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Details

Product Status	Active
Core Processor	HCS12
Core Size	16-Bit
Speed	25MHz
Connectivity	EBI/EMI, SCI, SPI
Peripherals	POR, PWM, WDT
Number of I/O	35
Program Memory Size	16KB (16K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	1K x 8
Voltage - Supply (Vcc/Vdd)	2.35V ~ 5.5V
Data Converters	A/D 8x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	52-LQFP
Supplier Device Package	52-QFP (10x10)
Purchase URL	https://www.e-xfl.com/pro/item?MUrl=&PartUrl=mc9s12gc16cpbe

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Chapter 7

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Chapter 8

Analog-to-Digital Converter (ATD10B8C) Block Description

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1.3.5.4 V_{DDA} , V_{SSA} — Power Supply Pins for ATD and VREG

V_{DDA} , V_{SSA} are the power supply and ground input pins for the voltage regulator reference and the analog to digital converter.

1.3.5.5 V_{RH} , V_{RL} — ATD Reference Voltage Input Pins

V_{RH} and V_{RL} are the reference voltage input pins for the analog to digital converter.

1.3.5.6 V_{DDPLL} , V_{SSPLL} — Power Supply Pins for PLL

Provides operating voltage and ground for the oscillator and the phased-locked loop. This allows the supply voltage to the oscillator and PLL to be bypassed independently. This 2.5V voltage is generated by the internal voltage regulator.

Table 1-6. Power and Ground Connection Summary

Mnemonic	Nominal Voltage (V)	Description
V_{DD1} , V_{DD2}	2.5	Internal power and ground generated by internal regulator. These also allow an external source to supply the core V_{DD}/V_{SS} voltages and bypass the internal voltage regulator. In the 48 and 52 LQFP packages V_{DD2} and V_{SS2} are not available.
V_{SS1} , V_{SS2}	0	
V_{DDR}	5.0	External power and ground, supply to internal voltage regulator.
V_{SSR}	0	
V_{DDX}	5.0	External power and ground, supply to pin drivers.
V_{SSX}	0	
V_{DDA}	5.0	Operating voltage and ground for the analog-to-digital converters and the reference for the internal voltage regulator, allows the supply voltage to the A/D to be bypassed independently.
V_{SSA}	0	
V_{RH}	5.0	Reference voltage low for the ATD converter. In the 48 and 52 LQFP packages V_{RL} is bonded to V_{SSA} .
V_{RL}	0	
V_{DDPLL}	2.5	Provides operating voltage and ground for the phased-locked loop. This allows the supply voltage to the PLL to be bypassed independently. Internal power and ground generated by internal regulator.
V_{SSPLL}	0	

NOTE

All V_{SS} pins must be connected together in the application. Because fast signal transitions place high, short-duration current demands on the power supply, use bypass capacitors with high-frequency characteristics and place them as close to the MCU as possible. Bypass requirements depend on MCU pin load.

