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### Applications of "[Embedded - Microcontrollers](#)"

#### Details

Product Status	Active
Core Processor	S08
Core Size	8-Bit
Speed	20MHz
Connectivity	I <sup>2</sup> C, SCI, SPI
Peripherals	LVD, POR, PWM, WDT
Number of I/O	12
Program Memory Size	4KB (4K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	256 x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.6V
Data Converters	A/D 8x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	16-TSSOP (0.173", 4.40mm Width)
Supplier Device Package	16-TSSOP
Purchase URL	<a href="https://www.e-xfl.com/product-detail/nxp-semiconductors/mc9s08qg4mdte">https://www.e-xfl.com/product-detail/nxp-semiconductors/mc9s08qg4mdte</a>

# Chapter 1

## Device Overview

### 1.1 Introduction

The MC9S08QG8 is a member of the low-cost, high-performance HCS08 Family of 8-bit microcontroller units (MCUs). All MCUs in the family use the enhanced HCS08 core and are available with a variety of modules, memory sizes, memory types, and package types. Refer to [Table 1-1](#) for features associated with each device in this series.

#### 1.1.1 Devices in the MC9S08QG8/4 Series

[Table 1-1](#) summarizes the features available in the MC9S08QG8/4 series of MCUs.

**Table 1-1. Devices in the MC9S08QG8/4 Series**

Feature	Device					
	MC9S08QG8			MC9S08QG4		
Package	24-Pin	16-Pin	8-Pin	24-Pin	16-Pin	8-Pin
FLASH	8K			4K		
RAM	512			256		
XOSC	yes	yes	no	yes	yes	no
ICS	yes			yes		
ACMP	yes			yes		
ADC	8-ch	8-ch	4-ch	8-ch	8-ch	4-ch
DBG	yes			yes	yes	yes
IIC	yes			yes		
IRQ	yes			yes		
KBI	8-pin	8-pin	4-pin	8-pin	8-pin	4-pin
MTIM	yes			yes		
SCI	yes	yes	no	yes	yes	no
SPI	yes	yes	no	yes	yes	no
TPM	2-ch	2-ch	1-ch	2-ch	2-ch	1-ch
I/O pins	12 I/O 1 Output only 1 Input only	12 I/O 1 Output only 1 Input only	4 I/O 1 Output only 1 Input only	12 I/O 1 Output only 1 Input only	12 I/O 1 Output only 1 Input only	4 I/O 1 Output only 1 Input only
Package Types	24 QFN	16 PDIP 16 QFN 16 TSSOP	8 DFN 8 SOIC	24 QFN	16 QFN 16 TSSOP	8 DFN 8 PDIP 8 SOIC

**NOTE**

When an alternative function is first enabled, it is possible to get a spurious edge to the module. User software should clear out any associated flags before interrupts are enabled. [Table 2-1](#) shows the priority if multiple modules are enabled. The highest priority module will have control over the pin. Selecting a higher priority pin function with a lower priority function already enabled can cause spurious edges to the lower priority module. It is recommended that all modules that share a pin be disabled before enabling another module.



### 4.3 Register Addresses and Bit Assignments

The registers in the MC9S08QG8/4 are divided into these groups:

- Direct-page registers are located in the first 96 locations in the memory map; these are accessible with efficient direct addressing mode instructions.
- High-page registers are used much less often, so they are located from 0x1800 and above in the memory map. This leaves more room in the direct page for more frequently used registers and RAM.
- The nonvolatile register area consists of a block of 16 locations in FLASH memory at 0xFFB0–0xFFBF. Nonvolatile register locations include:
  - NVPROT and NVOPT are loaded into working registers at reset.
  - An 8-byte backdoor comparison key that optionally allows a user to gain controlled access to secure memory.

Because the nonvolatile register locations are FLASH memory, they must be erased and programmed like other FLASH memory locations.

Direct-page registers can be accessed with efficient direct addressing mode instructions. Bit manipulation instructions can be used to access any bit in any direct-page register. [Table 4-2](#) is a summary of all user-accessible direct-page registers and control bits.

The direct page registers in [Table 4-2](#) can use the more efficient direct addressing mode that requires only the lower byte of the address. Because of this, the lower byte of the address in column one is shown in bold text. In [Table 4-3](#) and [Table 4-4](#), the whole address in column one is shown in bold. In [Table 4-2](#), [Table 4-3](#), and [Table 4-4](#), the register names in column two are shown in bold to set them apart from the bit names to the right. Cells that are not associated with named bits are shaded. A shaded cell with a 0 indicates this unused bit always reads as a 0. Shaded cells with dashes indicate unused or reserved bit locations that could read as 1s or 0s.

**Table 4-2. Direct-Page Register Summary**

Address	Register Name	Bit 7	6	5	4	3	2	1	Bit 0
0x0000	<b>PTAD</b>	0	0	PTAD5	PTAD4	PTAD3	PTAD2	PTAD1	PTAD0
0x0001	<b>PTADD</b>	0	0	PTADD5	PTADD4	PTADD3	PTADD2	PTADD1	PTADD0
0x0002	<b>PTBD</b>	PTBD7	PTBD6	PTBD5	PTBD4	PTBD3	PTBD2	PTBD1	PTBD0
0x0003	<b>PTBDD</b>	PTBDD7	PTBDD6	PTBDD5	PTBDD4	PTBDD3	PTBDD2	PTBDD1	PTBDD0
0x0004– 0x000B	Reserved	—	—	—	—	—	—	—	—
0x000C	<b>KBISC</b>	0	0	0	0	KBF	KBACK	KBIE	KBMOD
0x000D	<b>KBIPE</b>	KBIPE7	KBIPE6	KBIPE5	KBIPE4	KBIPE3	KBIPE2	KBIPE1	KBIPE0
0x000E	<b>KBIES</b>	KBEDG7	KBEDG6	KBEDG5	KBEDG4	KBEDG3	KBEDG2	KBEDG1	KBEDG0
0x000F	<b>IRQSC</b>	0	IRQPDD	0	IRQPE	IRQF	IRQACK	IRQIE	IRQMOD
0x0010	<b>ADCSC1</b>	COCO	AIEN	ADCO	ADCH				
0x0011	<b>ADCSC2</b>	ADACT	ADTRG	ACFE	ACFGT	—	—	—	—
0x0012	<b>ADCRH</b>	0	0	0	0	0	0	ADR9	ADR8
0x0013	<b>ADCRL</b>	ADR7	ADR6	ADR5	ADR4	ADR3	ADR2	ADR1	ADR0

