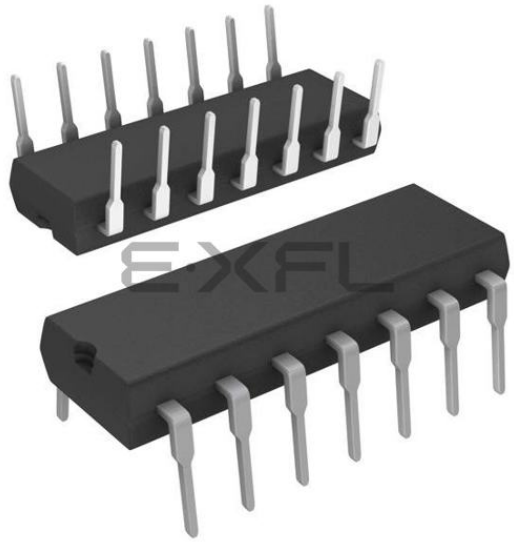

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Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	32MHz
Connectivity	I²C, LINbus, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	12
Program Memory Size	7KB (4K x 14)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	512 x 8
Voltage - Supply (Vcc/Vdd)	2.3V ~ 5.5V
Data Converters	A/D 8x10b; D/A 1x8b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Through Hole
Package / Case	14-DIP (0.300", 7.62mm)
Supplier Device Package	14-PDIP
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic16f1614-i-p

TABLE 1-2: PIC16(L)F1614 PINOUT DESCRIPTION (CONTINUED)

Name	Function	Input Type	Output Type	Description
RC1/AN5/C1IN1-/C2IN1-/ T4IN(1)/SMTSIG2(1)/SDI(1)	RC1	TTL/ST	CMOS/OD	General purpose I/O.
	AN5	AN	—	ADC Channel input.
	C1IN1-	AN	—	Comparator negative input.
	C2IN1-	AN	—	Comparator negative input.
	T4IN	TTL/ST	—	Timer4 input.
	SMTSIG2	TTL/ST	—	SMT2 signal input.
	CLCIN2	ST	—	Configurable Logic Cell source input.
	SDI	CMOS	—	SPI data input.
RC2/AN6/C1IN2-/C2IN2-	RC2	TTL/ST	CMOS/OD	General purpose I/O.
	AN6	AN	—	ADC Channel input.
	C1IN2-	AN	—	Comparator negative input.
	C2IN2-	AN	—	Comparator negative input.
RC3/AN7/C1IN3-/C2IN3-/T5G(1)/ CCP2(1)/CLCIN0(1)/ATCC(1)/$\overline{SS}$	RC3	TTL/ST	—	General purpose input with IOC and WPU.
	AN7	AN	—	ADC Channel input.
	C1IN3-	AN	—	Comparator negative input.
	C2IN3-	AN	—	Comparator negative input.
	T5G	ST	—	Timer5 Gate input.
	CCP2	TTL/ST	CMOS/OD	Capture/Compare/PWM2.
	CLCIN0	ST	—	Configurable Logic Cell source input.
	ATCC	ST	—	Angular Timer Capture/Compare input.
	$\overline{SS}$	ST	—	Slave Select input.
RC4/T3G(1)/CLCIN1(1)/CK(1)/ HIC4	RC4	TTL/ST	CMOS/OD	General purpose I/O.
	T3G	TTL/ST	—	Timer3 Gate input.
	CLCIN1	ST	—	Configurable Logic Cell source input.
	CK	ST	CMOS	USART synchronous clock.
	HIC4	TTL	CMOS	High Current I/O.
RC5/T3CKI(1)/CCP1(1)/RX(1)/ ATIN(1)/HIC5	RC5	TTL/ST	CMOS/OD	General purpose I/O.
	T3CKI	TTL/ST	—	Timer3 clock input.
	CCP1	TTL/ST	CMOS/OD	Capture/Compare/PWM1.
	RX	ST	—	USART asynchronous input.
	ATIN	TTL/ST	—	Angular Timer clock input.
	HIC5	TTL	—	High Current I/O.
VDD	VDD	Power	—	Positive supply.
VSS	VSS	Power	—	Ground reference.

Legend: AN = Analog input or output CMOS = CMOS compatible input or output OD = Open-Drain
TTL = TTL compatible input ST = Schmitt Trigger input with CMOS levels I^2C = Schmitt Trigger input with I^2C levels
HV = High Voltage XTAL = Crystal

Note 1: Default peripheral input. Input can be moved to any other pin with the PPS input selection registers.

2: All pin outputs default to PORT latch data. Any pin can be selected as a digital peripheral output with the PPS output selection registers. See Register 13-1.

3.3 Data Memory Organization

The data memory is partitioned in 32 memory banks with 128 bytes in a bank. Each bank consists of (Figure 3-2):

- 12 core registers
- 20 Special Function Registers (SFR)
- Up to 80 bytes of General Purpose RAM (GPR)
- 16 bytes of common RAM

The active bank is selected by writing the bank number into the Bank Select Register (BSR). Unimplemented memory will read as '0'. All data memory can be accessed either directly (via instructions that use the file registers) or indirectly via the two File Select Registers (FSR). See **Section 3.6 "Indirect Addressing"** for more information.

Data memory uses a 12-bit address. The upper five bits of the address define the Bank address and the lower seven bits select the registers/RAM in that bank.

