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Details

Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	4MHz
Connectivity	I ² C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	33
Program Memory Size	14KB (8K x 14)
Program Memory Type	OTP
EEPROM Size	-
RAM Size	368 x 8
Voltage - Supply (Vcc/Vdd)	2.5V ~ 6V
Data Converters	A/D 8x8b
Oscillator Type	External
Operating Temperature	0°C ~ 70°C (TA)
Mounting Type	Surface Mount
Package / Case	44-TQFP
Supplier Device Package	44-TQFP (10x10)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic16lc77-04-pt

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For register and module descriptions in this data sheet, device legends show which devices apply to those sections. As an example, the legend below would mean that the following section applies only to the PIC16C72, PIC16C73A and PIC16C74A devices.

Applicable Devices 72 73 73A 74 74A 76 77

12|13|13A|14|14A|16|11

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3.0 ARCHITECTURAL OVERVIEW

The high performance of the PIC16CXX family can be attributed to a number of architectural features commonly found in RISC microprocessors. To begin with, the PIC16CXX uses a Harvard architecture, in which, program and data are accessed from separate memories using separate buses. This improves bandwidth over traditional von Neumann architecture in which program and data are fetched from the same memory using the same bus. Separating program and data buses further allows instructions to be sized differently than the 8-bit wide data word. Instruction opcodes are 14-bits wide making it possible to have all single word instructions. A 14-bit wide program memory access bus fetches a 14-bit instruction in a single cycle. A twostage pipeline overlaps fetch and execution of instructions (Example 3-1). Consequently, all instructions (35) execute in a single cycle (200 ns @ 20 MHz) except for program branches.

The table below lists program memory (EPROM) and data memory (RAM) for each PIC16C7X device.

Device	Program Memory	Data Memory			
PIC16C72	2K x 14	128 x 8			
PIC16C73	4K x 14	192 x 8			
PIC16C73A	4K x 14	192 x 8			
PIC16C74	4K x 14	192 x 8			
PIC16C74A	4K x 14	192 x 8			
PIC16C76	8K x 14	368 x 8			
PIC16C77	8K x 14	386 x 8			

The PIC16CXX can directly or indirectly address its register files or data memory. All special function registers, including the program counter, are mapped in the data memory. The PIC16CXX has an orthogonal (symmetrical) instruction set that makes it possible to carry out any operation on any register using any addressing mode. This symmetrical nature and lack of 'special optimal situations' make programming with the PIC16CXX simple yet efficient. In addition, the learning curve is reduced significantly.

PIC16CXX devices contain an 8-bit ALU and working register. The ALU is a general purpose arithmetic unit. It performs arithmetic and Boolean functions between the data in the working register and any register file.

The ALU is 8-bits wide and capable of addition, subtraction, shift and logical operations. Unless otherwise mentioned, arithmetic operations are two's complement in nature. In two-operand instructions, typically one operand is the working register (W register). The other operand is a file register or an immediate constant. In single operand instructions, the operand is either the W register or a file register.

The W register is an 8-bit working register used for ALU operations. It is not an addressable register.

Depending on the instruction executed, the ALU may affect the values of the Carry (C), Digit Carry (DC), and Zero (Z) bits in the STATUS register. The C and DC bits operate as a borrow bit and a digit borrow out bit, respectively, in subtraction. See the SUBLW and SUBWF instructions for examples.





Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	Value on all other resets (3)
Bank 1	nk 1										
80h ⁽¹⁾	INDF	Addressing this location uses contents of FSR to address data memory (not a physical register)									0000 0000
81h	OPTION	RBPU	INTEDG	TOCS	TOSE	PSA	PS2	PS1	PS0	1111 1111	1111 1111
82h ⁽¹⁾	PCL	Program Co	ounter's (PC)	Least Signif	icant Byte					0000 0000	0000 0000
83h ⁽¹⁾	STATUS	IRP ⁽⁴⁾	RP1 ⁽⁴⁾	RP0	то	PD	Z	DC	С	0001 1xxx	000q quuu
84h ⁽¹⁾	FSR	Indirect data	a memory ac	Idress pointe	er					XXXX XXXX	uuuu uuuu
85h	TRISA	—	_	PORTA Dat	a Direction F	Register				11 1111	11 1111
86h	TRISB	PORTB Dat	ta Direction F	Register						1111 1111	1111 1111
87h	TRISC	PORTC Dat	ta Direction F	Register						1111 1111	1111 1111
88h	_	Unimpleme	nted							_	_
89h	_	Unimpleme	nted							_	—
8Ah ^(1,2)	PCLATH	—	_	_	Write Buffer	r for the uppe	er 5 bits of the	e PC		0 0000	0 0000
8Bh ⁽¹⁾	INTCON	GIE	PEIE	TOIE	INTE	RBIE	TOIF	INTF	RBIF	0000 000x	0000 000u
8Ch	PIE1	_	ADIE	_	_	SSPIE	CCP1IE	TMR2IE	TMR1IE	-0 0000	-0 0000
8Dh	_	Unimpleme	nted							_	_
8Eh	PCON	—	_	_	_	—	_	POR	BOR	dd	uu
8Fh	-	Unimpleme	nted							-	_
90h	_	Unimpleme	nted							_	_
91h	-	Unimpleme	nted							-	_
92h	PR2	Timer2 Peri	od Register							1111 1111	1111 1111
93h	SSPADD	Synchronou	us Serial Port	(I ² C mode)	Address Reg	gister				0000 0000	0000 0000
94h	SSPSTAT	—	-	D/Ā	Р	S	R/W	UA	BF	00 0000	00 0000
95h	—	Unimpleme	nted							—	—
96h	—	Unimpleme	nted							—	—
97h	-	Unimpleme	nted							-	_
98h	_	Unimpleme	nted							—	_
99h	—	Unimpleme	nted		—	—					
9Ah	-	Unimpleme	Jnimplemented								
9Bh	-	Unimpleme	Unimplemented —								
9Ch	_	Unimplemented									_
9Dh	_	Unimpleme	nted								—
9Eh	_	Unimpleme	nted							_	—
9Fh	ADCON1	-	-	_	—	—	PCFG2	PCFG1	PCFG0	000	000