## 4.5 FLASH

The FLASH memory is intended primarily for program storage. In-circuit programming allows the operating program to be loaded into the FLASH memory after final assembly of the application product. It is possible to program the entire array through the single-wire background debug interface. Because no special voltages are needed for FLASH erase and programming operations, in-application programming is also possible through other software-controlled communication paths. For a more detailed discussion of in-circuit and in-application programming, refer to the *HCS08 Family Reference Manual, Volume I*, Freescale Semiconductor document order number HCS08RMv1/D.

The first byte of a series of sequential bytes being programmed in burst mode will take the same amount of time to program as a byte programmed in standard mode. Subsequent bytes will program in the burst program time provided that the conditions above are met. In the case the next sequential address is the beginning of a new row, the program time for that byte will be the standard time instead of the burst time. This is because the high voltage to the array must be disabled and then enabled again. If a new burst command has not been queued before the current command completes, then the charge pump will be disabled and high voltage removed from the array.

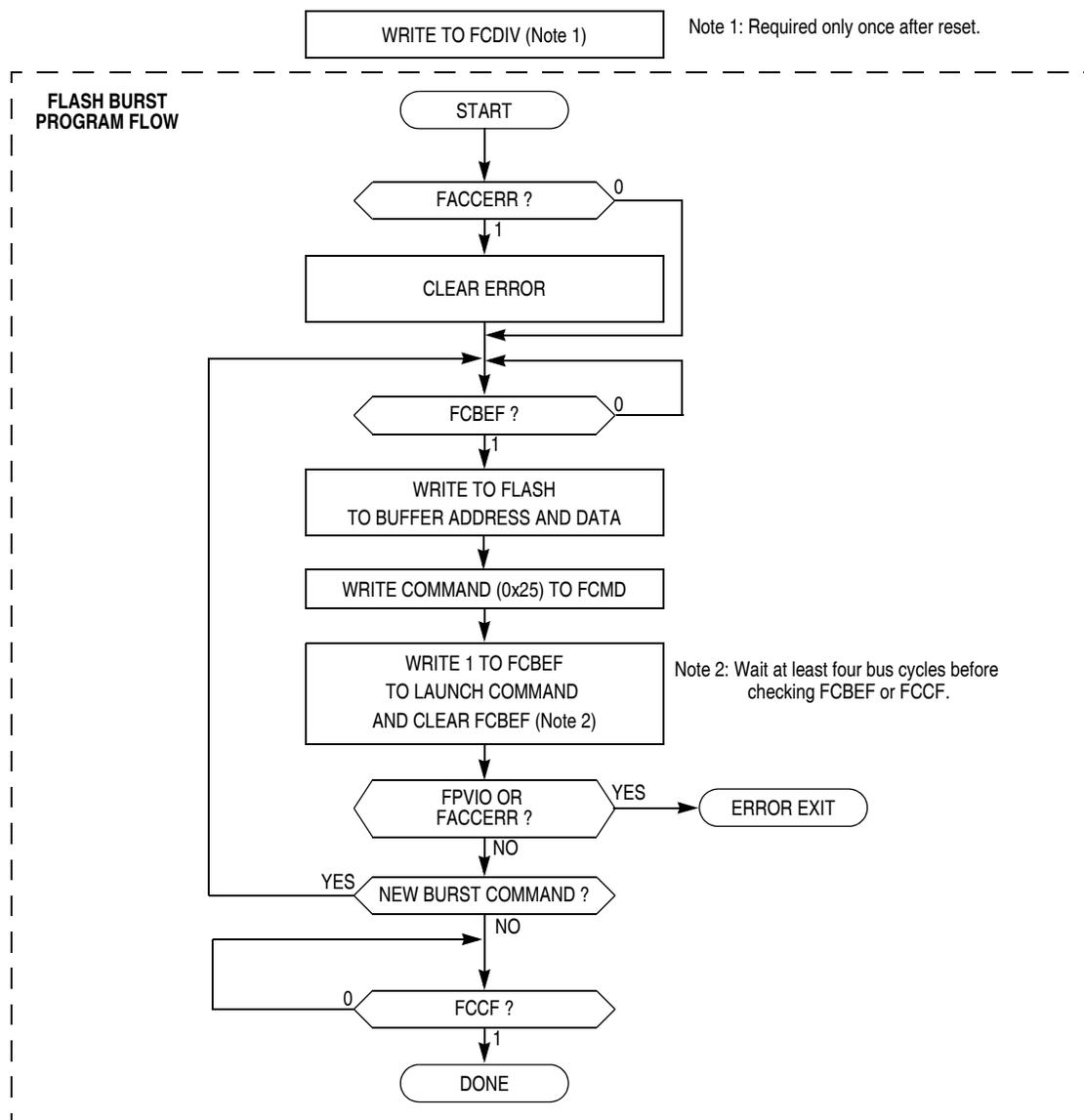


Figure 4-3. FLASH Burst Program Flowchart

### 6.4.2.3 Port A Drive Strength Select (PTADS)

An output pin can be selected to have high output drive strength by setting the corresponding bit in the drive strength select register (PTADS). When high drive is selected, a pin is capable of sourcing and sinking greater current. Even though every I/O pin can be selected as high drive, the user must ensure that the total current source and sink limits for the chip are not exceeded. Drive strength selection is intended to affect the DC behavior of I/O pins. However, the AC behavior is also affected. High drive allows a pin to drive a greater load with the same switching speed as a low drive enabled pin into a smaller load. Because of this the EMC emissions may be affected by enabling pins as high drive.

