSS	Ground	N/A
L18	VDD	Power	N/A
L19	VSS	Ground	N/A
L20	Reserved	NC	—
L21	Reserved	NC	—
L22	Reserved	NC	—
L23	SR2_TXD3/PE_TXD3/SG2_TX[4]	O	SXPVDD2
L24	$\overline{\text{SR2\_TXD3}}/\overline{\text{PE\_TXD3}}/\overline{\text{SG2\_TX}}$[4]	O	SXPVDD2
L25	SXCVSS2	Ground	N/A
L26	SXCVDD2	Power	N/A
L27	$\overline{\text{SR2\_RXD3}}/\overline{\text{PE\_RXD3}}/\overline{\text{SG2\_RX}}$[4]	I	SXCVDD2
L28	SR2_RXD3/PE_RXD3/SG2_RX[4]	I	SXCVDD2
M1	M2DQ8	I/O	GVDD2
M2	VSS	Ground	N/A
M3	GVDD2	Power	N/A
M4	M2DQ15	I/O	GVDD2

Table 1. Signal List by Ball Number (continued)

Ball Number	Signal Name[1,2]	Pin Type[10]	Power Rail Name
N23	SR2_TXD2/PE_TXD2/SG1_TX[4]	O	SXPVDD2
N24	$\overline{\text{SR2\_TXD2}}$/$\overline{\text{PE\_TXD2}}$/$\overline{\text{SG1\_TX}}$[4]	O	SXPVDD2
N25	SXCVDD2	Power	N/A
N26	SXCVSS2	Ground	N/A
N27	$\overline{\text{SR2\_RXD2}}$/$\overline{\text{PE\_RXD2}}$/$\overline{\text{SG1\_RX}}$[4]	I	SXCVDD2
N28	SR2_RXD2/PE_RXD2/SG1_RX[4]	I	SXCVDD2
P1	CLKIN	I	QVDD
P2	EE0	I	QVDD
P3	QVDD	Power	N/A
P4	VSS	Ground	N/A
P5	STOP_BS	I	QVDD
P6	QVDD	Power	N/A
P7	VSS	Ground	N/A
P8	PLL0_AVDD[9]	Power	VDD
P9	PLL2_AVDD[9]	Power	VDD
P10	VSS	Ground	N/A
P11	VDD	Power	N/A
P12	VSS	Ground	N/A
P13	VDD	Power	N/A
P14	VSS	Ground	N/A
P15	MVDD	Power	N/A
P16	VSS	Ground	N/A
P17	MVDD	Power	N/A
P18	VSS	Ground	N/A
P19	VDD	Power	N/A
P20	Reserved	NC	—
P21	Reserved	NC	—
P22	Reserved	NC	—
P23	SXPVDD2	Power	N/A
P24	SXPVSS2	Ground	N/A
P25	SR2_PLL_AGND[9]	Ground	SXCVSS2
P26	SR2_PLL_AVDD[9]	Power	SXCVDD2
P27	SXCVSS2	Ground	N/A
P28	SXCVDD2	Power	N/A
R1	VSS	Ground	N/A
R2	$\overline{\text{NMI}}$	I	QVDD
R3	$\overline{\text{NMI\_OUT}}$[6]	O	QVDD
R4	$\overline{\text{HRESET}}$[6,7]	I/O	QVDD
R5	$\overline{\text{INT\_OUT}}$[6]	O	QVDD
R6	EE1	O	QVDD
R7	VSS	Ground	N/A
R8	PLL1_AVDD[9]	Power	VDD
R9	VSS	Ground	N/A
R10	VDD	Power	N/A
R11	VSS	Non-user	N/A
R12	VDD	Power	N/A

Table 1. Signal List by Ball Number (continued)

Ball Number	Signal Name[1,2]	Pin Type[10]	Power Rail Name
U3	GVDD1	Power	N/A
U4	M1DQ15	I/O	GVDD1
U5	M1DQ1	I/O	GVDD1
U6	VSS	Ground	N/A
U7	GVDD1	Power	N/A
U8	M1DQ7	I/O	GVDD1
U9	M1DQ6	I/O	GVDD1
U10	VDD	Power	N/A
U11	VSS	Ground	N/A
U12	M3VDD	Power	N/A
U13	VSS	Ground	N/A
U14	VDD	Power	N/A
U15	VSS	Ground	N/A
U16	VDD	Power	N/A
U17	VSS	Ground	N/A
U18	VDD	Power	N/A
U19	VSS	Ground	N/A
U20	VSS	Ground	N/A
U21	VSS	Ground	N/A
U22	VSS	Non-user	N/A
U23	SR2_TXD0/PE_TXD0[4]	O	SXPVDD2
U24	$\overline{\text{SR2\_TXD0}}/\overline{\text{PE\_TXD0}}$[4]	O	SXPVDD2
U25	SXCVDD2	Power	N/A
U26	SXCVSS2	Ground	N/A
U27	$\overline{\text{SR2\_RXD0}}/\overline{\text{PE\_RXD0}}$[4]	I	SXCVDD2
U28	SR2_RXD0/PE_RXD0[4]	I	SXCVDD2
V1	M1DQ9	I/O	GVDD1
V2	M1DQ12	I/O	GVDD1
V3	M1DQ13	I/O	GVDD1
V4	$\overline{\text{M1DQS0}}$	I/O	GVDD1
V5	M1DQS0	I/O	GVDD1
V6	M1DM0	O	GVDD1
V7	M1DQ3	I/O	GVDD1
V8	M1DQ2	I/O	GVDD1
V9	M1DQ4	I/O	GVDD1
V10	VSS	Ground	N/A
V11	VDD	Power	N/A
V12	VSS	Ground	N/A
V13	VDD	Power	N/A
V14	VSS	Ground	N/A
V15	VDD	Power	N/A
V16	VSS	Ground	N/A
V17	VDD	Power	N/A
V18	VSS	Ground	N/A
V19	VDD	Power	N/A
V20	NVDD	Power	N/A

Table 1. Signal List by Ball Number (continued)

