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#### Details

Product Status	Active
Core Processor	HCS12
Core Size	16-Bit
Speed	25MHz
Connectivity	EBI/EMI, SCI, SPI
Peripherals	POR, PWM, WDT
Number of I/O	31
Program Memory Size	32KB (32K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	2K x 8
Voltage - Supply (Vcc/Vdd)	2.35V ~ 5.5V
Data Converters	A/D 8x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	48-LQFP
Supplier Device Package	48-LQFP (7x7)
Purchase URL	<a href="https://www.e-xfl.com/pro/item?MUrl=&amp;PartUrl=mc9s12gc32mfae">https://www.e-xfl.com/pro/item?MUrl=&amp;PartUrl=mc9s12gc32mfae</a>

## 1.2 Memory Map and Registers

### 1.2.1 Device Memory Map

Table 1-1 shows the device register map after reset. Figure 1-2 through Figure 1-6 illustrate the full device memory map.

**Table 1-1. Device Register Map Overview**

Address	Module	Size
0x0000–0x0017	Core (ports A, B, E, modes, inits, test)	24
0x0018	Reserved	1
0x0019	Voltage regulator (VREG)	1
0x001A–0x001B	Device ID register	2
0x001C–0x001F	Core (MEMSIZ, IRQ, HPRI0)	4
0x0020–0x002F	Core (DBG)	16
0x0030–0x0033	Core (PPAGE <sup>(1)</sup> )	4
0x0034–0x003F	Clock and reset generator (CRG)	12
0x0040–0x006F	Standard timer module (TIM)	48
0x0070–0x007F	Reserved	16
0x0080–0x009F	Analog-to-digital converter (ATD)	32
0x00A0–0x00C7	Reserved	40
0x00C8–0x00CF	Serial communications interface (SCI)	8
0x00D0–0x00D7	Reserved	8
0x00D8–0x00DF	Serial peripheral interface (SPI)	8
0x00E0–0x00FF	Pulse width modulator (PWM)	32
0x0100–0x010F	Flash control register	16
0x0110–0x013F	Reserved	48
0x0140–0x017F	Scalable controller area network (MSCAN) <sup>(2)</sup>	64
0x0180–0x023F	Reserved	192
0x0240–0x027F	Port integration module (PIM)	64
0x0280–0x03FF	Reserved	384

1. External memory paging is not supported on this device (Section 1.7.1, "PPAGE").

2. Not available on MC9S12GC Family devices

### 1.3.4.13 PE2 / $\overline{R/\overline{W}}$ — Port E I/O Pin [2] / Read/Write

In all modes this pin can be used as a general-purpose I/O and is an input with an active pull-up out of reset. If the read/write function is required it should be enabled by setting the RDWE bit in the PEAR register. External writes will not be possible until enabled. This pin is not available in the 48- / 52-pin package versions.

### 1.3.4.14 PE1 / $\overline{IRQ}$ — Port E Input Pin [1] / Maskable Interrupt Pin

The  $\overline{IRQ}$  input provides a means of applying asynchronous interrupt requests to the MCU. Either falling edge-sensitive triggering or level-sensitive triggering is program selectable (INTCR register).  $\overline{IRQ}$  is always enabled and configured to level-sensitive triggering out of reset. It can be disabled by clearing IRQEN bit (INTCR register). When the MCU is reset the  $\overline{IRQ}$  function is masked in the condition code register. This pin is always an input and can always be read. There is an active pull-up on this pin while in reset and immediately out of reset. The pull-up can be turned off by clearing PUPPE in the PUCR register.

### 1.3.4.15 PE0 / $\overline{XIRQ}$ — Port E input Pin [0] / Non Maskable Interrupt Pin

The  $\overline{XIRQ}$  input provides a means of requesting a non-maskable interrupt after reset initialization. During reset, the X bit in the condition code register (CCR) is set and any interrupt is masked until MCU software enables it. Because the  $\overline{XIRQ}$  input is level sensitive, it can be connected to a multiple-source wired-OR network. This pin is always an input and can always be read. There is an active pull-up on this pin while in reset and immediately out of reset. The pull-up can be turned off by clearing PUPPE in the PUCR register.

### 1.3.4.16 PAD[7:0] / AN[7:0] — Port AD I/O Pins [7:0]

PAD7–PAD0 are general purpose I/O pins and also analog inputs for the analog to digital converter. In order to use a PAD pin as a standard input, the corresponding ATDDIEN register bit must be set. These bits are cleared out of reset to configure the PAD pins for A/D operation.

When the A/D converter is active in multi-channel mode, port inputs are scanned and converted irrespective of Port AD configuration. Thus Port AD pins that are configured as digital inputs or digital outputs are also converted in the A/D conversion sequence.

### 1.3.4.17 PP[7] / KWP[7] — Port P I/O Pin [7]

PP7 is a general purpose input or output pin, shared with the keypad interrupt function. When configured as an input, it can generate interrupts causing the MCU to exit stop or wait mode. This pin is not available in the 48- / 52-pin package versions.

### 1.3.4.18 PP[6] / KWP[6]/ROMCTL — Port P I/O Pin [6]

PP6 is a general purpose input or output pin, shared with the keypad interrupt function. When configured as an input, it can generate interrupts causing the MCU to exit stop or wait mode. This pin is not available in the 48- / 52-pin package versions. During MCU expanded modes of operation, this pin is used to enable

### 2.3.2.1.6 Port T Polarity Select Register (PTTST)

Module Base + 0x0005

	7	6	5	4	3	2	1	0
R	PPST7	PPST6	PPST5	PPST4	PPST3	PPST2	PPST1	PPST0
W								
Reset	0	0	0	0	0	0	0	0

Figure 2-8. Port T Polarity Select Register (PPST)

Read: Anytime.

Write: Anytime.

Table 2-8. PPST Field Descriptions

Field	Description
7–0 PPST[7:0]	<p><b>Pull Select Port T</b> — This register selects whether a pull-down or a pull-up device is connected to the pin.</p> <p>0 A pull-up device is connected to the associated port T pin, if enabled by the associated bit in register PERT and if the port is used as input.</p> <p>1 A pull-down device is connected to the associated port T pin, if enabled by the associated bit in register PERT and if the port is used as input.</p>

### 2.3.2.1.7 Port T Module Routing Register (MODRR)

Module Base + 0x0007

	7	6	5	4	3	2	1	0
R	0	0	0	MODRR4	MODRR3	MODRR2	MODRR1	MODRR0
W								
Reset	—	—	—	0	0	0	0	0

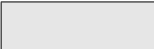
 = Unimplemented or Reserved

Figure 2-9. Port T Module Routing Register (MODRR)

Read: Anytime.

Write: Anytime.

#### NOTE

MODRR[4] must be kept clear on devices featuring a 4 channel PWM.

