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### Understanding [Embedded - FPGAs \(Field Programmable Gate Array\)](#)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

### Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

#### Details

Product Status	Obsolete
Number of LABs/CLBs	23780
Number of Logic Elements/Cells	504000
Total RAM Bits	27695104
Number of I/O	544
Number of Gates	-
Voltage - Supply	1.07V ~ 1.13V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	1152-BBGA, FCBGA Exposed Pad
Supplier Device Package	1152-FBGA (35x35)
Purchase URL	<a href="https://www.e-xfl.com/product-detail/intel/5agxfb7h6f35c6n">https://www.e-xfl.com/product-detail/intel/5agxfb7h6f35c6n</a>

## Transceiver Power Supply Operating Conditions

Table 1-4: Transceiver Power Supply Operating Conditions for Arria V Devices

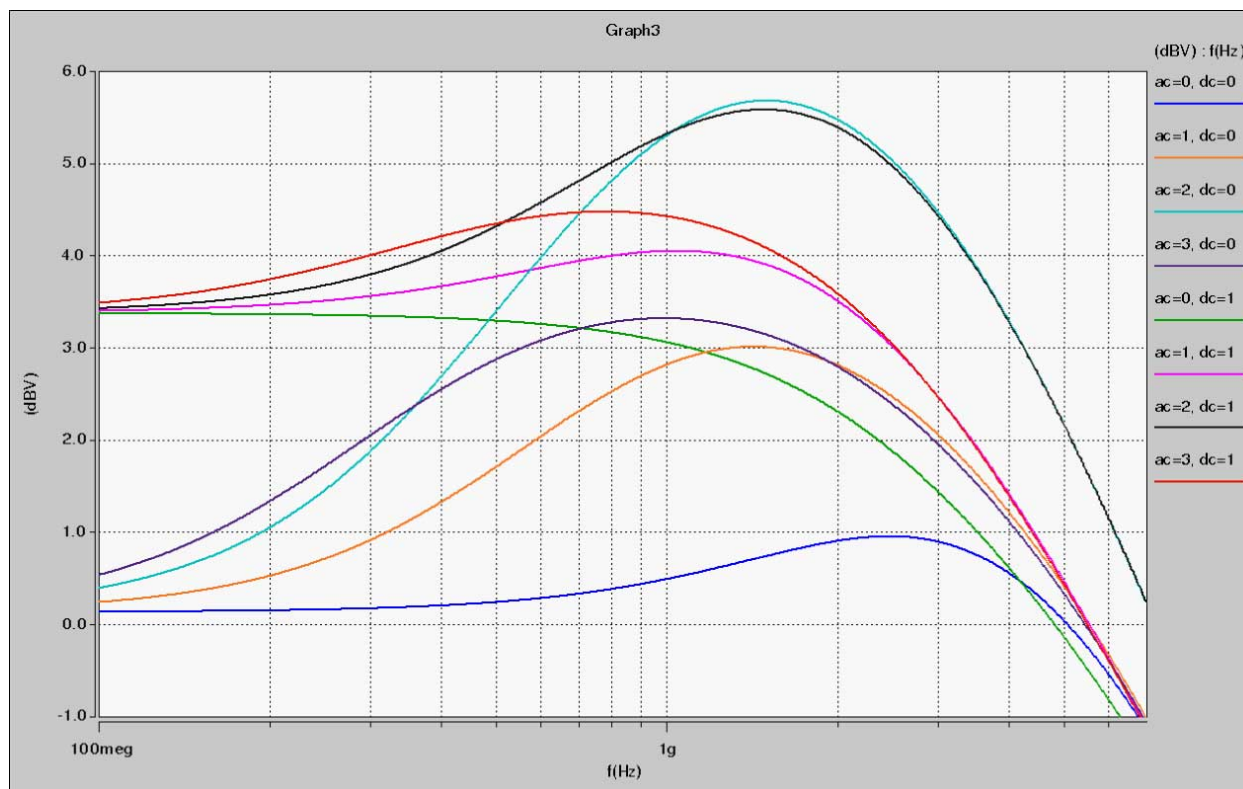
Symbol	Description	Minimum <sup>(5)</sup>	Typical	Maximum <sup>(5)</sup>	Unit
V <sub>CCA_GXBL</sub>	Transceiver high voltage power (left side)	2.375	2.500	2.625	V
V <sub>CCA_GXBR</sub>	Transceiver high voltage power (right side)				
V <sub>CCR_GXBL</sub>	GX and SX speed grades—receiver power (left side)	1.08/1.12	1.1/1.15 <sup>(6)</sup>	1.14/1.18	V
V <sub>CCR_GXBR</sub>	GX and SX speed grades—receiver power (right side)				
V <sub>CCR_GXBL</sub>	GT and ST speed grades—receiver power (left side)	1.17	1.20	1.23	V
V <sub>CCR_GXBR</sub>	GT and ST speed grades—receiver power (right side)				
V <sub>CCT_GXBL</sub>	GX and SX speed grades—transmitter power (left side)	1.08/1.12	1.1/1.15 <sup>(6)</sup>	1.14/1.18	V
V <sub>CCT_GXBR</sub>	GX and SX speed grades—transmitter power (right side)				
V <sub>CCT_GXBL</sub>	GT and ST speed grades—transmitter power (left side)	1.17	1.20	1.23	V
V <sub>CCT_GXBR</sub>	GT and ST speed grades—transmitter power (right side)				
V <sub>CCH_GXBL</sub>	Transmitter output buffer power (left side)	1.425	1.500	1.575	V
V <sub>CCH_GXBR</sub>	Transmitter output buffer power (right side)				

<sup>(5)</sup> The power supply value describes the budget for the DC (static) power supply tolerance and does not include the dynamic tolerance requirements. Refer to the PDN tool for the additional budget for the dynamic tolerance requirements.

