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Applications of "<u>Embedded -</u> <u>Microcontrollers</u>"

Details

Details	
Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	20MHz
Connectivity	I ² C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	33
Program Memory Size	14KB (8K x 14)
Program Memory Type	ОТР
EEPROM Size	-
RAM Size	368 x 8
Voltage - Supply (Vcc/Vdd)	4V ~ 6V
Data Converters	A/D 8x8b
Oscillator Type	External
Operating Temperature	0°C ~ 70°C (TA)
Mounting Type	Surface Mount
Package / Case	44-LCC (J-Lead)
Supplier Device Package	44-PLCC (16.59x16.59)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic16c77-20-l

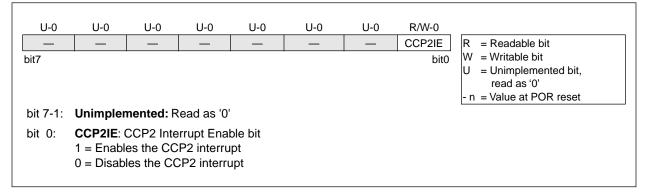
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4.2.2.6 PIE2 REGISTER Applicable Devices 72 73 73 74 74 76 77

This register contains the individual enable bit for the CCP2 peripheral interrupt.

FIGURE 4-14: PIE2 REGISTER (ADDRESS 8Dh)



7.2 Using Timer0 with an External Clock Applicable Devices 72 73 73A 74 74A 76 77

When an external clock input is used for Timer0, it must meet certain requirements. The requirements ensure the external clock can be synchronized with the internal phase clock (Tosc). Also, there is a delay in the actual incrementing of Timer0 after synchronization.

7.2.1 EXTERNAL CLOCK SYNCHRONIZATION

When no prescaler is used, the external clock input is the same as the prescaler output. The synchronization of T0CKI with the internal phase clocks is accomplished by sampling the prescaler output on the Q2 and Q4 cycles of the internal phase clocks (Figure 7-5). Therefore, it is necessary for T0CKI to be high for at least 2Tosc (and a small RC delay of 20 ns) and low for at least 2Tosc (and a small RC delay of 20 ns). Refer to the electrical specification of the desired device. When a prescaler is used, the external clock input is divided by the asynchronous ripple-counter type prescaler so that the prescaler output is symmetrical. For the external clock to meet the sampling requirement, the ripple-counter must be taken into account. Therefore, it is necessary for TOCKI to have a period of at least 4Tosc (and a small RC delay of 40 ns) divided by the prescaler value. The only requirement on TOCKI high and low time is that they do not violate the minimum pulse width requirement of 10 ns. Refer to parameters 40, 41 and 42 in the electrical specification of the desired device.

7.2.2 TMR0 INCREMENT DELAY

Since the prescaler output is synchronized with the internal clocks, there is a small delay from the time the external clock edge occurs to the time the Timer0 module is actually incremented. Figure 7-5 shows the delay from the external clock edge to the timer incrementing.

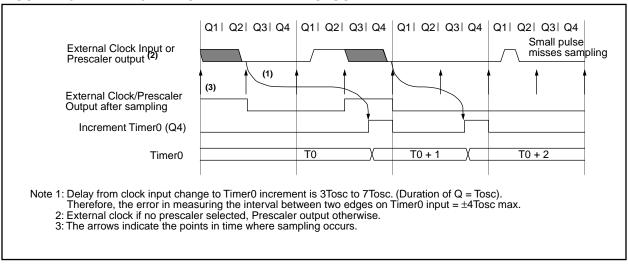
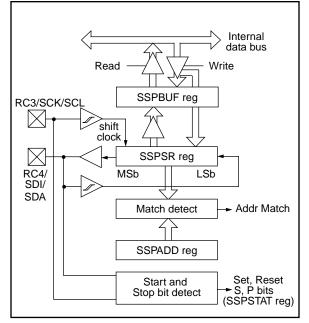


FIGURE 7-5: TIMER0 TIMING WITH EXTERNAL CLOCK

11.5 <u>SSP I²C Operation</u>

The SSP module in I²C mode fully implements all slave functions, except general call support, and provides interrupts on start and stop bits in hardware to facilitate firmware implementations of the master functions. The SSP module implements the standard mode specifications as well as 7-bit and 10-bit addressing. Two pins are used for data transfer. These are the RC3/SCK/SCL pin, which is the clock (SCL), and the RC4/SDI/SDA pin, which is the data (SDA). The user must configure these pins as inputs or outputs through the TRISC<4:3> bits. The SSP module functions are enabled by setting SSP Enable bit SSPEN (SSP-CON<5>).

FIGURE 11-24: SSP BLOCK DIAGRAM (I²C MODE)



The SSP module has five registers for ${\rm I}^2{\rm C}$ operation. These are the:

- SSP Control Register (SSPCON)
- SSP Status Register (SSPSTAT)
- Serial Receive/Transmit Buffer (SSPBUF)
- SSP Shift Register (SSPSR) Not directly accessible
- SSP Address Register (SSPADD)

The SSPCON register allows control of the I^2C operation. Four mode selection bits (SSPCON<3:0>) allow one of the following I^2C modes to be selected:

- I²C Slave mode (7-bit address)
- I²C Slave mode (10-bit address)
- I²C Slave mode (7-bit address), with start and stop bit interrupts enabled
- I²C Slave mode (10-bit address), with start and stop bit interrupts enabled
- I²C Firmware controlled Master Mode, slave is idle

Selection of any I²C mode, with the SSPEN bit set, forces the SCL and SDA pins to be open drain, provided these pins are programmed to inputs by setting the appropriate TRISC bits.

The SSPSTAT register gives the status of the data transfer. This information includes detection of a START or STOP bit, specifies if the received byte was data or address if the next byte is the completion of 10-bit address, and if this will be a read or write data transfer. The SSPSTAT register is read only.

The SSPBUF is the register to which transfer data is written to or read from. The SSPSR register shifts the data in or out of the device. In receive operations, the SSPBUF and SSPSR create a doubled buffered receiver. This allows reception of the next byte to begin before reading the last byte of received data. When the complete byte is received, it is transferred to the SSPBUF register and flag bit SSPIF is set. If another complete byte is received before the SSPBUF register is read, a receiver overflow has occurred and bit SSPOV (SSPCON<6>) is set and the byte in the SSPSR is lost.

The SSPADD register holds the slave address. In 10-bit mode, the user first needs to write the high byte of the address (1111 0 A9 A8 0). Following the high byte address match, the low byte of the address needs to be loaded (A7:A0).

11.5.2 MASTER MODE

Master mode of operation is supported in firmware using interrupt generation on the detection of the START and STOP conditions. The STOP (P) and START (S) bits are cleared from a reset or when the SSP module is disabled. The STOP (P) and START (S) bits will toggle based on the START and STOP conditions. Control of the I^2C bus may be taken when the P bit is set, or the bus is idle and both the S and P bits are clear.

In master mode the SCL and SDA lines are manipulated by clearing the corresponding TRISC<4:3> bit(s). The output level is always low, irrespective of the value(s) in PORTC<4:3>. So when transmitting data, a '1' data bit must have the TRISC<4> bit set (input) and a '0' data bit must have the TRISC<4> bit cleared (output). The same scenario is true for the SCL line with the TRISC<3> bit.

