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## Understanding [Embedded - FPGAs \(Field Programmable Gate Array\)](#)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

## Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications,

### Details

Product Status	Active
Number of LABs/CLBs	21000
Number of Logic Elements/Cells	84000
Total RAM Bits	3833856
Number of I/O	118
Number of Gates	-
Voltage - Supply	1.045V ~ 1.155V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	285-LFBGA, CSPBGA
Supplier Device Package	285-CSFBGA (10x10)
Purchase URL	<a href="https://www.e-xfl.com/product-detail/lattice-semiconductor/lfe5um-85f-7mg285c">https://www.e-xfl.com/product-detail/lattice-semiconductor/lfe5um-85f-7mg285c</a>

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## 2.2.2. Modes of Operation

Slices 0-2 have up to four potential modes of operation: Logic, Ripple, RAM and ROM. Slice 3 is not needed for RAM mode, it can be used in Logic, Ripple, or ROM modes.

### Logic Mode

In this mode, the LUTs in each slice are configured as 4-input combinatorial lookup tables. A LUT4 can have 16 possible input combinations. Any four input logic functions can be generated by programming this lookup table. Since there are two LUT4s per slice, a LUT5 can be constructed within one slice. Larger look-up tables such as LUT6, LUT7 and LUT8 can be constructed by concatenating other slices. Note that LUT8 requires more than four slices.

### Ripple Mode

Ripple mode supports the efficient implementation of small arithmetic functions. In ripple mode, the following functions can be implemented by each slice:

- Addition 2-bit
- Subtraction 2-bit
- Add/Subtract 2-bit using dynamic control
- Up counter 2-bit
- Down counter 2-bit
- Up/Down counter with asynchronous clear
- Up/Down counter with preload (sync)
- Ripple mode multiplier building block
- Multiplier support
- Comparator functions of A and B inputs
  - A greater-than-or-equal-to B
  - A not-equal-to B
  - A less-than-or-equal-to B

Ripple Mode includes an optional configuration that performs arithmetic using fast carry chain methods. In this configuration (also referred to as CCU2 mode) two additional signals, Carry Generate and Carry Propagate, are generated on a per slice basis to allow fast arithmetic functions to be constructed by concatenating Slices.

### RAM Mode

In this mode, a 16x4-bit distributed single port RAM (SPR) can be constructed in one PFU using each LUT block in Slice 0 and Slice 1 as a 16 x 2-bit memory in each slice. Slice 2 is used to provide memory address and control signals.

A 16 x 2-bit pseudo dual port RAM (PDPR) memory is created in one PFU by using one Slice as the read-write port and the other companion slice as the read-only port. The slice with the read-write port updates the SRAM data contents in both slices at the same write cycle.

ECP5/ECP5-5G devices support distributed memory initialization.

The Lattice design tools support the creation of a variety of different size memories. Where appropriate, the software will construct these using distributed memory primitives that represent the capabilities of the PFU. [Table 2.3](#) lists the number of slices required to implement different distributed RAM primitives. For more information about using RAM in ECP5/ECP5-5G devices, refer to [ECP5 and ECP5-5G Memory Usage Guide \(TN1264\)](#).

**Table 2.3. Number of Slices Required to Implement Distributed RAM**

	SPR 16 X 4	PDPR 16 X 4
Number of slices	3	6

Note: SPR = Single Port RAM, PDPR = Pseudo Dual Port RAM

### ROM Mode

ROM mode uses the LUT logic; hence, Slices 0 through 3 can be used in ROM mode. Preloading is accomplished through the programming interface during PFU configuration.

For more information, refer to [ECP5 and ECP5-5G Memory Usage Guide \(TN1264\)](#).

**Table 2.4** provides a description of the signals in the PLL blocks.

**Table 2.4. PLL Blocks Signal Descriptions**

Signal	Type	Description
CLKI	Input	Clock Input to PLL from external pin or routing
CLKI2	Input	Muxed clock input to PLL
SEL	Input	Input Clock select, selecting from CLKI and CLKI2 inputs
CLKFB	Input	PLL Feedback Clock
PHASESEL[1:0]	Input	Select which output to be adjusted on Phase by PHASEDIR, PHASESTEP, PHASELOADREG
PHASEDIR	Input	Dynamic Phase adjustment direction.
PHASESTEP	Input	Dynamic Phase adjustment step.
PHASELOADREG	Input	Load dynamic phase adjustment values into PLL.
CLKOP	Output	Primary PLL output clock (with phase shift adjustment)
CLKOS	Output	Secondary PLL output clock (with phase shift adjust)
CLKOS2	Output	Secondary PLL output clock2 (with phase shift adjust)
CLKOS3	Output	Secondary PLL output clock3 (with phase shift adjust)
LOCK	Output	PLL LOCK to CLKI, Asynchronous signal. Active high indicates PLL lock
STDBY	Input	Standby signal to power down the PLL
RST	Input	Resets the PLL
ENCLKOP	Input	Enable PLL output CLKOP
ENCLKOS	Input	Enable PLL output CLKOS
ENCLKOS2	Input	Enable PLL output CLKOS2
ENCLKOS3	Input	Enable PLL output CLKOS3

For more details on the PLL you can refer to the [ECP5 and ECP5-5G sysClock PLL/DLL Design and Usage Guide \(TN1263\)](#).

## 2.5. Clock Distribution Network

There are two main clock distribution networks for any member of the ECP5/ECP5-5G product family, namely Primary Clock (PCLK) and Edge Clock (ECLK). These clock networks have the clock sources come from many different sources, such as Clock Pins, PLL outputs, DLLDEL outputs, Clock divider outputs, SERDES/PCS clocks and some on chip generated clock signal. There are clock dividers (CLKDIV) blocks to provide the slower clock from these clock sources.

ECP5/ECP5-5G also supports glitchless dynamic enable function (DCC) for the PCLK Clock to save dynamic power. There are also some logics to allow dynamic glitchless selection between two clocks for the PCLK network (DCS).

Overview of Clocking Network is shown in [Figure 2.6](#) on page 20 for LFE5UM/LFE5UM5G-85 device.

### 2.5.1.2. Dynamic Clock Select

The Dynamic Clock Select (DCS) is a smart multiplexer function available in the primary clock routing. It switches between two independent input clock sources. Depending on the operation modes, it switches between two (2) independent input clock sources either with or without any glitches. This is achieved regardless of when the select signal is toggled. Both input clocks must be running to achieve functioning glitch-less DCS output clock, but it is not required running clocks when used as non-glitch-less normal clock multiplexer.

There are two DCS blocks per device that are fed to all quadrants. The inputs to the DCS block come from all the output of MIDMUXs and Clock from CIB located at the center of the PLC array core. The output of the DCS is connected to one of the inputs of Primary Clock Center MUX.

[Figure 2.7](#) shows the timing waveforms of the default DCS operating mode. The DCS block can be programmed to other modes. For more information about the DCS, refer to [ECP5 and ECP5-5G sysClock PLL/DLL Design and Usage Guide \(TN1263\)](#).

