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Details

Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	20MHz
Connectivity	-
Peripherals	Brown-out Detect/Reset, POR, WDT
Number of I/O	13
Program Memory Size	3.5KB (2K x 14)
Program Memory Type	OTP
EEPROM Size	128 x 8
RAM Size	128 x 8
Voltage - Supply (Vcc/Vdd)	3V ~ 5.5V
Data Converters	-
Oscillator Type	External
Operating Temperature	0°C ~ 70°C (TA)
Mounting Type	Surface Mount
Package / Case	20-SSOP (0.209", 5.30mm Width)
Supplier Device Package	20-SSOP
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic16ce625t-20-ss

3.1 Clocking Scheme/Instruction Cycle

The clock input (OSC1/CLKIN pin) is internally divided by four to generate four non-overlapping quadrature clocks namely Q1, Q2, Q3 and Q4. Internally, the program counter (PC) is incremented every Q1, the instruction is fetched from the program memory and latched into the instruction register in Q4. The instruction is decoded and executed during the following Q1 through Q4. The clocks and instruction execution flow is shown in Figure 3-2.

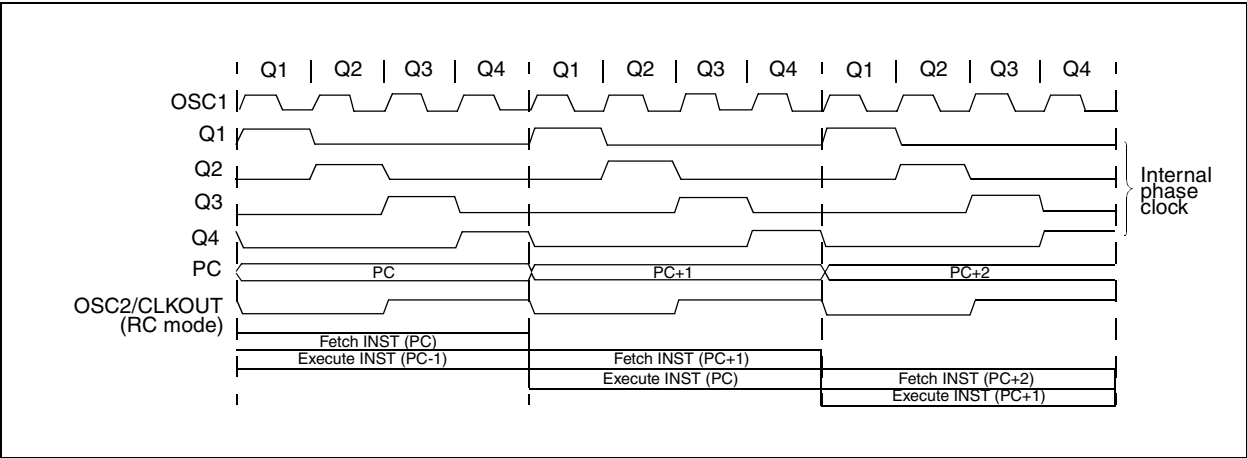
3.2 Instruction Flow/Pipelining

An “Instruction Cycle” consists of four Q cycles (Q1, Q2, Q3 and Q4). The instruction fetch and execute are pipelined such that fetch takes one instruction cycle, while decode and execute takes another instruction cycle. However, due to the pipelining, each instruction effectively executes in one cycle. If an instruction causes the program counter to change (i.e., GOTO) then two cycles are required to complete the instruction (Example 3-1).

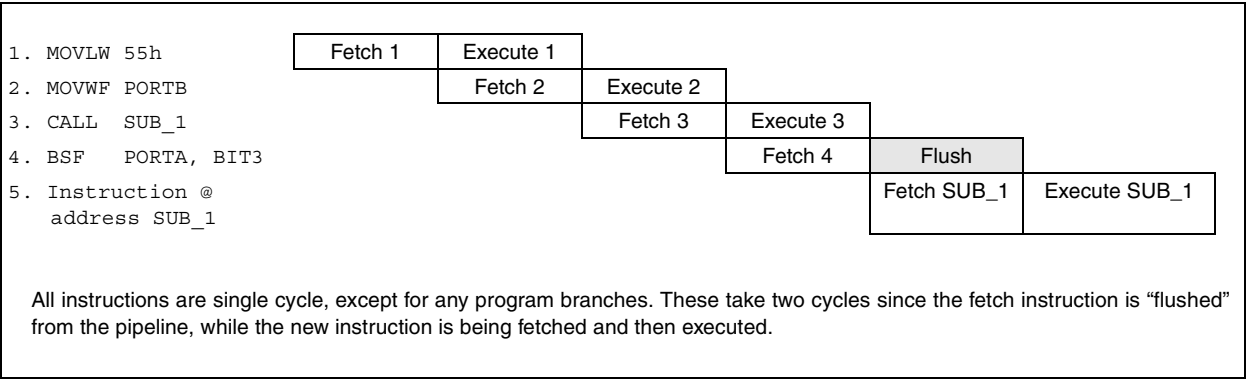
A fetch cycle begins with the program counter (PC) incrementing in Q1.

In the execution cycle, the fetched instruction is latched into the “Instruction Register (IR)” in cycle Q1. This instruction is then decoded and executed during the Q2, Q3, and Q4 cycles. Data memory is read during Q2 (operand read) and written during Q4 (destination write).

FIGURE 3-2: CLOCK/INSTRUCTION CYCLE



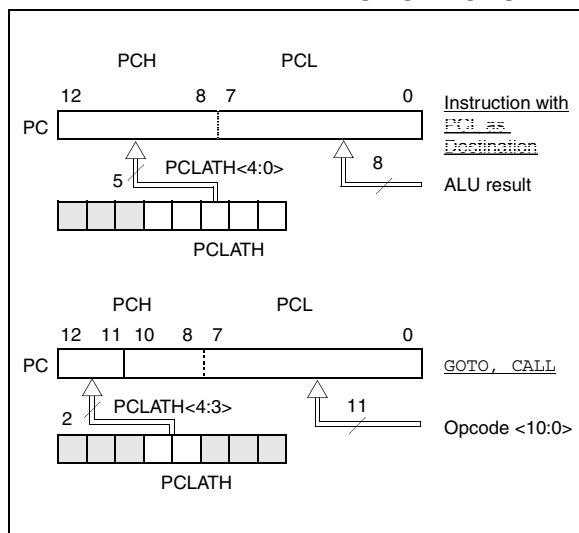
EXAMPLE 3-1: INSTRUCTION PIPELINE FLOW



4.3 PCL and PCLATH

The program counter (PC) is 13 bits wide. The low byte comes from the PCL register, which is a readable and writable register. The high byte (PC<12:8>) is not directly readable or writable and comes from PCLATH. On any reset, the PC is cleared. Figure 4-6 shows the two situations for the loading of the PC. The upper example in the figure shows how the PC is loaded on a write to PCL (PCLATH<4:0> → PCH). The lower example in the figure shows how the PC is loaded during a CALL or GOTO instruction (PCLATH<4:3> → PCH).

FIGURE 4-6: LOADING OF PC IN DIFFERENT SITUATIONS



4.3.1 COMPUTED GOTO

A computed GOTO is accomplished by adding an offset to the program counter (ADDWF PCL). When doing a table read using a computed GOTO method, care should be exercised if the table location crosses a PCL memory boundary (each 256 byte block). Refer to the application note, "Implementing a Table Read" (AN556).

4.3.2 STACK

The PIC16CE62X family has an 8 level deep x 13-bit wide hardware stack (Figure 4-2 and Figure 4-3). The stack space is not part of either program or data space and the stack pointer is not readable or writable. The PC is PUSHed onto the stack when a CALL instruction is executed or an interrupt causes a branch. The stack is POPed in the event of a RETURN, RETLW or a RETFIE instruction execution. PCLATH is not affected by a PUSH or POP operation.

The stack operates as a circular buffer. This means that after the stack has been PUSHed eight times, the ninth push overwrites the value that was stored from the first push. The tenth push overwrites the second push (and so on).

