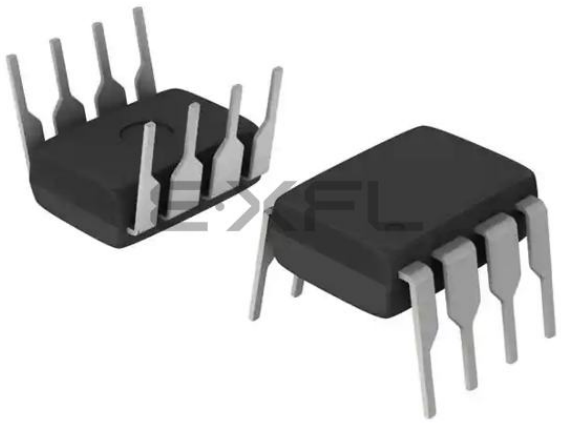

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Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	4MHz
Connectivity	-
Peripherals	POR, WDT
Number of I/O	5
Program Memory Size	1.75KB (1K x 14)
Program Memory Type	OTP
EEPROM Size	-
RAM Size	128 x 8
Voltage - Supply (Vcc/Vdd)	3V ~ 5.5V
Data Converters	A/D 4x8b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Through Hole
Package / Case	8-DIP (0.300", 7.62mm)
Supplier Device Package	8-PDIP
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic12c671-04e-p

1.0 GENERAL DESCRIPTION

The PIC12C67X devices are low-cost, high-performance, CMOS, fully-static, 8-bit microcontrollers with integrated analog-to-digital (A/D) converter and EEPROM data memory (EEPROM on PIC12CE67X versions only).

All PIC® microcontrollers employ an advanced RISC architecture. The PIC12C67X microcontrollers have enhanced core features, eight-level deep stack, and multiple internal and external interrupt sources. The separate instruction and data buses of the Harvard architecture allow a 14-bit wide instruction word with the separate 8-bit wide data. The two stage instruction pipeline allows all instructions to execute in a single cycle, except for program branches, which require two cycles. A total of 35 instructions (reduced instruction set) are available. Additionally, a large register set gives some of the architectural innovations used to achieve a very high performance.

PIC12C67X microcontrollers typically achieve a 2:1 code compression and a 4:1 speed improvement over other 8-bit microcontrollers in their class.

The PIC12C67X devices have 128 bytes of RAM, 16 bytes of EEPROM data memory (PIC12CE67X only), 5 I/O pins and 1 input pin. In addition a timer/counter is available. Also a 4-channel, high-speed, 8-bit A/D is provided. The 8-bit resolution is ideally suited for applications requiring low-cost analog interface, (i.e., thermostat control, pressure sensing, etc.)

The PIC12C67X devices have special features to reduce external components, thus reducing cost, enhancing system reliability and reducing power consumption. The Power-On Reset (POR), Power-up Timer (PWRT), and Oscillator Start-up Timer (OST) eliminate the need for external reset circuitry. There are five oscillator configurations to choose from, including INTRC precision internal oscillator mode and the power-saving LP (Low Power) oscillator mode. Power-saving SLEEP mode, Watchdog Timer and code protection features improve system cost, power and reliability. The SLEEP (power-down) feature provides a power-saving mode. The user can wake-up the chip from SLEEP through several external and internal interrupts and resets.

A highly reliable Watchdog Timer with its own on-chip RC oscillator provides protection against software lock-up.

A UV erasable windowed package version is ideal for code development, while the cost-effective One-Time-Programmable (OTP) version is suitable for production in any volume. The customer can take full advantage of Microchip's price leadership in OTP microcontrollers, while benefiting from the OTP's flexibility.

1.1 Applications

The PIC12C67X series fits perfectly in applications ranging from personal care appliances and security systems to low-power remote transmitters/receivers. The EPROM technology makes customizing application programs (transmitter codes, appliance settings, receiver frequencies, etc.) extremely fast and convenient, while the EEPROM data memory (PIC12CE67X only) technology allows for the changing of calibration factors and security codes. The small footprint packages, for through hole or surface mounting, make this microcontroller series perfect for applications with space limitations. Low-cost, low-power, high performance, ease of use and I/O flexibility make the PIC12C67X series very versatile even in areas where no microcontroller use has been considered before (i.e., timer functions, replacement of "glue" logic and PLD's in larger systems, coprocessor applications).

1.2 Family and Upward Compatibility

The PIC12C67X products are compatible with other members of the 14-bit PIC16CXXX families.

1.3 Development Support

The PIC12C67X devices are supported by a full-featured macro assembler, a software simulator, an in-circuit emulator, a low-cost development programmer and a full-featured programmer. A "C" compiler and fuzzy logic support tools are also available.

3.1 Clocking Scheme/Instruction Cycle

The clock input (from OSC1) is internally divided by four to generate four non-overlapping quadrature clocks, namely Q1, Q2, Q3 and Q4. Internally, the program counter (PC) is incremented every Q1, and the instruction is fetched from the program memory and latched into the instruction register in Q4. The instruction is decoded and executed during the following Q1 through Q4. The clocks and instruction execution flow is shown in Figure 3-2.

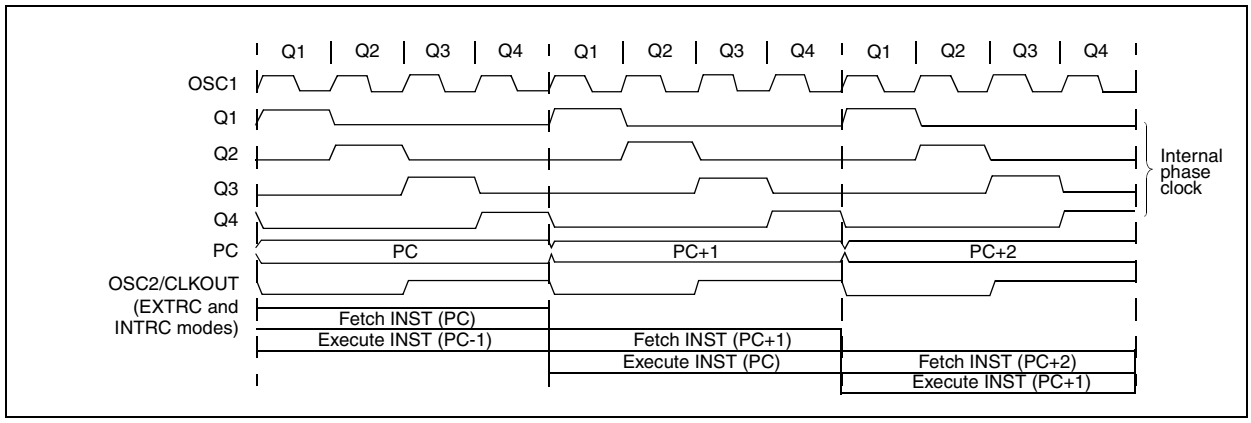
3.2 Instruction Flow/Pipelining

An “Instruction Cycle” consists of four Q cycles (Q1, Q2, Q3 and Q4). The instruction fetch and execute are pipelined such that fetch takes one instruction cycle, while decode and execute takes another instruction cycle. However, due to the pipelining, each instruction effectively executes in one cycle. If an instruction causes the program counter to change (i.e., `GOTO`), then two cycles are required to complete the instruction (Example 3-1).