The following events will cause SSP Interrupt Flag bit, SSPIF, to be set (SSP Interrupt if enabled):

- START condition
- STOP condition
- Data transfer byte transmitted/received

Master mode of operation can be done with either the slave mode idle (SSPM3:SSPM0 = 1011) or with the slave active. When both master and slave modes are enabled, the software needs to differentiate the source(s) of the interrupt.

11.5.3 MULTI-MASTER MODE

In multi-master mode, the interrupt generation on the detection of the START and STOP conditions allows the determination of when the bus is free. The STOP (P) and START (S) bits are cleared from a reset or when the SSP module is disabled. The STOP (P) and START (S) bits will toggle based on the START and STOP conditions. Control of the I^2C bus may be taken when bit P (SSPSTAT<4>) is set, or the bus is idle and both the S and P bits clear. When the bus is busy, enabling the SSP Interrupt will generate the interrupt when the STOP condition occurs.

In multi-master operation, the SDA line must be monitored to see if the signal level is the expected output level. This check only needs to be done when a high level is output. If a high level is expected and a low level is present, the device needs to release the SDA and SCL lines (set TRISC<4:3>). There are two stages where this arbitration can be lost, these are:

- Address Transfer
- Data Transfer

When the slave logic is enabled, the slave continues to receive. If arbitration was lost during the address transfer stage, communication to the device may be in progress. If addressed an ACK pulse will be generated. If arbitration was lost during the data transfer stage, the device will need to re-transfer the data at a later time.

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other resets
0Bh, 8Bh, 10Bh,18Bh	INTCON	GIE	PEIE	TOIE	INTE	RBIE	TOIF	INTF	RBIF	0000 000x	0000 000u
0Ch	PIR1	PSPIF ⁽¹⁾	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 0000	0000 0000
8Ch	PIE1	PSPIE ⁽¹⁾	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 0000	0000 0000
13h	SSPBUF	Synchrono	us Serial	Port Rece	eive Buffe	r/Transmit	Register			xxxx xxxx	uuuu uuuu
93h	SSPADD	Synchrono	us Serial	Port (I ² C	mode) Ad	ldress Re	gister			0000 0000	0000 0000
14h	SSPCON	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0	0000 0000	0000 0000
94h	SSPSTAT	SMP ⁽²⁾	CKE ⁽²⁾	D/Ā	Р	S	R/W	UA	BF	0000 0000	0000 0000
87h	TRISC	PORTC Da	ta Directi	on registe	er					1111 1111	1111 1111

TABLE 11-5: REGISTERS ASSOCIATED WITH I²C OPERATION

 $\label{eq:Legend: x = unknown, u = unchanged, - = unimplemented locations read as '0'. Shaded cells are not used by SSP module in SPI mode.$

Note 1: PSPIF and PSPIE are reserved on the PIC16C73/73A/76, always maintain these bits clear.

2: The SMP and CKE bits are implemented on the PIC16C76/77 only. All other PIC16C7X devices have these two bits unimplemented, read as '0'.

12.0 UNIVERSAL SYNCHRONOUS ASYNCHRONOUS RECEIVER TRANSMITTER (USART)

Applicable Devices

72 73 73A 74 74A 76 77

The Universal Synchronous Asynchronous Receiver Transmitter (USART) module is one of the two serial I/O modules. (USART is also known as a Serial Communications Interface or SCI). The USART can be configured as a full duplex asynchronous system that can communicate with peripheral devices such as CRT terminals and personal computers, or it can be configured as a half duplex synchronous system that can communicate with peripheral devices such as A/D or D/A integrated circuits, Serial EEPROMs etc.

The USART can be configured in the following modes:

- Asynchronous (full duplex)
- Synchronous Master (half duplex)
- Synchronous Slave (half duplex)

Bit SPEN (RCSTA<7>), and bits TRISC<7:6>, have to be set in order to configure pins RC6/TX/CK and RC7/ RX/DT as the Universal Synchronous Asynchronous Receiver Transmitter.

FIGURE 12-1: TXSTA: TRANSMIT STATUS AND CONTROL REGISTER (ADDRESS 98h)

R/W-0	R/W-0	R/W-0	R/W-0	U-0	R/W-0	R-1	R/W-0		
CSRC	TX9	TXEN	SYNC	—	BRGH	TRMT	TX9D	R = Readable bit	
bit7							bit0	W = Writable bit U = Unimplemented bit, read as '0' - n =Value at POR reset	
bit 7:	CSRC: Clo	ck Source	Select bit						
	Asynchrone Don't care	<u>ous mode</u>							
	$\frac{\text{Synchrono}}{1 = \text{Master}}$ $0 = \text{Slave n}$	mode (Clo				.G)			
bit 6:	TX9 : 9-bit 7 1 = Selects 0 = Selects	9-bit trans	mission						
bit 5:	TXEN: Transmit Enable bit 1 = Transmit enabled 0 = Transmit disabled Note: SREN/CREN overrides TXEN in SYNC mode.								
bit 4:	SYNC : US/ 1 = Synchr 0 = Asynch	onous mod	le						
bit 3:	Unimplem	ented: Rea	ad as '0'						
bit 2:	BRGH: Hig	h Baud Ra	te Select b	it					
	Asynchrone 1 = High sp								
	Note:	rience a h baud rate	igh rate of	receive er H = 0 can	rors. It is re	commende	ed that BRG	ode (BRGH = 1) may expe- H = 0. If you desire a higher a for additional information,	
	0 = Low sp	eed							
	Synchrono Unused in t								
bit 1:	TRMT : Trar 1 = TSR en 0 = TSR ful	npty	Register St	atus bit					
bit 0:	TX9D : 9th I	bit of transi	mit data. Ca	an be pari	ty bit.				

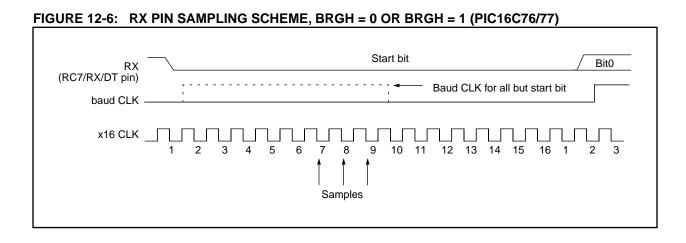
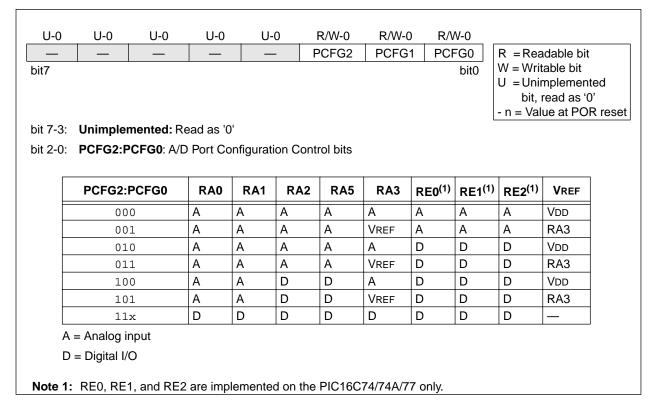


FIGURE 13-2: ADCON1 REGISTER (ADDRESS 9Fh)



13.8 Use of the CCP Trigger Applicable Devices 72 73 73A 74 74A 76 77

Note: In the PIC16C72, the "special event trigger" is implemented in the CCP1 module.