TABLE 4-1: PIC16C72 SPECIAL FUNCTION REGISTER SUMMARY (Cont.'d)

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented read as '0'.

Shaded locations are unimplemented, read as '0'.

Note 1: These registers can be addressed from either bank.

2: The upper byte of the program counter is not directly accessible. PCLATH is a holding register for the PC<12:8> whose contents are transferred to the upper byte of the program counter.

3: Other (non power-up) resets include external reset through MCLR and Watchdog Timer Reset.

4: The IRP and RP1 bits are reserved on the PIC16C72, always maintain these bits clear.

TABLE 5-4:	SUMMARY OF REGISTERS	ASSOCIATED WITH PORTB

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	Value on all other resets
06h, 106h	PORTB	RB7	RB6	RB5	RB4	RB3	RB2	RB1	RB0	xxxx xxxx	uuuu uuuu
86h, 186h	TRISB	PORTB	Data Directio	on Regist	er					1111 1111	1111 1111
81h, 181h	OPTION	RBPU	INTEDG	TOCS	TOSE	PSA	PS2	PS1	PS0	1111 1111	1111 1111

Legend: x = unknown, u = unchanged. Shaded cells are not used by PORTB.

5.5 PORTE and TRISE Register Applicable Devices 72/73/73A/74/74A/76/77

PORTE has three pins RE0/RD/AN5, RE1/WR/AN6 and RE2/CS/AN7, which are individually configurable as inputs or outputs. These pins have Schmitt Trigger input buffers.

I/O PORTE becomes control inputs for the microprocessor port when bit PSPMODE (TRISE<4>) is set. In this mode, the user must make sure that the TRISE<2:0> bits are set (pins are configured as digital inputs) and that register ADCON1 is configured for digital I/O. In this mode the input buffers are TTL.

Figure 5-9 shows the TRISE register, which also controls the parallel slave port operation.

PORTE pins are multiplexed with analog inputs. The operation of these pins is selected by control bits in the ADCON1 register. When selected as an analog input, these pins will read as '0's.

TRISE controls the direction of the RE pins, even when they are being used as analog inputs. The user must make sure to keep the pins configured as inputs when using them as analog inputs.



FIGURE 5-8: PORTE BLOCK DIAGRAM (IN I/O PORT MODE)



FIGURE 5-9: TRISE REGISTER (ADDRESS 89h)

R _0	R-0	P/\/_0	R/M/-0	11-0	P/\/_1	P/\/_1	₽/\\/_1			
IBF	OBF	IBOV	PSPMODE		bit2	bit1	bit0	R = Readable bit		
bit7					512	<u><u></u></u>	bit0	W = Writable bit U = Unimplemented bit, read as '0' - n = Value at POR reset		
bit 7 :	IBF: Input 1 = A word 0 = No wor	Buffer Full has been rd has beer	Status bit received and n received	is waiting t	o be read by	the CPU				
bit 6:	OBF : Outp 1 = The ou 0 = The ou	ut Buffer F tput buffer tput buffer	ull Status bit still holds a pi has been read	eviously w	ritten word					
bit 5:	 5: IBOV: Input Buffer Overflow Detect bit (in microprocessor mode) 1 = A write occurred when a previously input word has not been read (must be cleared in software) 0 = No overflow occurred 									
bit 4:	PSPMODE 1 = Paralle 0 = Genera	E: Parallel S I slave por al purpose	Slave Port Moo t mode I/O mode	de Select b	bit					
bit 3:	Unimplem	ented: Re	ad as '0'							
	PORTE D	Data Direc	ction Bits							
bit 2:	Bit2 : Direc 1 = Input 0 = Output	tion Contro	ol bit for pin RI	E2/CS/AN7	,					
bit 1:	Bit1 : Direc 1 = Input 0 = Output	tion Contro	ol bit for pin RI	E1/WR/AN	6					
bit 0:	Bit0 : Direc 1 = Input 0 = Output	tion Contro	ol bit for pin RI	E0/RD/AN5	5					

11.0 SYNCHRONOUS SERIAL PORT (SSP) MODULE

11.1 <u>SSP Module Overview</u>

The Synchronous Serial Port (SSP) module is a serial interface useful for communicating with other peripheral or microcontroller devices. These peripheral devices may be Serial EEPROMs, shift registers, display drivers, A/D converters, etc. The SSP module can operate in one of two modes:

- Serial Peripheral Interface (SPI)
- Inter-Integrated Circuit (I²C)

The SSP module in I^2C mode works the same in all PIC16C7X devices that have an SSP module. However the SSP Module in SPI mode has differences between the PIC16C76/77 and the other PIC16C7X devices.

