Lattice Semiconductor Corporation - LFE5U-45F-8BG256C Datasheet

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Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Active
Number of LABs/CLBs	11000
Number of Logic Elements/Cells	44000
Total RAM Bits	1990656
Number of I/O	197
Number of Gates	-
Voltage - Supply	1.045V ~ 1.155V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	256-LFBGA
Supplier Device Package	256-CABGA (14x14)
Purchase URL	https://www.e-xfl.com/product-detail/lattice-semiconductor/lfe5u-45f-8bg256c

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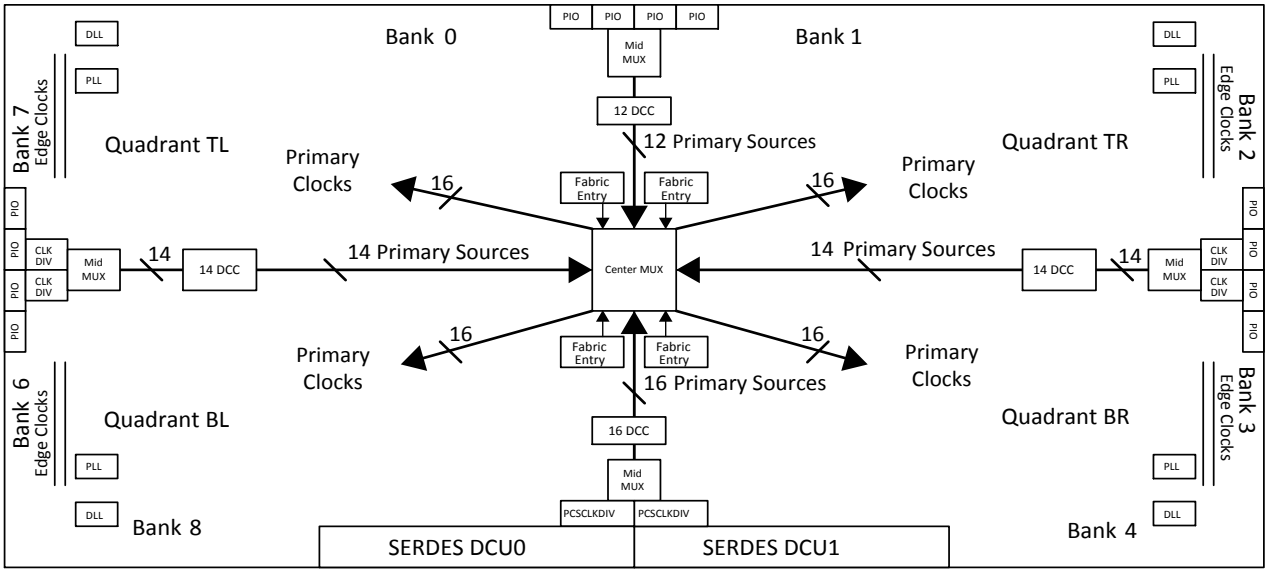

Figure 2.6. LFE5UM/LFE5UM5G-85 Clocking

2.5.1. Primary Clocks

The ECP5/ECP5-5G device family provides low-skew, high fan-out clock distribution to all synchronous elements in the FPGA fabric through the Primary Clock Network.

The primary clock network is divided into four clocking quadrants: Top Left (TL), Bottom Left (BL), Top Right (TR), and Bottom Right (BR). Each of these quadrants has 16 clocks that can be distributed to the fabric in the quadrant.

The Lattice Diamond software can automatically route each clock to one of the four quadrants up to a maximum of 16 clocks per quadrant. The user can change how the clocks are routed by specifying a preference in the Lattice Diamond software to locate the clock to specific. The ECP5/ECP5-5G device provides the user with a maximum of 64 unique clock input sources that can be routed to the primary Clock network.

Primary clock sources are:

- Dedicated clock input pins
- PLL outputs
- CLKDIV outputs
- Internal FPGA fabric entries (with minimum general routing)
- SERDES/PCS/PCSDIV clocks
- OSC clock

These sources are routed to one of four clock switches called a Mid Mux. The outputs of the Mid MUX are routed to the center of the FPGA where another clock switch, called the Center MUX, is used to route the primary clock sources to primary clock distribution to the ECP5/ECP5-5G fabric. These routing muxes are shown in Figure 2.6. Since there is a maximum of 60 unique clock input sources to the clocking quadrants, there are potentially 64 unique clock domains that can be used in the ECP5/ECP5-5G Device. For more information about the primary clock tree and connections, refer to ECP5 and ECP5-5G sysClock PLL/DLL Design and Usage Guide (TN1263).

2.5.1.1. Dynamic Clock Control

The Dynamic Clock Control (DCC), Quadrant Clock enable/disable feature allows internal logic control of the quadrant primary clock network. When a clock network is disabled, the clock signal is static and not toggle. All the logic fed by that clock will not toggle, reducing the overall power consumption of the device. The disable function will not create glitch and increase the clock latency to the primary clock network.

This DCC controls the clock sources from the Primary CLOCK MIDMUX before they are fed to the Primary Center MUXs that drive the quadrant clock network. For more information about the DCC, refer to ECP5 and ECP5-5G sysClock PLL/DLL Design and Usage Guide (TN1263).

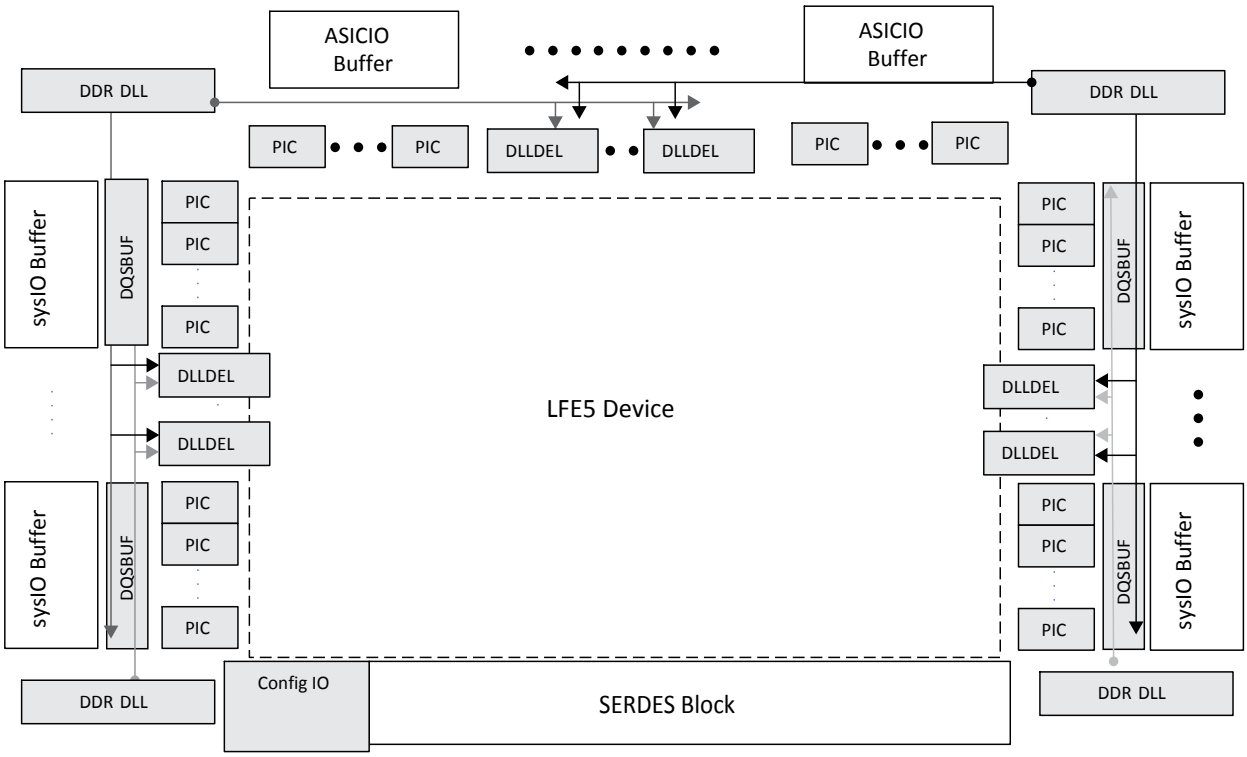

Figure 2.11. ECP5/ECP5-5G DLL Top Level View (For LFE-45 and LFE-85)

2.8. sysMEM Memory

ECP5/ECP5-5G devices contain a number of sysMEM Embedded Block RAM (EBR). The EBR consists of an 18 Kb RAM with memory core, dedicated input registers and output registers with separate clock and clock enable. Each EBR includes functionality to support true dual-port, pseudo dual-port, single-port RAM, ROM and FIFO buffers (via external PFUs).