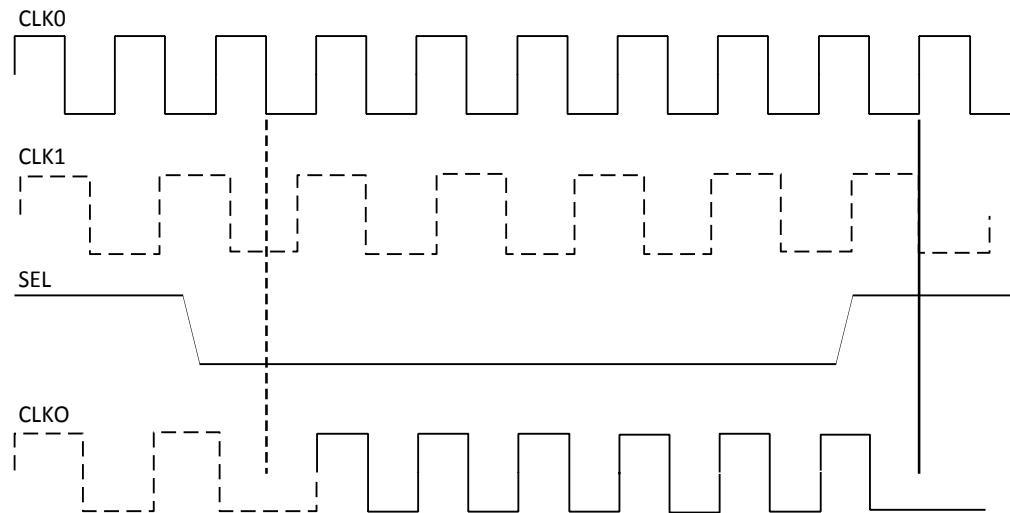


Figure 2.7. DCS Waveforms

### 2.5.2. Edge Clock

ECP5/ECP5-5G devices have a number of high-speed edge clocks that are intended for use with the PIOs in the implementation of high-speed interfaces. There are two ECLK networks per bank IO on the Left and Right sides of the devices.

Each Edge Clock can be sourced from the following:

- Dedicated Clock input pins (PCLK)
- DLLDEL output (Clock delayed by 90°)
- PLL outputs (CLKOP and CLKOS)
- ECLKBRIDGE
- Internal Nodes

In [Figure 2.15](#), note that A\_ALU, B\_ALU and C\_ALU are internal signals generated by combining bits from AA, AB, BA BB and C inputs. For further information, refer to [ECP5 and ECP5-5G sysDSP Usage Guide \(TN1267\)](#).

The ECP5/ECP5-5G sysDSP block supports the following basic elements.

- MULT (Multiply)
- MAC (Multiply, Accumulate)
- MULTADDSUB (Multiply, Addition/Subtraction)
- MULTADDSUBSUM (Multiply, Addition/Subtraction, Summation)

[Table 2.7](#) shows the capabilities of each of the ECP5/ECP5-5G slices versus the above functions.

**Table 2.7. Maximum Number of Elements in a Slice**

Width of Multiply	x9	x18	x36
MULT	4	2	1/2
MAC	1	1	—
MULTADDSUB	2	1	—
MULTADDSUBSUM	1*	1/2	—

\*Note: One slice can implement 1/2 9x9 m9x9addsubsum and two m9x9addsubsum with two slices.

Some options are available in the four elements. The input register in all the elements can be directly loaded or can be loaded as a shift register from previous operand registers. By selecting “dynamic operation” the following operations are possible:

- In the Add/Sub option the Accumulator can be switched between addition and subtraction on every cycle.
- The loading of operands can switch between parallel and serial operations.

For further information, refer to [ECP5 and ECP5-5G sysDSP Usage Guide \(TN1267\)](#).

## 2.10. Programmable I/O Cells

The programmable logic associated with an I/O is called a PIO. The individual PIO are connected to their respective sysIO buffers and pads. On the ECP5/ECP5-5G devices, the Programmable I/O cells (PIC) are assembled into groups of four PIO cells called a Programmable I/O Cell or PIC. The PICs are placed on all four sides of the device.

On all the ECP5/ECP5-5G devices, two adjacent PIOs can be combined to provide a complementary output driver pair. All PIO pairs can implement differential receivers. Half of the PIO pairs on the left and right edges of these devices can be configured as true LVDS transmit pairs.

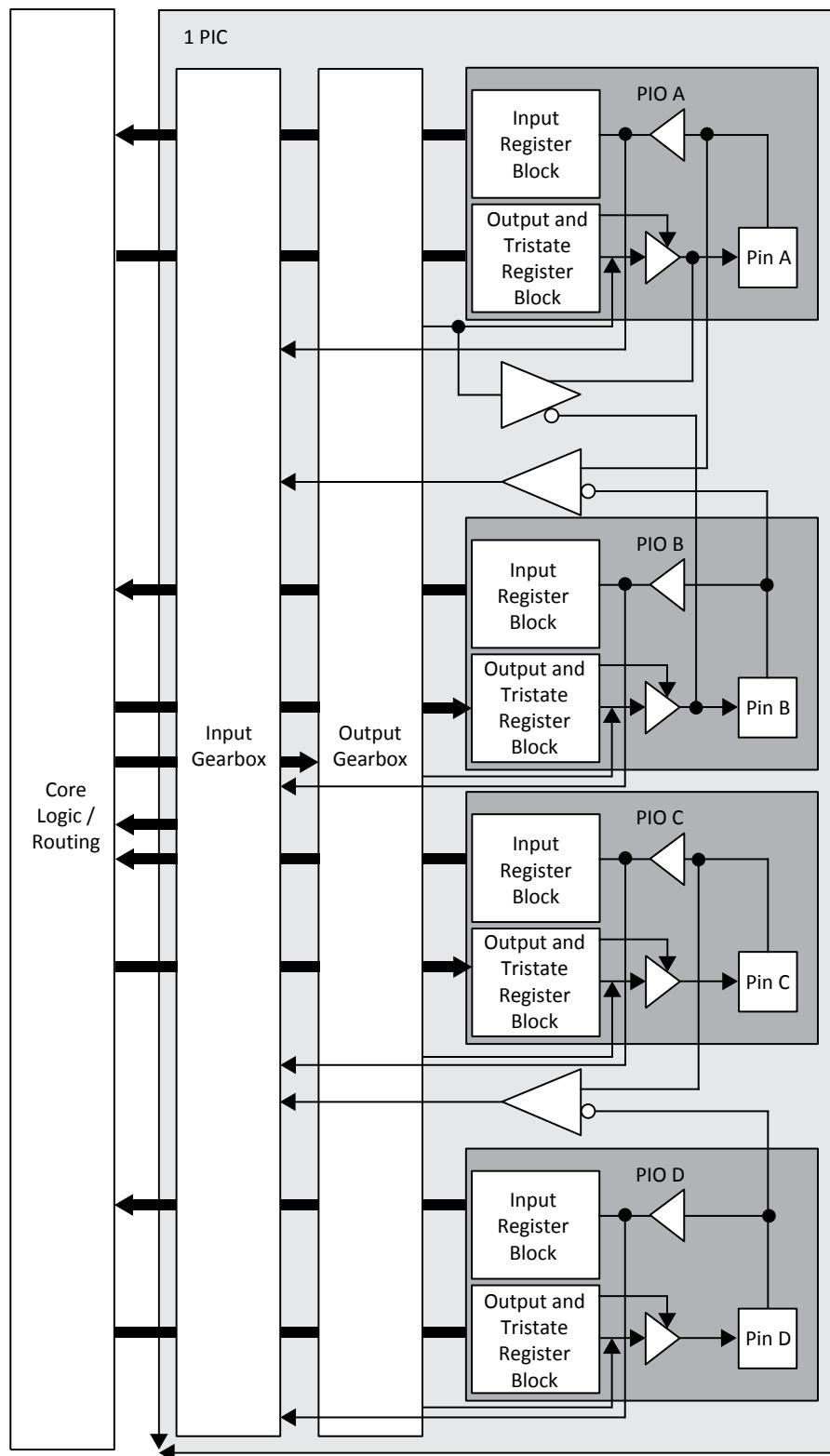
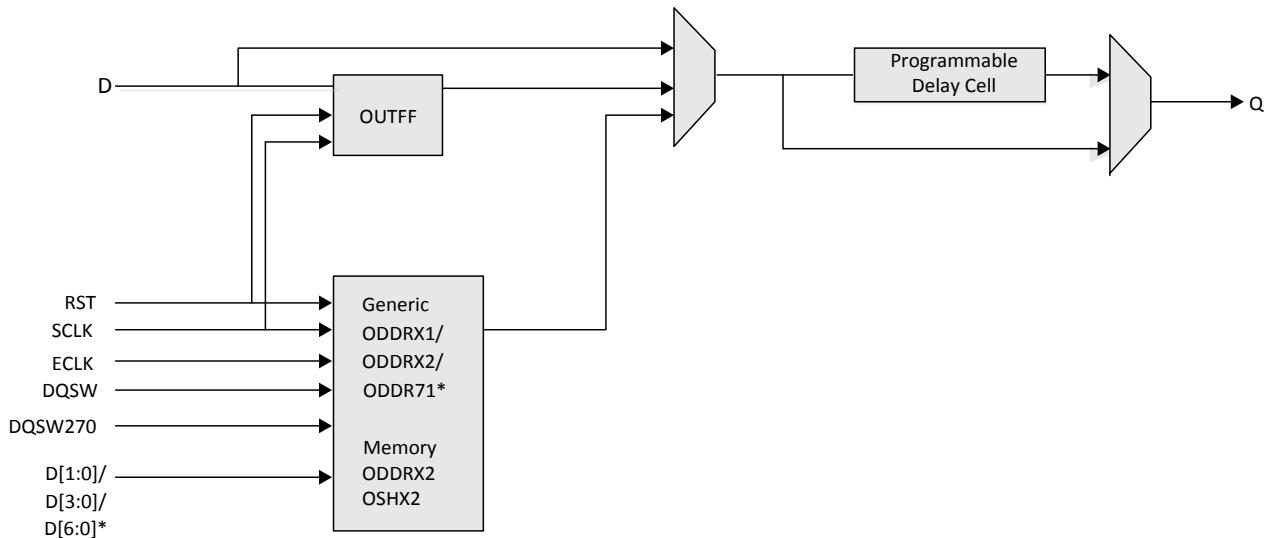