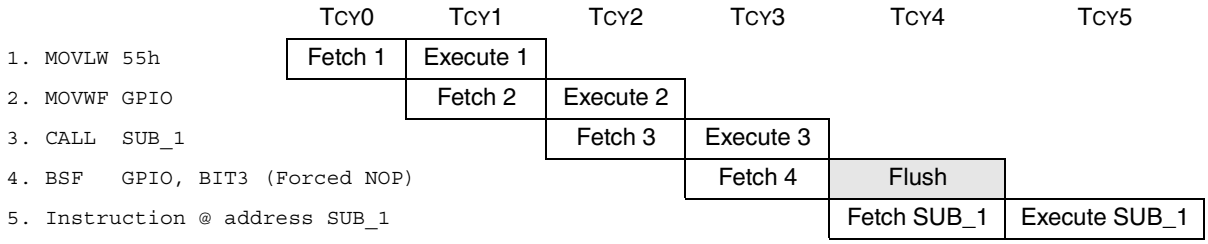
A fetch cycle begins with the program counter (PC) incrementing in Q1.

In the execution cycle, the fetched instruction is latched into the “Instruction Register" (IR) in cycle Q1. This instruction is then decoded and executed during the Q2, Q3, and Q4 cycles. Data memory is read during Q2 (operand read) and written during Q4 (destination write).

FIGURE 3-2: CLOCK/INSTRUCTION CYCLE

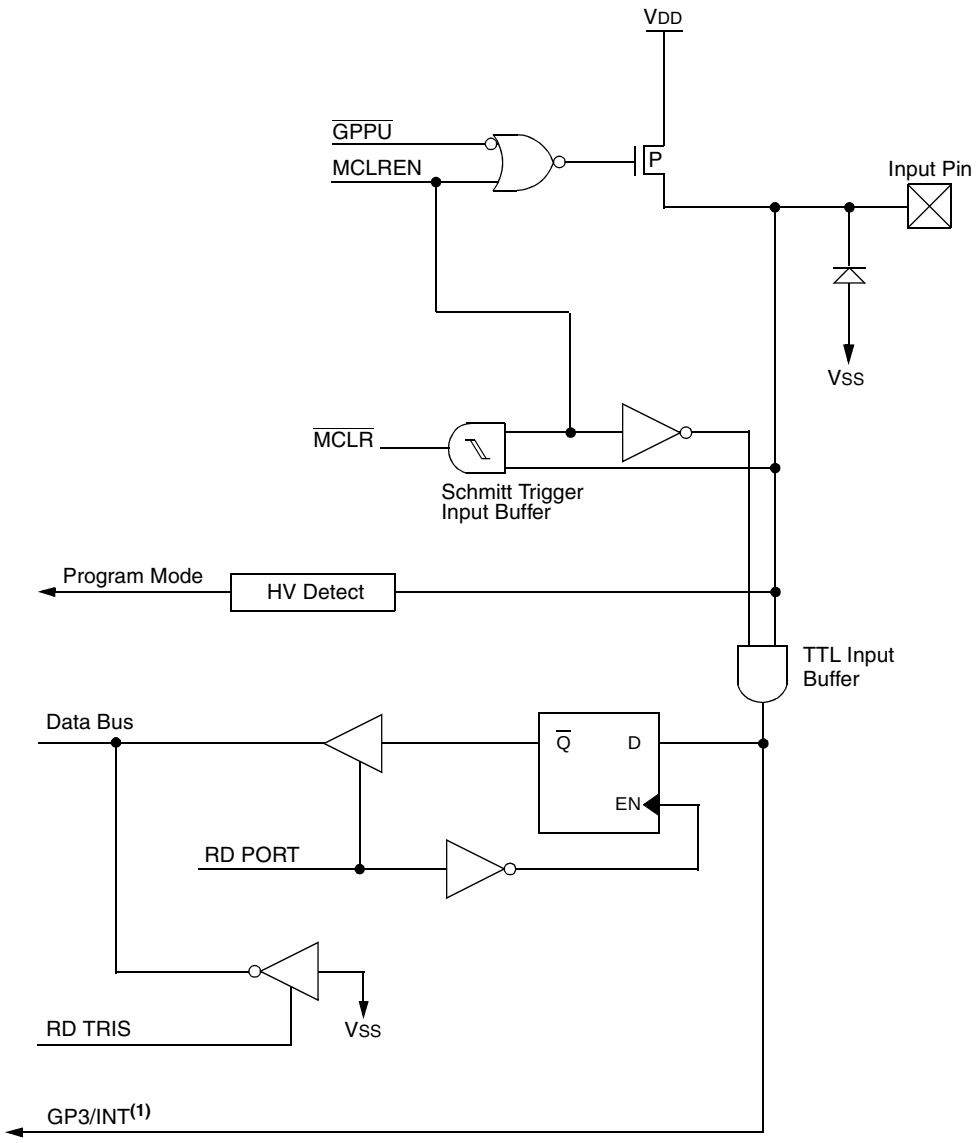


EXAMPLE 3-1: INSTRUCTION PIPELINE FLOW



All instructions are single cycle, except for any program branches. These take two cycles since the fetched instruction is “flushed” from the pipeline while the new instruction is being fetched and then executed.

FIGURE 5-3: BLOCK DIAGRAM OF GP3/$\overline{\text{MCLR}}$/VPP PIN



Note 1: Wake-up on pin change interrupt for GP3.

6.1.5 ACKNOWLEDGE

The EEPROM, when addressed, will generate an acknowledge after the reception of each byte. The processor must generate an extra clock pulse which is associated with this acknowledge bit.

Note:	Acknowledge bits are not generated if an internal programming cycle is in progress.

The device that acknowledges has to pull down the SDA line during the acknowledge clock pulse in such a way that the SDA line is stable LOW during the HIGH period of the acknowledge related clock pulse. Of course, setup and hold times must be taken into account. The processor must signal an end of data to the EEPROM by not generating an acknowledge bit on the last byte that has been clocked out of the EEPROM. In this case, the EEPROM must leave the data line HIGH to enable the processor to generate the STOP condition (Figure 6-4).

FIGURE 6-1: BLOCK DIAGRAM OF GPIO6 (SDA LINE)