2.8.1. sysMEM Memory Block

The sysMEM block can implement single port, dual port or pseudo dual port memories. Each block can be used in a variety of depths and widths as listed in Table 2.6 on page 25. FIFOs can be implemented in sysMEM EBR blocks by implementing support logic with PFUs. The EBR block facilitates parity checking by supporting an optional parity bit for each data byte. EBR blocks provide byte-enable support for configurations with 18-bit and 36-bit data widths. For more information, refer to ECP5 and ECP5-5G Memory Usage Guide (TN1264).

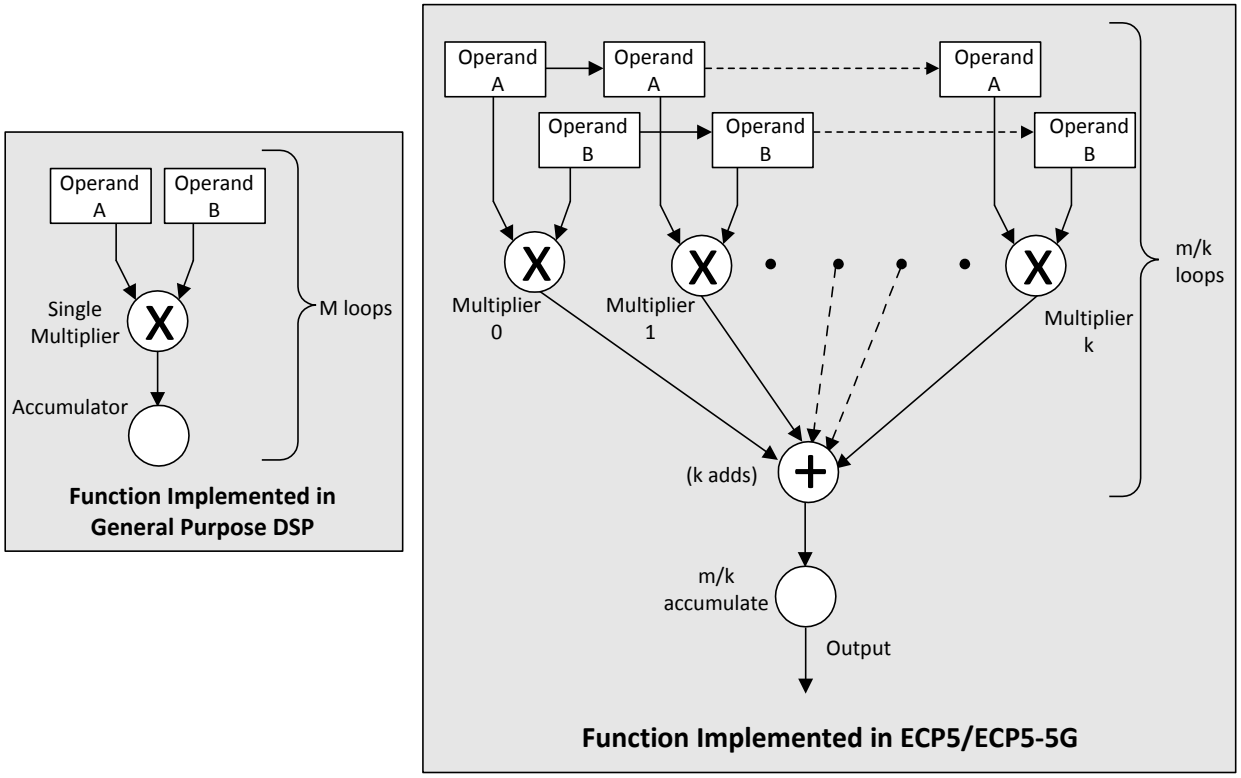

Figure 2.13. Comparison of General DSP and ECP5/ECP5-5G Approaches

2.9.2. sysDSP Slice Architecture Features

The ECP5/ECP5-5G sysDSP Slice has been significantly enhanced to provide functions needed for advanced processing applications. These enhancements provide improved flexibility and resource utilization.

The ECP5/ECP5-5G sysDSP Slice supports many functions that include the following:

- Symmetry support. The primary target application is wireless. 1D Symmetry is useful for many applications that use FIR filters when their coefficients have symmetry or asymmetry characteristics. The main motivation for using 1D symmetry is cost/size optimization. The expected size reduction is up to 2x.
  - Odd mode – Filter with Odd number of taps
  - Even mode – Filter with Even number of taps
  - Two dimensional (2D) symmetry mode – supports 2D filters for mainly video applications
- Dual-multiplier architecture. Lower accumulator overhead to half and the latency to half compared to single multiplier architecture
- Fully cascadable DSP across slices. Support for symmetric, asymmetric and non-symmetric filters.
- Multiply (one 18x36, two 18x18 or four 9x9 Multiplies per Slice)
- Multiply (36x36 by cascading across two sysDSP slices)
- Multiply Accumulate (supports one 18x36 multiplier result accumulation or two 18x18 multiplier result accumulation)
- Two Multiplies feeding one Accumulate per cycle for increased processing with lower latency (two 18x18 Multiplies feed into an accumulator that can accumulate up to 52 bits)
- Pipeline registers
- 1D Symmetry support. The coefficients of FIR filters have symmetry or negative symmetry characteristics.
  - Odd mode – Filter with Odd number of taps
  - Even mode – Filter with Even number of taps
- 2D Symmetry support. The coefficients of 2D FIR filters have symmetry or negative symmetry characteristics.
  - 3*3 and 3*5 – Internal DSP Slice support

- 5*5 and larger size 2D blocks – Semi internal DSP Slice support
- Flexible saturation and rounding options to satisfy a diverse set of applications situations
- Flexible cascading across DSP slices
  - Minimizes fabric use for common DSP and ALU functions
  - Enables implementation of FIR Filter or similar structures using dedicated sysDSP slice resources only
  - Provides matching pipeline registers
  - Can be configured to continue cascading from one row of sysDSP slices to another for longer cascade chains
- Flexible and Powerful Arithmetic Logic Unit (ALU) Supports:
  - Dynamically selectable ALU OPCODE
  - Ternary arithmetic (addition/subtraction of three inputs)
  - Bit-wise two-input logic operations (AND, OR, NAND, NOR, XOR and XNOR)
  - Eight flexible and programmable ALU flags that can be used for multiple pattern detection scenarios, such as, overflow, underflow and convergent rounding.
  - Flexible cascading across slices to get larger functions
- RTL Synthesis friendly synchronous reset on all registers, while still supporting asynchronous reset for legacy users
- Dynamic MUX selection to allow Time Division Multiplexing (TDM) of resources for applications that require processor-like flexibility that enables different functions for each clock cycle

For most cases, as shown in Figure 2.14, the ECP5/ECP5-5G sysDSP slice is backwards-compatible with the LatticeECP2™ and LatticeECP3™ sysDSP block, such that, legacy applications can be targeted to the ECP5/ ECP5-5G sysDSP slice. Figure 2.14 shows the diagram of sysDSP, and Figure 2.15 shows the detailed diagram.