3.3.1 CORE REGISTERS

The core registers contain the registers that directly affect the basic operation. The core registers occupy the first 12 addresses of every data memory bank (addresses x00h/x80h through x0Bh/x8Bh). These registers are listed below in Table 3-1. For detailed

TABLE 3-1: CORE REGISTERS

Addresses	BANKx
x00h or x80h	INDF0
x01h or x81h	INDF1
x02h or x82h	PCL
x03h or x83h	STATUS
x04h or x84h	FSR0L
x05h or x85h	FSR0H
x06h or x86h	FSR1L
x07h or x87h	FSR1H
x08h or x88h	BSR
x09h or x89h	WREG
x0Ah or x8Ah	PCLATH
x0Bh or x8Bh	INTCON

TABLE 3-14: SPECIAL FUNCTION REGISTER SUMMARY (CONTINUED)

Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
Bank 14											
70Ch to 710h	—	Unimplemented								—	—
711h	WDTCON0	—	—	WDTPS<4:0>					SEN	--qq qqqq	--qq qqqq
712h	WDTCON1	—	WDTCS<2:0>			—	WINDOW<2:0>			-qqq -qqq	-qqq -qqq
713h	WDTPSL	PSCNT<7:0>								0000 0000	0000 0000
714h	WDTPSH	PSCNT<15:8>								0000 0000	0000 0000
715h	WDTTMR	WDTTMR<4:0>					STATE	PSCNT<17:16>		0000 0000	0000 0000
716h	—	Unimplemented								—	—
717h	—	Unimplemented								—	—
718h	SCANLADRL	LADR<7:0>								0000 0000	0000 0000
719h	SCANLADRH	LADR<15:8>								0000 0000	0000 0000
71Ah	SCANHADRL	HADR<7:0>								1111 1111	1111 1111
71Bh	SCANHADRH	HADR<15:8>								1111 1111	1111 1111
71Ch	SCANCON0	EN	SCANGO	BUSY	INVALID	INTM	—	MODE<1:0>		0000 0-00	0000 0-00
71Dh	SCANTRIG					TSEL<3:0>				---- 0000	---- 0000
71Eh	—	Unimplemented								—	—
71Fh	—	Unimplemented								—	—

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, r = reserved. Shaded locations are unimplemented, read as '0'.

Note 1: PIC16F1614/8 only.

2: Unimplemented, read as '1'.

3: PIC16(L)F1614 only.

4: PIC16(L)F1618 only.

5.3 Clock Switching

The system clock source can be switched between external and internal clock sources via software using the System Clock Select (SCS) bits of the OSCCON register. The following clock sources can be selected using the SCS bits:

- Default system oscillator determined by FOSC bits in Configuration Words
- Internal Oscillator Block (INTOSC)

5.3.1 SYSTEM CLOCK SELECT (SCS) BITS

The System Clock Select (SCS) bits of the OSCCON register selects the system clock source that is used for the CPU and peripherals.

- When the SCS bits of the OSCCON register = 00, the system clock source is determined by value of the FOSC<1:0> bits in the Configuration Words.
- When the SCS bits of the OSCCON register = 1x, the system clock source is chosen by the internal oscillator frequency selected by the IRCF<3:0> bits of the OSCCON register. After a Reset, the SCS bits of the OSCCON register are always cleared.

When switching between clock sources, a delay is required to allow the new clock to stabilize. These oscillator delays are shown in Table 5-1.

TABLE 5-1: OSCILLATOR SWITCHING DELAYS

Switch From	Switch To	Frequency	Oscillator Delay
Sleep	LFINTOSC(1) MFINTOSC(1) HFINTOSC(1)	31 kHz 31.25 kHz-500 kHz 31.25 kHz-16 MHz	Oscillator Warm-up Delay (Tiosc st)
Sleep/POR	EC(1)	DC – 32 MHz	2 cycles
LFINTOSC	EC(1)	DC – 32 MHz	1 cycle of each
Any clock source	MFINTOSC(1) HFINTOSC(1)	31.25 kHz-500 kHz 31.25 kHz-16 MHz	2 μs (approx.)
Any clock source	LFINTOSC(1)	31 kHz	1 cycle of each
PLL inactive	PLL active	16-32 MHz	2 ms (approx.)

Note 1: PLL inactive.

20.2 ZCD Logic Output

The ZCD module includes a Status bit, which can be read to determine whether the current source or sink is active. The ZCDxOUT bit of the ZCDxCON register is set when the current sink is active, and cleared when the current source is active. The ZCDxOUT bit is affected by the polarity bit.

20.3 ZCD Logic Polarity

The ZCDxPOL bit of the ZCDxCON register inverts the ZCDxOUT bit relative to the current source and sink output. When the ZCDxPOL bit is set, a ZCDxOUT high indicates that the current source is active, and a low output indicates that the current sink is active.

The ZCDxPOL bit affects the ZCD interrupts. See **Section20.4 "ZCD Interrupts"**.

20.4 ZCD Interrupts

An interrupt will be generated upon a change in the ZCD logic output when the appropriate interrupt enables are set. A rising edge detector and a falling edge detector are present in the ZCD for this purpose.

The ZCDIF bit of the PIR3 register will be set when either edge detector is triggered and its associated enable bit is set. The ZCDxINTP enables rising edge interrupts and the ZCDxINTN bit enables falling edge interrupts. Both are located in the ZCDxCON register.

To fully enable the interrupt, the following bits must be set:

- ZCDIE bit of the PIE3 register
- ZCDxINTP bit of the ZCDxCON register (for a rising edge detection)
- ZCDxINTN bit of the ZCDxCON register (for a falling edge detection)
- PEIE and GIE bits of the INTCON register

Changing the ZCDxPOL bit will cause an interrupt, regardless of the level of the ZCDxEN bit.

The ZCDIF bit of the PIR3 register must be cleared in software as part of the interrupt service. If another edge is detected while this flag is being cleared, the flag will still be set at the end of the sequence.

20.5 Correcting for VCPINV offset

The actual voltage at which the ZCD switches is the reference voltage at the non-inverting input of the ZCD op amp. For external voltage source waveforms other than square waves, this voltage offset from zero causes the zero-cross event to occur either too early or too late. When the waveform is varying relative to VSS, then the zero cross is detected too early as the waveform falls and too late as the waveform rises. When the waveform is varying relative to VDD, then the zero cross is detected too late as the waveform rises and too early as the waveform falls. The actual offset time can be determined for sinusoidal waveforms with the corresponding equations shown in Equation 20-2.

EQUATION 20-2: ZCD EVENT OFFSET

When External Voltage Source is relative to Vss:

$$T_{OFFSET} = \frac{\text{asin}\left(\frac{Vcpinv}{V_{PEAK}}\right)}{2\pi \bullet Freq}$$

When External Voltage Source is relative to VDD:

$$T_{OFFSET} = \frac{\text{asin}\left(\frac{V_{DD}-Vcpinv}{V_{PEAK}}\right)}{2\pi \bullet Freq}$$

This offset time can be compensated for by adding a pull-up or pull-down biasing resistor to the ZCD pin. A pull-up resistor is used when the external voltage source is varying relative to VSS. A pull-down resistor is used when the voltage is varying relative to VDD. The resistor adds a bias to the ZCD pin so that the target external voltage source must go to zero to pull the pin voltage to the VCPINV switching voltage. The pull-up or pull-down value can be determined with the equations shown in Equation 20-3 or Equation 20-4.

