Lattice Semiconductor Corporation - LFE5U-25F-6BG256C Datasheet

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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Active
Number of LABs/CLBs	6000
Number of Logic Elements/Cells	24000
Total RAM Bits	1032192
Number of I/O	197
Number of Gates	-
Voltage - Supply	1.045V ~ 1.155V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	256-LFBGA
Supplier Device Package	256-CABGA (14x14)
Purchase URL	https://www.e-xfl.com/product-detail/lattice-semiconductor/lfe5u-25f-6bg256c

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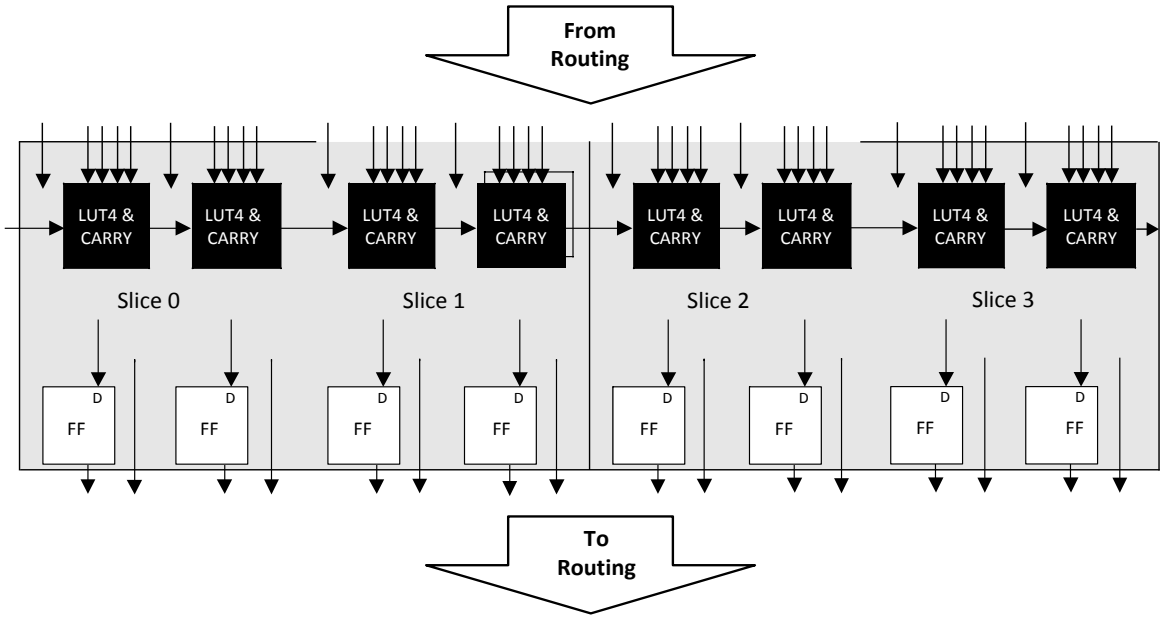

Figure 2.2. PFU Diagram

2.2.1. Slice

Each slice contains two LUT4s feeding two registers. In Distributed SRAM mode, Slice 0 through Slice 2 are configured as distributed memory, and Slice 3 is used as Logic or ROM. Table 2.1 shows the capability of the slices along with the operation modes they enable. In addition, each PFU contains logic that allows the LUTs to be combined to perform functions such as LUT5, LUT6, LUT7 and LUT8. There is control logic to perform set/reset functions (programmable as synchronous/ asynchronous), clock select, chip-select and wider RAM/ROM functions.

Table 2.1. Resources and Modes Available per Slice

Slice	PFU (Used in Distributed SRAM)		PFU (Not used as Distributed SRAM)	
	Resources	Modes	Resources	Modes
Slice 0	2 LUT4s and 2 Registers	RAM	2 LUT4s and 2 Registers	Logic, Ripple, ROM
Slice 1	2 LUT4s and 2 Registers	RAM	2 LUT4s and 2 Registers	Logic, Ripple, ROM
Slice 2	2 LUT4s and 2 Registers	RAM	2 LUT4s and 2 Registers	Logic, Ripple, ROM
Slice 3	2 LUT4s and 2 Registers	Logic, Ripple, ROM	2 LUT4s and 2 Registers	Logic, Ripple, ROM

Figure 2.3 shows an overview of the internal logic of the slice. The registers in the slice can be configured for positive/negative and edge triggered or level sensitive clocks.

Each slice has 14 input signals, 13 signals from routing and one from the carry-chain (from the adjacent slice or PFU). There are five outputs, four to routing and one to carry-chain (to the adjacent PFU). There are two inter slice/ PFU output signals that are used to support wider LUT functions, such as LUT6, LUT7 and LUT8. Table 2.2 and Figure 2.3 list the signals associated with all the slices. Figure 2.4 on page 16 shows the connectivity of the inter-slice/PFU signals that support LUT5, LUT6, LUT7 and LUT8.

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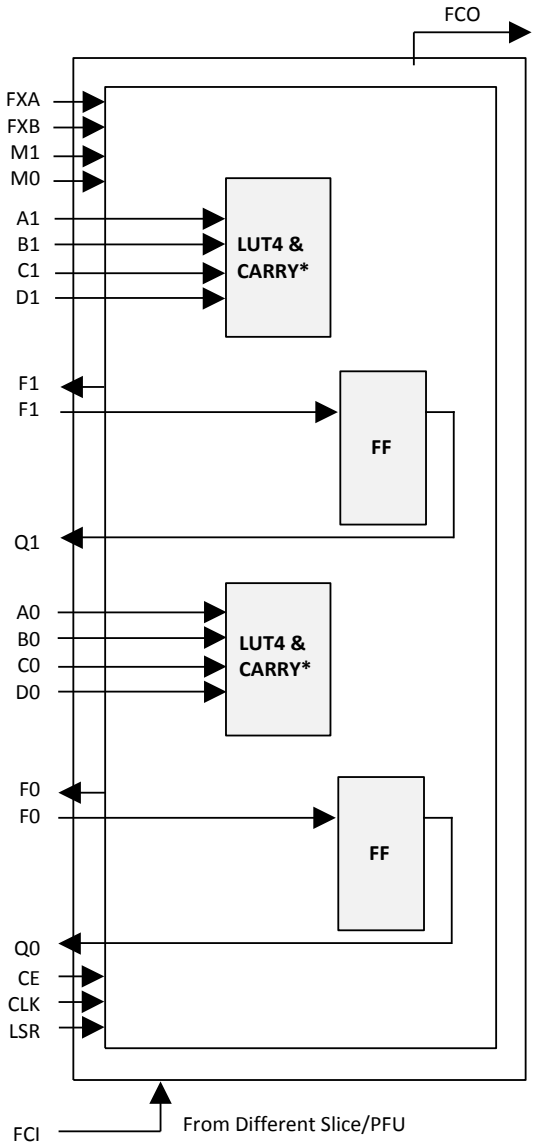

Notes: For Slices 0 and 1, memory control signals are generated from Slice 2 as follows:
WCK is CLK
WRE is from LSR
DI[3:2] for Slice 1 and DI[1:0] for Slice 0 data from Slice 2
WAD [A:D] is a 4-bit address from slice 2 LUT input

Figure 2.3. Slice Diagram

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2.7. DDRDLL

Every DDRDLL (master DLL block) can generate phase shift code representing the amount of delay in a delay block that corresponding to 90° phase of the reference clock input. The reference clock can be either from PLL, or input pin. This code is used in the DQSBUF block that controls a set of DQS pin groups to interface with DDR memory (slave DLL). There are two DDRDLLs that supply two sets of codes (for two different reference clock frequencies) to each side of the I/Os (at each of the corners). The DQSBUF uses this code to controls the DQS input of the DDR memory to 90° shift to clock DQs at the center of the data eye for DDR memory interface.