The register definitions and operational description of SPI mode has been split into two sections because of the differences between the PIC16C76/77 and the other PIC16C7X devices. The default reset values of both the SPI modules is the same regardless of the device:

11.2	SPI Mode for PIC16C72/73/73A/74/74A78	В
11.3	SPI Mode for PIC16C76/7783	3
11.4	I2C [™] Overview89	Э
11.5	SSP I2C Operation	3

Refer to Application Note AN578, "Use of the SSP Module in the l^2C Multi-Master Environment."

To enable the serial port, SSP enable bit SSPEN (SSPCON<5>) must be set. To reset or reconfigure SPI mode, clear enable bit SSPEN, re-initialize SSPCON register, and then set enable bit SSPEN. This configures the SDI, SDO, SCK, and SS pins as serial port pins. For the pins to behave as the serial port function, they must have their data direction bits (in the TRIS register) appropriately programmed. That is:

- SDI must have TRISC<4> set
- SDO must have TRISC<5> cleared
- SCK (Master mode) must have TRISC<3> cleared
- SCK (Slave mode) must have TRISC<3> set
- <u>SS</u> must have TRISA<5> set (if implemented)

Any serial port function that is not desired may be overridden by programming the corresponding data direction (TRIS) register to the opposite value. An example would be in master mode where you are only sending data (to a display driver), then both SDI and SS could be used as general purpose outputs by clearing their corresponding TRIS register bits.

Figure 11-4 shows a typical connection between two microcontrollers. The master controller (Processor 1) initiates the data transfer by sending the SCK signal. Data is shifted out of both shift registers on their programmed clock edge, and latched on the opposite edge of the clock. Both processors should be programmed to the same Clock Polarity (CKP), then both controllers would send and receive data at the same time. Whether the data is meaningful (or dummy data) depends on the application software. This leads to three scenarios for data transmission:

- Master sends data Slave sends dummy data
- Master sends data Slave sends data
- · Master sends dummy data Slave sends data

FIGURE 11-4: SPI MASTER/SLAVE CONNECTION

The master can initiate the data transfer at any time because it controls the SCK. The master determines when the slave (Processor 2) is to broadcast data by the software protocol.

In master mode the data is transmitted/received as soon as the SSPBUF register is written to. If the SPI is only going to receive, the SCK output could be disabled (programmed as an input). The SSPSR register will continue to shift in the signal present on the SDI pin at the programmed clock rate. As each byte is received, it will be loaded into the SSPBUF register as if a normal received byte (interrupts and status bits appropriately set). This could be useful in receiver applications as a "line activity monitor" mode.

In slave mode, the data is transmitted and received as the external clock pulses appear on SCK. When the last bit is latched interrupt flag bit SSPIF (PIR1<3>) is set.

The clock polarity is selected by appropriately programming bit CKP (SSPCON<4>). This then would give waveforms for SPI communication as shown in Figure 11-5 and Figure 11-6 where the MSB is transmitted first. In master mode, the SPI clock rate (bit rate) is user programmable to be one of the following:

- Fosc/4 (or TCY)
- Fosc/16 (or 4 TCY)
- Fosc/64 (or 16 TCY)
- Timer2 output/2

This allows a maximum bit clock frequency (at 20 MHz) of 5 MHz. When in slave mode the external clock must meet the minimum high and low times.

In sleep mode, the slave can transmit and receive data and wake the device from sleep.



11.3.1 SPI MODE FOR PIC16C76/77

The SPI mode allows 8-bits of data to be synchronously transmitted and received simultaneously. To accomplish communication, typically three pins are used:

- Serial Data Out (SDO) RC5/SDO
- Serial Data In (SDI) RC4/SDI/SDA
- Serial Clock (SCK) RC3/SCK/SCL

Additionally a fourth pin may be used when in a slave mode of operation:

Slave Select (SS) RA5/SS/AN4

When initializing the SPI, several options need to be specified. This is done by programming the appropriate control bits in the SSPCON register (SSPCON<5:0>) and SSPSTAT<7:6>. These control bits allow the following to be specified:

- Master Mode (SCK is the clock output)
- Slave Mode (SCK is the clock input)
- Clock Polarity (Idle state of SCK)
- Clock edge (output data on rising/falling edge of SCK)
- Clock Rate (Master mode only)
- Slave Select Mode (Slave mode only)