Ball Number	Signal Name[1,2]	Pin Type[10]	Power Rail Name
AB1	$\overline{\text{M1DQS2}}$	I/O	GVDD1
AB2	M1DQS2	I/O	GVDD1
AB3	M1DQ19	I/O	GVDD1
AB4	M1DM2	O	GVDD1
AB5	M1DQ21	I/O	GVDD1
AB6	M1DQ22	I/O	GVDD1
AB7	M1CKE0	O	GVDD1
AB8	M1A11	O	GVDD1
AB9	M1A7	O	GVDD1
AB10	$\overline{\text{M1CK2}}$	O	GVDD1
AB11	M1APAR_OUT	O	GVDD1
AB12	M1ODT1	O	GVDD1
AB13	$\overline{\text{M1APAR\_IN}}$	I	GVDD1
AB14	M1DQ43	I/O	GVDD1
AB15	M1DM5	O	GVDD1
AB16	M1DQ44	I/O	GVDD1
AB17	M1DQ40	I/O	GVDD1
AB18	M1DQ59	I/O	GVDD1
AB19	M1DM7	O	GVDD1
AB20	M1DQ60	I/O	GVDD1
AB21	VSS	Ground	N/A
AB22	GPIO31/I2C_SDA[5,8]	I/O	NVDD
AB23	GPIO27/TMR4/RCW_SRC0[5,8]	I/O	NVDD
AB24	GPIO25/TMR2/RCW_SRC1[5,8]	I/O	NVDD
AB25	GPIO24/TMR1/RCW_SRC2[5,8]	I/O	NVDD
AB26	GPIO10/$\overline{\text{IRQ10}}$/RC10[5,8]	I/O	NVDD
AB27	GPIO5/$\overline{\text{IRQ5}}$/RC5[5,8]	I/O	NVDD
AB28	GPIO0/$\overline{\text{IRQ0}}$/RC0[5,8]	I/O	NVDD
AC1	VSS	Ground	N/A
AC2	GVDD1	Power	N/A
AC3	M1DQ16	I/O	GVDD1
AC4	VSS	Ground	N/A
AC5	GVDD1	Power	N/A
AC6	M1DQ17	I/O	GVDD1
AC7	VSS	Ground	N/A
AC8	GVDD1	Power	N/A
AC9	M1BA2	O	GVDD1
AC10	VSS	Ground	N/A
AC11	GVDD1	Power	N/A
AC12	M1A4	O	GVDD1
AC13	VSS	Ground	N/A
AC14	GVDD1	Power	N/A
AC15	M1DQ42	I/O	GVDD1
AC16	VSS	Ground	N/A
AC17	GVDD1	Power	N/A
AC18	M1DQ58	I/O	GVDD1

2 Electrical Characteristics

This document contains detailed information on power considerations, DC/AC electrical characteristics, and AC timing specifications. For additional information, see the *MSC8151 Reference Manual*.

2.1 Maximum Ratings

In calculating timing requirements, adding a maximum value of one specification to a minimum value of another specification does not yield a reasonable sum. A maximum specification is calculated using a worst case variation of process parameter values in one direction. The minimum specification is calculated using the worst case for the same parameters in the opposite direction. Therefore, a “maximum” value for a specification never occurs in the same device with a “minimum” value for another specification; adding a maximum to a minimum represents a condition that can never exist.

Table 2 describes the maximum electrical ratings for the MSC8151.

Table 2. Absolute Maximum Ratings

Rating	Power Rail Name	Symbol	Value	Unit
Core supply voltage	VDD	V_{DD}	–0.3 to 1.1	V
PLL supply voltage[3]		V_{DDPLL0} V_{DDPLL1} V_{DDPLL2}	–0.3 to 1.1 –0.3 to 1.1 –0.3 to 1.1	V V V
M3 memory supply voltage	M3VDD	V_{DDM3}	–0.3 to 1.1	V
MAPLE-B supply voltage	MVDD	V_{DDM}	–0.3 to 1.1	V
DDR memory supply voltage • DDR2 mode • DDR3 mode	GVDD1, GVDD2	V_{DDDDR}	 –0.3 to 1.98 –0.3 to 1.65	 V V
DDR reference voltage	MVREF	MV_{REF}	–0.3 to 0.51 × V_{DDDDR}	V
Input DDR voltage		V_{INDDR}	–0.3 to V_{DDDDR} + 0.3	V
I/O voltage excluding DDR and RapidIO lines	NVDD, QVDD	V_{DDIO}	–0.3 to 2.625	V
Input I/O voltage		V_{INIO}	–0.3 to V_{DDIO} + 0.3	V
RapidIO pad voltage	SXPVDD1, SXPVDD2	V_{DDSXP}	–0.3 to 1.26	V
Rapid I/O core voltage	SXCVDD1, SXCVDD2	V_{DDSXC}	–0.3 to 1.21	V
Rapid I/O PLL voltage[3]		$V_{DDRIOPLL}$	–0.3 to 1.21	V
Input RapidIO I/O voltage		V_{INRIO}	–0.3 to V_{DDSXC} + 0.3	V
Operating temperature		T_J	–40 to 105	°C
Storage temperature range		T_{STG}	–55 to +150	°C

Notes:
1. Functional operating conditions are given in Table 3**.**
2. Absolute maximum ratings are stress ratings only, and functional operation at the maximum is not guaranteed. Stress beyond the listed limits may affect device reliability or cause permanent damage.
3. PLL supply voltage is specified at input of the filter and not at pin of the MSC8151 (see Figure 37 and Figure 38)

2.3 Thermal Characteristics

Table 4 describes thermal characteristics of the MSC8151 for the FC-PBGA packages.

Table 4. Thermal Characteristics for the MSC8151

Characteristic	Symbol	FC-PBGA 29 × 29 mm^2		Unit
		Natural Convection	200 ft/min (1 m/s) airflow	
Junction-to-ambient[1, 2]	$R_{\theta JA}$	18	12	°C/W
Junction-to-ambient, four-layer board[1, 2]	$R_{\theta JA}$	13	9	°C/W
Junction-to-board (bottom)[3]	$R_{\theta JB}$	5		°C/W
Junction-to-case[4]	$R_{\theta JC}$	0.6		°C/W

Notes:
1. Junction temperature is a function of die size, on-chip power dissipation, package thermal resistance, mounting site (board) temperature, ambient temperature, air flow, power dissipation of other components on the board, and board thermal resistance.
2. Junction-to-ambient thermal resistance determined per JEDEC JESD51-3 and JESDC51-6. Thermal test board meets JEDEC specification for the specified package.
3. Junction-to-board thermal resistance determined per JEDEC JESD 51-8. Thermal test board meets JEDEC specification for the specified package.
4. Junction-to-case at the top of the package determined using MIL- STD-883 Method 1012.1. The cold plate temperature is used for the case temperature. Reported value includes the thermal resistance of the interface layer

2.4 CLKIN Requirements

Table 5 summarizes the required characteristics for the CLKIN signal.

Table 5. CLKIN Requirements

Parameter/Condition[1]	Symbol	Min	Typ	Max	Unit	Notes
CLKIN duty cycle	—	40	—	60	%	2
CLKIN slew rate	—	1	—	4	V/ns	3
CLKIN peak period jitter	—	—	—	±150	ps	—
CLKIN jitter phase noise at –56 dBc	—	—	—	500	KHz	4
AC input swing limits	ΔV_{AC}	1.5	—	—	V	—
Input capacitance	C_{IN}	—	—	15	pf	—

Notes:
1. For clock frequencies, see the *Clock* chapter in the *MSC8151 Reference Manual.*
2. Measured at the rising edge and/or the falling edge at $V_{DDIO}/2$.
3. Slew rate as measured from ±20% to 80% of voltage swing at clock input.
4. Phase noise is calculated as FFT of TIE jitter.