The code is also sent to another slave DLL, DLLDEL, that takes a clock input and generates a 90° shift clock output to drive the clocking structure. This is useful to interface edge-aligned Generic DDR, where 90° clocking needs to be created. Figure 2.10 shows DDRDLL functional diagram.

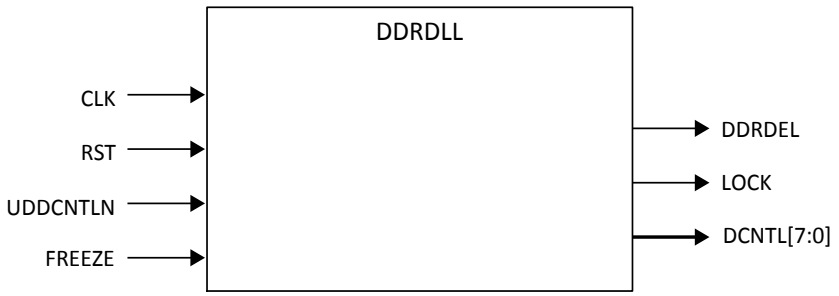

Figure 2.10. DDRDLL Functional Diagram

Table 2.5. DDRDLL Ports List

Port Name	Type	Description
CLK	Input	Reference clock input to the DDRDLL. Should run at the same frequency as the clock to the delay code.
RST	Input	Reset Input to the DDRDLL.
UDDCNTLN	Input	Update Control to update the delay code. The code is the DCNTL[7:0] outputs. These outputs are updated when the UDDCNTLN signal is LOW.
FREEZE	Input	FREEZE goes high and, without a glitch, turns off the DLL internal clock and the ring oscillator output clock. When FREEZE goes low, it turns them back on.
DDRDEL	Output	The delay codes from the DDRDLL to be used in DQSBUF or DLLDEL.
LOCK	Output	Lock output to indicate the DDRDLL has valid delay output.
DCNTL [7:0]	Output	The delay codes from the DDRDLL available for the user IP.

There are four identical DDRDLLs, one in each of the four corners in LFE5-85 and LFE5-45 devices, and two DDRDLLs in both LFE5-25 & LFE5-12 devices in the upper two corners. Each DDRDLL can generate delay code based on the reference frequency. The slave DLL (DQSBUF and DLLDEL) use the code to delay the signal, to create the phase shifted signal used for either DDR memory, to create 90° shift clock. Figure 2.11 shows the DDRDLL and the slave DLLs on the top level view.

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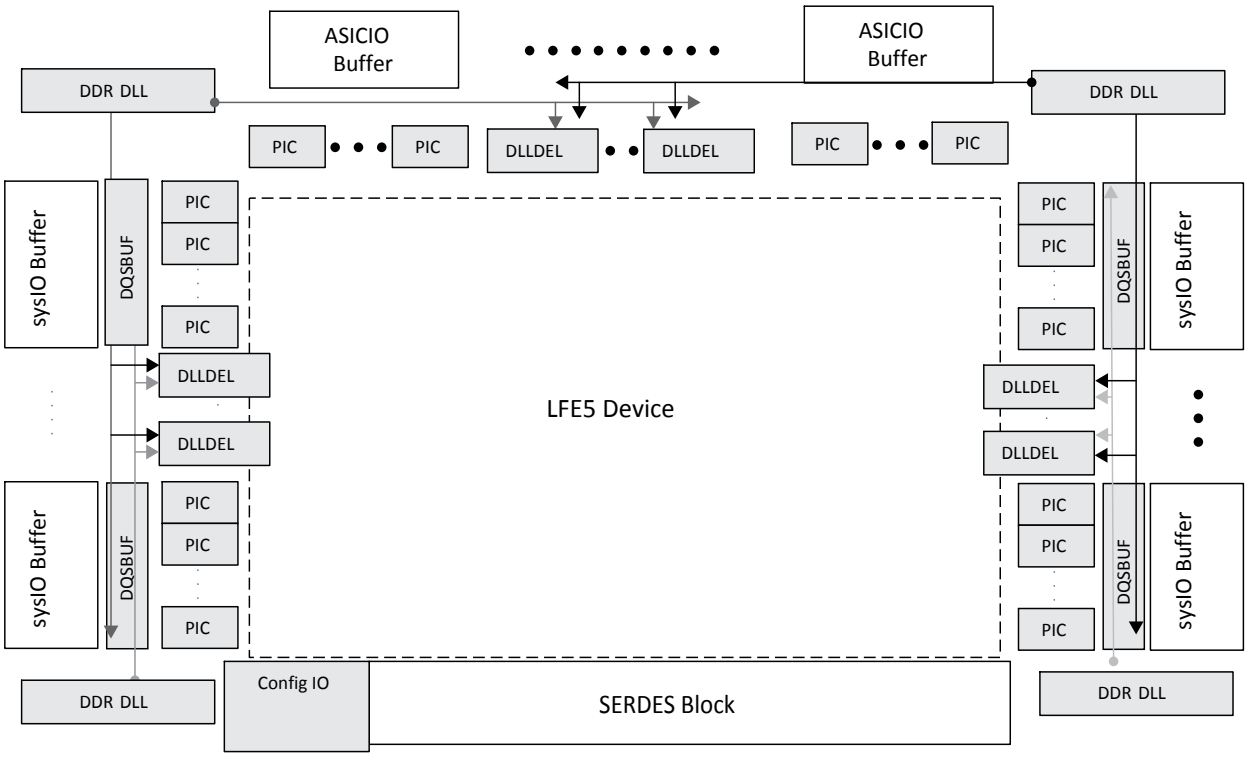

Figure 2.11. ECP5/ECP5-5G DLL Top Level View (For LFE-45 and LFE-85)

2.8. sysMEM Memory

ECP5/ECP5-5G devices contain a number of sysMEM Embedded Block RAM (EBR). The EBR consists of an 18 Kb RAM with memory core, dedicated input registers and output registers with separate clock and clock enable. Each EBR includes functionality to support true dual-port, pseudo dual-port, single-port RAM, ROM and FIFO buffers (via external PFUs).

2.8.1. sysMEM Memory Block

The sysMEM block can implement single port, dual port or pseudo dual port memories. Each block can be used in a variety of depths and widths as listed in Table 2.6 on page 25. FIFOs can be implemented in sysMEM EBR blocks by implementing support logic with PFUs. The EBR block facilitates parity checking by supporting an optional parity bit for each data byte. EBR blocks provide byte-enable support for configurations with 18-bit and 36-bit data widths. For more information, refer to ECP5 and ECP5-5G Memory Usage Guide (TN1264).

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Table 2.6. sysMEM Block Configurations

Memory Mode	Configurations
Single Port	16,384 x 1 8,192 x 2 4,096 x 4 2,048 x 9 1,024 x 18 512 x 36
True Dual Port	16,384 x 1 8,192 x 2 4,096 x 4 2,048 x 9 1,024 x 18
Pseudo Dual Port	16,384 x 1 8,192 x 2 4,096 x 4 2,048 x 9 1,024 x 18 512 x 36

2.8.2. Bus Size Matching

All of the multi-port memory modes support different widths on each of the ports. The RAM bits are mapped LSB word 0 to MSB word 0, LSB word 1 to MSB word 1, and so on. Although the word size and number of words for each port varies, this mapping scheme applies to each port.