<sup>(6)</sup> For data rate ≤ 3.2 Gbps, connect V<sub>CCR\_GXBL/R</sub>, V<sub>CCT\_GXBL/R</sub>, or V<sub>CCL\_GXBL/R</sub> to either 1.1-V or 1.15-V power supply. For data rate > 3.2 Gbps, connect V<sub>CCR\_GXBL/R</sub>, V<sub>CCT\_GXBL/R</sub>, or V<sub>CCL\_GXBL/R</sub> to a 1.15-V power supply. For details, refer to the Arria V GT, GX, ST, and SX Device Family Pin Connection Guidelines.

## CTLE Response at Data Rates $\leq 3.25$ Gbps across Supported AC Gain and DC Gain

Figure 1-3: CTLE Response at Data Rates  $\leq 3.25$  Gbps across Supported AC Gain and DC Gain for Arria V GX, GT, SX, and ST Devices



Symbol	V <sub>OD</sub> Setting <sup>(58)</sup>	V <sub>OD</sub> Value (mV)	V <sub>OD</sub> Setting <sup>(58)</sup>	V <sub>OD</sub> Value (mV)
	25	500	53	1060
	26	520	54	1080
	27	540	55	1100
	28	560	56	1120
	29	580	57	1140
	30	600	58	1160
	31	620	59	1180
	32	640	60	1200
	33	660		

## Transmitter Pre-Emphasis Levels

The following table lists the simulation data on the transmitter pre-emphasis levels in dB for the first post tap under the following conditions:

- Low-frequency data pattern—five 1s and five 0s
- Data rate—2.5 Gbps

The levels listed are a representation of possible pre-emphasis levels under the specified conditions only and the pre-emphasis levels may change with data pattern and data rate.

Arria V devices only support 1st post tap pre-emphasis with the following conditions:

- The 1st post tap pre-emphasis settings must satisfy  $|B| + |C| \leq 60$  where  $|B| = V_{OD}$  setting with termination value,  $R_{TERM} = 100 \Omega$  and  $|C| = 1st$  post tap pre-emphasis setting.
- $|B| - |C| > 5$  for data rates  $< 5$  Gbps and  $|B| - |C| > 8.25$  for data rates  $> 5$  Gbps.
- $(V_{MAX}/V_{MIN} - 1)\% < 600\%$ , where  $V_{MAX} = |B| + |C|$  and  $V_{MIN} = |B| - |C|$ .

Exception for PCIe Gen2 design:  $V_{OD}$  setting = 43 and pre-emphasis setting = 19 are allowed for PCIe Gen2 design with transmit de-emphasis – 6dB setting (`pipe_txdeemp = 1'b0`) using Altera PCIe Hard IP and PIPE IP cores.

<sup>(58)</sup> Convert these values to their binary equivalent form if you are using the dynamic reconfiguration mode for PMA analog controls.

## DSP Block Performance Specifications

Table 1-37: DSP Block Performance Specifications for Arria V Devices

Mode		Performance			Unit
		-I3, -C4	-I5, -C5	-C6	
Modes using One DSP Block	Independent $9 \times 9$ multiplication	370	310	220	MHz
	Independent $18 \times 19$ multiplication	370	310	220	MHz
	Independent $18 \times 25$ multiplication	370	310	220	MHz
	Independent $20 \times 24$ multiplication	370	310	220	MHz
	Independent $27 \times 27$ multiplication	310	250	200	MHz
	Two $18 \times 19$ multiplier adder mode	370	310	220	MHz
	$18 \times 18$ multiplier added summed with 36-bit input	370	310	220	MHz
Modes using Two DSP Blocks	Complex $18 \times 19$ multiplication	370	310	220	MHz

## Memory Block Performance Specifications

To achieve the maximum memory block performance, use a memory block clock that comes through global clock routing from an on-chip PLL and set to 50% output duty cycle. Use the Quartus Prime software to report timing for the memory block clocking schemes.

When you use the error detection cyclical redundancy check (CRC) feature, there is no degradation in  $f_{MAX}$ .

## High-Speed I/O Specifications

Table 1-40: High-Speed I/O Specifications for Arria V Devices

When  $J = 3$  to  $10$ , use the serializer/deserializer (SERDES) block. When  $J = 1$  or  $2$ , bypass the SERDES block.

For LVDS applications, you must use the PLLs in integer PLL mode.

The Arria V devices support the following output standards using true LVDS output buffer types on all I/O banks.

- True RSDS output standard with data rates of up to 360 Mbps
- True mini-LVDS output standard with data rates of up to 400 Mbps

Symbol		Condition	-I3, -C4			-I5, -C5			-C6			Unit
			Min	Typ	Max	Min	Typ	Max	Min	Typ	Max	
$f_{\text{HCLK\_in}}$ (input clock frequency) True Differential I/O Standards		Clock boost factor $W = 1$ to $40^{(72)}$	5	—	800	5	—	750	5	—	625	MHz
$f_{\text{HCLK\_in}}$ (input clock frequency) Single-Ended I/O Standards <sup>(73)</sup>		Clock boost factor $W = 1$ to $40^{(72)}$	5	—	625	5	—	625	5	—	500	MHz
$f_{\text{HCLK\_in}}$ (input clock frequency) Single-Ended I/O Standards <sup>(74)</sup>		Clock boost factor $W = 1$ to $40^{(72)}$	5	—	420	5	—	420	5	—	420	MHz
$f_{\text{HCLK\_OUT}}$ (output clock frequency)		—	5	—	$625^{(75)}$	5	—	$625^{(75)}$	5	—	$500^{(75)}$	MHz
Transmitter	True Differential I/O Standards - $f_{\text{HSDR}}$ (data rate)	SERDES factor $J = 3$ to $10^{(76)}$	<sup>(77)</sup>	—	1250	<sup>(77)</sup>	—	1250	<sup>(77)</sup>	—	1050	Mbps

<sup>(72)</sup> Clock boost factor ( $W$ ) is the ratio between the input data rate and the input clock rate.

<sup>(73)</sup> This applies to DPA and soft-CDR modes only.

