Lattice Semiconductor Corporation - LFE5UM-85F-6BG756C Datasheet

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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Active
Number of LABs/CLBs	21000
Number of Logic Elements/Cells	84000
Total RAM Bits	3833856
Number of I/O	365
Number of Gates	-
Voltage - Supply	1.045V ~ 1.155V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	756-FBGA
Supplier Device Package	756-CABGA (27x27)
Purchase URL	https://www.e-xfl.com/product-detail/lattice-semiconductor/lfe5um-85f-6bg756c

2. Architecture

2.1. Overview

Each ECP5/ECP5-5G device contains an array of logic blocks surrounded by Programmable I/O Cells (PIC). Interspersed between the rows of logic blocks are rows of sysMEM™ Embedded Block RAM (EBR) and rows of sysDSP™ Digital Signal Processing slices, as shown in Figure 2.1 on page 13. The LFE5-85 devices have three rows of DSP slices, the LFE5-45 devices have two rows, and both LFE5-25 and LFE5-12 devices have one. In addition, the LFE5UM/LFE5UM5G devices contain SERDES Duals on the bottom of the device.

The Programmable Functional Unit (PFU) contains the building blocks for logic, arithmetic, RAM and ROM functions. The PFU block is optimized for flexibility, allowing complex designs to be implemented quickly and efficiently. Logic Blocks are arranged in a two-dimensional array.

The ECP5/ECP5-5G devices contain one or more rows of sysMEM EBR blocks. sysMEM EBRs are large, dedicated 18 Kb fast memory blocks. Each sysMEM block can be configured in a variety of depths and widths as RAM or ROM. In addition, ECP5/ECP5-5G devices contain up to three rows of DSP slices. Each DSP slice has multipliers and adder/accumulators, which are the building blocks for complex signal processing capabilities.

The ECP5 devices feature up to four embedded 3.2 Gb/s SERDES channels, and the ECP5-5G devices feature up to four embedded 5 Gb/s SERDES channels. Each SERDES channel contains independent 8b/10b encoding / decoding, polarity adjust and elastic buffer logic. Each group of two SERDES channels, along with its Physical Coding Sublayer (PCS) block, creates a dual DCU (Dual Channel Unit). The functionality of the SERDES/PCS duals can be controlled by SRAM cell settings during device configuration or by registers that are addressable during device operation. The registers in every dual can be programmed via the SERDES Client Interface (SCI). These DCUs (up to two) are located at the bottom of the devices.

Each PIC block encompasses two PIOs (PIO pairs) with their respective sysI/O buffers. The sysI/O buffers of the ECP5/ECP5-5G devices are arranged in seven banks (eight banks for LFE5-85 devices in caBGA756 and caBGA554 packages), allowing the implementation of a wide variety of I/O standards. One of these banks (Bank 8) is shared with the programming interfaces. Half of the PIO pairs on the left and right edges of the device can be configured as LVDS transmit pairs, and all pairs on left and right can be configured as LVDS receive pairs. The PIC logic in the left and right banks also includes pre-engineered support to aid in the implementation of high speed source synchronous standards such as XGMII, 7:1 LVDS, along with memory interfaces including DDR3 and LPDDR3.

The ECP5/ECP5-5G registers in PFU and sysI/O can be configured to be SET or RESET. After power up and the device is configured, it enters into user mode with these registers SET/RESET according to the configuration setting, allowing the device entering to a known state for predictable system function.

Other blocks provided include PLLs, DLLs and configuration functions. The ECP5/ECP5-5G architecture provides up to four Delay-Locked Loops (DLLs) and up to four Phase-Locked Loops (PLLs). The PLL and DLL blocks are located at the corners of each device.

The configuration block that supports features such as configuration bit-stream decryption, transparent updates and dual-boot support is located at the bottom of each device, to the left of the SERDES blocks. Every device in the ECP5/ECP5-5G family supports a sysCONFIG™ ports located in that same corner, powered by Vccio8, allowing for serial or parallel device configuration.

In addition, every device in the family has a JTAG port. This family also provides an on-chip oscillator and soft error detect capability. The ECP5 devices use 1.1 V and ECP5UM5G devices use 1.2 V as their core voltage.

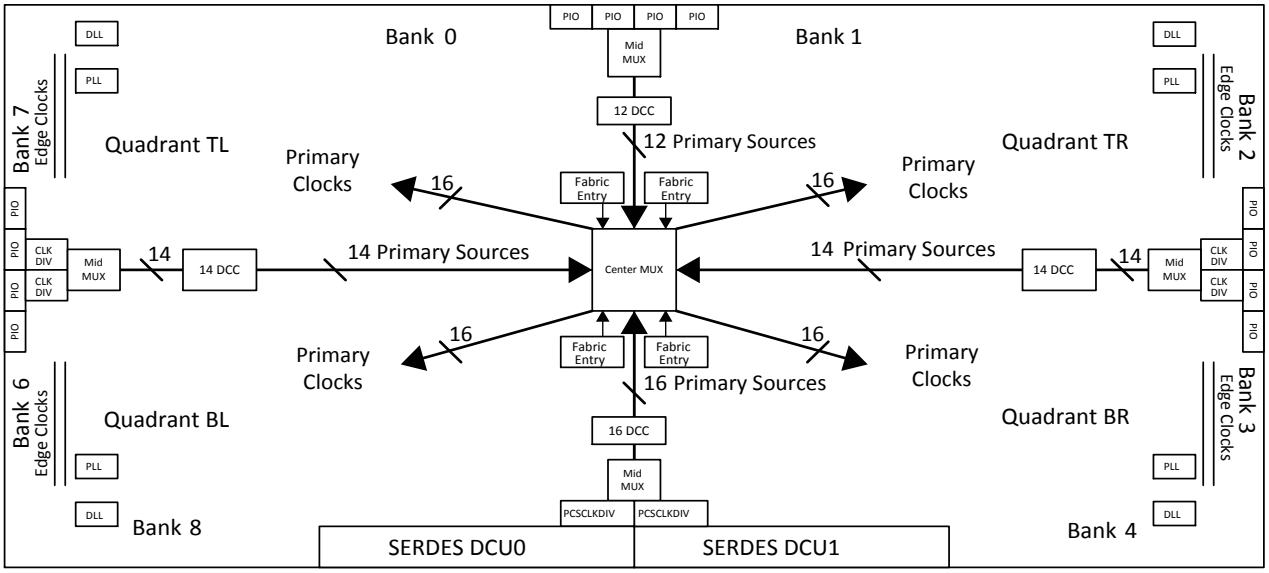


Figure 2.6. LFE5UM/LFE5UM5G-85 Clocking

2.5.1. Primary Clocks

The ECP5/ECP5-5G device family provides low-skew, high fan-out clock distribution to all synchronous elements in the FPGA fabric through the Primary Clock Network.

The primary clock network is divided into four clocking quadrants: Top Left (TL), Bottom Left (BL), Top Right (TR), and Bottom Right (BR). Each of these quadrants has 16 clocks that can be distributed to the fabric in the quadrant.

The Lattice Diamond software can automatically route each clock to one of the four quadrants up to a maximum of 16 clocks per quadrant. The user can change how the clocks are routed by specifying a preference in the Lattice Diamond software to locate the clock to specific. The ECP5/ECP5-5G device provides the user with a maximum of 64 unique clock input sources that can be routed to the primary Clock network.

Primary clock sources are:

- Dedicated clock input pins
- PLL outputs
- CLKDIV outputs
- Internal FPGA fabric entries (with minimum general routing)
- SERDES/PCS/PCSDIV clocks
- OSC clock

These sources are routed to one of four clock switches called a Mid Mux. The outputs of the Mid MUX are routed to the center of the FPGA where another clock switch, called the Center MUX, is used to route the primary clock sources to primary clock distribution to the ECP5/ECP5-5G fabric. These routing muxes are shown in Figure 2.6. Since there is a maximum of 60 unique clock input sources to the clocking quadrants, there are potentially 64 unique clock domains that can be used in the ECP5/ECP5-5G Device. For more information about the primary clock tree and connections, refer to ECP5 and ECP5-5G sysClock PLL/DLL Design and Usage Guide (TN1263).

2.5.1.1. Dynamic Clock Control

The Dynamic Clock Control (DCC), Quadrant Clock enable/disable feature allows internal logic control of the quadrant primary clock network. When a clock network is disabled, the clock signal is static and not toggle. All the logic fed by that clock will not toggle, reducing the overall power consumption of the device. The disable function will not create glitch and increase the clock latency to the primary clock network.

This DCC controls the clock sources from the Primary CLOCK MIDMUX before they are fed to the Primary Center MUXs that drive the quadrant clock network. For more information about the DCC, refer to ECP5 and ECP5-5G sysClock PLL/DLL Design and Usage Guide (TN1263).