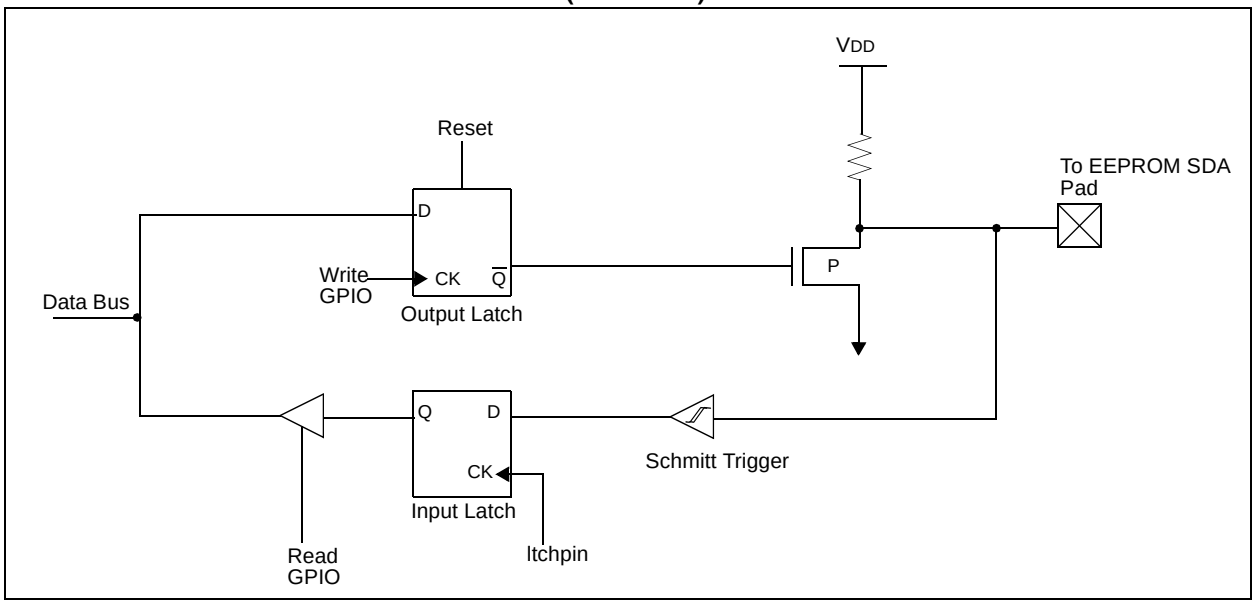
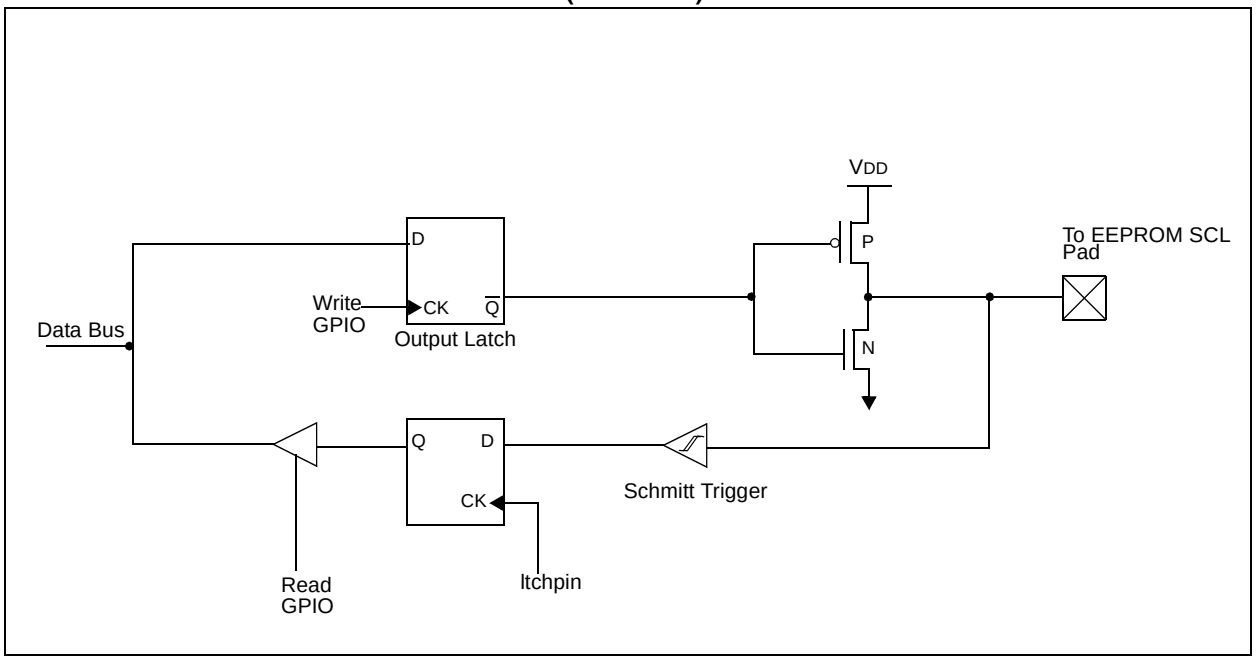


FIGURE 6-2: BLOCK DIAGRAM OF GPIO7 (SCL LINE)



NOTES:

REGISTER 8-2: ADCON1 REGISTER (ADDRESS 9Fh)

U-0	U-0	U-0	U-0	U-0	R/W-0	R/W-0	R/W-0
—	—	—	—	—	PCFG2	PCFG1	PCFG0
bit7							bit0

R = Readable bit
W = Writable bit
U = Unimplemented bit, read as '0'
- n = Value at POR reset

bit 7-2: **Unimplemented:** Read as '0'

bit 1-0: **PCFG<2:0>:** A/D Port Configuration Control bits

PCFG<2:0>	GP4	GP2	GP1	GP0	VREF
000(1)	A	A	A	A	VDD
001	A	A	VREF	A	GP1
010	D	A	A	A	VDD
011	D	A	VREF	A	GP1
100	D	D	A	A	VDD
101	D	D	VREF	A	GP1
110	D	D	D	A	VDD
111	D	D	D	D	VDD

A = Analog input

D = Digital I/O

Note 1: Value on reset.

2: Any instruction that reads a pin configured as an analog input will read a '0'.

8.1 A/D Sampling Requirements

For the A/D converter to meet its specified accuracy, the charge holding capacitor (CHOLD) must be allowed to fully charge to the input channel voltage level. The analog input model is shown in Figure 8-2. The source impedance (RS) and the internal sampling switch (RSS) impedance directly affect the time required to charge the capacitor CHOLD. The sampling switch (RSS) impedance varies over the device voltage (VDD), see Figure 8-2. **The maximum recommended impedance for analog sources is 10 kΩ.** After the analog input channel is selected (changed), this acquisition must be done before the conversion can be started.

To calculate the minimum acquisition time, Equation 8-1 may be used. This equation assumes that 1/2 LSb error is used (512 steps for the A/D). The 1/2 LSb error is the maximum error allowed for the A/D to meet its specified resolution.

EQUATION 8-1: A/D MINIMUM CHARGING TIME

$V_{HOLD} = (V_{REF} - (V_{REF}/512)) \bullet (1 - e^{(-Tc/C_{HOLD}(R_{IC} + R_{SS} + R_S))})$

or

$Tc = -(51.2\ pF)(1\ k\Omega + R_{SS} + R_S)\ \ln(1/511)$

Example 8-1 shows the calculation of the minimum required acquisition time TACQ. This calculation is based on the following system assumptions.

Rs = 10 kΩ

1/2 LSb error

VDD = 5V → Rss = 7 kΩ

Temp (system max.) = 50°C

VHOLD = 0 @ t = 0

Note 1: The reference voltage (VREF) has no effect on the equation, since it cancels itself out.

2: The charge holding capacitor (CHOLD) is not discharged after each conversion.

3: The maximum recommended impedance for analog sources is 10 kΩ. This is required to meet the pin leakage specification.

4: After a conversion has completed, a 2.0 TAD delay must complete before acquisition can begin again. During this time, the holding capacitor is not connected to the selected A/D input channel.