Note 1: There are no STATUS bits to indicate stack overflow or stack underflow conditions.

Note 2: There are no instruction/mnemonics called PUSH or POP. These are actions that occur from the execution of the CALL, RETURN, RETLW and RETFIE instructions or the vectoring to an interrupt address.

PIC16CE62X

NOTES:

FIGURE 5-3: BLOCK DIAGRAM OF RA3 PIN

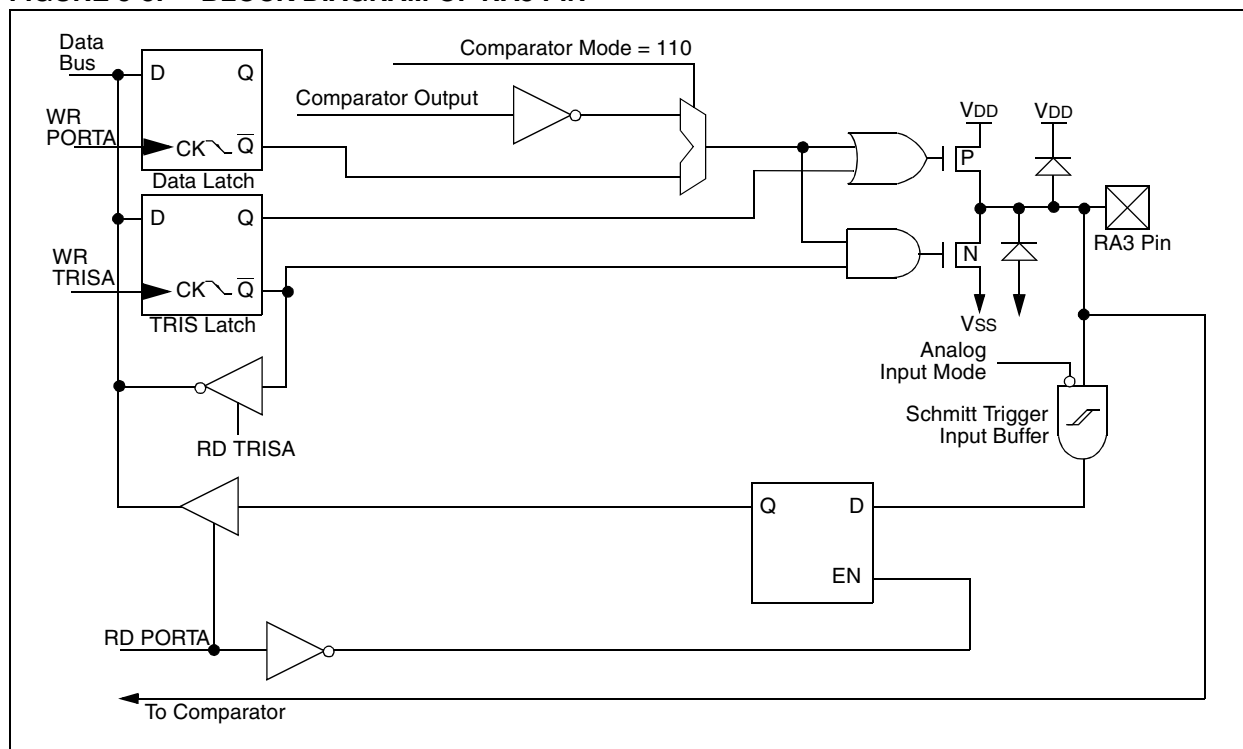


FIGURE 5-4: BLOCK DIAGRAM OF RA4 PIN

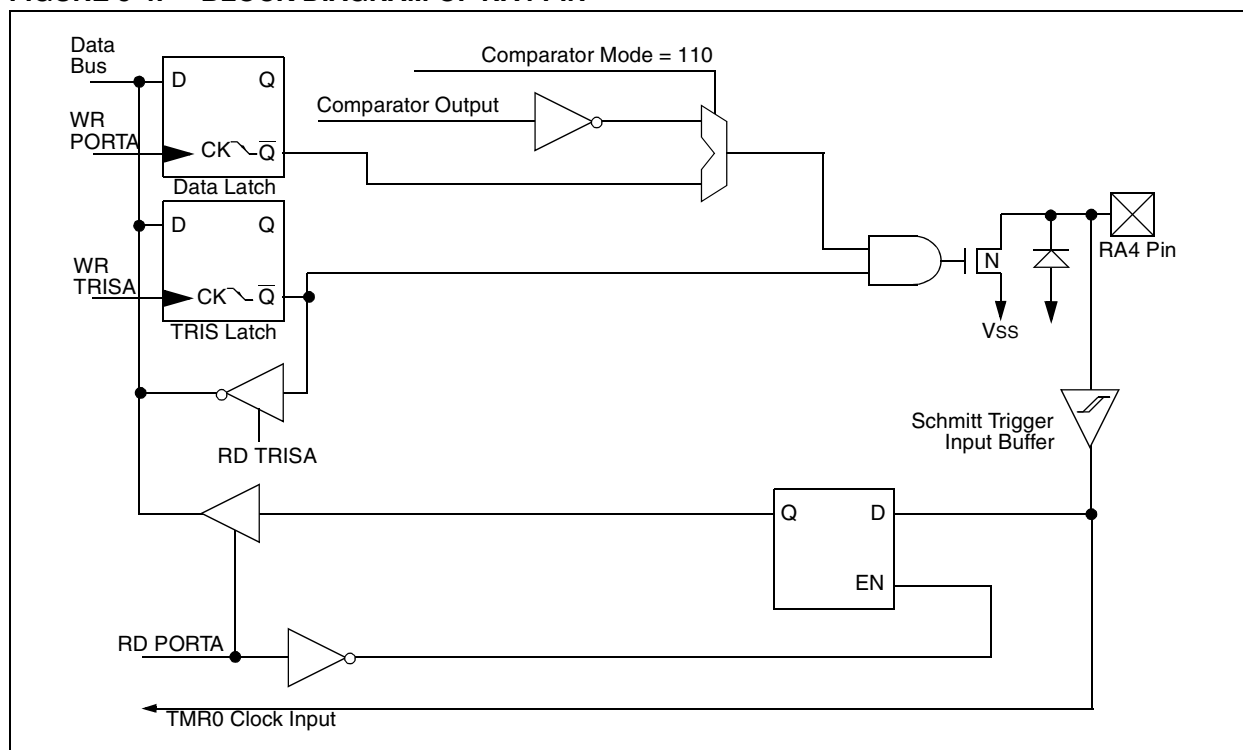


FIGURE 6-1: DATA TRANSFER SEQUENCE ON THE SERIAL BUS

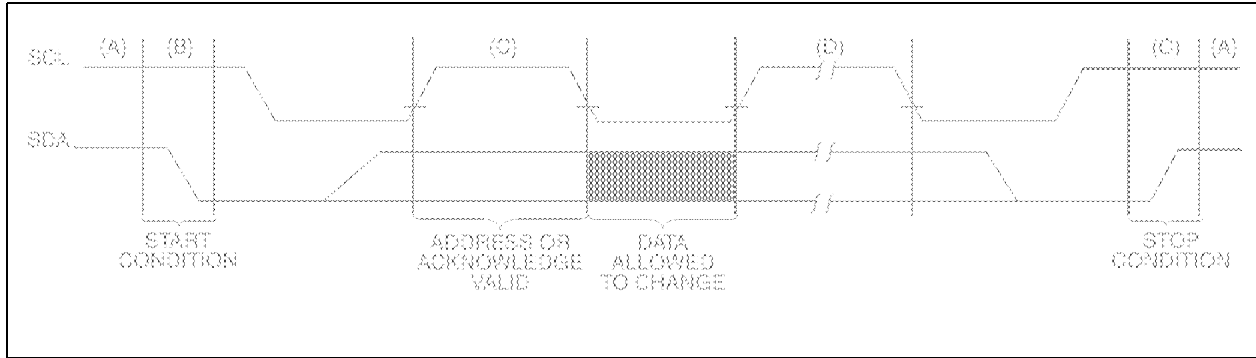
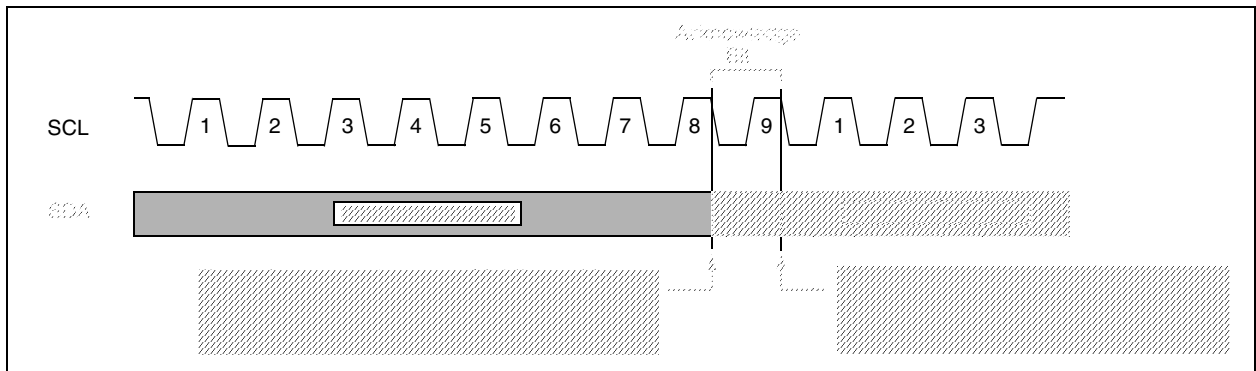


FIGURE 6-2: ACKNOWLEDGE TIMING

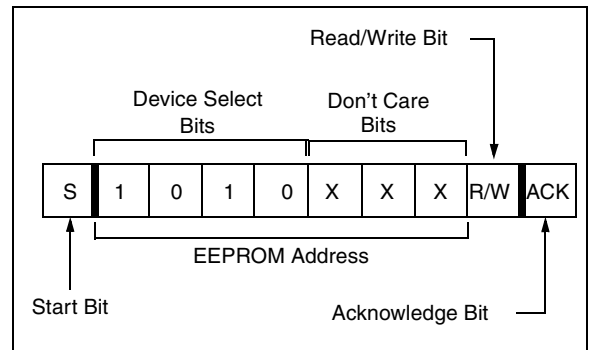


6.2 Device Addressing

After generating a START condition, the processor transmits a control byte consisting of a EEPROM address and a Read/Write bit that indicates what type of operation is to be performed. The EEPROM address consists of a 4-bit device code (1010) followed by three don't care bits.

The last bit of the control byte determines the operation to be performed. When set to a one, a read operation is selected, and when set to a zero, a write operation is selected. (Figure 6-3). The bus is monitored for its corresponding EEPROM address all the time. It generates an acknowledge bit if the EEPROM address was true and it is not in a programming mode.

FIGURE 6-3: CONTROL BYTE FORMAT



6.3 Write Operations

6.3.1 BYTE WRITE

Following the start signal from the processor, the device code (4 bits), the don't care bits (3 bits), and the R/W bit, which is a logic low, is placed onto the bus by the processor. This indicates to the EEPROM that a byte with a word address will follow after it has generated an acknowledge bit during the ninth clock cycle. Therefore, the next byte transmitted by the processor is the word address and will be written into the address pointer of the EEPROM. After receiving another acknowledge signal from the EEPROM, the processor will transmit the data word to be written into the addressed memory location. The EEPROM acknowledges again and the processor generates a stop condition. This initiates the internal write cycle, and during this time, the EEPROM will not generate acknowledge signals (Figure 6-5).