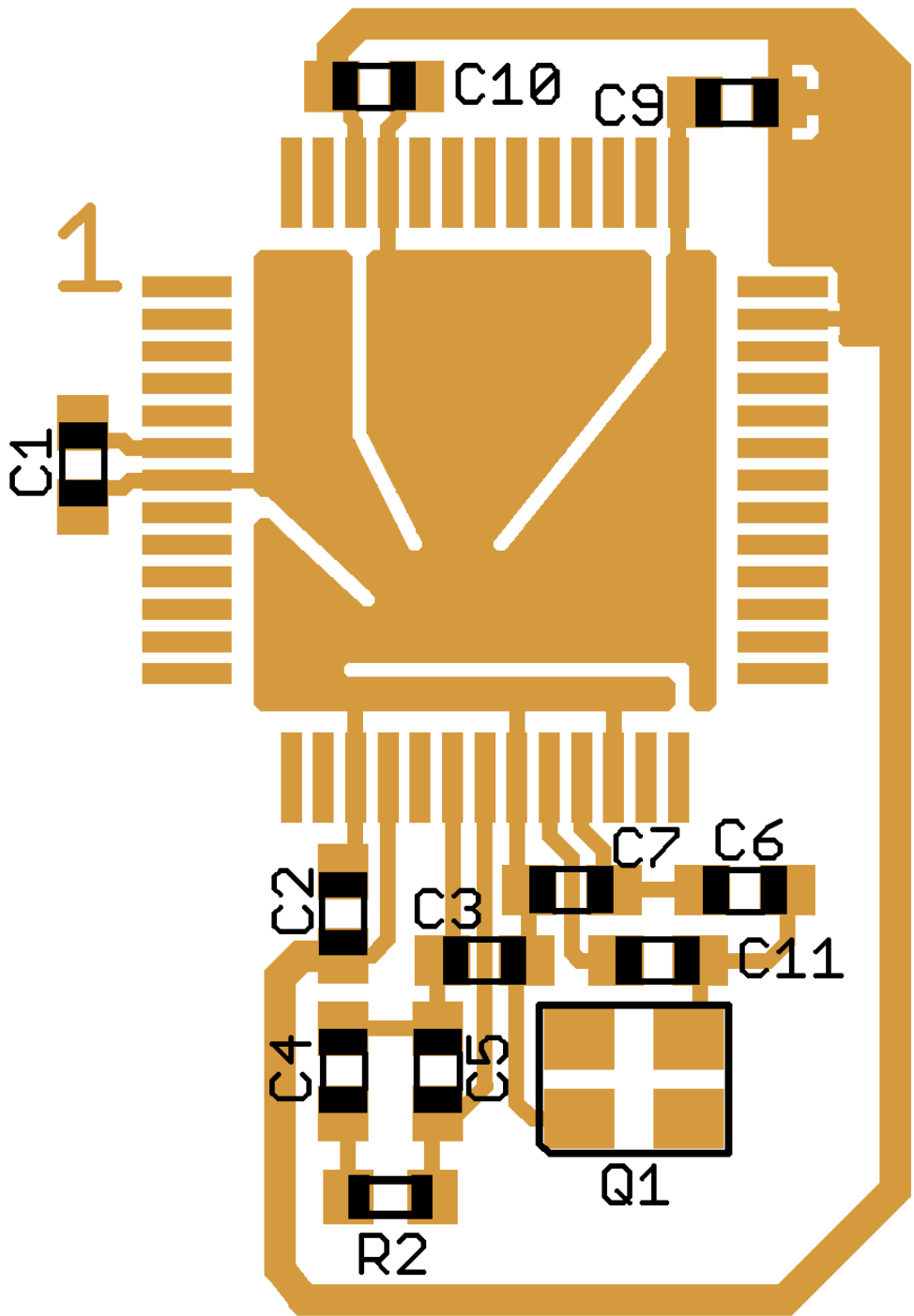


Figure 1-16. Recommended PCB Layout (52 LQFP) Colpitts Oscillator

Table 2-1. Pin Functions and Priorities (continued)

Port	Pin Name	Pin Function	Description	Pin Function after Reset
Port E	PE7	NOACC/ XCLKS/ GPIO	Refer to MEBI Block Guide.	
	PE6	IPIPE1/ MODB/ GPIO		
	PE5	IPIPE0/ MODA/ GPIO		
	PE4	ECLK/GPIO		
	PE3	LSTRB/ TAGLO/ GPIO		
	PE2	R/W/ GPIO		
	PE1	IRQ/GPI		
	PE0	XIRQ/GPI		

2.3 Memory Map and Registers

This section provides a detailed description of all registers.

2.3.1 Module Memory Map

Figure 2-2 shows the register map of the Port Integration Module.

Address	Name		Bit 7	6	5	4	3	2	1	Bit 0
0x0000	PTT	R	PTT7	PTT6	PTT5	PTT4	PTT3	PTT2	PTT1	PTT0
		W								
		TIM	IOC7	IOC6	IOC5	IOC4	IOC3	IOC2	IOC1	IOC0
0x0001	PTIT	PWM				PWM4	PWM3	PWM2	PWM1	PWM0
		R	PTIT7	PTIT6	PTIT5	PTIT4	PTIT3	PTIT2	PTIT1	PTIT0
0x0002	DDRT	W								
		R	DDRT7	DDRT6	DDRT5	DDRT4	DDRT3	DDRT2	DDRT1	DDRT0
0x0003	RDRT	W								
		R	RDRT7	RDRT6	RDRT5	RDRT4	RDRT3	RDRT2	RDRT1	RDRT0
0x0004	PERT	W								
		R	PERT7	PERT6	PERT5	PERT4	PERT3	PERT2	PERT1	PERT0
0x0005	PPST	W								
		R	PPST7	PPST6	PPST5	PPST4	PPST3	PPST2	PPST1	PPST0

= Unimplemented or Reserved

Figure 2-2. Quick Reference to PIM Registers (Sheet 1 of 3)

Table 3-14. Program Page Index Register Bits

PIX5	PIX4	PIX3	PIX2	PIX1	PIX0	Program Space Selected
0	0	0	0	0	0	16K page 0
0	0	0	0	0	1	16K page 1
0	0	0	0	1	0	16K page 2
0	0	0	0	1	1	16K page 3
.	.	.	.	.	.	.
.	.	.	.	.	.	.
.	.	.	.	.	.	.
.	.	.	.	.	.	.
.	.	.	.	.	.	.
1	1	1	1	0	0	16K page 60
1	1	1	1	0	1	16K page 61
1	1	1	1	1	0	16K page 62
1	1	1	1	1	1	16K page 63

3.4 Functional Description

The MMC sub-block performs four basic functions of the core operation: bus control, address decoding and select signal generation, memory expansion, and security decoding for the system. Each aspect is described in the following subsections.

3.4.1 Bus Control

The MMC controls the address bus and data buses that interface the core with the rest of the system. This includes the multiplexing of the input data buses to the core onto the main CPU read data bus and control of data flow from the CPU to the output address and data buses of the core. In addition, the MMC manages all CPU read data bus swapping operations.

3.4.2 Address Decoding

As data flows on the core address bus, the MMC decodes the address information, determines whether the internal core register or firmware space, the peripheral space or a memory register or array space is being addressed and generates the correct select signal. This decoding operation also interprets the mode of operation of the system and the state of the mapping control registers in order to generate the proper select. The MMC also generates two external chip select signals, emulation chip select ($\overline{\text{ECS}}$) and external chip select ($\overline{\text{XCS}}$).