- 5*5 and larger size 2D blocks – Semi internal DSP Slice support
- Flexible saturation and rounding options to satisfy a diverse set of applications situations
- Flexible cascading across DSP slices
  - Minimizes fabric use for common DSP and ALU functions
  - Enables implementation of FIR Filter or similar structures using dedicated sysDSP slice resources only
  - Provides matching pipeline registers
  - Can be configured to continue cascading from one row of sysDSP slices to another for longer cascade chains
- Flexible and Powerful Arithmetic Logic Unit (ALU) Supports:
  - Dynamically selectable ALU OPCODE
  - Ternary arithmetic (addition/subtraction of three inputs)
  - Bit-wise two-input logic operations (AND, OR, NAND, NOR, XOR and XNOR)
  - Eight flexible and programmable ALU flags that can be used for multiple pattern detection scenarios, such as, overflow, underflow and convergent rounding.
  - Flexible cascading across slices to get larger functions
- RTL Synthesis friendly synchronous reset on all registers, while still supporting asynchronous reset for legacy users
- Dynamic MUX selection to allow Time Division Multiplexing (TDM) of resources for applications that require processor-like flexibility that enables different functions for each clock cycle

For most cases, as shown in Figure 2.14, the ECP5/ECP5-5G sysDSP slice is backwards-compatible with the LatticeECP2™ and LatticeECP3™ sysDSP block, such that, legacy applications can be targeted to the ECP5/ ECP5-5G sysDSP slice. Figure 2.14 shows the diagram of sysDSP, and Figure 2.15 shows the detailed diagram.

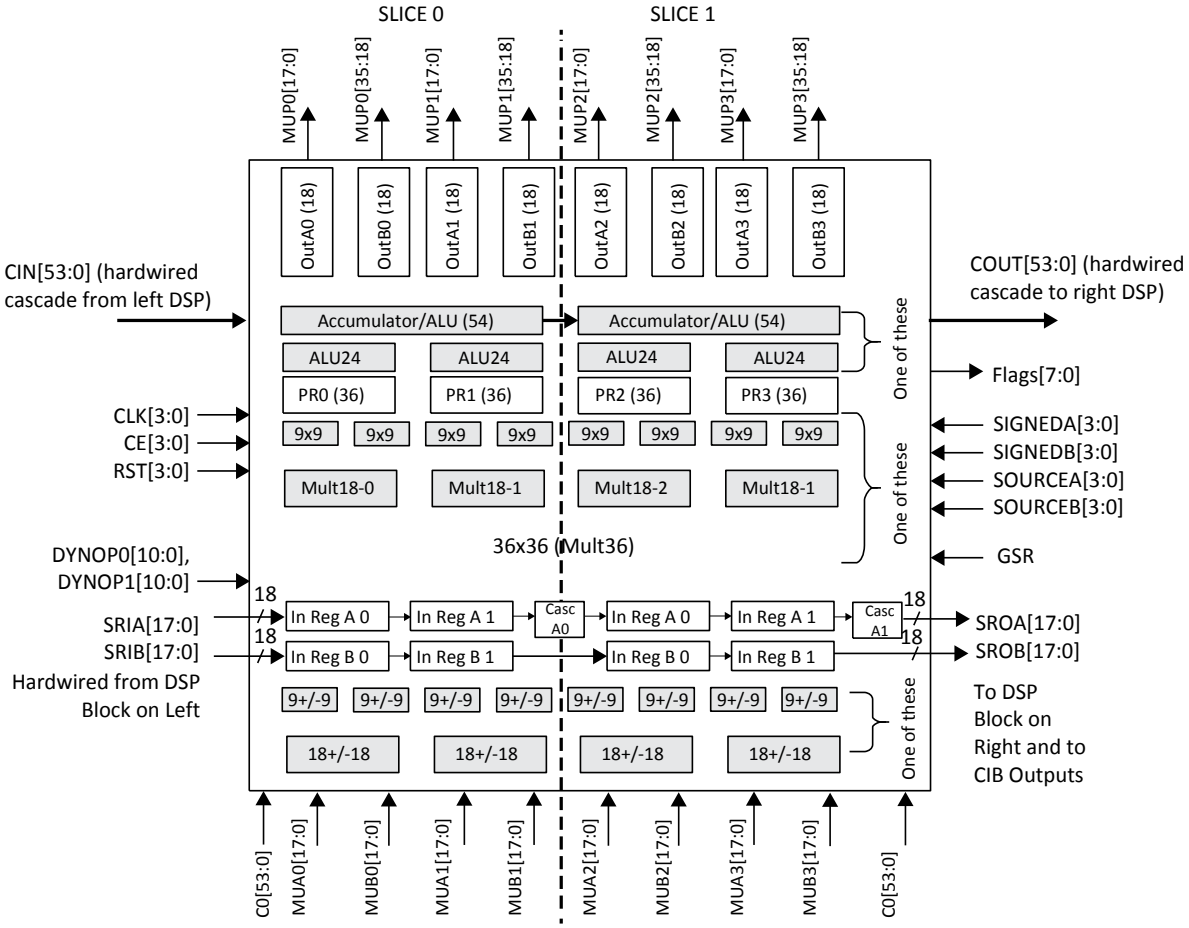


Figure 2.14. Simplified sysDSP Slice Block Diagram

2.11.1.1. Input FIFO

The ECP5/ECP5-5G PIO has dedicated input FIFO per single-ended pin for input data register for DDR Memory interfaces. The FIFO resides before the gearing logic. It transfers data from DQS domain to continuous ECLK domain. On the Write side of the FIFO, it is clocked by DQS clock which is the delayed version of the DQS Strobe signal from DDR memory. On the Read side of FIFO, it is clocked by ECLK. ECLK may be any high speed clock with identical frequency as DQS (the frequency of the memory chip). Each DQS group has one FIFO control block. It distributes FIFO read/write pointer to every PIC in same DQS group. DQS Grouping and DQS Control Block is described in DDR Memory Support section on page 35.

Table 2.8. Input Block Port Description

Name	Type	Description
D	Input	High Speed Data Input
Q[1:0]/Q[3:0]/Q[6:0]	Output	Low Speed Data to the device core
RST	Input	Reset to the Output Block
SCLK	Input	Slow Speed System Clock
ECLK	Input	High Speed Edge Clock
DQS	Input	Clock from DQS control Block used to clock DDR memory data
ALIGNWD	Input	Data Alignment signal from device core.

2.11.2. Output Register Block

The output register block registers signal from the core of the device before they are passed to the sysIO buffers.

ECP5/ECP5-5G output data path has output programmable flip flops and output gearing logic. On the left and right sides, the output register block can support 1x, 2x and 7:1 gearing enabling high speed DDR interfaces and DDR memory interfaces. On the top side, the banks support 1x gearing. ECP5/ECP5-5G output data path diagram is shown in Figure 2.19. The programmable delay cells are also available in the output data path.

For detailed description of the output register block modes and usage, refer to ECP5 and ECP5-5G High-Speed I/O Interface (TN1265).