EXAMPLE 8-1: CALCULATING THE MINIMUM REQUIRED SAMPLE TIME

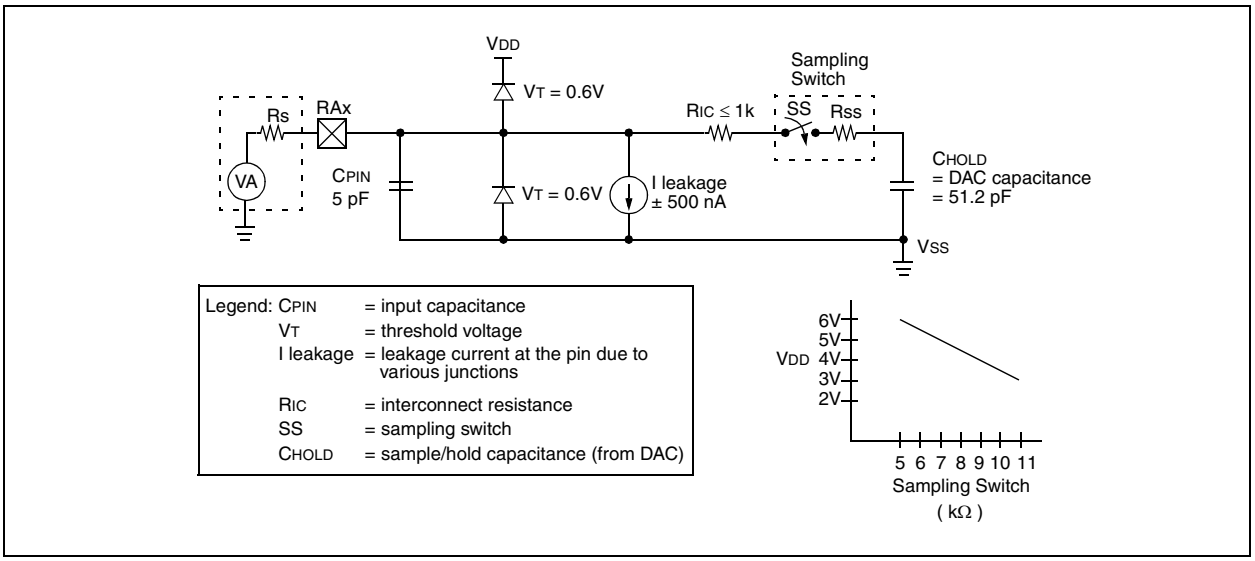
TACQ = Internal Amplifier Settling Time +
Holding Capacitor Charging Time +
Temperature Coefficient

TACQ = 5 μs + Tc + [(Temp - 25°C)(0.05 μs/°C)]

TC = -CHOLD (RIC + RSS + RS) ln(1/512)
-51.2 pF (1 kΩ + 7 kΩ + 10 kΩ) ln(0.0020)
-51.2 pF (18 kΩ) ln(0.0020)
-0.921 μs (-6.2146)
5.724 μs

TACQ = 5 μs + 5.724 μs + [(50°C - 25°C)(0.05 μs/°C)]
10.724 μs + 1.25 μs
11.974 μs

FIGURE 8-2: ANALOG INPUT MODEL



8.5 A/D Operation During Sleep

The A/D module can operate during SLEEP mode. This requires that the A/D clock source be set to RC (ADCS<1:0> = 11). When the RC clock source is selected, the A/D module waits one instruction cycle before starting the conversion. This allows the SLEEP instruction to be executed, which eliminates all digital switching noise from the conversion. When the conversion is completed, the GO/$\overline{\text{DONE}}$ bit will be cleared, and the result loaded into the ADRES Register. If the A/D interrupt is enabled, the device will wake-up from SLEEP. If the A/D interrupt is not enabled, the A/D module will then be turned off, although the ADON bit will remain set.

When the A/D clock source is another clock option (not RC), a SLEEP instruction will cause the present conversion to be aborted and the A/D module to be turned off, though the ADON bit will remain set.

Turning off the A/D places the A/D module in its lowest current consumption state.

Note:	For the A/D module to operate in SLEEP, the A/D clock source must be set to RC (ADCS<1:0> = 11). To perform an A/D conversion in SLEEP, the GO/$\overline{\text{DONE}}$ bit must be set, followed by the SLEEP instruction.

8.6 A/D Accuracy/Error

The overall accuracy of the A/D is less than ± 1 LSb for V_{DD} = 5V ± 10% and the analog V_{REF} = V_{DD}. This overall accuracy includes offset error, full scale error, and integral error. The A/D converter is monotonic over the full V_{DD} range. The resolution and accuracy may be less when either the analog reference (V_{DD}) is less than 5.0V or when the analog reference (V_{REF}) is less than V_{DD}.

The maximum pin leakage current is specified in the Device Data Sheet electrical specification, parameter #D060.

In systems where the device frequency is low, use of the A/D RC clock is preferred. At moderate to high frequencies, T_{AD} should be derived from the device oscillator. T_{AD} must not violate the minimum and should be ≤8 μs for preferred operation. This is because T_{AD}, when derived from T_{OSC}, is kept away from on-chip phase clock transitions. This reduces, to a large extent, the effects of digital switching noise. This is not possible with the RC derived clock. The loss of accuracy due to digital switching noise can be significant if many I/O pins are active.

In systems where the device will enter SLEEP mode after the start of the A/D conversion, the RC clock source selection is required. In this mode, the digital noise from the modules in SLEEP are stopped. This method gives high accuracy.

8.7 Effects of a Reset

A device reset forces all registers to their reset state. This forces the A/D module to be turned off, and any conversion is aborted. The value that is in the ADRES register is not modified for a Reset. The ADRES register will contain unknown data after a Power-on Reset.

8.8 Connection Considerations

If the input voltage exceeds the rail values (V_{SS} or V_{DD}) by greater than 0.2V, then the accuracy of the conversion is out of specification.

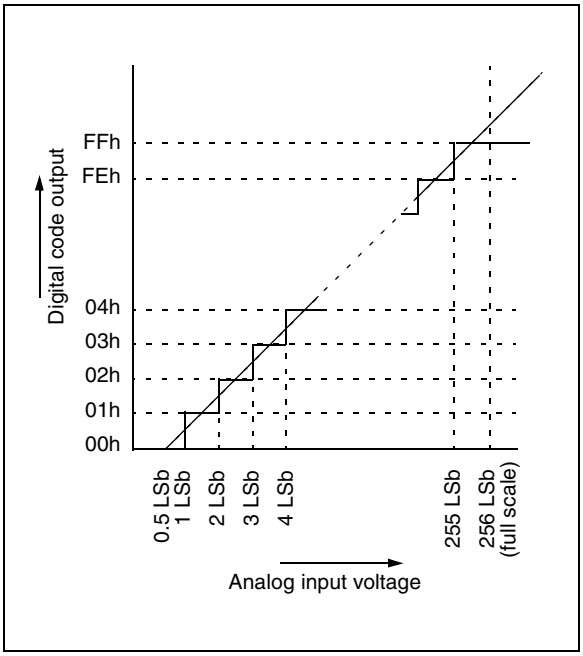
Note:	For the PIC12C67X, care must be taken when using the GP4 pin in A/D conversions due to its proximity to the OSC1 pin.

An external RC filter is sometimes added for anti-aliasing of the input signal. The R component should be selected to ensure that the total source impedance is kept under the 10 kΩ recommended specification. Any external components connected (via hi-impedance) to an analog input pin (capacitor, zener diode, etc.) should have very little leakage current at the pin.

8.9 Transfer Function

The ideal transfer function of the A/D converter is as follows: the first transition occurs when the analog input voltage (V_{AIN}) is 1 LSb (or Analog V_{REF} / 256) (Figure 8-3).

FIGURE 8-3: A/D TRANSFER FUNCTION



9.7 Watchdog Timer (WDT)

The Watchdog Timer is a free running, on-chip RC oscillator, which does not require any external components. This RC oscillator is separate from the RC oscillator of the OSC1/CLKIN pin. That means that the WDT will run, even if the clock on the OSC1/CLKIN and OSC2/CLKOUT pins of the device has been stopped, for example, by execution of a SLEEP instruction. During normal operation, a WDT time-out generates a device RESET (Watchdog Timer Reset). If the device is in SLEEP mode, a WDT time-out causes the device to wake-up and continue with normal operation (Watchdog Timer Wake-up). The WDT can be permanently disabled by clearing configuration bit WDTE (Section 9.1).