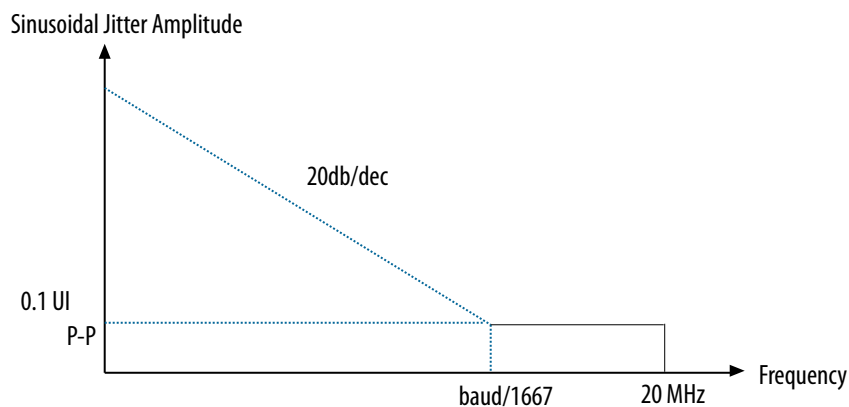
<sup>(74)</sup> This applies to non-DPA mode only.

<sup>(75)</sup> This is achieved by using the LVDS clock network.

<sup>(76)</sup> The  $F_{\text{max}}$  specification is based on the fast clock used for serial data. The interface  $F_{\text{max}}$  is also dependent on the parallel clock domain which is design dependent and requires timing analysis.

<sup>(77)</sup> The minimum specification depends on the clock source (for example, the PLL and clock pin) and the clock routing resource (global, regional, or local) that you use. The I/O differential buffer and input register do not have a minimum toggle rate.

Figure 1-6: LVDS Soft-CDR/DPA Sinusoidal Jitter Tolerance Specification for a Data Rate Less than 1.25 Gbps



## DLL Frequency Range Specifications

Table 1-43: DLL Frequency Range Specifications for Arria V Devices

Parameter	-I3, -C4	-I5, -C5	-C6	Unit
DLL operating frequency range	200 – 667	200 – 667	200 – 667	MHz

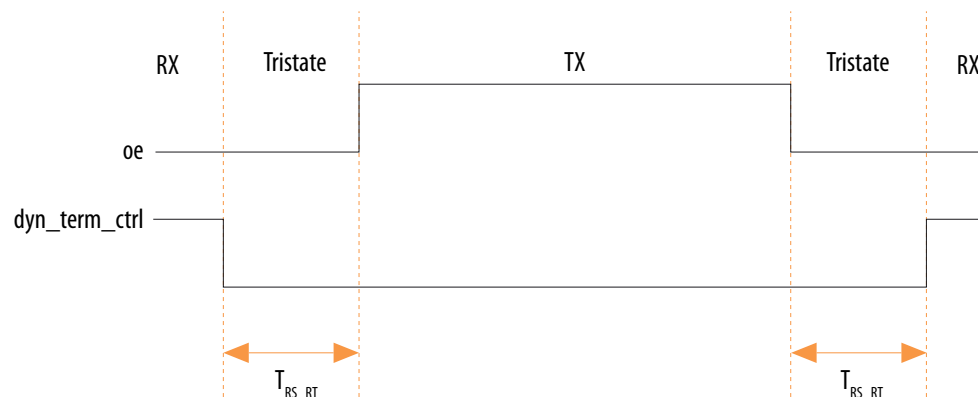
## DQS Logic Block Specifications

Table 1-44: DQS Phase Shift Error Specifications for DLL-Delayed Clock ( $t_{\text{DQS\_PSERR}}$ ) for Arria V Devices

This error specification is the absolute maximum and minimum error.

Number of DQS Delay Buffer	-I3, -C4	-I5, -C5	-C6	Unit
2	40	80	80	ps

Figure 1-7: Timing Diagram for oe and dyn\_term\_ctrl Signals



## Duty Cycle Distortion (DCD) Specifications

Table 1-47: Worst-Case DCD on Arria V I/O Pins

The output DCD cycle only applies to the I/O buffer. It does not cover the system DCD.

Symbol	-I3, -C4		-C5, -I5		-C6		Unit
	Min	Max	Min	Max	Min	Max	
Output Duty Cycle	45	55	45	55	45	55	%

## HPS Specifications

This section provides HPS specifications and timing for Arria V devices.

For HPS reset, the minimum reset pulse widths for the HPS cold and warm reset signals (HPS\_nRST and HPS\_nPOR) are six clock cycles of HPS\_CLK1.

## FPP Configuration Timing when DCLK-to-DATA[] &gt;1

Table 1-67: FPP Timing Parameters When DCLK-to-DATA[] Ratio is &gt;1 for Arria V Devices

Use these timing parameters when you use the decompression and design security features.

Symbol	Parameter	Minimum	Maximum	Unit
$t_{CF2CD}$	nCONFIG low to CONF_DONE low	—	600	ns
$t_{CF2ST0}$	nCONFIG low to nSTATUS low	—	600	ns
$t_{CFG}$	nCONFIG low pulse width	2	—	$\mu$ s
$t_{STATUS}$	nSTATUS low pulse width	268	1506 <sup>(98)</sup>	$\mu$ s
$t_{CF2ST1}$	nCONFIG high to nSTATUS high	—	1506 <sup>(99)</sup>	$\mu$ s
$t_{CF2CK}^{(100)}$	nCONFIG high to first rising edge on DCLK	1506	—	$\mu$ s
$t_{ST2CK}^{(100)}$	nSTATUS high to first rising edge of DCLK	2	—	$\mu$ s
$t_{DSU}$	DATA[] setup time before rising edge on DCLK	5.5	—	ns
$t_{DH}$	DATA[] hold time after rising edge on DCLK	$N - 1/f_{DCLK}^{(101)}$	—	s
$t_{CH}$	DCLK high time	$0.45 \times 1/f_{MAX}$	—	s
$t_{CL}$	DCLK low time	$0.45 \times 1/f_{MAX}$	—	s
$t_{CLK}$	DCLK period	$1/f_{MAX}$	—	s
$f_{MAX}$	DCLK frequency (FPP $\times 8/ \times 16$ )	—	125	MHz
$t_R$	Input rise time	—	40	ns
$t_F$	Input fall time	—	40	ns
$t_{CD2UM}$	CONF_DONE high to user mode <sup>(102)</sup>	175	437	$\mu$ s

<sup>(98)</sup> This value can be obtained if you do not delay configuration by extending the nCONFIG or nSTATUS low pulse width.