3.4.2.1 Select Priority and Mode Considerations

Although internal resources such as control registers and on-chip memory have default addresses, each can be relocated by changing the default values in control registers. Normally, I/O addresses, control registers,

Table 4-8. MODC, MODB, and MODA Write Capability⁽¹⁾

MODC	MODB	MODA	Mode	MODx Write Capability
0	0	0	Special single chip	MODC, MODB, and MODA write anytime but not to 110 ⁽²⁾
0	0	1	Emulation narrow	No write
0	1	0	Special test	MODC, MODB, and MODA write anytime but not to 110 ⁽²⁾
0	1	1	Emulation wide	No write
1	0	0	Normal single chip	MODC write never, MODB and MODA write once but not to 110
1	0	1	Normal expanded narrow	No write
1	1	0	Special peripheral	No write
1	1	1	Normal expanded wide	No write

1. No writes to the MOD bits are allowed while operating in a secure mode. For more details, refer to the device overview chapter.
2. If you are in a special single-chip or special test mode and you write to this register, changing to normal single-chip mode, then one allowed write to this register remains. If you write to normal expanded or emulation mode, then no writes remain.

4.3.2.10 Pull Control Register (PUCR)

Module Base + 0x000C

Starting address location affected by INITRG register setting.

	7	6	5	4	3	2	1	0
R	PUPKE	0	0	PUPEE	0	0	PUPBE	PUPAE
W								
Reset ¹	1	0	0	1	0	0	0	0

NOTES:

1. The default value of this parameter is shown. Please refer to the device overview chapter to determine the actual reset state of this register.

 = Unimplemented or Reserved

Figure 4-14. Pull Control Register (PUCR)

Read: Anytime (provided this register is in the map).

Write: Anytime (provided this register is in the map).

This register is used to select pull resistors for the pins associated with the core ports. Pull resistors are assigned on a per-port basis and apply to any pin in the corresponding port that is currently configured as an input. The polarity of these pull resistors is determined by chip integration. Please refer to the device overview chapter to determine the polarity of these resistors.

Table 7-3. DBGC1 Field Descriptions (continued)

Field	Description
3 DBGBRK	DBG Breakpoint Enable Bit — The DBGBRK bit controls whether the debugger will request a breakpoint based on comparator A and B to the CPU upon completion of a tracing session. Please refer to Section 7.4.3, “Breakpoints,” for further details. 0 CPU break request not enabled 1 CPU break request enabled
1:0 CAPMOD	Capture Mode Field — See Table 7-4 for capture mode field definitions. In LOOP1 mode, the debugger will automatically inhibit redundant entries into capture memory. In detail mode, the debugger is storing address and data for all cycles except program fetch (P) and free (f) cycles. In profile mode, the debugger is returning the address of the last instruction executed by the CPU on each access of trace buffer address. Refer to Section 7.4.2.6, “Capture Modes,” for more information.

Table 7-4. CAPMOD Encoding

CAPMOD	Description
00	Normal
01	LOOP1
10	DETAIL
11	PROFILE

Table 8-1. ATDCTL2 Field Descriptions (continued)

Field	Description
1 ASCIE	ATD Sequence Complete Interrupt Enable 0 ATD Sequence Complete interrupt requests are disabled. 1 ATD Interrupt will be requested whenever ASCIF = 1 is set.
0 ASCIF	ATD Sequence Complete Interrupt Flag — If ASCIE = 1 the ASCIF flag equals the SCF flag (see Section 8.3.2.7, “ATD Status Register 0 (ATDSTAT0)”), else ASCIF reads zero. Writes have no effect. 0 No ATD interrupt occurred 1 ATD sequence complete interrupt pending

Table 8-2. External Trigger Configurations

ETRIGLE	ETRIGP	External Trigger Sensitivity
0	0	Falling edge
0	1	Rising edge
1	0	Low level
1	1	High level

8.3.2.4 ATD Control Register 3 (ATDCTL3)

This register controls the conversion sequence length, FIFO for results registers and behavior in Freeze Mode. Writes to this register will abort current conversion sequence but will not start a new sequence.

Module Base + 0x0003

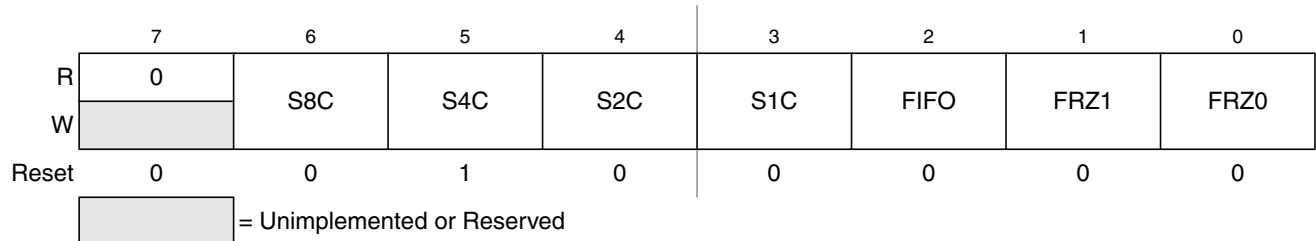


Figure 8-6. ATD Control Register 3 (ATDCTL3)

Read: Anytime

Write: Anytime

Table 8-3. ATDCTL3 Field Descriptions

Field	Description
6–3 S8C, S4C, S2C, S1C	Conversion Sequence Length — These bits control the number of conversions per sequence. Table 8-4 shows all combinations. At reset, S4C is set to 1 (sequence length is 4). This is to maintain software continuity to HC12 Family.

