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"[Embedded - Microcontrollers](#)" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

### Applications of "[Embedded - Microcontrollers](#)"

#### Details

Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	32MHz
Connectivity	I <sup>2</sup> C, LINbus, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	25
Program Memory Size	14KB (8K x 14)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	1K x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.6V
Data Converters	A/D 17x10b; D/A 1x5b, 1x8b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	28-VQFN Exposed Pad
Supplier Device Package	28-QFN (6x6)
Purchase URL	<a href="https://www.e-xfl.com/product-detail/microchip-technology/pic16lf1716t-i-ml">https://www.e-xfl.com/product-detail/microchip-technology/pic16lf1716t-i-ml</a>

## 3.6 Stack

All devices have a 16-level x 15-bit wide hardware stack (refer to Figure 3-1). The stack space is not part of either program or data space. The PC is PUSHed onto the stack when `CALL` or `CALLW` instructions are executed or an interrupt causes a branch. The stack is POPed in the event of a `RETURN`, `RETLW` or a `RETFIE` instruction execution. `PCLATH` is not affected by a `PUSH` or `POP` operation.

The stack operates as a circular buffer if the `STVREN` bit is programmed to '0' (Configuration Words). This means that after the stack has been PUSHed sixteen times, the seventeenth PUSH overwrites the value that was stored from the first PUSH. The eighteenth PUSH overwrites the second PUSH (and so on). The `STKOVF` and `STKUNF` flag bits will be set on an Overflow/Underflow, regardless of whether the Reset is enabled.

**Note:** There are no instructions/mnemonics called `PUSH` or `POP`. These are actions that occur from the execution of the `CALL`, `CALLW`, `RETURN`, `RETLW` and `RETFIE` instructions or the vectoring to an interrupt address.

### 3.6.1 ACCESSING THE STACK

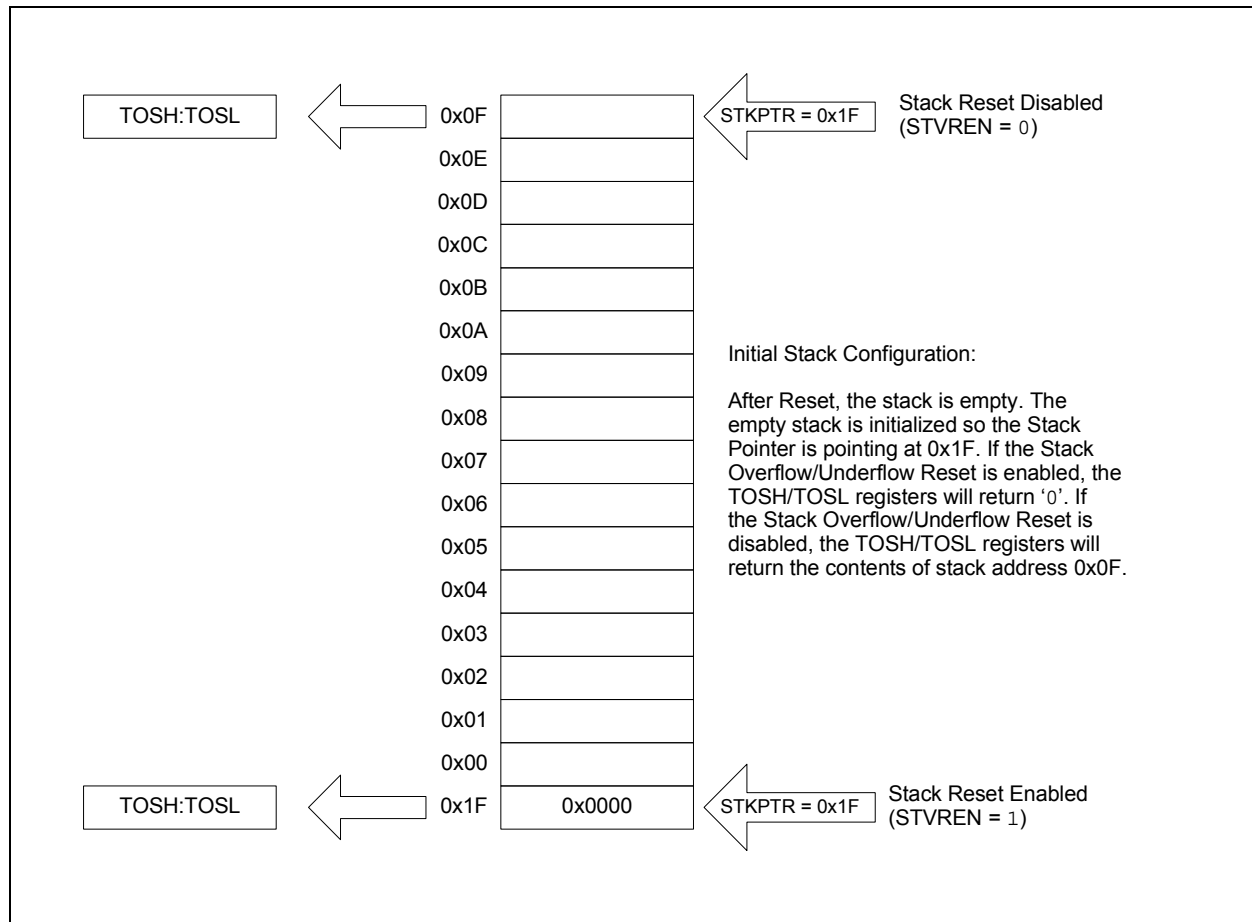
The stack is available through the `TOSH`, `TOSL` and `STKPTR` registers. `STKPTR` is the current value of the Stack Pointer. `TOSH:TOSL` register pair points to the TOP of the stack. Both registers are read/writable. `TOS` is split into `TOSH` and `TOSL` due to the 15-bit size of the PC. To access the stack, adjust the value of `STKPTR`, which will position `TOSH:TOSL`, then read/write to `TOSH:TOSL`. `STKPTR` is five bits to allow detection of overflow and underflow.

**Note:** Care should be taken when modifying the `STKPTR` while interrupts are enabled.

During normal program operation, `CALL`, `CALLW` and Interrupts will increment `STKPTR` while `RETLW`, `RETURN`, and `RETFIE` will decrement `STKPTR`. At any time, `STKPTR` can be inspected to see how much stack is left. The `STKPTR` always points at the currently used place on the stack. Therefore, a `CALL` or `CALLW` will increment the `STKPTR` and then write the PC, and a return will unload the PC and then decrement the `STKPTR`.

Reference Figure 3-5 through Figure 3-8 for examples of accessing the stack.

**FIGURE 3-5: ACCESSING THE STACK EXAMPLE 1**



## 4.3 Code Protection

Code protection allows the device to be protected from unauthorized access. Program memory protection is controlled independently. Internal access to the program memory is unaffected by any code protection setting.

### 4.3.1 PROGRAM MEMORY PROTECTION

The entire program memory space is protected from external reads and writes by the CP bit in Configuration Words. When  $\overline{CP} = 0$ , external reads and writes of program memory are inhibited and a read will return all '0's. The CPU can continue to read program memory, regardless of the protection bit settings. Writing the program memory is dependent upon the write protection setting. See **Section 4.4 “Write Protection”** for more information.

## 4.4 Write Protection

Write protection allows the device to be protected from unintended self-writes. Applications, such as boot loader software, can be protected while allowing other regions of the program memory to be modified.

The WRT<1:0> bits in Configuration Words define the size of the program memory block that is protected.