2.5.1.4 DDR Reference Current Draw

Table 9 lists the current draw characteristics for MV_{REF}.

Note: Values when used at recommended operating conditions (see Table 3).

Table 9. Current Draw Characteristics for MV_{REF}

Parameter / Condition	Symbol	Min	Max	Unit
Current draw for MV_{REFn} • DDR2 SDRAM • DDR3 SDRAM	I_{MVREFn}	—	 300 250	 µA µA

2.5.2 High-Speed Serial Interface (HSSI) DC Electrical Characteristics

The MSC8151 features an HSSI that includes two 4-channel SerDes ports used for high-speed serial interface applications (PCI Express, Serial RapidIO interfaces, and SGMII). This section and its subsections describe the common portion of the SerDes DC, including the DC requirements for the SerDes reference clocks and the SerDes data lane transmitter (Tx) and receiver (Rx) reference circuits. The data lane circuit specifications are specific for each supported interface, and they have individual subsections by protocol. The selection of individual data channel functionality is done via the Reset Configuration Word High Register (RCWHR) SerDes Protocol selection fields (S1P and S2P). Specific AC electrical characteristics are defined in Section 2.6.2, “HSSI AC Timing Specifications.”

2.5.2.1 Signal Term Definitions

The SerDes interface uses differential signalling to transfer data across the serial link. This section defines terms used in the description and specification of differential signals. Figure 4 shows how the signals are defined. For illustration purposes only, one SerDes lane is used in the description. Figure 4 shows the waveform for either a transmitter output (SR[1–2]_TX and $\overline{\text{SR[1–2]\_TX}}$) or a receiver input (SR[1–2]_RX and $\overline{\text{SR[1–2]\_RX}}$). Each signal swings between A volts and B volts where A > B.

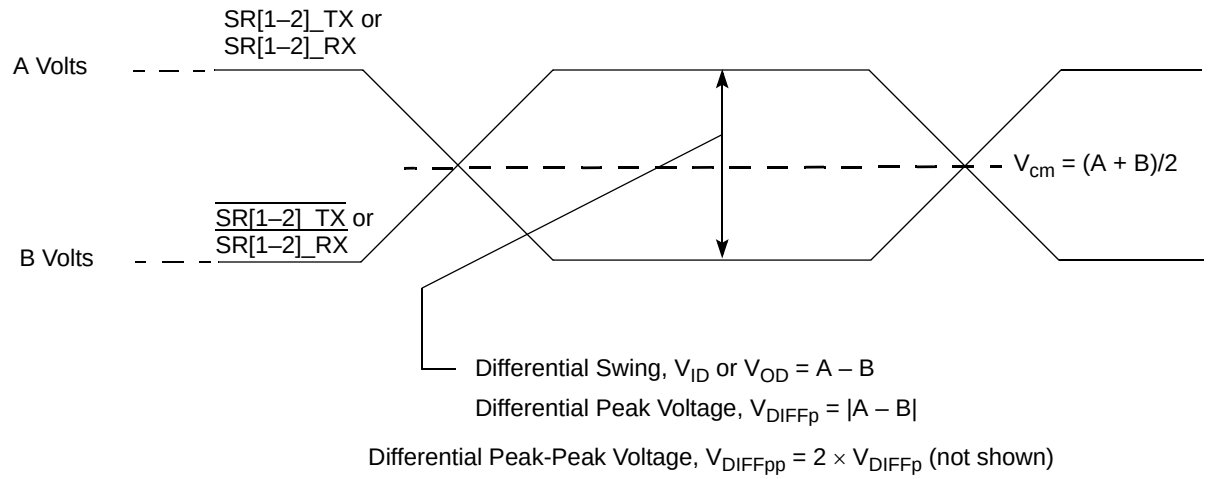


Figure 4. Differential Voltage Definitions for Transmitter or Receiver

Using this waveform, the definitions are listed in Table 10. To simplify the illustration, the definitions assume that the SerDes transmitter and receiver operate in a fully symmetrical differential signalling environment.

Table 10. Differential Signal Definitions

Term	Definition
Single-Ended Swing	The transmitter output signals and the receiver input signals SR[1–2]_TX, $\overline{\text{SR[1–2]\_TX}}$, SR[1–2]_RX and $\overline{\text{SR[1–2]\_RX}}$ each have a peak-to-peak swing of A – B volts. This is also referred to as each signal wire's single-ended swing.
Differential Output Voltage, V_{OD} (or **Differential Output Swing**):	The differential output voltage (or swing) of the transmitter, V_{OD}, is defined as the difference of the two complimentary output voltages: $V_{SR[1–2]\_TX} - V_{\overline{SR[1–2]\_TX}}$. The V_{OD} value can be either positive or negative.
Differential Input Voltage, V_{ID} (or **Differential Input Swing**)	The differential input voltage (or swing) of the receiver, V_{ID}, is defined as the difference of the two complimentary input voltages: $V_{SR[1–2]\_RX} - V_{\overline{SR[1–2]\_RX}}$. The V_{ID} value can be either positive or negative.
Differential Peak Voltage, V_{DIFFp}	The peak value of the differential transmitter output signal or the differential receiver input signal is defined as the differential peak voltage, $V_{DIFFp} = \|A – B\|$ volts.
Differential Peak-to-Peak, $V_{DIFFp-p}$	Since the differential output signal of the transmitter and the differential input signal of the receiver each range from A – B to –(A – B) volts, the peak-to-peak value of the differential transmitter output signal or the differential receiver input signal is defined as differential peak-to-peak voltage, $V_{DIFFp-p} = 2 \times V_{DIFFp} = 2 \times \|(A – B)\|$ volts, which is twice the differential swing in amplitude, or twice of the differential peak. For example, the output differential peak-peak voltage can also be calculated as $V_{TX-DIFFp-p} = 2 \times \|V_{OD}\|$.
Differential Waveform	The differential waveform is constructed by subtracting the inverting signal ($\overline{\text{SR[1–2]\_TX}}$, for example) from the non-inverting signal (SR[1–2]_TX, for example) within a differential pair. There is only one signal trace curve in a differential waveform. The voltage represented in the differential waveform is not referenced to ground. Refer to Figure 16 as an example for differential waveform.
Common Mode Voltage, V_{cm}	The common mode voltage is equal to half of the sum of the voltages between each conductor of a balanced interchange circuit and ground. In this example, for SerDes output, $V_{cm\_out} = (V_{SR[1–2]\_TX} + V_{\overline{SR[1–2]\_TX}}) \div 2 = (A + B) \div 2$, which is the arithmetic mean of the two complimentary output voltages within a differential pair. In a system, the common mode voltage may often differ from one component's output to the other's input. It may be different between the receiver input and driver output circuits within the same component. It is also referred to as the DC offset on some occasions.