9.1.2 Modes of Operation

This subsection lists and briefly describes all operating modes supported by the CRG.

- **Run mode**
All functional parts of the CRG are running during normal run mode. If RTI or COP functionality is required the individual bits of the associated rate select registers (COPCTL, RTICTL) have to be set to a nonzero value.
- **Wait mode**
This mode allows to disable the system and core clocks depending on the configuration of the individual bits in the CLKSEL register.
- **Stop mode**
Depending on the setting of the PSTP bit, stop mode can be differentiated between full stop mode (PSTP = 0) and pseudo-stop mode (PSTP = 1).
 - **Full stop mode**
The oscillator is disabled and thus all system and core clocks are stopped. The COP and the RTI remain frozen.
 - **Pseudo-stop mode**
The oscillator continues to run and most of the system and core clocks are stopped. If the respective enable bits are set the COP and RTI will continue to run, else they remain frozen.
- **Self-clock mode**
Self-clock mode will be entered if the clock monitor enable bit (CME) and the self-clock mode enable bit (SCME) are both asserted and the clock monitor in the oscillator block detects a loss of clock. As soon as self-clock mode is entered the CRGV4 starts to perform a clock quality check. Self-clock mode remains active until the clock quality check indicates that the required quality of the incoming clock signal is met (frequency and amplitude). Self-clock mode should be used for safety purposes only. It provides reduced functionality to the MCU in case a loss of clock is causing severe system conditions.

9.1.3 Block Diagram

Figure 9-1 shows a block diagram of the CRGV4.

Table 10-9. CANRFLG Register Field Descriptions

Field	Description
7 WUPIF	Wake-Up Interrupt Flag — If the MSCAN detects CAN bus activity while in sleep mode (see Section 10.4.5.4, “MSCAN Sleep Mode,”) and WUPE = 1 in CANTCTL0 (see Section 10.3.2.1, “MSCAN Control Register 0 (CANCTL0)”), the module will set WUPIF. If not masked, a wake-up interrupt is pending while this flag is set. 0 No wake-up activity observed while in sleep mode 1 MSCAN detected activity on the CAN bus and requested wake-up
6 CSCIF	CAN Status Change Interrupt Flag — This flag is set when the MSCAN changes its current CAN bus status due to the actual value of the transmit error counter (TEC) and the receive error counter (REC). An additional 4-bit (RSTAT[1:0], TSTAT[1:0]) status register, which is split into separate sections for TEC/REC, informs the system on the actual CAN bus status (see Section 10.3.2.6, “MSCAN Receiver Interrupt Enable Register (CANRIER)”). If not masked, an error interrupt is pending while this flag is set. CSCIF provides a blocking interrupt. That guarantees that the receiver/transmitter status bits (RSTAT/TSTAT) are only updated when no CAN status change interrupt is pending. If the TECs/RECs change their current value after the CSCIF is asserted, which would cause an additional state change in the RSTAT/TSTAT bits, these bits keep their status until the current CSCIF interrupt is cleared again. 0 No change in CAN bus status occurred since last interrupt 1 MSCAN changed current CAN bus status
5:4 RSTAT[1:0]	Receiver Status Bits — The values of the error counters control the actual CAN bus status of the MSCAN. As soon as the status change interrupt flag (CSCIF) is set, these bits indicate the appropriate receiver related CAN bus status of the MSCAN. The coding for the bits RSTAT1, RSTAT0 is: 00 RxOK: $0 \leq \text{receive error counter} \leq 96$ 01 RxWRN: $96 < \text{receive error counter} \leq 127$ 10 RxERR: $127 < \text{receive error counter}$ 11 Bus-off ⁽¹⁾ : transmit error counter > 255
3:2 TSTAT[1:0]	Transmitter Status Bits — The values of the error counters control the actual CAN bus status of the MSCAN. As soon as the status change interrupt flag (CSCIF) is set, these bits indicate the appropriate transmitter related CAN bus status of the MSCAN. The coding for the bits TSTAT1, TSTAT0 is: 00 TxOK: $0 \leq \text{transmit error counter} \leq 96$ 01 TxWRN: $96 < \text{transmit error counter} \leq 127$ 10 TxERR: $127 < \text{transmit error counter} \leq 255$ 11 Bus-Off: transmit error counter > 255
1 OVRIF	Overrun Interrupt Flag — This flag is set when a data overrun condition occurs. If not masked, an error interrupt is pending while this flag is set. 0 No data overrun condition 1 A data overrun detected
0 RXF ⁽²⁾	Receive Buffer Full Flag — RXF is set by the MSCAN when a new message is shifted in the receiver FIFO. This flag indicates whether the shifted buffer is loaded with a correctly received message (matching identifier, matching cyclic redundancy code (CRC) and no other errors detected). After the CPU has read that message from the RxFG buffer in the receiver FIFO, the RXF flag must be cleared to release the buffer. A set RXF flag prohibits the shifting of the next FIFO entry into the foreground buffer (RxFG). If not masked, a receive interrupt is pending while this flag is set. 0 No new message available within the RxFG 1 The receiver FIFO is not empty. A new message is available in the RxFG

1. Redundant information for the most critical CAN bus status which is “bus-off”. This only occurs if the Tx error counter exceeds a number of 255 errors. Bus-off affects the receiver state. As soon as the transmitter leaves its bus-off state the receiver state skips to RxOK too. Refer also to TSTAT[1:0] coding in this register.