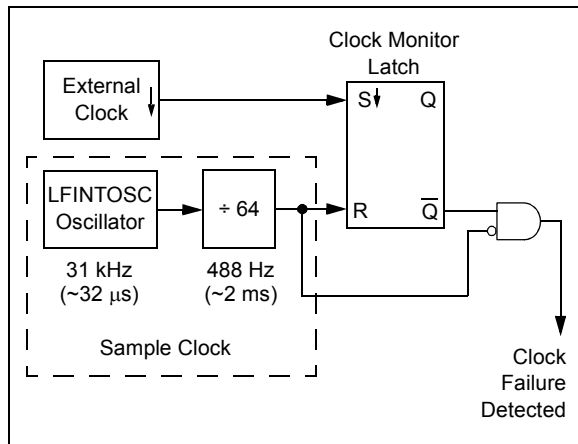
## 4.5 User ID

Four memory locations (8000h-8003h) are designated as ID locations where the user can store checksum or other code identification numbers. These locations are readable and writable during normal execution. See **Section 10.4 “User ID, Device ID and Configuration Word Access”** for more information on accessing these memory locations. For more information on checksum calculation, see the *“PIC16(L)F170X Memory Programming Specification”* (DS41683).

## 6.5 Fail-Safe Clock Monitor

The Fail-Safe Clock Monitor (FSCM) allows the device to continue operating should the external oscillator fail. The FSCM can detect oscillator failure any time after the Oscillator Start-up Timer (OST) has expired. The FSCM is enabled by setting the FCMEN bit in the Configuration Words. The FSCM is applicable to all external Oscillator modes (LP, XT, HS, EC, Secondary Oscillator and RC).

**FIGURE 6-9: FSCM BLOCK DIAGRAM**



### 6.5.1 FAIL-SAFE DETECTION

The FSCM module detects a failed oscillator by comparing the external oscillator to the FSCM sample clock. The sample clock is generated by dividing the LFINTOSC by 64. See Figure 6-9. Inside the fail detector block is a latch. The external clock sets the latch on each falling edge of the external clock. The sample clock clears the latch on each rising edge of the sample clock. A failure is detected when an entire half-cycle of the sample clock elapses before the external clock goes low.

### 6.5.2 FAIL-SAFE OPERATION

When the external clock fails, the FSCM switches the device clock to an internal clock source and sets the bit flag OSFIF of the PIR2 register. Setting this flag will generate an interrupt if the OSFIE bit of the PIE2 register is also set. The device firmware can then take steps to mitigate the problems that may arise from a failed clock. The system clock will continue to be sourced from the internal clock source until the device firmware successfully restarts the external oscillator and switches back to external operation.

The internal clock source chosen by the FSCM is determined by the IRCF<3:0> bits of the OSCCON register. This allows the internal oscillator to be configured before a failure occurs.

### 6.5.3 FAIL-SAFE CONDITION CLEARING

The Fail-Safe condition is cleared after a Reset, executing a SLEEP instruction or changing the SCS bits of the OSCCON register. When the SCS bits are changed, the OST is restarted. While the OST is running, the device continues to operate from the INTOSC selected in OSCCON. When the OST times out, the Fail-Safe condition is cleared after successfully switching to the external clock source. The OSFIF bit should be cleared prior to switching to the external clock source. If the Fail-Safe condition still exists, the OSFIF flag will again become set by hardware.

### 6.5.4 RESET OR WAKE-UP FROM SLEEP

The FSCM is designed to detect an oscillator failure after the Oscillator Start-up Timer (OST) has expired. The OST is used after waking up from Sleep and after any type of Reset. The OST is not used with the EC or RC Clock modes so that the FSCM will be active as soon as the Reset or wake-up has completed. When the FSCM is enabled, the Two-Speed Start-up is also enabled. Therefore, the device will always be executing code while the OST is operating.

**Note:** Due to the wide range of oscillator start-up times, the Fail-Safe circuit is not active during oscillator start-up (i.e., after exiting Reset or Sleep). After an appropriate amount of time, the user should check the Status bits in the OSCSTAT register to verify the oscillator start-up and that the system clock switchover has successfully completed.

## REGISTER 7-5: PIR1: PERIPHERAL INTERRUPT REQUEST REGISTER 1

R/W-0/0	R/W-0/0	R-0/0	R-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
TMR1GIF	ADIF	RCIF	TXIF	SSP1IF	CCP1IF	TMR2IF	TMR1IF
bit 7							bit 0

### Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

u = Bit is unchanged

x = Bit is unknown

-n/n = Value at POR and BOR/Value at all other Resets

'1' = Bit is set

'0' = Bit is cleared

bit 7	<b>TMR1GIF:</b> Timer1 Gate Interrupt Flag bit 1 = Interrupt is pending 0 = Interrupt is not pending
bit 6	<b>ADIF:</b> Analog-to-Digital Converter (ADC) Interrupt Flag bit 1 = Interrupt is pending 0 = Interrupt is not pending
bit 5	<b>RCIF:</b> USART Receive Interrupt Flag bit 1 = Interrupt is pending 0 = Interrupt is not pending
bit 4	<b>TXIF:</b> USART Transmit Interrupt Flag bit 1 = Interrupt is pending 0 = Interrupt is not pending
bit 3	<b>SSP1IF:</b> Synchronous Serial Port (MSSP) Interrupt Flag bit 1 = Interrupt is pending 0 = Interrupt is not pending
bit 2	<b>CCP1IF:</b> CCP1 Interrupt Flag bit 1 = Interrupt is pending 0 = Interrupt is not pending
bit 1	<b>TMR2IF:</b> Timer2 to PR2 Interrupt Flag bit 1 = Interrupt is pending 0 = Interrupt is not pending
bit 0	<b>TMR1IF:</b> Timer1 Overflow Interrupt Flag bit 1 = Interrupt is pending 0 = Interrupt is not pending

**Note:** Interrupt flag bits are set when an interrupt condition occurs, regardless of the state of its corresponding enable bit or the Global Enable bit, GIE, of the INTCON register. User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt.

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**REGISTER 16-3: CMOUT: COMPARATOR OUTPUT REGISTER**

U-0	U-0	U-0	U-0	U-0	U-0	R-0/0	R-0/0
—	—	—	—	—	—	MC2OUT	MC1OUT
bit 7						bit 0	

**Legend:**

R = Readable bit                      W = Writable bit                      U = Unimplemented bit, read as '0'  
u = Bit is unchanged                  x = Bit is unknown                  -n/n = Value at POR and BOR/Value at all other Resets  
'1' = Bit is set                          '0' = Bit is cleared

bit 7-2              **Unimplemented:** Read as '0'  
bit 1                **MC2OUT:** Mirror Copy of C2OUT bit  
bit 0                **MC1OUT:** Mirror Copy of C1OUT bit

**TABLE 16-3: SUMMARY OF REGISTERS ASSOCIATED WITH COMPARATOR MODULE**

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
ANSELA	—	—	ANSA5	ANSA4	ANSA3	ANSA2	ANSA1	ANSA0	120
ANSELB	—	—	ANSB5	ANSB4	ANSB3	ANSB2	ANSB1	ANSB0	126
CM1CON0	C1ON	C1OUT	—	C1POL	C1ZLF	C1SP	C1HYS	C1SYNC	160
CM2CON0	C2ON	C2OUT	—	C2POL	C2ZLF	C2SP	C2HYS	C2SYNC	160
CM1CON1	C1NTP	C1INTN	C1PCH<2:0>			C1NCH<2:0>			161
CM2CON1	C2NTP	C2INTN	C2PCH<2:0>			C2NCH<2:0>			161
CMOUT	—	—	—	—	—	—	MC2OUT	MC1OUT	162
FVRCON	FVREN	FVRRDY	TSEN	TSRNG	CDAFVR<1:0>		ADFVR<1:0>		151
DAC1CON0	DAC1EN	—	DAC1OE1	DAC1OE2	DAC1PSS<1:0>		—	DAC1NSS	249
DAC1CON1	DAC1R<7:0>								249
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	83
PIE2	OSFIE	C2IE	C1IE	—	BCL1IE	TMR6IE	TMR4IE	CCP2IE	85
PIR2	OSFIF	C2IF	C1IF	—	BCL1IF	TMR6IF	TMR4IF	CCP2IF	88
TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	119
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	125
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	130
RxyPPS	—	—	—	RxyPPS<4:0>					137