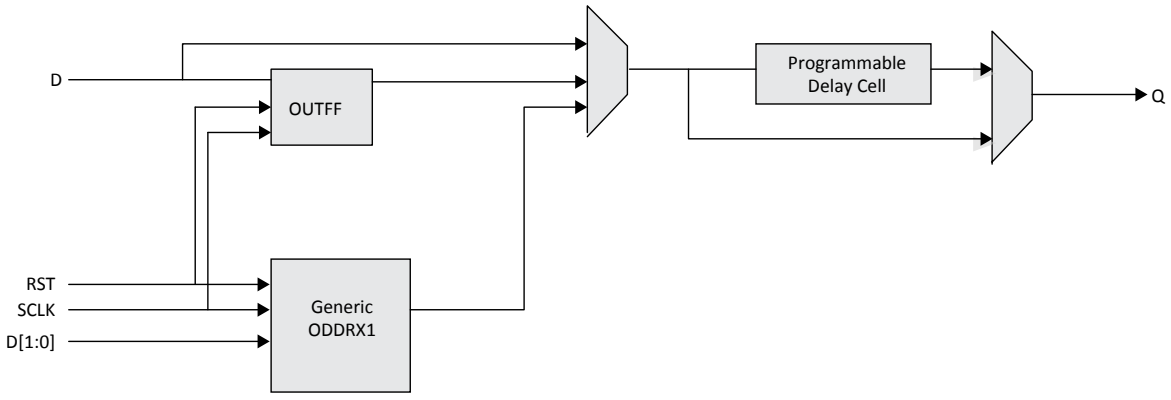


Figure 2.19. Output Register Block on Top Side

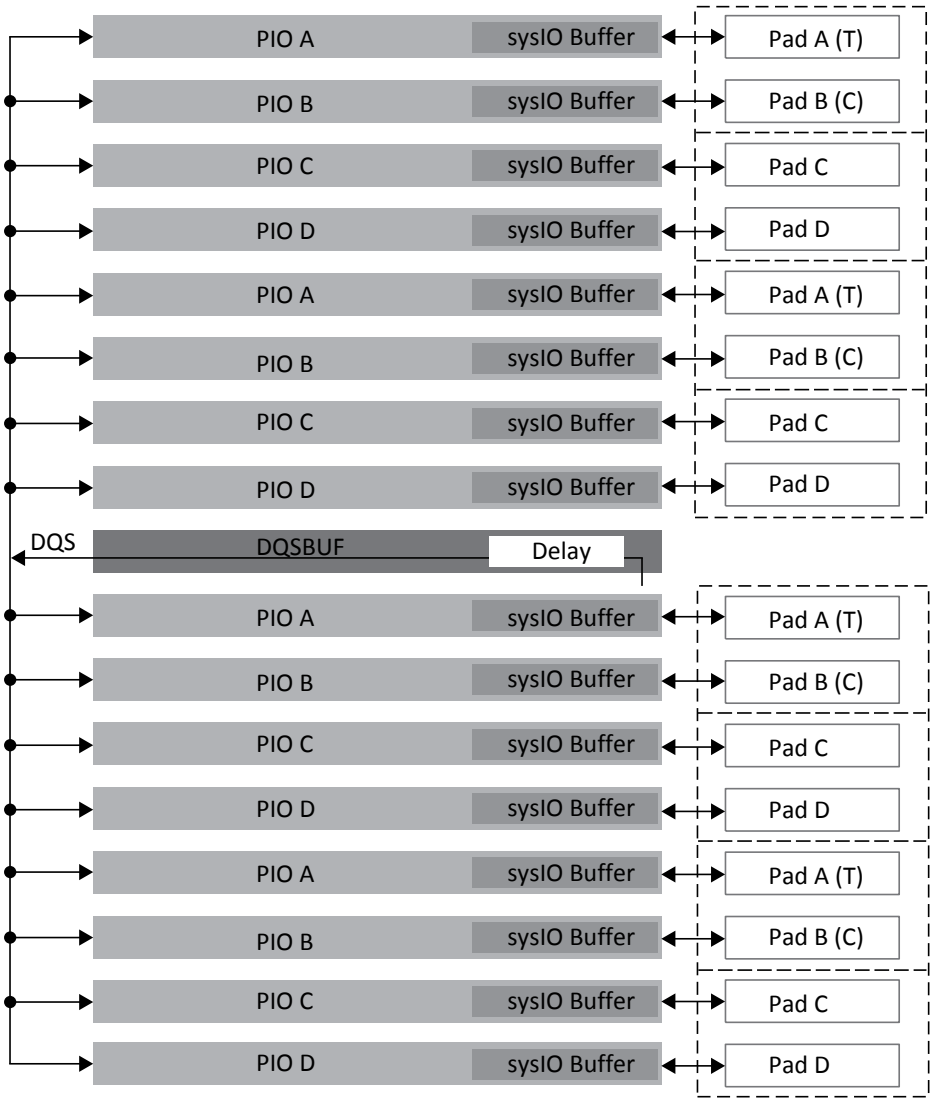


Figure 2.23. DQS Grouping on the Left and Right Edges

2.13.2. DLL Calibrated DQS Delay and Control Block (DQSBUF)

To support DDR memory interfaces (DDR2/3, LPDDR2/3), the DQS strobe signal from the memory must be used to capture the data (DQ) in the PIC registers during memory reads. This signal is output from the DDR memory device aligned to data transitions and must be time shifted before it can be used to capture data in the PIC. This time shifted is achieved by using DQSDEL programmable delay line in the DQS Delay Block (DQS read circuit). The DQSDEL is implemented as a slave delay line and works in conjunction with a master DDRDLL.

This block also includes slave delay line to generate delayed clocks used in the write side to generate DQ and DQS with correct phases within one DQS group. There is a third delay line inside this block used to provide write leveling feature for DDR write if needed.

Each of the read and write side delays can be dynamically shifted using margin control signals that can be controlled by the core logic.

FIFO Control Block shown in Figure 2.24 generates the Read and Write Pointers for the FIFO block inside the Input Register Block. These pointers are generated to control the DQS to ECLK domain crossing using the FIFO module.

Table 2.14. Available SERDES Duals per LFE5UM/LFE5UM5G Devices

Package	LFE5UM/LFE5UM5G-25	LFE5UM/LFE5UM5G-45	LFE5UM/LFE5UM5G-85
285 csfBGA	1	1	1
381 caBGA	1	2	2
554 caBGA	—	2	2
756 caBGA	—	—	2

2.15.1. SERDES Block

A SERDES receiver channel may receive the serial differential data stream, equalize the signal, perform Clock and Data Recovery (CDR) and de-serialize the data stream before passing the 8- or 10-bit data to the PCS logic. The SERDES transmitter channel may receive the parallel 8- or 10-bit data, serialize the data and transmit the serial bit stream through the differential drivers. Figure 2.28 shows a single-channel SERDES/PCS block. Each SERDES channel provides a recovered clock and a SERDES transmit clock to the PCS block and to the FPGA core logic.

Each transmit channel, receiver channel, and SERDES PLL shares the same power supply (VCCA). The output and input buffers of each channel have their own independent power supplies (VCCHTX and VCCHRX).

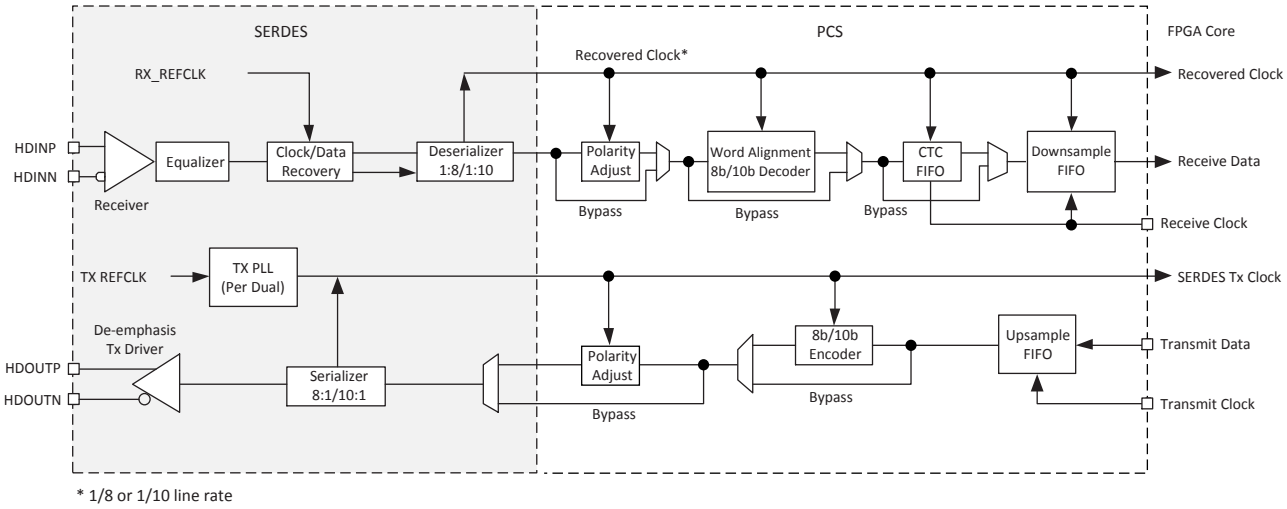


Figure 2.28. Simplified Channel Block Diagram for SERDES/PCS Block

2.15.2. PCS

As shown in Figure 2.28, the PCS receives the parallel digital data from the deserializer and selects the polarity, performs word alignment, decodes (8b/10b), provides Clock Tolerance Compensation and transfers the clock domain from the recovered clock to the FPGA clock via the Down Sample FIFO.

For the transmit channel, the PCS block receives the parallel data from the FPGA core, encodes it with 8b/10b, selects the polarity and passes the 8/10 bit data to the transmit SERDES channel.

The PCS also provides bypass modes that allow a direct 8-bit or 10-bit interface from the SERDES to the FPGA logic. The PCS interface to the FPGA can also be programmed to run at 1/2 speed for a 16-bit or 20-bit interface to the FPGA logic.

Some of the enhancements in LFE5UM/LFE5UM5G SERDES/PCS include:

- Higher clock/channel granularity: Dual channel architecture provides more clock resource per channel.
- Enhanced Tx de-emphasis: Programmable pre- and post-cursors improves Tx output signaling
- Bit-slip function in PCS: Improves logic needed to perform Word Alignment function

Refer to ECP5 and ECP5-5G SERDES/PCS Usage Guide (TN1261) for more information.

2.18. Device Configuration

All ECP5/ECP5-5G devices contain two ports that can be used for device configuration. The Test Access Port (TAP), which supports bit-wide configuration, and the sysCONFIG port, support dual-byte, byte and serial configuration. The TAP supports both the IEEE Standard 1149.1 Boundary Scan specification and the IEEE Standard 1532 In-System Configuration specification. There are 11 dedicated pins for TAP and sysConfig supports (TDI, TDO, TCK, TMS, CFG[2:0], PROGRAMN, DONE, INITN and CCLK). The remaining sysCONFIG pins are used as dual function pins. Refer to ECP5 and ECP5-5G sysCONFIG Usage Guide (TN1260) for more information about using the dual-use pins as general purpose I/Os.

There are various ways to configure an ECP5/ECP5-5G device:

- JTAG
- Standard Serial Peripheral Interface (SPI) – Interface to boot PROM Support x1, x2, x4 wide SPI memory interfaces.
- System microprocessor to drive a x8 CPU port SPCM mode
- System microprocessor to drive a serial slave SPI port (SSPI mode)
- Slave Serial model (SCM)

On power-up, the FPGA SRAM is ready to be configured using the selected sysCONFIG port. Once a configuration port is selected, it will remain active throughout that configuration cycle. The IEEE 1149.1 port can be activated any time after power-up by sending the appropriate command through the TAP port.

ECP5/ECP5-5G devices also support the Slave SPI Interface. In this mode, the FPGA behaves like a SPI Flash device (slave mode) with the SPI port of the FPGA to perform read-write operations.

2.18.1. Enhanced Configuration Options

ECP5/ECP5-5G devices have enhanced configuration features such as: decryption support, decompression support, TransFR™ I/O and dual-boot and multi-boot image support.