6.3.2 PAGE WRITE

The write control byte, word address and the first data byte are transmitted to the EEPROM in the same way as in a byte write. But instead of generating a stop condition, the processor transmits up to eight data bytes to the EEPROM, which are temporarily stored in the on-chip page buffer and will be written into the memory after the processor has transmitted a stop condition. After the receipt of each word, the three lower order address pointer bits are internally incremented by one. The higher order five bits of the word address remains constant. If the processor should transmit more than eight words prior to generating the stop condition, the address counter will roll over and the previously received data will be overwritten. As with the byte write operation, once the stop condition is received, an internal write cycle will begin (Figure 6-6).

6.4 Acknowledge Polling

Since the EEPROM will not acknowledge during a write cycle, this can be used to determine when the cycle is complete (this feature can be used to maximize bus throughput). Once the stop condition for a write command has been issued from the processor, the EEPROM initiates the internally timed write cycle. ACK polling can be initiated immediately. This involves the processor sending a start condition followed by the control byte for a write command (R/W = 0). If the device is still busy with the write cycle, then no ACK will be returned. If no ACK is returned, then the start bit and control byte must be re-sent. If the cycle is complete, then the device will return the ACK and the processor can then proceed with the next read or write command. See Figure 6-4 for flow diagram.

FIGURE 6-4: ACKNOWLEDGE POLLING FLOW

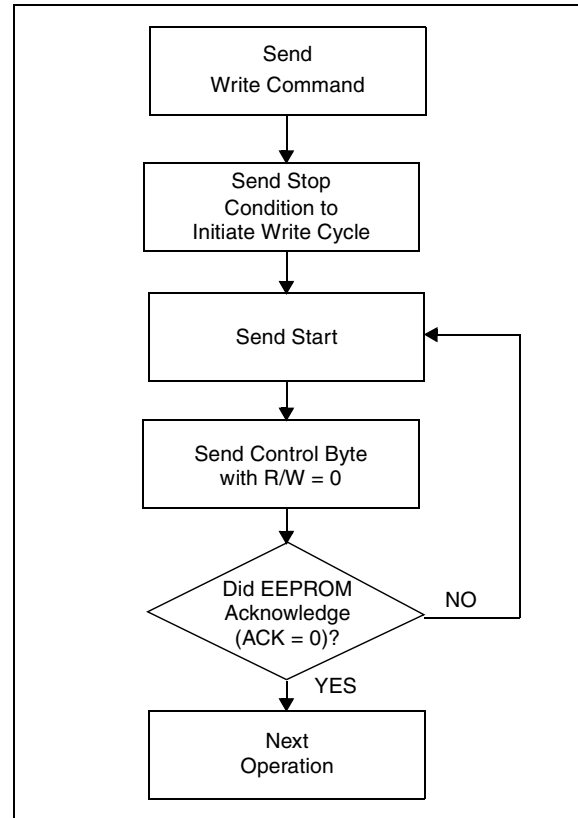
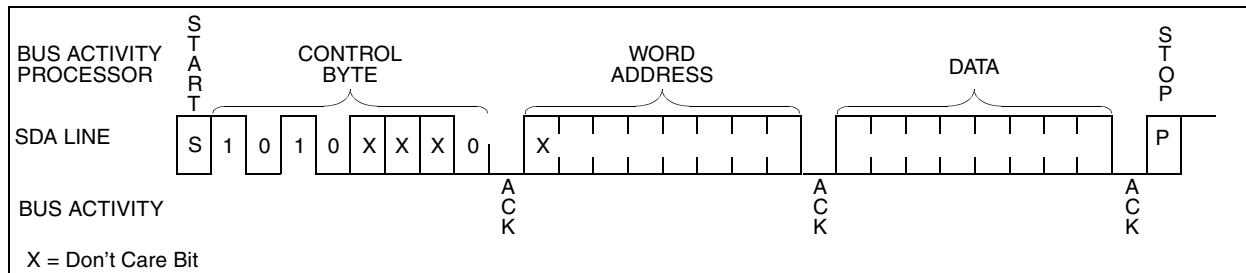