**Legend:** — = unimplemented location, read as '0'. Shaded cells are unused by the comparator module.

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## REGISTER 17-2: PWMxDCH: PWM DUTY CYCLE HIGH BITS

R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
PWMxDCH<7:0>							
bit 7							
bit 0							

### Legend:

R = Readable bit      W = Writable bit      U = Unimplemented bit, read as '0'  
u = Bit is unchanged      x = Bit is unknown      -n/n = Value at POR and BOR/Value at all other Resets  
'1' = Bit is set      '0' = Bit is cleared

bit 7-0      **PWMxDCH<7:0>**: PWM Duty Cycle Most Significant bits  
These bits are the MSBs of the PWM duty cycle. The two LSbs are found in PWMxDCL Register.

## REGISTER 17-3: PWMxDCL: PWM DUTY CYCLE LOW BITS

R/W-x/u	R/W-x/u	U-0	U-0	U-0	U-0	U-0	U-0
PWMxDCL<7:6>							
bit 7							
bit 0							

### Legend:

R = Readable bit      W = Writable bit      U = Unimplemented bit, read as '0'  
u = Bit is unchanged      x = Bit is unknown      -n/n = Value at POR and BOR/Value at all other Resets  
'1' = Bit is set      '0' = Bit is cleared

bit 7-6      **PWMxDCL<7:6>**: PWM Duty Cycle Least Significant bits  
These bits are the LSbs of the PWM duty cycle. The MSBs are found in PWMxDCH Register.

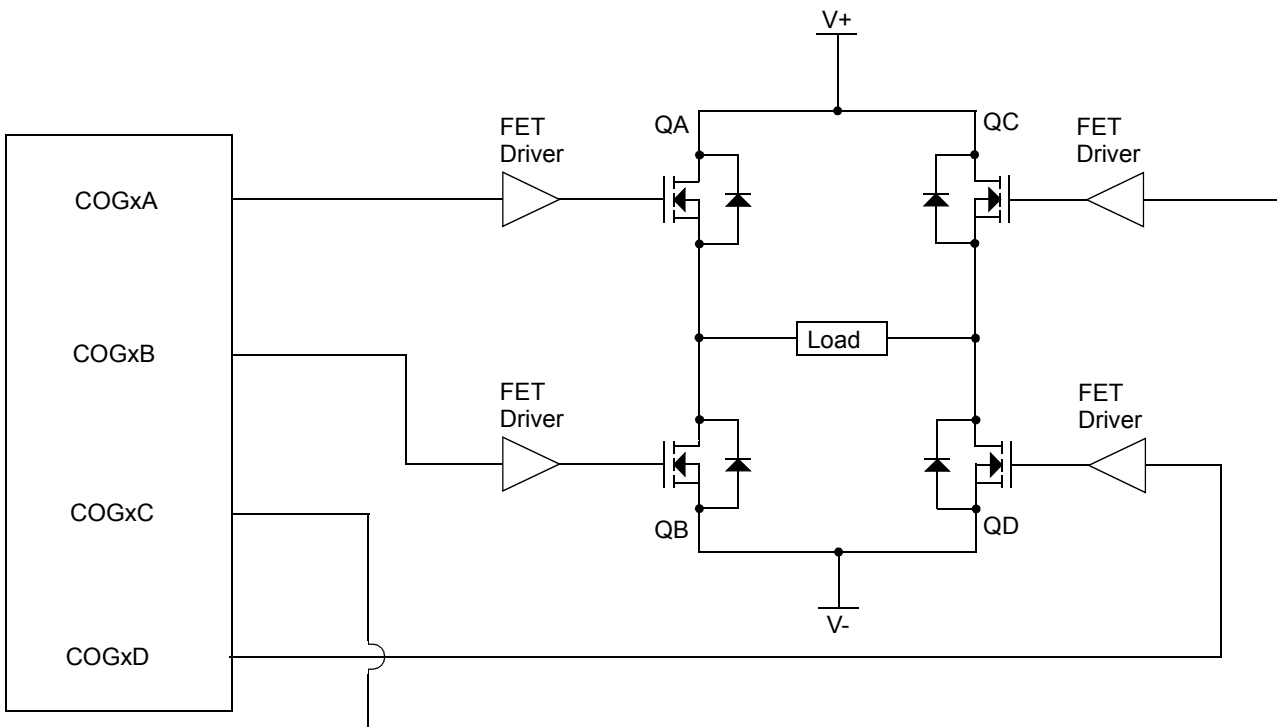
bit 5-0      **Unimplemented**: Read as '0'

## TABLE 17-3: SUMMARY OF REGISTERS ASSOCIATED WITH PWM

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
CCPTMRS	P4TSEL<1:0>		P3TSEL<1:0>		C2TSEL<1:0>		C1TSEL<1:0>		272
PR2	Timer2 module Period Register								268
PWM3CON	PWM3EN	—	PWM3OUT	PWM3POL	—	—	—	—	167
PWM3DCH	PWM3DCH<7:0>								168
PWM3DCL	PWM3DCL<7:6>		—	—	—	—	—	—	168
PWM4CON	PWM4EN	—	PWM4OUT	PWM4POL	—	—	—	—	167
PWM4DCH	PWM4DCH<7:0>								168
PWM4DCL	PWM4DCL<7:6>		—	—	—	—	—	—	168
TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	119
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	125
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	130
RxyPPS	—	—	—	RxyPPS<4:0>					137
T2CON	TOUTPS<3:0>					TMR2ON	T2CKPS<1:0>		270
TMR2	Timer2 module Register								268