2. To ensure data integrity, do not read the receive buffer registers while the RXF flag is cleared. For MCUs with dual CPUs, reading the receive buffer registers while the RXF flag is cleared may result in a CPU fault condition.

Table 12-9. PWMCTL Field Descriptions

Field	Description
6 CON45	Concatenate Channels 4 and 5 0 Channels 4 and 5 are separate 8-bit PWMs. 1 Channels 4 and 5 are concatenated to create one 16-bit PWM channel. Channel 4 becomes the high-order byte and channel 5 becomes the low-order byte. Channel 5 output pin is used as the output for this 16-bit PWM (bit 5 of port PWMP). Channel 5 clock select control bit determines the clock source, channel 5 polarity bit determines the polarity, channel 5 enable bit enables the output and channel 5 center aligned enable bit determines the output mode.
5 CON23	Concatenate Channels 2 and 3 0 Channels 2 and 3 are separate 8-bit PWMs. 1 Channels 2 and 3 are concatenated to create one 16-bit PWM channel. Channel 2 becomes the high-order byte and channel 3 becomes the low-order byte. Channel 3 output pin is used as the output for this 16-bit PWM (bit 3 of port PWMP). Channel 3 clock select control bit determines the clock source, channel 3 polarity bit determines the polarity, channel 3 enable bit enables the output and channel 3 center aligned enable bit determines the output mode.
4 CON01	Concatenate Channels 0 and 1 0 Channels 0 and 1 are separate 8-bit PWMs. 1 Channels 0 and 1 are concatenated to create one 16-bit PWM channel. Channel 0 becomes the high-order byte and channel 1 becomes the low-order byte. Channel 1 output pin is used as the output for this 16-bit PWM (bit 1 of port PWMP). Channel 1 clock select control bit determines the clock source, channel 1 polarity bit determines the polarity, channel 1 enable bit enables the output and channel 1 center aligned enable bit determines the output mode.
3 PSWAI	PWM Stops in Wait Mode — Enabling this bit allows for lower power consumption in wait mode by disabling the input clock to the prescaler. 0 Allow the clock to the prescaler to continue while in wait mode. 1 Stop the input clock to the prescaler whenever the MCU is in wait mode.
2 PFRZ	PWM Counters Stop in Freeze Mode — In freeze mode, there is an option to disable the input clock to the prescaler by setting the PFRZ bit in the PWMCTL register. If this bit is set, whenever the MCU is in freeze mode the input clock to the prescaler is disabled. This feature is useful during emulation as it allows the PWM function to be suspended. In this way, the counters of the PWM can be stopped while in freeze mode so that after normal program flow is continued, the counters are re-enabled to simulate real-time operations. Because the registers remain accessible in this mode, to re-enable the prescaler clock, either disable the PFRZ bit or exit freeze mode. 0 Allow PWM to continue while in freeze mode. 1 Disable PWM input clock to the prescaler whenever the part is in freeze mode. This is useful for emulation.

The baud rate generator is activated only when the SPI is in the master mode and a serial transfer is taking place. In the other cases, the divider is disabled to decrease I_{DD} current.

$$\text{BaudRateDivisor} = (\text{SPPR} + 1) \cdot 2^{(\text{SPR} + 1)}$$

Figure 14-11. Baud Rate Divisor Equation

14.4.5 Special Features

14.4.5.1 $\overline{SS}$ Output

The $\overline{SS}$ output feature automatically drives the $\overline{SS}$ pin low during transmission to select external devices and drives it high during idle to deselect external devices. When $\overline{SS}$ output is selected, the $\overline{SS}$ output pin is connected to the $\overline{SS}$ input pin of the external device.

The $\overline{SS}$ output is available only in master mode during normal SPI operation by asserting SSOE and MODFEN bit as shown in [Table 14-3](#).

The mode fault feature is disabled while $\overline{SS}$ output is enabled.

NOTE

Care must be taken when using the $\overline{SS}$ output feature in a multimaster system because the mode fault feature is not available for detecting system errors between masters.

14.4.5.2 Bidirectional Mode (MOSI or MISO)

The bidirectional mode is selected when the SPC0 bit is set in SPI Control Register 2 (see [Table 14-9](#)). In this mode, the SPI uses only one serial data pin for the interface with external device(s). The MSTR bit decides which pin to use. The MOSI pin becomes the serial data I/O (MOMI) pin for the master mode, and the MISO pin becomes serial data I/O (SISO) pin for the slave mode. The MISO pin in master mode and MOSI pin in slave mode are not used by the SPI.