2.8.3. RAM Initialization and ROM Operation

If desired, the contents of the RAM can be pre-loaded during device configuration. By preloading the RAM block during the chip configuration cycle and disabling the write controls, the sysMEM block can also be utilized as a ROM.

2.8.4. Memory Cascading

Larger and deeper blocks of RAM can be created using EBR sysMEM Blocks. Typically, the Lattice design tools cascade memory transparently, based on specific design inputs.

2.8.5. Single, Dual and Pseudo-Dual Port Modes

In all the sysMEM RAM modes the input data and address for the ports are registered at the input of the memory array. The output data of the memory is optionally registered at the output.

EBR memory supports the following forms of write behavior for single port or dual port operation:

- **Normal** – Data on the output appears only during a read cycle. During a write cycle, the data (at the current address) does not appear on the output. This mode is supported for all data widths.
- **Write Through** – A copy of the input data appears at the output of the same port during a write cycle. This mode is supported for all data widths.
- **Read-Before-Write** – When new data is written, the old content of the address appears at the output. This mode is supported for x9, x18, and x36 data widths.

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2.14. sysI/O Buffer

Each I/O is associated with a flexible buffer referred to as a sysI/O buffer. These buffers are arranged around the periphery of the device in groups referred to as banks. The sysI/O buffers allow users to implement the wide variety of standards that are found in today's systems including LVDS, HSUL, BLVDS, SSTL Class I and II, LVCMOS, LVTTL, LVPECL, and MIPI.

2.14.1. sysI/O Buffer Banks

ECP5/ECP5-5G devices have seven sysI/O buffer banks, two banks per side at Top, Left and Right, plus one at the bottom left side. The bottom left side bank (Bank 8) is a shared I/O bank. The I/Os in that bank contains both dedicated and shared I/O for sysConfig function. When a shared pin is not used for configuration, it is available as a user I/O. For LFE5-85 devices, there is an additional I/O bank (Bank 4) that is not available in other device in the family.

In ECP5/ECP5-5G devices, the Left and Right sides are tailored to support high performance interfaces, such as DDR2, DDR3, LPDDR2, LPDDR3 and other high speed source synchronous standards. The banks on the Left and Right sides of the devices feature LVDS input and output buffers, data-width gearing, and DQSBUF block to support DDR2/3 and LPDDR2/3 interfaces. The I/Os on the top and bottom banks do not have LVDS input and output buffer, and gearing logic, but can use LVCMOS to emulate most of differential output signaling.

Each sysIO bank has its own I/O supply voltage (V_{CCIO}). In addition, the banks on the Left and Right sides of the device, have voltage reference input (shared I/O pin), VREF1 per bank, which allow it to be completely independent of each other. The V_{REF} voltage is used to set the threshold for the referenced input buffers, such as SSTL. Figure 2.25 shows the seven banks and their associated supplies.

In ECP5/ECP5-5G devices, single-ended output buffers and ratioed input buffers (LVTTL, and LVCMOS) are powered using V_{CCIO}. LVTTL, LVCMOS33, LVCMOS25 and LVCMOS12 can also be set as fixed threshold inputs independent of V_{CCIO}.

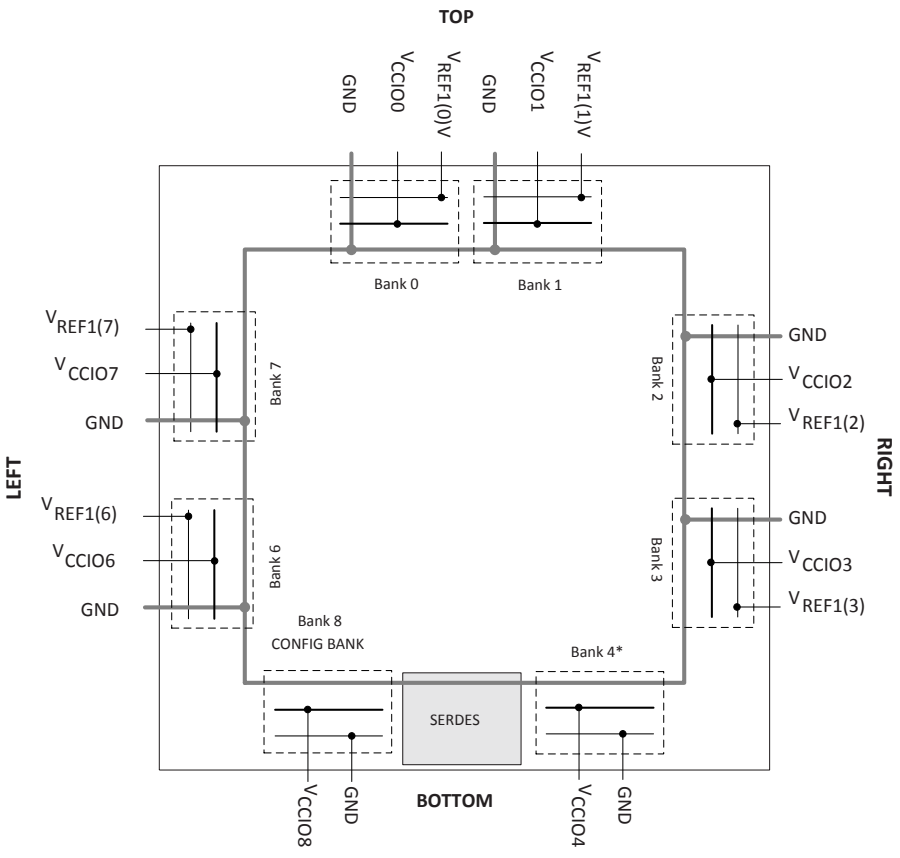

Figure 2.25. ECP5/ECP5-5G Device Family Banks

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2.14.4. On-Chip Programmable Termination

The ECP5/ECP5-5G devices support a variety of programmable on-chip terminations options, including:

- Dynamically switchable Single-Ended Termination with programmable resistor values of 50 Ω, 75 Ω, or 150 Ω.
- Common mode termination of 100 Ω for differential inputs.

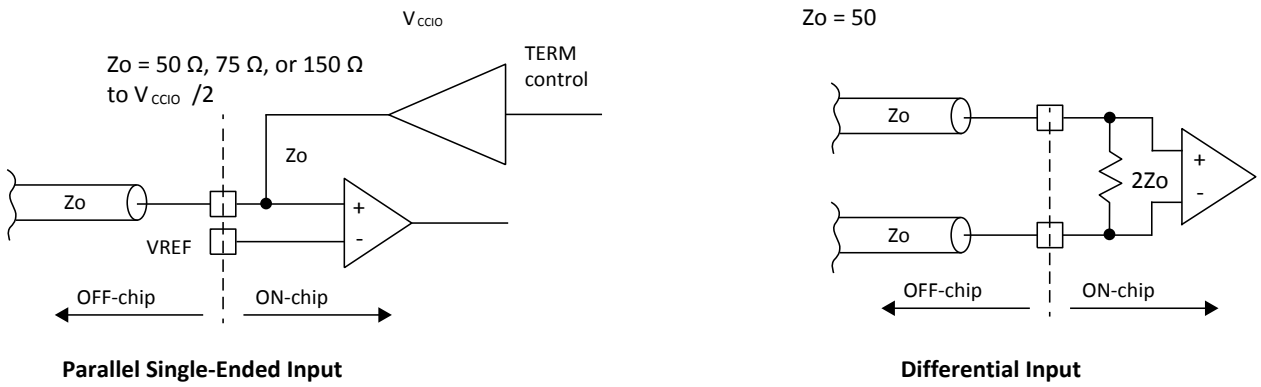

Figure 2.26. On-Chip Termination

See Table 2.12 for termination options for input modes.