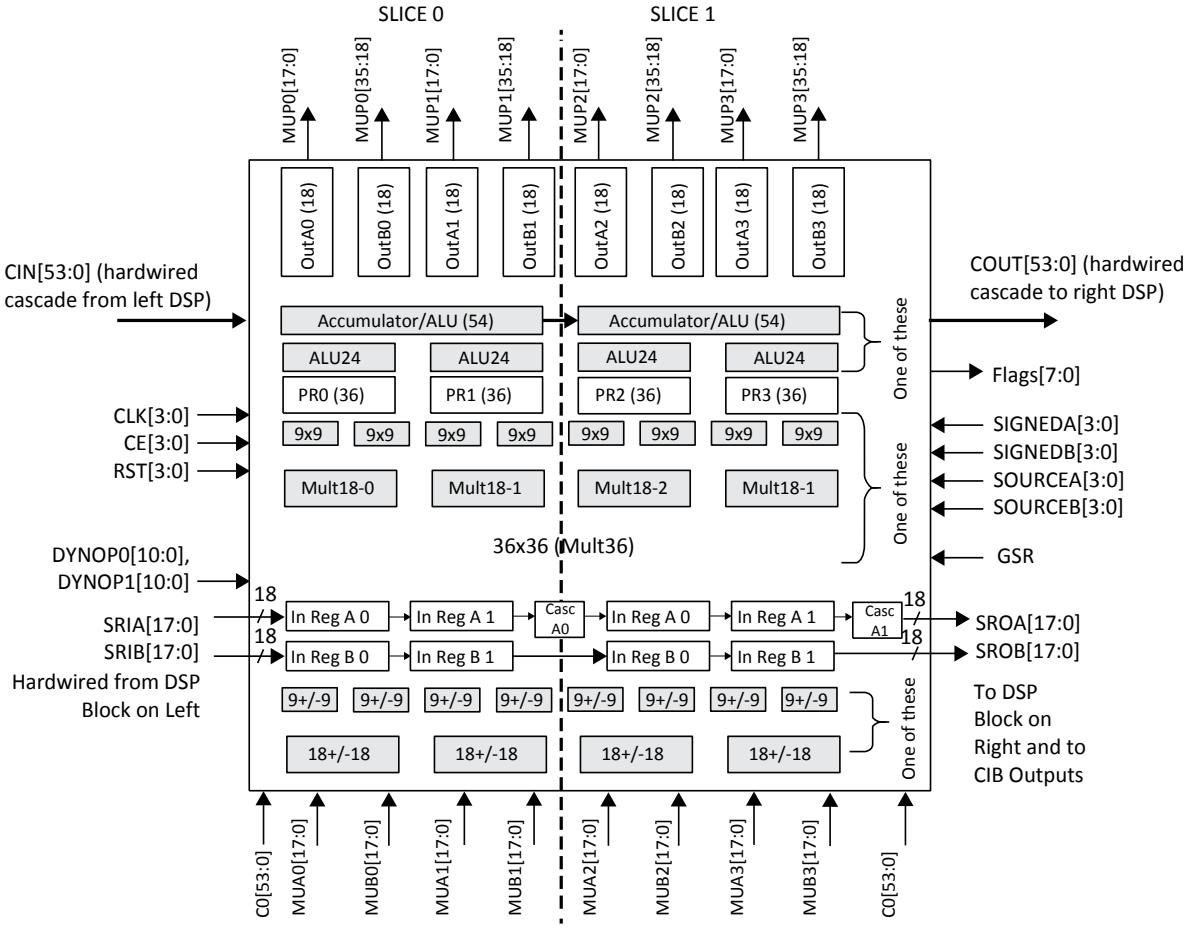

Figure 2.14. Simplified sysDSP Slice Block Diagram

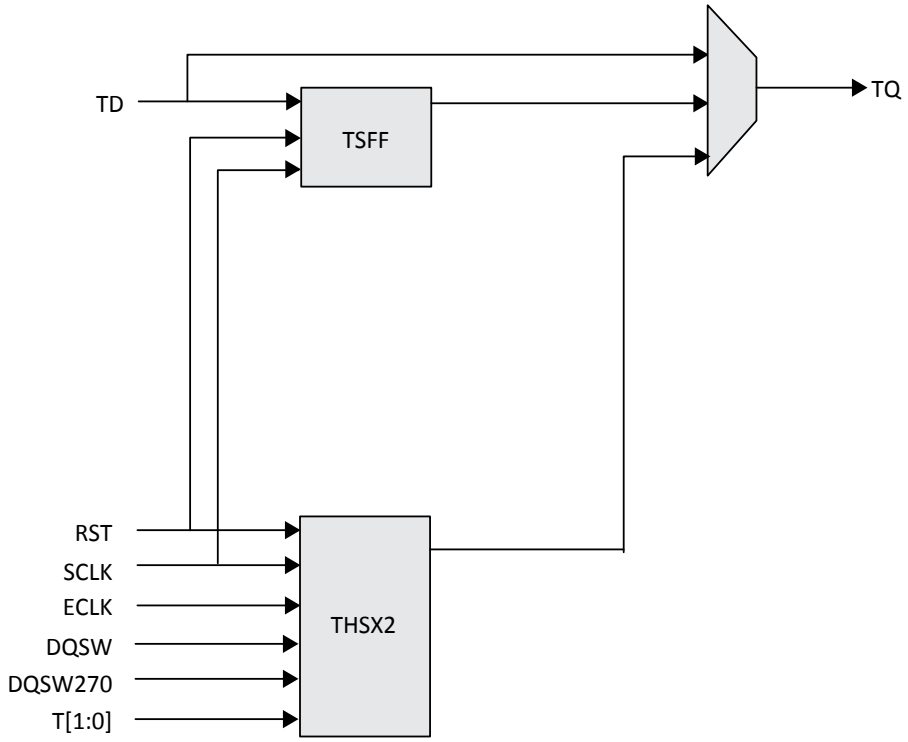

Figure 2.22. Tristate Register Block on Left and Right Sides

Table 2.10. Tristate Block Port Description

Name	Type	Description
TD	Input	Tristate Input to Tristate SDR Register
RST	Input	Reset to the Tristate Block
TD[1:0]	Input	Tristate input to TSHX2 function
SCLK	Input	Slow Speed System Clock
ECLK	Input	High Speed Edge Clock
DQSW	Input	Clock from DQS control Block used to generate DDR memory DQS output
DQSW270	Input	Clock from DQS control Block used to generate DDR memory DQ output
TQ	Output	Output of the Tristate block

2.13. DDR Memory Support

2.13.1. DQS Grouping for DDR Memory

Certain PICs have additional circuitry to allow the implementation of high-speed source synchronous and DDR2, DDR3, LPDDR2 or LPDDR3 memory interfaces. The support varies by the edge of the device as detailed below.

The left and right sides of the PIC have fully functional elements supporting DDR2, DDR3, LPDDR2 or LPDDR3 memory interfaces. Every 16 PIOs on the left and right sides are grouped into one DQS group, as shown in Figure 2.23 on page 36. Within each DQS group, there are two pre-placed pins for DQS and DQS# signals. The rest of the pins in the DQS group can be used as DQ signals and DM signal. The number of pins in each DQS group bonded out is package dependent. DQS groups with less than 11 pins bonded out can only be used for LPDDR2/3 Command/ Address busses. In DQS groups with more than 11 pins bonded out, up to two pre-defined pins are assigned to be used as "virtual" VCCIO, by driving these pins to HIGH, with the user connecting these pins to VCCIO power supply. These connections create "soft" connections to V_{CCIO} thru these output pins, and make better connections on VCCIO to help to reduce SSO noise. For details, refer to ECP5 and ECP5-5G High-Speed I/O Interface (TN1265).