<sup>(99)</sup> This value can be obtained if you do not delay configuration by externally holding nSTATUS low.

<sup>(100)</sup> If nSTATUS is monitored, follow the  $t_{ST2CK}$  specification. If nSTATUS is not monitored, follow the  $t_{CF2CK}$  specification.

<sup>(101)</sup>  $N$  is the DCLK-to-DATA[] ratio and  $f_{DCLK}$  is the DCLK frequency of the system.

<sup>(102)</sup> The minimum and maximum numbers apply only if you chose the internal oscillator as the clock source for initializing the device.

**Related Information**

- [PS Configuration Timing](#) on page 1-81
- [AS Configuration Timing](#)  
Provides the AS configuration timing waveform.

## DCLK Frequency Specification in the AS Configuration Scheme

**Table 1-69: DCLK Frequency Specification in the AS Configuration Scheme**

This table lists the internal clock frequency specification for the AS configuration scheme. The DCLK frequency specification applies when you use the internal oscillator as the configuration clock source. The AS multi-device configuration scheme does not support DCLK frequency of 100 MHz.

Parameter	Minimum	Typical	Maximum	Unit
DCLK frequency in AS configuration scheme	5.3	7.9	12.5	MHz
	10.6	15.7	25.0	MHz
	21.3	31.4	50.0	MHz
	42.6	62.9	100.0	MHz

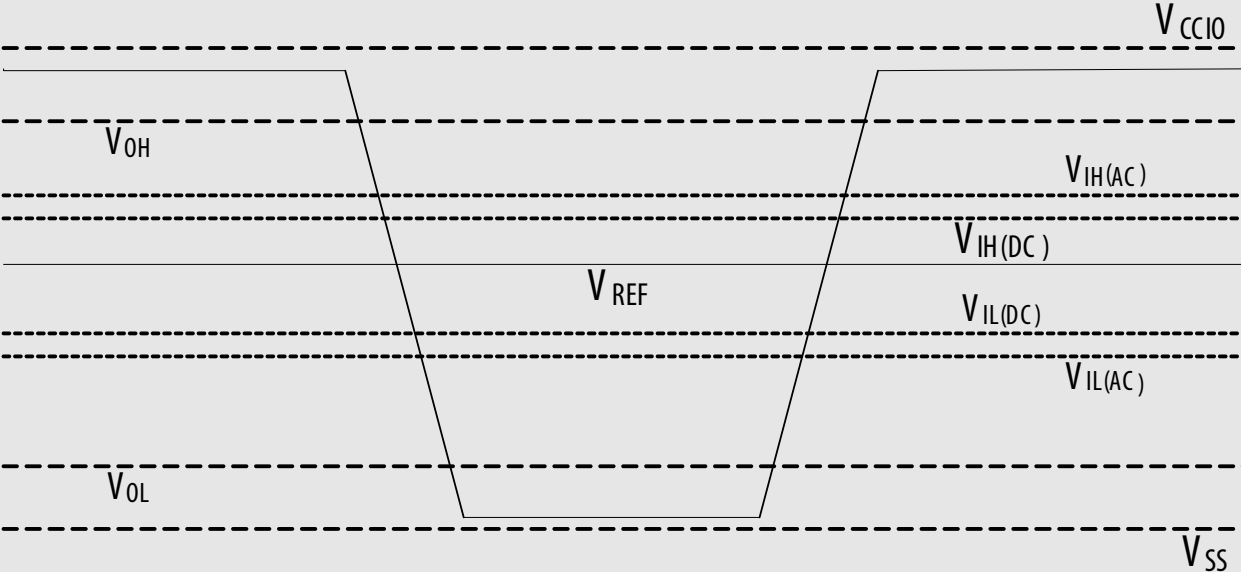
## PS Configuration Timing

**Table 1-70: PS Timing Parameters for Arria V Devices**

Symbol	Parameter	Minimum	Maximum	Unit
t <sub>CF2CD</sub>	nCONFIG low to CONF_DONE low	—	600	ns
t <sub>CF2ST0</sub>	nCONFIG low to nSTATUS low	—	600	ns
t <sub>CFG</sub>	nCONFIG low pulse width	2	—	μs
t <sub>STATUS</sub>	nSTATUS low pulse width	268	1506 <sup>(103)</sup>	μs
t <sub>CF2ST1</sub>	nCONFIG high to nSTATUS high	—	1506 <sup>(104)</sup>	μs

<sup>(103)</sup> You can obtain this value if you do not delay configuration by extending the nCONFIG or nSTATUS low pulse width.

<sup>(104)</sup> You can obtain this value if you do not delay configuration by externally holding nSTATUS low.

Term	Definition
Single-ended voltage referenced I/O standard	<p>The JEDEC standard for the SSTL and HSTL I/O defines both the AC and DC input signal values. The AC values indicate the voltage levels at which the receiver must meet its timing specifications. The DC values indicate the voltage levels at which the final logic state of the receiver is unambiguously defined. After the receiver input has crossed the AC value, the receiver changes to the new logic state.</p> <p>The new logic state is then maintained as long as the input stays beyond the DC threshold. This approach is intended to provide predictable receiver timing in the presence of input waveform ringing.</p> <p>Single-Ended Voltage Referenced I/O Standard</p> 
$t_C$	High-speed receiver/transmitter input and output clock period.
TCCS (channel-to-channel-skew)	The timing difference between the fastest and slowest output edges, including the $t_{CO}$ variation and clock skew, across channels driven by the same PLL. The clock is included in the TCCS measurement (refer to the Timing Diagram figure under SW in this table).
$t_{DUTY}$	High-speed I/O block—Duty cycle on high-speed transmitter output clock.