Table 2.12. On-Chip Termination Options for Input Modes

IO_TYPE	Terminate to $V_{CCIO}/2$*	Differential Termination Resistor*
LVDS25	—	100
BLVDS25	—	100
MLVDS	—	100
LVPECL33	—	100
subLVDS	—	100
SLVS	—	100
HSUL12	50, 75, 150	—
HSUL12D	—	100
SSTL135_I / II	50, 75, 150	—
SSTL135D_I / II	—	100
SSTL15_I / II	50, 75, 150	—
SSTL15D_I / II	—	100
SSTL18_I / II	50, 75, 150	—
SSTL18D_I / II	—	100

***Notes**:

TERMINATE to $V_{CCIO}/2$ (Single-Ended) and DIFFRENTIAL TERMINATION RESISTOR when turned on can only have one setting per bank. Only left and right banks have this feature.

Use of TERMINATE to $V_{CCIO}/2$ and DIFFRENTIAL TERMINATION RESISTOR are mutually exclusive in an I/O bank. On-chip termination tolerance ±20%.

Refer to ECP5 and ECP5-5G sysIO Usage Guide (TN1262) for on-chip termination usage and value ranges.

2.14.5. Hot Socketing

ECP5/ECP5-5G devices have been carefully designed to ensure predictable behavior during power-up and power-down. During power-up and power-down sequences, the I/Os remain in tristate until the power supply voltage is high enough to ensure reliable operation. In addition, leakage into I/O pins is controlled within specified limits. See the Hot Socketing Specifications section on page 48.

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Table 2.14. Available SERDES Duals per LFE5UM/LFE5UM5G Devices

Package	LFE5UM/LFE5UM5G-25	LFE5UM/LFE5UM5G-45	LFE5UM/LFE5UM5G-85
285 csfBGA	1	1	1
381 caBGA	1	2	2
554 caBGA	—	2	2
756 caBGA	—	—	2

2.15.1. SERDES Block

A SERDES receiver channel may receive the serial differential data stream, equalize the signal, perform Clock and Data Recovery (CDR) and de-serialize the data stream before passing the 8- or 10-bit data to the PCS logic. The SERDES transmitter channel may receive the parallel 8- or 10-bit data, serialize the data and transmit the serial bit stream through the differential drivers. Figure 2.28 shows a single-channel SERDES/PCS block. Each SERDES channel provides a recovered clock and a SERDES transmit clock to the PCS block and to the FPGA core logic.

Each transmit channel, receiver channel, and SERDES PLL shares the same power supply (VCCA). The output and input buffers of each channel have their own independent power supplies (VCCHTX and VCCHRX).

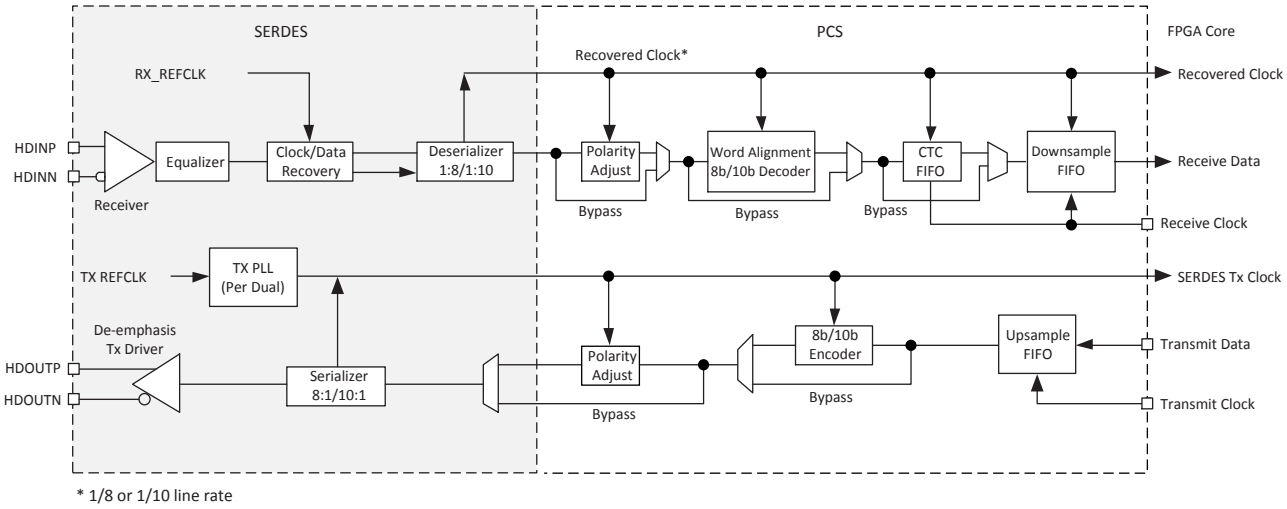

Figure 2.28. Simplified Channel Block Diagram for SERDES/PCS Block

2.15.2. PCS

As shown in Figure 2.28, the PCS receives the parallel digital data from the deserializer and selects the polarity, performs word alignment, decodes (8b/10b), provides Clock Tolerance Compensation and transfers the clock domain from the recovered clock to the FPGA clock via the Down Sample FIFO.

For the transmit channel, the PCS block receives the parallel data from the FPGA core, encodes it with 8b/10b, selects the polarity and passes the 8/10 bit data to the transmit SERDES channel.

The PCS also provides bypass modes that allow a direct 8-bit or 10-bit interface from the SERDES to the FPGA logic. The PCS interface to the FPGA can also be programmed to run at 1/2 speed for a 16-bit or 20-bit interface to the FPGA logic.

Some of the enhancements in LFE5UM/LFE5UM5G SERDES/PCS include:

- Higher clock/channel granularity: Dual channel architecture provides more clock resource per channel.
- Enhanced Tx de-emphasis: Programmable pre- and post-cursors improves Tx output signaling
- Bit-slip function in PCS: Improves logic needed to perform Word Alignment function

Refer to ECP5 and ECP5-5G SERDES/PCS Usage Guide (TN1261) for more information.

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2.15.3. SERDES Client Interface Bus

The SERDES Client Interface (SCI) is an IP interface that allows the user to change the configuration thru this interface. This is useful when the user needs to fine-tune some settings, such as input and output buffer that need to be optimized based on the channel characteristics. It is a simple register configuration interface that allows SERDES/PCS configuration without power cycling the device.

The Diamond design tools support all modes of the PCS. Most modes are dedicated to applications associated with a specific industry standard data protocol. Other more general purpose modes allow users to define their own operation. With these tools, the user can define the mode for each dual in a design.

Popular standards such as 10 Gb Ethernet, x4 PCI Express and 4x Serial RapidIO can be implemented using IP (available through Lattice), with two duals (Four SERDES channels and PCS) and some additional logic from the core.

The LFE5UM/LFE5UM5G devices support a wide range of protocols. Within the same dual, the LFE5UM/ LFE5UM5G devices support mixed protocols with semi-independent clocking as long as the required clock frequencies are integer x1, x2, or x11 multiples of each other. Table 2.15 lists the allowable combination of primary and secondary protocol combinations.