Table 2-9. MODRR Field Descriptions

Field	Description
4–0 MODRR[4:0]	<p><b>Module Routing Register Port T</b> — This register selects the module connected to port T.</p> <p>0 Associated pin is connected to TIM module</p> <p>1 Associated pin is connected to PWM module</p>

The commands are described as follows:

- **ACK\_ENABLE** — enables the hardware handshake protocol. The target will issue the ACK pulse when a CPU command is executed by the CPU. The ACK\_ENABLE command itself also has the ACK pulse as a response.
- **ACK\_DISABLE** — disables the ACK pulse protocol. In this case, the host needs to use the worst case delay time at the appropriate places in the protocol.

The default state of the BDM after reset is hardware handshake protocol disabled.

All the read commands will ACK (if enabled) when the data bus cycle has completed and the data is then ready for reading out by the BKGD serial pin. All the write commands will ACK (if enabled) after the data has been received by the BDM through the BKGD serial pin and when the data bus cycle is complete. See [Section 6.4.3, “BDM Hardware Commands,”](#) and [Section 6.4.4, “Standard BDM Firmware Commands,”](#) for more information on the BDM commands.

The ACK\_ENABLE sends an ACK pulse when the command has been completed. This feature could be used by the host to evaluate if the target supports the hardware handshake protocol. If an ACK pulse is issued in response to this command, the host knows that the target supports the hardware handshake protocol. If the target does not support the hardware handshake protocol the ACK pulse is not issued. In this case, the ACK\_ENABLE command is ignored by the target because it is not recognized as a valid command.

The BACKGROUND command will issue an ACK pulse when the CPU changes from normal to background mode. The ACK pulse related to this command could be aborted using the SYNC command.

The GO command will issue an ACK pulse when the CPU exits from background mode. The ACK pulse related to this command could be aborted using the SYNC command.

The GO\_UNTIL command is equivalent to a GO command with exception that the ACK pulse, in this case, is issued when the CPU enters into background mode. This command is an alternative to the GO command and should be used when the host wants to trace if a breakpoint match occurs and causes the CPU to enter active background mode. Note that the ACK is issued whenever the CPU enters BDM, which could be caused by a breakpoint match or by a BGND instruction being executed. The ACK pulse related to this command could be aborted using the SYNC command.

The TRACE1 command has the related ACK pulse issued when the CPU enters background active mode after one instruction of the application program is executed. The ACK pulse related to this command could be aborted using the SYNC command.

The TAGGO command will not issue an ACK pulse because this would interfere with the tagging function shared on the same pin.

Table 7-12. DBGCC Field Descriptions

Field	Description
15:0	<b>Comparator C Compare Bits</b> — The comparator C compare bits control whether comparator C will compare the address bus bits [15:0] to a logic 1 or logic 0. See <a href="#">Table 7-13</a> . 0 Compare corresponding address bit to a logic 0 1 Compare corresponding address bit to a logic 1 <b>Note:</b> This register will be cleared automatically when the DBG module is armed in LOOP1 mode.

Table 7-13. Comparator C Compares

PAGSEL	EXTCMP Compare	High-Byte Compare
x0	No compare	DBGCCCH[7:0] = AB[15:8]
x1	EXTCMP[5:0] = XAB[21:16]	DBGCCCH[7:0] = XAB[15:14],AB[13:8]

### 7.3.2.7 Debug Control Register 2 (DBGC2)

Module Base + 0x0028

Starting address location affected by INITRG register setting.

	7	6	5	4	3	2	1	0
R	BKABEN <sup>(1)</sup>	FULL	BDM	TAGAB	BKCEN <sup>(2)</sup>	TAGC <sup>2</sup>	RWCEN <sup>2</sup>	RWC <sup>2</sup>
W								
Reset	0	0	0	0	0	0	0	0

- When BKABEN is set (BKP mode), all bits in DBGC2 are available. When BKABEN is cleared and DBG is used in DBG mode, bits FULL and TAGAB have no meaning.
- These bits can be used in BKP mode and DBG mode (when capture mode is not set in LOOP1) to provide a third breakpoint.

Figure 7-13. Debug Control Register 2 (DBGC2)

Table 7-14. DBGC2 Field Descriptions

Field	Description
7 BKABEN	<b>Breakpoint Using Comparator A and B Enable</b> — This bit enables the breakpoint capability using comparator A and B, when set (BKP mode) the DBGGEN bit in DBGCC1 cannot be set. 0 Breakpoint module off 1 Breakpoint module on
6 FULL	<b>Full Breakpoint Mode Enable</b> — This bit controls whether the breakpoint module is in dual mode or full mode. In full mode, comparator A is used to match address and comparator B is used to match data. See <a href="#">Section 7.4.1.2, “Full Breakpoint Mode,”</a> for more details. 0 Dual address mode enabled 1 Full breakpoint mode enabled
5 BDM	<b>Background Debug Mode Enable</b> — This bit determines if the breakpoint causes the system to enter background debug mode (BDM) or initiate a software interrupt (SWI). 0 Go to software interrupt on a break request 1 Go to BDM on a break request

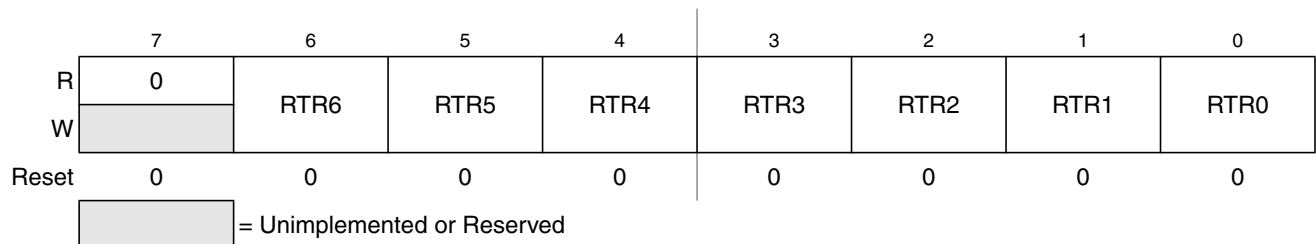
**Table 9-5. PLLCTL Field Descriptions (continued)**