Figure 2.16. Group of Four Programmable I/O Cells on Left/Right Sides



\*For 7:1 LVDS interface only. It is required to use PIO pair pins PIOA/B.

**Figure 2.20. Output Register Block on Left and Right Sides**

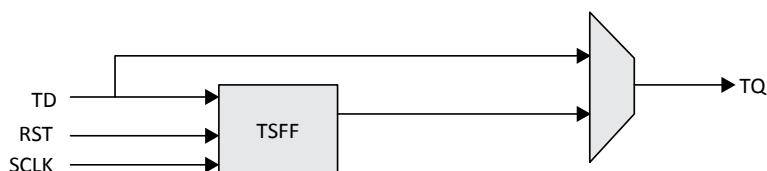
**Table 2.9. Output Block Port Description**

Name	Type	Description
Q	Output	High Speed Data Output
D	Input	Data from core to output SDR register
D[1:0]/D[3:0]/ D[6:0]	Input	Low Speed Data from device core to output DDR register
RST	Input	Reset to the Output Block
SCLK	Input	Slow Speed System Clock
ECLK	Input	High Speed Edge Clock
DQSW	Input	Clock from DQS control Block used to generate DDR memory DQS output
DQSW270	Input	Clock from DQS control Block used to generate DDR memory DQ output

## 2.12. Tristate Register Block

The tristate register block registers tristate control signals from the core of the device before they are passed to the sysIO buffers. The block contains a register for SDR operation. In SDR, TD input feeds one of the flip-flops that then feeds the output. In DDR operation used mainly for DDR memory interface can be implemented on the left and right sides of the device. Here two inputs feed the tristate registers clocked by both ECLK and SCLK.

Figure 2.21 and Figure 2.22 show the Tristate Register Block functions on the device. For detailed description of the tristate register block modes and usage, refer to [ECP5 and ECP5-5G High-Speed I/O Interface \(TN1265\)](#).



**Figure 2.21. Tristate Register Block on Top Side**

## 3. DC and Switching Characteristics

### 3.1. Absolute Maximum Ratings

**Table 3.1. Absolute Maximum Ratings**

Symbol	Parameter	Min	Max	Unit
$V_{CC}$	Supply Voltage	-0.5	1.32	V
$V_{CCA}$	Supply Voltage	-0.5	1.32	V
$V_{CCAUX}, V_{CCAUXA}$	Supply Voltage	-0.5	2.75	V
$V_{CCIO}$	Supply Voltage	-0.5	3.63	V
—	Input or I/O Transient Voltage Applied	-0.5	3.63	V
$V_{CCHRX}, V_{CCHTX}$	SERDES RX/TX Buffer Supply Voltages	-0.5	1.32	V
—	Voltage Applied on SERDES Pins	-0.5	1.80	V
$T_A$	Storage Temperature (Ambient)	-65	150	°C
$T_J$	Junction Temperature	—	+125	°C

**Notes:**

1. Stress above those listed under the “Absolute Maximum Ratings” may cause permanent damage to the device. Functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.
2. Compliance with the Lattice [Thermal Management](#) document is required.
3. All voltages referenced to GND.

### 3.2. Recommended Operating Conditions

**Table 3.2. Recommended Operating Conditions**

Symbol	Parameter		Min	Max	Unit
$V_{CC}^2$	Core Supply Voltage	ECP5	1.045	1.155	V
		ECP5-5G	1.14	1.26	V
$V_{CCAUX}^{2,4}$	Auxiliary Supply Voltage	—	2.375	2.625	V
$V_{CCIO}^{2,3}$	I/O Driver Supply Voltage	—	1.14	3.465	V
$V_{REF}^1$	Input Reference Voltage	—	0.5	1.0	V
$t_{JCOM}$	Junction Temperature, Commercial Operation	—	0	85	°C
$t_{JIND}$	Junction Temperature, Industrial Operation	—	-40	100	°C
<b>SERDES External Power Supply<sup>5</sup></b>					
$V_{CCA}$	SERDES Analog Power Supply	ECP5UM	1.045	1.155	V
		ECP5-5G	1.164	1.236	V
$V_{CCAUXA}$	SERDES Auxiliary Supply Voltage	—	2.374	2.625	V
$V_{CCHRX}^6$	SERDES Input Buffer Power Supply	ECP5UM	0.30	1.155	V
		ECP5-5G	0.30	1.26	V
$V_{CCHTX}$	SERDES Output Buffer Power Supply	ECP5UM	1.045	1.155	V
		ECP5-5G	1.14	1.26	V

**Notes:**

1. For correct operation, all supplies except  $V_{REF}$  must be held in their valid operation range. This is true independent of feature usage.
2. All supplies with same voltage, except SERDES Power Supplies, should be connected together.
3. See recommended voltages by I/O standard in [Table 3.4](#) on page 48.
4.  $V_{CCAUX}$  ramp rate must not exceed 30 mV/μs during power-up when transitioning between 0 V and 3 V.
5. Refer to [ECP5 and ECP5-5G SERDES/PCS Usage Guide \(TN1261\)](#) for information on board considerations for SERDES power supplies.
6.  $V_{CCHRX}$  is used for Rx termination. It can be biased to  $V_{cm}$  if external AC coupling is used. This voltage needs to meet all the HDin input voltage level requirements specified in the Rx section of this Data Sheet.