Date	Version	Changes
June 2012	2.0	<ul style="list-style-type: none"><li>• Updated for the Quartus II software v12.0 release:</li><li>• Restructured document.</li><li>• Updated “Supply Current and Power Consumption” section.</li><li>• Updated Table 20, Table 21, Table 24, Table 25, Table 26, Table 35, Table 39, Table 43, and Table 52.</li><li>• Added Table 22, Table 23, and Table 33.</li><li>• Added Figure 1–1 and Figure 1–2.</li><li>• Added “Initialization” and “Configuration Files” sections.</li></ul>
February 2012	1.3	<ul style="list-style-type: none"><li>• Updated Table 2–1.</li><li>• Updated Transceiver-FPGA Fabric Interface rows in Table 2–20.</li><li>• Updated <math>V_{CCP}</math> description.</li></ul>
December 2011	1.2	Updated Table 2–1 and Table 2–3.
November 2011	1.1	<ul style="list-style-type: none"><li>• Updated Table 2–1, Table 2–19, Table 2–26, and Table 2–36.</li><li>• Added Table 2–5.</li><li>• Added Figure 2–4.</li></ul>
August 2011	1.0	Initial release.

**Related Information**

- [PowerPlay Early Power Estimator User Guide](#)  
For more information about the EPE tool.
- [PowerPlay Power Analysis](#)  
For more information about PowerPlay power analysis.

**Power Consumption**

Altera offers two ways to estimate power consumption for a design—the Excel-based Early Power Estimator and the Quartus II PowerPlay Power Analyzer feature.

**Note:** You typically use the interactive Excel-based Early Power Estimator before designing the FPGA to get a magnitude estimate of the device power. The Quartus II PowerPlay Power Analyzer provides better quality estimates based on the specifics of the design after you complete place-and-route. The PowerPlay Power Analyzer can apply a combination of user-entered, simulation-derived, and estimated signal activities that, when combined with detailed circuit models, yields very accurate power estimates.

**Related Information**

- [PowerPlay Early Power Estimator User Guide](#)  
For more information about the EPE tool.
- [PowerPlay Power Analysis](#)  
For more information about PowerPlay power analysis.

**I/O Pin Leakage Current****Table 2-8: I/O Pin Leakage Current for Arria V GZ Devices**

If  $V_O = V_{CCIO}$  to  $V_{CCIO_{MAX}}$ , 100  $\mu A$  of leakage current per I/O is expected.

Symbol	Description	Conditions	Min	Typ	Max	Unit
$I_I$	Input pin	$V_I = 0\text{ V to }V_{CCIO_{MAX}}$	-30	—	30	$\mu A$
$I_{OZ}$	Tri-stated I/O pin	$V_O = 0\text{ V to }V_{CCIO_{MAX}}$	-30	—	30	$\mu A$

Table 2-19: Differential SSTL I/O Standards for Arria V GZ Devices

I/O Standard	$V_{CCIO}$ (V)			$V_{SWING(DC)}$ (V)		$V_{X(AC)}$ (V)			$V_{SWING(AC)}$ (V)	
	Min	Typ	Max	Min	Max	Min	Typ	Max	Min	Max
SSTL-2 Class I, II	2.375	2.5	2.625	0.3	$V_{CCIO} + 0.6$	$V_{CCIO}/2 - 0.2$	—	$V_{CCIO}/2 + 0.2$	0.62	$V_{CCIO} + 0.6$
SSTL-18 Class I, II	1.71	1.8	1.89	0.25	$V_{CCIO} + 0.6$	$V_{CCIO}/2 - 0.175$	—	$V_{CCIO}/2 + 0.175$	0.5	$V_{CCIO} + 0.6$
SSTL-15 Class I, II	1.425	1.5	1.575	0.2	<sup>(127)</sup>	$V_{CCIO}/2 - 0.15$	—	$V_{CCIO}/2 + 0.15$	0.35	—
SSTL-135 Class I, II	1.283	1.35	1.45	0.2	<sup>(127)</sup>	$V_{CCIO}/2 - 0.15$	$V_{CCIO}/2$	$V_{CCIO}/2 + 0.15$	$2(V_{IH(AC)} - V_{REF})$	$2(V_{IL(AC)} - V_{REF})$
SSTL-125 Class I, II	1.19	1.25	1.31	0.18	<sup>(127)</sup>	$V_{CCIO}/2 - 0.15$	$V_{CCIO}/2$	$V_{CCIO}/2 + 0.15$	$2(V_{IH(AC)} - V_{REF})$	—
SSTL-12 Class I, II	1.14	1.2	1.26	0.18	—	$V_{REF} - 0.15$	$V_{CCIO}/2$	$V_{REF} + 0.15$	-0.30	0.30

Table 2-20: Differential HSTL and HSUL I/O Standards for Arria V GZ Devices

I/O Standard	$V_{CCIO}$ (V)			$V_{DIF(DC)}$ (V)		$V_{X(AC)}$ (V)			$V_{CM(DC)}$ (V)			$V_{DIF(AC)}$ (V)	
	Min	Typ	Max	Min	Max	Min	Typ	Max	Min	Typ	Max	Min	Max
HSTL-18 Class I, II	1.71	1.8	1.89	0.2	—	0.78	—	1.12	0.78	—	1.12	0.4	—
HSTL-15 Class I, II	1.425	1.5	1.575	0.2	—	0.68	—	0.9	0.68	—	0.9	0.4	—

<sup>(127)</sup> The maximum value for  $V_{SWING(DC)}$  is not defined. However, each single-ended signal needs to be within the respective single-ended limits ( $V_{IH(DC)}$  and  $V_{IL(DC)}$ ).