Field	Description
5 AUTO	<b>Automatic Bandwidth Control Bit</b> — AUTO selects either the high bandwidth (acquisition) mode or the low bandwidth (tracking) mode depending on how close to the desired frequency the VCO is running. Write anytime except when PLLWAI=1, because PLLWAI sets the AUTO bit to 1. 0 Automatic mode control is disabled and the PLL is under software control, using ACQ bit. 1 Automatic mode control is enabled and ACQ bit has no effect.
4 ACQ	<b>Acquisition Bit</b> — Write anytime. If AUTO=1 this bit has no effect. 0 Low bandwidth filter is selected. 1 High bandwidth filter is selected.
2 PRE	<b>RTI Enable during Pseudo-Stop Bit</b> — PRE enables the RTI during pseudo-stop mode. Write anytime. 0 RTI stops running during pseudo-stop mode. 1 RTI continues running during pseudo-stop mode. <b>Note:</b> If the PRE bit is cleared the RTI dividers will go static while pseudo-stop mode is active. The RTI dividers will <u>not</u> initialize like in wait mode with RTIWAI bit set.
1 PCE	<b>COP Enable during Pseudo-Stop Bit</b> — PCE enables the COP during pseudo-stop mode. Write anytime. 0 COP stops running during pseudo-stop mode 1 COP continues running during pseudo-stop mode <b>Note:</b> If the PCE bit is cleared the COP dividers will go static while pseudo-stop mode is active. The COP dividers will <u>not</u> initialize like in wait mode with COPWAI bit set.
0 SCME	<b>Self-Clock Mode Enable Bit</b> — Normal modes: Write once —Special modes: Write anytime — SCME can not be cleared while operating in self-clock mode (SCM=1). 0 Detection of crystal clock failure causes clock monitor reset (see <a href="#">Section 9.5.1, “Clock Monitor Reset”</a> ). 1 Detection of crystal clock failure forces the MCU in self-clock mode (see <a href="#">Section 9.4.7.2, “Self-Clock Mode”</a> ).

### 9.3.2.8 CRG RTI Control Register (RTICTL)

This register selects the timeout period for the real-time interrupt.

Module Base + 0x0007



**Figure 9-11. CRG RTI Control Register (RTICTL)**

Read: anytime

Write: anytime

#### NOTE

A write to this register initializes the RTI counter.

## 9.4.2 System Clocks Generator

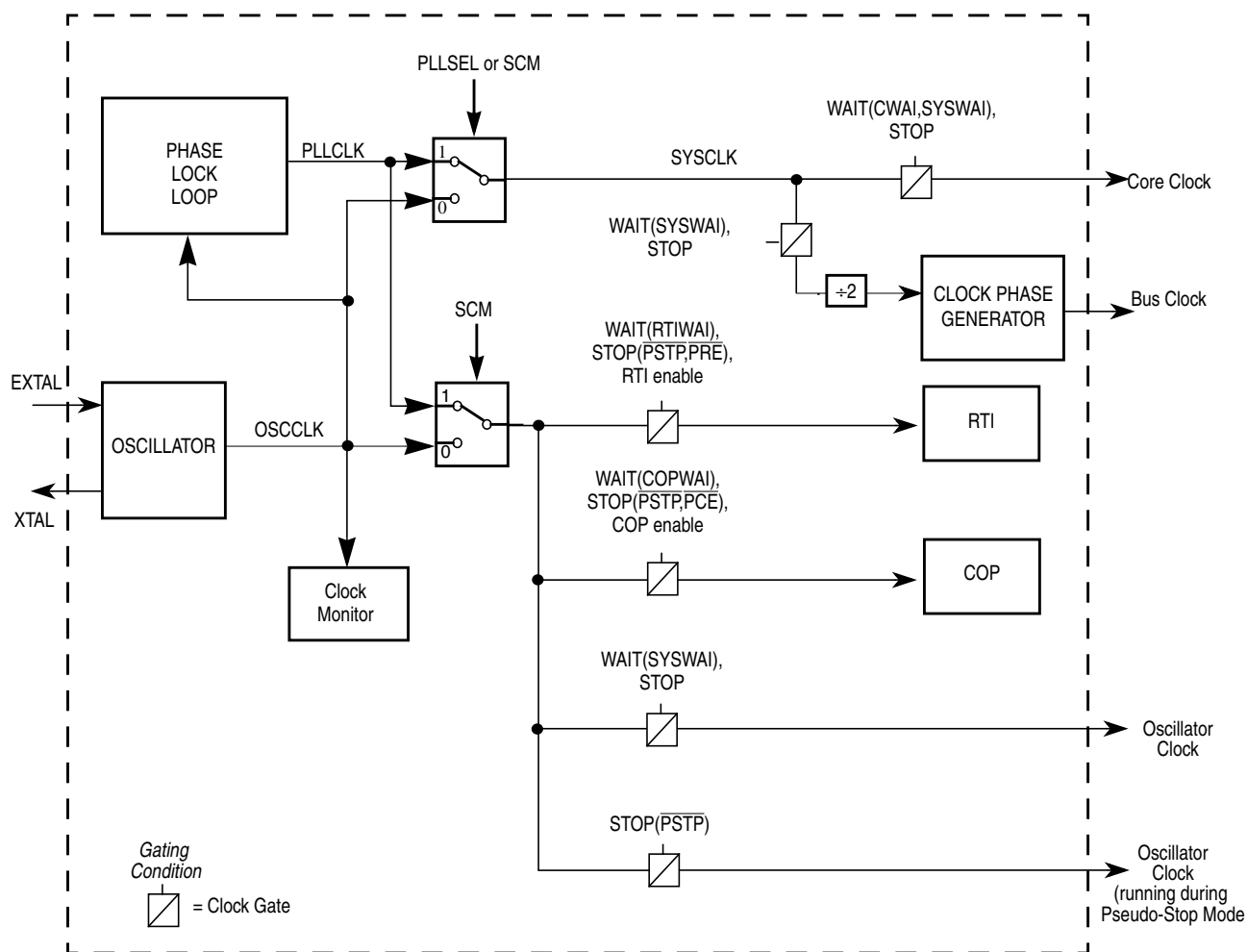


Figure 9-17. System Clocks Generator

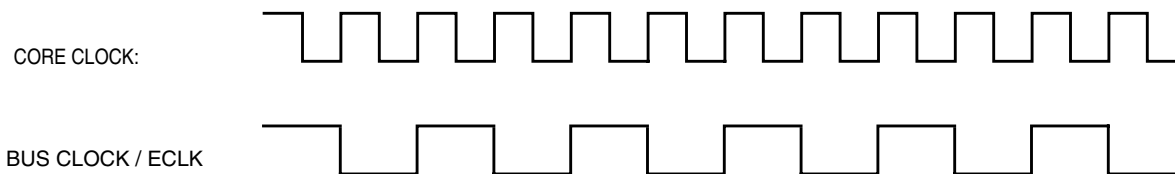
The clock generator creates the clocks used in the MCU (see Figure 9-17). The gating condition placed on top of the individual clock gates indicates the dependencies of different modes (stop, wait) and the setting of the respective configuration bits.

The peripheral modules use the bus clock. Some peripheral modules also use the oscillator clock. The memory blocks use the bus clock. If the MCU enters self-clock mode (see Section 9.4.7.2, “Self-Clock Mode”), oscillator clock source is switched to PLLCLK running at its minimum frequency  $f_{SCM}$ . The bus clock is used to generate the clock visible at the ECLK pin. The core clock signal is the clock for the CPU. The core clock is twice the bus clock as shown in Figure 9-18. But note that a CPU cycle corresponds to one bus clock.