9.7.1 WDT PERIOD

The WDT has a nominal time-out period of 18 ms (with no prescaler). The time-out periods vary with temperature, VDD and process variations from part to part (see DC specs). If longer time-out periods are desired, a prescaler with a division ratio of up to 1:128 can be assigned to the WDT under software control by writing to the OPTION register. Thus, time-out periods up to 2.3 seconds can be realized.

The CLRWDT and SLEEP instructions clear the WDT and the postscaler, if assigned to the WDT, and prevent it from timing out early and generating a premature device RESET condition.

The $\overline{TO}$ bit in the STATUS register will be cleared upon a Watchdog Timer time-out.

9.7.2 WDT PROGRAMMING CONSIDERATIONS

It should also be taken into account that under worst case conditions (VDD = Min., Temperature = Max., and max. WDT prescaler), it may take several seconds before a WDT time-out occurs.

Note:	When the prescaler is assigned to the WDT, always execute a CLRWDT instruction before changing the prescale value, otherwise a WDT reset may occur.

See Example 7-1 and Example 7-2 for changing prescaler between WDT and Timer0.

FIGURE 9-15: WATCHDOG TIMER BLOCK DIAGRAM

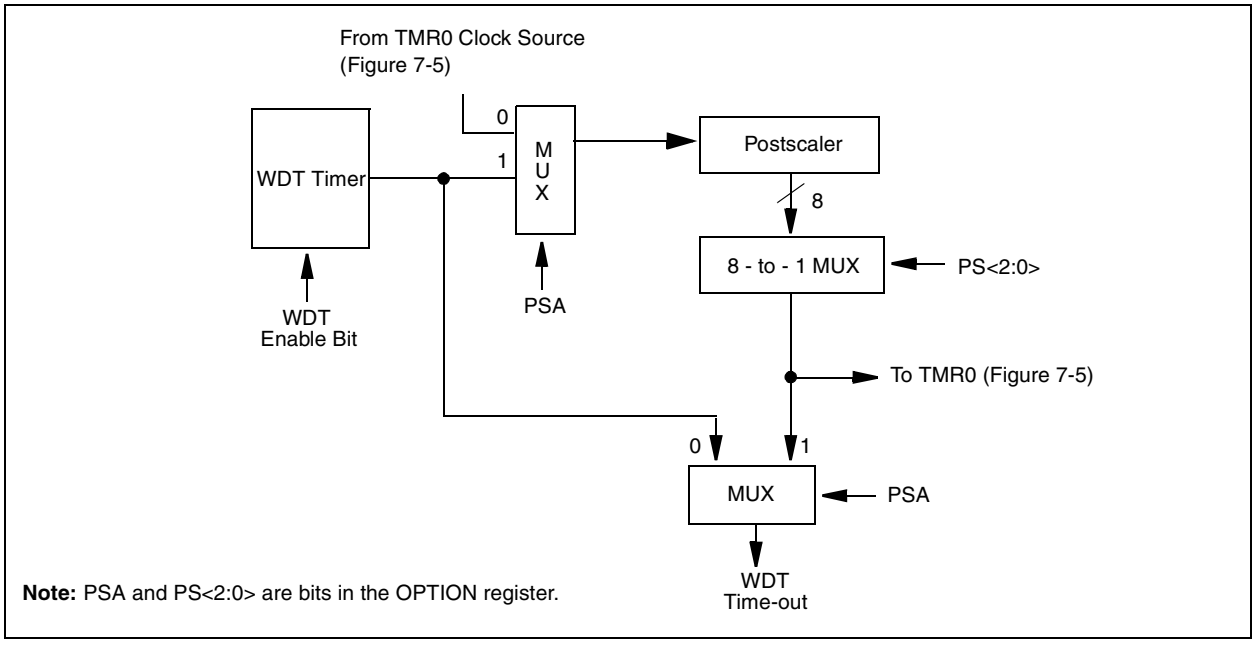


TABLE 9-8: SUMMARY OF WATCHDOG TIMER REGISTERS

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
2007h	Config. bits(1)	MCLRE	CP1	CP0	$\overline{PWRTE}$	$\overline{WDTE}$	FOSC2	FOSC1	FOSC0
81h	OPTION	$\overline{GPPU}$	INTEDG	T0CS	T0SE	PSA	PS2	PS1	PS0

Legend: Shaded cells are not used by the Watchdog Timer.
Note 1: See Register 9-1 for operation of these bits. Not all CP0 and CP1 bits are shown.

10.0 INSTRUCTION SET SUMMARY

Each PIC12C67X instruction is a 14-bit word divided into an OPCODE which specifies the instruction type and one or more operands which further specify the operation of the instruction. The PIC12C67X instruction set summary in Table 10-2 lists **byte-oriented**, **bit-oriented**, and **literal and control** operations. Table 10-1 shows the opcode field descriptions.

For **byte-oriented** instructions, 'f' represents a file register designator and 'd' represents a destination designator. The file register designator specifies which file register is to be used by the instruction.

The destination designator specifies where the result of the operation is to be placed. If 'd' is zero, the result is placed in the W register. If 'd' is one, the result is placed in the file register specified in the instruction.

For **bit-oriented** instructions, 'b' represents a bit field designator which selects the number of the bit affected by the operation, while 'f' represents the number of the file in which the bit is located.

For **literal and control** operations, 'k' represents an eight or eleven bit constant or literal value.

TABLE 10-1: OPCODE FIELD DESCRIPTIONS

Field	Description
f	Register file address (0x00 to 0x7F)
W	Working register (accumulator)
b	Bit address within an 8-bit file register
k	Literal field, constant data or label
x	Don't care location (= 0 or 1) The assembler will generate code with x = 0. It is the recommended form of use for compatibility with all Microchip software tools.
d	Destination select; d = 0: store result in W, d = 1: store result in file register f. Default is d = 1
label	Label name
TOS	Top of Stack
PC	Program Counter
PCLATH	Program Counter High Latch
GIE	Global Interrupt Enable bit
WDT	Watchdog Timer/Counter
$\overline{TO}$	Time-out bit
$\overline{PD}$	Power-down bit
dest	Destination either the W register or the specified register file location
[ ]	Options
( )	Contents
→	Assigned to
< >	Register bit field
∈	In the set of
italics	User defined term (font is courier)

The instruction set is highly orthogonal and is grouped into three basic categories:

- **Byte-oriented** operations
- **Bit-oriented** operations
- **Literal and control** operations

All instructions are executed within one single instruction cycle, unless a conditional test is true or the program counter is changed as a result of an instruction. In this case, the execution takes two instruction cycles with the second cycle executed as a NOP. One instruction cycle consists of four oscillator periods. Thus, for an oscillator frequency of 4 MHz, the normal instruction execution time is 1 μs. If a conditional test is true or the program counter is changed as a result of an instruction, the instruction execution time is 2 μs.

Table 10-2 lists the instructions recognized by the MPASM assembler.

Figure 10-1 shows the three general formats that the instructions can have.