An A/D conversion can be started by the "special event trigger" of the CCP2 module (CCP1 on the PIC16C72 only). This requires that the CCP2M3:CCP2M0 bits (CCP2CON<3:0>) be programmed as 1011 and that the A/D module is enabled (ADON bit is set). When the trigger occurs, the GO/DONE bit will be set, starting the A/D conversion, and the Timer1 counter will be reset to zero. Timer1 is reset to automatically repeat the A/D acquisition period with minimal software overhead (moving the ADRES to the desired location). The appropriate analog input channel must be selected and the minimum acquisition done before the "special event trigger" sets the GO/DONE bit (starts a conversion).

If the A/D module is not enabled (ADON is cleared), then the "special event trigger" will be ignored by the A/D module, but will still reset the Timer1 counter.

13.9 Connection Considerations Applicable Devices 72/73/73A/74/74A/76/77

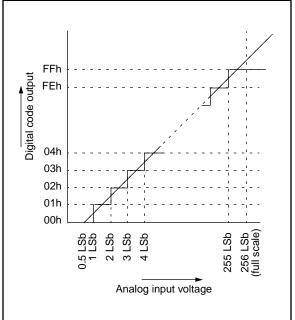
If the input voltage exceeds the rail values (VSS or VDD) by greater than 0.2V, then the accuracy of the conversion is out of specification.

An external RC filter is sometimes added for anti-aliasing of the input signal. The R component should be selected to ensure that the total source impedance is kept under the 10 k Ω recommended specification. Any external components connected (via hi-impedance) to an analog input pin (capacitor, zener diode, etc.) should have very little leakage current at the pin.

13.10 Transfer Function Applicable Devices 72 73 73 74 74 76 77

The ideal transfer function of the A/D converter is as follows: the first transition occurs when the analog input voltage (VAIN) is Analog VREF/256 (Figure 13-5).

FIGURE 13-5: A/D TRANSFER FUNCTION



13.11 References

A very good reference for understanding A/D converters is the "Analog-Digital Conversion Handbook" third edition, published by Prentice Hall (ISBN 0-13-03-2848-0).

FIGURE 14-2: CONFIGURATION WORD FOR PIC16C72/73A/74A/76/77

	P0 CP1	CP0	CP1	CP0	_	BODEN	CP1	CP0	PWRTE	WDTE	FOSC1		Register: Address	CONFIG 2007h
bit13												bit0	Address	200711
bit 13-8														
5-4:														
	10 = Upper half of program memory code protected 01 = Upper 3/4th of program memory code protected													
	01 = Opp 00 = All m					/ code pr	otected	1						
bit 7:	Unimpler	,		•										
bit 6:	BODEN:				hle hit	(1)								
511 0.	1 = BOR (
	0 = BOR (disable	d											
bit 3:	PWRTE:	Power-u	up Time	er Enab	le bit (1)								
	1 = PWR1													
	0 = PWRT	enabl	ed											
bit 2:	WDTE: W			Enabl	e bit									
	1 = WDT		-											
	0 = WDT		-											
bit 1-0:	FOSC1:F			tor Sele	ection	bits								
	11 = RC (10 = HS (
	10 = HSC 01 = XTC													
	01 = 100													
Note 1:	-					•				,	-	ess of the	value of bit F	PWRTE.
~	Ensure th					,								
2:	All of the	CP1:CF	20 pairs	s have t	to be g	jiven the	same \	alue to	o enable	the coo	de prote	ction sch	eme listed.	

14.2.3 EXTERNAL CRYSTAL OSCILLATOR CIRCUIT

Either a prepackaged oscillator can be used or a simple oscillator circuit with TTL gates can be built. Prepackaged oscillators provide a wide operating range and better stability. A well-designed crystal oscillator will provide good performance with TTL gates. Two types of crystal oscillator circuits can be used; one with series resonance, or one with parallel resonance.

Figure 14-5 shows implementation of a parallel resonant oscillator circuit. The circuit is designed to use the fundamental frequency of the crystal. The 74AS04 inverter performs the 180-degree phase shift that a parallel oscillator requires. The 4.7 k Ω resistor provides the negative feedback for stability. The 10 k Ω potentiometer biases the 74AS04 in the linear region. This could be used for external oscillator designs.

FIGURE 14-5: EXTERNAL PARALLEL RESONANT CRYSTAL OSCILLATOR CIRCUIT

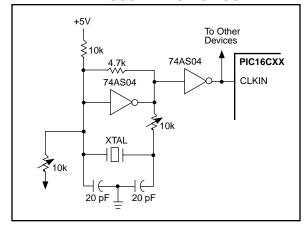
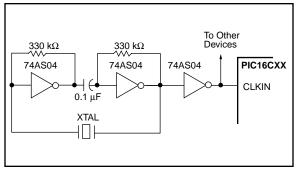


Figure 14-6 shows a series resonant oscillator circuit. This circuit is also designed to use the fundamental frequency of the crystal. The inverter performs a 180-degree phase shift in a series resonant oscillator circuit. The 330 k Ω resistors provide the negative feedback to bias the inverters in their linear region.

FIGURE 14-6: EXTERNAL SERIES RESONANT CRYSTAL OSCILLATOR CIRCUIT



14.2.4 RC OSCILLATOR

For timing insensitive applications the "RC" device option offers additional cost savings. The RC oscillator frequency is a function of the supply voltage, the resistor (Rext) and capacitor (Cext) values, and the operating temperature. In addition to this, the oscillator frequency will vary from unit to unit due to normal process parameter variation. Furthermore, the difference in lead frame capacitance between package types will also affect the oscillation frequency, especially for low Cext values. The user also needs to take into account variation due to tolerance of external R and C components used. Figure 14-7 shows how the R/C combination is connected to the PIC16CXX. For Rext values below 2.2 k Ω , the oscillator operation may become unstable, or stop completely. For very high Rext values (e.g. 1 M Ω), the oscillator becomes sensitive to noise, humidity and leakage. Thus, we recommend to keep Rext between 3 k Ω and 100 k Ω .

Although the oscillator will operate with no external capacitor (Cext = 0 pF), we recommend using values above 20 pF for noise and stability reasons. With no or small external capacitance, the oscillation frequency can vary dramatically due to changes in external capacitances, such as PCB trace capacitance or package lead frame capacitance.

See characterization data for desired device for RC frequency variation from part to part due to normal process variation. The variation is larger for larger R (since leakage current variation will affect RC frequency more for large R) and for smaller C (since variation of input capacitance will affect RC frequency more).

See characterization data for desired device for variation of oscillator frequency due to VDD for given Rext/ Cext values as well as frequency variation due to operating temperature for given R, C, and VDD values.

The oscillator frequency, divided by 4, is available on the OSC2/CLKOUT pin, and can be used for test purposes or to synchronize other logic (see Figure 3-4 for waveform).

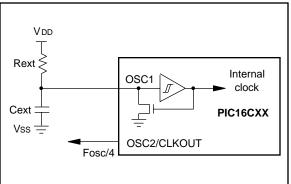


FIGURE 14-7: RC OSCILLATOR MODE

14.4.5 TIME-OUT SEQUENCE

On power-up the time-out sequence is as follows: First PWRT time-out is invoked after the POR time delay has expired. Then OST is activated. The total time-out will vary based on oscillator configuration and the status of the PWRT. For example, in RC mode with the PWRT disabled, there will be no time-out at all. Figure 14-10, Figure 14-11, and Figure 14-12 depict time-out sequences on power-up.