PLL clock mode is selected with PLLSEL bit in the CLKSEL register. When selected, the PLL output clock drives SYSCLK for the main system including the CPU and peripherals. The PLL cannot be turned off by clearing the PLLON bit, if the PLL clock is selected. When PLLSEL is changed, it takes a maximum



of 4 OSCCLK plus 4 PLLCLK cycles to make the transition. During the transition, all clocks freeze and CPU activity ceases.



**Figure 9-18. Core Clock and Bus Clock Relationship**

## 9.4.3 Clock Monitor (CM)

If no OSCCLK edges are detected within a certain time, the clock monitor within the oscillator block generates a clock monitor fail event. The CRGV4 then asserts self-clock mode or generates a system reset depending on the state of SCME bit. If the clock monitor is disabled or the presence of clocks is detected no failure is indicated by the oscillator block. The clock monitor function is enabled/disabled by the CME control bit.

## 9.4.4 Clock Quality Checker

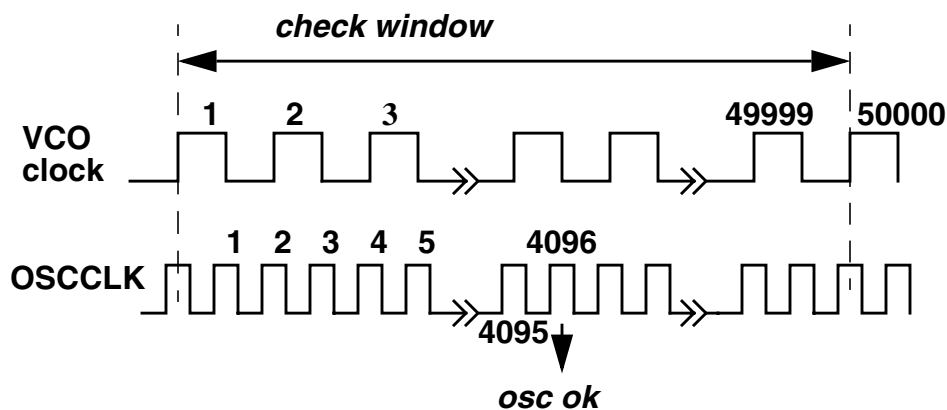
The clock monitor performs a coarse check on the incoming clock signal. The clock quality checker provides a more accurate check in addition to the clock monitor.

A clock quality check is triggered by any of the following events:

- Power-on reset (POR)
- Low voltage reset (LVR)
- Wake-up from full stop mode (exit full stop)
- Clock monitor fail indication (CM fail)

A time window of 50000 VCO clock cycles<sup>1</sup> is called *check window*.

A number greater equal than 4096 rising OSCCLK edges within a *check window* is called *osc ok*. Note that *osc ok* immediately terminates the current *check window*. See Figure 9-19 as an example.



**Figure 9-19. Check Window Example**

1. VCO clock cycles are generated by the PLL when running at minimum frequency  $f_{SCM}$ .

Register Name		Bit 7	6	5	4	3	2	1	Bit 0
0x0000 CANCTL0	R	RXFRM	RXACT	CSWAI	SYNCH	TIME	WUPE	SLPRQ	INITRQ
	W								
0x0001 CANCTL1	R	CANE	CLKSRC	LOOPB	LISTEN		WUPM	SLPAK	INITAK
	W								
0x0002 CANBTR0	R	SJW1	SJW0	BRP5	BRP4	BRP3	BRP2	BRP1	BRP0
	W								
0x0003 CANBTR1	R	SAMP	TSEG22	TSEG21	TSEG20	TSEG13	TSEG12	TSEG11	TSEG10
	W								
0x0004 CANRFLG	R	WUPIF	CSCIF	RSTAT1	RSTAT0	TSTAT1	TSTAT0	OVRIF	RXF
	W								
0x0005 CANRIER	R	WUPIE	CSCIE	RSTATE1	RSTATE0	TSTATE1	TSTATE0	OVRIE	RXFIE
	W								
0x0006 CANTFLG	R	0	0	0	0	0	TXE2	TXE1	TXE0
	W								
0x0007 CANTIER	R	0	0	0	0	0	TXEIE2	TXEIE1	TXEIE0
	W								
0x0008 CANTARQ	R	0	0	0	0	0	ABTRQ2	ABTRQ1	ABTRQ0
	W								
0x0009 CANTAACK	R	0	0	0	0	0	ABTAK2	ABTAK1	ABTAK0
	W								
0x000A CANTBSEL	R	0	0	0	0	0	TX2	TX1	TX0
	W								
0x000B CANIDAC	R	0	0	IDAM1	IDAM0	0	IDHIT2	IDHIT1	IDHIT0
	W								
0x000C–0x000D Reserved	R	0	0	0	0	0	0	0	0
	W								
0x000E CANRXERR	R	RXERR7	RXERR6	RXERR5	RXERR4	RXERR3	RXERR2	RXERR1	RXERR0
	W								
0x000F CANTXERR	R	TXERR7	TXERR6	TXERR5	TXERR4	TXERR3	TXERR2	TXERR1	TXERR0
	W								