FIGURE 6-5: BYTE WRITE



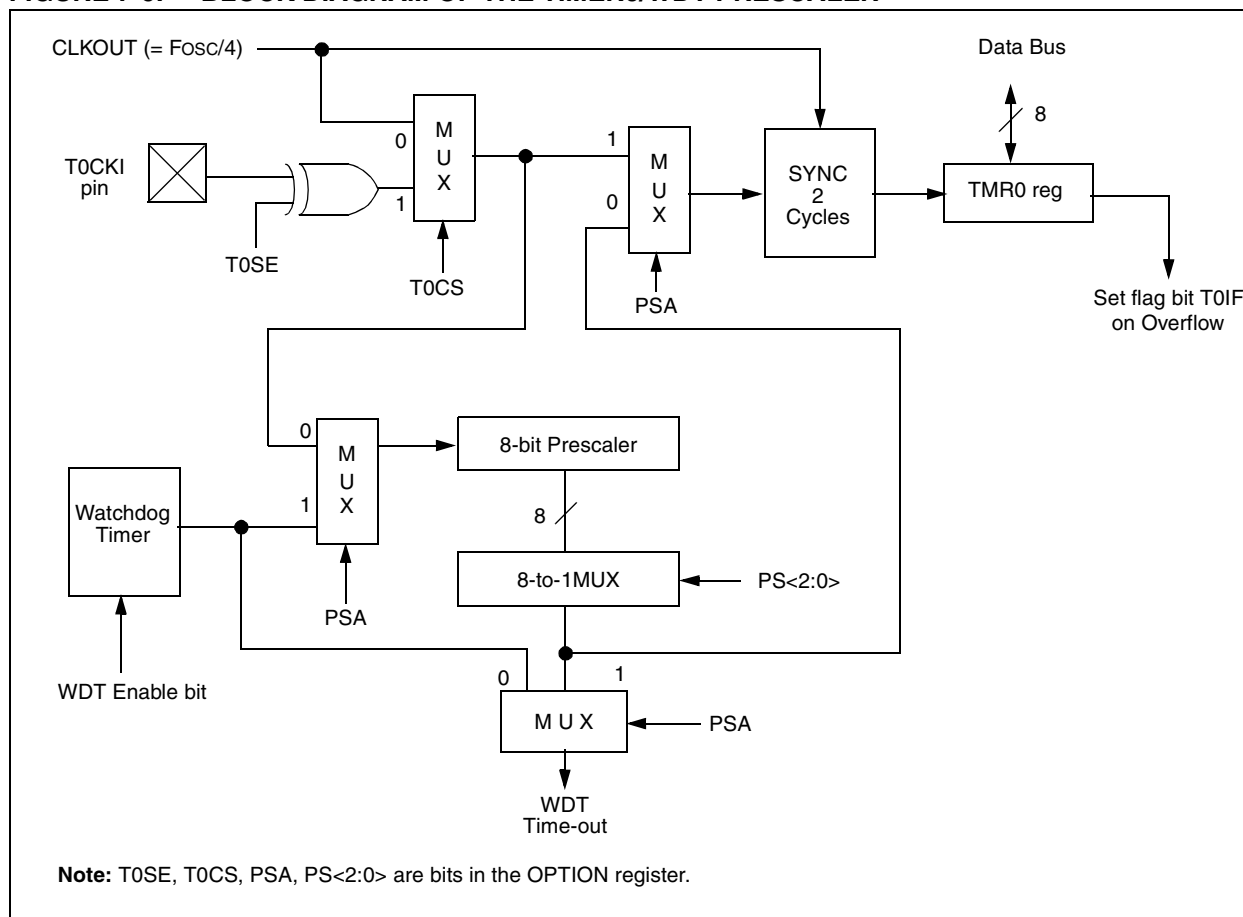
7.3 Prescaler

An 8-bit counter is available as a prescaler for the Timer0 module, or as a postscaler for the Watchdog Timer, respectively (Figure 7-6). For simplicity, this counter is being referred to as “prescaler” throughout this data sheet. Note that there is only one prescaler available which is mutually exclusive between the Timer0 module and the Watchdog Timer. Thus, a prescaler assignment for the Timer0 module means that there is no prescaler for the Watchdog Timer and vice-versa.

The PSA and PS<2:0> bits (OPTION<3:0>) determine the prescaler assignment and prescale ratio.

When assigned to the Timer0 module, all instructions writing to the TMR0 register (i.e., CLRF 1, MOVWF 1, BSF 1,x,...etc.) will clear the prescaler. When assigned to WDT, a CLRWDT instruction will clear the prescaler along with the Watchdog Timer. The prescaler is not readable or writable.

FIGURE 7-6: BLOCK DIAGRAM OF THE TIMER0/WDT PRESCALER



8.4 Comparator Response Time

Response time is the minimum time, after selecting a new reference voltage or input source, before the comparator output has a valid level. If the internal reference is changed, the maximum delay of the internal voltage reference must be considered when using the comparator outputs, otherwise the maximum delay of the comparators should be used (Table 13-1).

8.5 Comparator Outputs

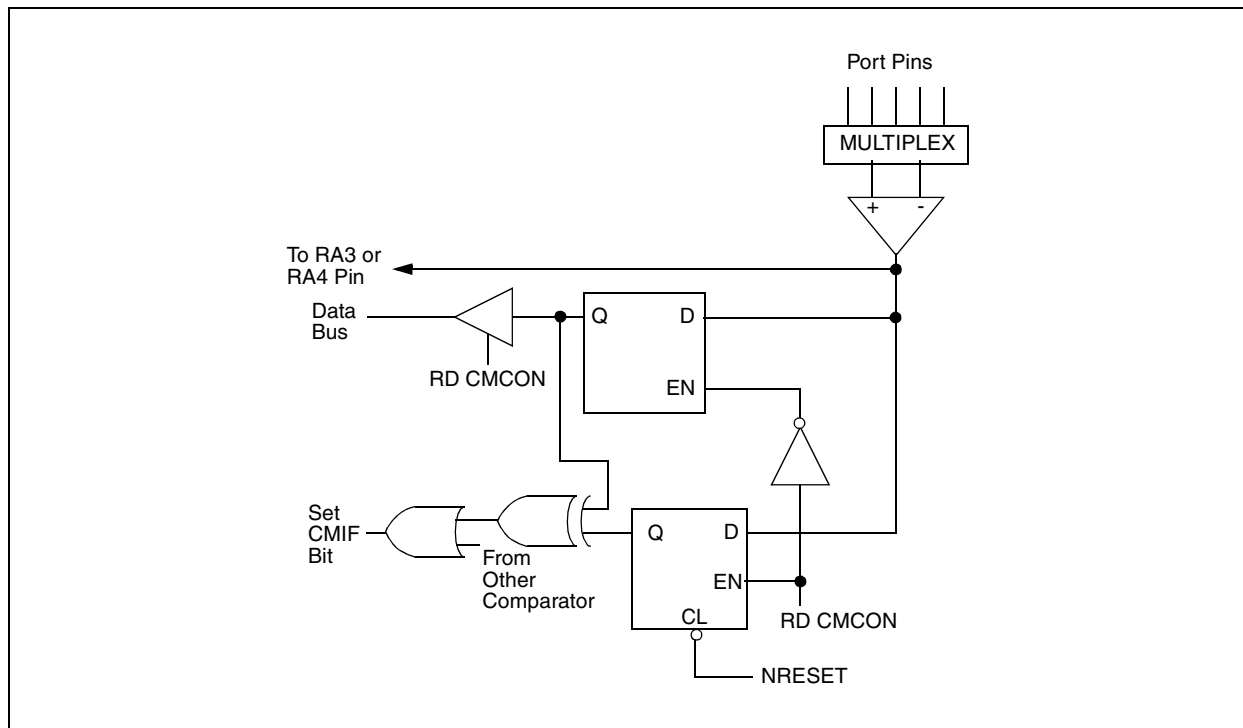
The comparator outputs are read through the CMCON register. These bits are read only. The comparator outputs may also be directly output to the RA3 and RA4 I/O pins. When the CM<2:0> = 110, multiplexors in the output path of the RA3 and RA4 pins will switch and the output of each pin will be the unsynchronized output of the comparator. The uncertainty of each of the comparators is related to the input offset voltage and the response time given in the specifications. Figure 8-3 shows the comparator output block diagram.

The TRISA bits will still function as an output enable/disable for the RA3 and RA4 pins while in this mode.