### 6.4.2.4 Port A Drive Strength Select (PTADS)

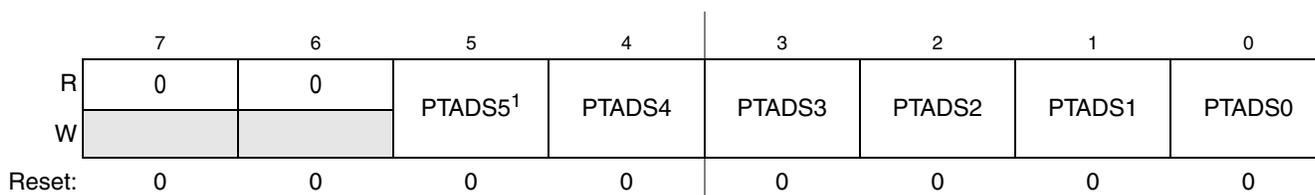


Figure 6-8. Drive Strength Selection for Port A Register (PTADS)

<sup>1</sup> PTADS5 has no effect on the input-only PTA5 pin.

Table 6-5. PTADS Register Field Descriptions

Field	Description
5:0 PTADS[5:0]	<p><b>Output Drive Strength Selection for Port A Bits</b> — Each of these control bits selects between low and high output drive for the associated PTA pin. For port A pins that are configured as inputs, these bits have no effect.</p> <p>0 Low output drive strength selected for port A bit n.</p> <p>1 High output drive strength selected for port A bit n.</p>

Table 7-2. . Instruction Set Summary (Sheet 3 of 9)

Source Form	Operation	Address Mode	Object Code	Cycles	Cyc-by-Cyc Details	Affect on CCR					
						VH	I	N	Z	C	
BRA <i>rel</i>	Branch Always (if I = 1)	REL	20 rr	3	ppp	--	---	---	---	---	
BRCLR <i>n,opr8a,rel</i>	Branch if Bit <i>n</i> in Memory Clear (if (Mn) = 0)	DIR (b0)	01 dd rr	5	rpppp	--	---	---	---	↑	
		DIR (b1)	03 dd rr	5	rpppp						
		DIR (b2)	05 dd rr	5	rpppp						
		DIR (b3)	07 dd rr	5	rpppp						
		DIR (b4)	09 dd rr	5	rpppp						
		DIR (b5)	0B dd rr	5	rpppp						
		DIR (b6)	0D dd rr	5	rpppp						
		DIR (b7)	0F dd rr	5	rpppp						
BRN <i>rel</i>	Branch Never (if I = 0)	REL	21 rr	3	ppp	--	---	---	---	---	
BRSET <i>n,opr8a,rel</i>	Branch if Bit <i>n</i> in Memory Set (if (Mn) = 1)	DIR (b0)	00 dd rr	5	rpppp	--	---	---	---	↑	
		DIR (b1)	02 dd rr	5	rpppp						
		DIR (b2)	04 dd rr	5	rpppp						
		DIR (b3)	06 dd rr	5	rpppp						
		DIR (b4)	08 dd rr	5	rpppp						
		DIR (b5)	0A dd rr	5	rpppp						
		DIR (b6)	0C dd rr	5	rpppp						
		DIR (b7)	0E dd rr	5	rpppp						
BSET <i>n,opr8a</i>	Set Bit <i>n</i> in Memory (Mn ← 1)	DIR (b0)	10 dd	5	rfwpp	--	---	---	---	---	
		DIR (b1)	12 dd	5	rfwpp						
		DIR (b2)	14 dd	5	rfwpp						
		DIR (b3)	16 dd	5	rfwpp						
		DIR (b4)	18 dd	5	rfwpp						
		DIR (b5)	1A dd	5	rfwpp						
		DIR (b6)	1C dd	5	rfwpp						
		DIR (b7)	1E dd	5	rfwpp						
BSR <i>rel</i>	Branch to Subroutine PC ← (PC) + \$0002 push (PCL); SP ← (SP) – \$0001 push (PCH); SP ← (SP) – \$0001 PC ← (PC) + <i>rel</i>	REL	AD rr	5	ssppp	--	---	---	---	---	
CBEQ <i>opr8a,rel</i> CBEQA # <i>opr8i,rel</i> CBEQX # <i>opr8i,rel</i> CBEQ <i>opr8,X+,rel</i> CBEQ <i>,X+,rel</i> CBEQ <i>opr8,SP,rel</i>	Compare and... Branch if (A) = (M) Branch if (A) = (M) Branch if (X) = (M) Branch if (A) = (M) Branch if (A) = (M) Branch if (A) = (M)	DIR	31 dd rr	5	rpppp	--	---	---	---	---	
		IMM	41 ii rr	4	pppp						
		IMM	51 ii rr	4	pppp						
		IX1+	61 ff rr	5	rpppp						
		IX+	71 rr	5	rfppp						
		SP1	9E 61 ff rr	6	prpppp						
CLC	Clear Carry Bit (C ← 0)	INH	98	1	p	--	---	---	---	0	
CLI	Clear Interrupt Mask Bit (I ← 0)	INH	9A	1	p	--	0	---	---	---	
CLR <i>opr8a</i> CLRA CLR X CLR X CLR <i>opr8,X</i> CLR <i>,X</i> CLR <i>opr8,SP</i>	Clear M ← \$00 A ← \$00 X ← \$00 H ← \$00 M ← \$00 M ← \$00 M ← \$00	DIR	3F dd	5	rfwpp	0	-	-	0	1	-
		INH	4F	1	p						
		INH	5F	1	p						
		INH	8C	1	p						
		IX1	6F ff	5	rfwpp						
		IX	7F	4	rfwp						
		SP1	9E 6F ff	6	prfwpp						

## 9.1.2 Features

Features of the ADC module include:

- Linear successive approximation algorithm with 10 bits resolution.
- Up to 28 analog inputs.
- Output formatted in 10- or 8-bit right-justified format.
- Single or continuous conversion (automatic return to idle after single conversion).
- Configurable sample time and conversion speed/power.
- Conversion complete flag and interrupt.
- Input clock selectable from up to four sources.
- Operation in wait or stop3 modes for lower noise operation.
- Asynchronous clock source for lower noise operation.
- Selectable asynchronous hardware conversion trigger.
- Automatic compare with interrupt for less-than, or greater-than or equal-to, programmable value.