 = Unimplemented or Reserved
  u = Unaffected

Figure 10-3. MSCAN Register Summary

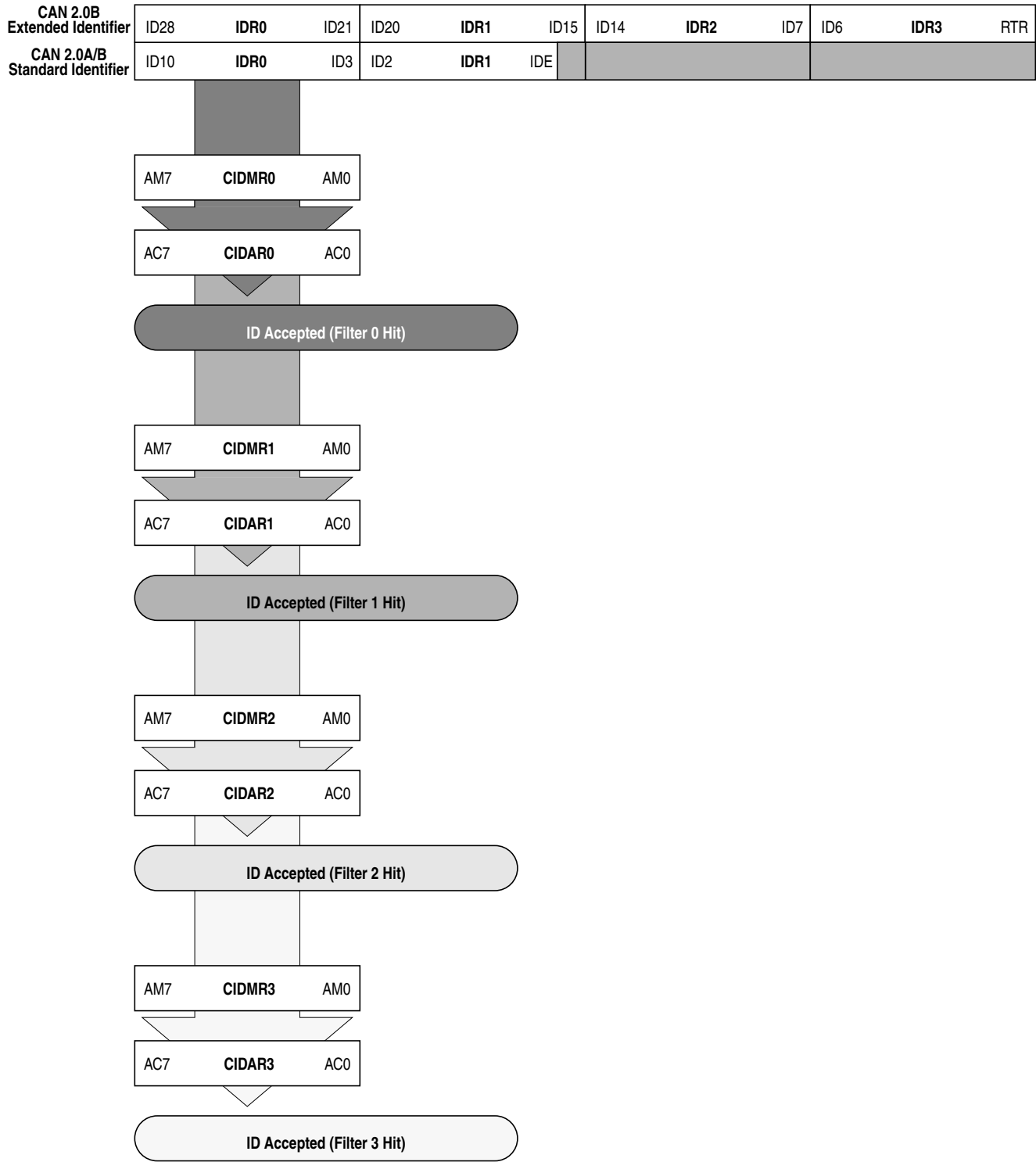


Figure 10-41. 8-bit Maskable Identifier Acceptance Filters

Table 12-3. PWMPOL Field Descriptions (continued)

Field	Description
3 PPOL3	<b>Pulse Width Channel 3 Polarity</b> 0 PWM channel 3 output is low at the beginning of the period, then goes high when the duty count is reached. 1 PWM channel 3 output is high at the beginning of the period, then goes low when the duty count is reached.
2 PPOL2	<b>Pulse Width Channel 2 Polarity</b> 0 PWM channel 2 output is low at the beginning of the period, then goes high when the duty count is reached. 1 PWM channel 2 output is high at the beginning of the period, then goes low when the duty count is reached.
1 PPOL1	<b>Pulse Width Channel 1 Polarity</b> 0 PWM channel 1 output is low at the beginning of the period, then goes high when the duty count is reached. 1 PWM channel 1 output is high at the beginning of the period, then goes low when the duty count is reached.
0 PPOL0	<b>Pulse Width Channel 0 Polarity</b> 0 PWM channel 0 output is low at the beginning of the period, then goes high when the duty count is reached. 1 PWM channel 0 output is high at the beginning of the period, then goes low when the duty count is reached.

### 12.3.2.3 PWM Clock Select Register (PWMCLK)

Each PWM channel has a choice of two clocks to use as the clock source for that channel as described below.

Module Base + 0x0002

	7	6	5	4	3	2	1	0
R	0	0	PCLK5	PCLK4	PCLK3	PCLK2	PCLK1	PCLK0
W								
Reset	0	0	0	0	0	0	0	0
	= Unimplemented or Reserved							

Figure 12-5. PWM Clock Select Register (PWMCLK)

Read: anytime

Write: anytime

#### NOTE

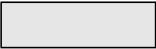
Register bits PCLK0 to PCLK5 can be written anytime. If a clock select is changed while a PWM signal is being generated, a truncated or stretched pulse can occur during the transition.

### 12.3.2.7 Reserved Register (PWMTST)

This register is reserved for factory testing of the PWM module and is not available in normal modes.

Module Base + 0x0006

	7	6	5	4	3	2	1	0
R	0	0	0	0	0	0	0	0
W								
Reset	0	0	0	0	0	0	0	0

 = Unimplemented or Reserved

**Figure 12-9. Reserved Register (PWMTST)**

Read: always read 0x0000 in normal modes

Write: unimplemented in normal modes

#### NOTE


Writing to this register when in special modes can alter the PWM functionality.

### 12.3.2.8 Reserved Register (PWMPRSC)

This register is reserved for factory testing of the PWM module and is not available in normal modes.

Module Base + 0x0007

	7	6	5	4	3	2	1	0
R	0	0	0	0	0	0	0	0
W								
Reset	0	0	0	0	0	0	0	0

 = Unimplemented or Reserved

**Figure 12-10. Reserved Register (PWMPRSC)**

Read: always read 0x0000 in normal modes

Write: unimplemented in normal modes

#### NOTE

Writing to this register when in special modes can alter the PWM functionality.

### 12.3.2.9 PWM Scale A Register (PWMSCLA)

PWMSCLA is the programmable scale value used in scaling clock A to generate clock SA. Clock SA is generated by taking clock A, dividing it by the value in the PWMSCLA register and dividing that by two.

$$\text{Clock SA} = \text{Clock A} / (2 * \text{PWMSCLA})$$

#### NOTE

When PWMSCLA = 0x0000, PWMSCLA value is considered a full scale value of 256. Clock A is thus divided by 512.

Any value written to this register will cause the scale counter to load the new scale value (PWMSCLA).

Module Base + 0x0008

	7	6	5	4	3	2	1	0
R								
W	Bit 7	6	5	4	3	2	1	Bit 0
Reset	0	0	0	0	0	0	0	0

Figure 12-11. PWM Scale A Register (PWMSCLA)

Read: anytime

Write: anytime (causes the scale counter to load the PWMSCLA value)

### 12.3.2.10 PWM Scale B Register (PWMSCLB)

PWMSCLB is the programmable scale value used in scaling clock B to generate clock SB. Clock SB is generated by taking clock B, dividing it by the value in the PWMSCLB register and dividing that by two.

$$\text{Clock SB} = \text{Clock B} / (2 * \text{PWMSCLB})$$

#### NOTE

When PWMSCLB = 0x0000, PWMSCLB value is considered a full scale value of 256. Clock B is thus divided by 512.

Any value written to this register will cause the scale counter to load the new scale value (PWMSCLB).