TransFR (Transparent Field Reconfiguration)

TransFR I/O (TFR) is a unique Lattice technology that allows users to update their logic in the field without interrupting system operation using a single ispVM command. TransFR I/O allows I/O states to be frozen during device configuration. This allows the device to be field updated with a minimum of system disruption and downtime. Refer to Minimizing System Interruption During Configuration Using TransFR Technology (TN1087) for details.

Dual-Boot and Multi-Boot Image Support

Dual-boot and multi-boot images are supported for applications requiring reliable remote updates of configuration data for the system FPGA. After the system is running with a basic configuration, a new boot image can be downloaded remotely and stored in a separate location in the configuration storage device. Any time after the update the ECP5/ECP5-5G devices can be re-booted from this new configuration file. If there is a problem, such as corrupt data during download or incorrect version number with this new boot image, the ECP5/ECP5-5G device can revert back to the original backup golden configuration and try again. This all can be done without power cycling the system. For more information, refer to ECP5 and ECP5-5G sysCONFIG Usage Guide (TN1260).

2.18.2. Single Event Upset (SEU) Support

ECP5/ECP5-5G devices support SEU mitigation with three supporting functions:

- SED – Soft Error Detect
- SEC – Soft Error Correction
- SEI – Soft Error Injection

ECP5/ECP5-5G devices have dedicated logic to perform Cycle Redundancy Code (CRC) checks. During configuration, the configuration data bitstream can be checked with the CRC logic block. In addition, the ECP5/ECP5-5G device can also be programmed to utilize a Soft Error Detect (SED) mode that checks for soft errors in configuration SRAM. The SED operation can be run in the background during user mode. If a soft error occurs, during user mode (normal operation) the device can be programmed to generate an error signal.

3.7. Hot Socketing Requirements

Table 3.6. Hot Socketing Requirements

Description	Min	Typ	Max	Unit
Input current per SERDES I/O pin when device is powered down and inputs driven.	—	—	8	mA
Input current per HDIN pin when device power supply is off, inputs driven[1, 2]	—	—	15	mA
Current per HDIN pin when device power ramps up, input driven[3]	—	—	50	mA
Current per HDOUT pin when device power supply is off, outputs pulled up[4]	—	—	30	mA

Notes:

1. Device is powered down with all supplies grounded, both HDINP and HDINN inputs driven by a CML driver with maximum allowed output V_{CCHTX}, 8b/10b data, no external AC coupling.
2. Each P and N input must have less than the specified maximum input current during hot plug. For a device with 2 DCU, the total input current would be 15 mA * 4 channels * 2 input pins per channel = 120 mA.
3. Device power supplies are ramping up (V_{CCA} and V_{CCAUX}), both HDINP and HDINN inputs are driven by a CML driver with maximum allowed output V_{CCHTX}, 8b/10b data, internal AC coupling.
4. Device is powered down with all supplies grounded. Both HDOUTP and HDOUN outputs are pulled up to V_{CCHTX} by the far end receiver termination of 50 Ω single ended.

3.8. ESD Performance

Refer to the ECP5 and ECP5-5G Product Family Qualification Summary for complete qualification data, including ESD performance.

3.9. DC Electrical Characteristics

Over Recommended Operating Conditions

Table 3.7. DC Electrical Characteristics

Symbol	Parameter	Condition	Min	Typ	Max	Unit
I_{IL}, I_{IH}[1, 4]	Input or I/O Low Leakage	$0 \le V_{IN} \le V_{CCIO}$	—	—	10	µA
I_{IH}[1, 3]	Input or I/O High Leakage	$V_{CCIO} < V_{IN} \le V_{IH(MAX)}$	—	—	100	µA
I_{PU}	I/O Active Pull-up Current, sustaining logic HIGH state	$0.7\ V_{CCIO} \le V_{IN} \le V_{CCIO}$	–30	—	—	µA
	I/O Active Pull-up Current, pulling down from logic HIGH state	$0 \le V_{IN} \le 0.7\ V_{CCIO}$	—	—	–150	µA
I_{PD}	I/O Active Pull-down Current, sustaining logic LOW state	$0 \le V_{IN} \le V_{IL}$ (MAX)	30	—	—	µA
	I/O Active Pull-down Current, pulling up from logic LOW state	$0 \le V_{IN} \le V_{CCIO}$	—	—	150	µA
C1	I/O Capacitance[2]	V_{CCIO} = 3.3 V, 2.5 V, 1.8 V, 1.5 V, 1.2 V, V_{CC} = 1.2 V, V_{IO} = 0 to $V_{IH(MAX)}$	—	5	8	pf
C2	Dedicated Input Capacitance[2]	V_{CCIO} = 3.3 V, 2.5 V, 1.8 V, 1.5 V, 1.2 V, V_{CC} = 1.2 V, V_{IO} = 0 to $V_{IH(MAX)}$	—	5	7	pf
V_{HYST}	Hysteresis for Single-Ended Inputs	V_{CCIO} = 3.3 V	—	300	—	mV
		V_{CCIO} = 2.5 V	—	250	—	mV

Notes:

1. Input or I/O leakage current is measured with the pin configured as an input or as an I/O with the output driver tristated. It is not measured with the output driver active. Bus maintenance circuits are disabled.
2. T_A 25 °C, f = 1.0 MHz.
3. Applicable to general purpose I/Os in top and bottom banks.
4. When used as V_{REF}, maximum leakage= 25 µA.

Table 3.10. ECP5-5G

Symbol	Description	Typ	Max	Unit
Standby (Power Down)				
I_{CCA-SB}	V_{CCA} Power Supply Current (Per Channel)	4	9.5	mA
$I_{CCHRX-SB}$[4]	V_{CCHRX}, Input Buffer Current (Per Channel)	—	0.1	mA
$I_{CCHTX-SB}$	V_{CCHTX}, Output Buffer Current (Per Channel)	—	0.9	mA
Operating (Data Rate = 5 Gb/s)				
I_{CCA-OP}	V_{CCA} Power Supply Current (Per Channel)	58	67	mA
$I_{CCHRX-OP}$[5]	V_{CCHRX}, Input Buffer Current (Per Channel)	0.4	0.5	mA
$I_{CCHTX-OP}$	V_{CCHTX}, Output Buffer Current (Per Channel)	10	13	mA
Operating (Data Rate = 3.2 Gb/s)				
I_{CCA-OP}	V_{CCA} Power Supply Current (Per Channel)	48	57	mA
$I_{CCHRX-OP}$[5]	V_{CCHRX}, Input Buffer Current (Per Channel)	0.4	0.5	mA
$I_{CCHTX-OP}$	V_{CCHTX}, Output Buffer Current (Per Channel)	10	13	mA
Operating (Data Rate = 2.5 Gb/s)				
I_{CCA-OP}	V_{CCA} Power Supply Current (Per Channel)	44	53	mA
$I_{CCHRX-OP}$[5]	V_{CCHRX}, Input Buffer Current (Per Channel)	0.4	0.5	mA
$I_{CCHTX-OP}$	V_{CCHTX}, Output Buffer Current (Per Channel)	10	13	mA
Operating (Data Rate = 1.25 Gb/s)				
I_{CCA-OP}	V_{CCA} Power Supply Current (Per Channel)	36	46	mA
$I_{CCHRX-OP}$[5]	V_{CCHRX}, Input Buffer Current (Per Channel)	0.4	0.5	mA
$I_{CCHTX-OP}$	V_{CCHTX}, Output Buffer Current (Per Channel)	10	13	mA
Operating (Data Rate = 270 Mb/s)				
I_{CCA-OP}	V_{CCA} Power Supply Current (Per Channel)	30	40	mA
$I_{CCHRX-OP}$[5]	V_{CCHRX}, Input Buffer Current (Per Channel)	0.4	0.5	mA
$I_{CCHTX-OP}$	V_{CCHTX}, Output Buffer Current (Per Channel)	8	10	mA

Notes:

1. Rx Equalization enabled, Tx De-emphasis (pre-cursor and post-cursor) disabled
2. Per Channel current is calculated with both channels on in a Dual, and divide current by two. If only one channel is on, current will be higher.
3. To calculate with Tx De-emphasis enabled, use the Diamond Power Calculator tool.
4. For $I_{CCHRX-SB}$, during Standby, input termination on Rx are disabled.
5. For $I_{CCHRX-OP}$, during operational, the max specified when external AC coupling is used. If externally DC coupled, the power is based on current pulled down by external driver when the input is driven to LOW.

3.17. Maximum I/O Buffer Speed

Over recommended operating conditions.