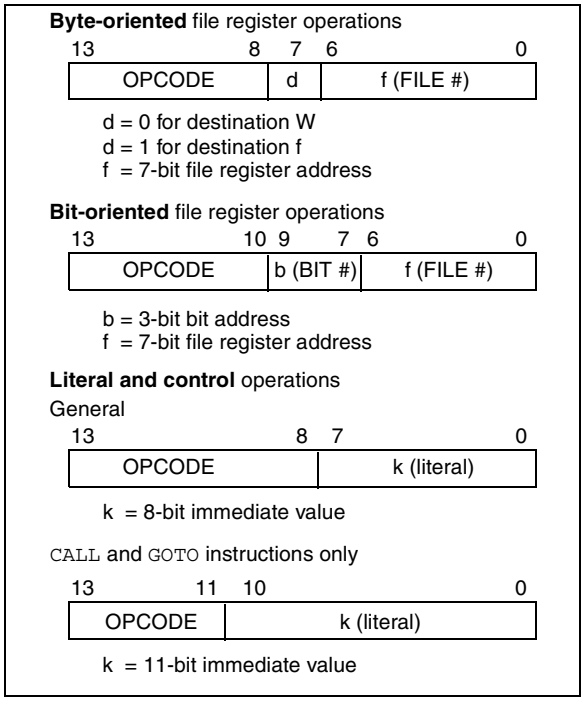
Note:	To maintain upward compatibility with future PIC12C67X products, do not use the OPTION and TRIS instructions.

All examples use the following format to represent a hexadecimal number:

0xhh

where h signifies a hexadecimal digit.

FIGURE 10-1: GENERAL FORMAT FOR INSTRUCTIONS



BCF **Bit Clear f**

Syntax: [*label*] BCF f,b

Operands: $0 \le f \le 127$
$0 \le b \le 7$

Operation: $0 \rightarrow (f<b>)$

Status Affected: None

Encoding:

01	00bb	bfff	ffff

Description: Bit 'b' in register 'f' is cleared.

Words: 1

Cycles: 1

Example

```
BCF     FLAG_REG, 7
```

Before Instruction
FLAG_REG = 0xC7
After Instruction
FLAG_REG = 0x47

BSF **Bit Set f**

Syntax: [*label*] BSF f,b

Operands: $0 \le f \le 127$
$0 \le b \le 7$

Operation: $1 \rightarrow (f<b>)$

Status Affected: None

Encoding:

01	01bb	bfff	ffff

Description: Bit 'b' in register 'f' is set.

Words: 1

Cycles: 1

Example

```
BSF     FLAG_REG,   7
```

Before Instruction
FLAG_REG = 0x0A
After Instruction
FLAG_REG = 0x8A

BTFSC **Bit Test, Skip if Clear**

Syntax: [*label*] BTFSC f,b

Operands: $0 \le f \le 127$
$0 \le b \le 7$

Operation: skip if (f<b>) = 0

Status Affected: None

Encoding:

01	10bb	bfff	ffff

Description: If bit 'b' in register 'f' is '0', then the next instruction is skipped.
If bit 'b' is '0', then the next instruction fetched during the current instruction execution is discarded, and a NOP is executed instead, making this a 2 cycle instruction.

Words: 1

Cycles: 1(2)

Example

```
HERE    BTFSC   FLAG,1
FALSE   GOTO    PROCESS_CO
TRUE    •       DE
        •
        •
```

Before Instruction
PC = address HERE
After Instruction
if FLAG<1> = 0,
PC = address TRUE
if FLAG<1>=1,
PC = address FALSE

NOTES:

12.0 ELECTRICAL SPECIFICATIONS FOR PIC12C67X

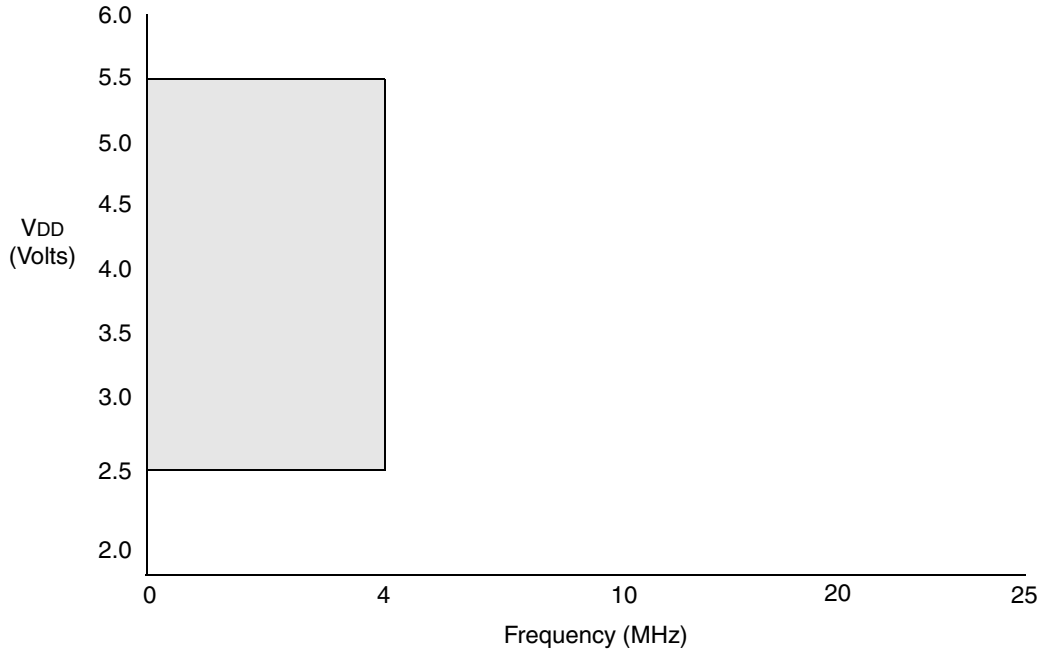
Absolute Maximum Ratings †

Ambient temperature under bias ..... –40° to +125°C
Storage temperature ..... –65°C to +150°C
Voltage on any pin with respect to V_{SS} (except V_{DD} and $\overline{MCLR}$) ..... –0.3V to (V_{DD} + 0.3V)
Voltage on V_{DD} with respect to V_{SS} ..... 0 to +7.0V
Voltage on $\overline{MCLR}$ with respect to V_{SS} (Note 2) ..... 0 to +14V
Total power dissipation (Note 1) ..... 700 mW
Maximum current out of V_{SS} pin ..... 200 mA
Maximum current into V_{DD} pin ..... 150 mA
Input clamp current, I_{IK} (V_I < 0 or V_I > V_{DD}) ..... ± 20 mA
Output clamp current, I_{OK} (V_O < 0 or V_O > V_{DD}) ..... ± 20 mA
Maximum output current sunk by any I/O pin ..... 25 mA
Maximum output current sourced by any I/O pin ..... 25 mA
Maximum current sunk by GPIO pins combined ..... 100 mA
Maximum current sourced by GPIO pins combined ..... 100 mA

Note 1: Power dissipation is calculated as follows: $P_{dis} = V_{DD} \times \{I_{DD} - \sum I_{OH}\} + \sum \{(V_{DD} - V_{OH}) \times I_{OH}\} + \sum(V_{OI} \times I_{OL})$.

† NOTICE: Stresses above those listed under “Absolute Maximum Ratings” may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operation listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

FIGURE 12-3: PIC12LC67X VOLTAGE-FREQUENCY GRAPH, -40°C ≤ TA ≤ +85°C



Note 1: The shaded region indicates the permissible combinations of voltage and frequency.

2: The maximum rated speed of the part limits the permissible combinations of voltage and frequency. Please reference the Product Identification System section for the maximum rated speed of the parts.

DC CHARACTERISTICS			Standard Operating Conditions (unless otherwise specified) Operating Temperature 0°C ≤ TA ≤ +70°C (commercial) –40°C ≤ TA ≤ +85°C (industrial) –40°C ≤ TA ≤ +125°C (extended)				
Parm No.	**Characteristic**	**Sym**	**Min**	**Typ(1)**	**Max**	**Units**	**Conditions**
	LP Oscillator Operating Frequency	FOSC	0		200	kHz	All temperatures
	INTRC/EXTRC Oscillator Operating Frequency		—		4(6)	MHz	All temperatures
	XT Oscillator Operating Frequency		0		4	MHz	All temperatures
	HS Oscillator Operating Frequency		0		10	MHz	All temperatures

* These parameters are characterized but not tested.