### 3.7. Hot Socketing Requirements

**Table 3.6. Hot Socketing Requirements**

Description	Min	Typ	Max	Unit
Input current per SERDES I/O pin when device is powered down and inputs driven.	—	—	8	mA
Input current per HDIN pin when device power supply is off, inputs driven <sup>1, 2</sup>	—	—	15	mA
Current per HDIN pin when device power ramps up, input driven <sup>3</sup>	—	—	50	mA
Current per HDOUT pin when device power supply is off, outputs pulled up <sup>4</sup>	—	—	30	mA

**Notes:**

1. Device is powered down with all supplies grounded, both HDINP and HDINN inputs driven by a CML driver with maximum allowed output  $V_{CCHTX}$ , 8b/10b data, no external AC coupling.
2. Each P and N input must have less than the specified maximum input current during hot plug. For a device with 2 DCU, the total input current would be  $15\text{ mA} * 4\text{ channels} * 2\text{ input pins per channel} = 120\text{ mA}$ .
3. Device power supplies are ramping up ( $V_{CCA}$  and  $V_{CCAUX}$ ), both HDINP and HDINN inputs are driven by a CML driver with maximum allowed output  $V_{CCHTX}$ , 8b/10b data, internal AC coupling.
4. Device is powered down with all supplies grounded. Both HDOUTP and HDOUN outputs are pulled up to  $V_{CCHTX}$  by the far end receiver termination of  $50\text{ }\Omega$  single ended.

### 3.8. ESD Performance

Refer to the [ECP5 and ECP5-5G Product Family Qualification Summary](#) for complete qualification data, including ESD performance.

### 3.9. DC Electrical Characteristics

Over Recommended Operating Conditions

**Table 3.7. DC Electrical Characteristics**

Symbol	Parameter	Condition	Min	Typ	Max	Unit
$I_{IL}, I_{IH}^{1, 4}$	Input or I/O Low Leakage	$0 \leq V_{IN} \leq V_{CCIO}$	—	—	10	$\mu\text{A}$
$I_{IH}^{1, 3}$	Input or I/O High Leakage	$V_{CCIO} < V_{IN} \leq V_{IH(\text{MAX})}$	—	—	100	$\mu\text{A}$
$I_{PU}$	I/O Active Pull-up Current, sustaining logic HIGH state	$0.7 V_{CCIO} \leq V_{IN} \leq V_{CCIO}$	-30	—	—	$\mu\text{A}$
	I/O Active Pull-up Current, pulling down from logic HIGH state	$0 \leq V_{IN} \leq 0.7 V_{CCIO}$	—	—	-150	$\mu\text{A}$
$I_{PD}$	I/O Active Pull-down Current, sustaining logic LOW state	$0 \leq V_{IN} \leq V_{IL(\text{MAX})}$	30	—	—	$\mu\text{A}$
	I/O Active Pull-down Current, pulling up from logic LOW state	$0 \leq V_{IN} \leq V_{CCIO}$	—	—	150	$\mu\text{A}$
C1	I/O Capacitance <sup>2</sup>	$V_{CCIO} = 3.3\text{ V}, 2.5\text{ V}, 1.8\text{ V}, 1.5\text{ V}, 1.2\text{ V}$ , $V_{CC} = 1.2\text{ V}$ , $V_{IO} = 0$ to $V_{IH(\text{MAX})}$	—	5	8	pf
C2	Dedicated Input Capacitance <sup>2</sup>	$V_{CCIO} = 3.3\text{ V}, 2.5\text{ V}, 1.8\text{ V}, 1.5\text{ V}, 1.2\text{ V}$ , $V_{CC} = 1.2\text{ V}$ , $V_{IO} = 0$ to $V_{IH(\text{MAX})}$	—	5	7	pf
$V_{HYST}$	Hysteresis for Single-Ended Inputs	$V_{CCIO} = 3.3\text{ V}$	—	300	—	mV
		$V_{CCIO} = 2.5\text{ V}$	—	250	—	mV

**Notes:**

1. Input or I/O leakage current is measured with the pin configured as an input or as an I/O with the output driver tristated. It is not measured with the output driver active. Bus maintenance circuits are disabled.
2.  $T_A 25^\circ\text{C}$ ,  $f = 1.0\text{ MHz}$ .
3. Applicable to general purpose I/Os in top and bottom banks.
4. When used as  $V_{REF}$ , maximum leakage=  $25\text{ }\mu\text{A}$ .

### 3.14.8. SLVS

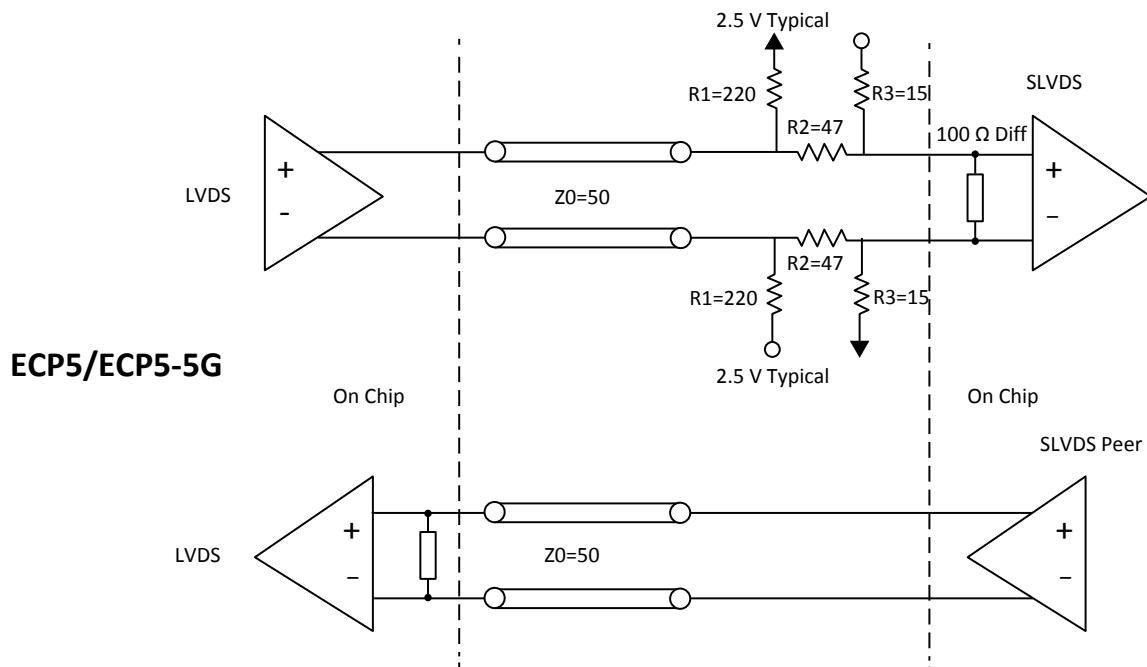
Scalable Low-Voltage Signaling (SLVS) is based on a point-to-point signaling method defined in the JEDEC JESD8-13 (SLVS-400) standard. This standard evolved from the traditional LVDS standard and relies on the advantage of its use of smaller voltage swings and a lower common-mode voltage. The 200 mV (400 mV p-p) SLVS swing contributes to a reduction in power.