Symbol	Parameter	Min	Typ	Max	Unit
$t_{\text{OUTPJ\_IO}}^{(173), (175)}$	Period Jitter for a clock output on a regular I/O in integer PLL ( $f_{\text{OUT}} \geq 100$ MHz)	—	—	600	ps (p-p)
	Period Jitter for a clock output on a regular I/O in integer PLL ( $f_{\text{OUT}} < 100$ MHz)	—	—	60	mUI (p-p)
$t_{\text{FOUTPJ\_IO}}^{(173), (175), (176)}$	Period Jitter for a clock output on a regular I/O in fractional PLL ( $f_{\text{OUT}} \geq 100$ MHz)	—	—	600	ps (p-p)
	Period Jitter for a clock output on a regular I/O in fractional PLL ( $f_{\text{OUT}} < 100$ MHz)	—	—	60	mUI (p-p)
$t_{\text{OUTCCJ\_IO}}^{(173), (175)}$	Cycle-to-cycle Jitter for a clock output on a regular I/O in integer PLL ( $f_{\text{OUT}} \geq 100$ MHz)	—	—	600	ps (p-p)
	Cycle-to-cycle Jitter for a clock output on a regular I/O in integer PLL ( $f_{\text{OUT}} < 100$ MHz)	—	—	60	mUI (p-p)
$t_{\text{FOUTCCJ\_IO}}^{(173), (175), (176)}$	Cycle-to-cycle Jitter for a clock output on a regular I/O in fractional PLL ( $f_{\text{OUT}} \geq 100$ MHz)	—	—	600	ps (p-p)
	Cycle-to-cycle Jitter for a clock output on a regular I/O in fractional PLL ( $f_{\text{OUT}} < 100$ MHz)	—	—	60	mUI (p-p)
$t_{\text{CASC\_OUTPJ\_DC}}^{(173), (177)}$	Period Jitter for a dedicated clock output in cascaded PLLs ( $f_{\text{OUT}} \geq 100$ MHz)	—	—	175	ps (p-p)
	Period Jitter for a dedicated clock output in cascaded PLLs ( $f_{\text{OUT}} < 100$ MHz)	—	—	17.5	mUI (p-p)
$dK_{\text{BIT}}$	Bit number of Delta Sigma Modulator (DSM)	8	24	32	Bits

<sup>(175)</sup> The external memory interface clock output jitter specifications use a different measurement method, which is available in the "Memory Output Clock Jitter Specification for Arria V GZ Devices" table.

<sup>(176)</sup> This specification only covered fractional PLL for low bandwidth. The  $f_{\text{VCO}}$  for fractional value range 0.05–0.95 must be  $\geq 1000$  MHz.

<sup>(177)</sup> The cascaded PLL specification is only applicable with the following condition:

- Upstream PLL:  $0.59\text{MHz} \leq \text{Upstream PLL BW} < 1$  MHz
- Downstream PLL:  $\text{Downstream PLL BW} > 2$  MHz

Symbol	Conditions	C3, I3L			C4, I4			Unit
		Min	Typ	Max	Min	Typ	Max	
$f_{\text{HCLK\_in}}$ (input clock frequency) True Differential I/O Standards <sup>(179)</sup>	Clock boost factor $W = 1$ to $40$ <sup>(180)</sup>	5	—	625	5	—	525	MHz
$f_{\text{HCLK\_in}}$ (input clock frequency) Single Ended I/O Standards	Clock boost factor $W = 1$ to $40$ <sup>(180)</sup>	5	—	625	5	—	525	MHz
$f_{\text{HCLK\_in}}$ (input clock frequency) Single Ended I/O Standards	Clock boost factor $W = 1$ to $40$ <sup>(180)</sup>	5	—	420	5	—	420	MHz
$f_{\text{HCLK\_OUT}}$ (output clock frequency)	—	5	—	625 <sup>(181)</sup>	5	—	525 <sup>(181)</sup>	MHz

### Transmitter High-Speed I/O Specifications

**Table 2-40: Transmitter High-Speed I/O Specifications for Arria V GZ Devices**

When  $J = 3$  to  $10$ , use the serializer/deserializer (SERDES) block.

When  $J = 1$  or  $2$ , bypass the SERDES block.

<sup>(179)</sup> This only applies to DPA and soft-CDR modes.

<sup>(180)</sup> Clock Boost Factor ( $W$ ) is the ratio between the input data rate to the input clock rate.

<sup>(181)</sup> This is achieved by using the LVDS clock network.

Symbol	Conditions	C3, I3L			C4, I4			Unit
		Min	Typ	Max	Min	Typ	Max	
True Differential I/O Standards - $f_{\text{HSDRDPA}}$ (data rate)	SERDES factor J = 3 to 10 (192), (193), (194), (195), (196), (197)	150	—	1250	150	—	1050	Mbps
	SERDES factor J $\geq 4$ LVDS RX with DPA (193), (195), (196), (197)	150	—	1600	150	—	1250	Mbps
	SERDES factor J = 2, uses DDR Registers	(198)	—	(199)	(198)	—	(199)	Mbps
	SERDES factor J = 1, uses SDR Register	(198)	—	(199)	(198)	—	(199)	Mbps
$f_{\text{HSDR}}$ (data rate)	SERDES factor J = 3 to 10	(198)	—	(200)	(198)	—	(200)	Mbps
	SERDES factor J = 2, uses DDR Registers	(198)	—	(199)	(198)	—	(199)	Mbps
	SERDES factor J = 1, uses SDR Register	(198)	—	(199)	(198)	—	(199)	Mbps

(192) The  $F_{\text{MAX}}$  specification is based on the fast clock used for serial data. The interface  $F_{\text{MAX}}$  is also dependent on the parallel clock domain which is design dependent and requires timing analysis.