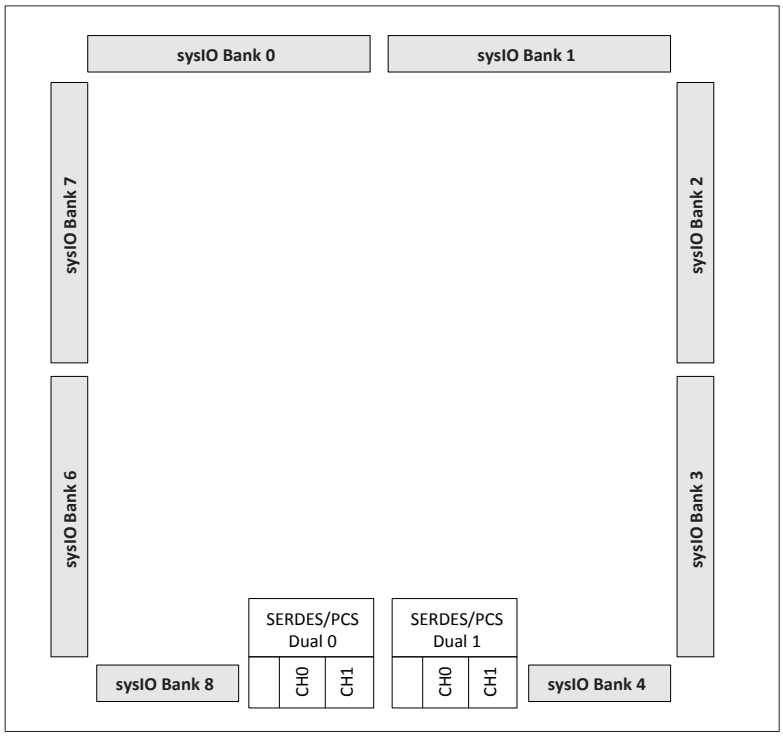

Figure 2.27. SERDES/PCS Duals (LFE5UM/LFE5UM5G-85)

Table 2.13. LFE5UM/LFE5UM5G SERDES Standard Support

Standard	Data Rate (Mb/s)	Number of General/Link Width	Encoding Style
PCI Express 1.1 and 2.0 2.02	2500	x1, x2, x4	8b10b
	5000 [2]	x1, x2	8b10b
Gigabit Ethernet	1250	x1	8b10b
SGMII	1250	x1	8b10b
	2500	x1	8b10b
XAUI	3125	x4	8b10b
CPRI-1 CPRI-2 CPRI-3 CPRI-4 CPRI-5	614.4 1228.8 2457.6 3072.0 4915.2 [2]	x1	8b10b
SD-SDI (259M, 344M) [1]	270	x1	NRZI/Scrambled
HD-SDI (292M)	1483.5 1485	x1	NRZI/Scrambled
3G-SDI (424M)	2967 2970	x1	NRZI/Scrambled
	5000	—	—
JESD204A/B	3125	x1	8b/10b

Notes:

1. For SD-SDI rate, the SERDES is bypassed and SERDES input signals are directly connected to the FPGA routing.
2. For ECP5-5G family devices only.

Table 2.14. Available SERDES Duals per LFE5UM/LFE5UM5G Devices

Package	LFE5UM/LFE5UM5G-25	LFE5UM/LFE5UM5G-45	LFE5UM/LFE5UM5G-85
285 csfBGA	1	1	1
381 caBGA	1	2	2
554 caBGA	—	2	2
756 caBGA	—	—	2

2.15.1. SERDES Block

A SERDES receiver channel may receive the serial differential data stream, equalize the signal, perform Clock and Data Recovery (CDR) and de-serialize the data stream before passing the 8- or 10-bit data to the PCS logic. The SERDES transmitter channel may receive the parallel 8- or 10-bit data, serialize the data and transmit the serial bit stream through the differential drivers. Figure 2.28 shows a single-channel SERDES/PCS block. Each SERDES channel provides a recovered clock and a SERDES transmit clock to the PCS block and to the FPGA core logic.

Each transmit channel, receiver channel, and SERDES PLL shares the same power supply (VCCA). The output and input buffers of each channel have their own independent power supplies (VCCHTX and VCCHRX).

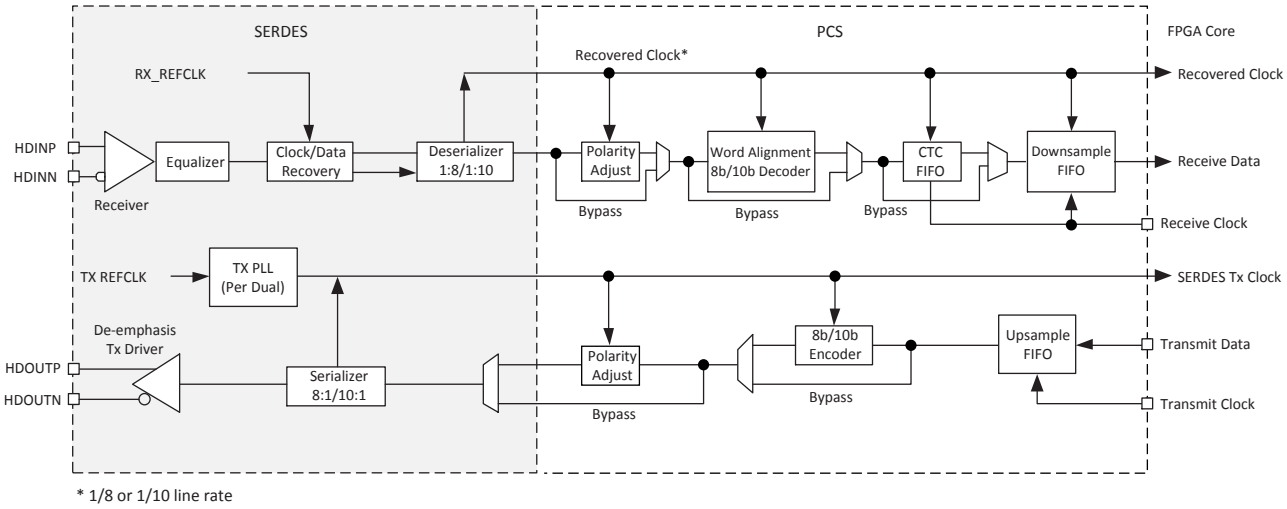

Figure 2.28. Simplified Channel Block Diagram for SERDES/PCS Block

2.15.2. PCS

As shown in Figure 2.28, the PCS receives the parallel digital data from the deserializer and selects the polarity, performs word alignment, decodes (8b/10b), provides Clock Tolerance Compensation and transfers the clock domain from the recovered clock to the FPGA clock via the Down Sample FIFO.

For the transmit channel, the PCS block receives the parallel data from the FPGA core, encodes it with 8b/10b, selects the polarity and passes the 8/10 bit data to the transmit SERDES channel.

The PCS also provides bypass modes that allow a direct 8-bit or 10-bit interface from the SERDES to the FPGA logic. The PCS interface to the FPGA can also be programmed to run at 1/2 speed for a 16-bit or 20-bit interface to the FPGA logic.

Some of the enhancements in LFE5UM/LFE5UM5G SERDES/PCS include:

- Higher clock/channel granularity: Dual channel architecture provides more clock resource per channel.
- Enhanced Tx de-emphasis: Programmable pre- and post-cursors improves Tx output signaling
- Bit-slip function in PCS: Improves logic needed to perform Word Alignment function

Refer to ECP5 and ECP5-5G SERDES/PCS Usage Guide (TN1261) for more information.

When an error is detected, and the user's error handling software determines the error did not create any risk to the system operation, the SEC tool allows the device to be re-configured in the background to correct the affected bit. This operation allows the user functions to continue to operate without stopping the system function.

Additional SEI tool is also available in the Diamond Software, by creating a frame of data to be programmed into the device in the background with one bit changed, without stopping the user functions on the device. This emulates an SEU situation, allowing the user to test and monitor its error handling software.

For further information on SED support, refer to LatticeECP3, ECP5 and ECP5-5G Soft Error Detection (SED)/Correction (SEC) Usage Guide (TN1184).

2.18.3. On-Chip Oscillator

Every ECP5/ECP5-5G device has an internal CMOS oscillator which is used to derive a Master Clock (MCLK) for configuration. The oscillator and the MCLK run continuously and are available to user logic after configuration is completed. The software default value of the MCLK is nominally 2.4 MHz. Table 2.16 lists all the available MCLK frequencies. When a different Master Clock is selected during the design process, the following sequence takes place:

1. Device powers up with a nominal Master Clock frequency of 2.4 MHz.
2. During configuration, users select a different master clock frequency.
3. The Master Clock frequency changes to the selected frequency once the clock configuration bits are received.
4. If the user does not select a master clock frequency, then the configuration bitstream defaults to the MCLK frequency of 2.4 MHz.