The ECP5/ECP5-5G devices can receive differential input up to 800 Mb/s with its LVDS input buffer. This LVDS input buffer is used to meet the SLVS input standard specified by the JEDEC standard. The SLVS output parameters are compared to ECP5/ECP5-5G LVDS input parameters, as listed in Table 3.18.

**Table 3.18. Input to SLVS**

Parameter	ECP5/ECP5-5G LVDS Input	SLVS Output	Unit
Vcm (min)	50	150	mV
Vcm (max)	2350	250	mV
Differential Voltage (min)	100	140	mV
Differential Voltage (max)	—	270	mV

ECP5/ECP5-5G does not support SLVS output. However, SLVS output can be created using ECP5/ECP5-5G LVDS outputs by level shift to meet the low Vcm/Vod levels required by SLVS. [Figure 3.5](#) shows how the LVDS output can be shifted external to meet SLVS levels.



**Figure 3.5. SLVS Interface**

### 3.15. Typical Building Block Function Performance

**Table 3.19. Pin-to-Pin Performance**

Function	-8 Timing	Unit
<b>Basic Functions</b>		
16-Bit Decoder	5.06	ns
32-Bit Decoder	6.08	ns
64-Bit Decoder	5.06	ns
4:1 Mux	4.45	ns
8:1 Mux	4.63	ns
16:1 Mux	4.81	ns
32:1 Mux	4.85	ns

**Notes:**

1. I/Os are configured with LVC MOS25 with  $V_{COO}=2.5$ , 12 mA drive.
2. These functions were generated using Lattice Diamond design software tool. Exact performance may vary with the device and the design software tool version. The design software tool uses internal parameters that have been characterized but are not tested on every device.
3. Commercial timing numbers are shown. Industrial numbers are typically slower and can be extracted from Lattice Diamond design software tool.

### 3.18. External Switching Characteristics

Over recommended commercial operating conditions.

**Table 3.22. ECP5/ECP5-5G External Switching Characteristics**

Parameter	Description	Device	-8		-7		-6		Unit			
			Min	Max	Min	Max	Min	Max				
<b>Clocks</b>												
<b>Primary Clock</b>												
f <sub>MAX_PRI</sub>	Frequency for Primary Clock Tree	—	—	370	—	303	—	257	MHz			
t <sub>W_PRI</sub>	Clock Pulse Width for Primary Clock	—	0.8	—	0.9	—	1.0	—	ns			
t <sub>SKEW_PRI</sub>	Primary Clock Skew within a Device	—	—	420	—	462	—	505	ps			
<b>Edge Clock</b>												
f <sub>MAX_EDGE</sub>	Frequency for Edge Clock Tree	—	—	400	—	350	—	312	MHz			
t <sub>W_EDGE</sub>	Clock Pulse Width for Edge Clock	—	1.175	—	1.344	—	1.50	—	ns			
t <sub>SKEW_EDGE</sub>	Edge Clock Skew within a Bank	—	—	160	—	180	—	200	ps			
<b>Generic SDR Input</b>												
<b>General I/O Pin Parameters Using Dedicated Primary Clock Input without PLL</b>												
t <sub>CO</sub>	Clock to Output - PIO Output Register	All Devices	—	5.4	—	6.1	—	6.8	ns			
t <sub>SU</sub>	Clock to Data Setup - PIO Input Register	All Devices	0	—	0	—	0	—	ns			
t <sub>H</sub>	Clock to Data Hold - PIO Input Register	All Devices	2.7	—	3	—	3.3	—	ns			
t <sub>SU_DEL</sub>	Clock to Data Setup - PIO Input Register with Data Input Delay	All Devices	1.2	—	1.33	—	1.46	—	ns			
t <sub>H_DEL</sub>	Clock to Data Hold - PIO Input Register with Data Input Delay	All Devices	0	—	0	—	0	—	ns			
f <sub>MAX_IO</sub>	Clock Frequency of I/O and PFU Register	All Devices	—	400	—	350	—	312	MHz			
<b>General I/O Pin Parameters Using Dedicated Primary Clock Input with PLL</b>												
t <sub>COPLL</sub>	Clock to Output - PIO Output Register	All Devices	—	3.5	—	3.8	—	4.1	ns			
t <sub>SUPPLL</sub>	Clock to Data Setup - PIO Input Register	All Devices	0.7	—	0.78	—	0.85	—	ns			
t <sub>HPLL</sub>	Clock to Data Hold - PIO Input Register	All Devices	0.8	—	0.89	—	0.98	—	ns			
t <sub>SU_DEPLPLL</sub>	Clock to Data Setup - PIO Input Register with Data Input Delay	All Devices	1.6	—	1.78	—	1.95	—	ns			

### 3.19. sysCLOCK PLL Timing

Over recommended operating conditions.

**Table 3.23. sysCLOCK PLL Timing**

Parameter	Descriptions	Conditions	Min	Max	Units
$f_{IN}$	Input Clock Frequency (CLKI, CLKFB)	—	8	400	MHz
$f_{OUT}$	Output Clock Frequency (CLKOP, CLKOS)	—	3.125	400	MHz
$f_{VCO}$	PLL VCO Frequency	—	400	800	MHz
$f_{PFD}^3$	Phase Detector Input Frequency	—	10	400	MHz
<b>AC Characteristics</b>					
$t_{DT}$	Output Clock Duty Cycle	—	45	55	%
$t_{PH4}$	Output Phase Accuracy	—	-5	5	%
$t_{OPJIT}^1$	Output Clock Period Jitter	$f_{OUT} \geq 100$ MHz	—	100	ps p-p
		$f_{OUT} < 100$ MHz	—	0.025	UIPP
	Output Clock Cycle-to-Cycle Jitter	$f_{OUT} \geq 100$ MHz	—	200	ps p-p
		$f_{OUT} < 100$ MHz	—	0.050	UIPP
	Output Clock Phase Jitter	$f_{PFD} \geq 100$ MHz	—	200	ps p-p
		$f_{PFD} < 100$ MHz	—	0.011	UIPP
$t_{SPO}$	Static Phase Offset	Divider ratio = integer	—	400	ps p-p
$t_w$	Output Clock Pulse Width	At 90% or 10%	0.9	—	ns
$t_{LOCK}^2$	PLL Lock-in Time	—	—	15	ms
$t_{UNLOCK}$	PLL Unlock Time	—	—	50	ns
$t_{IPJIT}$	Input Clock Period Jitter	$f_{PFD} \geq 20$ MHz	—	1,000	ps p-p
		$f_{PFD} < 20$ MHz	—	0.02	UIPP
$t_{HI}$	Input Clock High Time	90% to 90%	0.5	—	ns
$t_{LO}$	Input Clock Low Time	10% to 10%	0.5	—	ns
$t_{RST}$	RST / Pulse Width	—	1	—	ms
$t_{RSTREC}$	RST Recovery Time	—	1	—	ns
$t_{LOAD\_REG}$	Min Pulse for CIB_LOAD_REG	—	10	—	ns
$t_{ROTATE-SETUP}$	Min time for CIB dynamic phase controls to be stable before CIB_ROTATE	—	5	—	ns
$t_{ROTATE-WD}$	Min pulse width for CIB_ROTATE to maintain "0" or	—	4	—	VCO cycles

**Notes:**

1. Jitter sample is taken over 10,000 samples for Periodic jitter, and 2,000 samples for Cycle-to-Cycle jitter of the primary PLL output with clean reference clock with no additional I/O toggling.
2. Output clock is valid after  $t_{LOCK}$  for PLL reset and dynamic delay adjustment.
3. Period jitter and cycle-to-cycle jitter numbers are guaranteed for  $f_{PFD} > 10$  MHz. For  $f_{PFD} < 10$  MHz, the jitter numbers may not be met in certain conditions.