Since the time-outs occur from the POR pulse, if $\overline{\text{MCLR}}$ is kept low long enough, the time-outs will expire. Then bringing $\overline{\text{MCLR}}$ high will begin execution immediately (Figure 14-11). This is useful for testing purposes or to synchronize more than one PIC16CXX device operating in parallel.

Table 14-7 shows the reset conditions for some special function registers, while Table 14-8 shows the reset conditions for all the registers.

14.4.6 POWER CONTROL/STATUS REGISTER (PCON)

-	Applicable Devices							
72	73	73A	74	74A	76	77		

The Power Control/Status Register, PCON has up to two bits, depending upon the device. Bit0 is not implemented on the PIC16C73 or PIC16C74.

Bit0 is Brown-out Reset Status bit, BOR. Bit BOR is unknown on a Power-on Reset. It must then be set by the user and checked on subsequent resets to see if bit BOR cleared, indicating a BOR occurred. The BOR bit is a "Don't Care" bit and is not necessarily predictable if the Brown-out Reset circuitry is disabled (by clearing bit BODEN in the Configuration Word).

Bit1 is POR (Power-on Reset Status bit). It is cleared on a Power-on Reset and unaffected otherwise. The user must set this bit following a Power-on Reset.

TABLE 14-3: TIME-OUT IN VARIOUS SITUATIONS, PIC16C73/74

Oscillator Configuration	Powe	Wake-up from SLEEP	
	PWRTE = 1	PWRTE = 0	
XT, HS, LP	72 ms + 1024Tosc	1024Tosc	1024 Tosc
RC	72 ms	_	

TABLE 14-4: TIME-OUT IN VARIOUS SITUATIONS, PIC16C72/73A/74A/76/77

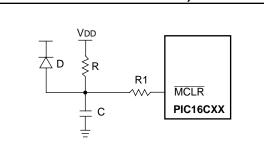
Oscillator Configuration	Power	r-up	Brown-out	Wake-up from SLEEP		
	PWRTE = 0	PWRTE = 1	Brown-out			
XT, HS, LP	72 ms + 1024Tosc	1024Tosc	72 ms + 1024Tosc	1024Tosc		
RC	72 ms		72 ms	_		

TABLE 14-5: STATUS BITS AND THEIR SIGNIFICANCE, PIC16C73/74

POR	TO	PD				
0	1	1	Power-on Reset			
0	0	х	Illegal, TO is set on POR			
0	x	0	Illegal, PD is set on POR			
1	0	1	WDT Reset			
1	0	0	WDT Wake-up			
1	u	u	MCLR Reset during normal operation			
1	1	0	MCLR Reset during SLEEP or interrupt wake-up from SLEEP			

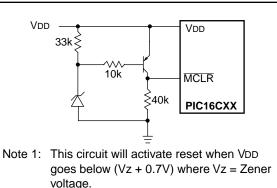
Legend: u = unchanged, x = unknown

FIGURE 14-13: EXTERNAL POWER-ON RESET CIRCUIT (FOR SLOW VDD POWER-UP)



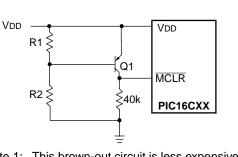
- Note 1: External Power-on Reset circuit is required only if VDD power-up slope is too slow. The diode D helps discharge the capacitor quickly when VDD powers down.
 - R < 40 kΩ is recommended to make sure that voltage drop across R does not violate the device's electrical specification.
 - 3: $R1 = 100\Omega$ to 1 k Ω will limit any current flowing into \overline{MCLR} from external capacitor C in the event of \overline{MCLR}/VPP pin breakdown due to Electrostatic Discharge (ESD) or Electrical Overstress (EOS).

FIGURE 14-14: EXTERNAL BROWN-OUT PROTECTION CIRCUIT 1



- 2: Internal brown-out detection on the PIC16C72/73A/74A/76/77 should be disabled when using this circuit.
- 3: Resistors should be adjusted for the characteristics of the transistor.

FIGURE 14-15: EXTERNAL BROWN-OUT PROTECTION CIRCUIT 2



Note 1: This brown-out circuit is less expensive, albeit less accurate. Transistor Q1 turns off when VDD is below a certain level such that:

$$V_{DD} \bullet \frac{R1}{R1 + R2} = 0.7V$$

- 2: Internal brown-out detection on the PIC16C72/73A/74A/76/77 should be disabled when using this circuit.
- 3: Resistors should be adjusted for the characteristics of the transistor.