Note 1: When reading the PORT register, all pins configured as analog inputs will read as a '0'. Pins configured as digital inputs will convert an analog input according to the Schmitt Trigger input specification.

2: Analog levels on any pin that is defined as a digital input may cause the input buffer to consume more current than is specified.

FIGURE 8-3: COMPARATOR OUTPUT BLOCK DIAGRAM



8.6 Comparator Interrupts

The comparator interrupt flag is set whenever there is a change in the output value of either comparator. Software will need to maintain information about the status of the output bits, as read from CMCON<7:6>, to determine the actual change that has occurred. The CMIF bit, PIR1<6>, is the comparator interrupt flag. The CMIF bit must be reset by clearing '0'. Since it is also possible to write a '1' to this register, a simulated interrupt may be initiated.

The CMIE bit (PIE1<6>) and the PEIE bit (INTCON<6>) must be set to enable the interrupt. In addition, the GIE bit must also be set. If any of these bits are clear, the interrupt is not enabled, though the CMIF bit will still be set if an interrupt condition occurs.

Note: If a change in the CMCON register (C1OUT or C2OUT) should occur when a read operation is being executed (start of the Q2 cycle), then the CMIF (PIR1<6>) interrupt flag may not get set.

The user, in the interrupt service routine, can clear the interrupt in the following manner:

- Any read or write of CMCON. This will end the mismatch condition.
- Clear flag bit CMIF.

A mismatch condition will continue to set flag bit CMIF. Reading CMCON will end the mismatch condition, and allow flag bit CMIF to be cleared.

8.7 Comparator Operation During SLEEP

When a comparator is active and the device is placed in SLEEP mode, the comparator remains active and the interrupt is functional if enabled. This interrupt will

wake-up the device from SLEEP mode when enabled. While the comparator is powered-up, higher sleep currents than shown in the power down current specification will occur. Each comparator that is operational will consume additional current as shown in the comparator specifications. To minimize power consumption while in SLEEP mode, turn off the comparators, CM<2:0> = 111, before entering sleep. If the device wakes-up from sleep, the contents of the CMCON register are not affected.

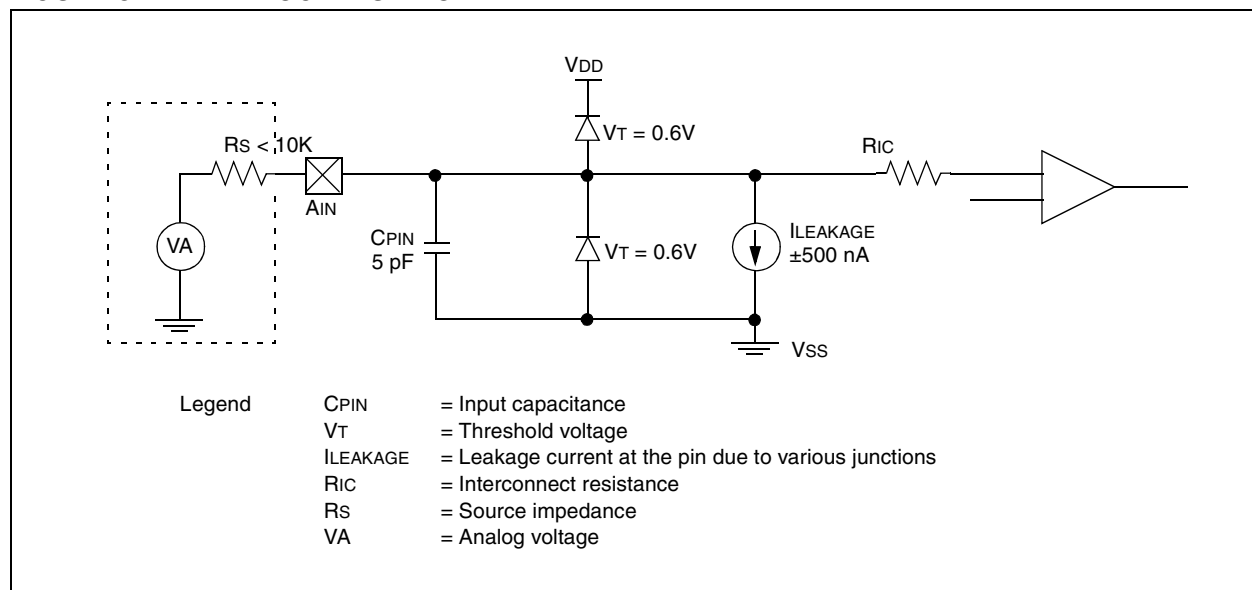
8.8 Effects of a RESET

A device reset forces the CMCON register to its reset state. This forces the comparator module to be in the comparator reset mode, CM<2:0> = 000. This ensures that all potential inputs are analog inputs. Device current is minimized when analog inputs are present at reset time. The comparators will be powered-down during the reset interval.

8.9 Analog Input Connection Considerations

A simplified circuit for an analog input is shown in Figure 8-4. Since the analog pins are connected to a digital output, they have reverse biased diodes to VDD and VSS. The analog input therefore, must be between VSS and VDD. If the input voltage deviates from this range by more than 0.6V in either direction, one of the diodes is forward biased and a latch-up may occur. A maximum source impedance of 10 kΩ is recommended for the analog sources. Any external component connected to an analog input pin, such as a capacitor or a Zener diode, should have very little leakage current.

FIGURE 8-4: ANALOG INPUT MODEL



PIC16CE62X

TABLE 8-1: REGISTERS ASSOCIATED WITH COMPARATOR MODULE

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR	Value on All Other Resets
1Fh	CMCON	C2OUT	C1OUT	—	—	CIS	CM2	CM1	CM0	00-- 0000	00-- 0000
9Fh	VRCON	VREN	VROE	VRR	—	VR3	VR2	VR1	VR0	000- 0000	000- 0000
0Bh	INTCON	GIE	PEIE	T0IE	INTE	RBIE	T0IF	INTF	RBIF	0000 000x	0000 000u
0Ch	PIR1	—	CMIF	—	—	—	—	—	—	-0-- ----	-0-- ----
8Ch	PIE1	—	CMIE	—	—	—	—	—	—	-0-- ----	-0-- ----
85h	TRISA	—	—	—	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	---1 1111	---1 1111

Legend: - = Unimplemented, read as "0", x = Unknown, u = unchanged

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TABLE 10-5: INITIALIZATION CONDITION FOR SPECIAL REGISTERS

Condition	Program Counter	STATUS Register	PCON Register
Power-on Reset	000h	0001 1xxx	---- --0x
MCLR reset during normal operation	000h	000u uuuu	---- --uu
MCLR reset during SLEEP	000h	0001 0uuu	---- --uu
WDT reset	000h	0000 uuuu	---- --uu
WDT Wake-up	PC + 1	uuu0 0uuu	---- --uu
Brown-out Reset	000h	000x xuuu	---- --u0
Interrupt Wake-up from SLEEP	PC + 1 ⁽¹⁾	uuu1 0uuu	---- --uu

Legend: u = unchanged, x = unknown, - = unimplemented bit, reads as '0'.

Note 1: When the wake-up is due to an interrupt and global enable bit, GIE is set and the PC is loaded with the interrupt vector (0004h) after execution of PC+1.