### 3.21. SERDES/PCS Block Latency

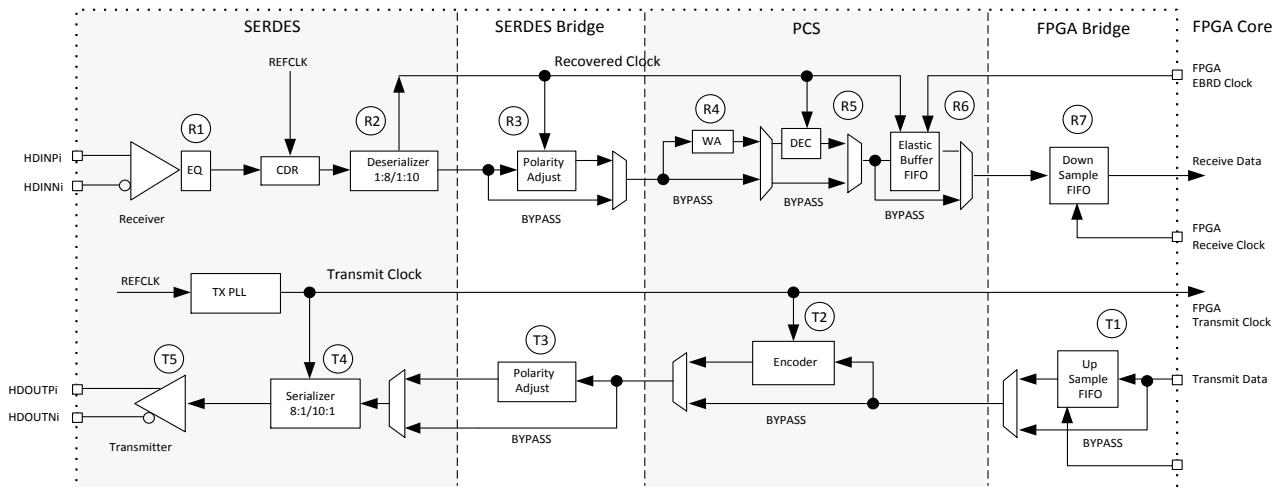
Table 3.26 describes the latency of each functional block in the transmitter and receiver. Latency is given in parallel clock cycles. Figure 3.13 shows the location of each block.

**Table 3.26. SERDES/PCS Latency Breakdown**

Item	Description	Min	Avg	Max	Fixed	Bypass	Unit <sup>3</sup>
<b>Transmit Data Latency<sup>1</sup></b>							
T1	FPGA Bridge - Gearing disabled with same clocks	3	—	4	—	1	byte clk
	FPGA Bridge - Gearing enabled	5	—	7	—	—	word clk
T2	8b10b Encoder	—	—	—	2	1	byte clk
T3	SERDES Bridge transmit	—	—	—	2	1	byte clk
T4	Serializer: 8-bit mode	—	—	—	15 + Δ1	—	UI + ps
	Serializer: 10-bit mode	—	—	—	18 + Δ1	—	UI + ps
T5	Pre-emphasis ON	—	—	—	1 + Δ2	—	UI + ps
	Pre-emphasis OFF	—	—	—	0 + Δ3	—	UI + ps
<b>Receive Data Latency<sup>2</sup></b>							
R1	Equalization ON	—	—	—	Δ1	—	UI + ps
	Equalization OFF	—	—	—	Δ2	—	UI + ps
R2	Deserializer: 8-bit mode	—	—	—	10 + Δ3	—	UI + ps
	Deserializer: 10-bit mode	—	—	—	12 + Δ3	—	UI + ps
R3	SERDES Bridge receive	—	—	—	2	—	byte clk
R4	Word alignment	3.1	—	4	—	1	byte clk
R5	8b10b decoder	—	—	—	1	0	byte clk
R6	Clock Tolerance Compensation	7	15	23	—	1	byte clk
R7	FPGA Bridge - Gearing disabled with same clocks	4	—	5	—	1	byte clk
	FPGA Bridge - Gearing enabled	7	—	9	—	—	word clk

**Notes:**

1.  $\Delta 1 = -245$  ps,  $\Delta 2 = +88$  ps,  $\Delta 3 = +112$  ps.
2.  $\Delta 1 = +118$  ps,  $\Delta 2 = +132$  ps,  $\Delta 3 = +700$  ps.
3. byte clk = 8UIs (8-bit mode), or 10 UIs (10-bit mode); word clk = 16UIs (8-bit mode), or 20 UIs (10-bit mode).



**Figure 3.13. Transmitter and Receiver Latency Block Diagram**

## 3.27. XAUI/CPRI LV E.30 Electrical and Timing Characteristics

### 3.27.1. AC and DC Characteristics

Over recommended operating conditions.

**Table 3.33. Transmit**

Symbol	Description	Test Conditions	Min	Typ	Max	Unit
T <sub>RF</sub>	Differential rise/fall time	20% to 80%	—	80	—	ps
Z <sub>TX_DIFF_DC</sub>	Differential impedance	—	80	100	120	Ω
J <sub>TX_DDJ</sub> <sup>2, 3</sup>	Output data deterministic jitter	—	—	—	0.17	UI
J <sub>TX_TJ</sub> <sup>1, 2, 3</sup>	Total output data jitter	—	—	—	0.35	UI

**Notes:**

1. Total jitter includes both deterministic jitter and random jitter.
2. Jitter values are measured with each CML output AC coupled into a 50 Ω impedance (100 Ω differential impedance).
3. Jitter and skew are specified between differential crossings of the 50% threshold of the reference signal.

Over recommended operating conditions.

**Table 3.34. Receive and Jitter Tolerance**

Symbol	Description	Test Conditions	Min	Typ	Max	Unit
RL <sub>RX_DIFF</sub>	Differential return loss	From 100 MHz to 3.125 GHz	10	—	—	dB
RL <sub>RX_CM</sub>	Common mode return loss	From 100 MHz to 3.125 GHz	6	—	—	dB
Z <sub>RX_DIFF</sub>	Differential termination resistance	—	80	100	120	Ω
J <sub>RX_DJ</sub> <sup>1, 2, 3</sup>	Deterministic jitter tolerance (peak-to-peak)	—	—	—	0.37	UI
J <sub>RX_RJ</sub> <sup>1, 2, 3</sup>	Random jitter tolerance (peak-to-peak)	—	—	—	0.18	UI
J <sub>RX_SJ</sub> <sup>1, 2, 3</sup>	Sinusoidal jitter tolerance (peak-to-peak)	—	—	—	0.10	UI
J <sub>RX_TJ</sub> <sup>1, 2, 3</sup>	Total jitter tolerance (peak-to-peak)	—	—	—	0.65	UI
T <sub>RX_EYE</sub>	Receiver eye opening	—	0.35	—	—	UI

**Notes:**

1. Total jitter includes deterministic jitter, random jitter and sinusoidal jitter.
2. Jitter values are measured with each high-speed input AC coupled into a 50 Ω impedance.
3. Jitter and skew are specified between differential crossings of the 50% threshold of the reference signal.