Module Base + 0x0009

	7	6	5	4	3	2	1	0
R								
W	Bit 7	6	5	4	3	2	1	Bit 0
Reset	0	0	0	0	0	0	0	0

Figure 12-12. PWM Scale B Register (PWMSCLB)

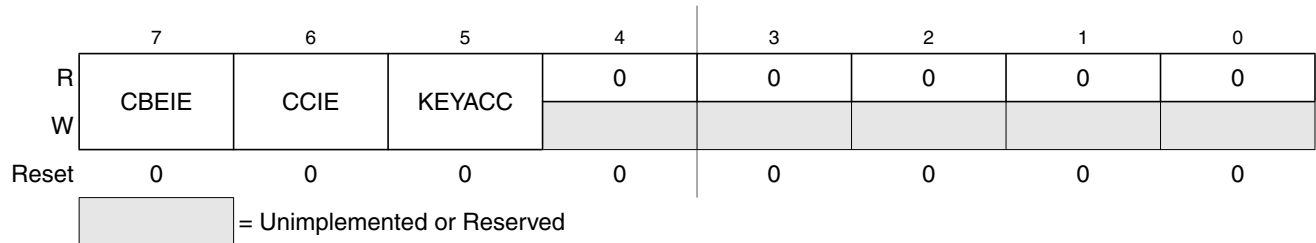
Read: anytime

Write: anytime (causes the scale counter to load the PWMSCLB value).

### 18.3.2.4 Flash Configuration Register (FCNFG)

The FCNFG register enables the Flash interrupts and gates the security backdoor key writes.

Module Base + 0x0003



**Figure 18-7. Flash Configuration Register (FCNFG)**

CBEIE, CCIE, and KEYACC are readable and writable while remaining bits read 0 and are not writable. KEYACC is only writable if the KEYEN bit in the FSEC register is set to the enabled state (see [Section 18.3.2.2](#)).

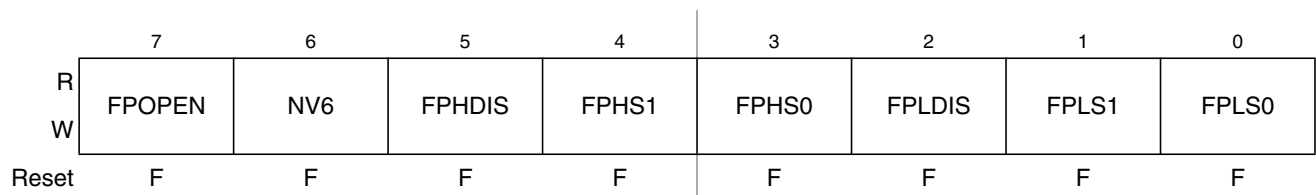
**Table 18-7. FCNFG Field Descriptions**

Field	Description
7 CBEIE	<b>Command Buffer Empty Interrupt Enable</b> — The CBEIE bit enables the interrupts in case of an empty command buffer in the Flash module. 0 Command Buffer Empty interrupts disabled 1 An interrupt will be requested whenever the CBEIF flag is set (see <a href="#">Section 18.3.2.6</a> )
6 CCIE	<b>Command Complete Interrupt Enable</b> — The CCIE bit enables the interrupts in case of all commands being completed in the Flash module. 0 Command Complete interrupts disabled 1 An interrupt will be requested whenever the CCIF flag is set (see <a href="#">Section 18.3.2.6</a> )
5 KEYACC	<b>Enable Security Key Writing.</b> 0 Flash writes are interpreted as the start of a command write sequence 1 Writes to the Flash array are interpreted as a backdoor key while reads of the Flash array return invalid data

### 18.3.2.5 Flash Protection Register (FPROT)

The FPROT register defines which Flash sectors are protected against program or erase.

Module Base + 0x0004

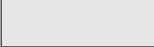


**Figure 18-8. Flash Protection Register (FPROT)**

The FPROT register is readable in normal and special modes. FPOPEN can only be written from a 1 to a 0. [FPLS\[1:0\] can be written anytime until FPLDIS is cleared.](#) FPHS[1:0] can be written anytime until

Module Base + 0x000F

	7	6	5	4	3	2	1	0
R	0	0	0	0	0	0	0	0
W								
Reset	0	0	0	0	0	0	0	0

 = Unimplemented or Reserved

**Figure 18-20. RESERVED6**

All bits read 0 and are not writable.

## 18.4 Functional Description

### 18.4.1 Flash Command Operations

Write operations are used for the program, erase, and erase verify algorithms described in this section. The program and erase algorithms are controlled by a state machine whose timebase FCLK is derived from the oscillator clock via a programmable divider. The FCMD register as well as the associated FADDR and FDATA registers operate as a buffer and a register (2-stage FIFO) so that a new command along with the necessary data and address can be stored to the buffer while the previous command is still in progress. This pipelined operation allows a time optimization when programming more than one word on a specific row, as the high voltage generation can be kept active in between two programming commands. The pipelined operation also allows a simplification of command launching. Buffer empty as well as command completion are signalled by flags in the FSTAT register with corresponding interrupts generated, if enabled.

The next sections describe:

- How to write the FCLKDIV register
- Command write sequence used to program, erase or erase verify the Flash array
- Valid Flash commands
- Errors resulting from illegal Flash operations

#### 18.4.1.1 Writing the FCLKDIV Register

Prior to issuing any Flash command after a reset, it is first necessary to write the FCLKDIV register to divide the oscillator clock down to within the 150-kHz to 200-kHz range. Since the program and erase timings are also a function of the bus clock, the FCLKDIV determination must take this information into account.

If we define:

- FCLK as the clock of the Flash timing control block
- Tbus as the period of the bus clock
- INT(x) as taking the integer part of x (e.g., INT(4.323) = 4),