The SSP consists of a transmit/receive Shift Register (SSPSR) and a buffer register (SSPBUF). The SSPSR shifts the data in and out of the device, MSb first. The SSPBUF holds the data that was written to the SSPSR until the received data is ready. Once the 8-bits of data have been received, that byte is moved to the SSPBUF register. Then the buffer full detect bit BF (SSPSTAT<0>) and interrupt flag bit SSPIF (PIR1<3>) are set. This double buffering of the received data (SSPBUF) allows the next byte to start reception before reading the data that was just received. Any write to the SSPBUF register during transmission/reception of data will be ignored, and the write collision detect bit WCOL (SSPCON<7>) will be set. User software must clear the WCOL bit so that it can be determined if the following write(s) to the SSPBUF register completed successfully. When the application software is expecting to receive valid data, the SSPBUF should be read before the next byte of data to transfer is written to the SSPBUF. Buffer full bit BF (SSPSTAT<0>) indicates when SSPBUF has been loaded with the received data (transmission is complete). When the SSPBUF is read, bit BF is cleared. This data may be irrelevant if the SPI is only a transmitter. Generally the SSP Interrupt is used to determine when the transmission/reception has completed. The SSPBUF must be read and/or written. If the interrupt method is not going to be used, then software polling can be done to ensure that a write collision does not occur. Example 11-2 shows the loading of the SSPBUF (SSPSR) for data transmission. The shaded instruction is only required if the received data is meaningful.

EXAMPLE 11-2: LOADING THE SSPBUF (SSPSR) REGISTER (PIC16C76/77)

	BCF BSF	STATUS , STATUS ,	RP1 RP0	;Specify Bank 1
LOOP	BTFSS	SSPSTAT	, BF	;Has data been ;received ;(transmit
				;complete)?
	GOTO	LOOP		;No
	BCF	STATUS,	RP0	;Specify Bank 0
	MOVF	SSPBUF,	W	;W reg = contents ; of SSPBUF
	MOVWF	RXDATA		;Save in user RAM
	MOVF	TXDATA,	W	;W reg = contents ; of TXDATA
	MOVWF	SSPBUF		;New data to xmit

The block diagram of the SSP module, when in SPI mode (Figure 11-9), shows that the SSPSR is not directly readable or writable, and can only be accessed from addressing the SSPBUF register. Additionally, the SSP status register (SSPSTAT) indicates the various status conditions.

FIGURE 11-9: SSP BLOCK DIAGRAM (SPI MODE)(PIC16C76/77)



FIGURE 11-13: SPI MODE TIMING (SLAVE MODE WITH CKE = 1) (PIC16C76/77)



Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value o POR, BOR	n:	Valu all o res	e on other sets
0Bh,8Bh. 10Bh,18Bh	INTCON	GIE	PEIE	TOIE	INTE	RBIE	TOIF	INTF	RBIF	0000 00	0x	0000	000u
0Ch	PIR1	PSPIF ⁽¹⁾	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 00	000	0000	0000
8Ch	PIE1	PSPIE ⁽¹⁾	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 00	000	0000	0000
87h	TRISC	PORTC Da	ta Directio	on Registe	er					1111 11	.11	1111	1111
13h	SSPBUF	Synchronou	us Serial F	Port Recei	ve Buff	er/Transm	it Registe	r		XXXX XX	xx	uuuu	uuuu
14h	SSPCON	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0	0000 00	000	0000	0000
85h	TRISA	—	_	PORTA I	PORTA Data Direction Register						.11	11	1111
94h	SSPSTAT	SMP	CKE	D/Ā	Р	S	R/W	UA	BF	0000 00	000	0000	0000

Legend: x = unknown, u = unchanged, - = unimplemented read as '0'. Shaded cells are not used by the SSP in SPI mode. Note 1: Bits PSPIE and PSPIF are reserved on the PIC16C76, always maintain these bits clear.

11.4.2 ADDRESSING I²C DEVICES

There are two address formats. The simplest is the 7-bit address format with a R/\overline{W} bit (Figure 11-15). The more complex is the 10-bit address with a R/\overline{W} bit (Figure 11-16). For 10-bit address format, two bytes must be transmitted with the first five bits specifying this to be a 10-bit address.

FIGURE 11-15: 7-BIT ADDRESS FORMAT



FIGURE 11-16: I²C 10-BIT ADDRESS FORMAT



11.4.3 TRANSFER ACKNOWLEDGE

All data must be transmitted per byte, with no limit to the number of bytes transmitted per data transfer. After each byte, the slave-receiver generates an acknowledge bit (\overline{ACK}) (Figure 11-17). When a slave-receiver doesn't acknowledge the slave address or received data, the master must abort the transfer. The slave must leave SDA high so that the master can generate the STOP condition (Figure 11-14).

FIGURE 11-17: SLAVE-RECEIVER ACKNOWLEDGE



If the master is receiving the data (master-receiver), it generates an acknowledge signal for each received byte of data, except for the last byte. To signal the end of data to the slave-transmitter, the master does not generate an acknowledge (not acknowledge). The slave then releases the SDA line so the master can generate the STOP condition. The master can also generate the STOP condition during the acknowledge pulse for valid termination of data transfer.