To illustrate these definitions using real values, consider the example of a current mode logic (CML) transmitter that has a common mode voltage of 2.25 V and outputs, TD and $\overline{\text{TD}}$. If these outputs have a swing from 2.0 V to 2.5 V, the peak-to-peak voltage swing of each signal (TD or $\overline{\text{TD}}$) is 500 mV p-p, which is referred to as the single-ended swing for each signal. Because the differential signaling environment is fully symmetrical in this example, the transmitter output differential swing (V_{OD}) has the same amplitude as each signal single-ended swing. The differential output signal ranges between 500 mV and –500 mV. In other words, V_{OD} is 500 mV in one phase and –500 mV in the other phase. The peak differential voltage (V_{DIFFp}) is 500 mV. The peak-to-peak differential voltage ($V_{DIFFp-p}$) is 1000 mV p-p.

2.5.2.2 SerDes Reference Clock Receiver Characteristics

The SerDes reference clock inputs are applied to an internal PLL whose output creates the clock used by the corresponding SerDes lanes. The SerDes reference clock inputs are SR1_REF_CLK/$\overline{\text{SR1\_REF\_CLK}}$ or SR2_REF_CLK/$\overline{\text{SR2\_REF\_CLK}}$. Figure 5 shows a receiver reference diagram of the SerDes reference clocks.

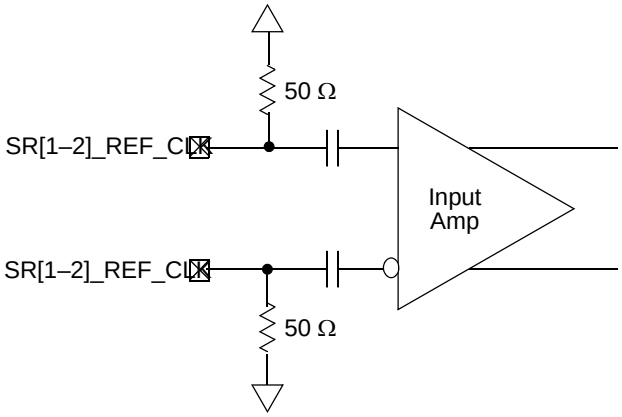


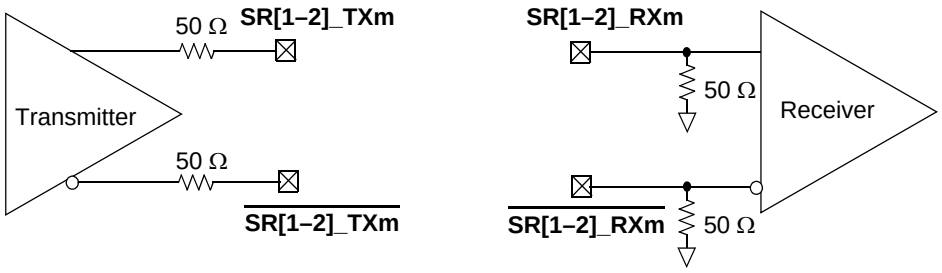
Figure 5. Receiver of SerDes Reference Clocks

The characteristics of the clock signals are as follows:

- The supply voltage requirements for V_{DDSXC} are as specified in **Table 3.**
- The SerDes reference clock receiver reference circuit structure is as follows:
  - The SR[1–2]_REF_CLK and $\overline{\text{SR[1–2]\_REF\_CLK}}$ are internally AC-coupled differential inputs as shown in Figure 5. Each differential clock input (SR[1–2]_REF_CLK or $\overline{\text{SR[1–2]\_REF\_CLK}}$) has on-chip 50-Ω termination to GND_{SXC} followed by on-chip AC-coupling.
  - The external reference clock driver must be able to drive this termination.
  - The SerDes reference clock input can be either differential or single-ended. Refer to the differential mode and single-ended mode descriptions below for detailed requirements.
- The maximum average current requirement also determines the common mode voltage range.
  - When the SerDes reference clock differential inputs are DC coupled externally with the clock driver chip, the maximum average current allowed for each input pin is 8 mA. In this case, the exact common mode input voltage is not critical as long as it is within the range allowed by the maximum average current of 8 mA because the input is AC-coupled on-chip.
  - This current limitation sets the maximum common mode input voltage to be less than 0.4 V (0.4 V / 50 = 8 mA) while the minimum common mode input level is 0.1 V above GND_{SXC}. For example, a clock with a 50/50 duty cycle can be produced by a clock driver with output driven by its current source from 0 mA to 16 mA (0–0.8 V), such that each phase of the differential input has a single-ended swing from 0 V to 800 mV with the common mode voltage at 400 mV.
  - If the device driving the SR[1–2]_REF_CLK and $\overline{\text{SR[1–2]\_REF\_CLK}}$ inputs cannot drive 50 Ω to GND_{SXC} DC or the drive strength of the clock driver chip exceeds the maximum input current limitations, it must be AC-coupled externally.
- The input amplitude requirement is described in detail in the following sections.

2.5.2.3 SerDes Transmitter and Receiver Reference Circuits

Figure 6 shows the reference circuits for SerDes data lane transmitter and receiver.



Note: The [1–2] indicates the specific SerDes Interface (1 or 2) and the m indicates the specific channel within that interface (0,1,2,3). Actual signals are assigned by the HRCW assignments at reset (see **Chapter 5**, *Reset* in the reference manual for details)

Figure 6. SerDes Transmitter and Receiver Reference Circuits

2.5.3 DC-Level Requirements for SerDes Interfaces

The following subsections define the DC-level requirements for the SerDes reference clocks, the PCI Express data lines, the Serial RapidIO data lines, and the SGMII data lines.

2.5.3.1 DC-Level Requirements for SerDes Reference Clocks

The DC-level requirement for the SerDes reference clock inputs is different depending on the signaling mode used to connect the clock driver chip and SerDes reference clock inputs, as described below:

- Differential Mode
  - The input amplitude of the differential clock must be between 400 mV and 1600 mV differential peak-peak (or between 200 mV and 800 mV differential peak). In other words, each signal wire of the differential pair must have a single-ended swing of less than 800 mV and greater than 200 mV. This requirement is the same for both external DC-coupled or AC-coupled connection.
  - For an external DC-coupled connection, the maximum average current requirements sets the requirement for average voltage (common mode voltage) as between 100 mV and 400 mV. Figure 7 shows the SerDes reference clock input requirement for DC-coupled connection scheme.