## 3.28. CPRI LV E.24/SGMII(2.5Gbps) Electrical and Timing Characteristics

### 3.28.1. AC and DC Characteristics

**Table 3.35. Transmit**

Symbol	Description	Test Conditions	Min	Typ	Max	Unit
T <sub>RF</sub> <sup>1</sup>	Differential rise/fall time	20% to 80%	—	80	—	ps
Z <sub>TX_DIFF_DC</sub>	Differential impedance	—	80	100	120	Ω
J <sub>TX_DDJ</sub> <sup>3, 4</sup>	Output data deterministic jitter	—	—	—	0.17	UI
J <sub>TX_TJ</sub> <sup>2, 4</sup>	Total output data jitter	—	—	—	0.35	UI

**Notes:**

1. Rise and Fall times measured with board trace, connector and approximately 2.5 pf load.
2. Total jitter includes both deterministic jitter and random jitter. The random jitter is the total jitter minus the actual deterministic jitter.
3. Jitter values are measured with each CML output AC coupled into a 50 Ω impedance (100 Ω differential impedance).
4. Jitter and skew are specified between differential crossings of the 50% threshold of the reference signal.

### 3.30. SMPTE SD/HD-SDI/3G-SDI (Serial Digital Interface) Electrical and Timing Characteristics

#### 3.30.1. AC and DC Characteristics

**Table 3.39. Transmit**

Symbol	Description	Test Conditions	Min	Typ	Max	Unit
BR <sub>SDO</sub>	Serial data rate	—	270	—	2975	Mb/s
T <sub>JALIGNMENT</sub> <sup>2</sup>	Serial output jitter, alignment	270 Mb/s <sup>6</sup>	—	—	0.2	UI
T <sub>JALIGNMENT</sub> <sup>2</sup>	Serial output jitter, alignment	1485 Mb/s	—	—	0.2	UI
T <sub>JALIGNMENT</sub> <sup>1, 2</sup>	Serial output jitter, alignment	2970 Mb/s	—	—	0.3	UI
T <sub>JTIMING</sub>	Serial output jitter, timing	270 Mb/s <sup>6</sup>	—	—	0.2	UI
T <sub>JTIMING</sub>	Serial output jitter, timing	1485 Mb/s	—	—	1	UI
T <sub>JTIMING</sub>	Serial output jitter, timing	2970 Mb/s	—	—	2	UI

**Notes:**

1. Timing jitter is measured in accordance with SMPTE serial data transmission standards.
2. Jitter is defined in accordance with SMPTE RP1 184-1996 as: jitter at an equipment output in the absence of input jitter.
3. All Tx jitter are measured at the output of an industry standard cable driver, with the Lattice SERDES device configured to 50 Ω output impedance connecting to the external cable driver with differential signaling.
4. The cable driver drives: RL=75 Ω, AC-coupled at 270, 1485, or 2970 Mb/s.
5. All LFE5UM/LFE5UM5G devices are compliant with all SMPTE compliance tests, except 3G-SDI Level-A pathological compliance pattern test.
6. 270 Mb/s is supported with Rate Divider only.

**Table 3.40. Receive**

Symbol	Description	Test Conditions	Min	Typ	Max	Unit
BR <sub>SDI</sub>	Serial input data rate	—	270	—	2970	Mb/s

**Table 3.41. Reference Clock**

Symbol	Description	Test Conditions	Min	Typ	Max	Unit
F <sub>VCLK</sub>	Video output clock frequency	—	54	—	148.5	MHz
DC <sub>V</sub>	Duty cycle, video clock	—	45	50	55	%

**Note:** SD-SDI (270 Mb/s) is supported with Rate Divider only. For Single Rate: Reference Clock = 54 MHz and Rate Divider = /2. For Tri-Rate: Reference Clock = 148.5 MHz and Rate Divider = /11.

### 3.31. sysCONFIG Port Timing Specifications

Over recommended operating conditions.

**Table 3.42. ECP5/ECP5-5G sysCONFIG Port Timing Specifications**

Symbol	Parameter		Min	Max	Unit
<b>POR, Configuration Initialization, and Wakeup</b>					
$t_{ICFG}$	Time from the Application of $V_{CC}$ , $V_{CCAUX}$ or $V_{CCIO8}$ (whichever is the last) to the rising edge of INITN	—	—	33	ms
$t_{VMC}$	Time from $t_{ICFG}$ to the valid Master CCLK	—	—	5	us
$t_{cz}$	CCLK from Active to High-Z	—	—	300	ns
<b>Master CCLK</b>					
$f_{MCLK}$	Frequency	All selected frequencies	-20	20	%
$t_{MCLK-DC}$	Duty Cycle	All selected frequencies	40	60	%
<b>All Configuration Modes</b>					
$t_{PRGM}$	PROGRAMN LOW pulse accepted	—	110	—	ns
$t_{PRGMRJ}$	PROGRAMN LOW pulse rejected	—	—	50	ns
$t_{INITL}$	INITN LOW time	—	—	55	ns
$t_{DPPINT}$	PROGRAMN LOW to INITN LOW	—	—	70	ns
$t_{DPPDONE}$	PROGRAMN LOW to DONE LOW	—	—	80	ns
$t_{IODISS}$	PROGRAMN LOW to I/O Disabled	—	—	150	ns
<b>Slave SPI</b>					
$f_{CCLK}$	CCLK input clock frequency	—	—	60	MHz
$t_{CCLKH}$	CCLK input clock pulselength HIGH	—	6	—	ns
$t_{CCLKL}$	CCLK input clock pulselength LOW	—	6	—	ns
$t_{STSU}$	CCLK setup time	—	1	—	ns
$t_{STH}$	CCLK hold time	—	1	—	ns
$t_{STCO}$	CCLK falling edge to valid output	—	—	10	ns
$t_{STOZ}$	CCLK falling edge to valid disable	—	—	10	ns
$t_{STOV}$	CCLK falling edge to valid enable	—	—	10	ns
$t_{SCS}$	Chip Select HIGH time	—	25	—	ns
$t_{SCSS}$	Chip Select setup time	—	3	—	ns
$t_{SCSH}$	Chip Select hold time	—	3	—	ns
<b>Master SPI</b>					
$f_{CCLK}$	Max selected CCLK output frequency	—	—	62	MHz
$t_{CCLKH}$	CCLK output clock pulse width HIGH	—	3.5	—	ns
$t_{CCLKL}$	CCLK output clock pulse width LOW	—	3.5	—	ns
$t_{STSU}$	CCLK setup time	—	5	—	ns
$t_{STH}$	CCLK hold time	—	1	—	ns
$t_{CSSPI}$	INITN HIGH to Chip Select LOW	—	100	200	ns
$t_{CFGX}$	INITN HIGH to first CCLK edge	—	—	150	ns
<b>Slave Serial</b>					
$f_{CCLK}$	CCLK input clock frequency	—	—	66	MHz
$t_{SSCH}$	CCLK input clock pulse width HIGH	—	5	—	ns
$t_{SSCL}$	CCLK input clock pulse width LOW	—	5	—	ns
$t_{SUSCDI}$	CCLK setup time	—	0.5	—	ns
$t_{HSCDI}$	CCLK hold time	—	1.5	—	ns