This internal oscillator is available to the user by routing it as an input clock to the clock tree. For further information on the use of this oscillator for configuration or user mode, refer to ECP5 and ECP5-5G sysCONFIG Usage Guide (TN1260) and ECP5 and ECP5-5G sysClock PLL/DLL Design and Usage Guide (TN1263).

Table 2.16. Selectable Master Clock (MCLK) Frequencies during Configuration (Nominal)

MCLK Frequency (MHz)
2.4
4.8
9.7
19.4
38.8
62

2.19. Density Shifting

The ECP5/ECP5-5G family is designed to ensure that different density devices in the same family and in the same package have the same pinout. Furthermore, the architecture ensures a high success rate when performing design migration from lower density devices to higher density devices. In many cases, it is also possible to shift a lower utilization design targeted for a high-density device to a lower density device. However, the exact details of the final resource utilization will impact the likelihood of success in each case. An example is that some user I/Os may become No Connects in smaller devices in the same package. Refer to the ECP5/ECP5-5G Pin Migration Tables and Diamond software for specific restrictions and limitations.

3.13. sysI/O Single-Ended DC Electrical Characteristics

Table 3.12. Single-Ended DC Characteristics

Input/Output Standard	V_{IL}		V_{IH}		V_{OL} Max (V)	V_{OH} Min (V)	I_{OL}[1] (mA)	I_{OH}[1] (mA)
	Min (V)	Max (V)	Min (V)	Max (V)				
LVCMOS33	–0.3	0.8	2.0	3.465	0.4	V_{CCIO} – 0.4	16, 12, 8, 4	–16, –12, –8, –4
LVCMOS25	–0.3	0.7	1.7	3.465	0.4	V_{CCIO} – 0.4	12, 8, 4	–12, –8, –4
LVCMOS18	–0.3	0.35 V_{CCIO}	0.65 V_{CCIO}	3.465	0.4	V_{CCIO} – 0.4	12, 8, 4	–12, –8, –4
LVCMOS15	–0.3	0.35 V_{CCIO}	0.65 V_{CCIO}	3.465	0.4	V_{CCIO} – 0.4	8, 4	–8, –4
LVCMOS12	–0.3	0.35 V_{CCIO}	0.65 V_{CCIO}	3.465	0.4	V_{CCIO} – 0.4	8, 4	–8, –4
LVTTL33	–0.3	0.8	2.0	3.465	0.4	V_{CCIO} – 0.4	16, 12, 8, 4	–16, –12, –8, –4
SSTL18_I (DDR2 Memory)	–0.3	V_{REF} – 0.125	V_{REF} + 0.125	3.465	0.4	V_{CCIO} – 0.4	6.7	–6.7
SSTL18_II	–0.3	V_{REF} –	V_{REF} + 0.125	3.465	0.28	V_{CCIO} – 0.28	13.4	–13.4
SSTL15 _I (DDR3 Memory)	–0.3	V_{REF} – 0.1	V_{REF} + 0.1	3.465	0.31	V_{CCIO} – 0.31	7.5	–7.5
SSTL15_II (DDR3 Memory)	–0.3	V_{REF} – 0.1	V_{REF} + 0.1	3.465	0.31	V_{CCIO} – 0.31	8.8	–8.8
SSTL135_I (DDR3L Memory)	–0.3	V_{REF} – 0.09	V_{REF} + 0.09	3.465	0.27	V_{CCIO} – 0.27	7	–7
SSTL135_II (DDR3L Memory)	–0.3	V_{REF} – 0.09	V_{REF} + 0.09	3.465	0.27	V_{CCIO} – 0.27	8	–8
MIPI D-PHY (LP)[3]	–0.3	0.55	0.88	3.465	—	—	—	—
HSUL12 (LPDDR2/3 Memory)	–0.3	V_{REF} – 0.1	V_{REF} + 0.1	3.465	0.3	V_{CCIO} – 0.3	4	–4

Notes:

1. For electromigration, the average DC current drawn by the I/O pads within a bank of I/Os shall not exceed 10 mA per I/O (All I/Os used in the same V_{CCIO}).
2. Not all IO types are supported in all banks. Refer to ECP5 and ECP5-5G sysIO Usage Guide (TN1262) for details.
3. MIPI D-PHY LP input can be implemented by powering VCCIO to 1.5V, and select MIPI LP primitive to meet MIPI Alliance spec on V_{IH} and V_{IL}. It can also be implemented as LVCMOS12 with VCCIO at 1.2V, which would meet V_{IH}/V_{IL} spec on LVCOM12.

Table 3.22. ECP5/ECP5-5G External Switching Characteristics *(Continued)*

Parameter	Description	Device	–8		–7		–6		Unit
			Min	Max	Min	Max	Min	Max	
Generic DDR Output									
Generic DDRX1 Outputs With Clock and Data Centered at Pin (GDDRX1_TX.SCLK.Centered) Using PCLK Clock Input - Figure 3.6									
$t_{DVB\_GDDRX1\_centered}$	Data Output Valid before CLK Output	All Devices	–0.67	—	–0.67	—	–0.67	—	ns + 1/2 UI
$t_{DVA\_GDDRX1\_centered}$	Data Output Valid after CLK Output	All Devices	–0.67	—	–0.67	—	–0.67	—	ns + 1/2 UI
$f_{DATA\_GDDRX1\_centered}$	GDDRX1 Data Rate	All Devices	—	500	—	500	—	500	Mb/s
$f_{MAX\_GDDRX1\_centered}$	GDDRX1 CLK Frequency (SCLK)	All Devices	—	250	—	250	—	250	MHz
Generic DDRX1 Outputs With Clock and Data Aligned at Pin (GDDRX1_TX.SCLK.Aligned) Using PCLK Clock Input - Figure 3.9									
$t_{DIB\_GDDRX1\_aligned}$	Data Output Invalid before CLK Output	All Devices	–0.3	—	–0.3	—	–0.3	—	ns
$t_{DIA\_GDDRX1\_aligned}$	Data Output Invalid after CLK Output	All Devices	—	0.3	—	0.3	—	0.3	ns
$f_{DATA\_GDDRX1\_aligned}$	GDDRX1 Data Rate	All Devices	—	500	—	500	—	500	Mb/s
$f_{MAX\_GDDRX1\_aligned}$	GDDRX1 CLK Frequency (SCLK)	All Devices	—	250	—	250	—	250	MHz
Generic DDRX2 Outputs With Clock and Data Centered at Pin (GDDRX2_TX.ECLK.Centered) Using PCLK Clock Input, Left and Right sides Only - Figure 3.8									
$t_{DVB\_GDDRX2\_centered}$	Data Output Valid Before CLK Output	All Devices	–0.442	—	–0.56	—	–0.676	—	ns + 1/2 UI
$t_{DVA\_GDDRX2\_centered}$	Data Output Valid After CLK Output	All Devices	—	0.442	—	0.56	—	0.676	ns + 1/2 UI
$f_{DATA\_GDDRX2\_centered}$	GDDRX2 Data Rate	All Devices	—	800	—	700	—	624	Mb/s
$f_{MAX\_GDDRX2\_centered}$	GDDRX2 CLK Frequency (ECLK)	All Devices	—	400	—	350	—	312	MHz
Generic DDRX2 Outputs With Clock and Data Aligned at Pin (GDDRX2_TX.ECLK.Aligned) Using PCLK Clock Input, Left and Right sides Only - Figure 3.9									
$t_{DIB\_GDDRX2\_aligned}$	Data Output Invalid before CLK Output	All Devices	–0.16	—	–0.18	—	–0.2	—	ns
$t_{DIA\_GDDRX2\_aligned}$	Data Output Invalid after CLK Output	All Devices	—	0.16	—	0.18	—	0.2	ns
$f_{DATA\_GDDRX2\_aligned}$	GDDRX2 Data Rate	All Devices	—	800	—	700	—	624	Mb/s
$f_{MAX\_GDDRX2\_aligned}$	GDDRX2 CLK Frequency (ECLK)	All Devices	—	400	—	350	—	312	MHz
Video DDRX71 Outputs With Clock and Data Aligned at Pin (GDDRX71_TX.ECLK) Using PLL Clock Input, Left and Right sides Only - Figure 3.12									
$t_{DIB\_LVDS71\_i}$	Data Output Invalid before CLK Output	All Devices	–0.16	—	–0.18	—	–0.2	—	ns + (i) * UI
$t_{DIA\_LVDS71\_i}$	Data Output Invalid after CLK Output	All Devices	—	0.16	—	0.18	—	0.2	ns + (i) * UI
f_{DATA_LVDS71}	DDR71 Data Rate	All Devices	—	756	—	620	—	525	Mb/s
f_{MAX_LVDS71}	DDR71 CLK Frequency (ECLK)	All Devices	—	378	—	310	—	262.5	MHz
Memory Interface									
DDR2/DDR3/DDR3L/LPDDR2/LPDDR3 READ (DQ Input Data are Aligned to DQS)									
t_{DVBDQ_DDR2} t_{DVBDQ_DDR3} t_{DVBDQ_DDR3L} t_{DVBDQ_LPDDR2} t_{DVBDQ_LPDDR3}	Data Output Valid before DQS Input	All Devices	—	–0.26	—	–0.317	—	–0.374	ns + 1/2 UI
t_{DVADQ_DDR2} t_{DVADQ_DDR3} t_{DVADQ_DDR3L} t_{DVADQ_LPDDR2} t_{DVADQ_LPDDR3}	Data Output Valid after DQS Input	All Devices	0.26	—	0.317	—	0.374	—	ns + 1/2 UI