EQUATION 20-3: ZCD PULL-UP/DOWN

When External Signal is relative to Vss:

$$R_{PULLUP} = \frac{R_{SERIES}(V_{PULLUP} - Vcpinv)}{Vcpinv}$$

When External Signal is relative to VDD:

$$R_{PULLDOWN} = \frac{R_{SERIES}(Vcpinv)}{(V_{DD} - Vcpinv)}$$

21.2 Register Definitions: Option Register

REGISTER 21-1: OPTION_REG: OPTION REGISTER

R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1
$\overline{\text{WPUEN}}$	INTEDG	TMR0CS	TMR0SE	PSA	PS<2:0>		
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7 **$\overline{\text{WPUEN}}$:** Weak Pull-Up Enable bit
1 = All weak pull-ups are disabled (except $\overline{\text{MCLR}}$, if it is enabled)
0 = Weak pull-ups are enabled by individual WPUx latch values

bit 6 **INTEDG:** Interrupt Edge Select bit
1 = Interrupt on rising edge of INT pin
0 = Interrupt on falling edge of INT pin

bit 5 **TMR0CS:** Timer0 Clock Source Select bit
1 = Transition on T0CKI pin
0 = Internal instruction cycle clock (FOSC/4)

bit 4 **TMR0SE:** Timer0 Source Edge Select bit
1 = Increment on high-to-low transition on T0CKI pin
0 = Increment on low-to-high transition on T0CKI pin

bit 3 **PSA:** Prescaler Assignment bit
1 = Prescaler is not assigned to the Timer0 module
0 = Prescaler is assigned to the Timer0 module

bit 2-0 **PS<2:0>:** Prescaler Rate Select bits

Bit Value	Timer0 Rate
000	1 : 2
001	1 : 4
010	1 : 8
011	1 : 16
100	1 : 32
101	1 : 64
110	1 : 128
111	1 : 256

TABLE 21-1: SUMMARY OF REGISTERS ASSOCIATED WITH TIMER0

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
ADCON2	TRIGSEL<4:0>					—	—	—	197
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	97
OPTION_REG	$\overline{\text{WPUEN}}$	INTEDG	TMR0CS	TMR0SE	PSA	PS<2:0>			222
TMR0	Holding Register for the 8-bit Timer0 Count								220*
TRISA	—	—	TRISA5	TRISA4	—(1)	TRISA2	TRISA1	TRISA0	151

Legend: — = Unimplemented location, read as '0'. Shaded cells are not used by the Timer0 module.
* Page provides register information.
Note 1: Unimplemented, read as '1'.

TABLE 23-5: SUMMARY OF REGISTERS ASSOCIATED WITH TIMER2

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
CCP1CON	EN	—	OUT	FMT	MODE<3:0>				352
CCP2CON	EN	—	OUT	FMT	MODE<3:0>				352
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	97
PIE1	TMR1GIE	ADIE	—	—	—	CCP1IE	TMR2IE	TMR1IE	98
PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSP1IF	CCP1IF	TMR2IF	TMR1IF	103
PR2	Timer2 Module Period Register								235*
TMR2	Holding Register for the 8-bit TMR2 Register								235*
T2CON	ON	CKPS<2:0>			OUTPS<3:0>				254
T2CLKCON	—	—	—	—	CS<3:0>				253
T2RST	—	—	—	—	RSEL<3:0>				256
T2HLT	PSYNC	CKPOL	CKSYNC	MODE<4:0>					255
PR4	Timer4 Module Period Register								235*
TMR4	Holding Register for the 8-bit TMR4 Register								235*
T4CON	ON	CKPS<2:0>			OUTPS<3:0>				254
T4CLKCON	—	—	—	—	CS<3:0>				253
T4RST	—	—	—	—	RSEL<3:0>				256
T4HLT	PSYNC	CKPOL	CKSYNC	MODE<4:0>					255
PR6	Timer6 Module Period Register								235*
TMR6	Holding Register for the 8-bit TMR6 Register								235*
T6CON	ON	CKPS<2:0>			OUTPS<3:0>				254
T6CLKCON	—	—	—	—	—	T6CS<2:0>			253
T6RST	—	—	—	—	RSEL<3:0>				256
T6HLT	PSYNC	CKPOL	CKSYNC	MODE<4:0>					255

Legend: — = unimplemented location, read as '0'. Shaded cells are not used for Timer2 module.

* Page provides register information.

24.2.3 SPI MASTER MODE

The master can initiate the data transfer at any time because it controls the SCK line. The master determines when the slave (Processor 2, Figure 24-5) is to broadcast data by the software protocol.

In Master mode, the data is transmitted/received as soon as the SSPxBUF register is written to. If the SPI is only going to receive, the SDO output could be disabled (programmed as an input). The SSPSR register will continue to shift in the signal present on the SDI pin at the programmed clock rate. As each byte is received, it will be loaded into the SSPxBUF register as if a normal received byte (interrupts and Status bits appropriately set).

The clock polarity is selected by appropriately programming the CKP bit of the SSPxCON1 register and the CKE bit of the SSPxSTAT register. This then, would give waveforms for SPI communication as shown in Figure 24-6, Figure 24-8, Figure 24-9 and Figure 24-10, where the MSB is transmitted first. In Master mode, the SPI clock rate (bit rate) is user programmable to be one of the following:

- FOSC/4 (or TCY)
- FOSC/16 (or 4 * TCY)
- FOSC/64 (or 16 * TCY)
- Timer2 output/2
- FOSC/(4 * (SSPxADD + 1))

Figure 24-6 shows the waveforms for Master mode.

When the CKE bit is set, the SDO data is valid before there is a clock edge on SCK. The change of the input sample is shown based on the state of the SMP bit. The time when the SSPxBUF is loaded with the received data is shown.

Note:	In Master mode the clock signal output to the SCK pin is also the clock signal input to the peripheral. The pin selected for output with the RxyPPS register must also be selected as the peripheral input with the SSPCLKPPS register.