Note 1: Data in Typical ("Typ") column is based on characterization results at 25°C. This data is for design guidance only and is not tested.

2: This is the limit to which VDD can be lowered in SLEEP mode without losing RAM data.

3: The supply current is mainly a function of the operating voltage and frequency. Other factors such as bus loading, oscillator type, bus rate, internal code execution pattern, and temperature also have an impact on the current consumption.

a) The test conditions for all IDD measurements in active operation mode are:
OSC1 = external square wave, from rail-to-rail; all I/O pins tristated, pulled to VSS, T0CKI = VDD,
$\overline{MCLR}$ = VDD; WDT disabled.

b) For standby current measurements, the conditions are the same, except that the device is in SLEEP mode.

4: For EXTRC osc configuration, current through REXT is not included. The current through the resistor can be estimated by the formula:
Ir = VDD/2REXT (mA) with REXT in kOhm.

5: The power-down current in SLEEP mode does not depend on the oscillator type. Power-down current is measured with the part in SLEEP mode, with all I/O pins in hi-impedance state and tied to VDD or VSS.

6: INTRC calibration value is for 4MHz nominal at 5V, 25°C.

12.2 DC Characteristics: PIC12LC671/672 (Commercial, Industrial) PIC12LCE673/674 (Commercial, Industrial)

DC CHARACTERISTICS			Standard Operating Conditions (unless otherwise specified) Operating temperature 0°C ≤ TA ≤ +70°C (commercial) –40°C ≤ TA ≤ +85°C (industrial)				
Param No.	**Characteristic**	**Sym**	**Min**	**Typ†**	**Max**	**Units**	**Conditions**
D001	**Supply Voltage**	VDD	2.5		5.5	V	
D002	**RAM Data Retention Voltage(2)**	VDR		1.5*		V	Device in SLEEP mode
D003	**VDD Start Voltage to ensure Power-on Reset**	VPOR		VSS		V	See section on Power-on Reset for details
D004	**VDD Rise Rate to ensure Power-on Reset**	SVDD	0.05*			V/ms	See section on Power-on Reset for details
D010	**Supply Current(3)**	IDD	—	0.4	2.1	mA	FOSC = 4MHz, VDD = 2.5V XT and EXTRC mode (Note 4)
D010C			—	0.4	2.1	mA	FOSC = 4MHz, VDD = 2.5V INTRC mode (Note 6)
D010A			—	15	33	µA	FOSC = 32kHz, VDD = 2.5V, WDT disabled LP mode, Industrial Temperature
D020	**Power-down Current(5)**	IPD					
D021			—	0.2	5	µA	VDD = 2.5V, Commercial
D021B			—	0.2	6	µA	VDD = 2.5V, Industrial
	Watchdog Timer Current	ΔIWDT	—	2.0	4	µA	VDD = 2.5V, Commercial
				2.0	6	µA	VDD = 2.5V, Industrial
	LP Oscillator Operating Frequency	FOSC	0		200	kHz	All temperatures
	INTRC/EXTRC Oscillator Operating Frequency		—		4(6)	MHz	All temperatures
	XT Oscillator Operating Frequency		0		4	MHz	All temperatures
	HS Oscillator Operating Frequency		0		10	MHz	All temperatures

* These parameters are characterized but not tested.

Note 1: Data in Typical ("Typ") column is based on characterization results at 25°C. This data is for design guidance only and is not tested.

2: This is the limit to which VDD can be lowered in SLEEP mode without losing RAM data.

3: The supply current is mainly a function of the operating voltage and frequency. Other factors such as bus loading, oscillator type, bus rate, internal code execution pattern, and temperature also have an impact on the current consumption.

a) The test conditions for all IDD measurements in active operation mode are:
OSC1 = external square wave, from rail-to-rail; all I/O pins tristated, pulled to VSS, T0CKI = VDD,
$\overline{MCLR}$ = VDD; WDT disabled.

b) For standby current measurements, the conditions are the same, except that the device is in SLEEP mode.

4: For EXTRC osc configuration, current through REXT is not included. The current through the resistor can be estimated by the formula:
Ir = VDD/2REXT (mA) with REXT in kOhm.

5: The power-down current in SLEEP mode does not depend on the oscillator type. Power-down current is measured with the part in SLEEP mode, with all I/O pins in hi-impedance state and tied to VDD or VSS.

6: INTRC calibration value is for 4MHz nominal at 5V, 25°C.

TABLE 12-7: A/D CONVERTER CHARACTERISTICS: PIC12C671/672-04/PIC12CE673/674-04 (COMMERCIAL, INDUSTRIAL, EXTENDED) PIC12C671/672-10/PIC12CE673/674-10 (COMMERCIAL, INDUSTRIAL, EXTENDED) PIC12LC671/672-04/PIC12LCE673/674-04 (COMMERCIAL, INDUSTRIAL)

Param No.	Sym	Characteristic		Min	Typ†	Max	Units	Conditions
A01	NR	Resolution		—	—	8-bits	bit	VREF = VDD = 5.12V, VSS ≤ VAIN ≤ VREF
A02	EABS	Total absolute error		—	—	< ±1	LSb	VREF = VDD = 5.12V, VSS ≤ VAIN ≤ VREF
A03	EIL	Integral linearity error		—	—	< ±1	LSb	VREF = VDD = 5.12V, VSS ≤ VAIN ≤ VREF
A04	EDL	Differential linearity error		—	—	< ±1	LSb	VREF = VDD = 5.12V, VSS ≤ VAIN ≤ VREF
A05	EFS	Full scale error		—	—	< ±1	LSb	VREF = VDD = 5.12V, VSS ≤ VAIN ≤ VREF
A06	EOFF	Offset error		—	—	< ±1	LSb	VREF = VDD = 5.12V, VSS ≤ VAIN ≤ VREF
A10	—	Monotonicity		—	guaranteed (Note 3)	—	—	VSS ≤ VAIN ≤ VREF
A20	VREF	Reference voltage		2.5V	—	VDD + 0.3	V	
A25	VAIN	Analog input voltage		VSS - 0.3	—	VREF + 0.3	V	
A30	ZAIN	Recommended impedance of analog voltage source		—	—	10.0	kΩ	
A40	IAD	A/D conversion current (VDD)	PIC12**C**67X	—	180	—	μA	Average current consumption when A/D is on. (Note 1)
			PIC12**LC**67X	—	90	—	μA	
A50	IREF	VREF input current (Note 2)		10	—	1000	μA	During VAIN acquisition. Based on differential of VHOLD to VAIN to charge CHOLD, see Section 8.1.
				—	—	10	μA	During A/D Conversion cycle

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: When A/D is off, it will not consume any current other than minor leakage current. The power-down current spec includes any such leakage from the A/D module.

2: VREF current is from GP1 pin or VDD pin, whichever is selected as reference input.

3: The A/D conversion result never decreases with an increase in the Input Voltage, and has no missing codes.

TABLE 12-9: EEPROM MEMORY BUS TIMING REQUIREMENTS - PIC12CE673/674 ONLY.