COMF	Complement f	DECFSZ	Decrement f, Skip if 0
Syntax:	[<i>label</i>] COMF f,d	Syntax:	[<i>label</i>] DECFSZ f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in \left[0,1\right] \end{array}$	Operands:	$0 \le f \le 127$ $d \in [0,1]$
Operation:	$(\overline{f}) \rightarrow$ (destination)	Operation:	(f) - 1 \rightarrow (destination);
Status Affected:	Z		skip if result = 0
Encoding:	00 1001 dfff ffff	Status Affected:	None
Description:	The contents of register 'f' are comple- mented. If 'd' is 0 the result is stored in	Encoding:	00 1011 dfff ffff
	W. If 'd' is 1 the result is stored back in register 'f'.	Description:	The contents of register 'f' are decre- mented. If 'd' is 0 the result is placed in the W register. If 'd' is 1 the result is placed
Words:	1		back in register 'f'. If the result is 1, the next instruction, is
Cycles:	1		executed. If the result is 0, then a NOP is executed instead making it a 2TCY instruc-
Q Cycle Activity:	Q1 Q2 Q3 Q4		tion.
	Decode Read Process Write to register data destination	Words:	1
	f	Cycles:	1(2)
		Q Cycle Activity:	Q1 Q2 Q3 Q4
Example	COMF REG1,0 Before Instruction		Decode Read register 'f' Process Write to destination
	REG1 = 0x13 After Instruction	If Skip:	(2nd Cycle)
	REG1 = 0x13		Q1 Q2 Q3 Q4
	W = 0xEC		No-No-No-OperationOperationOperation
DECF	Decrement f	E	
DECF Syntax:	Decrement f [<i>label</i>] DECF f,d	Example	HERE DECFSZ CNT, 1 GOTO LOOP
_		Example	
Syntax:	[<i>label</i>] DECF f,d $0 \le f \le 127$	Example	GOTO LOOP
Syntax: Operands:	[<i>label</i>] DECF f,d $0 \le f \le 127$ $d \in [0,1]$	Example	CONTINUE • • • • • • • • • • • • • • • • • • •
Syntax: Operands: Operation:	[<i>label</i>] DECF f,d $0 \le f \le 127$ $d \in [0,1]$ (f) - 1 \rightarrow (destination)	Example	GOTO LOOP CONTINUE • • • • • • • • • • • • • • • • • • •
Syntax: Operands: Operation: Status Affected:	[<i>label</i>] DECF f,d $0 \le f \le 127$ $d \in [0,1]$ (f) - 1 \rightarrow (destination) Z 00 0011 dfff fff Decrement register 'f'. If 'd' is 0 the result is stored in the W register. If 'd' is	Example	GOTO LOOP CONTINUE • • • Before Instruction PC = address HERE After Instruction CNT = CNT - 1 if CNT = 0, PC = address CONTINUE
Syntax: Operands: Operation: Status Affected: Encoding: Description:	[<i>label</i>] DECF f,d $0 \le f \le 127$ $d \in [0,1]$ (f) - 1 \rightarrow (destination) Z 00 0011 dfff ffff Decrement register 'f' If 'd' is 0 the	Example	$\begin{array}{rcl} & GOTO & LOOP \\ \hline CONTINUE & & \\ & & \\ & & \\ & & \\ \end{array}$ Before Instruction $\begin{array}{rcl} PC & = & address \ here \\ After Instruction \\ CNT & = & CNT \ -1 \\ & if \ CNT & = & 0, \\ PC & = & address \ CONTINUE \\ & if \ CNT \neq & 0, \end{array}$
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words:	$\begin{matrix} [label] & DECF \ f,d \\ 0 \leq f \leq 127 \\ d \in [0,1] \\ (f) - 1 \rightarrow (destination) \\ Z \\ \hline \hline 00 & 0011 & dfff & ffff \\ \hline Decrement \ register \ 'f'. \ If \ 'd' \ is \ 0 \ the \\ result \ is \ stored \ in \ the \ W \ register. \ If \ 'd' \ is \\ 1 \ the \ result \ is \ stored \ back \ in \ register \ 'f'. \\ \end{matrix}$	Example	GOTO LOOP CONTINUE • • • Before Instruction PC = address HERE After Instruction CNT = CNT - 1 if CNT = 0, PC = address CONTINUE
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles:	[<i>label</i>] DECF f,d $0 \le f \le 127$ $d \in [0,1]$ (f) - 1 \rightarrow (destination) Z Decrement register 'f'. If 'd' is 0 the result is stored in the W register. If 'd' is 1 the result is stored back in register 'f'. 1	Example	$\begin{array}{rcl} & GOTO & LOOP \\ \hline CONTINUE & & \\ & & \\ & & \\ & & \\ \end{array}$ Before Instruction $\begin{array}{rcl} PC & = & address \ here \\ After Instruction \\ CNT & = & CNT \ -1 \\ & if \ CNT & = & 0, \\ PC & = & address \ CONTINUE \\ & if \ CNT \neq & 0, \end{array}$
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words:	[<i>label</i>] DECF f,d $0 \le f \le 127$ $d \in [0,1]$ (f) - 1 \rightarrow (destination) Z 00 0011 dfff ffff Decrement register 'f'. If 'd' is 0 the result is stored in the W register. If 'd' is 1 the result is stored back in register 'f'.	Example	$\begin{array}{rcl} & GOTO & LOOP \\ \hline CONTINUE & & \\ & & \\ & & \\ & & \\ \end{array}$ Before Instruction $\begin{array}{rcl} PC & = & address \ here \\ After Instruction \\ CNT & = & CNT \ -1 \\ & if \ CNT & = & 0, \\ PC & = & address \ CONTINUE \\ & if \ CNT \neq & 0, \end{array}$
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles:	$[label] DECF f,d$ $0 \le f \le 127$ $d \in [0,1]$ (f) - 1 \rightarrow (destination) Z $00 0011 dfff ffff$ Decrement register 'f'. If 'd' is 0 the result is stored in the W register. If 'd' is 1 the result is stored back in register 'f'. 1 1 2 2 2 2 2 2 2 2 2 2 3 2 4 2 2 2 2 3 2 4 2 2 2 3 2 4 2 2	Example	$\begin{array}{rcl} & GOTO & LOOP \\ \hline CONTINUE & & \\ & & \\ & & \\ & & \\ \end{array}$ Before Instruction $\begin{array}{rcl} PC & = & address \ here \\ After Instruction \\ CNT & = & CNT \ -1 \\ & if \ CNT & = & 0, \\ PC & = & address \ CONTINUE \\ & if \ CNT \neq & 0, \end{array}$
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity:	[<i>label</i>] DECF f,d 0 ≤ f ≤ 127 d ∈ [0,1] (f) - 1 → (destination) Z 00 0011 dfff ffff Decrement register 'f'. If 'd' is 0 the result is stored in the W register. If 'd' is 1 the result is stored back in register 'f'. 1 1 Q1 Q2 Q3 Q4 Decode Read Process Write to data destination DECF CNT, 1 Before Instruction	Example	$\begin{array}{rcl} & GOTO & LOOP \\ \hline CONTINUE & & \\ & & \\ & & \\ & & \\ \end{array}$ Before Instruction $\begin{array}{rcl} PC & = & address \ here \\ After Instruction \\ CNT & = & CNT \ -1 \\ & if \ CNT & = & 0, \\ PC & = & address \ CONTINUE \\ & if \ CNT \neq & 0, \end{array}$
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity:	[<i>label</i>] DECF f,d $0 \le f \le 127$ $d \in [0,1]$ (f) - 1 \rightarrow (destination) Z $\boxed{00 0011 dfff ffff}$ Decrement register 'f'. If 'd' is 0 the result is stored in the W register. If 'd' is 1 the result is stored back in register 'f'. 1 1 $\boxed{Q1 Q2 Q3 Q4}$ $\boxed{Decode Read Process Write to \ destination \ 'f'}$ DECF CNT, 1	Example	$\begin{array}{rcl} & GOTO & LOOP \\ \hline CONTINUE & & \\ & & \\ & & \\ & & \\ \end{array}$ Before Instruction $\begin{array}{rcl} PC & = & address \ here \\ After Instruction \\ CNT & = & CNT \ -1 \\ & if \ CNT & = & 0, \\ PC & = & address \ CONTINUE \\ & if \ CNT \neq & 0, \end{array}$
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity:	[<i>label</i>] DECF f,d 0 ≤ f ≤ 127 d ∈ [0,1] (f) - 1 → (destination) Z 00 0011 dfff ffff Decrement register 'f'. If 'd' is 0 the result is stored in the W register. If 'd' is 1 the result is stored back in register 'f'. 1 1 2 Q1 Q2 Q3 Q4 Decode Read Process Write to data destination DECF CNT, 1 Before Instruction CNT = 0x01 Z = 0 After Instruction	Example	$\begin{array}{rcl} & GOTO & LOOP \\ \hline CONTINUE & & \\ & & \\ & & \\ & & \\ \end{array}$ Before Instruction $\begin{array}{rcl} PC & = & address \ here \\ After Instruction \\ CNT & = & CNT \ -1 \\ & if \ CNT & = & 0, \\ PC & = & address \ CONTINUE \\ & if \ CNT \neq & 0, \end{array}$
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity:	[<i>label</i>] DECF f,d 0 ≤ f ≤ 127 d ∈ [0,1] (f) - 1 → (destination) Z 00 0011 dfff ffff Decrement register 'f'. If 'd' is 0 the result is stored in the W register. If 'd' is 1 the result is stored back in register 'f'. 1 1 Q1 Q2 Q3 Q4 Decode Read Process Write to data destination DECF CNT, 1 Before Instruction CNT = 0x01 Z = 0	Example	$\begin{array}{rcl} & GOTO & LOOP \\ \hline CONTINUE & & \\ & & \\ & & \\ & & \\ \end{array}$ Before Instruction $\begin{array}{rcl} PC & = & address \ here \\ After Instruction \\ CNT & = & CNT \ -1 \\ & if \ CNT & = & 0, \\ PC & = & address \ CONTINUE \\ & if \ CNT \neq & 0, \end{array}$