## 9.1.3 Block Diagram

Figure 9-2 provides a block diagram of the ADC module

## 9.4.5 Automatic Compare Function

The compare function can be configured to check for either an upper limit or lower limit. After the input is sampled and converted, the result is added to the two's complement of the compare value (ADCCVH and ADCCVL). When comparing to an upper limit (ACFGT = 1), if the result is greater-than or equal-to the compare value, COCO is set. When comparing to a lower limit (ACFGT = 0), if the result is less than the compare value, COCO is set. The value generated by the addition of the conversion result and the two's complement of the compare value is transferred to ADCRH and ADCRL.

Upon completion of a conversion while the compare function is enabled, if the compare condition is not true, COCO is not set and no data is transferred to the result registers. An ADC interrupt is generated upon the setting of COCO if the ADC interrupt is enabled (AIEN = 1).

### NOTE

The compare function can be used to monitor the voltage on a channel while the MCU is in either wait or stop3 mode. The ADC interrupt will wake the MCU when the compare condition is met.

## 9.4.6 MCU Wait Mode Operation

The WAIT instruction puts the MCU in a lower power-consumption standby mode from which recovery is very fast because the clock sources remain active. If a conversion is in progress when the MCU enters wait mode, it continues until completion. Conversions can be initiated while the MCU is in wait mode by means of the hardware trigger or if continuous conversions are enabled.

The bus clock, bus clock divided by two, and ADACK are available as conversion clock sources while in wait mode. The use of ALTCLK as the conversion clock source in wait is dependent on the definition of ALTCLK for this MCU. Consult the module introduction for information on ALTCLK specific to this MCU.

A conversion complete event sets the COCO and generates an ADC interrupt to wake the MCU from wait mode if the ADC interrupt is enabled (AIEN = 1).

## 9.4.7 MCU Stop3 Mode Operation

The STOP instruction is used to put the MCU in a low power-consumption standby mode during which most or all clock sources on the MCU are disabled.

### 9.4.7.1 Stop3 Mode With ADACK Disabled

If the asynchronous clock, ADACK, is not selected as the conversion clock, executing a STOP instruction aborts the current conversion and places the ADC in its idle state. The contents of ADCRH and ADCRL are unaffected by stop3 mode. After exiting from stop3 mode, a software or hardware trigger is required to resume conversions.

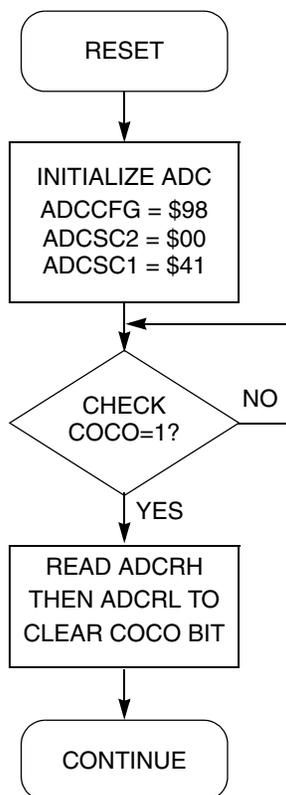


Figure 9-14. Initialization Flowchart for Example

## 9.6 Application Information

This section contains information for using the ADC module in applications. The ADC has been designed to be integrated into a microcontroller for use in embedded control applications requiring an A/D converter.

### 9.6.1 External Pins and Routing

The following sections discuss the external pins associated with the ADC module and how they should be used for best results.

#### 9.6.1.1 Analog Supply Pins

The ADC module has analog power and ground supplies ( $V_{DDAD}$  and  $V_{SSAD}$ ) which are available as separate pins on some devices. On other devices,  $V_{SSAD}$  is shared on the same pin as the MCU digital  $V_{SS}$ , and on others, both  $V_{SSAD}$  and  $V_{DDAD}$  are shared with the MCU digital supply pins. In these cases, there are separate pads for the analog supplies which are bonded to the same pin as the corresponding digital supply so that some degree of isolation between the supplies is maintained.

When available on a separate pin, both  $V_{DDAD}$  and  $V_{SSAD}$  must be connected to the same voltage potential as their corresponding MCU digital supply ( $V_{DD}$  and  $V_{SS}$ ) and must be routed carefully for maximum noise immunity and bypass capacitors placed as near as possible to the package.

## 13.1.2 Features

Timer system features include:

- 8-bit up-counter
  - Free-running or 8-bit modulo limit
  - Software controllable interrupt on overflow
  - Counter reset bit (TRST)
  - Counter stop bit (TSTP)
- Four software selectable clock sources for input to prescaler:
  - System bus clock — rising edge
  - Fixed frequency clock (XCLK) — rising edge
  - External clock source on the TCLK pin — rising edge
  - External clock source on the TCLK pin — falling edge
- Nine selectable clock prescale values:
  - Clock source divide by 1, 2, 4, 8, 16, 32, 64, 128, or 256

## 13.1.3 Modes of Operation

This section defines the MTIM's operation in stop, wait and background debug modes.

### 13.1.3.1 MTIM in Wait Mode

The MTIM continues to run in wait mode if enabled before executing the WAIT instruction. Therefore, the MTIM can be used to bring the MCU out of wait mode if the timer overflow interrupt is enabled. For lowest possible current consumption, the MTIM should be stopped by software if not needed as an interrupt source during wait mode.

### 13.1.3.2 MTIM in Stop Modes

The MTIM is disabled in all stop modes, regardless of the settings before executing the STOP instruction. Therefore, the MTIM cannot be used as a wake up source from stop modes.

Waking from stop1 and stop2 modes, the MTIM will be put into its reset state. If stop3 is exited with a reset, the MTIM will be put into its reset state. If stop3 is exited with an interrupt, the MTIM continues from the state it was in when stop3 was entered. If the counter was active upon entering stop3, the count will resume from the current value.