(193) Arria V GZ RX LVDS will need DPA. For Arria V GZ TX LVDS, the receiver side component must have DPA.

(194) Arria V GZ LVDS serialization and de-serialization factor needs to be x4 and above.

(195) Requires package skew compensation with PCB trace length.

(196) Do not mix single-ended I/O buffer within LVDS I/O bank.

(197) Chip-to-chip communication only with a maximum load of 5 pF.

(198) The minimum specification depends on the clock source (for example, the PLL and clock pin) and the clock routing resource (global, regional, or local) that you use. The I/O differential buffer and input register do not have a minimum toggle rate.

(199) The maximum ideal data rate is the SERDES factor (J) x the PLL maximum output frequency ( $f_{\text{OUT}}$ ) provided you can close the design timing and the signal integrity simulation is clean.

(200) You can estimate the achievable maximum data rate for non-DPA mode by performing link timing closure analysis. You must consider the board skew margin, transmitter delay margin, and receiver sampling margin to determine the maximum data rate supported.

Number of DQS Delay Buffers	C3, I3L	C4, I4	Unit
4	120	128	ps

## Memory Output Clock Jitter Specifications

**Table 2-50: Memory Output Clock Jitter Specification for Arria V GZ Devices**

The clock jitter specification applies to the memory output clock pins generated using differential signal-splitter and DDIO circuits clocked by a PLL output routed on a PHY, regional, or global clock network as specified. Altera recommends using PHY clock networks whenever possible.

The clock jitter specification applies to the memory output clock pins clocked by an integer PLL.

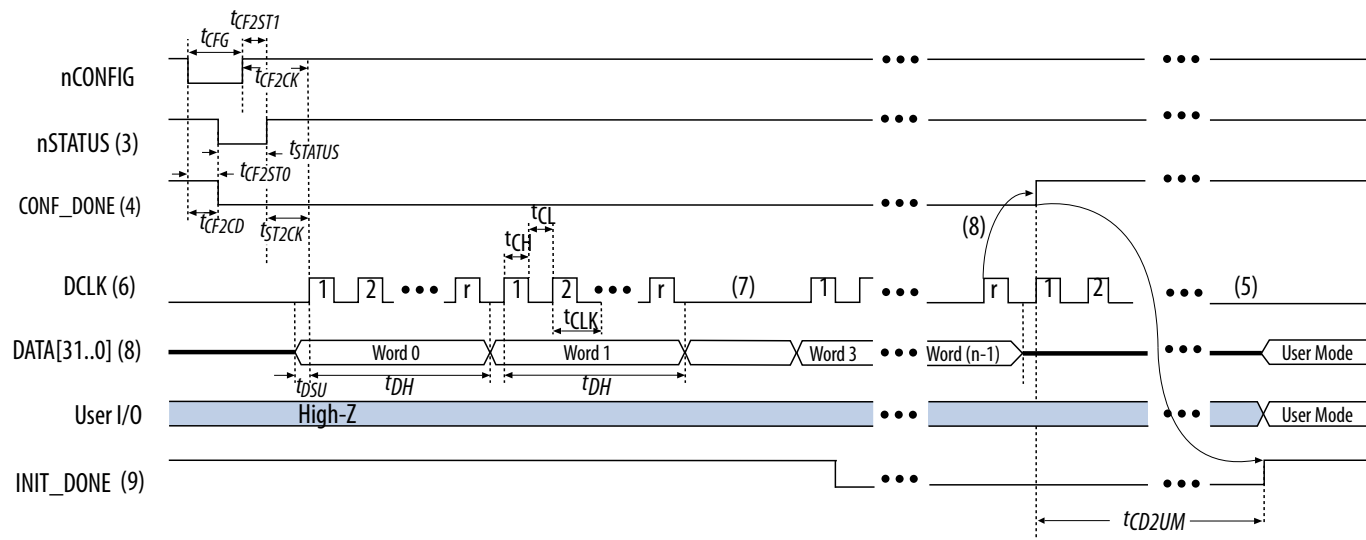
The memory output clock jitter is applicable when an input jitter of 30 ps peak-to-peak is applied with bit error rate (BER) -12, equivalent to 14 sigma.

Clock Network	Parameter	Symbol	C3, I3L		C4, I4		Unit
			Min	Max	Min	Max	
Regional	Clock period jitter	$t_{JIT(per)}$	-55	55	-55	55	ps
	Cycle-to-cycle period jitter	$t_{JIT(cc)}$	-110	110	-110	110	ps
	Duty cycle jitter	$t_{JIT(duty)}$	-82.5	82.5	-82.5	82.5	ps
Global	Clock period jitter	$t_{JIT(per)}$	-82.5	82.5	-82.5	82.5	ps
	Cycle-to-cycle period jitter	$t_{JIT(cc)}$	-165	165	-165	165	ps
	Duty cycle jitter	$t_{JIT(duty)}$	-90	90	-90	90	ps
PHY Clock	Clock period jitter	$t_{JIT(per)}$	-30	30	-35	35	ps
	Cycle-to-cycle period jitter	$t_{JIT(cc)}$	-60	60	-70	70	ps
	Duty cycle jitter	$t_{JIT(duty)}$	-45	45	-56	56	ps

## FPP Configuration Timing when DCLK to DATA[] &gt; 1

Figure 2-8: FPP Configuration Timing Waveform When the DCLK-to-DATA[] Ratio is &gt;1 ,

Timing when using a MAX II device, MAX V device, or microprocessor as an external host.