2.16. Flexible Dual SERDES Architecture

The LFE5UM/LFE5UM5G SERDES architecture is a dual channel-based architecture. For most SERDES settings and standards, the whole dual (consisting of two SERDES channels) is treated as a unit. This helps in silicon area savings, better utilization, higher granularity on clock/SERDES channel and overall lower cost.

However, for some specific standards, the LFE5UM/LFE5UM5G dual-channel architecture provides flexibility; more than one standard can be supported within the same dual.

Table 2.15 lists the standards that can be mixed and matched within the same dual. In general, the SERDES standards whose nominal data rates are either the same or a defined subset of each other, can be supported within the same dual. The two Protocol columns of the table define the different combinations of protocols that can be implemented together within a Dual.

Table 2.15. LFE5UM/LFE5UM5G Mixed Protocol Support

Protocol		Protocol
PCI Express 1.1	with	SGMII
PCI Express 1.1	with	Gigabit Ethernet
CPRI-3	with	CPRI-2 and CPRI-1
3G-SDI	with	HD-SDI and SD-SDI

There are some restrictions to be aware of when using spread spectrum clocking. When a dual shares a PCI Express x1 channel with a non-PCI Express channel, ensure that the reference clock for the dual is compatible with all protocols within the dual. For example, a PCI Express spread spectrum reference clock is not compatible with most Gigabit Ethernet applications because of tight CTC ppm requirements.

While the LFE5UM/LFE5UM5G architecture will allow the mixing of a PCI Express channel and a Gigabit Ethernet, or SGMII channel within the same dual, using a PCI Express spread spectrum clocking as the transmit reference clock will cause a violation of the Gigabit Ethernet, and SGMII transmit jitter specifications.

For further information on SERDES, refer to ECP5 and ECP5-5G SERDES/PCS Usage Guide (TN1261).

2.17. IEEE 1149.1-Compliant Boundary Scan Testability

All ECP5/ECP5-5G devices have boundary scan cells that are accessed through an IEEE 1149.1 compliant Test Access Port (TAP). This allows functional testing of the circuit board on which the device is mounted through a serial scan path that can access all critical logic nodes. Internal registers are linked internally, allowing test data to be shifted in and loaded directly onto test nodes, or test data to be captured and shifted out for verification. The test access port consists of dedicated I/Os: TDI, TDO, TCK and TMS. The test access port uses VCCIO8 for power supply.

For more information, refer to ECP5 and ECP5-5G sysCONFIG Usage Guide (TN1260).

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Table 3.20. Register-to-Register Performance

Function	–8 Timing	Unit
Basic Functions		
16-Bit Decoder	441	MHz
32-Bit Decoder	441	MHz
64-Bit Decoder	332	MHz
4:1 Mux	441	MHz
8:1 Mux	441	MHz
16:1 Mux	441	MHz
32:1 Mux	441	MHz
8-Bit Adder	441	MHz
16-Bit Adder	441	MHz
64-Bit Adder	441	MHz
16-Bit Counter	384	MHz
32-Bit Counter	317	MHz
64-Bit Counter	263	MHz
64-Bit Accumulator	288	MHz
Embedded Memory Functions		
1024x18 True-Dual Port RAM (Write Through or Normal), with EBR Output Registers	272	MHz
1024x18 True-Dual Port RAM (Read-Before-Write), with EBR Output Registers	214	MHz
Distributed Memory Functions		
16 x 2 Pseudo-Dual Port or 16 x 4 Single Port RAM (One PFU)	441	MHz
16 x 4 Pseudo-Dual Port (Two PFUs)	441	MHz
DSP Functions		
9 x 9 Multiplier (All Registers)	225	MHz
18 x 18 Multiplier (All Registers)	225	MHz
36 x 36 Multiplier (All Registers)	225	MHz
18 x 18 Multiply-Add/Sub (All Registers)	225	MHz
18 x 18 Multiply/Accumulate (Input and Output Registers)	225	MHz

Notes:

1. These functions were generated using Lattice Diamond design software tool. Exact performance may vary with the device and the design software tool version. The design software tool uses internal parameters that have been characterized but are not tested on every device.
2. Commercial timing numbers are shown. Industrial numbers are typically slower and can be extracted from Lattice Diamond design software tool.

3.16. Derating Timing Tables

Logic timing provided in the following sections of this data sheet and the Diamond design tools are worst case numbers in the operating range. Actual delays at nominal temperature and voltage for best case process, can be much better than the values given in the tables. The Diamond design tool can provide logic timing numbers at a particular temperature and voltage.

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3.18. External Switching Characteristics

Over recommended commercial operating conditions.

Table 3.22. ECP5/ECP5-5G External Switching Characteristics

Parameter	Description	Device	–8		–7		–6		Unit
			Min	Max	Min	Max	Min	Max	
Clocks									
Primary Clock									
f_{MAX_PRI}	Frequency for Primary Clock Tree	—	—	370	—	303	—	257	MHz
t_{W_PRI}	Clock Pulse Width for Primary Clock	—	0.8	—	0.9	—	1.0	—	ns
t_{SKEW_PRI}	Primary Clock Skew within a Device	—	—	420	—	462	—	505	ps
Edge Clock									
f_{MAX_EDGE}	Frequency for Edge Clock Tree	—	—	400	—	350	—	312	MHz
t_{W_EDGE}	Clock Pulse Width for Edge Clock	—	1.175	—	1.344	—	1.50	—	ns
t_{SKEW_EDGE}	Edge Clock Skew within a Bank	—	—	160	—	180	—	200	ps
Generic SDR Input									
General I/O Pin Parameters Using Dedicated Primary Clock Input without PLL									
t_{CO}	Clock to Output - PIO Output Register	All Devices	—	5.4	—	6.1	—	6.8	ns
t_{SU}	Clock to Data Setup - PIO Input Register	All Devices	0	—	0	—	0	—	ns
t_{H}	Clock to Data Hold - PIO Input Register	All Devices	2.7	—	3	—	3.3	—	ns
t_{SU_DEL}	Clock to Data Setup - PIO Input Register with Data Input Delay	All Devices	1.2	—	1.33	—	1.46	—	ns
t_{H_DEL}	Clock to Data Hold - PIO Input Register with Data Input Delay	All Devices	0	—	0	—	0	—	ns
f_{MAX_IO}	Clock Frequency of I/O and PFU Register	All Devices	—	400	—	350	—	312	MHz
General I/O Pin Parameters Using Dedicated Primary Clock Input with PLL									
t_{COPLL}	Clock to Output - PIO Output Register	All Devices	—	3.5	—	3.8	—	4.1	ns
t_{SUPLL}	Clock to Data Setup - PIO Input Register	All Devices	0.7	—	0.78	—	0.85	—	ns
t_{HPLL}	Clock to Data Hold - PIO Input Register	All Devices	0.8	—	0.89	—	0.98	—	ns
t_{SU_DELPLL}	Clock to Data Setup - PIO Input Register with Data Input Delay	All Devices	1.6	—	1.78	—	1.95	—	ns

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3.25. PCI Express Electrical and Timing Characteristics

3.25.1. PCIe (2.5 Gb/s) AC and DC Characteristics

Over recommended operating conditions.