Table 14-9. Normal Mode and Bidirectional Mode

When SPE = 1	Master Mode MSTR = 1	Slave Mode MSTR = 0
Normal Mode SPC0 = 0		
Bidirectional Mode SPC0 = 1		

Module Base + 0x0009

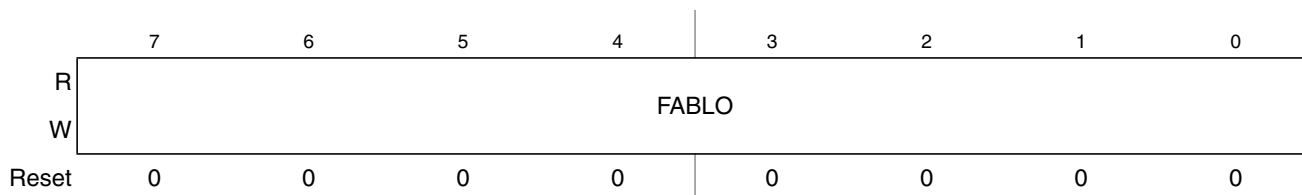


Figure 18-14. Flash Address Low Register (FADDRLO)

In normal modes, all FABHI and FABLO bits read 0 and are not writable. In special modes, the FABHI and FABLO bits are readable and writable. For sector erase, the MCU address bits [8:0] are ignored. For mass erase, any address within the Flash array is valid to start the command.

18.3.2.10 Flash Data Register (FDATA)

FDATAHI and FDATALO are the Flash data registers.

Module Base + 0x000A

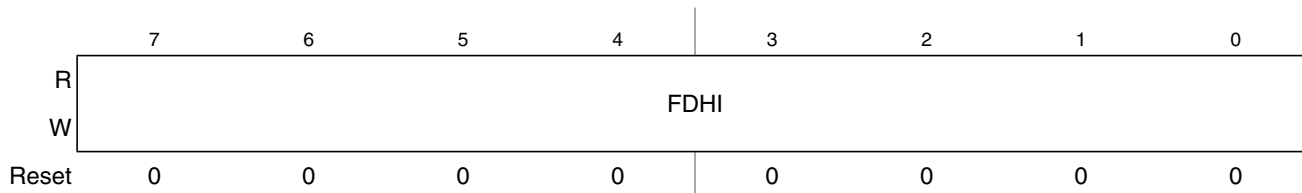


Figure 18-15. Flash Data High Register (FDATAHI)

Module Base + 0x000B

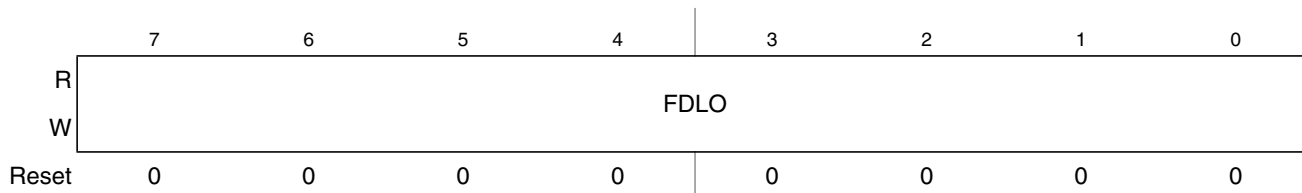


Figure 18-16. Flash Data Low Register (FDATALO)

In normal modes, all FDATAHI and FDATALO bits read 0 and are not writable. In special modes, all FDATAHI and FDATALO bits are readable and writable when writing to an address within the Flash address range.

18.3.2.11 RESERVED3

This register is reserved for factory testing and is not accessible to the user.

18.4.4 Flash Reset Sequence

On each reset, the Flash module executes a reset sequence to hold CPU activity while loading the following registers from the Flash array memory according to [Table 18-1](#):

- FPROT — Flash Protection Register (see [Section 18.3.2.5](#))
- FSEC — Flash Security Register (see [Section 18.3.2.2](#))

18.4.4.1 Reset While Flash Command Active

If a reset occurs while any Flash command is in progress, that command will be immediately aborted. The state of the word being programmed or the sector/array being erased is not guaranteed.

18.4.5 Interrupts

The Flash module can generate an interrupt when all Flash commands have completed execution or the Flash address, data, and command buffers are empty.

Table 18-17. Flash Interrupt Sources

Interrupt Source	Interrupt Flag	Local Enable	Global (CCR) Mask
Flash Address, Data, and Command Buffers are empty	CBEIF (FSTAT register)	CBEIE	I Bit
All Flash commands have completed execution	CCIF (FSTAT register)	CCIE	I Bit

NOTE

Vector addresses and their relative interrupt priority are determined at the MCU level.

18.4.5.1 Description of Interrupt Operation

[Figure 18-26](#) shows the logic used for generating interrupts.

The Flash module uses the CBEIF and CCIF flags in combination with the enable bits CBIE and CCIE to discriminate for the generation of interrupts.