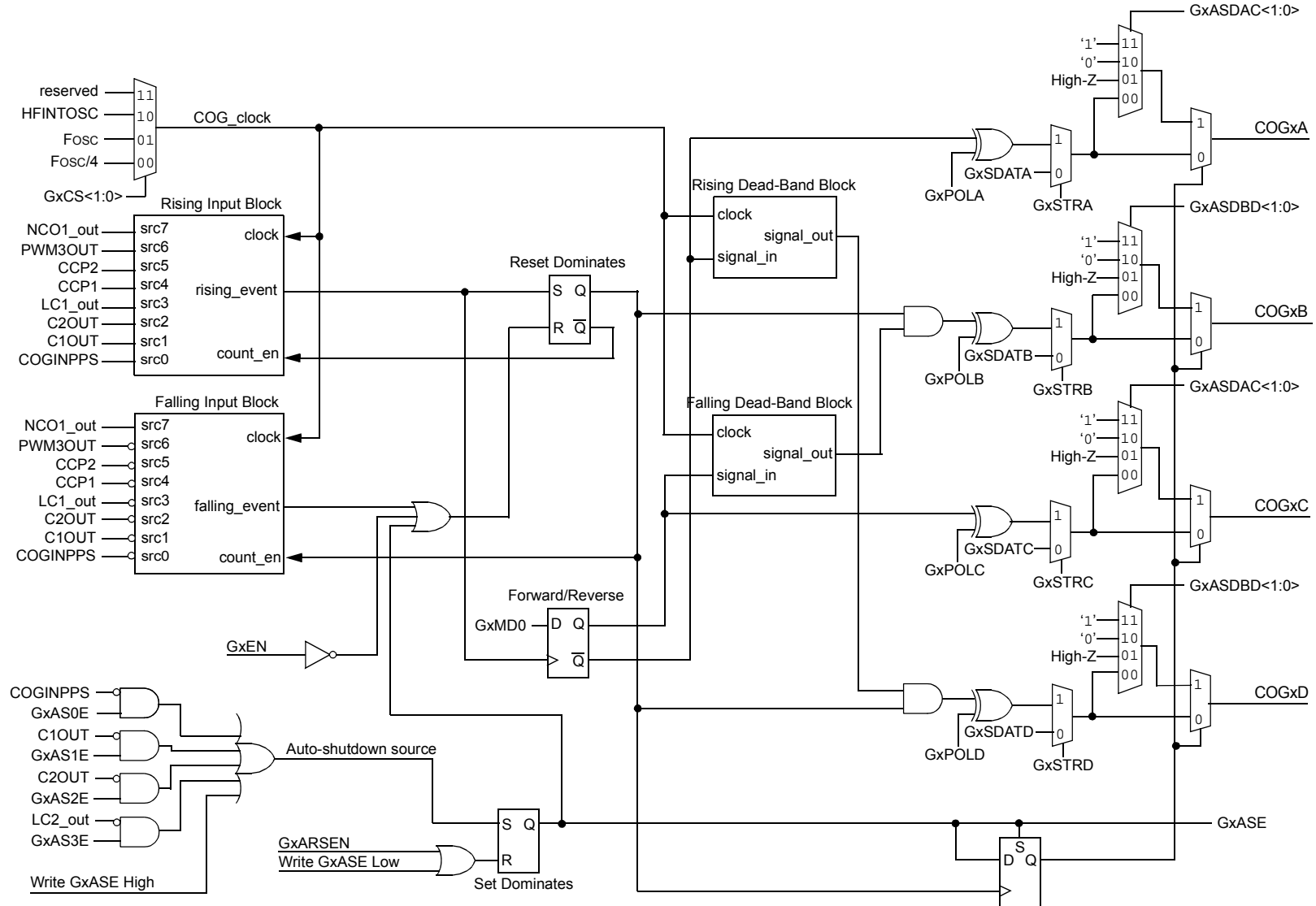
**Legend:** — = Unimplemented locations, read as '0', u = unchanged, x = unknown. Shaded cells are not used by the PWM.

FIGURE 18-1: EXAMPLE OF FULL-BRIDGE APPLICATION





**FIGURE 18-4: SIMPLIFIED COG BLOCK DIAGRAM (FULL-BRIDGE MODES, FORWARD: GXMD = 2, REVERSE: GXMD = 3)**



## REGISTER 18-2: COGxCON1: COG CONTROL REGISTER 1

R/W-0/0	R/W-0/0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
GxRDBS	GxFDBS	—	—	GxPOLD	GxPOLC	GxPOLB	GxPOLA
bit 7							bit 0

### Legend:

R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	q = Value depends on condition

bit 7	<b>GxRDBS:</b> COGx Rising Event Dead-band Timing Source Select bit 1 = Delay chain and COGxDBR are used for dead-band timing generation 0 = COGx_clock and COGxDBR are used for dead-band timing generation
bit 6	<b>GxFDBS:</b> COGx Falling Event Dead-band Timing Source select bit 1 = Delay chain and COGxDF are used for dead-band timing generation 0 = COGx_clock and COGxDBF are used for dead-band timing generation
bit 5-4	<b>Unimplemented:</b> Read as '0'.
bit 3	<b>GxPOLD:</b> COGxD Output Polarity Control bit 1 = Active level of COGxD output is low 0 = Active level of COGxD output is high
bit 2	<b>GxPOLC:</b> COGxC Output Polarity Control bit 1 = Active level of COGxC output is low 0 = Active level of COGxC output is high
bit 1	<b>GxPOLB:</b> COGxB Output Polarity Control bit 1 = Active level of COGxB output is low 0 = Active level of COGxB output is high
bit 0	<b>GxPOLA:</b> COGxA Output Polarity Control bit 1 = Active level of COGxA output is low 0 = Active level of COGxA output is high

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Refer to **Section 29.2.4 “Auto-Conversion Trigger”** for more information.

- Note 1:** The Auto-conversion Trigger from the CCP module does not set interrupt flag bit TMR1IF of the PIR1 register.
- 2:** Removing the match condition by changing the contents of the CCPRxH and CCPRxL register pair, between the clock edge that generates the Auto-conversion Trigger and the clock edge that generates the Timer1 Reset, will preclude the Reset from occurring.

## 29.2.5 COMPARE DURING SLEEP

The Compare mode is dependent upon the system clock (Fosc) for proper operation. Since Fosc is shut down during Sleep mode, the Compare mode will not function properly during Sleep.

## 29.3 PWM Overview

Pulse-Width Modulation (PWM) is a scheme that provides power to a load by switching quickly between fully on and fully off states. The PWM signal resembles a square wave where the high portion of the signal is considered the on state and the low portion of the signal is considered the off state. The high portion, also known as the pulse width, can vary in time and is defined in steps. A larger number of steps applied, which lengthens the pulse width, also supplies more power to the load. Lowering the number of steps applied, which shortens the pulse width, supplies less power. The PWM period is defined as the duration of one complete cycle or the total amount of on and off time combined.

PWM resolution defines the maximum number of steps that can be present in a single PWM period. A higher resolution allows for more precise control of the pulse width time and in turn the power that is applied to the load.

The term duty cycle describes the proportion of the on time to the off time and is expressed in percentages, where 0% is fully off and 100% is fully on. A lower duty cycle corresponds to less power applied and a higher duty cycle corresponds to more power applied.

Figure 29-3 shows a typical waveform of the PWM signal.

## 29.3.1 STANDARD PWM OPERATION

The standard PWM function described in this section is available and identical for all CCP modules.

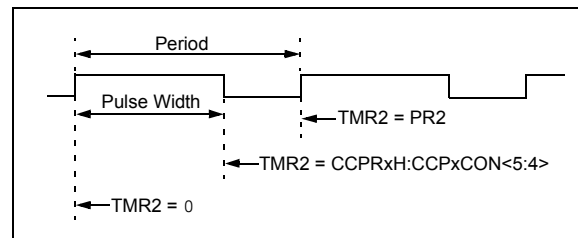
The standard PWM mode generates a Pulse-Width Modulation (PWM) signal on the CCPx pin with up to 10 bits of resolution. The period, duty cycle, and resolution are controlled by the following registers:

- PR2 registers
- T2CON registers
- CCPRxL registers
- CCPxCON registers

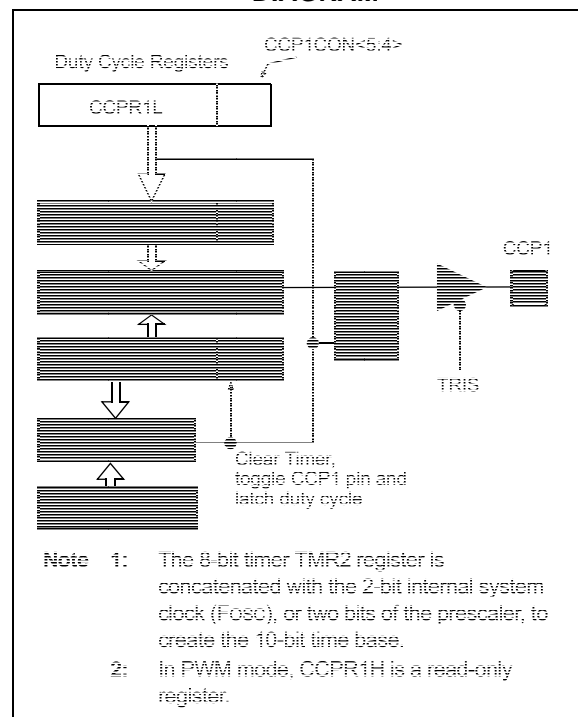
Figure 29-4 shows a simplified block diagram of PWM operation.