Table 3.30. PCIe (2.5 Gb/s)

Symbol	Description	Test Conditions	Min	Typ	Max	Unit
Transmit[1]						
UI	Unit interval	—	399.88	400	400.12	ps
$V_{TX-DIFF\_P-P}$	Differential peak-to-peak output	—	0.8	1.0	1.2	V
$V_{TX-DE-RATIO}$	De-emphasis differential output voltage ratio	—	–3	–3.5	–4	dB
$V_{TX-CM-AC\_P}$	RMS AC peak common-mode output voltage	—	—	—	20	mV
$V_{TX-RCV-DETECT}$	Amount of voltage change allowed during receiver detection	—	—	—	600	mV
$V_{TX-CM-DC}$	Tx DC common mode voltage	—	0	—	V_{CCHTX}	V
$I_{TX-SHORT}$	Output short circuit current	V_{TX-D+}=0.0 V V_{TX-D-}=0.0 V	—	—	90	mA
$Z_{TX-DIFF-DC}$	Differential output impedance	—	80	100	120	Ω
$RL_{TX-DIFF}$	Differential return loss	—	10	—	—	dB
RL_{TX-CM}	Common mode return loss	—	6.0	—	—	dB
$T_{TX-RISE}$	Tx output rise time	20% to 80%	0.125	—	—	UI
$T_{TX-FALL}$	Tx output fall time	20% to 80%	0.125	—	—	UI
$L_{TX-SKEW}$	Lane-to-lane static output skew for all lanes in port/link	—	—	—	1.3	ns
T_{TX-EYE}	Transmitter eye width	—	0.75	—	—	UI
$T_{TX-EYE-MEDIAN-TO-MAX-JITTER}$	Maximum time between jitter median and maximum deviation from median	—	—	—	0.125	UI
Receive[1, 2]						
UI	Unit Interval	—	399.88	400	400.12	ps
$V_{RX-DIFF\_P-P}$	Differential peak-to-peak input voltage	—	0.34[3]	—	1.2	V
$V_{RX-IDLE-DET-DIFF\_P-P}$	Idle detect threshold voltage	—	65	—	340[3]	mV
$V_{RX-CM-AC\_P}$	RMS AC peak common-mode input voltage	—	—	—	150	mV
$Z_{RX-DIFF-DC}$	DC differential input impedance	—	80	100	120	Ω
Z_{RX-DC}	DC input impedance	—	40	50	60	Ω
$Z_{RX-HIGH-IMP-DC}$	Power-down DC input impedance	—	200K	—	—	Ω
$RL_{RX-DIFF}$	Differential return loss	—	10	—	—	dB
RL_{RX-CM}	Common mode return loss	—	6.0	—	—	dB

Notes:

1. Values are measured at 2.5 Gb/s.
2. Measured with external AC-coupling on the receiver.
3. Not in compliance with PCI Express 1.1 standard.

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3.26. CPRI LV2 E.48 Electrical and Timing Characteristics – Preliminary

Table 3.32. CPRI LV2 E.48 Electrical and Timing Characteristics

Symbol	Description	Test Conditions	Min	Typ	Max	Unit
Transmit						
UI	Unit Interval	—	203.43	203.45	203.47	ps
T_{DCD}	Duty Cycle Distortion	—	—	—	0.05	UI
J_{UBHPJ}	Uncorrelated Bounded High Probability Jitter	—	—	—	0.15	UI
J_{TOTAL}	Total Jitter	—	—	—	0.3	UI
$Z_{RX-DIFF-DC}$	DC differential Impedance	—	80	—	120	Ω
T_{SKEW}	Skew between differential signals	—	—	—	9	ps
$R_{LTX-DIFF}$	Tx Differential Return Loss (S22), including package and silicon	100 MHz < freq < 3.6864 GHz	—	—	–8	dB
		3.6864 GHz < freq < 4.9152 GHz	—	—	–8 + 16.6 *log (freq/3.6864)	dB
R_{LTX-CM}	Tx Common Mode Return Loss, including package and silicon	100 MHz < freq < 3.6864 GHz	6	—	—	dB
$I_{TX-SHORT}$	Transmitter short-circuit current	—	—	—	100	mA
$T_{RISE\_FALL-DIFF}$	Differential Rise and Fall Time	—		—	—	ps
$L_{TX-SKEW}$	Lane-to-lane output skew	—	—	—		ps
Receive						
UI	Unit Interval	—	203.43	203.45	203.47	ps
$V_{RX-DIFF-PP}$	Differential Rx peak-peak voltage	—	—	—	1.2	V, p-p
$V_{RX-EYE\_Y1\_Y2}$	Receiver eye opening mask, Y1 and Y2	—	62.5	—	375	mV, diff
V_{RX-EYE_X1}	Receiver eye opening mask, X1	—	—	—	0.3	UI
T_{RX-TJ}	Receiver total jitter tolerance (not including sinusoidal)	—	—	—	0.6	UI
$R_{LRX-DIFF}$	Receiver differential Return Loss, package plus silicon	100 MHz < freq < 3.6864 GHz	—	—	–8	dB
		3.6864 GHz < freq < 4.9152 GHz	—	—	–8 + 16.6 *log (freq/3.6864)	dB
R_{LRX-CM}	Receiver common mode Return Loss, package plus silicon	—	6	—	—	dB
$Z_{RX-DIFF-DC}$	Receiver DC differential impedance	—	80	100	120	Ω

Note: Data is measured with PRBS7 data pattern, not with PRBS-31 pattern.

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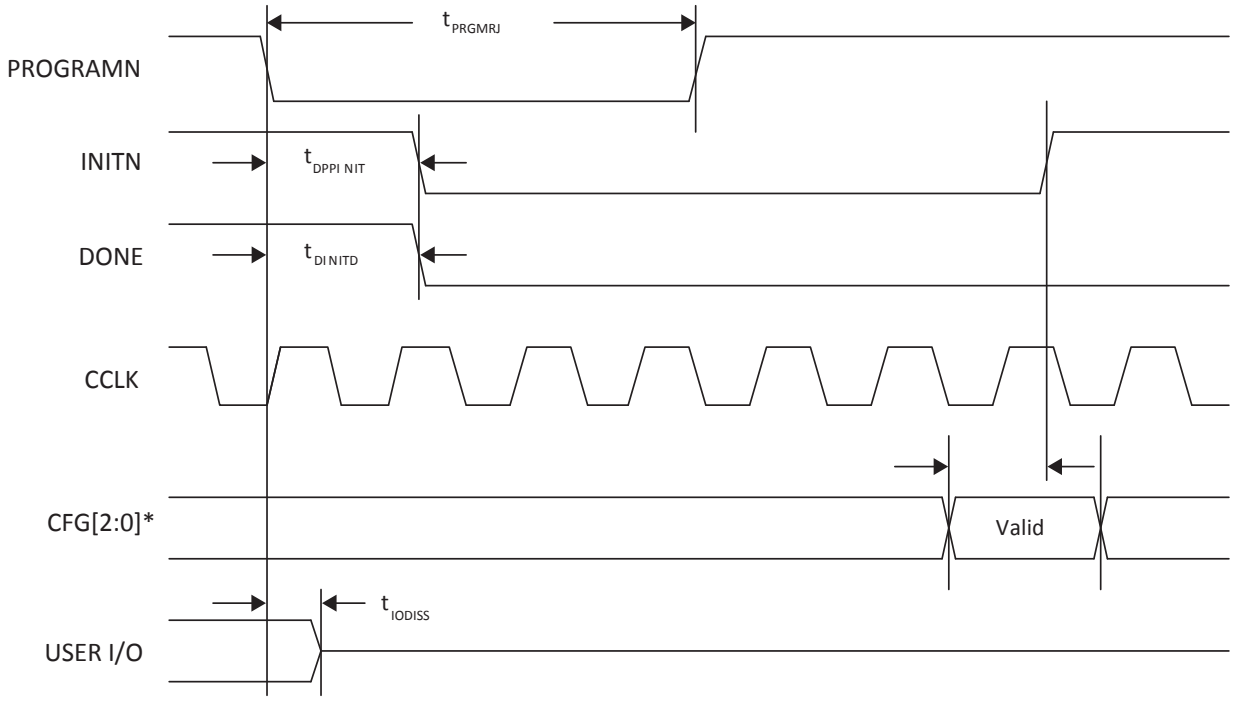

*The CFG pins are normally static (hardwired).