Table 3.22. ECP5/ECP5-5G External Switching Characteristics *(Continued)*

3.22. SERDES High-Speed Data Receiver

Table 3.27. Serial Input Data Specifications

Symbol	Description	Min	Typ	Max	Unit
$V_{RX-DIFF-S}$	Differential input sensitivity	150	—	1760	mV, p-p
V_{RX-IN}	Input levels	0	—	V_{CCA} +0.5[2]	V
$V_{RX-CM-DCCM}$	Input common mode range (internal DC coupled mode)	0.6	—	V_{CCA}	V
$V_{RX-CM-ACCM}$	Input common mode range (internal AC coupled mode)[2]	0.1	—	V_{CCA} +0.2	V
$T_{RX-RELOCK}$	SCDR re-lock time[1]	—	1000	—	Bits
$Z_{RX-TERM}$	Input termination 50/75 Ω /High Z	–20%	50/75/5 K	+20%	Ω
RL_{RX-RL}	Return loss (without package)	—	—	–10	dB

Notes:
1. This is the typical number of bit times to re-lock to a new phase or frequency within ±300 ppm, assuming 8b10b encoded data.
2. Up to 1.655 for ECP5, and 1.76 for ECP5-5G.

3.23. Input Data Jitter Tolerance

A receiver's ability to tolerate incoming signal jitter is very dependent on jitter type. High speed serial interface standards have recognized the dependency on jitter type and have specifications to indicate tolerance levels for different jitter types as they relate to specific protocols. Sinusoidal jitter is considered to be a worst case jitter type.

Table 3.28. Receiver Total Jitter Tolerance Specification

Description	Frequency	Condition	Min	Typ	Max	Unit
Deterministic	5 Gb/s	400 mV differential eye	—	—	TBD	UI, p-p
Random		400 mV differential eye	—	—	TBD	UI, p-p
Total		400 mV differential eye	—	—	TBD	UI, p-p
Deterministic	3.125 Gb/s	400 mV differential eye	—	—	0.37	UI, p-p
Random		400 mV differential eye	—	—	0.18	UI, p-p
Total		400 mV differential eye	—	—	0.65	UI, p-p
Deterministic	2.5 Gb/s	400 mV differential eye	—	—	0.37	UI, p-p
Random		400 mV differential eye	—	—	0.18	UI, p-p
Total		400 mV differential eye	—	—	0.65	UI, p-p
Deterministic	1.25 Gb/s	400 mV differential eye	—	—	0.37	UI, p-p
Random		400 mV differential eye	—	—	0.18	UI, p-p
Total		400 mV differential eye	—	—	0.65	UI, p-p

Notes:
1. Jitter tolerance measurements are done with protocol compliance tests: 3.125 Gb/s - XAUI Standard, 2.5 Gb/s - PCIe Standard, 1.25 Gb/s - SGMII Standard.
2. For ECP5-5G family devices only.

3.24. SERDES External Reference Clock

The external reference clock selection and its interface are a critical part of system applications for this product. Table 3.29 specifies reference clock requirements, over the full range of operating conditions.

Table 3.29. External Reference Clock Specification (refclkp/refclkn)

Symbol	Description	Min	Typ	Max	Unit
F_{REF}	Frequency range	50	—	320	MHz
$F_{REF-PPM}$	Frequency tolerance[1]	–1000	—	1000	ppm
$V_{REF-IN-SE}$	Input swing, single-ended clock[2, 4]	200	—	V_{CCAUXA}	mV, p-p
$V_{REF-IN-DIFF}$	Input swing, differential clock	200	—	$2*V_{CCAUXA}$	mV, p-p differential
V_{REF-IN}	Input levels	0	—	V_{CCAUXA} + 0.4	V
D_{REF}	Duty cycle[3]	40	—	60	%
T_{REF-R}	Rise time (20% to 80%)	200	500	1000	ps
T_{REF-F}	Fall time (80% to 20%)	200	500	1000	ps
$Z_{REF-IN-TERM-DIFF}$	Differential input termination	–30%	100/HiZ	+30%	Ω
$C_{REF-IN-CAP}$	Input capacitance	—	—	7	pF

Notes:

1. Depending on the application, the PLL_LOL_SET and CDR_LOL_SET control registers may be adjusted for other tolerance values as described in ECP5 and ECP5-5G SERDES/PCS Usage Guide (TN1261).
2. The signal swing for a single-ended input clock must be as large as the p-p differential swing of a differential input clock to get the same gain at the input receiver. With single-ended clock, a reference voltage needs to be externally connected to CLKREFN pin, and the input voltage needs to be swung around this reference voltage.
3. Measured at 50% amplitude.
4. Single-ended clocking is achieved by applying a reference voltage V_{REF} on REFCLKN input, with the clock applied to REFCLKP input pin. V_{REF} should be set to mid-point of the REFCLKP voltage swing.

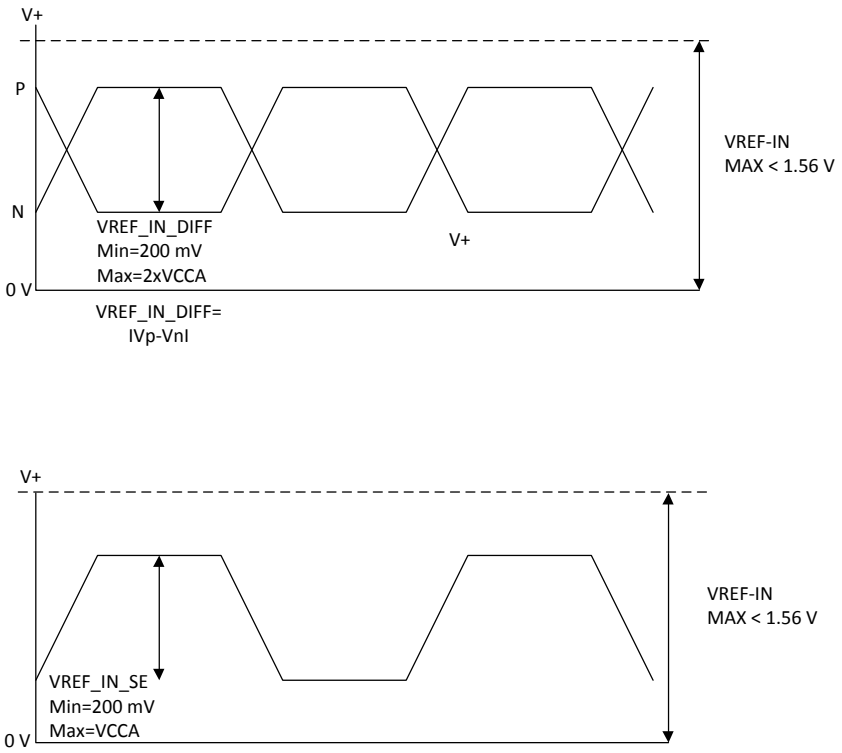

Figure 3.14. SERDES External Reference Clock Waveforms

3.25. PCI Express Electrical and Timing Characteristics

3.25.1. PCIe (2.5 Gb/s) AC and DC Characteristics

Over recommended operating conditions.