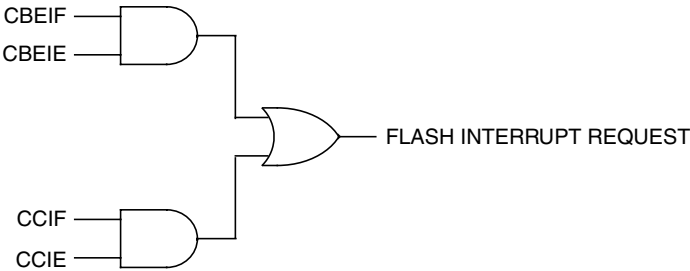


Figure 18-26. Flash Interrupt Implementation

For a detailed description of these register bits, refer to [Section 18.3.2.4](#), “Flash Configuration Register (FCNFG)” and [Section 18.3.2.6](#), “Flash Status Register (FSTAT)”.

FPHDIS is cleared. The FPROT register is loaded from Flash address 0xFF0D during the reset sequence, indicated by F in [Figure 19-10](#).

To change the Flash protection that will be loaded on reset, the upper sector of the Flash array must be unprotected, then the Flash protection byte located at Flash address 0xFF0D must be written to.

A protected Flash sector is disabled by FPHDIS and [FPLDIS](#) while the size of the protected sector is defined by FPHS[1:0] and [FPLS\[1:0\]](#) in the FPROT register.

Trying to alter any of the protected areas will result in a protect violation error and the PVIOL flag will be set in the FSTAT register (see [Section 19.3.2.6](#)). A mass erase of the whole Flash array is only possible when protection is fully disabled by setting the FPOPEN, [FPLDIS](#), and FPHDIS bits. An attempt to mass erase a Flash array while protection is enabled will set the PVIOL flag in the FSTAT register.

Table 19-9. FPROT Field Descriptions

Field	Description
7 FPOPEN	Protection Function for Program or Erase — It is possible using the FPOPEN bit to either select address ranges to be protected using FPHDIS, FPLDIS, FPHS[1:0] and FPLS[1:0] or to select the same ranges to be unprotected. When FPOPEN is set, FPxDIS enables the ranges to be protected, whereby clearing FPxDIS enables protection for the range specified by the corresponding FPxS[1:0] bits. When FPOPEN is cleared, FPxDIS defines unprotected ranges as specified by the corresponding FPxS[1:0] bits. In this case, setting FPxDIS enables protection. Thus the effective polarity of the FPxDIS bits is swapped by the FPOPEN bit as shown in Table 19-10 . This function allows the main part of the Flash array to be protected while a small range can remain unprotected for EEPROM emulation. 0 The FPHDIS and FPLDIS bits define Flash address ranges to be unprotected 1 The FPHDIS and FPLDIS bits define Flash address ranges to be protected
6 NV6	Nonvolatile Flag Bit — The NV6 bit should remain in the erased state for future enhancements.
5 FPHDIS	Flash Protection Higher Address Range Disable — The FPHDIS bit determines whether there is a protected/unprotected area in the higher space of the Flash address map. 0 Protection/unprotection enabled 1 Protection/unprotection disabled
4–3 FPHS[1:0]	Flash Protection Higher Address Size — The FPHS[1:0] bits determine the size of the protected/unprotected sector as shown in Table 19-11 . The FPHS[1:0] bits can only be written to while the FPHDIS bit is set.
2 FPLDIS	Flash Protection Lower Address Range Disable — The FPLDIS bit determines whether there is a protected/unprotected sector in the lower space of the Flash address map. 0 Protection/unprotection enabled 1 Protection/unprotection disabled
1–0 FPLS[1:0]	Flash Protection Lower Address Size — The FPLS[1:0] bits determine the size of the protected/unprotected sector as shown in Table 19-12 . The FPLS[1:0] bits can only be written to while the FPLDIS bit is set.

19.4.1.2 Command Write Sequence

The Flash command controller is used to supervise the command write sequence to execute program, erase, and erase verify algorithms.

Before starting a command write sequence, the ACCERR and PVIOL flags in the FSTAT register must be clear and the CBEIF flag should be tested to determine the state of the address, data, and command buffers. If the CBEIF flag is set, indicating the buffers are empty, a new command write sequence can be started. If the CBEIF flag is clear, indicating the buffers are not available, a new command write sequence will overwrite the contents of the address, data, and command buffers.

A command write sequence consists of three steps which must be strictly adhered to with writes to the Flash module not permitted between the steps. However, Flash register and array reads are allowed during a command write sequence. The basic command write sequence is as follows:

1. Write to a valid address in the Flash array memory.
2. Write a valid command to the FCMD register.
3. Clear the CBEIF flag in the FSTAT register by writing a 1 to CBEIF to launch the command.

The address written in step 1 will be stored in the FADDR registers and the data will be stored in the FDATA registers. When the CBEIF flag is cleared in step 3, the CCIF flag is cleared by the Flash command controller indicating that the command was successfully launched. For all command write sequences, the CBEIF flag will set after the CCIF flag is cleared indicating that the address, data, and command buffers are ready for a new command write sequence to begin. A buffered command will wait for the active operation to be completed before being launched. Once a command is launched, the completion of the command operation is indicated by the setting of the CCIF flag in the FSTAT register. The CCIF flag will set upon completion of all active and buffered commands.