**Note:** The corresponding TRIS bit must be cleared to enable the PWM output on the CCPx pin.

**FIGURE 29-3: CCP PWM OUTPUT SIGNAL**



**FIGURE 29-4: SIMPLIFIED PWM BLOCK DIAGRAM**



## 30.5.4 SLAVE MODE 10-BIT ADDRESS RECEPTION

This section describes a standard sequence of events for the MSSP module configured as an I<sup>2</sup>C slave in 10-bit Addressing mode.

Figure 30-20 is used as a visual reference for this description.

This is a step by step process of what must be done by slave software to accomplish I<sup>2</sup>C communication.

1. Bus starts Idle.
2. Master sends Start condition; S bit of SSPSTAT is set; SSPIF is set if interrupt on Start detect is enabled.
3. Master sends matching high address with R/W bit clear; UA bit of the SSPSTAT register is set.
4. Slave sends  $\overline{ACK}$  and SSPIF is set.
5. Software clears the SSPIF bit.
6. Software reads received address from SSPBUF clearing the BF flag.
7. Slave loads low address into SSPADD, releasing SCL.
8. Master sends matching low address byte to the slave; UA bit is set.

**Note:** Updates to the SSPADD register are not allowed until after the  $\overline{ACK}$  sequence.

9. Slave sends  $\overline{ACK}$  and SSPIF is set.

**Note:** If the low address does not match, SSPIF and UA are still set so that the slave software can set SSPADD back to the high address. BF is not set because there is no match. CKP is unaffected.

10. Slave clears SSPIF.
11. Slave reads the received matching address from SSPBUF clearing BF.
12. Slave loads high address into SSPADD.
13. Master clocks a data byte to the slave and clocks out the slaves  $\overline{ACK}$  on the 9th SCL pulse; SSPIF is set.
14. If SEN bit of SSPCON2 is set, CKP is cleared by hardware and the clock is stretched.
15. Slave clears SSPIF.
16. Slave reads the received byte from SSPBUF clearing BF.
17. If SEN is set the slave sets CKP to release the SCL.
18. Steps 13-17 repeat for each received byte.
19. Master sends Stop to end the transmission.

## 30.5.5 10-BIT ADDRESSING WITH ADDRESS OR DATA HOLD

Reception using 10-bit addressing with AHEN or DHEN set is the same as with 7-bit modes. The only difference is the need to update the SSPADD register using the UA bit. All functionality, specifically when the CKP bit is cleared and SCL line is held low are the same. Figure 30-21 can be used as a reference of a slave in 10-bit addressing with AHEN set.

Figure 30-22 shows a standard waveform for a slave transmitter in 10-bit Addressing mode.

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## 30.6.6.1 BF Status Flag

In Transmit mode, the BF bit of the SSPSTAT register is set when the CPU writes to SSPBUF and is cleared when all eight bits are shifted out.

## 30.6.6.2 WCOL Status Flag

If the user writes the SSPBUF when a transmit is already in progress (i.e., SSPSR is still shifting out a data byte), the WCOL bit is set and the contents of the buffer are unchanged (the write does not occur).

WCOL must be cleared by software before the next transmission.

## 30.6.6.3 ACKSTAT Status Flag

In Transmit mode, the ACKSTAT bit of the SSPCON2 register is cleared when the slave has sent an Acknowledge ( $\overline{\text{ACK}} = 0$ ) and is set when the slave does not Acknowledge ( $\text{ACK} = 1$ ). A slave sends an Acknowledge when it has recognized its address (including a general call), or when the slave has properly received its data.

## 30.6.6.4 Typical transmit sequence:

1. The user generates a Start condition by setting the SEN bit of the SSPCON2 register.
2. SSPIF is set by hardware on completion of the Start.
3. SSPIF is cleared by software.
4. The MSSP module will wait the required start time before any other operation takes place.
5. The user loads the SSPBUF with the slave address to transmit.
6. Address is shifted out the SDA pin until all eight bits are transmitted. Transmission begins as soon as SSPBUF is written to.
7. The MSSP module shifts in the  $\overline{\text{ACK}}$  bit from the slave device and writes its value into the ACKSTAT bit of the SSPCON2 register.
8. The MSSP module generates an interrupt at the end of the ninth clock cycle by setting the SSPIF bit.
9. The user loads the SSPBUF with eight bits of data.
10. Data is shifted out the SDA pin until all eight bits are transmitted.
11. The MSSP module shifts in the  $\overline{\text{ACK}}$  bit from the slave device and writes its value into the ACKSTAT bit of the SSPCON2 register.
12. Steps 8-11 are repeated for all transmitted data bytes.
13. The user generates a Stop or Restart condition by setting the PEN or RSEN bits of the SSPCON2 register. Interrupt is generated once the Stop/Restart condition is complete.

## 31.1.2.4 Receive Framing Error

Each character in the receive FIFO buffer has a corresponding framing error Status bit. A framing error indicates that a Stop bit was not seen at the expected time. The framing error status is accessed via the FERR bit of the RCSTA register. The FERR bit represents the status of the top unread character in the receive FIFO. Therefore, the FERR bit must be read before reading the RCREG.

The FERR bit is read-only and only applies to the top unread character in the receive FIFO. A framing error (FERR = 1) does not preclude reception of additional characters. It is not necessary to clear the FERR bit. Reading the next character from the FIFO buffer will advance the FIFO to the next character and the next corresponding framing error.

The FERR bit can be forced clear by clearing the SPEN bit of the RCSTA register which resets the EUSART. Clearing the CREN bit of the RCSTA register does not affect the FERR bit. A framing error by itself does not generate an interrupt.

<b>Note:</b> If all receive characters in the receive FIFO have framing errors, repeated reads of the RCREG will not clear the FERR bit.
--

## 31.1.2.5 Receive Overrun Error

The receive FIFO buffer can hold two characters. An overrun error will be generated if a third character, in its entirety, is received before the FIFO is accessed. When this happens the OERR bit of the RCSTA register is set. The characters already in the FIFO buffer can be read but no additional characters will be received until the error is cleared. The error must be cleared by either clearing the CREN bit of the RCSTA register or by resetting the EUSART by clearing the SPEN bit of the RCSTA register.

## 31.1.2.6 Receiving 9-Bit Characters

The EUSART supports 9-bit character reception. When the RX9 bit of the RCSTA register is set the EUSART will shift nine bits into the RSR for each character received. The RX9D bit of the RCSTA register is the ninth and Most Significant data bit of the top unread character in the receive FIFO. When reading 9-bit data from the receive FIFO buffer, the RX9D data bit must be read before reading the eight Least Significant bits from the RCREG.

## 31.1.2.7 Address Detection

A special Address Detection mode is available for use when multiple receivers share the same transmission line, such as in RS-485 systems. Address detection is enabled by setting the ADDEN bit of the RCSTA register.

Address detection requires 9-bit character reception. When address detection is enabled, only characters with the ninth data bit set will be transferred to the receive FIFO buffer, thereby setting the RCIF interrupt bit. All other characters will be ignored.

Upon receiving an address character, user software determines if the address matches its own. Upon address match, user software must disable address detection by clearing the ADDEN bit before the next Stop bit occurs. When user software detects the end of the message, determined by the message protocol used, software places the receiver back into the Address Detection mode by setting the ADDEN bit.