Table 3.21. ECP5/ECP5-5G Maximum I/O Buffer Speed

Buffer	Description	Max	Unit
Maximum Input Frequency			
LVDS25	LVDS, V_{CCIO} = 2.5 V	400	MHz
MLVDS25	MLVDS, Emulated, V_{CCIO} = 2.5 V	400	MHz
BLVDS25	BLVDS, Emulated, V_{CCIO} = 2.5 V	400	MHz
MIPI D-PHY (HS Mode)	MIPI Video	400	MHz
SLVS	SLVS similar to MIPI	400	MHz
Mini LVDS	Mini LVDS	400	MHz
LVPECL33	LVPECL, Emulated, V_{CCIO} = 3.3 V	400	MHz
SSTL18 (all supported classes)	SSTL_18 class I, II, V_{CCIO} = 1.8 V	400	MHz
SSTL15 (all supported classes)	SSTL_15 class I, II, V_{CCIO} = 1.5 V	400	MHz
SSTL135 (all supported classes)	SSTL_135 class I, II, V_{CCIO} = 1.35 V	400	MHz
HSUL12 (all supported classes)	HSUL_12 class I, II, V_{CCIO} = 1.2 V	400	MHz
LVTTL33	LVTTL, V_{CCIO} = 3.3 V	200	MHz
LVCMOS33	LVCMOS, V_{CCIO} = 3.3 V	200	MHz
LVCMOS25	LVCMOS, V_{CCIO} = 2.5 V	200	MHz
LVCMOS18	LVCMOS, V_{CCIO} = 1.8 V	200	MHz
LVCMOS15	LVCMOS 1.5, V_{CCIO} = 1.5 V	200	MHz
LVCMOS12	LVCMOS 1.2, V_{CCIO} = 1.2 V	200	MHz
Maximum Output Frequency			
LVDS25E	LVDS, Emulated, V_{CCIO} = 2.5 V	150	MHz
LVDS25	LVDS, V_{CCIO} = 2.5 V	400	MHz
MLVDS25	MLVDS, Emulated, V_{CCIO} = 2.5 V	150	MHz
BLVDS25	BLVDS, Emulated, V_{CCIO} = 2.5 V	150	MHz
LVPECL33	LVPECL, Emulated, V_{CCIO} = 3.3 V	150	MHz
SSTL18 (all supported classes)	SSTL_18 class I, II, V_{CCIO} = 1.8 V	400	MHz
SSTL15 (all supported classes)	SSTL_15 class I, II, V_{CCIO} = 1.5 V	400	MHz
SSTL135 (all supported classes)	SSTL_135 class I, II, V_{CCIO} = 1.35 V	400	MHz
HSUL12 (all supported classes)	HSUL12 class I, II, V_{CCIO} = 1.2 V	400	MHz
LVTTL33	LVTTL, VCCIO = 3.3 V	150	MHz
LVCMOS33 (For all drives)	LVCMOS, 3.3 V	150	MHz
LVCMOS25 (For all drives)	LVCMOS, 2.5 V	150	MHz
LVCMOS18 (For all drives)	LVCMOS, 1.8 V	150	MHz
LVCMOS15 (For all drives)	LVCMOS, 1.5 V	150	MHz
LVCMOS12 (For all drives)	LVCMOS, 1.2 V	150	MHz

Notes:

1. These maximum speeds are characterized but not tested on every device.
2. Maximum I/O speed for differential output standards emulated with resistors depends on the layout.
3. LVCMOS timing is measured with the load specified in Switching Test Conditions, Table 3.44 on page 90.
4. All speeds are measured at fast slew.
5. Actual system operation may vary depending on user logic implementation.
6. Maximum data rate equals 2 times the clock rate when utilizing DDR.

3.19. sysCLOCK PLL Timing

Over recommended operating conditions.

Table 3.23. sysCLOCK PLL Timing

Parameter	Descriptions	Conditions	Min	Max	Units
f_{IN}	Input Clock Frequency (CLKI, CLKFB)	—	8	400	MHz
f_{OUT}	Output Clock Frequency (CLKOP, CLKOS)	—	3.125	400	MHz
f_{VCO}	PLL VCO Frequency	—	400	800	MHz
f_{PFD}[3]	Phase Detector Input Frequency	—	10	400	MHz
AC Characteristics					
t_{DT}	Output Clock Duty Cycle	—	45	55	%
t_{PH4}	Output Phase Accuracy	—	–5	5	%
t_{OPJIT}[1]	Output Clock Period Jitter	$f_{OUT} \geq 100$ MHz	—	100	ps p-p
		$f_{OUT} < 100$ MHz	—	0.025	UIPP
	Output Clock Cycle-to-Cycle Jitter	$f_{OUT} \geq 100$ MHz	—	200	ps p-p
		$f_{OUT} < 100$ MHz	—	0.050	UIPP
	Output Clock Phase Jitter	$f_{PFD} \geq 100$ MHz	—	200	ps p-p
		$f_{PFD} < 100$ MHz	—	0.011	UIPP
t_{SPO}	Static Phase Offset	Divider ratio = integer	—	400	ps p-p
t_W	Output Clock Pulse Width	At 90% or 10%	0.9	—	ns
t_{LOCK}[2]	PLL Lock-in Time	—	—	15	ms
t_{UNLOCK}	PLL Unlock Time	—	—	50	ns
t_{IPJIT}	Input Clock Period Jitter	$f_{PFD} \geq 20$ MHz	—	1,000	ps p-p
		$f_{PFD} < 20$ MHz	—	0.02	UIPP
t_{HI}	Input Clock High Time	90% to 90%	0.5	—	ns
t_{LO}	Input Clock Low Time	10% to 10%	0.5	—	ns
t_{RST}	RST/ Pulse Width	—	1	—	ms
t_{RSTREC}	RST Recovery Time	—	1	—	ns
t_{LOAD_REG}	Min Pulse for CIB_LOAD_REG	—	10	—	ns
$t_{ROTATE-SETUP}$	Min time for CIB dynamic phase controls to be stable fore CIB_ROTATE	—	5	—	ns
$t_{ROTATE-WD}$	Min pulse width for CIB_ROTATE to maintain “0” or	—	4	—	VCO cycles

Notes:

1. Jitter sample is taken over 10,000 samples for Periodic jitter, and 2,000 samples for Cycle-to-Cycle jitter of the primary PLL output with clean reference clock with no additional I/O toggling.
2. Output clock is valid after t_{LOCK} for PLL reset and dynamic delay adjustment.
3. Period jitter and cycle-to-cycle jitter numbers are guaranteed for $f_{PFD} > 10$ MHz. For $f_{PFD} < 10$ MHz, the jitter numbers may not be met in certain conditions.

3.25. PCI Express Electrical and Timing Characteristics

3.25.1. PCIe (2.5 Gb/s) AC and DC Characteristics

Over recommended operating conditions.

Table 3.30. PCIe (2.5 Gb/s)

Symbol	Description	Test Conditions	Min	Typ	Max	Unit
Transmit[1]						
UI	Unit interval	—	399.88	400	400.12	ps
$V_{TX-DIFF\_P-P}$	Differential peak-to-peak output	—	0.8	1.0	1.2	V
$V_{TX-DE-RATIO}$	De-emphasis differential output voltage ratio	—	–3	–3.5	–4	dB
$V_{TX-CM-AC\_P}$	RMS AC peak common-mode output voltage	—	—	—	20	mV
$V_{TX-RCV-DETECT}$	Amount of voltage change allowed during receiver detection	—	—	—	600	mV
$V_{TX-CM-DC}$	Tx DC common mode voltage	—	0	—	V_{CCHTX}	V
$I_{TX-SHORT}$	Output short circuit current	V_{TX-D+}=0.0 V V_{TX-D-}=0.0 V	—	—	90	mA
$Z_{TX-DIFF-DC}$	Differential output impedance	—	80	100	120	Ω
$RL_{TX-DIFF}$	Differential return loss	—	10	—	—	dB
RL_{TX-CM}	Common mode return loss	—	6.0	—	—	dB
$T_{TX-RISE}$	Tx output rise time	20% to 80%	0.125	—	—	UI
$T_{TX-FALL}$	Tx output fall time	20% to 80%	0.125	—	—	UI
$L_{TX-SKEW}$	Lane-to-lane static output skew for all lanes in port/link	—	—	—	1.3	ns
T_{TX-EYE}	Transmitter eye width	—	0.75	—	—	UI
$T_{TX-EYE-MEDIAN-TO-MAX-JITTER}$	Maximum time between jitter median and maximum deviation from median	—	—	—	0.125	UI
Receive[1, 2]						
UI	Unit Interval	—	399.88	400	400.12	ps
$V_{RX-DIFF\_P-P}$	Differential peak-to-peak input voltage	—	0.34[3]	—	1.2	V
$V_{RX-IDLE-DET-DIFF\_P-P}$	Idle detect threshold voltage	—	65	—	340[3]	mV
$V_{RX-CM-AC\_P}$	RMS AC peak common-mode input voltage	—	—	—	150	mV
$Z_{RX-DIFF-DC}$	DC differential input impedance	—	80	100	120	Ω
Z_{RX-DC}	DC input impedance	—	40	50	60	Ω
$Z_{RX-HIGH-IMP-DC}$	Power-down DC input impedance	—	200K	—	—	Ω
$RL_{RX-DIFF}$	Differential return loss	—	10	—	—	dB
RL_{RX-CM}	Common mode return loss	—	6.0	—	—	dB

Notes:

1. Values are measured at 2.5 Gb/s.
2. Measured with external AC-coupling on the receiver.
3. Not in compliance with PCI Express 1.1 standard.

3.25.2. PCIe (5 Gb/s) – Preliminary AC and DC Characteristics

Over recommended operating conditions.