Chapter 20

96 Kbyte Flash Module (S12FTS96KV1)

20.1 Introduction

The **FTS128K1FTS96K** module implements a **12896** Kbyte Flash (nonvolatile) memory. The Flash memory contains one array of **12896** Kbytes organized as **1024768** rows of **128128** bytes with an erase sector size of eight rows (**10241024** bytes). The Flash array may be read as either bytes, aligned words, or misaligned words. Read access time is one bus cycle for byte and aligned word, and two bus cycles for misaligned words.

The Flash array is ideal for program and data storage for single-supply applications allowing for field reprogramming without requiring external voltage sources for program or erase. Program and erase functions are controlled by a command driven interface. The Flash module supports both mass erase and sector erase. An erased bit reads 1 and a programmed bit reads 0. The high voltage required to program and erase is generated internally. It is not possible to read from a Flash array while it is being erased or programmed.

CAUTION

A Flash word must be in the erased state before being programmed.
Cumulative programming of bits within a Flash word is not allowed.

20.1.1 Glossary

Command Write Sequence — A three-step MCU instruction sequence to program, erase, or erase verify the Flash array memory.

20.1.2 Features

- **12896** Kbytes of Flash memory comprised of one **12896** Kbyte array divided into **12896** sectors of **10241024** bytes
- Automated program and erase algorithm
- Interrupts on Flash command completion and command buffer empty
- Fast sector erase and word program operation
- 2-stage command pipeline for faster multi-word program times
- Flexible protection scheme to prevent accidental program or erase
- Single power supply for Flash program and erase operations
- Security feature to prevent unauthorized access to the Flash array memory



A.2 ATD Characteristics

This section describes the characteristics of the analog-to-digital converter.

V_{RL} is not available as a separate pin in the 48- and 52-pin versions. In this case the internal V_{RL} pad is bonded to the V_{SSA} pin.

The ATD is specified and tested for both the 3.3V and 5V range. For ranges between 3.3V and 5V the ATD accuracy is generally the same as in the 3.3V range but is not tested in this range in production test.

A.2.1 ATD Operating Characteristics In 5V Range

The [Table A-10](#) shows conditions under which the ATD operates.

The following constraints exist to obtain full-scale, full range results: $V_{SSA} \leq V_{RL} \leq V_{IN} \leq V_{RH} \leq V_{DDA}$. This constraint exists since the sample buffer amplifier can not drive beyond the power supply levels that it ties to. If the input level goes outside of this range it will effectively be clipped.

Table A-10. ATD Operating Characteristics

Conditions are shown in Table A-4 unless otherwise noted. Supply Voltage $5V-10\% \leq V_{DDA} \leq 5V+10\%$							
Num	C	Rating	Symbol	Min	Typ	Max	Unit
1	D	Reference Potential Low High	V_{RL} V_{RH}	V_{SSA} $V_{DDA}/2$	— —	$V_{DDA}/2$ V_{DDA}	V V
2	C	Differential Reference Voltage ⁽¹⁾	$V_{RH}-V_{RL}$	4.75	5.0	5.25	V
3	D	ATD Clock Frequency	f_{ATDCLK}	0.5	—	2.0	MHz
4	D	ATD 10-Bit Conversion Period Clock Cycles ⁽²⁾ Conv, Time at 2.0MHz ATD Clock f_{ATDCLK}	N_{CONV10} T_{CONV10}	14 7	— —	28 14	Cycles μs
5	D	ATD 8-Bit Conversion Period Clock Cycles ² Conv, Time at 2.0MHz ATD Clock f_{ATDCLK}	N_{CONV10} T_{CONV10}	12 6	— —	26 13	Cycles μs
5	D	Recovery Time ($V_{DDA}=5.0$ Volts)	t_{REC}	—	—	20	μs
6	P	Reference Supply current	I_{REF}	—	—	0.375	mA

1. Full accuracy is not guaranteed when differential voltage is less than 4.75V

2. The minimum time assumes a final sample period of 2 ATD clocks cycles while the maximum time assumes a final sample period of 16 ATD clocks.

Appendix D

Derivative Differences

The Device User Guide provides information about the MC9S12C-Family and the MC9S12GC-Family. The C-Family and the GC-Family offer an extensive range of package, temperature and speed options. The members of the GC-Family are a subset of the C-family that do not feature a CAN module.

Table D-1. shows a feature overview of the C and GC family members.

Table D-1. List of MC9S12C and MC9S12GC Family members⁽¹⁾

Flash	RAM	Device	CAN	SCI	SPI	A/D	PWM	Timer
128K	4K	MC9S12C128	1	1	1	8ch	6ch	8ch
		MC9S12GC128	—	1	1	8ch	6ch	8ch
96K	4K	MC9S12C96	1	1	1	8ch	6ch	8ch
		MC9S12GC96	—	1	1	8ch	6ch	8ch
64K	4K	MC9S12C64	1	1	1	8ch	6ch	8ch
		MC9S12GC64	—	1	1	8ch	6ch	8ch
32K	2K	MC9S12C32	1	1	1	8ch	6ch	8ch
		MC9S12GC32	—	1	1	8ch	6ch	8ch
16K	1K	MC9S12GC16	—	1	1	8ch	6ch	8ch

1. All family members are available in 80QFP, 52LQFP and 48LQFP package options