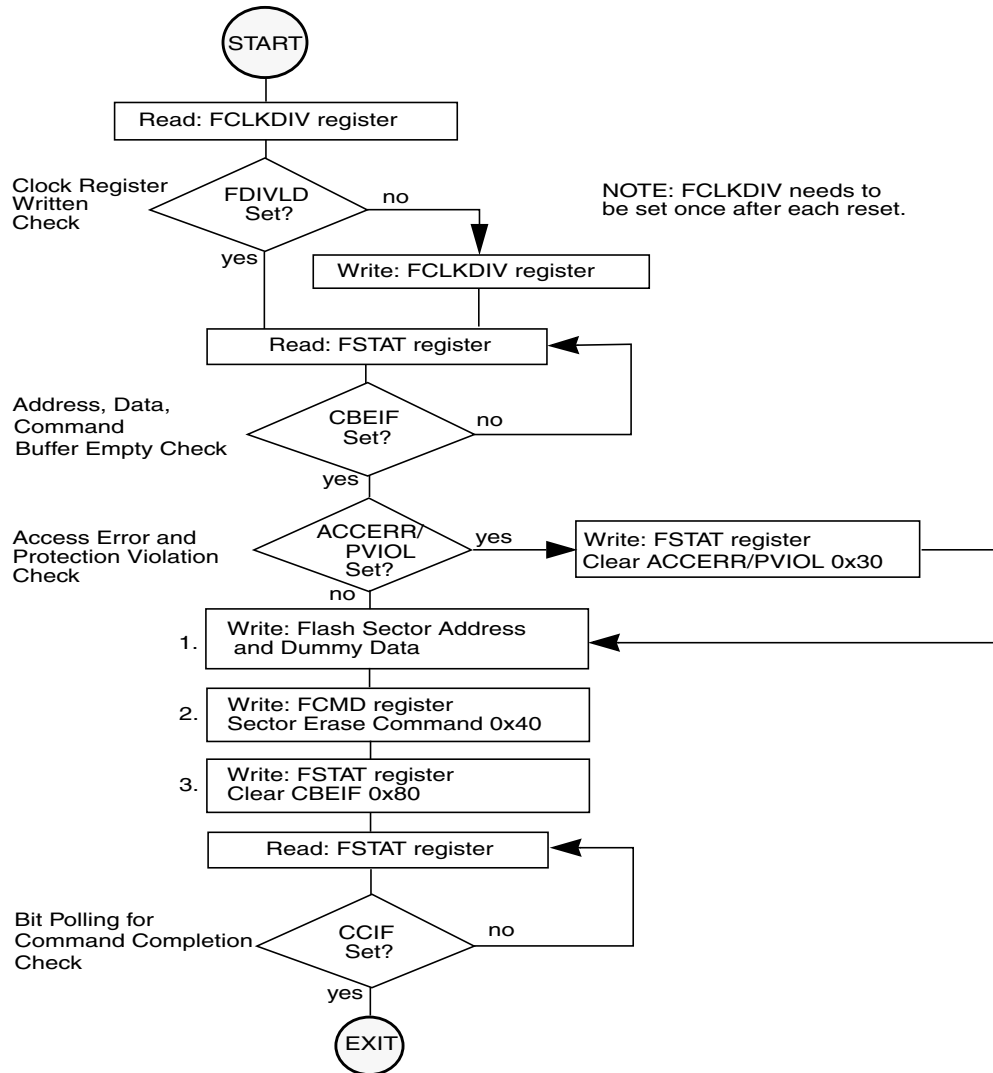


Figure 19-27. Example Sector Erase Command Flow

Table 20-10. Flash Protection Function

FPOPEN	FPHDIS	FPHS[1]	FPHS[0]	FPLDIS	FPLS[1]	FPLS[0]	Function <sup>(1)</sup>
1	1	x	x	1	x	x	No protection
1	1	x	x	0	x	x	Protect low range
1	0	x	x	1	x	x	Protect high range
1	0	x	x	0	x	x	Protect high and low ranges
0	1	x	x	1	x	x	Full Flash array protected
0	0	x	x	1	x	x	Unprotected high range
0	1	x	x	0	x	x	Unprotected low range
0	0	x	x	0	x	x	Unprotected high and low ranges

1. For range sizes refer to [Table 20-11](#) and [Table 20-12](#) or .

Table 20-11. Flash Protection Higher Address Range

FPHS[1:0]	Address Range	Range Size
00	0xF800–0xFFFF	2 Kbytes
01	0xF000–0xFFFF	4 Kbytes
10	0xE000–0xFFFF	8 Kbytes
11	0xC000–0xFFFF	16 Kbytes

Table 20-12. Flash Protection Lower Address Range

FPLS[1:0]	Address Range	Range Size
00	0x4000–0x43FF	1 Kbyte
01	0x4000–0x47FF	2 Kbytes
10	0x4000–0x4FFF	4 Kbytes
11	0x4000–0x5FFF	8 Kbytes

[Figure 20-11](#) illustrates all possible protection scenarios. Although the protection scheme is loaded from the Flash array after reset, it is allowed to change in normal modes. This protection scheme can be used by applications requiring re-programming in single chip mode while providing as much protection as possible if no re-programming is required.

## 21.4.1.4 Illegal Flash Operations

### 21.4.1.4.1 Access Error

The ACCERR flag in the FSTAT register will be set during the command write sequence if any of the following illegal Flash operations are performed causing the command write sequence to immediately abort:

1. Writing to the Flash address space before initializing the FCLKDIV register
2. Writing a misaligned word or a byte to the valid Flash address space
3. Writing to the Flash address space while CBEIF is not set
4. Writing a second word to the Flash address space before executing a program or erase command on the previously written word
5. Writing to any Flash register other than FCMD after writing a word to the Flash address space
6. Writing a second command to the FCMD register before executing the previously written command
7. Writing an invalid command to the FCMD register
8. Writing to any Flash register other than FSTAT (to clear CBEIF) after writing to the FCMD register
9. The part enters stop mode and a program or erase command is in progress. The command is aborted and any pending command is killed
10. When security is enabled, a command other than mass erase originating from a non-secure memory or from the background debug mode is written to the FCMD register
11. A 0 is written to the CBEIF bit in the FSTAT register to abort a command write sequence.

The ACCERR flag will not be set if any Flash register is read during the command write sequence. If the Flash array is read during execution of an algorithm (CCIF=0), the Flash module will return invalid data and the ACCERR flag will not be set. If an ACCERR flag is set in the FSTAT register, the Flash command controller is locked. It is not possible to launch another command until the ACCERR flag is cleared.

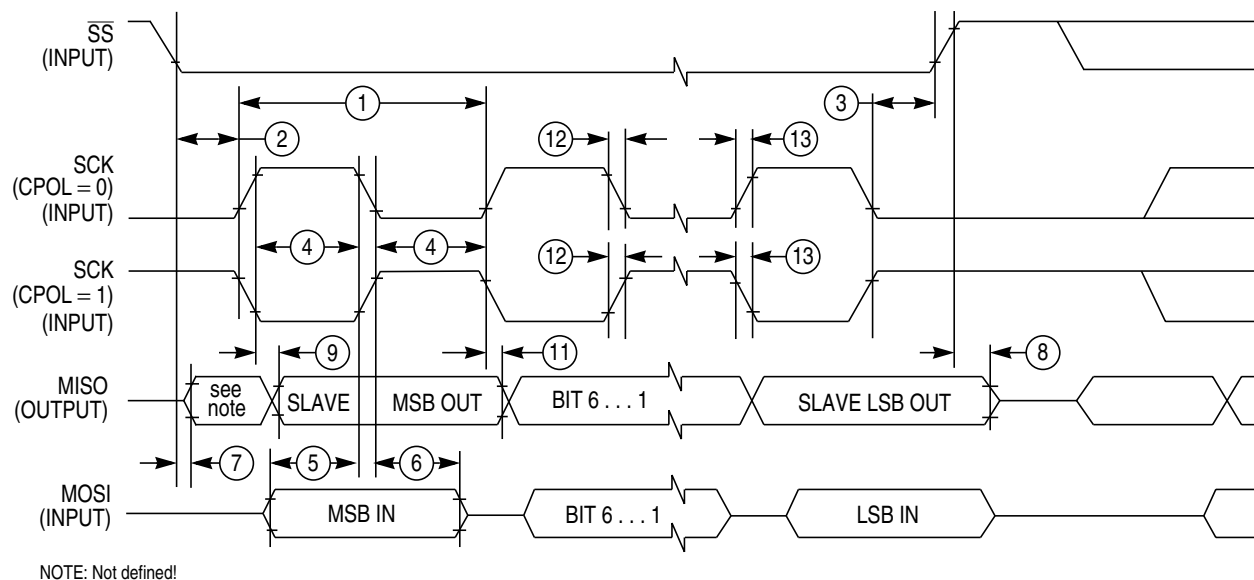
### 21.4.1.4.2 Protection Violation

The PVIOL flag in the FSTAT register will be set during the command write sequence after the word write to the Flash address space if any of the following illegal Flash operations are performed, causing the command write sequence to immediately abort:

1. Writing a Flash address to program in a protected area of the Flash array (see [Section 21.3.2.5](#)).
2. Writing a Flash address to erase in a protected area of the Flash array.
3. Writing the mass erase command to the FCMD register while any protection is enabled.