If the slave needs to delay the transmission of the next byte, holding the SCL line low will force the master into a wait state. Data transfer continues when the slave releases the SCL line. This allows the slave to move the received data or fetch the data it needs to transfer before allowing the clock to start. This wait state technique can also be implemented at the bit level, Figure 11-18. The slave will inherently stretch the clock, when it is a transmitter, but will not when it is a receiver. The slave will have to clear the SSPCON<4> bit to enable clock stretching when it is a receiver.



FIGURE 11-18: DATA TRANSFER WAIT STATE

11.4.4 MULTI-MASTER

The I^2C protocol allows a system to have more than one master. This is called multi-master. When two or more masters try to transfer data at the same time, arbitration and synchronization occur.

11.4.4.1 ARBITRATION

Arbitration takes place on the SDA line, while the SCL line is high. The master which transmits a high when the other master transmits a low loses arbitration (Figure 11-22), and turns off its data output stage. A master which lost arbitration can generate clock pulses until the end of the data byte where it lost arbitration. When the master devices are addressing the same device, arbitration continues into the data.

FIGURE 11-22: MULTI-MASTER ARBITRATION (TWO MASTERS)



Masters that also incorporate the slave function, and have lost arbitration must immediately switch over to slave-receiver mode. This is because the winning master-transmitter may be addressing it.

Arbitration is not allowed between:

- A repeated START condition
- A STOP condition and a data bit
- A repeated START condition and a STOP condition

Care needs to be taken to ensure that these conditions do not occur.

11.2.4.2 Clock Synchronization

Clock synchronization occurs after the devices have started arbitration. This is performed using a wired-AND connection to the SCL line. A high to low transition on the SCL line causes the concerned devices to start counting off their low period. Once a device clock has gone low, it will hold the SCL line low until its SCL high state is reached. The low to high transition of this clock may not change the state of the SCL line, if another device clock is still within its low period. The SCL line is held low by the device with the longest low period. Devices with shorter low periods enter a high wait-state, until the SCL line comes high. When the SCL line comes high, all devices start counting off their high periods. The first device to complete its high period will pull the SCL line low. The SCL line high time is determined by the device with the shortest high period, Figure 11-23.

FIGURE 11-23: CLOCK SYNCHRONIZATION



The ADRES register contains the result of the A/D conversion. When the A/D conversion is complete, the result is loaded into the ADRES register, the GO/DONE bit (ADCON0<2>) is cleared, and A/D interrupt flag bit ADIF is set. The block diagrams of the A/D module are shown in Figure 13-3.

After the A/D module has been configured as desired, the selected channel must be acquired before the conversion is started. The analog input channels must have their corresponding TRIS bits selected as an input. To determine acquisition time, see Section 13.1. After this acquisition time has elapsed the A/D conversion can be started. The following steps should be followed for doing an A/D conversion:

- 1. Configure the A/D module:
 - Configure analog pins / voltage reference / and digital I/O (ADCON1)
 - Select A/D input channel (ADCON0)
 - Select A/D conversion clock (ADCON0)
 - Turn on A/D module (ADCON0)
- 2. Configure A/D interrupt (if desired):
 - Clear ADIF bit
 - Set ADIE bit
 - Set GIE bit

FIGURE 13-3: A/D BLOCK DIAGRAM

- 3. Wait the required acquisition time.
- 4. Start conversion:Set GO/DONE bit (ADCON0)
- 5. Wait for A/D conversion to complete, by either:
 Polling for the GO/DONE bit to be cleared
 - OR
 - Waiting for the A/D interrupt
- Read A/D Result register (ADRES), clear bit ADIF if required.
- 7. For next conversion, go to step 1 or step 2 as required. The A/D conversion time per bit is defined as TAD. A minimum wait of 2TAD is required before next acquisition starts.



PIC16C7X

Register	Applicable Devices							Power-on Reset, Brown-out Reset	MCLR Resets WDT Reset	Wake-up via WDT or Interrupt
SSPADD	72	73	73A	74	74A	76	77	0000 0000	0000 0000	uuuu uuuu
SSPSTAT	72	73	73A	74	74A	76	77	00 0000	00 0000	uu uuuu
TXSTA	72	73	73A	74	74A	76	77	0000 -010	0000 -010	uuuu -uuu
SPBRG	72	73	73A	74	74A	76	77	0000 0000	0000 0000	uuuu uuuu
ADCON1	72	73	73A	74	74A	76	77	000	000	uuu

TABLE 14-8: INITIALIZATION CONDITIONS FOR ALL REGISTERS (Cont.'d)

Legend: u = unchanged, x = unknown, - = unimplemented bit, read as '0', q = value depends on condition Note 1: One or more bits in INTCON, PIR1 and/or PIR2 will be affected (to cause wake-up).

2: When the wake-up is due to an interrupt and the GIE bit is set, the PC is loaded with the interrupt vector (0004h).

3: See Table 14-7 for reset value for specific condition.