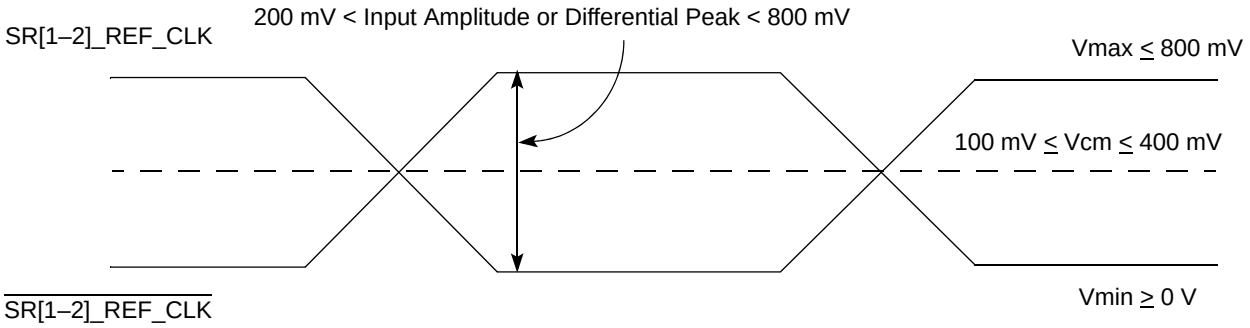


Figure 7. Differential Reference Clock Input DC Requirements (External DC-Coupled)

Figure 11 shows the DDR2 and DDR3 SDRAM interface input timing diagram.

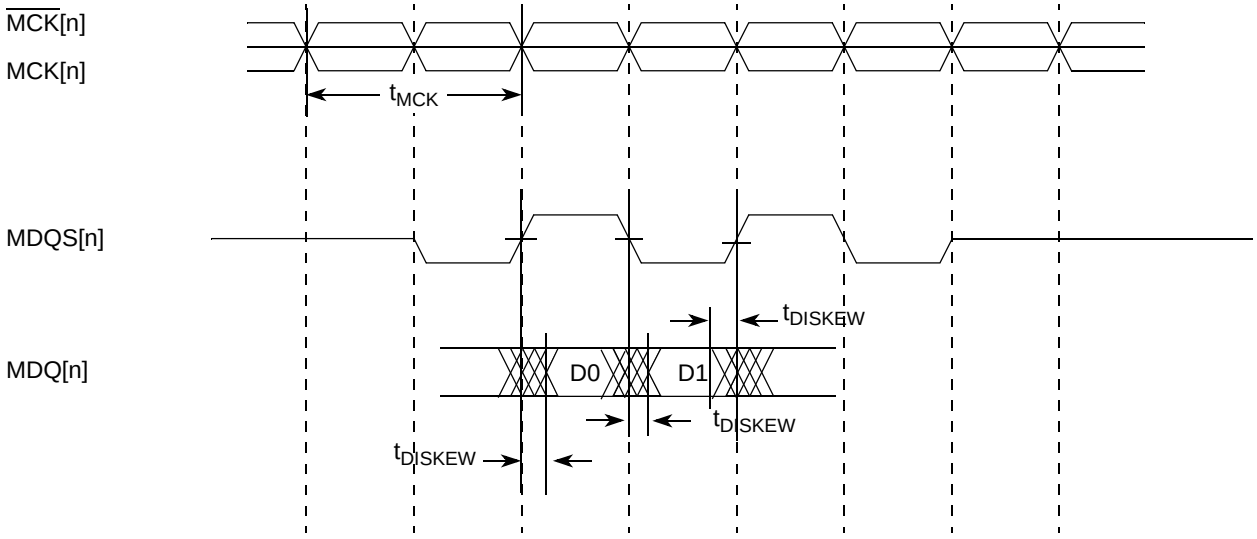


Figure 11. DDR2 and DDR3 SDRAM Interface Input Timing Diagram

2.6.1.2 DDR SDRAM Output AC Timing Specifications

Table 21 provides the output AC timing specifications for the DDR SDRAM interface.

Table 21. DDR SDRAM Output AC Timing Specifications

Parameter	Symbol [1]	Min	Max	Unit	Notes
MCK[n] cycle time	t_{MCK}	2.5	5	ns	2
ADDR/CMD output setup with respect to MCK • 800 MHz data rate • 667 MHz data rate	t_{DDKHAS}	 0.917 1.10	 — —	 ns ns	3
ADDR/CMD output hold with respect to MCK • 800 MHz data rate • 667 MHz data rate	t_{DDKHAX}	 0.767 1.02	 — —	 ns ns	3
$\overline{\text{MCS}}$n output setup with respect to MCK • 800 MHz data rate • 667 MHz data rate	t_{DDKHCS}	 0.917 1.10	 — —	 ns ns	3
$\overline{\text{MCS}}$n output hold with respect to MCK • 800 MHz data rate • 667 MHz data rate	t_{DDKHCX}	 0.767 1.02	 — —	 ns ns	3
MCK to MDQS Skew • 800 MHz data rate • 667 MHz data rate	t_{DDKHMH}	 –0.4 –0.6	 0.375 0.6	ns	4
MDQ/MECC/MDM output setup with respect to MDQS • 800 MHz • 667 MHz	t_{DDKHDS}, t_{DDKLDS}	 300 375	 — —	 ps ps	5
MDQ/MECC/MDM output hold with respect to MDQS • 800 MHz • 667 MHz	t_{DDKHDX}, t_{DDKLDX}	 300 375	 — —	 ps ps	5
MDQS preamble	t_{DDKHMP}	$-0.9 \times t_{MCK}$	—	ns	—
MDQS postamble	t_{DDKHME}	$-0.4 \times t_{MCK}$	$-0.6 \times t_{MCK}$	ns	—

Table 21. DDR SDRAM Output AC Timing Specifications (continued)

Parameter	Symbol [1]	Min	Max	Unit	Notes

Notes:

1. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state) (reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. Output hold time can be read as DDR timing (DD) from the rising or falling edge of the reference clock (KH or KL) until the output went invalid (AX or DX). For example, t_{DDKHAS} symbolizes DDR timing (DD) for the time t_{MCK} memory clock reference (K) goes from the high (H) state until outputs (A) are setup (S) or output valid time. Also, t_{DDKLDX} symbolizes DDR timing (DD) for the time t_{MCK} memory clock reference (K) goes low (L) until data outputs (D) are invalid (X) or data output hold time.
2. All MCK/$\overline{\text{MCK}}$ referenced measurements are made from the crossing of the two signals.
3. ADDR/CMD includes all DDR SDRAM output signals except MCK/$\overline{\text{MCK}}$, $\overline{\text{MCS}}$, and MDQ/MECC/MDM/MDQS.
4. Note that t_{DDKHMH} follows the symbol conventions described in note 1. For example, t_{DDKHMH} describes the DDR timing (DD) from the rising edge of the MCK(n) clock (KH) until the MDQS signal is valid (MH). t_{DDKHMH} can be modified through control of the DQSS override bits in the TIMING_CFG_2 register. This will typically be set to the same delay as the clock adjust in the CLK_CNTL register. The timing parameters listed in the table assume that these two parameters have been set to the same adjustment value. See the *MSC8151 Reference Manual* for a description and understanding of the timing modifications enabled by use of these bits.
5. Determined by maximum possible skew between a data strobe (MDQS) and any corresponding bit of data (MDQ), ECC (MECC), or data mask (MDM). The data strobe should be centered inside of the data eye at the pins of the MSC8151.
6. At recommended operating conditions with V_{DDDDR} (1.5 V or 1,8 V) ± 5%.