TABLE 10-6: INITIALIZATION CONDITION FOR REGISTERS

Register	Address	Power-on Reset	- MCLR Reset during normal operation - MCLR Reset during SLEEP - WDT Reset - Brown-out Reset ⁽¹⁾	- Wake-up from SLEEP through interrupt - Wake-up from SLEEP through WDT time-out
W	-	xxxx xxxx	uuuu uuuu	uuuu uuuu
INDF	00h	-	-	-
TMR0	01h	xxxx xxxx	uuuu uuuu	uuuu uuuu
PCL	02h	0000 0000	0000 0000	PC + 1 ⁽³⁾
STATUS	03h	0001 1xxx	000q quuu ⁽⁴⁾	uuuq quuu ⁽⁴⁾
FSR	04h	xxxx xxxx	uuuu uuuu	uuuu uuuu
PORTA	05h	---x xxxx	---u uuuu	---u uuuu
PORTB	06h	xxxx xxxx	uuuu uuuu	uuuu uuuu
CMCON	1Fh	00-- 0000	00-- 0000	uu-- uuuu
PCLATH	0Ah	---0 0000	---0 0000	---u uuuu
INTCON	0Bh	0000 000x	0000 000u	uuuu uqqq ⁽²⁾
PIR1	0Ch	-0-- ----	-0-- ----	-q-- ---- ^(2,5)
OPTION	81h	1111 1111	1111 1111	uuuu uuuu
TRISA	85h	---1 1111	---1 1111	---u uuuu
TRISB	86h	1111 1111	1111 1111	uuuu uuuu
PIE1	8Ch	-0-- ----	-0-- ----	-u-- ----
PCON	8Eh	---- --0x	---- --uq ^(1,6)	---- --uu
EEINTF	90h	---- -111	---- -111	---- -111
VRCON	9Fh	000- 0000	000- 0000	uuu- uuuu

Legend: u = unchanged, x = unknown, - = unimplemented bit, reads as '0', q = value depends on condition.

Note 1: If VDD goes too low, Power-on Reset will be activated and registers will be affected differently.

2: One or more bits in INTCON, PIR1 and/or PIR2 will be affected (to cause wake-up).

3: When the wake-up is due to an interrupt and the GIE bit is set, the PC is loaded with the interrupt vector (0004h).

4: See Table 10-5 for reset value for specific condition.

5: If wake-up was due to comparator input changing, then bit 6 = 1. All other interrupts generating a wake-up will cause bit 6 = u.

6: If reset was due to brown-out, then PCON bit 0 = 0. All other resets will cause bit 0 = u.

10.5 Interrupts

The PIC16CE62X has 4 sources of interrupt:

- External interrupt RB0/INT
- TMR0 overflow interrupt
- PortB change interrupts (pins RB<7:4>)
- Comparator interrupt

The interrupt control register (INTCON) records individual interrupt requests in flag bits. It also has individual and global interrupt enable bits.

A global interrupt enable bit, GIE (INTCON<7>) enables (if set) all un-masked interrupts or disables (if cleared) all interrupts. Individual interrupts can be disabled through their corresponding enable bits in INTCON register. GIE is cleared on reset.

The “return from interrupt” instruction, RETFIE, exits interrupt routine, as well as sets the GIE bit, which re-enable RB0/INT interrupts.

The INT pin interrupt, the RB port change interrupt and the TMR0 overflow interrupt flags are contained in the INTCON register.

The peripheral interrupt flag is contained in the special register PIR1. The corresponding interrupt enable bit is contained in special registers PIE1.

When an interrupt is responded to, the GIE is cleared to disable any further interrupt, the return address is pushed into the stack and the PC is loaded with 0004h. Once in the interrupt service routine, the source(s) of

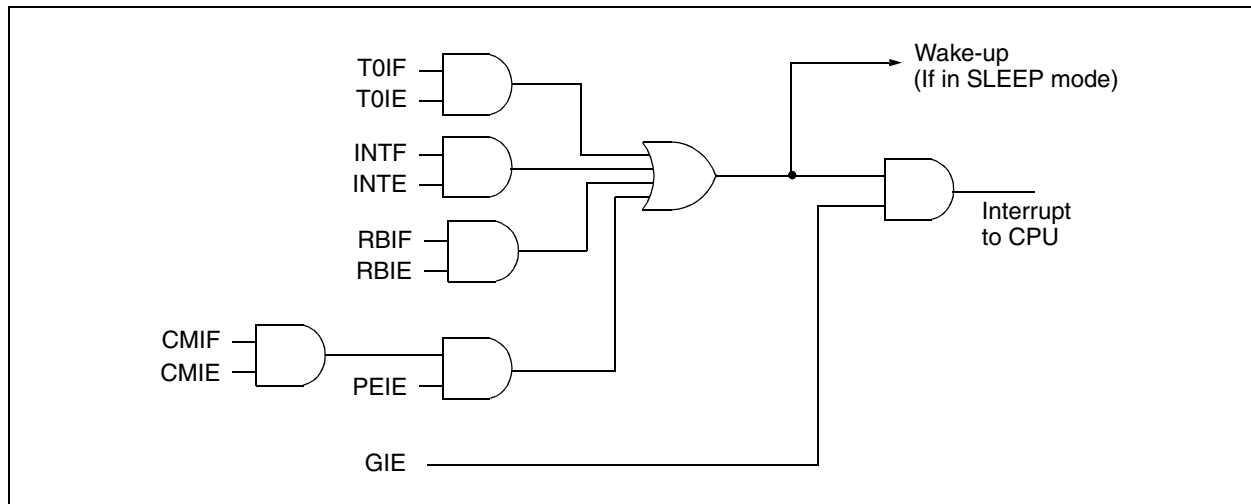
the interrupt can be determined by polling the interrupt flag bits. The interrupt flag bit(s) must be cleared in software before re-enabling interrupts to avoid RB0/INT recursive interrupts.

For external interrupt events, such as the INT pin or PORTB change interrupt, the interrupt latency will be three or four instruction cycles. The exact latency depends on when the interrupt event occurs (Figure 10-16). The latency is the same for one or two cycle instructions. Once in the interrupt service routine the source(s) of the interrupt can be determined by polling the interrupt flag bits. The interrupt flag bit(s) must be cleared in software before re-enabling interrupts to avoid multiple interrupt requests.

Note 1: Individual interrupt flag bits are set, regardless of the status of their corresponding mask bit or the GIE bit.

2: When an instruction that clears the GIE bit is executed, any interrupts that were pending for execution in the next cycle are ignored. The CPU will execute a NOP in the cycle immediately following the instruction which clears the GIE bit. The interrupts which were ignored are still pending to be serviced when the GIE bit is set again.

FIGURE 10-15: INTERRUPT LOGIC



10.6 Context Saving During Interrupts

During an interrupt, only the return PC value is saved on the stack. Typically, users may wish to save key registers during an interrupt (i.e. W register and STATUS register). This will have to be implemented in software.

Example 10-1 stores and restores the STATUS and W registers. The user register, W_TEMP, must be defined in both banks and must be defined at the same offset from the bank base address (i.e., W_TEMP is defined at 0x70 in Bank 0 and it must also be defined at 0xF0 in Bank 1). The user register, STATUS_TEMP, must be defined in Bank 0. The Example 10-1:

- Stores the W register
- Stores the STATUS register in Bank 0
- Executes the ISR code
- Restores the STATUS (and bank select bit register)
- Restores the W register

EXAMPLE 10-1: SAVING THE STATUS AND W REGISTERS IN RAM

```
MOVWF    W_TEMP        ;copy W to temp register,
                        ;could be in either bank

SWAPF    STATUS,W       ;swap status to be saved into W

BCF      STATUS,RP0     ;change to bank 0 regardless
                        ;of current bank

MOVWF    STATUS_TEMP    ;save status to bank 0
                        ;register

:
:   (ISR)
:

SWAPF    STATUS_TEMP,W  ;swap STATUS_TEMP register
                        ;into W, sets bank to original
                        ;state

MOVWF    STATUS         ;move W into STATUS register

SWAPF    W_TEMP,F       ;swap W_TEMP

SWAPF    W_TEMP,W       ;swap W_TEMP into W
```

10.7 Watchdog Timer (WDT)

The Watchdog Timer is a free running on-chip RC oscillator which does not require any external components. This RC oscillator is separate from the RC oscillator of the CLKIN pin. That means that the WDT will run, even if the clock on the OSC1 and OSC2 pins of the device have been stopped, for example, by execution of a SLEEP instruction. During normal operation, a WDT time-out generates a device RESET. If the device is in SLEEP mode, a WDT time-out causes the device to wake-up and continue with normal operation. The WDT can be permanently disabled by programming the configuration bit WDTE as clear (Section 10.1).