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**TABLE 31-5: BAUD RATES FOR ASYNCHRONOUS MODES (CONTINUED)**

BAUD RATE	SYNC = 0, BRGH = 1, BRG16 = 1 or SYNC = 1, BRG16 = 1											
	Fosc = 32.000 MHz			Fosc = 20.000 MHz			Fosc = 18.432 MHz			Fosc = 11.0592 MHz		
	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)
300	300.0	0.00	26666	300.0	0.00	16665	300.0	0.00	15359	300.0	0.00	9215
1200	1200	0.00	6666	1200	-0.01	4166	1200	0.00	3839	1200	0.00	2303
2400	2400	0.01	3332	2400	0.02	2082	2400	0.00	1919	2400	0.00	1151
9600	9604	0.04	832	9597	-0.03	520	9600	0.00	479	9600	0.00	287
10417	10417	0.00	767	10417	0.00	479	10425	0.08	441	10433	0.16	264
19.2k	19.18k	-0.08	416	19.23k	0.16	259	19.20k	0.00	239	19.20k	0.00	143
57.6k	57.55k	-0.08	138	57.47k	-0.22	86	57.60k	0.00	79	57.60k	0.00	47
115.2k	115.9k	0.64	68	116.3k	0.94	42	115.2k	0.00	39	115.2k	0.00	23

BAUD RATE	SYNC = 0, BRGH = 1, BRG16 = 1 or SYNC = 1, BRG16 = 1											
	Fosc = 8.000 MHz			Fosc = 4.000 MHz			Fosc = 3.6864 MHz			Fosc = 1.000 MHz		
	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)
300	300.0	0.00	6666	300.0	0.01	3332	300.0	0.00	3071	300.1	0.04	832
1200	1200	-0.02	1666	1200	0.04	832	1200	0.00	767	1202	0.16	207
2400	2401	0.04	832	2398	0.08	416	2400	0.00	383	2404	0.16	103
9600	9615	0.16	207	9615	0.16	103	9600	0.00	95	9615	0.16	25
10417	10417	0	191	10417	0.00	95	10473	0.53	87	10417	0.00	23
19.2k	19.23k	0.16	103	19.23k	0.16	51	19.20k	0.00	47	19.23k	0.16	12
57.6k	57.14k	-0.79	34	58.82k	2.12	16	57.60k	0.00	15	—	—	—
115.2k	117.6k	2.12	16	111.1k	-3.55	8	115.2k	0.00	7	—	—	—

## 31.5.1.5 Synchronous Master Reception

Data is received at the RX/DT pin. The RX/DT pin output driver is automatically disabled when the EUSART is configured for synchronous master receive operation.

In Synchronous mode, reception is enabled by setting either the Single Receive Enable bit (SREN of the RCSTA register) or the Continuous Receive Enable bit (CREN of the RCSTA register).

When SREN is set and CREN is clear, only as many clock cycles are generated as there are data bits in a single character. The SREN bit is automatically cleared at the completion of one character. When CREN is set, clocks are continuously generated until CREN is cleared. If CREN is cleared in the middle of a character the CK clock stops immediately and the partial character is discarded. If SREN and CREN are both set, then SREN is cleared at the completion of the first character and CREN takes precedence.

To initiate reception, set either SREN or CREN. Data is sampled at the RX/DT pin on the trailing edge of the TX/CK clock pin and is shifted into the Receive Shift Register (RSR). When a complete character is received into the RSR, the RCIF bit is set and the character is automatically transferred to the two character receive FIFO. The Least Significant eight bits of the top character in the receive FIFO are available in RCREG. The RCIF bit remains set as long as there are unread characters in the receive FIFO.

<b>Note:</b> If the RX/DT function is on an analog pin, the corresponding ANSEL bit must be cleared for the receiver to function.
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## 31.5.1.6 Slave Clock

Synchronous data transfers use a separate clock line, which is synchronous with the data. A device configured as a slave receives the clock on the TX/CK line. The TX/CK pin output driver is automatically disabled when the device is configured for synchronous slave transmit or receive operation. Serial data bits change on the leading edge to ensure they are valid at the trailing edge of each clock. One data bit is transferred for each clock cycle. Only as many clock cycles should be received as there are data bits.

<b>Note:</b> If the device is configured as a slave and the TX/CK function is on an analog pin, the corresponding ANSEL bit must be cleared.
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## 31.5.1.7 Receive Overrun Error

The receive FIFO buffer can hold two characters. An overrun error will be generated if a third character, in its entirety, is received before RCREG is read to access the FIFO. When this happens the OERR bit of the RCSTA register is set. Previous data in the FIFO will not be overwritten. The two characters in the FIFO buffer can be read, however, no additional characters will be received until the error is cleared. The OERR bit can only be cleared by clearing the overrun condition. If the overrun error occurred when the SREN bit is set and CREN is clear then the error is cleared by reading RCREG. If the overrun occurred when the CREN bit is set then the error condition is cleared by either clearing the CREN bit of the RCSTA register or by clearing the SPEN bit which resets the EUSART.

## 31.5.1.8 Receiving 9-bit Characters

The EUSART supports 9-bit character reception. When the RX9 bit of the RCSTA register is set the EUSART will shift nine bits into the RSR for each character received. The RX9D bit of the RCSTA register is the ninth, and Most Significant, data bit of the top unread character in the receive FIFO. When reading 9-bit data from the receive FIFO buffer, the RX9D data bit must be read before reading the eight Least Significant bits from the RCREG.

## 31.5.1.9 Synchronous Master Reception Setup:

1. Initialize the SPBRGH, SPBRGL register pair for the appropriate baud rate. Set or clear the BRGH and BRG16 bits, as required, to achieve the desired baud rate.
2. Clear the ANSEL bit for the RX pin (if applicable).
3. Enable the synchronous master serial port by setting bits SYNC, SPEN and CSRC.
4. Ensure bits CREN and SREN are clear.
5. If interrupts are desired, set the RCIE bit of the PIE1 register and the GIE and PEIE bits of the INTCON register.
6. If 9-bit reception is desired, set bit RX9.
7. Start reception by setting the SREN bit or for continuous reception, set the CREN bit.
8. Interrupt flag bit RCIF will be set when reception of a character is complete. An interrupt will be generated if the enable bit RCIE was set.
9. Read the RCSTA register to get the ninth bit (if enabled) and determine if any error occurred during reception.
10. Read the 8-bit received data by reading the RCREG register.
11. If an overrun error occurs, clear the error by either clearing the CREN bit of the RCSTA register or by clearing the SPEN bit which resets the EUSART.