Table 3.31. PCIe (5 Gb/s)

Symbol	Description	Test Conditions	Min	Typ	Max	Unit
Transmit[1]						
UI	Unit Interval	—	199.94	200	200.06	ps
$B_{WTX-PKG-PLL2}$	Tx PLL bandwidth corresponding to PKGTX-PLL2	—	5	—	16	MHz
$P_{KGTX-PLL2}$	Tx PLL Peaking	—	—	—	1	dB
$V_{TX-DIFF-PP}$	Differential p-p Tx voltage swing	—	0.8	—	1.2	V, p-p
$V_{TX-DIFF-PP-LOW}$	Low power differential p-p Tx voltage swing	—	0.4	—	1.2	V, p-p
$V_{TX-DE-RATIO-3.5dB}$	Tx de-emphasis level ratio at 3.5dB	—	3	—	4	dB
$V_{TX-DE-RATIO-6dB}$	Tx de-emphasis level ratio at 6dB	—	5.5	—	6.5	dB
$T_{MIN-PULSE}$	Instantaneous lone pulse width	—		—	—	UI
$T_{TX-RISE-FALL}$	Transmitter rise and fall time	—		—	—	UI
T_{TX-EYE}	Transmitter Eye, including all jitter sources	—	0.75	—	—	UI
T_{TX-DJ}	Tx deterministic jitter > 1.5 MHz	—	—	—	0.15	UI
T_{TX-RJ}	Tx RMS jitter < 1.5 MHz	—	—	—	3	ps, RMS
$T_{RF-MISMATCH}$	Tx rise/fall time mismatch	—	—	—		UI
$R_{LTX-DIFF}$	Tx Differential Return Loss, including package and silicon	50 MHz < freq < 1.25 GHz	10	—	—	dB
		1.25 GHz < freq < 2.5 GHz	8	—	—	dB
R_{LTX-CM}	Tx Common Mode Return Loss, including package and silicon	50 MHz < freq < 2.5 GHz	6	—	—	dB
$Z_{TX-DIFF-DC}$	DC differential Impedance	—	—	—	120	Ω
$V_{TX-CM-AC-PP}$	Tx AC peak common mode voltage, peak-peak	—	—	—		mV, p-p
$I_{TX-SHORT}$	Transmitter short-circuit current	—	—	—	90	mA
$V_{TX-DC-CM}$	Transmitter DC common-mode voltage	—	0	—	1.2	V
$V_{TX-IDLE-DIFF-DC}$	Electrical Idle Output DC voltage	—	0	—	5	mV
$V_{TX-IDLE-DIFF-AC-p}$	Electrical Idle Differential Output peak voltage	—	—	—		mV
$V_{TX-RCV-DETECT}$	Voltage change allowed during Receiver Detect	—	—	—	600	mV
$T_{TX-IDLE-MIN}$	Min. time in Electrical Idle	—	20	—	—	ns
$T_{TX-IDLE-SET-TO-IDLE}$	Max. time from EI Order Set to valid Electrical Idle	—	—	—	8	ns
$T_{TX-IDLE-TO-DIFF-DATA}$	Max. time from Electrical Idle to valid differential output	—	—	—	8	ns
$L_{TX-SKEW}$	Lane-to-lane output skew	—	—	—		ps

3.27. XAUI/CPRI LV E.30 Electrical and Timing Characteristics

3.27.1. AC and DC Characteristics

Over recommended operating conditions.

Table 3.33. Transmit

Symbol	Description	Test Conditions	Min	Typ	Max	Unit
T_{RF}	Differential rise/fall time	20% to 80%	—	80	—	ps
$Z_{TX\_DIFF\_DC}$	Differential impedance	—	80	100	120	Ω
J_{TX_DDJ}[2, 3]	Output data deterministic jitter	—	—	—	0.17	UI
J_{TX_TJ}[1, 2, 3]	Total output data jitter	—	—	—	0.35	UI

Notes:
1. Total jitter includes both deterministic jitter and random jitter.
2. Jitter values are measured with each CML output AC coupled into a 50 Ω impedance (100 Ω differential impedance).
3. Jitter and skew are specified between differential crossings of the 50% threshold of the reference signal.

Over recommended operating conditions.

Table 3.34. Receive and Jitter Tolerance

Symbol	Description	Test Conditions	Min	Typ	Max	Unit
RL_{RX_DIFF}	Differential return loss	From 100 MHz to 3.125 GHz	10	—	—	dB
RL_{RX_CM}	Common mode return loss	From 100 MHz to 3.125 GHz	6	—	—	dB
Z_{RX_DIFF}	Differential termination resistance	—	80	100	120	Ω
J_{RX_DJ}[1, 2, 3]	Deterministic jitter tolerance (peak-to-peak)	—	—	—	0.37	UI
J_{RX_RJ}[1, 2, 3]	Random jitter tolerance (peak-to-peak)	—	—	—	0.18	UI
J_{RX_SJ}[1, 2, 3]	Sinusoidal jitter tolerance (peak-to-peak)	—	—	—	0.10	UI
J_{RX_TJ}[1, 2, 3]	Total jitter tolerance (peak-to-peak)	—	—	—	0.65	UI
T_{RX_EYE}	Receiver eye opening	—	0.35	—	—	UI

Notes:
1. Total jitter includes deterministic jitter, random jitter and sinusoidal jitter.
2. Jitter values are measured with each high-speed input AC coupled into a 50 Ω impedance.
3. Jitter and skew are specified between differential crossings of the 50% threshold of the reference signal.

3.28. CPRI LV E.24/SGMII(2.5Gbps) Electrical and Timing Characteristics

3.28.1. AC and DC Characteristics

Table 3.35. Transmit

Symbol	Description	Test Conditions	Min	Typ	Max	Unit
T_{RF}[1]	Differential rise/fall time	20% to 80%	—	80	—	ps
$Z_{TX\_DIFF\_DC}$	Differential impedance	—	80	100	120	Ω
J_{TX_DDJ}[3, 4]	Output data deterministic jitter	—	—	—	0.17	UI
J_{TX_TJ}[2, 3, 4]	Total output data jitter	—	—	—	0.35	UI

Notes:
1. Rise and Fall times measured with board trace, connector and approximately 2.5 pf load.
2. Total jitter includes both deterministic jitter and random jitter. The random jitter is the total jitter minus the actual deterministic jitter.
3. Jitter values are measured with each CML output AC coupled into a 50 Ω impedance (100 Ω differential impedance).
4. Jitter and skew are specified between differential crossings of the 50% threshold of the reference signal.

Table 3.36. Receive and Jitter Tolerance

Symbol	Description	Test Conditions	Min	Typ	Max	Unit
RL_{RX_DIFF}	Differential return loss	From 100 MHz to 2.5 GHz	10	—	—	dB
RL_{RX_CM}	Common mode return loss	From 100 MHz to 2.5 GHz	6	—	—	dB
Z_{RX_DIFF}	Differential termination resistance	—	80	100	120	Ω
J_{RX_DJ}[2, 3, 4]	Deterministic jitter tolerance (peak-to-peak)	—	—	—	0.37	UI
J_{RX_RJ}[2, 3, 4]	Random jitter tolerance (peak-to-peak)	—	—	—	0.18	UI
J_{RX_SJ}[2, 3, 4]	Sinusoidal jitter tolerance (peak-to-peak)	—	—	—	0.10	UI
J_{RX_TJ}[1, 2, 3, 4]	Total jitter tolerance (peak-to-peak)	—	—	—	0.65	UI
T_{RX_EYE}	Receiver eye opening	—	0.35	—	—	UI

Notes:
1. Total jitter includes deterministic jitter, random jitter and sinusoidal jitter.
2. Jitter values are measured with each high-speed input AC coupled into a 50 Ω impedance.
3. Jitter and skew are specified between differential crossings of the 50% threshold of the reference signal.
4. Jitter tolerance, Differential Input Sensitivity and Receiver Eye Opening parameters are characterized when Full Rx Equalization is enabled.

3.29. Gigabit Ethernet/SGMII(1.25Gbps)/CPRI LV E.12 Electrical and Timing Characteristics

3.29.1. AC and DC Characteristics

Table 3.37. Transmit

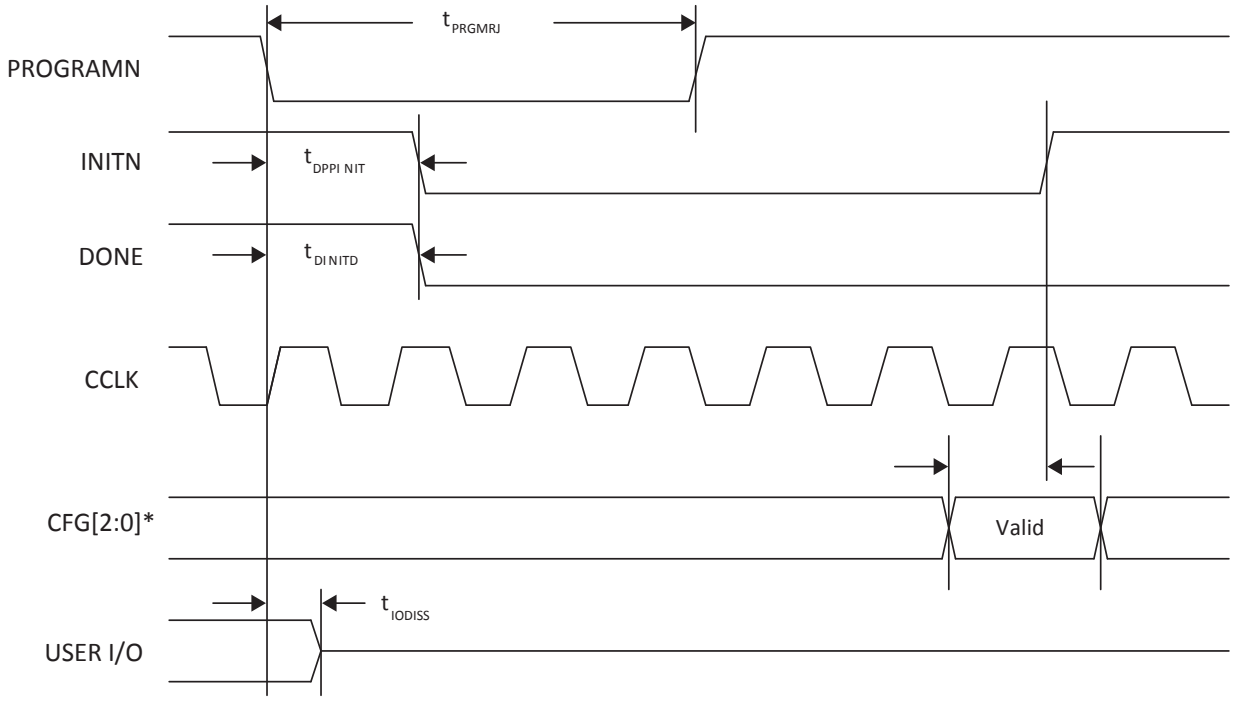
Symbol	Description	Test Conditions	Min	Typ	Max	Unit
T_{RF}	Differential rise/fall time	20% to 80%	—	80	—	ps
$Z_{TX\_DIFF\_DC}$	Differential impedance	—	80	100	120	Ω
J_{TX_DDJ}[2, 3]	Output data deterministic jitter	—	—	—	0.10	UI
J_{TX_TJ}[1, 2, 3]	Total output data jitter	—	—	—	0.24	UI