10.7.1 WDT PERIOD

The WDT has a nominal time-out period of 18 ms, (with no prescaler). The time-out periods vary with temperature, VDD and process variations from part to part (see DC specs). If longer time-out periods are desired, a prescaler with a division ratio of up to 1:128 can be assigned to the WDT under software control by writing to the OPTION register. Thus, time-out periods up to 2.3 seconds can be realized.

The CLRWDT and SLEEP instructions clear the WDT and the postscaler, if assigned to the WDT, and prevent it from timing out and generating a device RESET.

The $\overline{\text{TO}}$ bit in the STATUS register will be cleared upon a Watchdog Timer time-out.

10.7.2 WDT PROGRAMMING CONSIDERATIONS

It should also be taken in account that under worst case conditions (VDD = Min., Temperature = Max., max. WDT prescaler), it may take several seconds before a WDT time-out occurs.

11.1 Instruction Descriptions

ADDLW Add Literal and W

Syntax:	[<i>label</i>] ADDLW k				
Operands:	$0 \leq k \leq 255$				
Operation:	$(W) + k \rightarrow (W)$				
Status Affected:	C, DC, Z				
Encoding:	<table border="1"><tr><td>11</td><td>111x</td><td>kkkk</td><td>kkkk</td></tr></table>	11	111x	kkkk	kkkk
11	111x	kkkk	kkkk		
Description:	The contents of the W register are added to the eight bit literal 'k' and the result is placed in the W register.				
Words:	1				
Cycles:	1				
Example	ADDLW 0x15 Before Instruction W = 0x10 After Instruction W = 0x25				

ANDLW AND Literal with W

Syntax:	[<i>label</i>] ANDLW <i>k</i>				
Operands:	$0 \leq k \leq 255$				
Operation:	(W) .AND. (k) $\rightarrow$ (W)				
Status Affected:	Z				
Encoding:	<table border="1"><tr><td>11</td><td>1001</td><td>kkkk</td><td>kkkk</td></tr></table>	11	1001	kkkk	kkkk
11	1001	kkkk	kkkk		
Description:	The contents of W register are AND'ed with the eight bit literal 'k'. The result is placed in the W register.				
Words:	1				
Cycles:	1				
Example	ANDLW 0x5F Before Instruction W = 0xA3 After Instruction W = 0x03				

ADDWF Add W and f

Syntax:	[<i>label</i>] ADDWF <i>f</i> , <i>d</i>				
Operands:	$0 \leq f \leq 127$ $d \in [0,1]$				
Operation:	$(W) + (f) \rightarrow (dest)$				
Status Affected:	C, DC, Z				
Encoding:	<table><tr><td>00</td><td>0111</td><td>dfff</td><td>ffff</td></tr></table>	00	0111	dfff	ffff
00	0111	dfff	ffff		
Description:	Add the contents of the W register with register 'f'. If 'd' is 0, the result is stored in the W register. If 'd' is 1, the result is stored back in register 'f'.				
Words:	1				
Cycles:	1				
Example	<pre> ADDWF FSR, 0 Before Instruction W = 0x17 FSR = 0xC2 After Instruction W = 0xD9 FSR = 0xC2</pre>				

ANDWF AND W with f

Syntax:	[<i>label</i>] ANDWF <i>f</i> , <i>d</i>				
Operands:	$0 \leq f \leq 127$ $d \in [0,1]$				
Operation:	(W) .AND. (f) $\rightarrow$ (dest)				
Status Affected:	Z				
Encoding:	<table border="1"><tr><td>00</td><td>0101</td><td>dfff</td><td>ffff</td></tr></table>	00	0101	dfff	ffff
00	0101	dfff	ffff		
Description:	AND the W register with register 'f'. If 'd' is 0, the result is stored in the W register. If 'd' is 1, the result is stored back in register 'f'.				
Words:	1				
Cycles:	1				
Example	<pre> ANDWF FSR, 1 Before Instruction W = 0x17 FSR = 0xC2 After Instruction W = 0x17 FSR = 0x02</pre>				

TABLE 12-1: DEVELOPMENT TOOLS FROM MICROCHIP

	PIC12CXX	PIC14000	PIC16C5X	PIC16C6X	PIC16CXX	PIC16F62X	PIC16C7X	PIC16C7XX	PIC16C8X	PIC16F8XX	PIC16C9XX	PIC17C4X	PIC17C7XX	PIC18CXX2	24CXX/ 25CXX/ 93CXX	HCSXX	MCRFXX	MCP2510
Demo Boards and Eval Kits	MPLAB® Integrated Development Environment	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓			✓
	MPLAB® C17 Compiler											✓	✓					
	MPLAB® C18 Compiler													✓				
	MPASM/MPLINK	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓		
Emulators	MPLAB®-ICE	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓			
	PICMASTER/PICMASTER-CE	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓			
Debuggers	ICEPIC™ Low-Cost In-Circuit Emulator	✓	✓	✓	✓	✓	✓	✓	✓		✓							
	MPLAB®-ICD In-Circuit Debugger				✓		✓			✓								
Programmers	PICSTART® Plus Low-Cost Universal Dev. Kit	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓			
	PRO MATE® II Universal Programmer	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓		
Demo Boards and Eval Kits	SIMICE	✓	✓	✓														
	PICDEM-1		✓	✓			†		✓			✓						
	PICDEM-2				†		†							✓				
	PICDEM-3										✓							
	PICDEM-14A		✓										✓					
	PICDEM-17												✓					
	KEELOO® Evaluation Kit														✓			
	KEELOO Transponder Kit														✓			
	microID™ Programmer's Kit															✓		
	125 kHz microID Developer's Kit																✓	
	125 kHz Anticollision microID Developer's Kit																✓	
	13.56 MHz Anticollision microID Developer's Kit																✓	
	MCP2510 CAN Developer's Kit																✓	✓

* Contact the Microchip Technology Inc. web site at www.microchip.com for information on how to use the MPLAB®-ICD In-Circuit Debugger (DV164001) with PIC16C62, 63, 64, 65, 72, 73, 74, 76, 77

** Contact Microchip Technology Inc. for availability date.

† Development tool is available on select devices.