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FIGURE 34-1: VOLTAGE FREQUENCY GRAPH,  $-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$ , PIC16F1713/6 ONLY

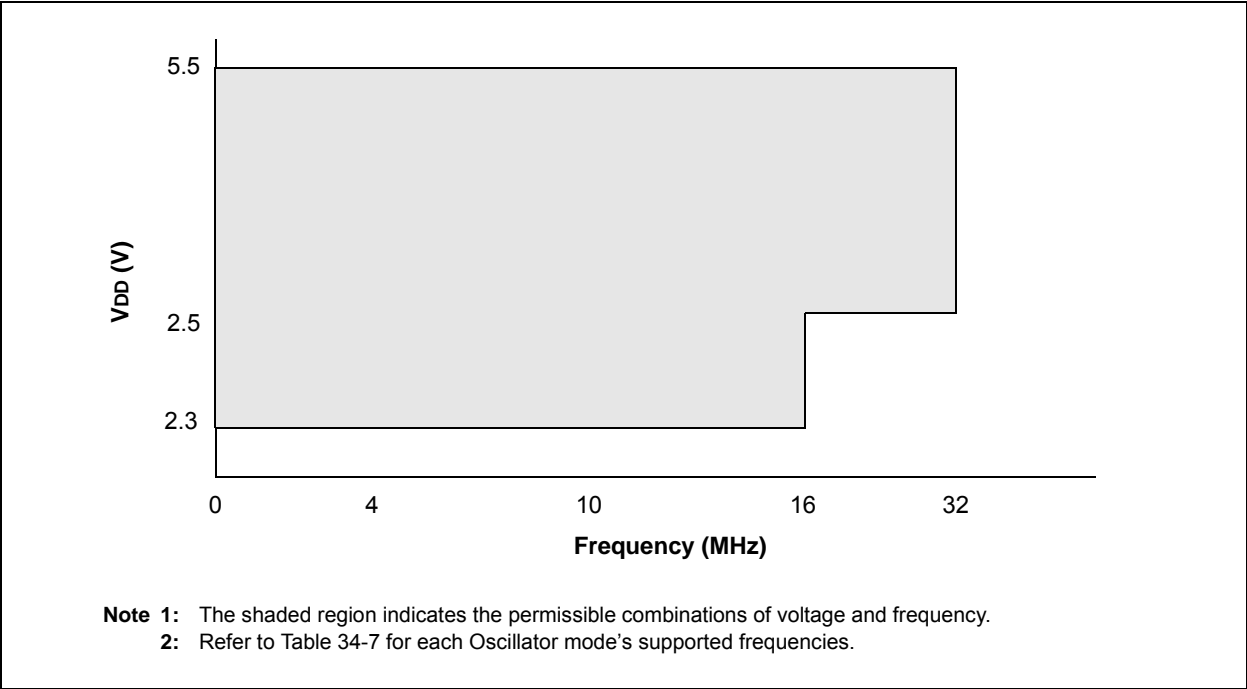
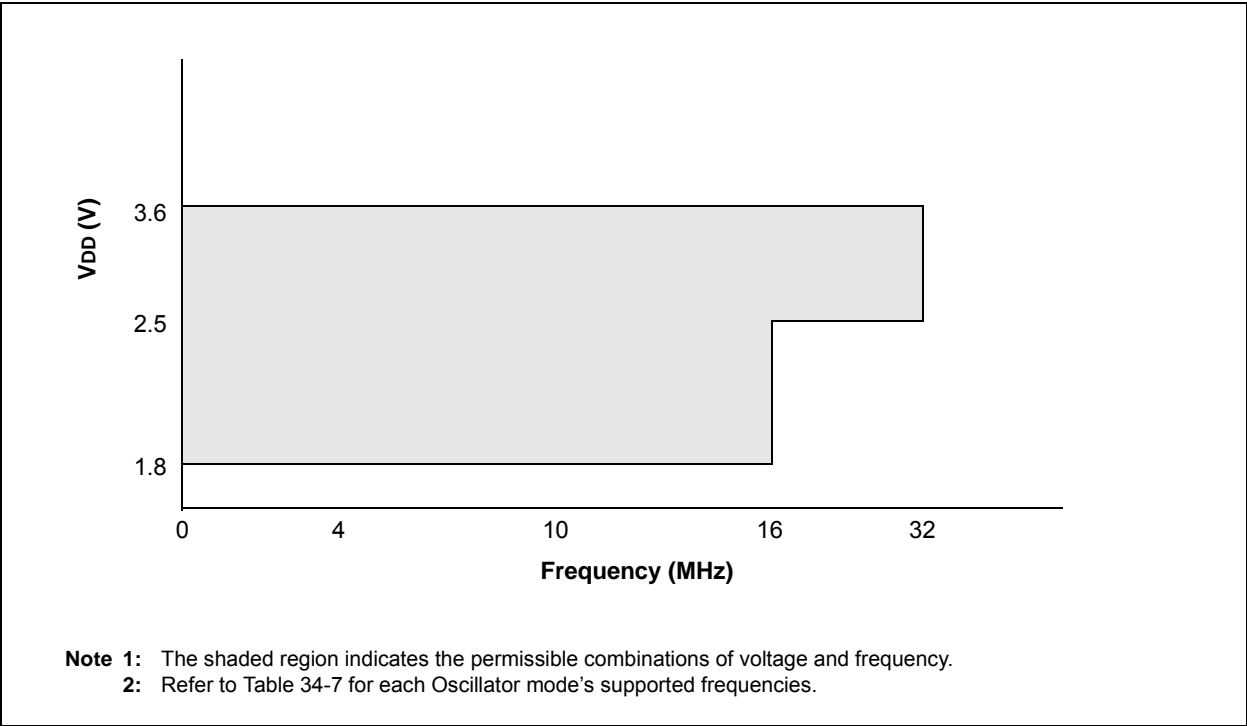


FIGURE 34-2: VOLTAGE FREQUENCY GRAPH,  $-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$ , PIC16LF1713/6 ONLY



**TABLE 34-9: PLL CLOCK TIMING SPECIFICATIONS**

Standard Operating Conditions (unless otherwise stated)							
Param No.	Sym.	Characteristic	Min.	Typ†	Max.	Units	Conditions
F10	FOSC	Oscillator Frequency Range	4	—	8	MHz	
F11	FSYS	On-Chip VCO System Frequency	16	—	32	MHz	
F12	TRC	PLL Start-up Time (Lock Time)	—	—	2	ms	
F13*	$\Delta$ CLK	CLKOUT Stability (Jitter)	-0.25%	—	+0.25%	%	

\* These parameters are characterized but not tested.

† Data in “Typ” column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

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**TABLE 34-26: I<sup>2</sup>C BUS DATA REQUIREMENTS**

Standard Operating Conditions (unless otherwise stated)							
Param. No.	Symbol	Characteristic		Min.	Max.	Units	Conditions
SP100*	THIGH	Clock high time	100 kHz mode	4.0	—	μs	Device must operate at a minimum of 1.5 MHz
			400 kHz mode	0.6	—	μs	Device must operate at a minimum of 10 MHz
			SSP module	1.5Tcy	—		
SP101*	TLOW	Clock low time	100 kHz mode	4.7	—	μs	Device must operate at a minimum of 1.5 MHz
			400 kHz mode	1.3	—	μs	Device must operate at a minimum of 10 MHz
			SSP module	1.5Tcy	—		
SP102*	TR	SDA and SCL rise time	100 kHz mode	—	1000	ns	
			400 kHz mode	20 + 0.1Cb	300	ns	Cb is specified to be from 10-400 pF
SP103*	TF	SDA and SCL fall time	100 kHz mode	—	250	ns	
			400 kHz mode	20 + 0.1Cb	250	ns	Cb is specified to be from 10-400 pF
SP106*	THD:DAT	Data input hold time	100 kHz mode	0	—	ns	
			400 kHz mode	0	0.9	μs	
SP107*	TSU:DAT	Data input setup time	100 kHz mode	250	—	ns	(Note 2)
			400 kHz mode	100	—	ns	
SP109*	TAA	Output valid from clock	100 kHz mode	—	3500	ns	(Note 1)
			400 kHz mode	—	—	ns	
SP110*	TBUF	Bus free time	100 kHz mode	4.7	—	μs	Time the bus must be free before a new transmission can start
			400 kHz mode	1.3	—	μs	
SP111	Cb	Bus capacitive loading		—	400	pF	

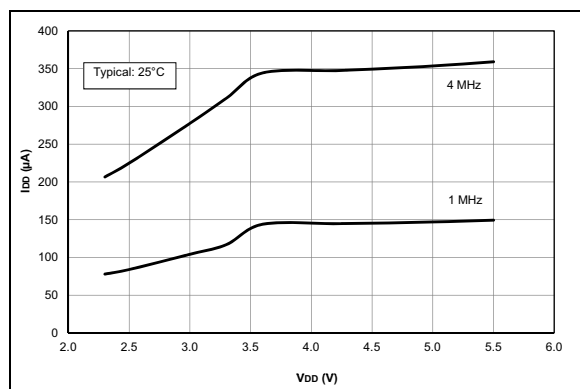
\* These parameters are characterized but not tested.