Applicable Devices 72 73 73A 74 74A 76 77

17.4 **Timing Parameter Symbology**

The timing parameter symbols have been created following one of the following formats:

1. TppS2pp	S	3. Tcc:st	(I ² C specifications only)					
2. TppS		4. Ts	(I ² C specifications only)					
Т								
F	Frequency	Т	Time					
Lowercas	e letters (pp) and their meanings:							
рр								
сс	CCP1	osc	OSC1					
ck	CLKOUT	rd	RD					
cs	CS	rw	RD or WR					
di	SDI	sc	SCK					
do	SDO	SS	SS					
dt	Data in	tO	TOCKI					
io	I/O port	t1	T1CKI					
mc	MCLR	wr	WR					
Uppercas	e letters and their meanings:							
S								
F	Fall	P	Period					
н	High	R	Rise					
1	Invalid (Hi-impedance)	V	Valid					
L	Low	z	Hi-impedance					
I ² C only								
AA	output access	High	High					
BUF	Bus free	Low	Low					
Tcc:st (l	C specifications only)	1						
CC								
HD	Hold	SU	Setup					
ST								
DAT	DATA input hold	STO	STOP condition					
STA	START condition							
FIGURE 17	7-1: LOAD CONDITIONS							
	Load condition 1		Load condition 2					
	VDD/2							
	Ĭ							
	\leq RL							
	<	_						
	↓							
	· ···· ↓		··· ↓					
	Vss		Vss					
	$R_{\rm L} = 464\Omega$							
	$\Omega = 50 \text{ pc} \text{for all pipe events}$	2802						
GL = 50 pF for all pins except OSC2								

15 pF for OSC2 output

PIC16C7X

Applicable Devices 72 73 73A 74 74A 76 77

19.0 ELECTRICAL CHARACTERISTICS FOR PIC16C73A/74A

Absolute Maximum Ratings †

Ambient temperature under bias	55 to +125°C
Storage temperature	65°C to +150°C
Voltage on any pin with respect to Vss (except VDD, MCLR. and RA4)	0.3V to (VDD + 0.3V)
Voltage on VDD with respect to VSS	0.3 to +7.5V
Voltage on MCLR with respect to Vss (Note 2)	0 to +14V
Voltage on RA4 with respect to Vss	0 to +14V
Total power dissipation (Note 1)	1.0W
Maximum current out of Vss pin	300 mA
Maximum current into VDD pin	250 mA
Input clamp current, IIK (VI < 0 or VI > VDD)	±20 mA
Output clamp current, IOK (VO < 0 or VO > VDD)	±20 mA
Maximum output current sunk by any I/O pin	25 mA
Maximum output current sourced by any I/O pin	25 mA
Maximum current sunk by PORTA, PORTB, and PORTE (combined) (Note 3)	200 mA
Maximum current sourced by PORTA, PORTB, and PORTE (combined) (Note 3)	200 mA
Maximum current sunk by PORTC and PORTD (combined) (Note 3)	200 mA
Maximum current sourced by PORTC and PORTD (combined) (Note 3)	200 mA
Note 1: Power dissipation is calculated as follows: Pdis = VDD x {IDD - \sum IOH} + \sum {(VDD - VOH)	H) x IOH} + Σ (VOI x IOL)

Note 2: Voltage spikes below Vss at the MCLR pin, inducing currents greater than 80 mA, may cause latch-up. Thus, a series resistor of 50-100Ω should be used when applying a "low" level to the MCLR pin rather than pulling this pin directly to Vss.

Note 3: PORTD and PORTE are not implemented on the PIC16C73A.

† NOTICE: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operation listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

TABLE 19-1: CROSS REFERENCE OF DEVICE SPECS FOR OSCILLATOR CONFIGURATIONS AND FREQUENCIES OF OPERATION (COMMERCIAL DEVICES)