Note: For the ADDR/CMD setup and hold specifications in Table 21, it is assumed that the clock control register is set to adjust the memory clocks by ½ applied cycle.

Figure 12 shows the DDR SDRAM output timing for the MCK to MDQS skew measurement (t_{DDKHMH}).

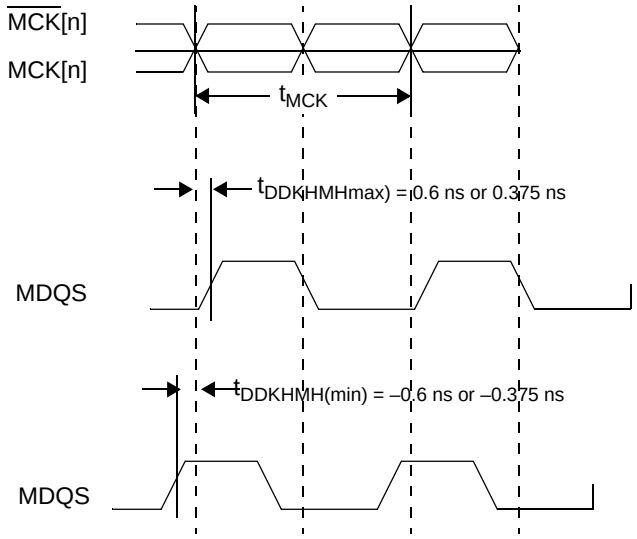


Figure 12. MCK to MDQS Timing

Figure 13 shows the DDR SDRAM output timing diagram.

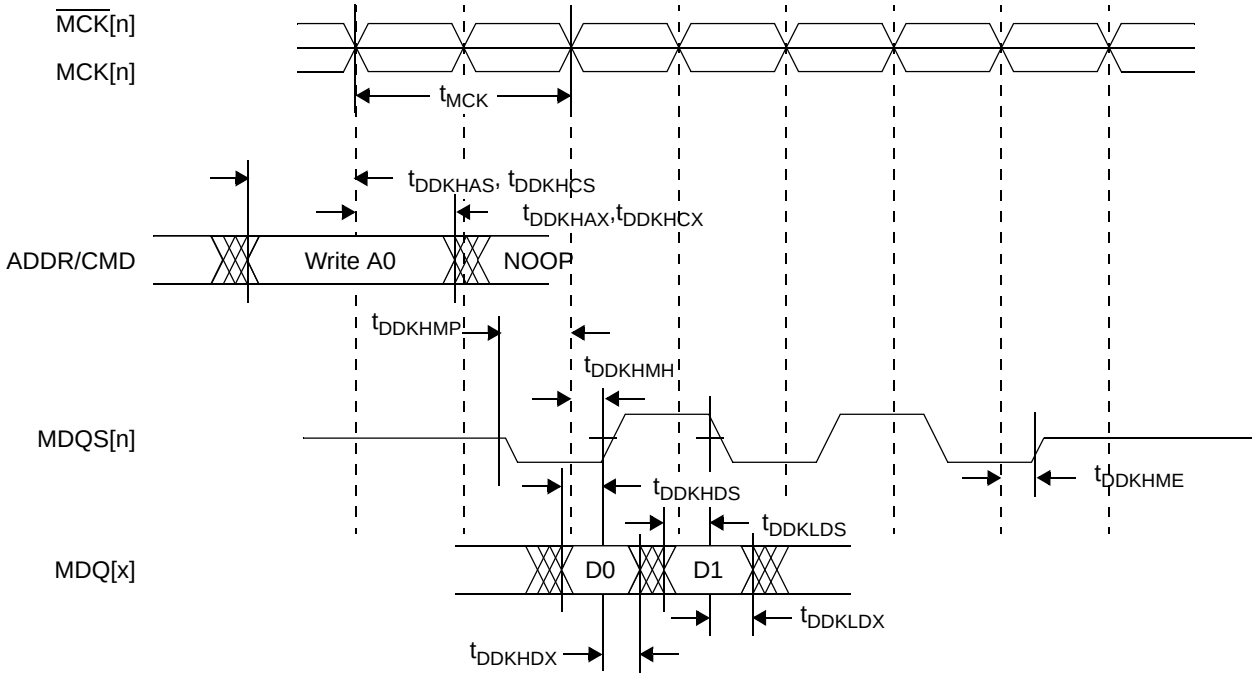


Figure 13. DDR SDRAM Output Timing

Figure 14 provides the AC test load for the DDR2 and DDR3 controller bus.

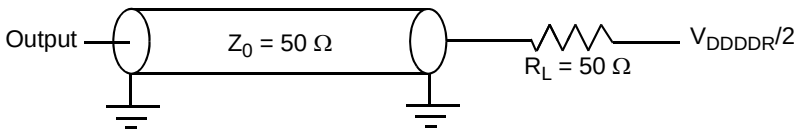


Figure 14. DDR2 and DDR3 Controller Bus AC Test Load

2.6.1.3 DDR2 and DDR3 SDRAM Differential Timing Specifications

This section describes the DC and AC differential timing specifications for the DDR2 and DDR3 SDRAM controller interface. Figure 15 shows the differential timing specification.

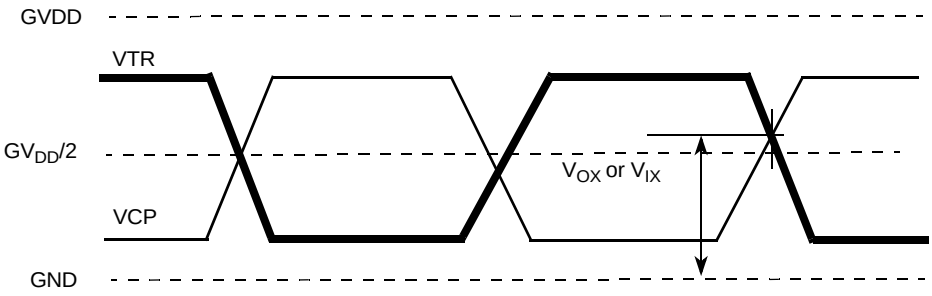


Figure 15. DDR2 and DDR3 SDRAM Differential Timing Specifications

Note: VTR specifies the true input signal (such as MCK or MDQS) and VCP is the complementary input signal (such as $\overline{\text{MCK}}$ or $\overline{\text{MDQS}}$).

Figure 21 shows the TDM transmit signal timing.