24.3 I²C MODE OVERVIEW

The Inter-Integrated Circuit (I^2C) bus is a multi-master serial data communication bus. Devices communicate in a master/slave environment where the master devices initiate the communication. A slave device is controlled through addressing.

The I^2C bus specifies two signal connections:

- Serial Clock (SCL)
- Serial Data (SDA)

Figure 24-11 shows the block diagram of the MSSP module when operating in I^2C mode.

Both the SCL and SDA connections are bidirectional open-drain lines, each requiring pull-up resistors for the supply voltage. Pulling the line to ground is considered a logical zero and letting the line float is considered a logical one.

Figure 24-11 shows a typical connection between two processors configured as master and slave devices.

The I^2C bus can operate with one or more master devices and one or more slave devices.

There are four potential modes of operation for a given device:

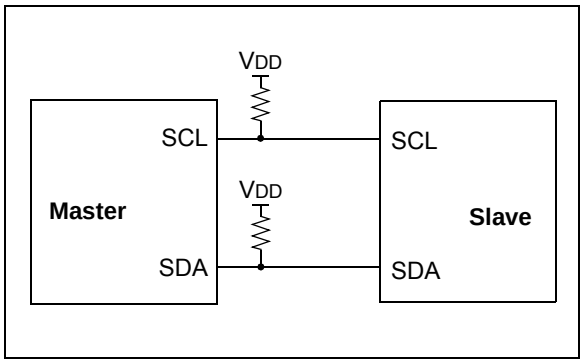
- Master Transmit mode
  (master is transmitting data to a slave)
- Master Receive mode
  (master is receiving data from a slave)
- Slave Transmit mode
  (slave is transmitting data to a master)
- Slave Receive mode
  (slave is receiving data from the master)

To begin communication, a master device starts out in Master Transmit mode. The master device sends out a Start bit followed by the address byte of the slave it intends to communicate with. This is followed by a single Read/Write bit, which determines whether the master intends to transmit to or receive data from the slave device.

If the requested slave exists on the bus, it will respond with an Acknowledge bit, otherwise known as an $\overline{ACK}$. The master then continues in either Transmit mode or Receive mode and the slave continues in the complement, either in Receive mode or Transmit mode, respectively.

A Start bit is indicated by a high-to-low transition of the SDA line while the SCL line is held high. Address and data bytes are sent out, Most Significant bit (MSb) first. The Read/Write bit is sent out as a logical one when the master intends to read data from the slave, and is sent out as a logical zero when it intends to write data to the slave.

FIGURE 24-11: I²C MASTER/ SLAVE CONNECTION



The Acknowledge bit ($\overline{ACK}$) is an active-low signal, which holds the SDA line low to indicate to the transmitter that the slave device has received the transmitted data and is ready to receive more.

The transition of a data bit is always performed while the SCL line is held low. Transitions that occur while the SCL line is held high are used to indicate Start and Stop bits.

If the master intends to write to the slave, then it repeatedly sends out a byte of data, with the slave responding after each byte with an $\overline{ACK}$ bit. In this example, the master device is in Master Transmit mode and the slave is in Slave Receive mode.

If the master intends to read from the slave, then it repeatedly receives a byte of data from the slave, and responds after each byte with an $\overline{ACK}$ bit. In this example, the master device is in Master Receive mode and the slave is Slave Transmit mode.

On the last byte of data communicated, the master device may end the transmission by sending a Stop bit. If the master device is in Receive mode, it sends the Stop bit in place of the last $\overline{ACK}$ bit. A Stop bit is indicated by a low-to-high transition of the SDA line while the SCL line is held high.

In some cases, the master may want to maintain control of the bus and re-initiate another transmission. If so, the master device may send another Start bit in place of the Stop bit or last $\overline{ACK}$ bit when it is in receive mode.

The I^2C bus specifies three message protocols;

- Single message where a master writes data to a slave.
- Single message where a master reads data from a slave.
- Combined message where a master initiates a minimum of two writes, or two reads, or a combination of writes and reads, to one or more slaves.

24.6.7 I²C MASTER MODE RECEPTION

Master mode reception (Figure 24-29) is enabled by programming the Receive Enable bit, RCEN bit of the SSPxCON2 register.

Note:	The MSSP module must be in an Idle state before the RCEN bit is set or the RCEN bit will be disregarded.

The Baud Rate Generator begins counting and on each rollover, the state of the SCL pin changes (high-to-low/ low-to-high) and data is shifted into the SSPSR. After the falling edge of the eighth clock, the receive enable flag is automatically cleared, the contents of the SSPSR are loaded into the SSPxBUF, the BF flag bit is set, the SSPxIF flag bit is set and the Baud Rate Generator is suspended from counting, holding SCL low. The MSSP is now in Idle state awaiting the next command. When the buffer is read by the CPU, the BF flag bit is automatically cleared. The user can then send an Acknowledge bit at the end of reception by setting the Acknowledge Sequence Enable, ACKEN bit of the SSPxCON2 register.

24.6.7.1 BF Status Flag

In receive operation, the BF bit is set when an address or data byte is loaded into SSPxBUF from SSPSR. It is cleared when the SSPxBUF register is read.

24.6.7.2 SSPOV Status Flag

In receive operation, the SSPOV bit is set when eight bits are received into the SSPSR and the BF flag bit is already set from a previous reception.

24.6.7.3 WCOL Status Flag

If the user writes the SSPxBUF when a receive is already in progress (i.e., SSPSR is still shifting in a data byte), the WCOL bit is set and the contents of the buffer are unchanged (the write does not occur).

24.6.7.4 Typical Receive Sequence:

1. The user generates a Start condition by setting the SEN bit of the SSPxCON2 register.
2. SSPxIF is set by hardware on completion of the Start.
3. SSPxIF is cleared by software.
4. User writes SSPxBUF with the slave address to transmit and the R/$\overline{W}$ bit set.
5. Address is shifted out the SDA pin until all eight bits are transmitted. Transmission begins as soon as SSPxBUF is written to.
6. The MSSP module shifts in the $\overline{ACK}$ bit from the slave device and writes its value into the ACKSTAT bit of the SSPxCON2 register.
7. The MSSP module generates an interrupt at the end of the ninth clock cycle by setting the SSPxIF bit.
8. User sets the RCEN bit of the SSPxCON2 register and the master clocks in a byte from the slave.
9. After the eighth falling edge of SCL, SSPxIF and BF are set.
10. Master clears SSPxIF and reads the received byte from SSPxBUF, clears BF.
11. Master sets $\overline{ACK}$ value sent to slave in ACKDT bit of the SSPxCON2 register and initiates the $\overline{ACK}$ by setting the ACKEN bit.
12. Master's $\overline{ACK}$ is clocked out to the slave and SSPxIF is set.
13. User clears SSPxIF.
14. Steps 8-13 are repeated for each received byte from the slave.
15. Master sends a not $\overline{ACK}$ or Stop to end communication.