PIC16C7X

Applicab	e Devices 72 73 73A 74 74A 76	6 77								
17.3 DC	C Characteristics: PIC16C72-0 PIC16C72-1 PIC16C72-2 PIC16C72-2 PIC16LC72-	0 (Co 0 (Co	mmercia mmercia	ul, In ul, In	dustrial dustrial	, Exte , Exte	nded)			
DC CHARA	ACTERISTICS	Standard Operating Conditions (unless otherwise stated)Operating temperature -40° C \leq TA \leq +125°C for extended, -40° C \leq TA \leq +85°C for industrial and 0° C \leq TA \leq +70°C for commercialOperating voltage VDD range as described in DC spec Section 17.1								
Param	Characteristic	and Se Sym	ction 17.2 Min	Тур	Max	Units	Conditions			
No.	Characteristic	Sym		1 1	Ινίαλ	Units	Conditions			
D030 D030A	Input Low Voltage I/O ports with TTL buffer	VIL	Vss Vss	-	0.15Vdd 0.8V	V V	For entire VDD range $4.5 \le VDD \le 5.5V$			
D031 D032 D033	with Schmitt Trigger buffer MCLR, OSC1 (in RC mode) OSC1 (in XT, HS and LP)		Vss Vss Vss		0.2Vdd 0.2Vdd 0.3Vdd	V V V	Note1			
D040 D040A	Input High Voltage I/O ports with TTL buffer	Vih	2.0 0.25Vdd + 0.8V	- -	Vdd Vdd		$4.5 \le VDD \le 5.5V$ For entire VDD range			
D041 D042 D042A D043	with Schmitt Trigger buffer MCLR OSC1 (XT, HS and LP) OSC1 (in RC mode)		0.8VDD 0.8VDD 0.7VDD 0.9VDD		Vdd Vdd Vdd Vdd	V	For entire VDD range Note1			
D070	PORTB weak pull-up current	IPURB	50	250		μA	VDD = 5V, VPIN = VSS			
D060	Input Leakage Current (Notes 2, 3) I/O ports	lı∟	-	-	±1		$Vss \le VPIN \le VDD$, Pin at hi-impedance			
D061 D063	MCLR, RA4/T0CKI OSC1		-	-	±5 ±5	μA	Vss \leq VPIN \leq VDD Vss \leq VPIN \leq VDD, XT, HS and LP osc configuration			
	Output Low Voltage									
D080 D080A	I/O ports	Vol	-	-	0.6 0.6		IOL = 8.5 mA, VDD = 4.5V, -40°C to +85°C IOL = 7.0 mA, VDD = 4.5V,			
D083	OSC2/CLKOUT (RC osc config)		-	-	0.6	V	-40°C to +125°C loL = 1.6 mA, VDD = 4.5V, -40°C to +85°C			
D083A	ese parameters are characterized but		-	-	0.6	V	IOL = 1.2 mA, VDD = 4.5V, -40°C to +125°C			

These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: In RC oscillator configuration, the OSC1/CLKIN pin is a Schmitt trigger input. It is not recommended that the PIC16C7X be driven with external clock in RC mode.

2: The leakage current on the MCLR/VPP pin is strongly dependent on the applied voltage level. The specified levels represent normal operating conditions. Higher leakage current may be measured at different input voltages.

3: Negative current is defined as current sourced by the pin.

Applicable Devices 72 73 73A 74 74A 76 77

17.4 **Timing Parameter Symbology**

The timing parameter symbols have been created following one of the following formats:

1. TppS2p	pS	3. TCC:ST	(I ² C specifications only)
2. TppS		4. Ts	(I ² C specifications only)
т			
F	Frequency	Т	Time
Lowerca	ase letters (pp) and their meanings:		
рр			
CC	CCP1	OSC	OSC1
ck	CLKOUT	rd	RD
CS	CS	rw	\overline{RD} or \overline{WR}
di	SDI	SC	SCK
do	SDO	SS	SS
dt	Data in	tO	TOCKI
io	I/O port	t1	T1CKI
mc	MCLR	wr	WR
Upperca	ase letters and their meanings:		
S			
F	Fall	P	Period
Н	High	R	Rise
I	Invalid (Hi-impedance)	V	Valid
L	Low	Z	Hi-impedance
I ² C only			
AA	output access	High	High
BUF	Bus free	Low	Low
Tcc:st ((I ² C specifications only)	I	
CC			
HD	Hold	SU	Setup
ST			Comp
DAT	DATA input hold	STO	STOP condition
STA	START condition		
FIGURE 1	17-1: LOAD CONDITIONS		
	Load condition 1		Load condition 2
	N = = /0		
	J		
	\leq RL		
	$ \leq $	N	
	I → I		X
		F	
	• • • • • • • • • • • • • • • • • • • •	,	··· ↓
	Vss		Vss
	RL = 464 Ω		
	$C_L = 50 \text{ pF}$ for all pins ex	cept 0502	

15 pF for OSC2 output

Applic	able Devices 72 73 73A 74	74A 76	6 77				
18.3	PIC16 PIC16	6C73/7	4-04 (Co 4-10 (Co 4-20 (Co 74-04 (Co	omr omr	nercial, l nercial, l	Indust Indust	rial) rial)
							less otherwise stated)
DC CH4	ARACTERISTICS	•			0°C	; ≤	$TA \le +85^{\circ}C$ for industrial and $TA \le +70^{\circ}C$ for commercial ibed in DC spec Section 18.1 and
Param No.	Characteristic	Sym	Min	Тур	Max	Units	Conditions
INO.				†			
	Input Low Voltage I/O ports	VIL					
D030	with TTL buffer	VIL	Vss	-	0.15VDD	v	For entire VDD range
D030A			VSS	_	0.10VDD	v	$4.5V \le VDD \le 5.5V$
D031	with Schmitt Trigger buffer		VSS	-	0.2VDD	v	
D032	MCLR, OSC1 (in RC mode)		Vss	-	0.2VDD	v	
D033	OSC1 (in XT, HS and LP)		Vss	-	0.3VDD	V	Note1
	Input High Voltage						
	I/O ports	Vih		-			
D040	with TTL buffer		2.0	-	Vdd	V	$4.5V \le VDD \le 5.5V$
D040A			0.25VDD + 0.8V	-	Vdd	V	For entire VDD range
D041	with Schmitt Trigger buffer		0.8Vdd	-	Vdd	v	For entire VDD range
D041	MCLR		0.8VDD	_	VDD	v	Tor entire VDD range
D042A	OSC1 (XT, HS and LP)		0.7VDD	-	VDD	v	Note1
D043	OSC1 (in RC mode)		0.9VDD	-	VDD	v	
D070	PORTB weak pull-up current	IPURB	50	250		μA	VDD = 5V, VPIN = VSS
	Input Leakage Current (Notes 2, 3)					Pr. 1	
D060	I/O ports	lıL	-	-	±1	μA	Vss \leq VPIN \leq VDD, Pin at hi-impedance
D061	MCLR, RA4/T0CKI		-	-	±5	μA	$Vss \le VPIN \le VDD$
D063	OSC1		-	-	±5	μA	Vss \leq VPIN \leq VDD, XT, HS and LP osc configuration
	Output Low Voltage	1					
D080	I/O ports	Vol	-	-	0.6	V	IOL = 8.5 mA, VDD = 4.5V, -40°C to +85°C
D083	OSC2/CLKOUT (RC osc config)		-	-	0.6	V	lOL = 1.6 mA, VDD = 4.5V, -40°C to +85°C
	Output High Voltage						
D090	I/O ports (Note 3)	Voh	Vdd - 0.7	-	-	V	IOH = -3.0 mA, VDD = 4.5V, -40°С to +85°С
D092	OSC2/CLKOUT (RC osc config)		Vdd - 0.7	-	-	V	IOH = -1.3 mA, VDD = 4.5V, -40°С to +85°С
D150*	Open-Drain High Voltage	VOD	-	-	14	V	RA4 pin
L		L	L	I	I	I	· ·

These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: In RC oscillator configuration, the OSC1/CLKIN pin is a Schmitt Trigger input. It is not recommended that the PIC16C7X be driven with external clock in RC mode.