**Note 1:** As a transmitter, the device must provide this internal minimum delay time to bridge the undefined region (min. 300 ns) of the falling edge of SCL to avoid unintended generation of Start or Stop conditions.

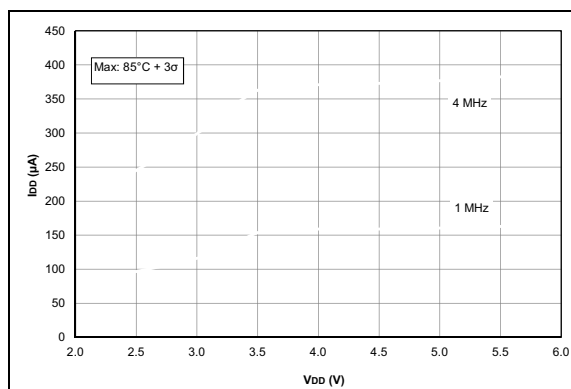
**2:** A Fast mode (400 kHz) I<sup>2</sup>C bus device can be used in a Standard mode (100 kHz) I<sup>2</sup>C bus system, but the requirement TSU:DAT ≥ 250 ns must then be met. This will automatically be the case if the device does not stretch the low period of the SCL signal. If such a device does stretch the low period of the SCL signal, it must output the next data bit to the SDA line TR max. + TSU:DAT = 1000 + 250 = 1250 ns (according to the Standard mode I<sup>2</sup>C bus specification), before the SCL line is released.

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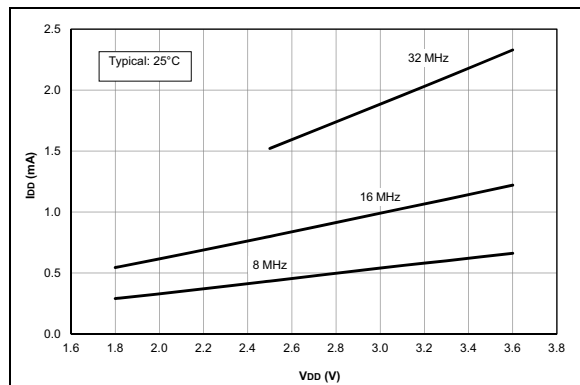
**Note:** Unless otherwise noted,  $V_{IN} = 5V$ ,  $F_{OSC} = 300\text{ kHz}$ ,  $C_{IN} = 0.1\text{ }\mu\text{F}$ ,  $T_A = 25^\circ\text{C}$ .



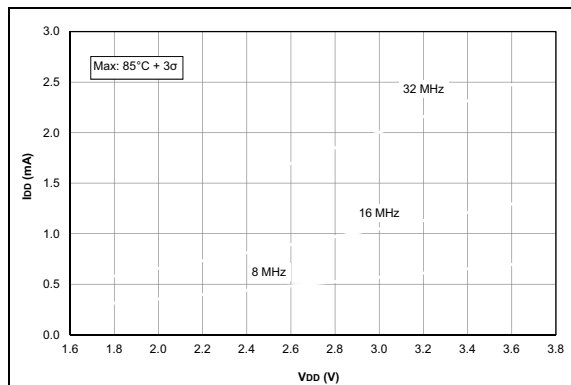
**FIGURE 35-13:**  $I_{DD}$  Typical, EC Oscillator MP Mode, PIC16F1713/6 Only.



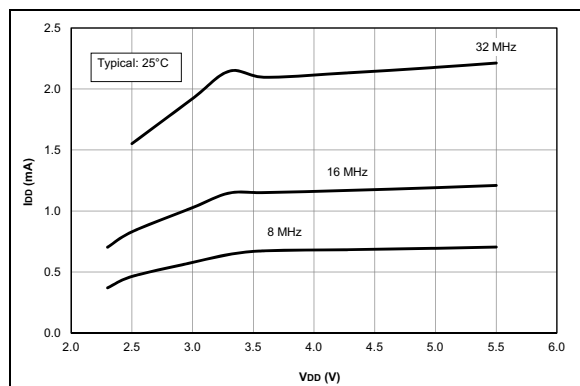
**FIGURE 35-14:**  $I_{DD}$  Maximum, EC Oscillator MP Mode, PIC16F1713/6 Only.



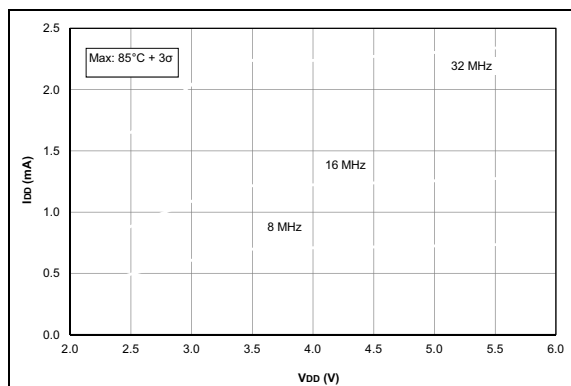
**FIGURE 35-15:**  $I_{DD}$  Typical, EC Oscillator HP Mode, PIC16LF1713/6 Only.



**FIGURE 35-16:**  $I_{DD}$  Maximum, EC Oscillator HP Mode, PIC16LF1713/6 Only.



**FIGURE 35-17:**  $I_{DD}$  Typical, EC Oscillator HP Mode, PIC16F1713/6 Only.



**FIGURE 35-18:**  $I_{DD}$  Maximum, EC Oscillator HP Mode, PIC16F1713/6 Only.

## 36.11 Demonstration/Development Boards, Evaluation Kits, and Starter Kits

A wide variety of demonstration, development and evaluation boards for various PIC MCUs and dsPIC DSCs allows quick application development on fully functional systems. Most boards include prototyping areas for adding custom circuitry and provide application firmware and source code for examination and modification.

The boards support a variety of features, including LEDs, temperature sensors, switches, speakers, RS-232 interfaces, LCD displays, potentiometers and additional EEPROM memory.

The demonstration and development boards can be used in teaching environments, for prototyping custom circuits and for learning about various microcontroller applications.

In addition to the PICDEM™ and dsPICDEM™ demonstration/development board series of circuits, Microchip has a line of evaluation kits and demonstration software for analog filter design, KEELOQ® security ICs, CAN, IrDA®, PowerSmart battery management, SEEVAL® evaluation system, Sigma-Delta ADC, flow rate sensing, plus many more.

Also available are starter kits that contain everything needed to experience the specified device. This usually includes a single application and debug capability, all on one board.

Check the Microchip web page ([www.microchip.com](http://www.microchip.com)) for the complete list of demonstration, development and evaluation kits.

## 36.12 Third-Party Development Tools

Microchip also offers a great collection of tools from third-party vendors. These tools are carefully selected to offer good value and unique functionality.

- Device Programmers and Gang Programmers from companies, such as SoftLog and CCS
- Software Tools from companies, such as Gimpel and Trace Systems
- Protocol Analyzers from companies, such as Saleae and Total Phase
- Demonstration Boards from companies, such as MikroElektronika, Digilent® and Olimex
- Embedded Ethernet Solutions from companies, such as EZ Web Lynx, WIZnet and IPLogika®