Figure 3.20. Configuration from PROGRAMN Timing

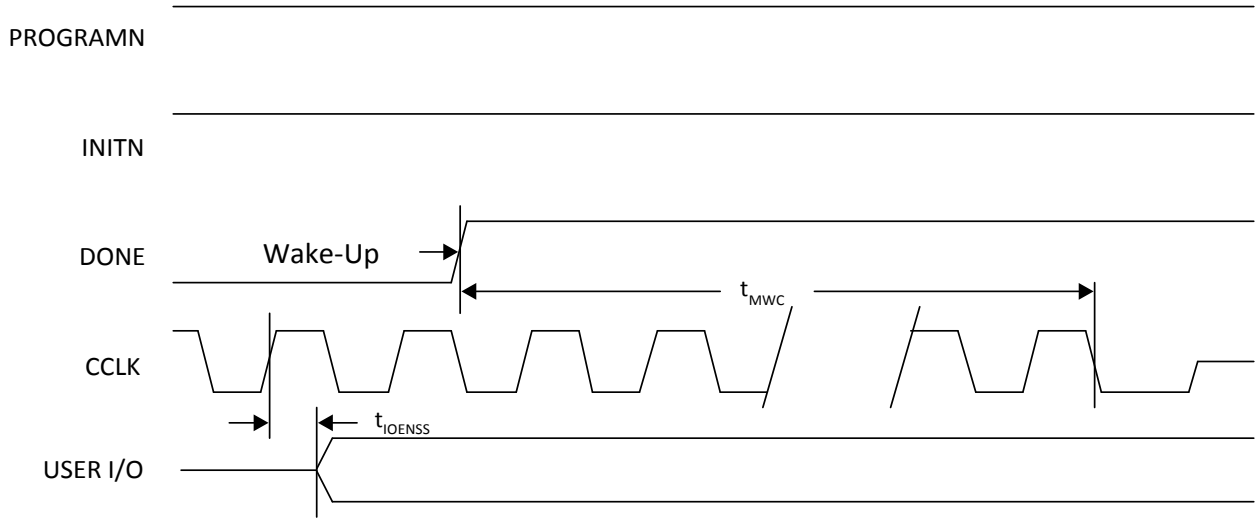

Figure 3.21. Wake-Up Timing

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Table 3.44. Test Fixture Required Components, Non-Terminated Interfaces

Test Condition	R_1	R_2	C_L	Timing Ref.	V_T
LVTTL and other LVCMOS settings (L ≥ H, H ≥ L)	∞	∞	0 pF	LVCMOS 3.3 = 1.5 V	—
				LVCMOS 2.5 = $V_{CCIO}/2$	—
				LVCMOS 1.8 = $V_{CCIO}/2$	—
				LVCMOS 1.5 = $V_{CCIO}/2$	—
				LVCMOS 1.2 = $V_{CCIO}/2$	—
LVCMOS 2.5 I/O (Z ≥ H)	∞	1 MΩ	0 pF	$V_{CCIO}/2$	—
LVCMOS 2.5 I/O (Z ≥ L)	1 MΩ	∞	0 pF	$V_{CCIO}/2$	V_{CCIO}
LVCMOS 2.5 I/O (H ≥ Z)	∞	100	0 pF	V_{OH} – 0.10	—
LVCMOS 2.5 I/O (L ≥ Z)	100	∞	0 pF	V_{OL} + 0.10	V_{CCIO}

Note: Output test conditions for all other interfaces are determined by the respective standards.

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Part number	Grade	Package	Pins	Temp.	LUTs (K)	SERDES
LFE5U-45F-8BG554C	–8	Lead free caBGA	554	Commercial	44	No
LFE5U-85F-6MG285C	–6	Lead free csfBGA	285	Commercial	84	No
LFE5U-85F-7MG285C	–7	Lead free csfBGA	285	Commercial	84	No
LFE5U-85F-8MG285C	–8	Lead free csfBGA	285	Commercial	84	No
LFE5U-85F-6BG381C	–6	Lead free caBGA	381	Commercial	84	No
LFE5U-85F-7BG381C	–7	Lead free caBGA	381	Commercial	84	No
LFE5U-85F-8BG381C	–8	Lead free caBGA	381	Commercial	84	No
LFE5U-85F-6BG554C	–6	Lead free caBGA	554	Commercial	84	No
LFE5U-85F-7BG554C	–7	Lead free caBGA	554	Commercial	84	No
LFE5U-85F-8BG554C	–8	Lead free caBGA	554	Commercial	84	No
LFE5U-85F-6BG756C	–6	Lead free caBGA	756	Commercial	84	No
LFE5U-85F-7BG756C	–7	Lead free caBGA	756	Commercial	84	No
LFE5U-85F-8BG756C	–8	Lead free caBGA	756	Commercial	84	No
LFE5UM-25F-6MG285C	–6	Lead free csfBGA	285	Commercial	24	Yes
LFE5UM-25F-7MG285C	–7	Lead free csfBGA	285	Commercial	24	Yes
LFE5UM-25F-8MG285C	–8	Lead free csfBGA	285	Commercial	24	Yes
LFE5UM-25F-6BG381C	–6	Lead free caBGA	381	Commercial	24	Yes
LFE5UM-25F-7BG381C	–7	Lead free caBGA	381	Commercial	24	Yes
LFE5UM-25F-8BG381C	–8	Lead free caBGA	381	Commercial	24	Yes
LFE5UM-45F-6MG285C	–6	Lead free csfBGA	285	Commercial	44	Yes
LFE5UM-45F-7MG285C	–7	Lead free csfBGA	285	Commercial	44	Yes
LFE5UM-45F-8MG285C	–8	Lead free csfBGA	285	Commercial	44	Yes
LFE5UM-45F-6BG381C	–6	Lead free caBGA	381	Commercial	44	Yes
LFE5UM-45F-7BG381C	–7	Lead free caBGA	381	Commercial	44	Yes
LFE5UM-45F-8BG381C	–8	Lead free caBGA	381	Commercial	44	Yes
LFE5UM-45F-6BG554C	–6	Lead free caBGA	554	Commercial	44	Yes
LFE5UM-45F-7BG554C	–7	Lead free caBGA	554	Commercial	44	Yes
LFE5UM-45F-8BG554C	–8	Lead free caBGA	554	Commercial	44	Yes
LFE5UM-85F-6MG285C	–6	Lead free csfBGA	285	Commercial	84	Yes
LFE5UM-85F-7MG285C	–7	Lead free csfBGA	285	Commercial	84	Yes
LFE5UM-85F-8MG285C	–8	Lead free csfBGA	285	Commercial	84	Yes
LFE5UM-85F-6BG381C	–6	Lead free caBGA	381	Commercial	84	Yes
LFE5UM-85F-7BG381C	–7	Lead free caBGA	381	Commercial	84	Yes
LFE5UM-85F-8BG381C	–8	Lead free caBGA	381	Commercial	84	Yes
LFE5UM-85F-6BG554C	–6	Lead free caBGA	554	Commercial	84	Yes
LFE5UM-85F-7BG554C	–7	Lead free caBGA	554	Commercial	84	Yes
LFE5UM-85F-8BG554C	–8	Lead free caBGA	554	Commercial	84	Yes
LFE5UM-85F-6BG756C	–6	Lead free caBGA	756	Commercial	84	Yes
LFE5UM-85F-7BG756C	–7	Lead free caBGA	756	Commercial	84	Yes
LFE5UM-85F-8BG756C	–8	Lead free caBGA	756	Commercial	84	Yes
LFE5UM5G-25F-8MG285C	–8	Lead free csfBGA	285	Commercial	24	Yes
LFE5UM5G-25F-8BG381C	–8	Lead free caBGA	381	Commercial	24	Yes
LFE5UM5G-45F-8MG285C	–8	Lead free csfBGA	285	Commercial	44	Yes
LFE5UM5G-45F-8BG381C	–8	Lead free caBGA	381	Commercial	44	Yes
LFE5UM5G-45F-8BG554C	–8	Lead free caBGA	554	Commercial	44	Yes
LFE5UM5G-85F-8MG285C	–8	Lead free csfBGA	285	Commercial	84	Yes