2: The leakage current on the MCLR pin is strongly dependent on the applied voltage level. The specified levels represent normal operating conditions. Higher leakage current may be measured at different input voltages.

3: Negative current is defined as current sourced by the pin.

PIC16C7X

Applicable Devices 72 73 73A 74 74A 76 77

19.0 ELECTRICAL CHARACTERISTICS FOR PIC16C73A/74A

Absolute Maximum Ratings †

-	
Ambient temperature under bias	55 to +125°C
Storage temperature	65°C to +150°C
Voltage on any pin with respect to Vss (except VDD, MCLR. and RA4)	0.3V to (VDD + 0.3V)
Voltage on VDD with respect to Vss	0.3 to +7.5V
Voltage on MCLR with respect to Vss (Note 2)	0 to +14V
Voltage on RA4 with respect to Vss	0 to +14V
Total power dissipation (Note 1)	1.0W
Maximum current out of Vss pin	300 mA
Maximum current into VDD pin	250 mA
Input clamp current, Iк (VI < 0 or VI > VDD)	±20 mA
Output clamp current, loк (Vo < 0 or Vo > VDD)	
Maximum output current sunk by any I/O pin	25 mA
Maximum output current sourced by any I/O pin	25 mA
Maximum current sunk by PORTA, PORTB, and PORTE (combined) (Note 3)	200 mA
Maximum current sourced by PORTA, PORTB, and PORTE (combined) (Note 3)	200 mA
Maximum current sunk by PORTC and PORTD (combined) (Note 3)	200 mA
Maximum current sourced by PORTC and PORTD (combined) (Note 3)	200 mA
Note 1: Power dissipation is calculated as follows: Pdis = VDD x {IDD - \sum IOH} + \sum {(VDD -	VOH) x IOH} + Σ (VOI x IOL)

Note 2: Voltage spikes below Vss at the MCLR pin, inducing currents greater than 80 mA, may cause latch-up. Thus, a series resistor of 50-100Ω should be used when applying a "low" level to the MCLR pin rather than pulling this pin directly to Vss.

Note 3: PORTD and PORTE are not implemented on the PIC16C73A.

† NOTICE: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operation listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

TABLE 19-1: CROSS REFERENCE OF DEVICE SPECS FOR OSCILLATOR CONFIGURATIONS AND FREQUENCIES OF OPERATION (COMMERCIAL DEVICES)

osc	PIC16C73A-04 PIC16C74A-04	PIC16C73A-10 PIC16C74A-10	PIC16C73A-20 PIC16C74A-20	PIC16LC73A-04 PIC16LC74A-04	JW Devices
RC	VDD: 4.0V to 6.0V IDD: 5 mA max. at 5.5V IPD: 16 μA max. at 4V Freq: 4 MHz max.	VDD: 4.5V to 5.5V IDD: 2.7 mA typ. at 5.5V IPD: 1.5 μA typ. at 4V Freq: 4 MHz max.	VDD: 4.5V to 5.5V IDD: 2.7 mA typ. at 5.5V IPD: 1.5 μA typ. at 4V Freq: 4 MHz max.	VDD: 2.5V to 6.0V IDD: 3.8 mA max. at 3.0V IPD: 5 μA max. at 3V Freq: 4 MHz max.	VDD: 4.0V to 6.0V IDD: 5 mA max. at 5.5V IPD: 16 μA max. at 4V Freq: 4 MHz max.
хт	VDD: 4.0V to 6.0V IDD: 5 mA max. at 5.5V IPD: 16 μA max. at 4V Freq: 4 MHz max.	VDD: 4.5V to 5.5V IDD: 2.7 mA typ. at 5.5V IPD: 1.5 μA typ. at 4V Freq: 4 MHz max.	VDD: 4.5V to 5.5V IDD: 2.7 mA typ. at 5.5V IPD: 1.5 μA typ. at 4V Freq: 4 MHz max.	VDD: 2.5V to 6.0V IDD: 3.8 mA max. at 3.0V IPD: 5 μA max. at 3V Freq: 4 MHz max.	VDD: 4.0V to 6.0V IDD: 5 mA max. at 5.5V IPD: 16 μA max. at 4V Freq: 4 MHz max.
HS	VDD: 4.5V to 5.5V IDD: 13.5 mA typ. at 5.5V IPD: 1.5 μA typ. at 4.5V Freq: 4 MHz max.	VDD: 4.5V to 5.5V IDD: 10 mA max. at 5.5V IPD: 1.5 μA typ. at 4.5V Freq: 10 MHz max.	VDD: 4.5V to 5.5V IDD: 20 mA max. at 5.5V IPD: 1.5 μA typ. at 4.5V Freq: 20 MHz max.	Not recommended for use in HS mode	VDD: 4.5V to 5.5V IDD: 20 mA max. at 5.5V IPD: 1.5 μA typ. at 4.5V Freq: 20 MHz max.
LP	VDD: 4.0V to 6.0V IDD: 52.5 μA typ. at 32 kHz, 4.0V IPD: 0.9 μA typ. at 4.0V Freq: 200 kHz max.	Not recommended for use in LP mode	Not recommended for use in LP mode	VDD: 2.5V to 6.0V IDD: 48 μA max. at 32 kHz, 3.0V IPD: 5.0 μA max. at 3.0V Freq: 200 kHz max.	VDD: 2.5V to 6.0V IDD: 48 μA max. at 32 kHz, 3.0V IPD: 5.0 μA max. at 3.0V Freq: 200 kHz max.

The shaded sections indicate oscillator selections which are tested for functionality, but not for MIN/MAX specifications. It is recommended that the user select the device type that ensures the specifications required.

Applicable Devices 72 73 73A 74 74A 76 77

19.1 DC Characteristics: PIC16C73A/74A-04 (Commercial, Industrial, Extended) PIC16C73A/74A-10 (Commercial, Industrial, Extended) PIC16C73A/74A-20 (Commercial, Industrial, Extended)