25.1.2 EUSART ASYNCHRONOUS RECEIVER

The Asynchronous mode is typically used in RS-232 systems. The receiver block diagram is shown in Figure 25-2. The data is received on the RX/DT pin and drives the data recovery block. The data recovery block is actually a high-speed shifter operating at 16 times the baud rate, whereas the serial Receive Shift Register (RSR) operates at the bit rate. When all eight or nine bits of the character have been shifted in, they are immediately transferred to a two character First-In-First-Out (FIFO) memory. The FIFO buffering allows reception of two complete characters and the start of a third character before software must start servicing the EUSART receiver. The FIFO and RSR registers are not directly accessible by software. Access to the received data is via the RCxREG register.

25.1.2.1 Enabling the Receiver

The EUSART receiver is enabled for asynchronous operation by configuring the following three control bits:

- CREN = 1
- SYNC = 0
- SPEN = 1

All other EUSART control bits are assumed to be in their default state.

Setting the CREN bit of the RCxSTA register enables the receiver circuitry of the EUSART. Clearing the SYNC bit of the TXxSTA register configures the EUSART for asynchronous operation. Setting the SPEN bit of the RCxSTA register enables the EUSART. The programmer must set the corresponding TRIS bit to configure the RX/DT I/O pin as an input.

Note:	If the RX/DT function is on an analog pin, the corresponding ANSEL bit must be cleared for the receiver to function.

25.1.2.2 Receiving Data

The receiver data recovery circuit initiates character reception on the falling edge of the first bit. The first bit, also known as the Start bit, is always a zero. The data recovery circuit counts one-half bit time to the center of the Start bit and verifies that the bit is still a zero. If it is not a zero then the data recovery circuit aborts character reception, without generating an error, and resumes looking for the falling edge of the Start bit. If the Start bit zero verification succeeds then the data recovery circuit counts a full bit time to the center of the next bit. The bit is then sampled by a majority detect circuit and the resulting '0' or '1' is shifted into the RSR. This repeats until all data bits have been sampled and shifted into the RSR. One final bit time is measured and the level sampled. This is the Stop bit, which is always a '1'. If the data recovery circuit samples a '0' in the Stop bit position then a framing error is set for this character, otherwise the framing error is cleared for this character. See **Section 25.1.2.4 "Receive Framing Error"** for more information on framing errors.

Immediately after all data bits and the Stop bit have been received, the character in the RSR is transferred to the EUSART receive FIFO and the RCIF interrupt flag bit of the PIR1 register is set. The top character in the FIFO is transferred out of the FIFO by reading the RCxREG register.

Note:	If the receive FIFO is overrun, no additional characters will be received until the overrun condition is cleared. See **Section 25.1.2.5 "Receive Overrun Error"** for more information on overrun errors.

25.1.2.3 Receive Interrupts

The RCIF interrupt flag bit of the PIR1 register is set whenever the EUSART receiver is enabled and there is an unread character in the receive FIFO. The RCIF interrupt flag bit is read-only, it cannot be set or cleared by software.

RCIF interrupts are enabled by setting all of the following bits:

- RCIE, Interrupt Enable bit of the PIE1 register
- PEIE, Peripheral Interrupt Enable bit of the INTCON register
- GIE, Global Interrupt Enable bit of the INTCON register

The RCIF interrupt flag bit will be set when there is an unread character in the FIFO, regardless of the state of interrupt enable bits.

25.1.2.4 Receive Framing Error

Each character in the receive FIFO buffer has a corresponding framing error Status bit. A framing error indicates that a Stop bit was not seen at the expected time. The framing error status is accessed via the FERR bit of the RCxSTA register. The FERR bit represents the status of the top unread character in the receive FIFO. Therefore, the FERR bit must be read before reading the RCxREG.

The FERR bit is read-only and only applies to the top unread character in the receive FIFO. A framing error (FERR = 1) does not preclude reception of additional characters. It is not necessary to clear the FERR bit. Reading the next character from the FIFO buffer will advance the FIFO to the next character and the next corresponding framing error.

The FERR bit can be forced clear by clearing the SPEN bit of the RCxSTA register which resets the EUSART. Clearing the CREN bit of the RCxSTA register does not affect the FERR bit. A framing error by itself does not generate an interrupt.

Note:	If all receive characters in the receive FIFO have framing errors, repeated reads of the RCxREG will not clear the FERR bit.

25.1.2.5 Receive Overrun Error

The receive FIFO buffer can hold two characters. An overrun error will be generated if a third character, in its entirety, is received before the FIFO is accessed. When this happens the OERR bit of the RCxSTA register is set. The characters already in the FIFO buffer can be read but no additional characters will be received until the error is cleared. The error must be cleared by either clearing the CREN bit of the RCxSTA register or by resetting the EUSART by clearing the SPEN bit of the RCxSTA register.

25.1.2.6 Receiving 9-Bit Characters

The EUSART supports 9-bit character reception. When the RX9 bit of the RCxSTA register is set, the EUSART will shift nine bits into the RSR for each character received. The RX9D bit of the RCxSTA register is the ninth and Most Significant data bit of the top unread character in the receive FIFO. When reading 9-bit data from the receive FIFO buffer, the RX9D data bit must be read before reading the eight Least Significant bits from the RCxREG.

25.1.2.7 Address Detection

A special Address Detection mode is available for use when multiple receivers share the same transmission line, such as in RS-485 systems. Address detection is enabled by setting the ADDEN bit of the RCxSTA register.