osc	PIC16C73A-04 PIC16C74A-04	PIC16C73A-10 PIC16C73A-20 PIC16C74A-10 PIC16C74A-20		PIC16LC73A-04 PIC16LC74A-04	JW Devices	
RC	VDD: 4.0V to 6.0V IDD: 5 mA max. at 5.5V IPD: 16 μA max. at 4V Freq: 4 MHz max.	VDD: 4.5V to 5.5V IDD: 2.7 mA typ. at 5.5V IPD: 1.5 μA typ. at 4V Freq: 4 MHz max.	VDD: 4.5V to 5.5V IDD: 2.7 mA typ. at 5.5V IPD: 1.5 μA typ. at 4V Freq: 4 MHz max.	VDD: 2.5V to 6.0V IDD: 3.8 mA max. at 3.0V IPD: 5 μA max. at 3V Freq: 4 MHz max.	VDD: 4.0V to 6.0V IDD: 5 mA max. at 5.5V IPD: 16 μA max. at 4V Freq: 4 MHz max.	
хт	VDD: 4.0V to 6.0V IDD: 5 mA max. at 5.5V IPD: 16 μA max. at 4V Freq: 4 MHz max.	VDD: 4.5V to 5.5V IDD: 2.7 mA typ. at 5.5V IPD: 1.5 μA typ. at 4V Freq: 4 MHz max.	VDD: 4.5V to 5.5V IDD: 2.7 mA typ. at 5.5V IPD: 1.5 μA typ. at 4V Freq: 4 MHz max.	VDD: 2.5V to 6.0V IDD: 3.8 mA max. at 3.0V IPD: 5 μA max. at 3V Freq: 4 MHz max.	VDD: 4.0V to 6.0V IDD: 5 mA max. at 5.5V IPD: 16 μA max. at 4V Freq: 4 MHz max.	
	VDD: 4.5V to 5.5V	VDD: 4.5V to 5.5V	VDD: 4.5V to 5.5V		VDD: 4.5V to 5.5V	
HS	IDD: 13.5 mA typ. at 5.5V IPD: 1.5 μA typ. at 4.5V	IPD: 1.5 μA typ. at 4.5V	IPD: 1.5 μA typ. at 4.5V	Not recommended for use in HS mode	IDD: 20 mA max. at 5.5V IPD: 1.5 μA typ. at 4.5V	
	Freq: 4 MHz max.	Freq: 10 MHz max.	Freq: 20 MHz max.		Freq: 20 MHz max.	
LP	VDD: 4.0V to 6.0V IDD: 52.5 μA typ. at 32 kHz, 4.0V IPD: 0.9 μA typ. at 4.0V Freq: 200 kHz max.	Not recommended for use in LP mode	Not recommended for use in LP mode	VDD: 2.5V to 6.0V IDD: 48 μA max. at 32 kHz, 3.0V IPD: 5.0 μA max. at 3.0V Freq: 200 kHz max.	VDD: 2.5V to 6.0V IDD: 48 μA max. at 32 kHz, 3.0V IPD: 5.0 μA max. at 3.0V Freq: 200 kHz max.	

The shaded sections indicate oscillator selections which are tested for functionality, but not for MIN/MAX specifications. It is recommended that the user select the device type that ensures the specifications required.

Applicable Devices 72 73 73A 74 74A 76 77

FIGURE 19-6: TIMER0 AND TIMER1 EXTERNAL CLOCK TIMINGS



TABLE 19-5: TIMER0 AND TIMER1 EXTERNAL CLOCK REQUIREMENTS

Param	Sym	Characteristic			Min	Typ†	Мах	Units	Conditions
NO.									
40*	Tt0H	T0CKI High Pulse Width		No Prescaler	0.5TCY + 20	—		ns	Must also meet
				With Prescaler	10	—	—	ns	parameter 42
41*	Tt0L	T0CKI Low Pulse W	/idth	No Prescaler	0.5Tcy + 20	—	—	ns	Must also meet
				With Prescaler	10	—	—	ns	parameter 42
42*	Tt0P	T0CKI Period		No Prescaler	TCY + 40	—	—	ns	
				With Prescaler	Greater of:	-	—	ns	N = prescale value
					20 or <u>TCY + 40</u>				(2, 4,, 256)
			-		N				
45*	Tt1H	T1CKI High Time	Synchronous, P	Prescaler = 1	0.5TCY + 20	-	—	ns	Must also meet
			Synchronous,	PIC16 C 7X	15	—	—	ns	parameter 47
			Prescaler =	PIC16 LC 7X	25	-	-	ns	
			Asynchronous	PIC16 C 7X	30	_	_	ns	
				PIC16LC7X	50	_		ns	-
46*	Tt1L	T1CKI Low Time	Synchronous, F	Prescaler = 1	0.5TCY + 20	_	_	ns	Must also meet
			Synchronous,	PIC16 C 7X	15	_	_	ns	parameter 47
			Prescaler =	PIC16 LC 7X	25	—	—	ns	
			2,4,8						
			Asynchronous	PIC16 C 7X	30		—	ns	
				PIC16 LC 7X	50	—	—	ns	
47*	Tt1P	T1CKI input period	Synchronous	PIC16 C 7X	Greater of:			ns	N = prescale value
					30 OR <u>TCY + 40</u>				(1, 2, 4, 8)
					N				
				PIC16 LC 7X	Greater of:				N = prescale value
					50 OR <u>TCY + 40</u>				(1, 2, 4, 8)
					N				
			Asynchronous	PIC16 C 7X	60			ns	
				PIC16 LC 7X	100			ns	
	Ft1	Timer1 oscillator inp	out frequency rar	nge	DC	-	200	kHz	
		(oscillator enabled b	by setting bit T1C						
48	TCKEZtmr1	Delay from external	clock edge to tir	ner increment	2Tosc	—	7Tosc	—	

These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Applicable Devices 72 73 73A 74 74A 76 77

20.4 <u>Timing Parameter Symbology</u>

The timing parameter symbols have been created following one of the following formats:

1. TppS2	ppS	3. TCC:ST	(I ² C specifications only)							
2. TppS		4. Ts	(I ² C specifications only)							
Т										
F	Frequency	Т	Time							
Lower	case letters (pp) and their meanings:									
рр										
сс	CCP1	osc	OSC1							
ck	CLKOUT	rd	RD							
CS	CS	rw	RD or WR							
di	SDI	sc	SCK							
do	SDO	SS	SS							
dt	Data in	tO	TOCKI							
io	I/O port	t1	T1CKI							
mc	MCLR	wr	WR							
Upper	case letters and their meanings:									
S										
F	Fall	P	Period							
H	High	R	Rise							
1	Invalid (Hi-impedance)	V	Valid							
L	Low	Z	Hi-impedance							
I ² C only										
AA	output access	High	High							
BUF	Bus free	Low	Low							
Tcc:st	(I ² C specifications only)	_								
CC										
HD	Hold	SU	Setup							
ST										
DAT	DATA input hold	STO	STOP condition							
STA	START condition									
FIGURE	20-1: LOAD CONDITIONS									
			Logd condition 2							
	Load condition 1		Load condition 2							
VDD/2 RL										
	PL = 4640									
	$C_1 = 50 \text{ pc}$ for all pipe avaant $OSO2$ but inc									
	CL = 50 pF for all pins except OSC2, but including PORTD and PORTE outputs as ports									
	15 pF for OSC2 output									
Note: PORTD and PORTE are not implemented on the PIC16C76.										

Applicable Devices 72 73 73A 74 74A 76 77

FIGURE 20-15: USART SYNCHRONOUS TRANSMISSION (MASTER/SLAVE) TIMING



TABLE 20-11: USART SYNCHRONOUS TRANSMISSION REQUIREMENTS

Param No.	Sym	Characteristic	Min	Тур†	Max	Units	Conditions	
120	TckH2dtV	SYNC XMIT (MASTER & SLAVE) Clock high to data out valid	PIC16 C 76/77		_	80 100	ns	
121	Tckrf	Clock out rise time and fall time (Master Mode)	PIC16 C 76/77		_	45	ns	
122	Tdtrf	Data out rise time and fall time	PIC16 LC 76/77 PIC16 C 76/77		-	50 45	ns ns	
			PIC16 LC 76/77	—	-	50	ns	

†: Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

FIGURE 20-16: USART SYNCHRONOUS RECEIVE (MASTER/SLAVE) TIMING



TABLE 20-12: USART SYNCHRONOUS RECEIVE REQUIREMENTS

Parameter No.	Sym	Characteristic	Min	Тур†	Max	Units	Conditions
125	TdtV2ckL	SYNC RCV (MASTER & SLAVE) Data setup before CK \downarrow (DT setup time)	15	_	_	ns	
126	TckL2dtl	Data hold after CK \downarrow (DT hold time)	15	—	—	ns	

†: Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Applicable Devices 72 73 73A 74 74A 76 77

TABLE 20-13: A/D CONVERTER CHARACTERISTICS:

PIC16C76/77-04 (Commercial, Industrial, Extended) PIC16C76/77-10 (Commercial, Industrial, Extended) PIC16C76/77-20 (Commercial, Industrial, Extended) PIC16LC76/77-04 (Commercial, Industrial)

Param No.	Sym	Characteristic		Min	Тур†	Max	Units	Conditions
A01	Nr	Resolution		_	8-bits	bit	$\label{eq:VREF} \begin{array}{l} VREF = VDD = 5.12V,\\ VSS \leq VAIN \leq VREF \end{array}$	
A02	Eabs	Total Absolute error			_	<±1	LSb	$\label{eq:VREF} \begin{array}{l} VREF = VDD = 5.12V,\\ VSS \leq VAIN \leq VREF \end{array}$
A03	EIL	Integral linearity error			_	<±1	LSb	$\label{eq:VREF} \begin{array}{l} VREF = VDD = 5.12V,\\ VSS \leq VAIN \leq VREF \end{array}$
A04	Edl	Differential linearity error	ſ		_	<±1	LSb	$\label{eq:VREF} \begin{array}{l} VREF = VDD = 5.12V,\\ VSS \leq VAIN \leq VREF \end{array}$
A05	Efs	Full scale error		_	<±1	LSb	$\label{eq:VREF} \begin{array}{l} VREF = VDD = 5.12V,\\ VSS \leq VAIN \leq VREF \end{array}$	
A06	Eoff	Offset error		_	_	<±1	LSb	$\label{eq:VREF} \begin{array}{l} VREF = VDD = 5.12V,\\ VSS \leq VAIN \leq VREF \end{array}$
A10	_	Monotonicity		—	guaranteed	_	_	$VSS \leq VAIN \leq VREF$
A20	Vref	Reference voltage		3.0V	—	Vdd + 0.3	V	
A25	VAIN	Analog input voltage		Vss - 0.3	—	Vref + 0.3	V	
A30	Zain	Recommended impedance of analog voltage source		—	—	10.0	kΩ	
A40	IAD	A/D conversion current	PIC16 C 76/77	—	180	_	μΑ	Average current consump-
		(VDD)	PIC16 LC 76/77	—	90	_	μΑ	tion when A/D is on. (Note 1)
A50	IREF	VREF input current (Note 2)		10	_	1000	μA	During VAIN acquisition. Based on differential of VHOLD to VAIN to charge CHOLD, see Section 13.1.
					_	10	μA	During A/D Conversion cycle

These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: When A/D is off, it will not consume any current other than minor leakage current. The power-down current spec includes any such leakage from the A/D module.

2: VREF current is from RA3 pin or VDD pin, whichever is selected as reference input.