### 13.1.3.3 MTIM in Active Background Mode

The MTIM suspends all counting until the microcontroller returns to normal user operating mode. Counting resumes from the suspended value as long as an MTIM reset did not occur (TRST written to a 1 or MTIMMOD written).

**Figure 13-3. MTIM Register Summary**

Name		7	6	5	4	3	2	1	0
MTIMSC	R	TOF	TOIE	0	TSTP	0	0	0	0
	W			TRST					
MTIMCLK	R	0	0	CLKS		PS			
	W								
MTIMCNT	R	COUNT							
	W								
MTIMMOD	R	MOD							
	W								

Each MTIM includes four registers:

- An 8-bit status and control register
- An 8-bit clock configuration register
- An 8-bit counter register
- An 8-bit modulo register

Refer to the direct-page register summary in the [Memory](#) chapter of this data sheet for the absolute address assignments for all MTIM registers. This section refers to registers and control bits only by their names and relative address offsets.

Some MCUs may have more than one MTIM, so register names include placeholder characters to identify which MTIM is being referenced.

### 13.3.2 MTIM Clock Configuration Register (MTIMCLK)

MTIMCLK contains the clock select bits (CLKS) and the prescaler select bits (PS).

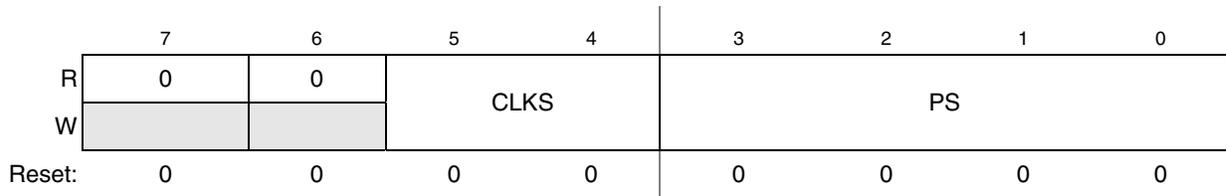


Figure 13-5. MTIM Clock Configuration Register

Table 13-3. MTIM Clock Configuration Register Field Description

Field	Description
7:6	Unused register bits, always read 0.
5:4 CLKS	<p><b>Clock Source Select</b> — These two read/write bits select one of four different clock sources as the input to the MTIM prescaler. Changing the clock source while the counter is active does not clear the counter. The count continues with the new clock source. Reset clears CLKS to 000.</p> <p>00 Encoding 0. Bus clock (BUSCLK)                      01 Encoding 1. Fixed-frequency clock (XCLK)                      10 Encoding 3. External source (TCLK pin), falling edge                      11 Encoding 4. External source (TCLK pin), rising edge                      All other encodings default to the bus clock (BUSCLK).</p>
3:0 PS	<p><b>Clock Source Prescaler</b> — These four read/write bits select one of nine outputs from the 8-bit prescaler. Changing the prescaler value while the counter is active does not clear the counter. The count continues with the new prescaler value. Reset clears PS to 0000.</p> <p>0000 Encoding 0. MTIM clock source ÷ 1                      0001 Encoding 1. MTIM clock source ÷ 2                      0010 Encoding 2. MTIM clock source ÷ 4                      0011 Encoding 3. MTIM clock source ÷ 8                      0100 Encoding 4. MTIM clock source ÷ 16                      0101 Encoding 5. MTIM clock source ÷ 32                      0110 Encoding 6. MTIM clock source ÷ 64                      0111 Encoding 7. MTIM clock source ÷ 128                      1000 Encoding 8. MTIM clock source ÷ 256                      All other encodings default to MTIM clock source ÷ 256.</p>

**Table 14-5. SCIS1 Register Field Descriptions (continued)**

Field	Description
5 RDRF	<p><b>Receive Data Register Full Flag</b> — RDRF becomes set when a character transfers from the receive shifter into the receive data register (SCID). To clear RDRF, read SCIS1 with RDRF = 1 and then read the SCI data register (SCID).</p> <p>0 Receive data register empty. 1 Receive data register full.</p>
4 IDLE	<p><b>Idle Line Flag</b> — IDLE is set when the SCI receive line becomes idle for a full character time after a period of activity. When ILT = 0, the receiver starts counting idle bit times after the start bit. So if the receive character is all 1s, these bit times and the stop bit time count toward the full character time of logic high (10 or 11 bit times depending on the M control bit) needed for the receiver to detect an idle line. When ILT = 1, the receiver doesn't start counting idle bit times until after the stop bit. So the stop bit and any logic high bit times at the end of the previous character do not count toward the full character time of logic high needed for the receiver to detect an idle line.</p> <p>To clear IDLE, read SCIS1 with IDLE = 1 and then read the SCI data register (SCID). After IDLE has been cleared, it cannot become set again until after a new character has been received and RDRF has been set. IDLE will get set only once even if the receive line remains idle for an extended period.</p> <p>0 No idle line detected. 1 Idle line was detected.</p>
3 OR	<p><b>Receiver Overrun Flag</b> — OR is set when a new serial character is ready to be transferred to the receive data register (buffer), but the previously received character has not been read from SCID yet. In this case, the new character (and all associated error information) is lost because there is no room to move it into SCID. To clear OR, read SCIS1 with OR = 1 and then read the SCI data register (SCID).</p> <p>0 No overrun. 1 Receive overrun (new SCI data lost).</p>
2 NF	<p><b>Noise Flag</b> — The advanced sampling technique used in the receiver takes seven samples during the start bit and three samples in each data bit and the stop bit. If any of these samples disagrees with the rest of the samples within any bit time in the frame, the flag NF will be set at the same time as the flag RDRF gets set for the character. To clear NF, read SCIS1 and then read the SCI data register (SCID).</p> <p>0 No noise detected. 1 Noise detected in the received character in SCID.</p>
1 FE	<p><b>Framing Error Flag</b> — FE is set at the same time as RDRF when the receiver detects a logic 0 where the stop bit was expected. This suggests the receiver was not properly aligned to a character frame. To clear FE, read SCIS1 with FE = 1 and then read the SCI data register (SCID).</p> <p>0 No framing error detected. This does not guarantee the framing is correct. 1 Framing error.</p>
0 PF	<p><b>Parity Error Flag</b> — PF is set at the same time as RDRF when parity is enabled (PE = 1) and the parity bit in the received character does not agree with the expected parity value. To clear PF, read SCIS1 and then read the SCI data register (SCID).</p> <p>0 No parity error. 1 Parity error.</p>