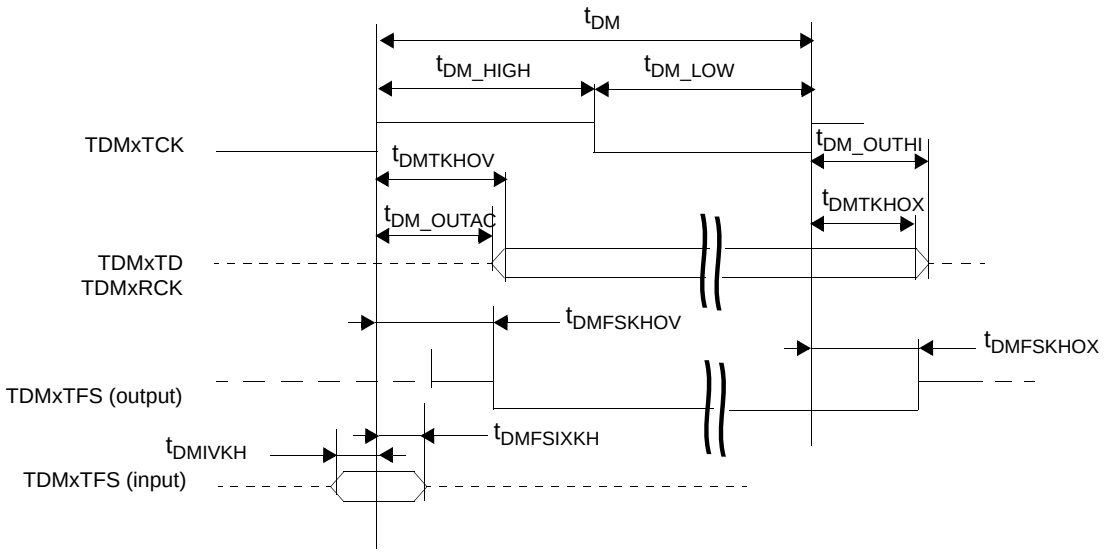


Figure 21. TDM Transmit Signals

Figure 22 provides the AC test load for the TDM/SI.

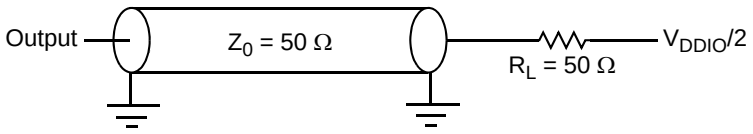


Figure 22. TDM AC Test Load

2.6.4 Timers AC Timing Specifications

Table 32 lists the timer input AC timing specifications.

Table 32. Timers Input AC Timing Specifications

Characteristics	Symbol	Minimum	Unit	Notes
Timers inputs—minimum pulse width	T_{TIWID}	8	ns	1, 2

Notes:
1. The maximum allowed frequency of timer outputs is 125 MHz. Configure the timer modules appropriately.
2. Timer inputs and outputs are asynchronous to any visible clock. Timer outputs should be synchronized before use by any external synchronous logic. Timer inputs are required to be valid for at least t_{TIWID} ns to ensure proper operation.

Note: For recommended operating conditions, see Table 3.

Figure 23 shows the AC test load for the timers.

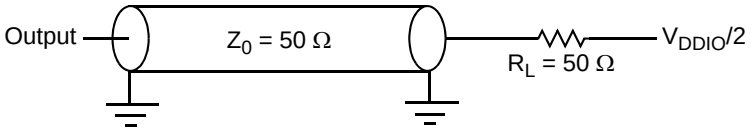


Figure 23. Timer AC Test Load

2.6.6 SPI Timing

Table 36 lists the SPI input and output AC timing specifications.

Table 36. SPI AC Timing Specifications

Parameter	Symbol [1]	Min	Max	Unit	Note
SPI outputs valid—Master mode (internal clock) delay	t_{NIKHOV}	—	6	ns	2
SPI outputs hold—Master mode (internal clock) delay	t_{NIKHOX}	0.5	—	ns	2
SPI outputs valid—Slave mode (external clock) delay	t_{NEKHOV}	—	12	ns	2
SPI outputs hold—Slave mode (external clock) delay	t_{NEKHOX}	2	—	ns	2
SPI inputs—Master mode (internal clock) input setup time	t_{NIIVKH}	12	—	ns	—
SPI inputs—Master mode (internal clock) input hold time	t_{NIIXKH}	0	—	ns	—
SPI inputs—Slave mode (external clock) input setup time	t_{NEIVKH}	4	—	ns	—
SPI inputs—Slave mode (external clock) input hold time	t_{NEIXKH}	2	—	ns	—

Notes:
1. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{NIKHOX} symbolizes the internal timing (NI) for the time SPICLK clock reference (K) goes to the high state (H) until outputs (O) are invalid (X).
2. Output specifications are measured from the 50% level of the rising edge of SPICLK to the 50% level of the signal. Timings are measured at the pin.

Figure 26 provides the AC test load for the SPI.

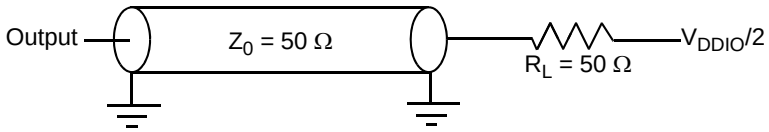
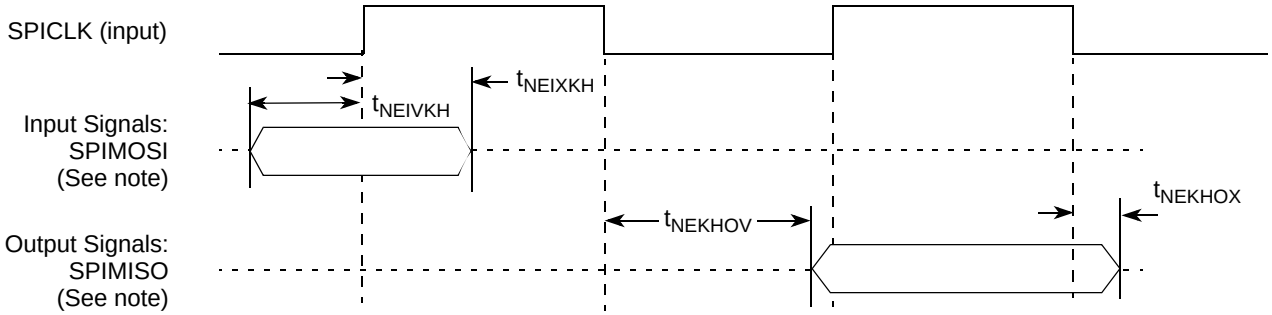


Figure 26. SPI AC Test Load

Figure 27 and Figure 28 represent the AC timings from Table 36. Note that although the specifications generally reference the rising edge of the clock, these AC timing diagrams also apply when the falling edge is the active edge.

Figure 27 shows the SPI timings in slave mode (external clock).