Notes:
1. Total jitter includes both deterministic jitter and random jitter. The random jitter is the total jitter minus the actual deterministic jitter.
2. Jitter values are measured with each CML output AC coupled into a 50 Ω impedance (100 Ω differential impedance).
3. Jitter and skew are specified between differential crossings of the 50% threshold of the reference signal.

Table 3.38. Receive and Jitter Tolerance

Symbol	Description	Test Conditions	Min	Typ	Max	Unit
RL_{RX_DIFF}	Differential return loss	From 100 MHz to 1.25 GHz	10	—	—	dB
RL_{RX_CM}	Common mode return loss	From 100 MHz to 1.25 GHz	6	—	—	dB
Z_{RX_DIFF}	Differential termination resistance	—	80	100	120	Ω
J_{RX_DJ}[1, 2, 3, 4]	Deterministic jitter tolerance (peak-to-peak)	—	—	—	0.34	UI
J_{RX_RJ}[1, 2, 3, 4]	Random jitter tolerance (peak-to-peak)	—	—	—	0.26	UI
J_{RX_SJ}[1, 2, 3, 4]	Sinusoidal jitter tolerance (peak-to-peak)	—	—	—	0.11	UI
J_{RX_TJ}[1, 2, 3, 4]	Total jitter tolerance (peak-to-peak)	—	—	—	0.71	UI
T_{RX_EYE}	Receiver eye opening	—	0.29	—	—	UI

Notes:
1. Total jitter includes deterministic jitter, random jitter and sinusoidal jitter.
2. Jitter values are measured with each high-speed input AC coupled into a 50 Ω impedance.
3. Jitter and skew are specified between differential crossings of the 50% threshold of the reference signal.
4. Jitter tolerance, Differential Input Sensitivity and Receiver Eye Opening parameters are characterized when Full Rx Equalization is enabled.



*The CFG pins are normally static (hardwired).

Figure 3.20. Configuration from PROGRAMN Timing

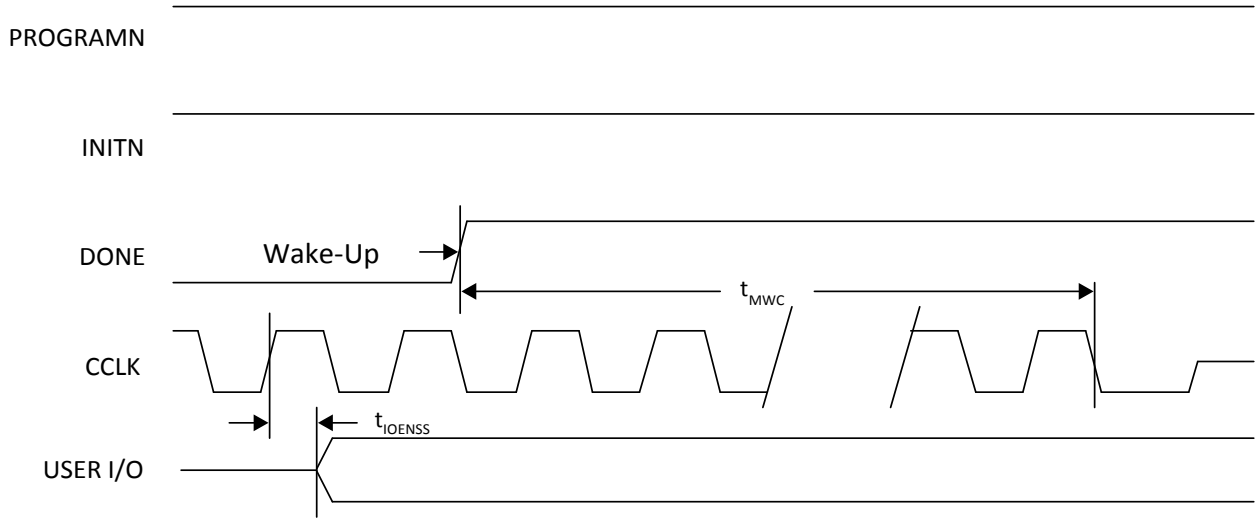
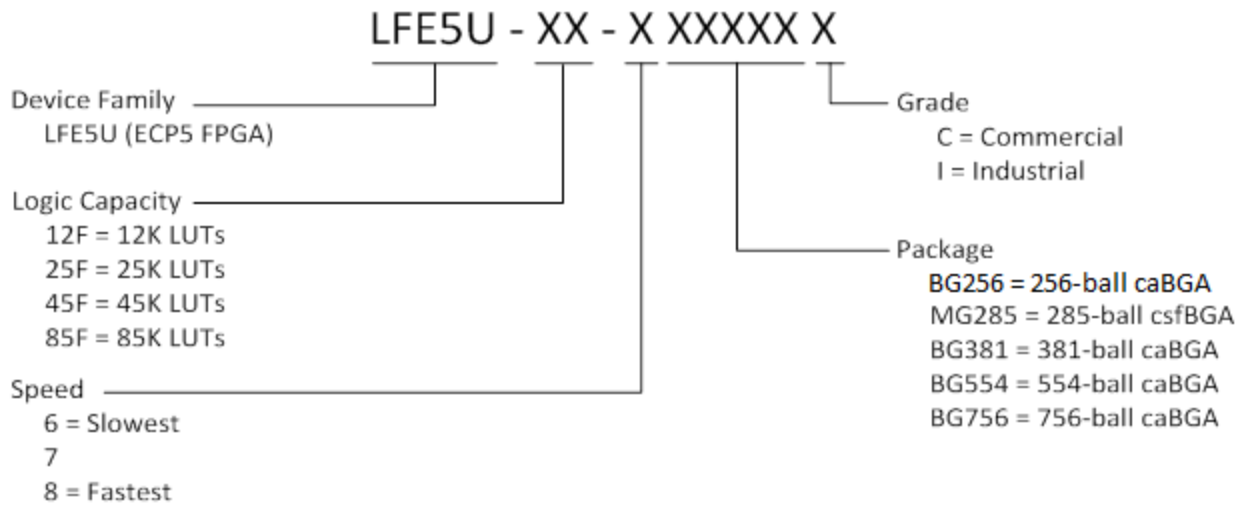
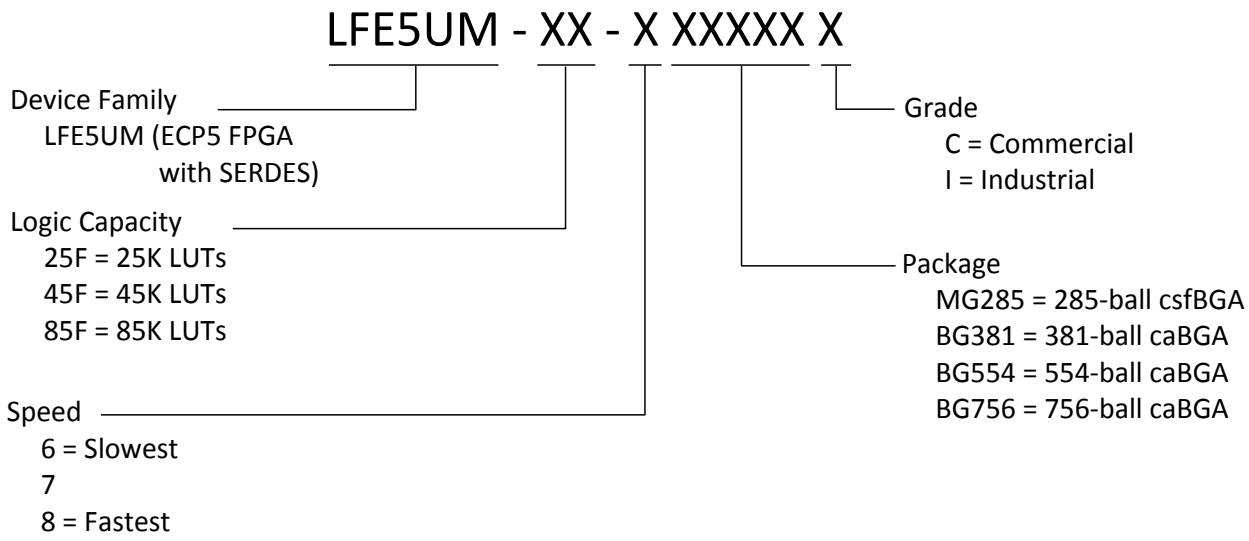