**Table 15-5. SPI Baud Rate Prescaler Divisor**

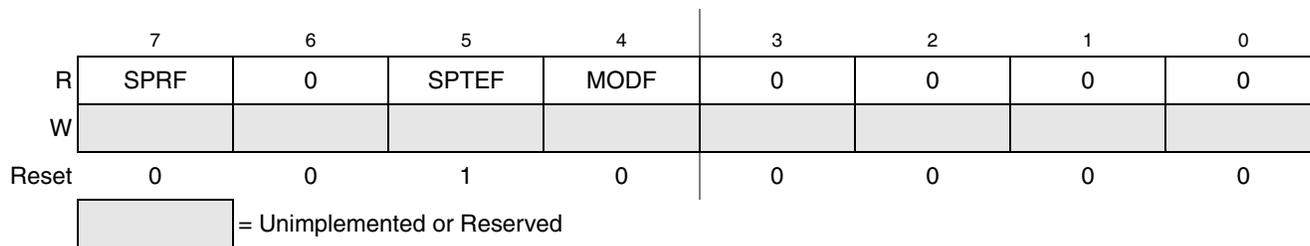
SPPR2:SPPR1:SPPR0	Prescaler Divisor
0:0:0	1
0:0:1	2
0:1:0	3
0:1:1	4
1:0:0	5
1:0:1	6
1:1:0	7
1:1:1	8

**Table 15-6. SPI Baud Rate Divisor**

SPR2:SPR1:SPR0	Rate Divisor
0:0:0	2
0:0:1	4
0:1:0	8
0:1:1	16
1:0:0	32
1:0:1	64
1:1:0	128
1:1:1	256

### 15.4.4 SPI Status Register (SPIS)

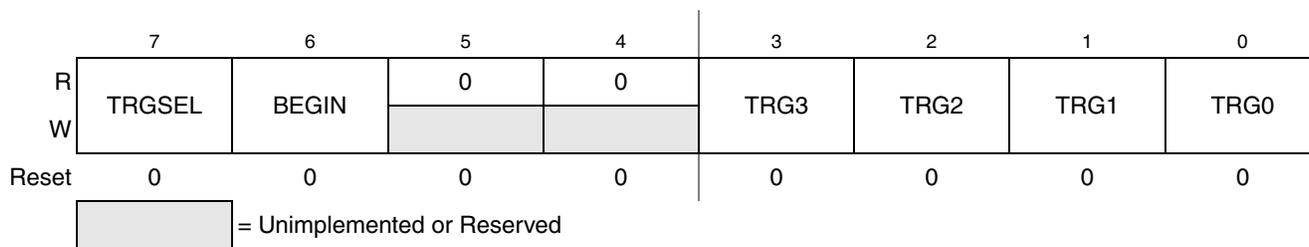
This register has three read-only status bits. Bits 6, 3, 2, 1, and 0 are not implemented and always read 0. Writes have no meaning or effect.



**Figure 15-8. SPI Status Register (SPIS)**

### 17.4.3.8 Debug Trigger Register (DBGT)

This register can be read any time, but may be written only if ARM = 0, except bits 4 and 5 are hard-wired to 0s.



**Figure 17-8. Debug Trigger Register (DBGT)**

**Table 17-5. DBGT Register Field Descriptions**

Field	Description
7 TRGSEL	<p><b>Trigger Type</b> — Controls whether the match outputs from comparators A and B are qualified with the opcode tracking logic in the debug module. If TRGSEL is set, a match signal from comparator A or B must propagate through the opcode tracking logic and a trigger event is only signalled to the FIFO logic if the opcode at the match address is actually executed.</p> <p>0 Trigger on access to compare address (force) 1 Trigger if opcode at compare address is executed (tag)</p>
6 BEGIN	<p><b>Begin/End Trigger Select</b> — Controls whether the FIFO starts filling at a trigger or fills in a circular manner until a trigger ends the capture of information. In event-only trigger modes, this bit is ignored and all debug runs are assumed to be begin traces.</p> <p>0 Data stored in FIFO until trigger (end trace) 1 Trigger initiates data storage (begin trace)</p>
3:0 TRG[3:0]	<p><b>Select Trigger Mode</b> — Selects one of nine triggering modes, as described below.</p> <p>0000 A-only 0001 A OR B 0010 A Then B 0011 Event-only B (store data) 0100 A then event-only B (store data) 0101 A AND B data (full mode) 0110 A AND NOT B data (full mode) 0111 Inside range: <math>A \leq \text{address} \leq B</math> 1000 Outside range: <math>\text{address} &lt; A</math> or <math>\text{address} &gt; B</math> 1001 – 1111 (No trigger)</p>

### 17.4.3.9 Debug Status Register (DBGS)

This is a read-only status register.

# Appendix A

## Electrical Characteristics

### A.1 Introduction

This section contains electrical and timing specifications.

### A.2 Absolute Maximum Ratings

Absolute maximum ratings are stress ratings only, and functional operation at the maxima is not guaranteed. Stress beyond the limits specified in [Table A-1](#) may affect device reliability or cause permanent damage to the device. For functional operating conditions, refer to the remaining tables in this section.

This device contains circuitry protecting against damage due to high static voltage or electrical fields; however, it is advised that normal precautions be taken to avoid application of any voltages higher than maximum-rated voltages to this high-impedance circuit. Reliability of operation is enhanced if unused inputs are tied to an appropriate logic voltage level (for instance, either  $V_{SS}$  or  $V_{DD}$ ) or the programmable pull-up resistor associated with the pin is enabled.