Note: measured with SPMODE[CI] = 0, SPMODE[CP] = 0

Figure 27. SPI AC Timing in Slave Mode (External Clock)

Figure 28 shows the SPI timings in master mode (internal clock).

3 Hardware Design Considerations

The following sections discuss areas to consider when the MSC8151 device is designed into a system.

3.1 Power Supply Ramp-Up Sequence

The following subsections describe the required device initialization sequence.

3.1.1 Clock, Reset, and Supply Coordination

Starting the device requires coordination between several inputs including: clock, reset, and power supplies. Follow this guidelines when starting up an MSC8151 device:

- $\overline{\text{PORESET}}$ and $\overline{\text{TRST}}$ must be asserted externally for the duration of the supply ramp-up, using the V_{DDIO} supply. $\overline{\text{TRST}}$ deassertion does not have to be synchronized with $\overline{\text{PORESET}}$ deassertion. However, $\overline{\text{TRST}}$ must be deasserted before normal operation begins to ensure correct functionality of the device.
- CLKIN should toggle at least 32 cycles before $\overline{\text{PORESET}}$ deassertion to guarantee correct device operation. The 32 cycles should only be counted from the time after V_{DDIO} reaches its nominal value (see timing 1 in Figure 33).
- CLKIN should either be stable low during ramp-up of V_{DDIO} supply (and start its swings after ramp-up) or should swing within V_{DDIO} range during V_{DDIO} ramp-up, so its amplitude grows as V_{DDIO} grows during ramp-up.

Figure 33 shows a sequence in which V_{DDIO} ramps-up after V_{DD} and CLKIN begins to toggle with the raise of V_{DDIO} supply.

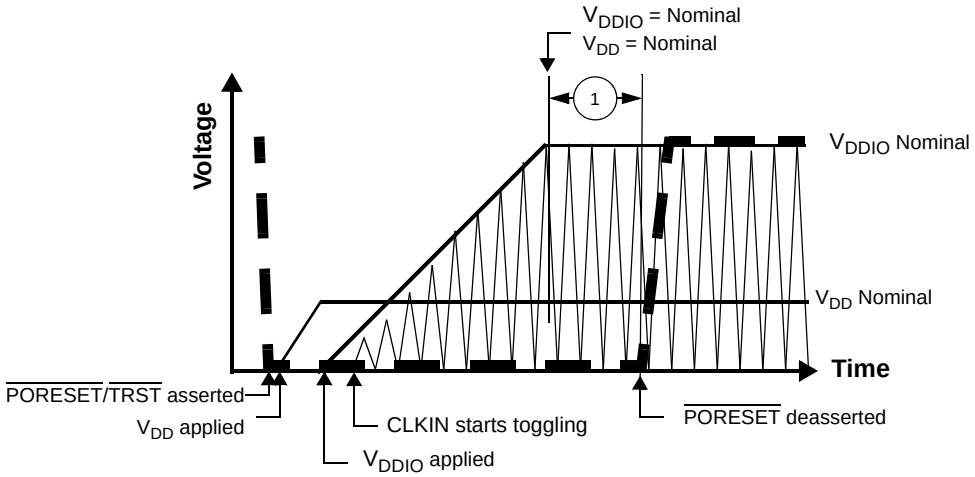


Figure 33. Supply Ramp-Up Sequence with V_{DD} Ramping Before V_{DDIO} and CLKIN Starting With V_{DDIO}

Note: For details on power-on reset flow and duration, see the *Reset* chapter in the *MSC8151 Reference Manual*.

2. After the above rails rise to 90% of their nominal voltage, the following I/O power rails may rise in any sequence (see Figure 34): QVDD, NVDD, GVDD1, and GVDD2.

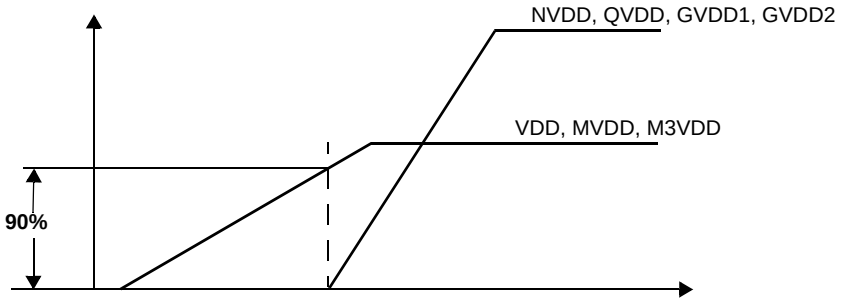


Figure 34. Supply Ramp-Up Sequence

Notes:
1. If the M3 memory is not used, M3VDD can be tied to GND.
2. If the MAPLE-B is not used, MVDD can be tied to GND.
3. If the HSSI port1 is not used, SXCVDD1and SXPVDD1 must be connected to the designated power supplies.
4. If the HSSI port2 is not used, SXCVDD2 and SXPVDD2 must be connected to the designated power supplies.
5. If the DDR port 1 interface is not used, it is recommended that GVDD1 be left unconnected.
6. If the DDR port 2 interface is not used, it is recommended that GVDD2 be left unconnected.

3.1.4 Reset Guidelines

When a debugger is not used, implement the connection scheme shown in Figure 35.

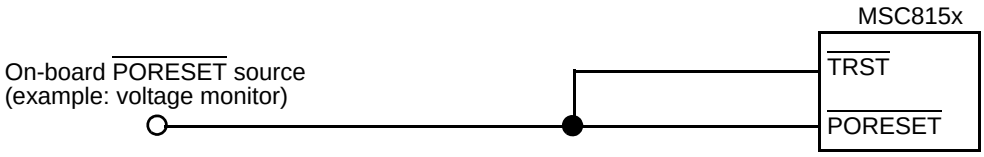


Figure 35. Reset Connection in Functional Application

When a debugger is used, implement the connection scheme shown in Figure 36.

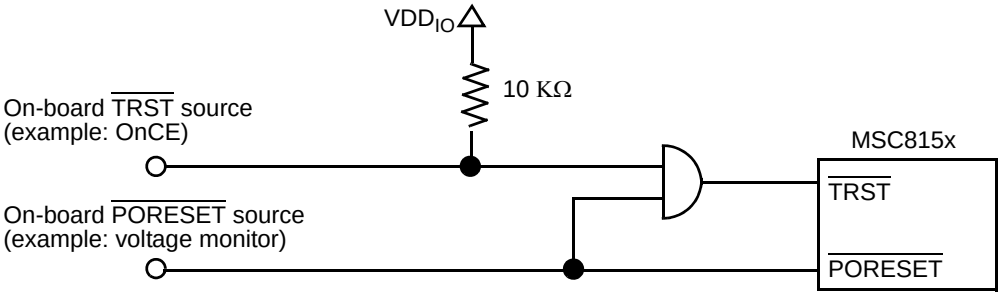


Figure 36. Reset Connection in Debugger Application

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