Table 3.30. PCIe (2.5 Gb/s)

Symbol	Description	Test Conditions	Min	Typ	Max	Unit
Transmit[1]						
UI	Unit interval	—	399.88	400	400.12	ps
$V_{TX-DIFF\_P-P}$	Differential peak-to-peak output	—	0.8	1.0	1.2	V
$V_{TX-DE-RATIO}$	De-emphasis differential output voltage ratio	—	–3	–3.5	–4	dB
$V_{TX-CM-AC\_P}$	RMS AC peak common-mode output voltage	—	—	—	20	mV
$V_{TX-RCV-DETECT}$	Amount of voltage change allowed during receiver detection	—	—	—	600	mV
$V_{TX-CM-DC}$	Tx DC common mode voltage	—	0	—	V_{CCHTX}	V
$I_{TX-SHORT}$	Output short circuit current	V_{TX-D+}=0.0 V V_{TX-D-}=0.0 V	—	—	90	mA
$Z_{TX-DIFF-DC}$	Differential output impedance	—	80	100	120	Ω
$RL_{TX-DIFF}$	Differential return loss	—	10	—	—	dB
RL_{TX-CM}	Common mode return loss	—	6.0	—	—	dB
$T_{TX-RISE}$	Tx output rise time	20% to 80%	0.125	—	—	UI
$T_{TX-FALL}$	Tx output fall time	20% to 80%	0.125	—	—	UI
$L_{TX-SKEW}$	Lane-to-lane static output skew for all lanes in port/link	—	—	—	1.3	ns
T_{TX-EYE}	Transmitter eye width	—	0.75	—	—	UI
$T_{TX-EYE-MEDIAN-TO-MAX-JITTER}$	Maximum time between jitter median and maximum deviation from median	—	—	—	0.125	UI
Receive[1, 2]						
UI	Unit Interval	—	399.88	400	400.12	ps
$V_{RX-DIFF\_P-P}$	Differential peak-to-peak input voltage	—	0.34[3]	—	1.2	V
$V_{RX-IDLE-DET-DIFF\_P-P}$	Idle detect threshold voltage	—	65	—	340[3]	mV
$V_{RX-CM-AC\_P}$	RMS AC peak common-mode input voltage	—	—	—	150	mV
$Z_{RX-DIFF-DC}$	DC differential input impedance	—	80	100	120	Ω
Z_{RX-DC}	DC input impedance	—	40	50	60	Ω
$Z_{RX-HIGH-IMP-DC}$	Power-down DC input impedance	—	200K	—	—	Ω
$RL_{RX-DIFF}$	Differential return loss	—	10	—	—	dB
RL_{RX-CM}	Common mode return loss	—	6.0	—	—	dB

Notes:

1. Values are measured at 2.5 Gb/s.
2. Measured with external AC-coupling on the receiver.
3. Not in compliance with PCI Express 1.1 standard.

3.25.2. PCIe (5 Gb/s) – Preliminary AC and DC Characteristics

Over recommended operating conditions.

Table 3.31. PCIe (5 Gb/s)

Symbol	Description	Test Conditions	Min	Typ	Max	Unit
Transmit[1]						
UI	Unit Interval	—	199.94	200	200.06	ps
$B_{WTX-PKG-PLL2}$	Tx PLL bandwidth corresponding to PKGTX-PLL2	—	5	—	16	MHz
$P_{KGTX-PLL2}$	Tx PLL Peaking	—	—	—	1	dB
$V_{TX-DIFF-PP}$	Differential p-p Tx voltage swing	—	0.8	—	1.2	V, p-p
$V_{TX-DIFF-PP-LOW}$	Low power differential p-p Tx voltage swing	—	0.4	—	1.2	V, p-p
$V_{TX-DE-RATIO-3.5dB}$	Tx de-emphasis level ratio at 3.5dB	—	3	—	4	dB
$V_{TX-DE-RATIO-6dB}$	Tx de-emphasis level ratio at 6dB	—	5.5	—	6.5	dB
$T_{MIN-PULSE}$	Instantaneous lone pulse width	—		—	—	UI
$T_{TX-RISE-FALL}$	Transmitter rise and fall time	—		—	—	UI
T_{TX-EYE}	Transmitter Eye, including all jitter sources	—	0.75	—	—	UI
T_{TX-DJ}	Tx deterministic jitter > 1.5 MHz	—	—	—	0.15	UI
T_{TX-RJ}	Tx RMS jitter < 1.5 MHz	—	—	—	3	ps, RMS
$T_{RF-MISMATCH}$	Tx rise/fall time mismatch	—	—	—		UI
$R_{LTX-DIFF}$	Tx Differential Return Loss, including package and silicon	50 MHz < freq < 1.25 GHz	10	—	—	dB
		1.25 GHz < freq < 2.5 GHz	8	—	—	dB
R_{LTX-CM}	Tx Common Mode Return Loss, including package and silicon	50 MHz < freq < 2.5 GHz	6	—	—	dB
$Z_{TX-DIFF-DC}$	DC differential Impedance	—	—	—	120	Ω
$V_{TX-CM-AC-PP}$	Tx AC peak common mode voltage, peak-peak	—	—	—		mV, p-p
$I_{TX-SHORT}$	Transmitter short-circuit current	—	—	—	90	mA
$V_{TX-DC-CM}$	Transmitter DC common-mode voltage	—	0	—	1.2	V
$V_{TX-IDLE-DIFF-DC}$	Electrical Idle Output DC voltage	—	0	—	5	mV
$V_{TX-IDLE-DIFF-AC-p}$	Electrical Idle Differential Output peak voltage	—	—	—		mV
$V_{TX-RCV-DETECT}$	Voltage change allowed during Receiver Detect	—	—	—	600	mV
$T_{TX-IDLE-MIN}$	Min. time in Electrical Idle	—	20	—	—	ns
$T_{TX-IDLE-SET-TO-IDLE}$	Max. time from EI Order Set to valid Electrical Idle	—	—	—	8	ns
$T_{TX-IDLE-TO-DIFF-DATA}$	Max. time from Electrical Idle to valid differential output	—	—	—	8	ns
$L_{TX-SKEW}$	Lane-to-lane output skew	—	—	—		ps

3.26. CPRI LV2 E.48 Electrical and Timing Characteristics – Preliminary

Table 3.32. CPRI LV2 E.48 Electrical and Timing Characteristics

Symbol	Description	Test Conditions	Min	Typ	Max	Unit
Transmit						
UI	Unit Interval	—	203.43	203.45	203.47	ps
T_{DCD}	Duty Cycle Distortion	—	—	—	0.05	UI
J_{UBHPJ}	Uncorrelated Bounded High Probability Jitter	—	—	—	0.15	UI
J_{TOTAL}	Total Jitter	—	—	—	0.3	UI
$Z_{RX-DIFF-DC}$	DC differential Impedance	—	80	—	120	Ω
T_{SKEW}	Skew between differential signals	—	—	—	9	ps
$R_{LTX-DIFF}$	Tx Differential Return Loss (S22), including package and silicon	100 MHz < freq < 3.6864 GHz	—	—	–8	dB
		3.6864 GHz < freq < 4.9152 GHz	—	—	–8 + 16.6 *log (freq/3.6864)	dB
R_{LTX-CM}	Tx Common Mode Return Loss, including package and silicon	100 MHz < freq < 3.6864 GHz	6	—	—	dB
$I_{TX-SHORT}$	Transmitter short-circuit current	—	—	—	100	mA
$T_{RISE\_FALL-DIFF}$	Differential Rise and Fall Time	—		—	—	ps
$L_{TX-SKEW}$	Lane-to-lane output skew	—	—	—		ps
Receive						
UI	Unit Interval	—	203.43	203.45	203.47	ps
$V_{RX-DIFF-PP}$	Differential Rx peak-peak voltage	—	—	—	1.2	V, p-p
$V_{RX-EYE\_Y1\_Y2}$	Receiver eye opening mask, Y1 and Y2	—	62.5	—	375	mV, diff
V_{RX-EYE_X1}	Receiver eye opening mask, X1	—	—	—	0.3	UI
T_{RX-TJ}	Receiver total jitter tolerance (not including sinusoidal)	—	—	—	0.6	UI
$R_{LRX-DIFF}$	Receiver differential Return Loss, package plus silicon	100 MHz < freq < 3.6864 GHz	—	—	–8	dB
		3.6864 GHz < freq < 4.9152 GHz	—	—	–8 + 16.6 *log (freq/3.6864)	dB
R_{LRX-CM}	Receiver common mode Return Loss, package plus silicon	—	6	—	—	dB
$Z_{RX-DIFF-DC}$	Receiver DC differential impedance	—	80	100	120	Ω