**Table A-1. Absolute Maximum Ratings**

Rating	Symbol	Value	Unit
Supply voltage	$V_{DD}$	-0.3 to +3.8	V
Maximum current into $V_{DD}$	$I_{DD}$	120	mA
Digital input voltage	$V_{In}$	-0.3 to $V_{DD} + 0.3$	V
Instantaneous maximum current Single pin limit (applies to all port pins) <sup>1, 2, 3</sup>	$I_D$	± 25	mA
Storage temperature range	$T_{stg}$	-55 to 150	°C

<sup>1</sup> Input must be current limited to the value specified. To determine the value of the required current-limiting resistor, calculate resistance values for positive ( $V_{DD}$ ) and negative ( $V_{SS}$ ) clamp voltages, then use the larger of the two resistance values.

<sup>2</sup> All functional non-supply pins are internally clamped to  $V_{SS}$  and  $V_{DD}$ .

<sup>3</sup> Power supply must maintain regulation within operating  $V_{DD}$  range during instantaneous and operating maximum current conditions. If positive injection current ( $V_{In} > V_{DD}$ ) is greater than  $I_{DD}$ , the injection current may flow out of  $V_{DD}$  and could result in external power supply going out of regulation. Ensure external  $V_{DD}$  load will shunt current greater than maximum injection current. This will be the greatest risk when the MCU is not consuming power. Examples are: if no system clock is present, or if the clock rate is very low (which would reduce overall power consumption).

## A.5 DC Characteristics

This section includes information about power supply requirements and I/O pin characteristics.

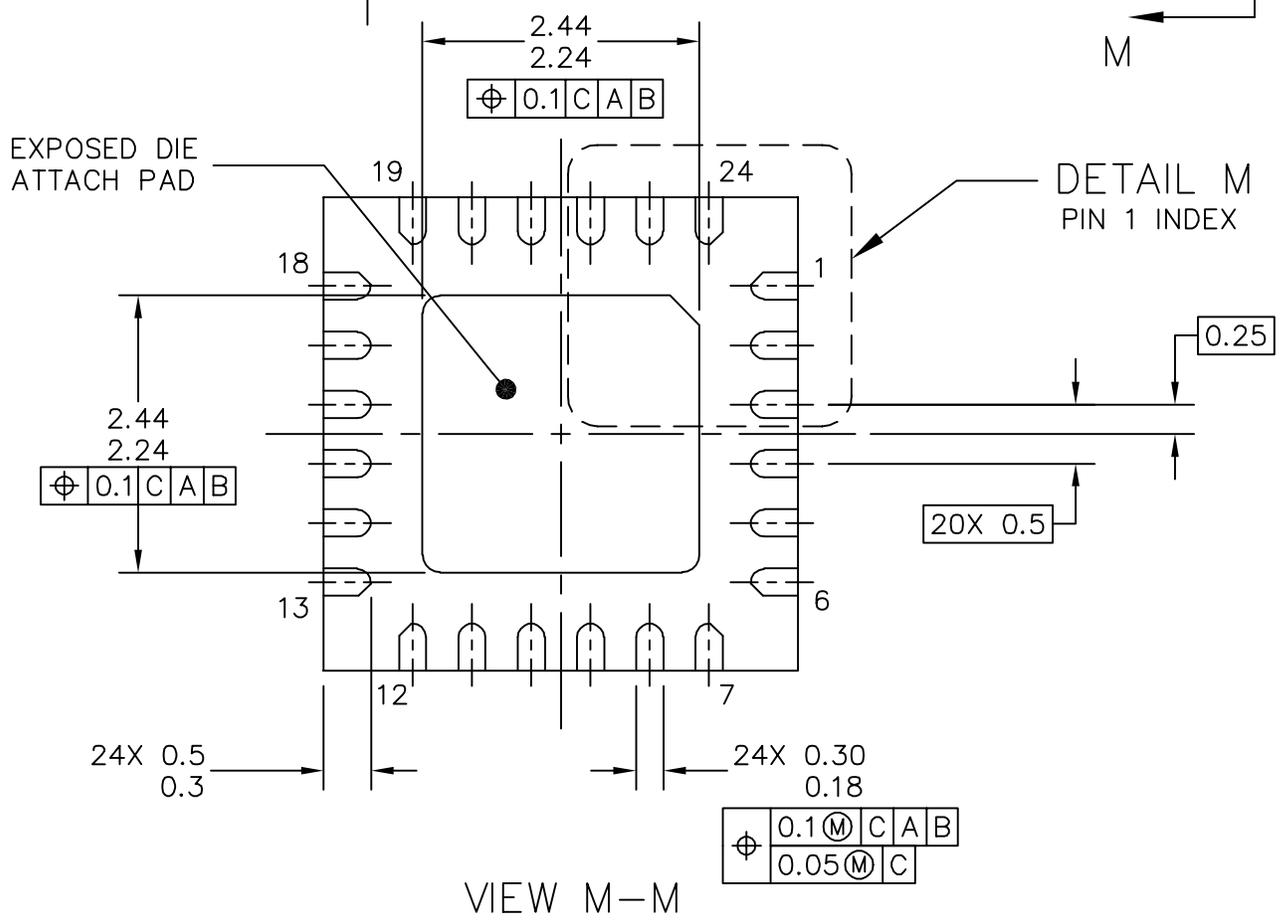
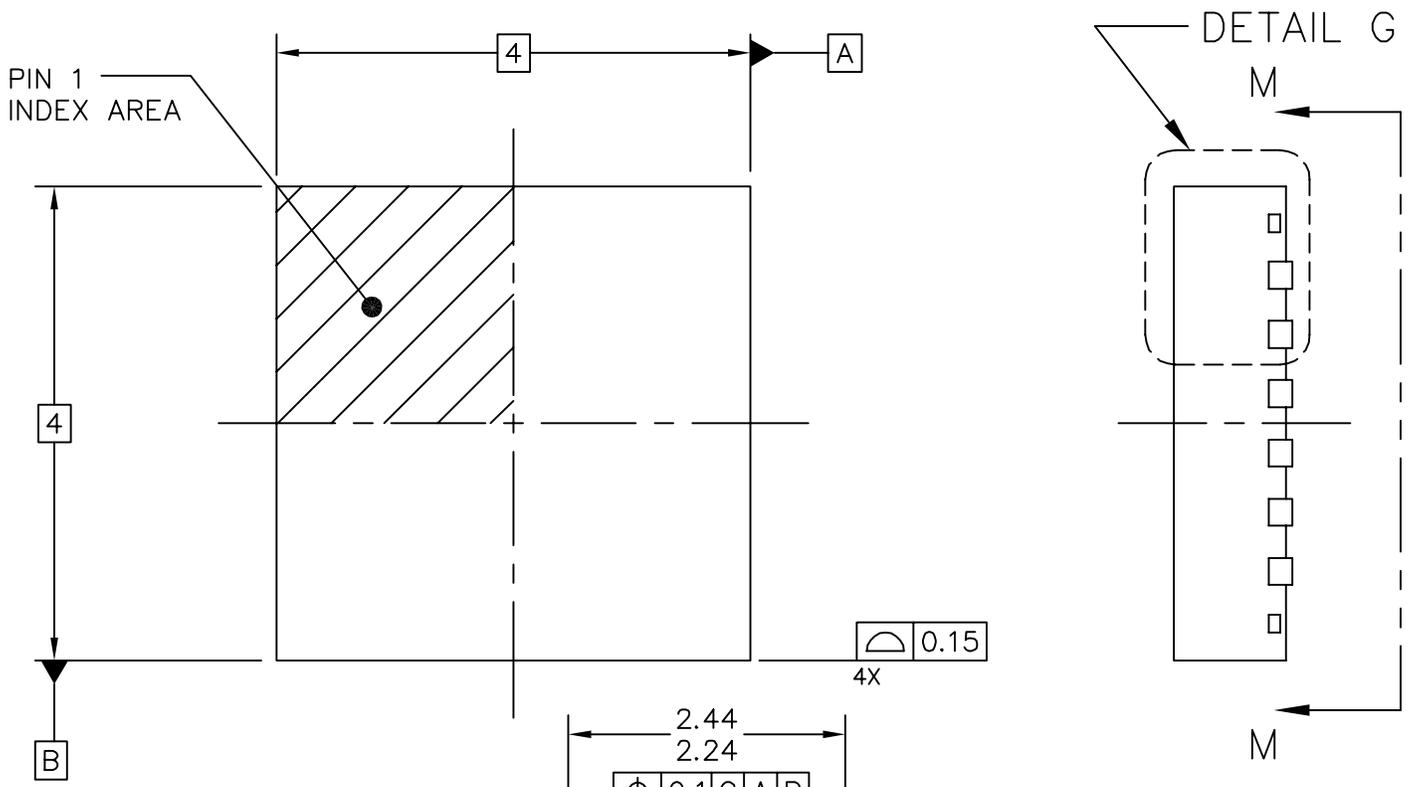
**Table A-5. Operating Range**