13.0 ELECTRICAL SPECIFICATIONS

Absolute Maximum Ratings †

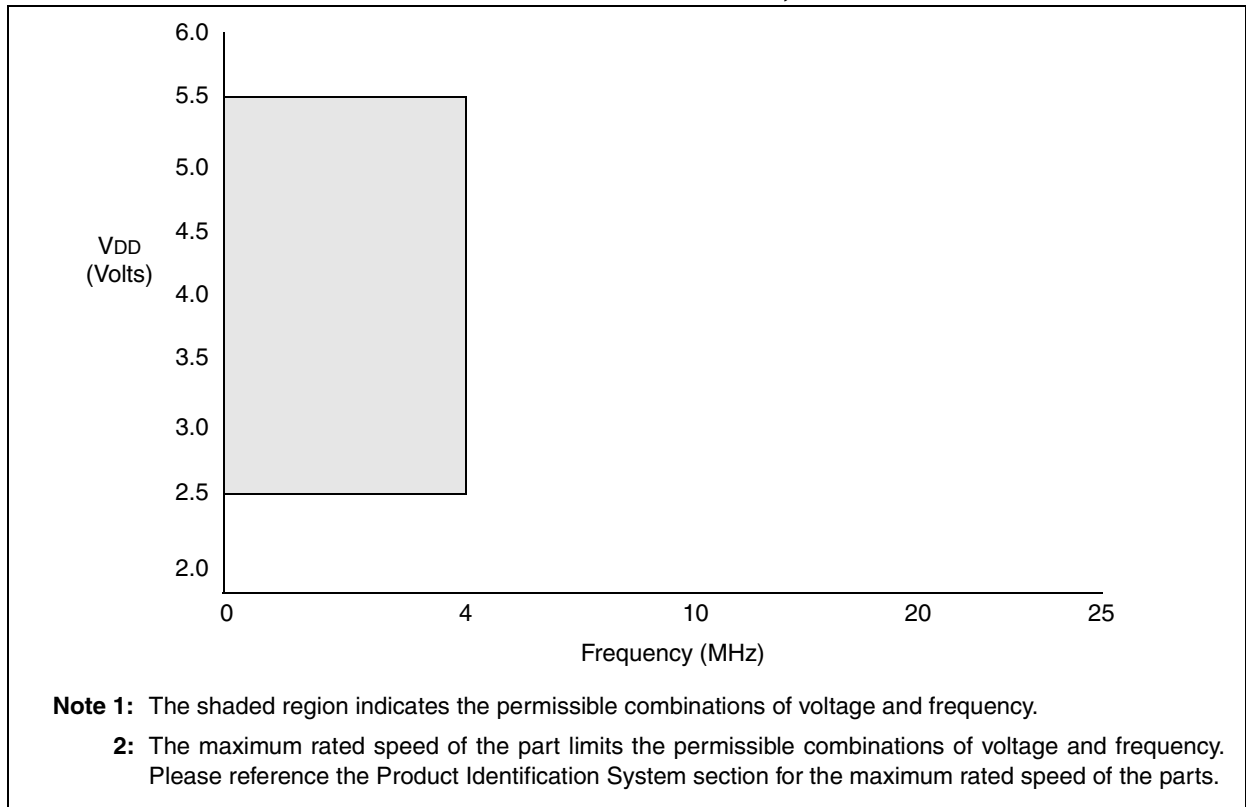
Ambient Temperature under bias	-40° to +125°C
Storage Temperature	-65° to +150°C
Voltage on any pin with respect to VSS (except VDD and $\overline{\text{MCLR}}$).....	-0.6V to VDD +0.6V
Voltage on VDD with respect to VSS	0 to +7.0V
Voltage on RA4 with respect to VSS.....	8.5V
Voltage on $\overline{\text{MCLR}}$ with respect to VSS (Note 2).....	0 to +14V
Voltage on RA4 with respect to VSS.....	8.5V
Total power Dissipation (Note 1)	1.0W
Maximum Current out of VSS pin.....	300 mA
Maximum Current into VDD pin	250 mA
Input Clamp Current, I _{IK} (V _I < 0 or V _I > VDD)	±20 mA
Output Clamp Current, I _{OK} (V _O < 0 or V _O > VDD).....	±20 mA
Maximum Output Current sunk by any I/O pin	25 mA
Maximum Output Current sourced by any I/O pin.....	25 mA
Maximum Current sunk by PORTA and PORTB	200 mA
Maximum Current sourced by PORTA and PORTB	200 mA

Note 1: Power dissipation is calculated as follows: $P_{DIS} = V_{DD} \times \{I_{DD} - \sum I_{OH}\} + \sum \{(V_{DD} - V_{OH}) \times I_{OH}\} + \sum (V_{OL} \times I_{OL})$

2: Voltage spikes below VSS at the $\overline{\text{MCLR}}$ pin, inducing currents greater than 80 mA, may cause latch-up. Thus, a series resistor of 50-100 Ω should be used when applying a "low" level to the $\overline{\text{MCLR}}$ pin rather than pulling this pin directly to VSS.

† **NOTICE:** Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operation listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

FIGURE 13-3: PIC16LCE62X VOLTAGE-FREQUENCY GRAPH, $-40^{\circ}\text{C} \leq T_A < +125^{\circ}\text{C}$



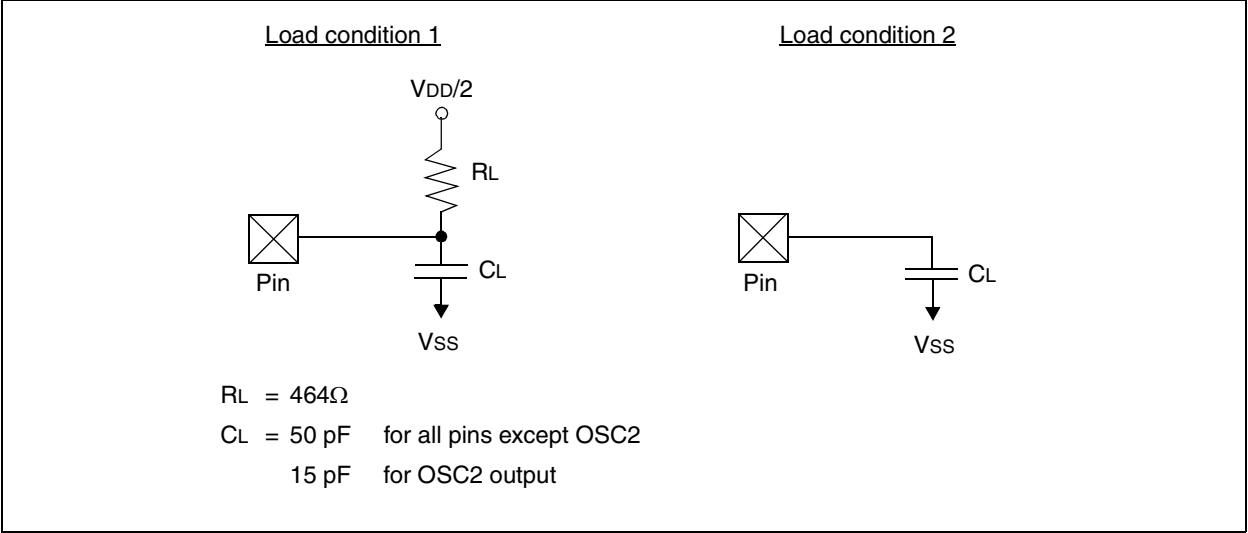
13.4 Timing Parameter Symbology

The timing parameter symbols have been created with one of the following formats:

- 1. TppS2ppS
- 2. TppS

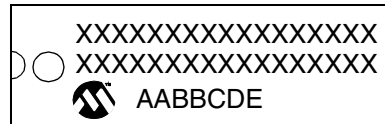
T			
F	Frequency	T	Time
Lowercase subscripts (pp) and their meanings:			
pp			
ck	CLKOUT	osc	OSC1
io	I/O port	t0	T0CKI
mc	MCLR		
Uppercase letters and their meanings:			
S			
F	Fall	P	Period
H	High	R	Rise
I	Invalid (Hi-impedance)	V	Valid
L	Low	Z	Hi-Impedance

FIGURE 13-4: LOAD CONDITIONS

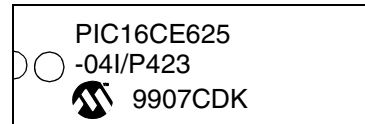


14.1 Package Marking Information

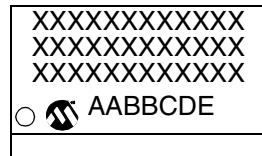
18-Lead PDIP



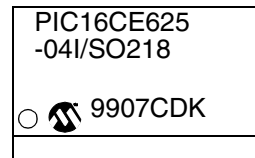
Example



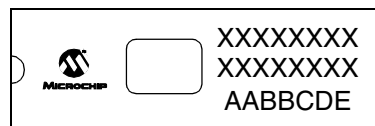
18-Lead SOIC (.300")



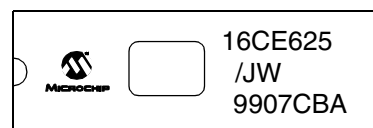
Example



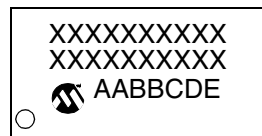
18-Lead CERDIP Windowed



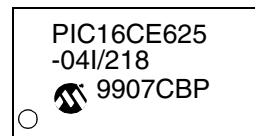
Example



20-Lead SSOP



Example



Legend:	XX...X	Customer-specific information
	Y	Year code (last digit of calendar year)
	YY	Year code (last 2 digits of calendar year)
	WW	Week code (week of January 1 is week '01')
	NNN	Alphanumeric traceability code
	(e3)	Pb-free JEDEC designator for Matte Tin (Sn)
	*	This package is Pb-free. The Pb-free JEDEC designator (e3) can be found on the outer packaging for this package.

Note: In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line, thus limiting the number of available characters for customer-specific information.

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