Pin Information Summary		LFE5UM/ LFE5UM5G-25		LFE5UM/LFE5UM5G-45			LFE5UM/LFE5UM5G-85			
Pin Type		285 csfBGA	381 caBGA	285 csfBGA	381 caBGA	554 caBGA	285 csfBGA	381 caBGA	554 caBGA	756 caBGA
TAP		4	4	4	4	4	4	4	4	4
Miscellaneous Dedicated Pins		7	7	7	7	7	7	7	7	7
GND		83	59	83	59	113	83	59	113	166
NC		1	8	1	2	33	1	0	17	29
Reserved		0	2	0	2	4	0	2	4	4
SERDES		14	28	14	28	28	14	28	28	28
VCCA (SERDES)	VCCA0	2	2	2	2	6	2	2	6	8
	VCCA1	0	2	0	2	6	0	2	6	9
VCCAUX (SERDES)	VCCAUXA0	2	2	2	2	2	2	2	2	2
	VCCAUXA1	0	2	0	2	2	0	2	2	2
GNDA (SERDES)		26	26	26	26	49	26	26	49	60
Total Balls		285	381	285	381	554	285	381	554	756
High Speed Differential Input / Output Pairs	Bank 0	0	0	0	0	0	0	0	0	0
	Bank 1	0	0	0	0	0	0	0	0	0
	Bank 2	10/8	16/8	10/8	16/8	16/8	10/8	17/9	16/8	24/12
	Bank 3	14/7	16/8	14/7	16/8	24/12	14/7	16/8	24/12	32/16
	Bank 4	0	0	0	0	0	0	0	0	0
	Bank 6	13/6	16/8	13/6	16/8	24/12	13/6	16/8	24/12	32/16
	Bank 7	8/6	16/8	8/6	16/8	16/8	8/6	16/8	16/8	24/12
	Bank 8	0	0	0	0	0	0	0	0	0
Total High Speed Differential I/O Pairs		45/2	64/32	45/27	64/3	80/40	45/27	65/3	80/40	112/5
DQS Groups (> 11 pins in group)	Bank 0	0	0	0	0	0	0	0	0	0
	Bank 1	0	0	0	0	0	0	0	0	0
	Bank 2	1	2	1	2	2	1	2	2	3
	Bank 3	2	2	2	2	3	2	2	3	4
	Bank 4	0	0	0	0	0	0	0	0	0
	Bank 6	2	2	2	2	3	2	2	3	4
	Bank 7	1	2	1	2	2	1	2	2	3
	Bank 8	0	0	0	0	0	0	0	0	0
Total DQS Groups		6	8	6	8	10	6	8	10	14

#### 4.3.2. LFE5U

Pin Information Summary		LFE5U-12			LFE5U-25			LFE5U-45					LFE5U-85				
Pin Type		256 caBGA	285 csfBGA	381 caBGA	256 caBGA	285 csfBGA	381 caBGA	256 caBGA	285 csfBGA	381 caBGA	554 caBGA	285 csfBG	381 caBGA	554 caBGA	756 caBG		
General Purpose Inputs/Outputs per Bank	Bank 0	24	6	24	24	6	24	24	6	27	32	6	27	32	56		
	Bank 1	32	6	32	32	6	32	32	6	33	40	6	33	40	48		
	Bank 2	32	21	32	32	21	32	32	21	32	32	21	34	32	48		
	Bank 3	32	28	32	32	28	32	32	28	33	48	28	33	48	64		
	Bank 4	0	0	0	0	0	0	0	0	0	0	0	0	0	14		
	Bank 6	32	26	32	32	26	32	32	26	33	48	26	33	48	64		
	Bank 7	32	18	32	32	18	32	32	18	32	32	18	32	32	48		
	Bank 8	13	13	13	13	13	13	13	13	13	13	13	13	13	13		
Total Single-Ended User		197	118	197	197	118	197	197	118	203	245	118	205	259	365		
VCC		6	13	20	6	13	20	6	13	20	24	13	20	24	36		
VCCAUX (Core)		2	3	4	2	3	4	2	3	4	9	3	4	9	8		
VCCIO	Bank 0	2	1	2	2	1	2	2	1	2	3	1	2	3	4		
	Bank 1	2	1	2	2	1	2	2	1	2	3	1	2	3	4		
	Bank 2	2	2	3	2	2	3	2	2	3	4	2	3	4	4		
	Bank 3	2	2	3	2	2	3	2	2	3	3	2	3	3	4		
	Bank 4	0	0	0	0	0	0	0	0	0	0	0	0	0	2		
	Bank 6	2	2	3	2	2	3	2	2	3	4	2	3	4	4		
	Bank 7	2	2	3	2	2	3	2	2	3	3	2	3	3	4		
	Bank 8	1	2	2	1	2	2	1	2	2	2	2	2	2	2		
TAP		4	4	4	4	4	4	4	4	4	4	4	4	4	4		
Miscellaneous Dedicated		7	7	7	7	7	7	7	7	7	7	7	7	7	7		
GND		27	123	99	27	123	99	27	123	99	198	123	99	198	267		
NC		0	1	26	0	1	26	0	1	26	33	1	26	33	29		
Reserved		0	4	6	0	4	6	0	4	6	12	4	6	12	12		
Total Balls		256	285	381	256	285	381	256	285	381	554	285	381	554	756		
High Speed Differential Input / Output Pairs	Bank 0	0	0	0	0	0	0	0	0	0	0	0	0	0	0		
	Bank 0	0	0	0	0	0	0	0	0	0	0	0	0	0	0		
	Bank	16/8	10/8	16/8	16/8	10/8	16/8	16/8	10/8	16/8	16/8	10/8	17/9	16/8			
	Bank	16/8	14/7	16/8	16/8	14/7	16/8	16/8	14/7	16/8	24/12	14/7	16/8	24/1			
	Bank	0	0	0	0	0	0	0	0	0	0	0	0	0			
	Bank	16/8	13/6	16/8	16/8	13/6	16/8	16/8	13/6	16/8	24/12	13/6	16/8	24/1			
	Bank	16/8	8/6	16/8	16/8	8/6	16/8	16/8	8/6	16/8	16/8	8/6	16/8	16/8			
	Bank	0	0	0	0	0	0	0	0	0	0	0	0	0			
Total High Speed		64/32	45/27	64/32	64/32	45/27	64/32	64/32	45/27	64/32	80/40	45/27	65/33	80/40	112/		
DQS Groups (>11 pins in group)	Bank 0	0	0	0	0	0	0	0	0	0	0	0	0	0	0		
	Bank 0	0	0	0	0	0	0	0	0	0	0	0	0	0	0		
	Bank 2	1	2	2	1	2	2	1	2	2	1	2	2	1	2		
	Bank 2	2	2	2	2	2	2	2	2	2	2	3	2	2	3		
	Bank 0	0	0	0	0	0	0	0	0	0	0	0	0	0	0		
	Bank 2	2	2	2	2	2	2	2	2	2	2	3	2	2	3		
	Bank 2	1	2	2	2	1	2	2	1	2	2	1	2	1	2		
	Bank 0	0	0	0	0	0	0	0	0	0	0	0	0	0	0		
Total DQS Groups		8	6	8	8	6	8	8	6	8	10	6	8	10	14		