AC Characteristics	Standard Operating Conditions (unless otherwise specified) Operating Temperature 0°C ≤ TA ≤ +70°C, Vcc = 3.0V to 5.5V (commercial) –40°C ≤ TA ≤ +85°C, Vcc = 3.0V to 5.5V (industrial) –40°C ≤ TA ≤ +125°C, Vcc = 4.5V to 5.5V (extended) Operating Voltage VDD range is described in Section 12.1				
Parameter	**Symbol**	**Min**	**Max**	**Units**	**Conditions**
Clock frequency	FCLK	— — —	100 100 400	kHz	4.5V ≤ Vcc ≤ 5.5V (E Temp range) 3.0V ≤ Vcc ≤ 4.5V 4.5V ≤ Vcc ≤ 5.5V
Clock high time	THIGH	4000 4000 600	— — —	ns	4.5V ≤ Vcc ≤ 5.5V (E Temp range) 3.0V ≤ Vcc ≤ 4.5V 4.5V ≤ Vcc ≤ 5.5V
Clock low time	TLOW	4700 4700 1300	— — —	ns	4.5V ≤ Vcc ≤ 5.5V (E Temp range) 3.0V ≤ Vcc ≤ 4.5V 4.5V ≤ Vcc ≤ 5.5V
SDA and SCL rise time (Note 1)	TR	— — —	1000 1000 300	ns	4.5V ≤ Vcc ≤ 5.5V (E Temp range) 3.0V ≤ Vcc ≤ 4.5V 4.5V ≤ Vcc ≤ 5.5V
SDA and SCL fall time	TF	—	300	ns	(Note 1)
START condition hold time	THD:STA	4000 4000 600	— — —	ns	4.5V ≤ Vcc ≤ 5.5V (E Temp range) 3.0V ≤ Vcc ≤ 4.5V 4.5V ≤ Vcc ≤ 5.5V
START condition setup time	TSU:STA	4700 4700 600	— — —	ns	4.5V ≤ Vcc ≤ 5.5V (E Temp range) 3.0V ≤ Vcc ≤ 4.5V 4.5V ≤ Vcc ≤ 5.5V
Data input hold time	THD:DAT	0	—	ns	(Note 2)
Data input setup time	TSU:DAT	250 250 100	— — —	ns	4.5V ≤ Vcc ≤ 5.5V (E Temp range) 3.0V ≤ Vcc ≤ 4.5V 4.5V ≤ Vcc ≤ 5.5V
STOP condition setup time	TSU:STO	4000 4000 600	— — —	ns	4.5V ≤ Vcc ≤ 5.5V (E Temp range) 3.0V ≤ Vcc ≤ 4.5V 4.5V ≤ Vcc ≤ 5.5V
Output valid from clock (Note 2)	TAA	— — —	3500 3500 900	ns	4.5V ≤ Vcc ≤ 5.5V (E Temp range) 3.0V ≤ Vcc ≤ 4.5V 4.5V ≤ Vcc ≤ 5.5V
Bus free time: Time the bus must be free before a new transmission can start	TBUF	4700 4700 1300	— — —	ns	4.5V ≤ Vcc ≤ 5.5V (E Temp range) 3.0V ≤ Vcc ≤ 4.5V 4.5V ≤ Vcc ≤ 5.5V
Output fall time from VIH minimum to VIL maximum	TOF	20+0.1 CB	250	ns	(Note 1), CB ≤ 100 pF
Input filter spike suppression (SDA and SCL pins)	TSP	—	50	ns	(Notes 1, 3)
Write cycle time	TWC	—	4	ms	
Endurance		1M	—	cycles	25°C, VCC = 5.0V, Block Mode (Note 4)

Note 1: Not 100% tested. CB = total capacitance of one bus line in pF.

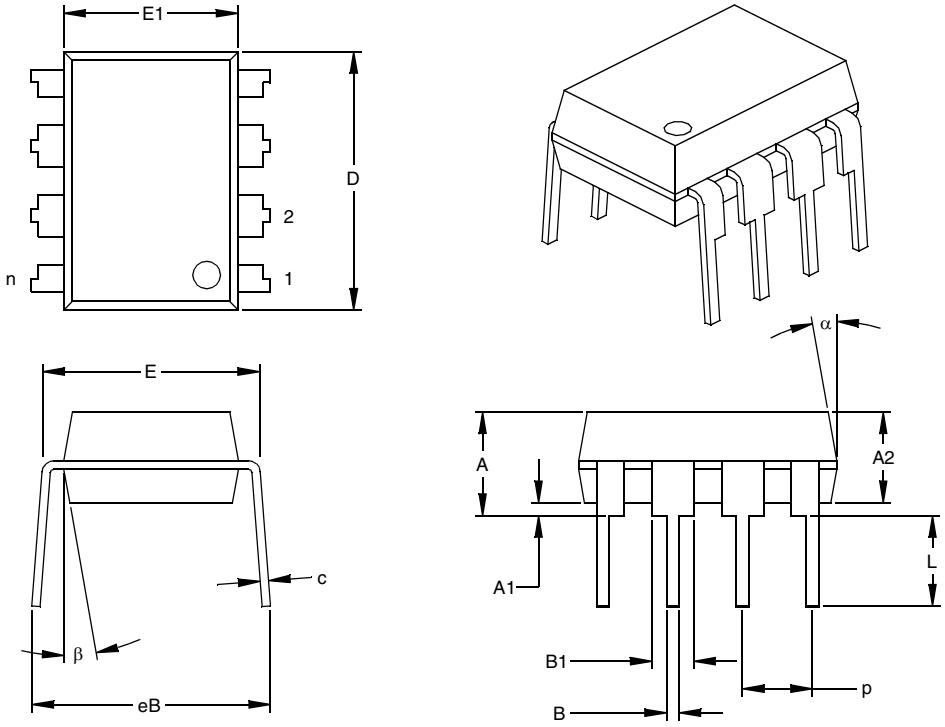
2: As a transmitter, the device must provide an internal minimum delay time to bridge the undefined region (minimum 300 ns) of the falling edge of SCL and avoid unintended generation of START or STOP conditions.

3: The combined TSP and VHYS specifications are due to new Schmitt Trigger inputs which provide improved noise spike suppression. This eliminates the need for a TI specification for standard operation.

4: This parameter is not tested but ensured by characterization. For endurance estimates in a specific application, please consult the Total Endurance Model which can be obtained on Microchip's website.

8-Lead Plastic Dual In-line (P) – 300 mil (PDIP)

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Units		INCHES*			MILLIMETERS		
Dimension Limits		MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n		8			8	
Pitch	p		.100			2.54	
Top to Seating Plane	A	.140	.155	.170	3.56	3.94	4.32
Molded Package Thickness	A2	.115	.130	.145	2.92	3.30	3.68
Base to Seating Plane	A1	.015			0.38		
Shoulder to Shoulder Width	E	.300	.313	.325	7.62	7.94	8.26
Molded Package Width	E1	.240	.250	.260	6.10	6.35	6.60
Overall Length	D	.360	.373	.385	9.14	9.46	9.78
Tip to Seating Plane	L	.125	.130	.135	3.18	3.30	3.43
Lead Thickness	c	.008	.012	.015	0.20	0.29	0.38
Upper Lead Width	B1	.045	.058	.070	1.14	1.46	1.78
Lower Lead Width	B	.014	.018	.022	0.36	0.46	0.56
Overall Row Spacing	eB	.310	.370	.430	7.87	9.40	10.92
Mold Draft Angle Top	α	5	10	15	5	10	15
Mold Draft Angle Bottom	β	5	10	15	5	10	15

*Controlling Parameter

Notes:

Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010” (0.254mm) per side.

JEDEC Equivalent: MS-001

Drawing No. C04-018

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