Figure 3.21. Wake-Up Timing

5. Ordering Information

5.1. ECP5/ECP5-5G Part Number Description





Part number	Grade	Package	Pins	Temp.	LUTs (K)	SERDES
LFE5UM5G-85F-8BG381C	–8	Lead free caBGA	381	Commercial	84	Yes
LFE5UM5G-85F-8BG554C	–8	Lead free caBGA	554	Commercial	84	Yes
LFE5UM5G-85F-8BG756C	–8	Lead free caBGA	756	Commercial	84	Yes

5.2.2. Industrial

Part number	Grade	Package	Pins	Temp.	LUTs (K)	SERDES
LFE5U-12F-6BG256I	–6	Lead free caBGA	256	Industrial	12	No
LFE5U-12F-7BG256I	–7	Lead free caBGA	256	Industrial	12	No
LFE5U-12F-8BG256I	–8	Lead free caBGA	256	Industrial	12	No
LFE5U-12F-6MG285I	–6	Lead free csfBGA	285	Industrial	12	No
LFE5U-12F-7MG285I	–7	Lead free csfBGA	285	Industrial	12	No
LFE5U-12F-8MG285I	–8	Lead free csfBGA	285	Industrial	12	No
LFE5U-12F-6BG381I	–6	Lead free caBGA	381	Industrial	12	No
LFE5U-12F-7BG381I	–7	Lead free caBGA	381	Industrial	12	No
LFE5U-12F-8BG381I	–8	Lead free caBGA	381	Industrial	12	No
LFE5U-25F-6BG256I	–6	Lead free caBGA	256	Industrial	24	No
LFE5U-25F-7BG256I	–7	Lead free caBGA	256	Industrial	24	No
LFE5U-25F-8BG256I	–8	Lead free caBGA	256	Industrial	24	No
LFE5U-25F-6MG285I	–6	Lead free csfBGA	285	Industrial	24	No
LFE5U-25F-7MG285I	–7	Lead free csfBGA	285	Industrial	24	No
LFE5U-25F-8MG285I	–8	Lead free csfBGA	285	Industrial	24	No
LFE5U-25F-6BG381I	–6	Lead free caBGA	381	Industrial	24	No
LFE5U-25F-7BG381I	–7	Lead free caBGA	381	Industrial	24	No
LFE5U-25F-8BG381I	–8	Lead free caBGA	381	Industrial	24	No
LFE5U-45F-6BG256I	–6	Lead free caBGA	256	Industrial	44	No
LFE5U-45F-7BG256I	–7	Lead free caBGA	256	Industrial	44	No
LFE5U-45F-8BG256I	–8	Lead free caBGA	256	Industrial	44	No
LFE5U-45F-6MG285I	–6	Lead free csfBGA	285	Industrial	44	No
LFE5U-45F-7MG285I	–7	Lead free csfBGA	285	Industrial	44	No
LFE5U-45F-8MG285I	–8	Lead free csfBGA	285	Industrial	44	No
LFE5U-45F-6BG381I	–6	Lead free caBGA	381	Industrial	44	No
LFE5U-45F-7BG381I	–7	Lead free caBGA	381	Industrial	44	No
LFE5U-45F-8BG381I	–8	Lead free caBGA	381	Industrial	44	No
LFE5U-45F-6BG554I	–6	Lead free caBGA	554	Industrial	44	No
LFE5U-45F-7BG554I	–7	Lead free caBGA	554	Industrial	44	No
LFE5U-45F-8BG554I	–8	Lead free caBGA	554	Industrial	44	No
LFE5U-85F-6MG285I	–6	Lead free csfBGA	285	Industrial	84	No
LFE5U-85F-7MG285I	–7	Lead free csfBGA	285	Industrial	84	No
LFE5U-85F-8MG285I	–8	Lead free csfBGA	285	Industrial	84	No
LFE5U-85F-6BG381I	–6	Lead free caBGA	381	Industrial	84	No
LFE5U-85F-7BG381I	–7	Lead free caBGA	381	Industrial	84	No
LFE5U-85F-8BG381I	–8	Lead free caBGA	381	Industrial	84	No
LFE5U-85F-6BG554I	–6	Lead free caBGA	554	Industrial	84	No
LFE5U-85F-7BG554I	–7	Lead free caBGA	554	Industrial	84	No
LFE5U-85F-8BG554I	–8	Lead free caBGA	554	Industrial	84	No

Supplemental Information

For Further Information

A variety of technical notes for the ECP5/ECP5-5G family are available.

- High-Speed PCB Design Considerations (TN1033)
- Transmission of High-Speed Serial Signals Over Common Cable Media (TN1066)
- PCB Layout Recommendations for BGA Packages (TN1074)
- Minimizing System Interruption During Configuration Using TransFR Technology (TN1087)
- Electrical Recommendations for Lattice SERDES (FPGA-TN-02077)
- LatticeECP3, ECP-5 and ECP5-5G Soft Error Detection (SED)/Correction (SEC) Usage Guide (TN1184)
- Using TraceID (TN1207)
- Sub-LVDS Signaling Using Lattice Devices (TN1210)
- Advanced Security Encryption Key Programming Guide for ECP5, ECP5-5G, LatticeECP3, and LatticeECP2/MS Devices (TN1215)
- LatticeECP3, LatticeECP2/M, ECP5 and ECP5-5G Dual Boot and Multiple Boot Feature (TN1216)
- ECP5 and ECP5-5G sysCONFIG Usage Guide (TN1260)
- ECP5 and ECP5-5G SERDES/PCS Usage Guide (TN1261)
- ECP5 and ECP5-5G sysIO Usage Guide (TN1262)
- ECP5 and ECP5-5G sysClock PLL/DLL Design and Usage Guide (TN1263)
- ECP5 and ECP5-5G Memory Usage Guide (TN1264)
- ECP5 and ECP5-5G High-Speed I/O Interface (TN1265)
- Power Consumption and Management for ECP5 and ECP5-5G Devices (TN1266)
- ECP5 and ECP5-5G sysDSP Usage Guide (TN1267)
- ECP5 and ECP5-5G Hardware Checklist (FPGA-TN-02038)
- Solder Reflow Guide for Surface Mount Devices (FPGA-TN-02041)
- ECP5 and ECP5-5G PCI Express Soft IP Ease of Use Guidelines (FPGA-TN-02045)
- Programming External SPI Flash through JTAG for ECP5/ECP5-5G (FPGA-TN-02050)
- Adding Scalable Power and Thermal Management to ECP5 Using L-ASC10 (AN6095)

For further information on interface standards refer to the following websites:

- JEDEC Standards (LVTTL, LVCMOS, SSTL): www.jedec.org
- PCI: www.pcisig.com

(Continued)

Date	Version	Section	Change Summary
August 2014	1.2	All	Changed document status from Advance to Preliminary.
		General Description	Updated Features section. • Deleted Serial RapidIO protocol under Embedded SERDES. • Corrected data rate under Pre-Engineered Source Synchronous
			Changed DD3. LPDDR3 to DDR2/3, LPDDR2/3. Mentioned transmit de-emphasis “pre- and post-cursors”.
		Architecture	Updated Overview section. • Revised description of PFU blocks. • Specified SRAM cell settings in describing the control of SERDES/PCS duals.
			Updated SERDES and Physical Coding Sublayer section. • Changed PCI Express 2.0 to PCI Express Gen1 and Gen2. • Deleted Serial RapidIO protocol. • Updated Table 2.13. LFE5UM/LFE5UM5G SERDES Standard Support.
			Updated On-Chip Oscillator section. • Deleted “130 MHz ±15% CMOS” oscillator. • Updated Table 2.16. Selectable Master Clock (MCLK) Frequencies during Configuration (Nominal)
		DC and Switching Characteristics	Updated Absolute Maximum Ratings section. Added supply voltages V_{CCA} and V_{CCAUXA}.
			Updated sysI/O Recommended Operating Conditions section. Revised HSULD12D VCCIO values and removed table note.
			Updated sysI/O Single-Ended DC Electrical Characteristics section. Revised some values for SSTL15 _I, SSTL15 _II, SSTL135_I, SSTL15_II, and HSUL12.
			Updated External Switching Characteristics section. Changed parameters to t_{SKEW_PR} V_{CCA} and t_{SKEW_EDGE} and added LFE5-85 as device.
			Updated ECP5 Family Timing Adders section. Added SSTL135_II buffer type data. Removed LVCMOS33_20mA, LVCMOS25_20mA, LVCMOS25_16mA, LVCMOS25D_16mA, and LVCMOS18_16mA buffer types. Changed buffer type to LVCMOS12_4mA and LVCMOS12_8mA.
			Updated Maximum I/O Buffer Speed section. Revised Max values.
			Updated sysCLOCK PLL Timing section. Revised t_{DT} Min and Max values. Revised t_{OPJIT} Max value. Revised number of samples in table note 1.
			Updated SERDES High-Speed Data Transmitter section. Updated Table 3.24. Serial Output Timing and Levels and Table 3.25. Channel Output Jitter.