Address detection requires 9-bit character reception. When address detection is enabled, only characters with the ninth data bit set will be transferred to the receive FIFO buffer, thereby setting the RCIF interrupt bit. All other characters will be ignored.

Upon receiving an address character, user software determines if the address matches its own. Upon address match, user software must disable address detection by clearing the ADDEN bit before the next Stop bit occurs. When user software detects the end of the message, determined by the message protocol used, software places the receiver back into the Address Detection mode by setting the ADDEN bit.

TABLE 25-9: SUMMARY OF REGISTERS ASSOCIATED WITH SYNCHRONOUS SLAVE TRANSMISSION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
ANSELA	—	—	—	ANSA4	—	ANSA2	ANSA1	ANSA0	152
ANSELB(1)	—	—	ANSB5	ANSB4	—	—	—	—	159
ANSELC	ANSC7(1)	ANSC6(1)	—	—	ANSC3	ANSC2	ANSC1	ANSC0	166
BAUD1CON	ABDOVF	RCIDL	—	SCKP	BRG16	—	WUE	ABDEN	323
CKPPS	—	—	—	CKPPS<4:0>					174, 172
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	97
PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSP1IE	CCP1IE	TMR2IE	TMR1IE	98
PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSP1IF	CCP1IF	TMR2IF	TMR1IF	103
RC1STA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	322
RXPPS	—	—	—	RXPPS<4:0>					174, 172
RxyPPS	—	—	—	RxyPPS<4:0>					172
TRISA	—	—	TRISA5	TRISA4	—(2)	TRISA2	TRISA1	TRISA0	151
TRISB(1)	TRISB7	TRISB6	TRISB5	TRISB4	—	—	—	—	158
TRISC	TRISC7(1)	TRISC6(1)	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	165
TX1REG	EUSART Transmit Data Register								313*
TX1STA	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	321

Legend: — = unimplemented location, read as '0'. Shaded cells are not used for synchronous slave transmission.

* Page provides register information.

Note 1: PIC16(L)F1618 only.

2: Unimplemented, read as '1'.

26.2 Compare Mode

The Compare mode function described in this section is available and identical for all CCP modules.

Compare mode makes use of the 16-bit Timer1 resource. The 16-bit value of the CCPRxH:CCPRxL register pair is constantly compared against the 16-bit value of the TMR1H:TMR1L register pair. When a match occurs, one of the following events can occur:

- Toggle the CCPx output
- Set the CCPx output
- Clear the CCPx output
- Pulse the CCPx output
- Generate a Software Interrupt
- Optionally Reset TMR1

The action on the pin is based on the value of the MODE<3:0> control bits of the CCPxCON register. At the same time, the interrupt flag CCPxIF bit is set.

All Compare modes can generate an interrupt.

Figure 26-2 shows a simplified diagram of the compare operation.

26.2.1 CCPx PIN CONFIGURATION

The user must configure the CCPx pin as an output by clearing the associated TRIS bit.

FIGURE 26-2: COMPARE MODE OPERATION BLOCK DIAGRAM

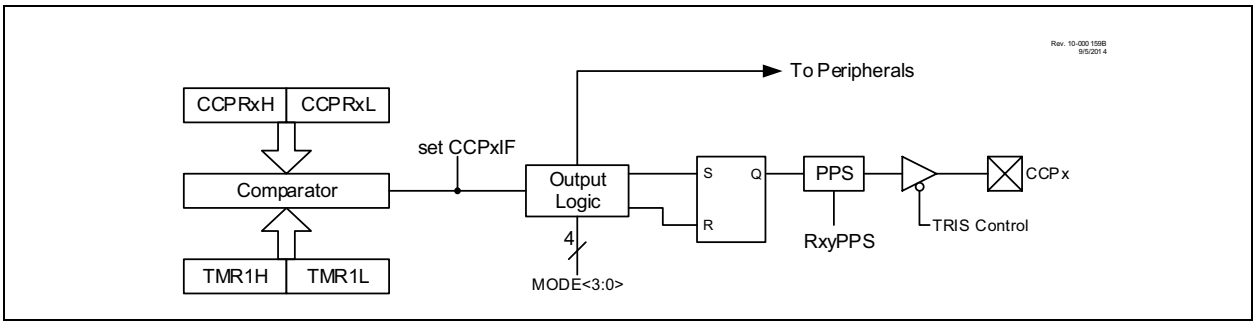


TABLE 26-1: EXAMPLE PWM FREQUENCIES AND RESOLUTIONS (FOSC = 20 MHz)

PWM Frequency	1.22 kHz	4.88 kHz	19.53 kHz	78.12 kHz	156.3 kHz	208.3 kHz
Timer Prescale	16	4	1	1	1	1
PR2 Value	0xFF	0xFF	0xFF	0x3F	0x1F	0x17
Maximum Resolution (bits)	10	10	10	8	7	6

TABLE 26-2: EXAMPLE PWM FREQUENCIES AND RESOLUTIONS (FOSC = 8 MHz)

PWM Frequency	1.22 kHz	4.90 kHz	19.61 kHz	76.92 kHz	153.85 kHz	200.0 kHz
Timer Prescale	16	4	1	1	1	1
PR2 Value	0x65	0x65	0x65	0x19	0x0C	0x09
Maximum Resolution (bits)	8	8	8	6	5	5

26.4.5 CHANGES IN SYSTEM CLOCK FREQUENCY

The PWM frequency is derived from the system clock frequency. Any changes in the system clock frequency will result in changes to the PWM frequency. See **Section5.0 “Oscillator Module”** for additional details.

26.4.6 EFFECTS OF RESET

Any Reset will force all ports to Input mode and the CCP registers to their Reset states.

26.4.7 PWM OUTPUT

The output of the CCP in PWM mode is the PWM signal generated by the module and described above. This output is available as an input signal to the CWG, as an auto-conversion trigger for the ADC, as an external Reset signal for the TMR2 modules, as a window input to the SMT, and as an input to the CLC module. In addition, the CCPx pin output can be mapped to output pins through the use of PPS (see **Section13.2 “PPS Outputs”**).