Note: Data is measured with PRBS7 data pattern, not with PRBS-31 pattern.

3.27. XAUI/CPRI LV E.30 Electrical and Timing Characteristics

3.27.1. AC and DC Characteristics

Over recommended operating conditions.

Table 3.33. Transmit

Symbol	Description	Test Conditions	Min	Typ	Max	Unit
T_{RF}	Differential rise/fall time	20% to 80%	—	80	—	ps
$Z_{TX\_DIFF\_DC}$	Differential impedance	—	80	100	120	Ω
J_{TX_DDJ}[2, 3]	Output data deterministic jitter	—	—	—	0.17	UI
J_{TX_TJ}[1, 2, 3]	Total output data jitter	—	—	—	0.35	UI

Notes:
1. Total jitter includes both deterministic jitter and random jitter.
2. Jitter values are measured with each CML output AC coupled into a 50 Ω impedance (100 Ω differential impedance).
3. Jitter and skew are specified between differential crossings of the 50% threshold of the reference signal.

Over recommended operating conditions.

Table 3.34. Receive and Jitter Tolerance

Symbol	Description	Test Conditions	Min	Typ	Max	Unit
RL_{RX_DIFF}	Differential return loss	From 100 MHz to 3.125 GHz	10	—	—	dB
RL_{RX_CM}	Common mode return loss	From 100 MHz to 3.125 GHz	6	—	—	dB
Z_{RX_DIFF}	Differential termination resistance	—	80	100	120	Ω
J_{RX_DJ}[1, 2, 3]	Deterministic jitter tolerance (peak-to-peak)	—	—	—	0.37	UI
J_{RX_RJ}[1, 2, 3]	Random jitter tolerance (peak-to-peak)	—	—	—	0.18	UI
J_{RX_SJ}[1, 2, 3]	Sinusoidal jitter tolerance (peak-to-peak)	—	—	—	0.10	UI
J_{RX_TJ}[1, 2, 3]	Total jitter tolerance (peak-to-peak)	—	—	—	0.65	UI
T_{RX_EYE}	Receiver eye opening	—	0.35	—	—	UI

Notes:
1. Total jitter includes deterministic jitter, random jitter and sinusoidal jitter.
2. Jitter values are measured with each high-speed input AC coupled into a 50 Ω impedance.
3. Jitter and skew are specified between differential crossings of the 50% threshold of the reference signal.

3.28. CPRI LV E.24/SGMII(2.5Gbps) Electrical and Timing Characteristics

3.28.1. AC and DC Characteristics

Table 3.35. Transmit

Symbol	Description	Test Conditions	Min	Typ	Max	Unit
T_{RF}[1]	Differential rise/fall time	20% to 80%	—	80	—	ps
$Z_{TX\_DIFF\_DC}$	Differential impedance	—	80	100	120	Ω
J_{TX_DDJ}[3, 4]	Output data deterministic jitter	—	—	—	0.17	UI
J_{TX_TJ}[2, 3, 4]	Total output data jitter	—	—	—	0.35	UI

Notes:
1. Rise and Fall times measured with board trace, connector and approximately 2.5 pf load.
2. Total jitter includes both deterministic jitter and random jitter. The random jitter is the total jitter minus the actual deterministic jitter.
3. Jitter values are measured with each CML output AC coupled into a 50 Ω impedance (100 Ω differential impedance).
4. Jitter and skew are specified between differential crossings of the 50% threshold of the reference signal.

Table 3.44. Test Fixture Required Components, Non-Terminated Interfaces

Test Condition	R_1	R_2	C_L	Timing Ref.	V_T
LVTTL and other LVCMOS settings (L ≥ H, H ≥ L)	∞	∞	0 pF	LVCMOS 3.3 = 1.5 V	—
				LVCMOS 2.5 = $V_{CCIO}/2$	—
				LVCMOS 1.8 = $V_{CCIO}/2$	—
				LVCMOS 1.5 = $V_{CCIO}/2$	—
				LVCMOS 1.2 = $V_{CCIO}/2$	—
LVCMOS 2.5 I/O (Z ≥ H)	∞	1 MΩ	0 pF	$V_{CCIO}/2$	—
LVCMOS 2.5 I/O (Z ≥ L)	1 MΩ	∞	0 pF	$V_{CCIO}/2$	V_{CCIO}
LVCMOS 2.5 I/O (H ≥ Z)	∞	100	0 pF	V_{OH} – 0.10	—
LVCMOS 2.5 I/O (L ≥ Z)	100	∞	0 pF	V_{OL} + 0.10	V_{CCIO}

Note: Output test conditions for all other interfaces are determined by the respective standards.

4.2. PICs and DDR Data (DQ) Pins Associated with the DDR Strobe (DQS) Pin

PICs Associated with DQS Strobe	PIO within PIC	DDR Strobe (DQS) and Data (DQ) Pins
For Left and Right Edges of the Device Only		
P[L/R] [n–6]	A	DQ
	B	DQ
	C	DQ
	D	DQ
P[L/R] [n–3]	A	DQ
	B	DQ
	C	DQ
	D	DQ
P[L/R] [n]	A	DQS (P)
	B	DQS (N)
	C	DQ
	D	DQ
P[L/R] [n+3]	A	DQ
	B	DQ
	C	DQ
	D	DQ

Note: "n" is a row PIC number.

4.3. Pin Information Summary

4.3.1. LFE5UM/LFE5UM5G

Pin Information Summary		LFE5UM/ LFE5UM5G-25		LFE5UM/LFE5UM5G-45			LFE5UM/LFE5UM5G-85			
Pin Type		285 csfBG	381 caBGA	285 csfBGA	381 caBG	554 caBGA	285 csfBGA	381 caBG	554 caBGA	756 caBGA
General Purpose Inputs/Outputs per Bank	Bank 0	6	24	6	27	32	6	27	32	56
	Bank 1	6	32	6	33	40	6	33	40	48
	Bank 2	21	32	21	32	32	21	34	32	48
	Bank 3	28	32	28	33	48	28	33	48	64
	Bank 4	0	0	0	0	0	0	0	14	24
	Bank 6	26	32	26	33	48	26	33	48	64
	Bank 7	18	32	18	32	32	18	32	32	48
	Bank 8	13	13	13	13	13	13	13	13	13
Total Single-Ended User I/O		118	197	118	203	245	118	205	259	365
VCC		13	20	13	20	24	13	20	24	36
VCCAUX (Core)		3	4	3	4	9	3	4	9	8
VCCIO	Bank 0	1	2	1	2	3	1	2	3	4
	Bank 1	1	2	1	2	3	1	2	3	4
	Bank 2	2	3	2	3	4	2	3	4	4
	Bank 3	2	3	2	3	3	2	3	3	4
	Bank 4	0	0	0	0	0	0	0	2	2
	Bank 6	2	3	2	3	4	2	3	4	4
	Bank 7	2	3	2	3	3	2	3	3	4
	Bank 8	2	2	2	2	2	2	2	2	2