DC CHARACTERISTICS				Standard Operating Conditions (unless otherwise stated)Operating temperature -40° C $\leq TA \leq +125^{\circ}$ C for extended, -40° C $\leq TA \leq +85^{\circ}$ C for industrial and 0° C $\leq TA \leq +70^{\circ}$ C for commercial				
Param No.	Characteristic	Sym	Min	Тур†	Max	Units	Conditions	
D001 D001A	Supply Voltage	Vdd	4.0 4.5	-	6.0 5.5	V V	XT, RC and LP osc configuration HS osc configuration	
D002*	RAM Data Retention Voltage (Note 1)	Vdr	-	1.5	-	V		
D003	VDD start voltage to ensure internal Power-on Reset signal	VPOR	-	Vss	-	V	See section on Power-on Reset for details	
D004*	VDD rise rate to ensure internal Power-on Reset signal	Svdd	0.05	-	-	V/ms	See section on Power-on Reset for details	
D005	Brown-out Reset Voltage	Bvdd	3.7	4.0	4.3	V	BODEN bit in configuration word enabled	
			3.7	4.0	4.4	V	Extended Range Only	
D010	Supply Current (Note 2,5)	IDD	-	2.7	5	mA	XT, RC osc configuration Fosc = 4 MHz, VDD = 5.5V (Note 4)	
D013			-	10	20	mA	HS osc configuration Fosc = 20 MHz, VDD = 5.5V	
D015*	Brown-out Reset Current (Note 6)	Δ Ibor	-	350	425	μA	BOR enabled VDD = 5.0V	
D020 D021 D021A D021B	Power-down Current (Note 3,5)	IPD	- - - -	10.5 1.5 1.5 2.5	42 16 19 19	μΑ μΑ μΑ μΑ	$\label{eq:VDD} \begin{array}{l} VDD = 4.0V, WDT \text{ enabled}, -40^\circC \text{ to } +85^\circC \\ VDD = 4.0V, WDT \text{ disabled}, -0^\circC \text{ to } +70^\circC \\ VDD = 4.0V, WDT \text{ disabled}, -40^\circC \text{ to } +85^\circC \\ VDD = 4.0V, WDT \text{ disabled}, -40^\circC \text{ to } +125^\circC \end{array}$	
D023*	Brown-out Reset Current (Note 6)	Δ Ibor	-	350	425	μA	BOR enabled VDD = 5.0V	

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: This is the limit to which VDD can be lowered without losing RAM data.

2: The supply current is mainly a function of the operating voltage and frequency. Other factors such as I/O pin loading and switching rate, oscillator type, internal code execution pattern, and temperature also have an impact on the current consumption.

The test conditions for all IDD measurements in active operation mode are:

OSC1 = external square wave, from rail to rail; all I/O pins tristated, pulled to VDD

 $\overline{\text{MCLR}}$ = VDD; WDT enabled/disabled as specified.

3: The power-down current in SLEEP mode does not depend on the oscillator type. Power-down current is measured with the part in SLEEP mode, with all I/O pins in hi-impedance state and tied to VDD and Vss.

4: For RC osc configuration, current through Rext is not included. The current through the resistor can be estimated by the formula Ir = VDD/2Rext (mA) with Rext in kOhm.

5: Timer1 oscillator (when enabled) adds approximately 20 µA to the specification. This value is from characterization and is for design guidance only. This is not tested.

6: The ∆ current is the additional current consumed when this peripheral is enabled. This current should be added to the base IDD or IPD measurement.

Applicable Devices 72 73 73A 74 74A 76 77

20.2 DC Characteristics: PIC16LC76/77-04 (Commercial, Industrial)

DC CHARACTERISTICSStandard Operating Conditions (unless otherwise stated) Operating temperature $-40^{\circ}C$ $\leq TA \leq +85^{\circ}C$ for industrial and $0^{\circ}C$ $\leq TA \leq +70^{\circ}C$ for commercial						$0^{\circ}C \leq TA \leq +85^{\circ}C$ for industrial and	
Param No.	Characteristic	Sym	Min	Тур†	Max	Units	Conditions
D001	Supply Voltage	Vdd	2.5	-	6.0	V	LP, XT, RC osc configuration (DC - 4 MHz)
D002*	RAM Data Retention Voltage (Note 1)	Vdr	-	1.5	-	V	
D003	VDD start voltage to ensure internal Power-on Reset signal	Vpor	-	Vss	-	V	See section on Power-on Reset for details
D004*	VDD rise rate to ensure internal Power-on Reset signal	Svdd	0.05	-	-	V/ms	See section on Power-on Reset for details
D005	Brown-out Reset Voltage	Bvdd	3.7	4.0	4.3	V	BODEN bit in configuration word enabled
D010	Supply Current (Note 2,5)	IDD	-	2.0	3.8	mA	XT, RC osc configuration Fosc = 4 MHz, VDD = 3.0V (Note 4)
D010A			-	22.5	48	μA	LP osc configuration Fosc = 32 kHz, VDD = 3.0V, WDT disabled
D015*	Brown-out Reset Current (Note 6)	Δ IBOR	-	350	425	μA	BOR enabled VDD = 5.0V
D020 D021 D021A	Power-down Current (Note 3,5)	IPD	- - -	7.5 0.9 0.9	30 5 5	μΑ μΑ μΑ	VDD = $3.0V$, WDT enabled, $-40^{\circ}C$ to $+85^{\circ}C$ VDD = $3.0V$, WDT disabled, $0^{\circ}C$ to $+70^{\circ}C$ VDD = $3.0V$, WDT disabled, $-40^{\circ}C$ to $+85^{\circ}C$
D023*	Brown-out Reset Current (Note 6)	Δ IBOR	-	350	425	μΑ	BOR enabled VDD = 5.0V

These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: This is the limit to which VDD can be lowered without losing RAM data.

2: The supply current is mainly a function of the operating voltage and frequency. Other factors such as I/O pin loading and switching rate, oscillator type, internal code execution pattern, and temperature also have an impact on the current consumption.

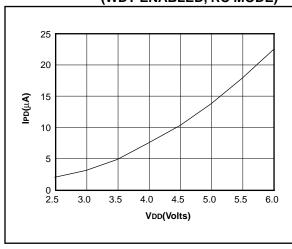
The test conditions for all IDD measurements in active operation mode are:

OSC1 = external square wave, from rail to rail; all I/O pins tristated, pulled to VDD \overline{MCLR} = VDD; WDT enabled/disabled as specified.

- 3: The power-down current in SLEEP mode does not depend on the oscillator type. Power-down current is measured with the part in SLEEP mode, with all I/O pins in hi-impedance state and tied to VDD and VSs.
- 4: For RC osc configuration, current through Rext is not included. The current through the resistor can be estimated by the formula Ir = VDD/2Rext (mA) with Rext in kOhm.
- 5: Timer1 oscillator (when enabled) adds approximately 20 µA to the specification. This value is from characterization and is for design guidance only. This is not tested.
- 6: The ∆ current is the additional current consumed when this peripheral is enabled. This current should be added to the base IDD or IPD measurement.

 Applicable Devices
 72
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 74A
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FIGURE 21-3: TYPICAL IPD vs. VDD @ 25°C (WDT ENABLED, RC MODE)





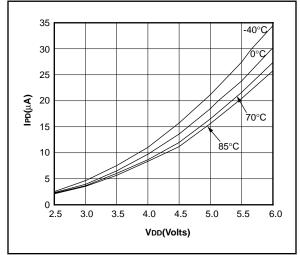
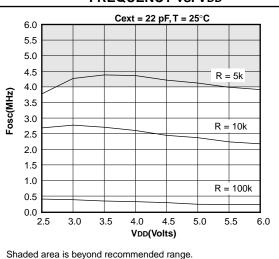


FIGURE 21-5: TYPICAL RC OSCILLATOR FREQUENCY vs. VDD





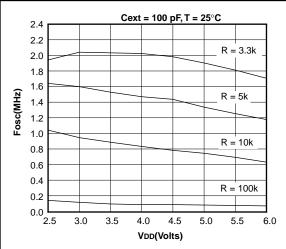


FIGURE 21-7: TYPICAL RC OSCILLATOR FREQUENCY vs. VDD