FIGURE 30-4: GATED TIMER MODE REPEAT ACQUISITION TIMING DIAGRAM

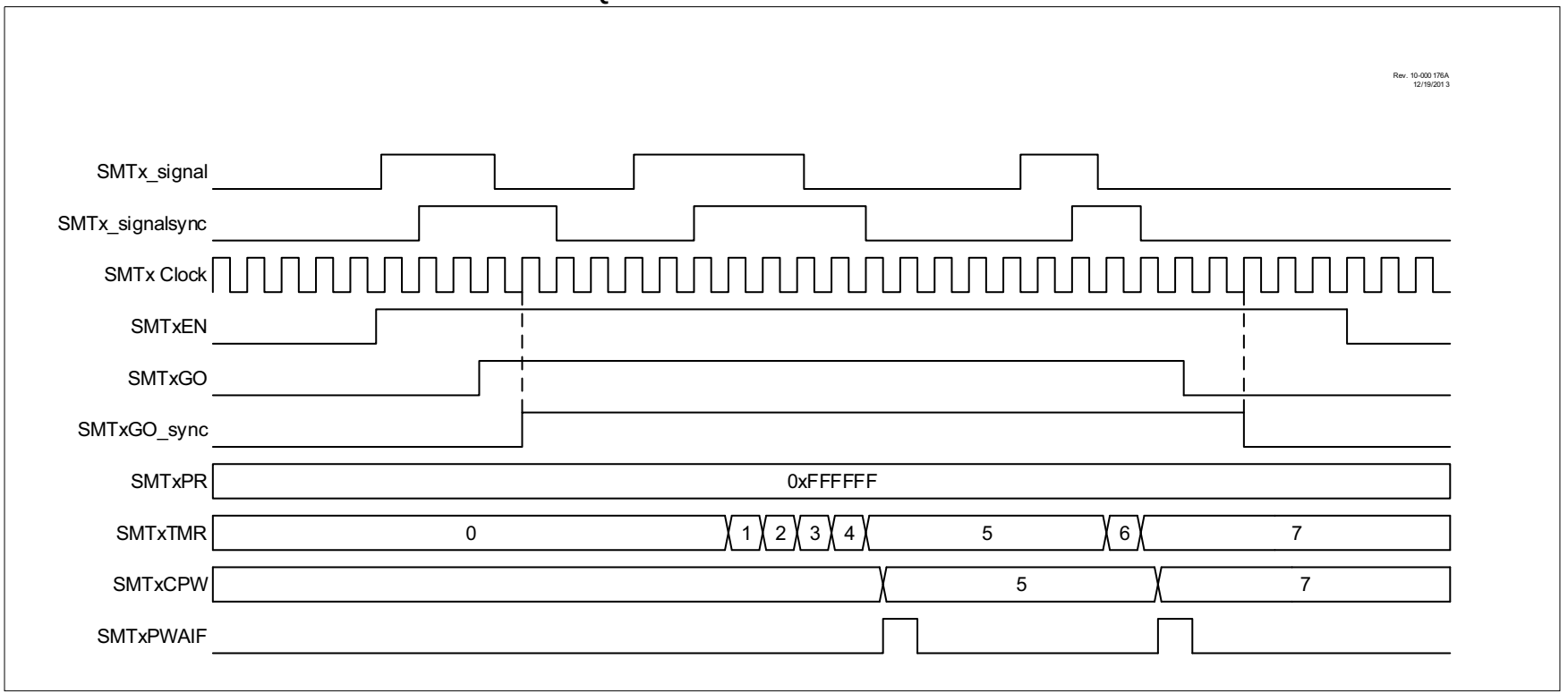


TABLE 30-3: SUMMARY OF REGISTERS ASSOCIATED WITH SMTx

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
PIE4	SCANIE	CRCIE	SMT2PWAIE	SMT2PRAIE	SMT2IE	SMT1PWAIE	SMT1PRAIE	SMT1IE	101
PIR4	SCANIF	CRCIF	SMT2PWAIF	SMT2PRAIF	SMT2IF	SMT1PWAIF	SMT1PRAIF	SMT1IF	106
SMT1CLK	—	—	—	—	—	CSEL<2:0>			438
SMT1CON0	EN	—	STP	WPOL	SPOL	CPOL	SMT1PS<1:0>		435
SMT1CON1	SMT1GO	REPEAT	—	—	MODE<3:0>				436
SMT1CPRH	SMT1CPR<15:8>								444
SMT1CPRL	SMT1CPR<7:0>								444
SMT1CPRU	SMT1CPR<23:16>								444
SMT1CPWH	SMT1CPW<15:8>								445
SMT1CPWL	SMT1CPW<7:0>								445
SMT1CPWU	SMT1CPW<23:16>								445
SMT1PRH	SMT1PR<15:8>								446
SMT1PRL	SMT1PR<7:0>								446
SMT1PRU	SMT1PR<23:16>								446
SMT1SIG	—	—	—	SSEL<4:0>					441
SMT1STAT	CPRUP	CPWUP	RST	—	—	TS	WS	AS	437
SMT1TMRH	SMT1TMR<15:8>								443
SMT1TMRL	SMT1TMR<7:0>								443
SMT1TMRU	SMT1TMR<23:16>								443
SMT1WIN	—	—	—	WSEL<4:0>					439
SMT2CLK	—	—	—	—	—	CSEL<2:0>			438
SMT2CON0	EN	—	STP	WPOL	SPOL	CPOL	SMT2PS<1:0>		435
SMT2CON1	SMT2GO	REPEAT	—	—	MODE<3:0>				436
SMT2CPRH	SMT2CPR<15:8>								444
SMT2CPRL	SMT2CPR<7:0>								444
SMT2CPRU	SMT2CPR<23:16>								444
SMT2CPWH	SMT2CPW<15:8>								445
SMT2CPWL	SMT2CPW<7:0>								445
SMT2CPWU	SMT2CPW<23:16>								445
SMT2PRH	SMT2PR<15:8>								446
SMT2PRL	SMT2PR<7:0>								446
SMT2PRU	SMT2PR<23:16>								446
SMT2SIG	—	—	—	—	—	SSEL<2:0>			441
SMT2STAT	CPRUP	CPWUP	RST	—	—	TS	WS	AS	437
SMT2TMRH	SMT2TMR<15:8>								443
SMT2TMRL	SMT2TMR<7:0>								443
SMT2TMRU	SMT2TMR<23:16>								443
SMT2WIN	—	—	—	WSEL<4:0>					438

Legend: x = unknown, u = unchanged, – = unimplemented read as '0', q = value depends on condition. Shaded cells are not used for SMTx module.