## Notes:

1. To find out the DCLK-to-DATA[] ratio for your system, refer to the "DCLK-to-DATA[] Ratio for Arria V GZ Devices" table.
2. The beginning of this waveform shows the device in user mode. In user mode, nCONFIG, nSTATUS, and CONF\_DONE are at logic high levels. When nCONFIG is pulled low, a reconfiguration cycle begins.
3. After power-up, the Arria V GZ device holds nSTATUS low for the time as specified by the POR delay.
4. After power-up, before and during configuration, CONF\_DONE is low.
5. Do not leave DCLK floating after configuration is complete. DCLK is ignored after configuration is complete. It can toggle high or low if required.
6. "r" denotes the DCLK-to-DATA[] ratio. For the DCLK-to-DATA[] ratio based on the decompression and the design security feature enable settings, refer to the "DCLK-to-DATA[] Ratio for Arria V GZ Devices" table.
7. If needed, pause DCLK by holding it low. When DCLK restarts, the external host must provide data on the DATA[31..0] pins prior to sending the first DCLK rising edge.
8. To ensure a successful configuration, send the entire configuration data to the Arria V GZ device. CONF\_DONE is released high after the Arria V GZ device receives all the configuration data successfully. After CONF\_DONE goes high, send two additional falling edges on DCLK to begin initialization and enter user mode.
9. After the option bit to enable the INIT\_DONE pin is configured into the device, the INIT\_DONE goes low.

Table 2-62: Uncompressed .rbf Sizes for Arria V GZ Devices

Variant	Member Code	Configuration .rbf Size (bits)	IOCSR .rbf Size (bits) <sup>(223)</sup>
Arria V GZ	E1	137,598,880	562,208
	E3	137,598,880	562,208
	E5	213,798,880	561,760
	E7	213,798,880	561,760

Table 2-63: Minimum Configuration Time Estimation for Arria V GZ Devices

Variant	Member Code	Active Serial <sup>(224)</sup>			Fast Passive Parallel <sup>(225)</sup>		
		Width	DCLK (MHz)	Min Config Time (ms)	Width	DCLK (MHz)	Min Config Time (ms)
Arria V GZ	E1	4	100	344	32	100	43
	E3	4	100	344	32	100	43
	E5	4	100	534	32	100	67
	E7	4	100	534	32	100	67

## Remote System Upgrades Circuitry Timing Specification

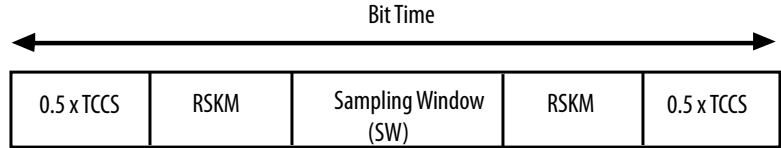
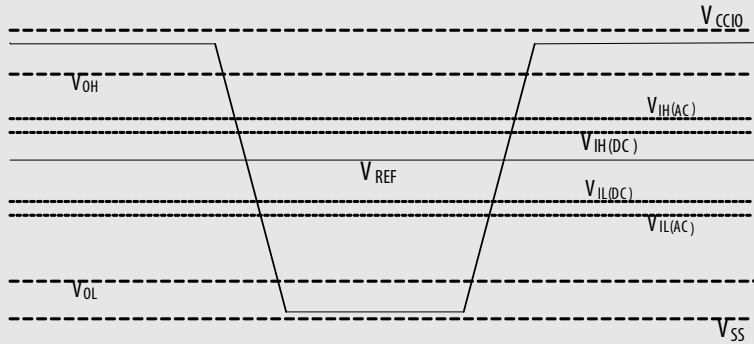
Table 2-64: Remote System Upgrade Circuitry Timing Specifications

Parameter	Minimum	Maximum	Unit
$t_{RU\_nCONFIG}$ <sup>(226)</sup>	250	—	ns
$t_{RU\_nRSTIMER}$ <sup>(227)</sup>	250	—	ns

<sup>(223)</sup> The IOCSR .rbf size is specifically for the Configuration via Protocol (CvP) feature.

<sup>(224)</sup> DCLK frequency of 100 MHz using external CLKUSR.

<sup>(225)</sup> Max FPGA FPP bandwidth may exceed bandwidth available from some external storage or control logic.

Term	Definition
$R_L$	Receiver differential input discrete resistor (external to the Arria V GZ device).
SW (sampling window)	<p>Timing Diagram—the period of time during which the data must be valid in order to capture it correctly. The setup and hold times determine the ideal strobe position within the sampling window, as shown:</p>  <p>The diagram shows a horizontal timeline. A double-headed arrow labeled 'Bit Time' spans the entire duration. Below this, a sequence of five rectangular blocks is shown: '0.5 x TCCS', 'RSKM', 'Sampling Window (SW)', 'RSKM', and '0.5 x TCCS'. The 'Sampling Window (SW)' is the central block, and the 'RSKM' blocks are positioned immediately before and after it. The '0.5 x TCCS' blocks are at the far left and right ends of the timeline.</p>
Single-ended voltage referenced I/O standard	<p>The JEDEC standard for SSTL and HSTL I/O defines both the AC and DC input signal values. The AC values indicate the voltage levels at which the receiver must meet its timing specifications. The DC values indicate the voltage levels at which the final logic state of the receiver is unambiguously defined. After the receiver input has crossed the AC value, the receiver changes to the new logic state.</p> <p>The new logic state is then maintained as long as the input stays beyond the DC threshold. This approach is intended to provide predictable receiver timing in the presence of input waveform ringing:</p> <p>Single-Ended Voltage Referenced I/O Standard</p>  <p>The diagram shows a trapezoidal waveform representing a signal transition. The signal starts at a high level, drops to a low level, and then rises back to a high level. The diagram includes several horizontal dashed lines representing voltage levels: <math>V_{OH}</math> (top), <math>V_{OH(AC)}</math> (just below <math>V_{OH}</math>), <math>V_{OH(DC)}</math> (below <math>V_{OH(AC)}</math>), <math>V_{REF}</math> (middle), <math>V_{IL(DC)}</math> (just above <math>V_{REF}</math>), <math>V_{IL(AC)}</math> (just below <math>V_{IL(DC)}</math>), <math>V_{SS}</math> (bottom), and <math>V_{CCIO}</math> (top right, above <math>V_{OH}</math>).</p>