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Part number	Grade	Package	Pins	Temp.	LUTs (K)	SERDES
LFE5U-85F-6BG756I	–6	Lead free caBGA	756	Industrial	84	No
LFE5U-85F-7BG756I	–7	Lead free caBGA	756	Industrial	84	No
LFE5U-85F-8BG756I	–8	Lead free caBGA	756	Industrial	84	No
LFE5UM-25F-6MG285I	–6	Lead free csfBGA	285	Industrial	24	Yes
LFE5UM-25F-7MG285I	–7	Lead free csfBGA	285	Industrial	24	Yes
LFE5UM-25F-8MG285I	–8	Lead free csfBGA	285	Industrial	24	Yes
LFE5UM-25F-6BG381I	–6	Lead free caBGA	381	Industrial	24	Yes
LFE5UM-25F-7BG381I	–7	Lead free caBGA	381	Industrial	24	Yes
LFE5UM-25F-8BG381I	–8	Lead free caBGA	381	Industrial	24	Yes
LFE5UM-45F-6MG285I	–6	Lead free csfBGA	285	Industrial	44	Yes
LFE5UM-45F-7MG285I	–7	Lead free csfBGA	285	Industrial	44	Yes
LFE5UM-45F-8MG285I	–8	Lead free csfBGA	285	Industrial	44	Yes
LFE5UM-45F-6BG381I	–6	Lead free caBGA	381	Industrial	44	Yes
LFE5UM-45F-7BG381I	–7	Lead free caBGA	381	Industrial	44	Yes
LFE5UM-45F-8BG381I	–8	Lead free caBGA	381	Industrial	44	Yes
LFE5UM-45F-6BG554I	–6	Lead free caBGA	554	Industrial	44	Yes
LFE5UM-45F-7BG554I	–7	Lead free caBGA	554	Industrial	44	Yes
LFE5UM-45F-8BG554I	–8	Lead free caBGA	554	Industrial	44	Yes
LFE5UM-85F-6MG285I	–6	Lead free csfBGA	285	Industrial	84	Yes
LFE5UM-85F-7MG285I	–7	Lead free csfBGA	285	Industrial	84	Yes
LFE5UM-85F-8MG285I	–8	Lead free csfBGA	285	Industrial	84	Yes
LFE5UM-85F-6BG381I	–6	Lead free caBGA	381	Industrial	84	Yes
LFE5UM-85F-7BG381I	–7	Lead free caBGA	381	Industrial	84	Yes
LFE5UM-85F-8BG381I	–8	Lead free caBGA	381	Industrial	84	Yes
LFE5UM-85F-6BG554I	–6	Lead free caBGA	554	Industrial	84	Yes
LFE5UM-85F-7BG554I	–7	Lead free caBGA	554	Industrial	84	Yes
LFE5UM-85F-8BG554I	–8	Lead free caBGA	554	Industrial	84	Yes
LFE5UM-85F-6BG756I	–6	Lead free caBGA	756	Industrial	84	Yes
LFE5UM-85F-7BG756I	–7	Lead free caBGA	756	Industrial	84	Yes
LFE5UM-85F-8BG756I	–8	Lead free caBGA	756	Industrial	84	Yes
LFE5UM5G-25F-8MG285I	–8	Lead free csfBGA	285	Industrial	24	Yes
LFE5UM5G-25F-8BG381I	–8	Lead free caBGA	381	Industrial	24	Yes
LFE5UM5G-45F-8MG285I	–8	Lead free csfBGA	285	Industrial	44	Yes
LFE5UM5G-45F-8BG381I	–8	Lead free caBGA	381	Industrial	44	Yes
LFE5UM5G-45F-8BG554I	–8	Lead free caBGA	554	Industrial	44	Yes
LFE5UM5G-85F-8MG285I	–8	Lead free csfBGA	285	Industrial	84	Yes
LFE5UM5G-85F-8BG381I	–8	Lead free caBGA	381	Industrial	84	Yes
LFE5UM5G-85F-8BG554I	–8	Lead free caBGA	554	Industrial	84	Yes
LFE5UM5G-85F-8BG756I	–8	Lead free caBGA	756	Industrial	84	Yes

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Supplemental Information

For Further Information

A variety of technical notes for the ECP5/ECP5-5G family are available.

- High-Speed PCB Design Considerations (TN1033)
- Transmission of High-Speed Serial Signals Over Common Cable Media (TN1066)
- PCB Layout Recommendations for BGA Packages (TN1074)
- Minimizing System Interruption During Configuration Using TransFR Technology (TN1087)
- Electrical Recommendations for Lattice SERDES (FPGA-TN-02077)
- LatticeECP3, ECP-5 and ECP5-5G Soft Error Detection (SED)/Correction (SEC) Usage Guide (TN1184)
- Using TraceID (TN1207)
- Sub-LVDS Signaling Using Lattice Devices (TN1210)
- Advanced Security Encryption Key Programming Guide for ECP5, ECP5-5G, LatticeECP3, and LatticeECP2/MS Devices (TN1215)
- LatticeECP3, LatticeECP2/M, ECP5 and ECP5-5G Dual Boot and Multiple Boot Feature (TN1216)
- ECP5 and ECP5-5G sysCONFIG Usage Guide (TN1260)
- ECP5 and ECP5-5G SERDES/PCS Usage Guide (TN1261)
- ECP5 and ECP5-5G sysIO Usage Guide (TN1262)
- ECP5 and ECP5-5G sysClock PLL/DLL Design and Usage Guide (TN1263)
- ECP5 and ECP5-5G Memory Usage Guide (TN1264)
- ECP5 and ECP5-5G High-Speed I/O Interface (TN1265)
- Power Consumption and Management for ECP5 and ECP5-5G Devices (TN1266)
- ECP5 and ECP5-5G sysDSP Usage Guide (TN1267)
- ECP5 and ECP5-5G Hardware Checklist (FPGA-TN-02038)
- Solder Reflow Guide for Surface Mount Devices (FPGA-TN-02041)
- ECP5 and ECP5-5G PCI Express Soft IP Ease of Use Guidelines (FPGA-TN-02045)
- Programming External SPI Flash through JTAG for ECP5/ECP5-5G (FPGA-TN-02050)
- Adding Scalable Power and Thermal Management to ECP5 Using L-ASC10 (AN6095)

For further information on interface standards refer to the following websites:

- JEDEC Standards (LVTTL, LVCMOS, SSTL): www.jedec.org
- PCI: www.pcisig.com

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