31.2.1 SINGLE-PULSE MODE

The operation of Single-Pulse mode is illustrated in Figure 31-1. The calculations on the input signal are done in a few distinct steps. First, there is a divider that divides the module clock by the ATxRES register pair and uses the resulting signal to increment a period counter. This operation is expressed by Equation 31-2. This equation differs slightly from that of Equation 31-1 because the counters include the count of zero. To compensate for this, the number written to the resolution register, ATxRES, must be one less than the desired resolution.

EQUATION 31-2:

$$ATxPER = \frac{\frac{F(ATxclk)}{F(ATxsig)}}{(ATxRES + 1)}$$

Variables in Equation 31-2 are as follows:

- ATxPER is the value of the period counter latched by the input signal.
- ATxRES is the user-specified resolution. The phase counter will count up to this value.
- F(ATxclk) is the ATx clock frequency.
- F(ATxsig) is the input signal frequency.

The second step in the angular timer's operation is the creation of the phase clock, which is also illustrated in Figure 31-1. The input clock is divided by the ATxPER value, latched-in during the previous step, and the resulting signal is used to increment the phase counter. This signal also is used as the phase clock output, and for setting the PHSIF interrupt flag bit of the ATxIR0 register. The result is that the phase counter counts from zero to a final value expressed in Equation 31-3, outputting a pulse each time the counter increments. The value of the phase counter can be accessed by software by reading the ATxPHS register pair. However, because of the synchronization required, in order for reads of this register pair to be accurate, the instruction clock (FOSC/4) needs to be at least 3x the ATx_phsclk output frequency.

EQUATION 31-3:

$$ATxPHS(final) = \frac{\left(\frac{F(ATxclk)}{F(ATxsig)}\right)}{(ATxPER + 1)}$$

The variables in Equation 31-3 are as follows:

- ATxPHS(final) is the maximum value that the phase counter will reach before being reset by the input signal. As noted in Equation 31-1, this will equal ATxRES.
- ATxPER is the maximum value of the period counter.
- F(ATxclk) is the ATx clock frequency.
- F(ATxsig) is the input signal frequency.

Notice that the division is ATxPER + 1. Ideally, this would be just ATxPER but the divider includes zero in the count. In most applications, ATxPER is a large number so the error introduced by adding one is negligible.

ATxPHS counting from 0 to ATxRES is useful when the input signal represents a rotation (for example, a motor or A/C mains). In this case, the input signal is understood to provide a period pulse every 360 degrees. Since the phase clock equally divides the signal period into a number of intervals determined by the ATxRES register pair, each pulse on the phase clock output marks a fixed phase angle in that rotation, as expressed by Equation 31-4.

EQUATION 31-4:

$$AngleResolution = \frac{360degrees}{ATxRES + 1}$$

ATxRES can then be used with the instantaneous value of the ATxPHS register pair to get the instantaneous angle of the rotation using Equation 31-5.

EQUATION 31-5:

$$Angle = 360degrees \bullet \frac{ATxPHS}{ATxRES + 1}$$

31.2.2 MULTI-PULSE MODE

The operation of Multi-Pulse mode is illustrated in Figure 31-3. The calculations on the input signal are similar to those in Single-Pulse mode, with the primary difference relating to when the ATxPHS register pair is reset.

The period counter is latched into the ATxPER register pair and reset on every input pulse except the pulse immediately following a missing pulse. The first active pulse following a missing pulse triggers all of the following:

- Period clock output
- PERIF interrupt
- Phase counter reset

The result is a period clock output that has a period length equal to the time between missing pulses (e.g., a missing tooth in a gear). This leads to a significantly different relation between ATxRES and the maximum phase count, ATxPHS, as shown in Equation 31-6.

EQUATION 31-6:

$$ATxPHS(final) = ATxRES\left(\frac{MissP}{PulseP}\right)$$

REGISTER 31-11: ATxPHSH: ANGULAR TIMER PHASE COUNTER HIGH REGISTER

U-0	U-0	U-0	U-0	U-0	U-0	R-x/x	R-x/x
—	—	—	—	—	—	PHS<9:8>	
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set	‘0’ = Bit is cleared	q = Value depends on condition

bit 7-2 **Unimplemented:** Read as ‘0’

bit 1-0 **PHS<9:8>:** Most Significant bits of ATxPHS. ATxPHS is the instantaneous value of the phase counter.

REGISTER 31-12: ATxPHSL: ANGULAR TIMER PHASE COUNTER LOW REGISTER

R-x/x	R-x/x	R-x/x	R-x/x	R-x/x	R-x/x	R-x/x	R-x/x
PHS<7:0>							
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set	‘0’ = Bit is cleared	q = Value depends on condition

bit 7-0 **PHS<7:0>:** Least Significant bits of ATxPHS. ATxPHS is the instantaneous value of the phase counter.

REGISTER 32-6: PIDxK1H: PID K1 HIGH REGISTER

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
K1<15:8>							
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set	‘0’ = Bit is cleared	q = Value depends on condition

bit 7-0 **K1<15:8>:** K1 upper eight bits. K1 is the 16-bit user-controlled coefficient calculated from Kp + Ki + Kd

REGISTER 32-7: PIDxK1L: PID K1 LOW REGISTER

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
K1<7:0>							
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set	‘0’ = Bit is cleared	q = Value depends on condition

bit 7-0 **K1<7:0>:** K1 lower eight bits. K1 is the 16-bit user-controlled coefficient calculated from Kp + Ki + Kd

REGISTER 32-8: PIDxK2H: PID K2 HIGH REGISTER

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
K2<15:8>							
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set	‘0’ = Bit is cleared	q = Value depends on condition

bit 7-0 K2<15:8>: K2 upper eight bits. K2 is the 16-bit user-controlled coefficient calculated from -(Kp + 2Kd)

REGISTER 32-9: PIDxK2L: PID K2 LOW REGISTER

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
K2<7:0>							
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set	‘0’ = Bit is cleared	q = Value depends on condition

bit 7-0 **K2<7:0>:** K2 lower eight bits. K2 is the 16-bit user-controlled coefficient calculated from -(Kp + 2Kd)