If the PVIOL flag is set, the Flash command controller is locked. It is not possible to launch another command until the PVIOL flag is cleared.

In Figure A-9 the timing diagram for slave mode with transmission format CPHA=1 is depicted.



**Figure A-9. SPI Slave Timing (CPHA=1)**

In Table A-22 the timing characteristics for slave mode are listed.

**Table A-22. SPI Slave Mode Timing Characteristics**

Num	C	Characteristic	Symbol	Min	Typ	Max	Unit
1	D	SCK Frequency	$f_{sck}$	DC	—	1/4	$f_{bus}$
1	P	SCK Period	$t_{sck}$	4	—	$\infty$	$t_{bus}$
2	D	Enable Lead Time	$t_{lead}$	4	—	—	$t_{bus}$
3	D	Enable Lag Time	$t_{lag}$	4	—	—	$t_{bus}$
4	D	Clock (SCK) High or Low Time	$t_{wsck}$	4	—	—	$t_{bus}$
5	D	Data Setup Time (Inputs)	$t_{su}$	8	—	—	ns
6	D	Data Hold Time (Inputs)	$t_{hi}$	8	—	—	ns
7	D	Slave Access Time (time to data active)	$t_a$	—	—	20	ns
8	D	Slave MISO Disable Time	$t_{dis}$	—	—	22	ns
9	D	Data Valid after SCK Edge	$t_{vsck}$	—	—	$30 + t_{bus}^{(1)}$	ns
10	D	Data Valid after SS fall	$t_{vss}$	—	—	$30 + t_{bus}^{(1)}$	ns
11	D	Data Hold Time (Outputs)	$t_{ho}$	20	—	—	ns
12	D	Rise and Fall Time Inputs	$t_{rfi}$	—	—	8	ns
13	D	Rise and Fall Time Outputs	$t_{rfo}$	—	—	8	ns

1.  $t_{bus}$  added due to internal synchronization delay

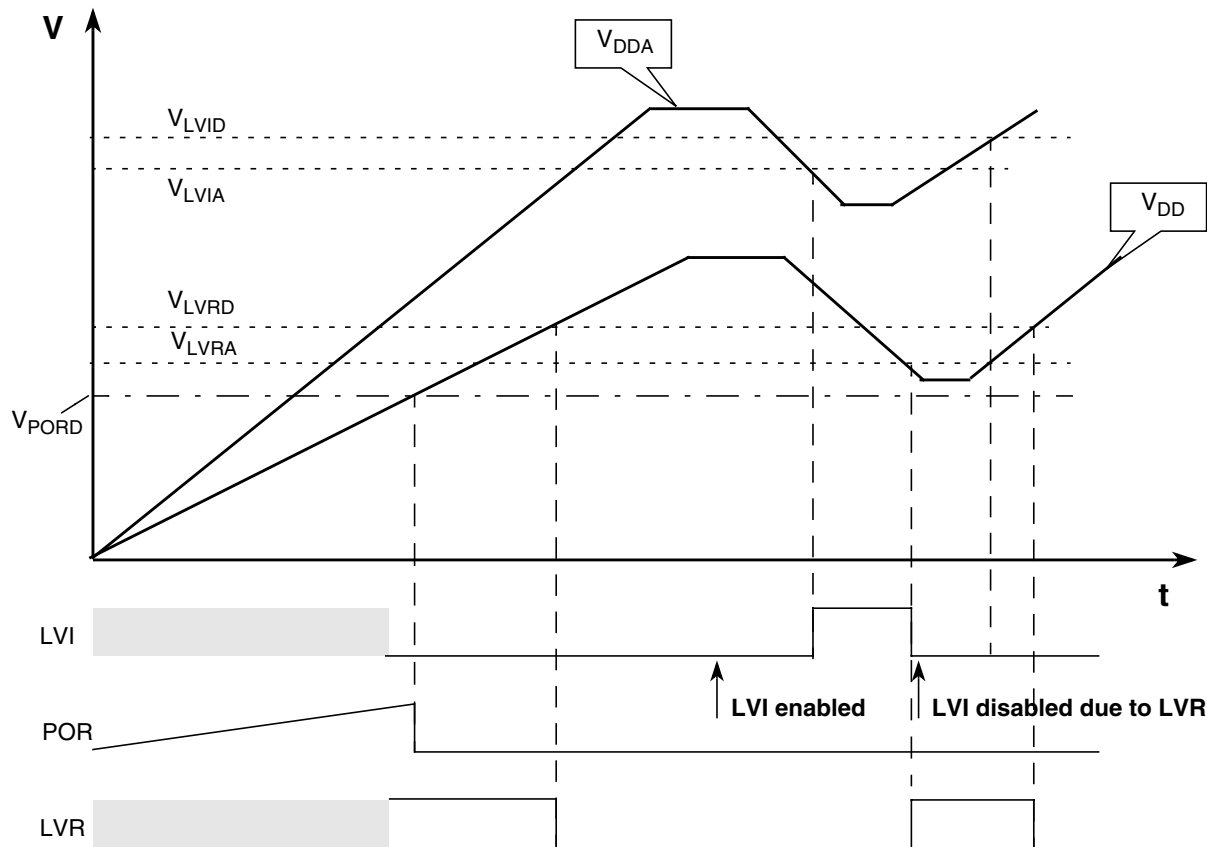


Figure A-10. Voltage Regulator — Chip Power-up and Voltage Drops (not scaled)

### A.7.3 Output Loads

#### A.7.3.1 Resistive Loads

The on-chip voltage regulator is intended to supply the internal logic and oscillator circuits allows no external DC loads.

#### A.7.3.2 Capacitive Loads

The capacitive loads are specified in [Table A-24](#). Ceramic capacitors with X7R dielectricum are required.

Table A-24. Voltage Regulator — Capacitive Loads

Num	Characteristic	Symbol	Min	Typical	Max	Unit
1	$V_{DD}$ external capacitive load	$C_{DDext}$	400	440	12000	nF
2	$V_{DDPLL}$ external capacitive load	$C_{DDPLLext}$	90	220	5000	nF