Parameter	Symbol	Min	Typical	Max	Unit
Supply voltage (run, wait and stop modes.)	$V_{DD}$	1.8 <sup>1</sup>		3.6	V
Temperature	C	-40	—	85	°C
	M	-40	—	125	

<sup>1</sup> As the supply voltage rises, the LVD circuit will hold the MCU in reset until the supply has risen above  $V_{LVDL}$ .

**Table A-6. DC Characteristics**

Parameter	Symbol	Min	Typical	Max	Unit
Minimum RAM retention supply voltage applied to $V_{DD}$	$V_{RAM}$	$V_{POR}$ <sup>1, 2</sup>		—	V
Low-voltage detection threshold — high range ( $V_{DD}$ falling) ( $V_{DD}$ rising)	$V_{LVDH}$	2.08	2.1	2.2	V
		2.16	2.19	2.27	
Low-voltage detection threshold — low range ( $V_{DD}$ falling) ( $V_{DD}$ rising)	$V_{LVDL}$	1.80	1.82	1.91	V
		1.88	1.90	1.99	
Low-voltage warning threshold — high range ( $V_{DD}$ falling) ( $V_{DD}$ rising)	$V_{LVWH}$	2.35	2.40	2.5	V
		2.35	2.40	2.5	
Low-voltage warning threshold — low range ( $V_{DD}$ falling) ( $V_{DD}$ rising)	$V_{LVWL}$	2.08	2.1	2.2	V
		2.16	2.19	2.27	
Power on reset (POR) re-arm voltage	$V_{POR}$		1.4		V
Bandgap Voltage Reference	$V_{BG}$	1.18	1.20	1.21	V
Input high voltage ( $V_{DD} > 2.3$ V) (all digital inputs)	$V_{IH}$	$0.70 \times V_{DD}$		—	V
Input high voltage ( $1.8$ V $\leq V_{DD} \leq 2.3$ V) (all digital inputs)		$0.85 \times V_{DD}$		—	
Input low voltage ( $V_{DD} > 2.3$ V) (all digital inputs)	$V_{IL}$	—		$0.35 \times V_{DD}$	V
Input low voltage ( $1.8$ V $\leq V_{DD} \leq 2.3$ V) (all digital inputs)		—		$0.30 \times V_{DD}$	
Input hysteresis (all digital inputs)	$V_{hys}$	$0.06 \times V_{DD}$		—	V
Input leakage current (Per pin) $V_{In} = V_{DD}$ or $V_{SS}$ , all input only pins	$ I_{In} $	—	0.025	1.0	$\mu$ A
High impedance (off-state) leakage current (per pin) $V_{In} = V_{DD}$ or $V_{SS}$ , all input/output	$ I_{OZ} $	—	0.025	1.0	$\mu$ A
Internal pullup resistors <sup>3,4</sup>	$R_{PU}$	17.5		52.5	k $\Omega$



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