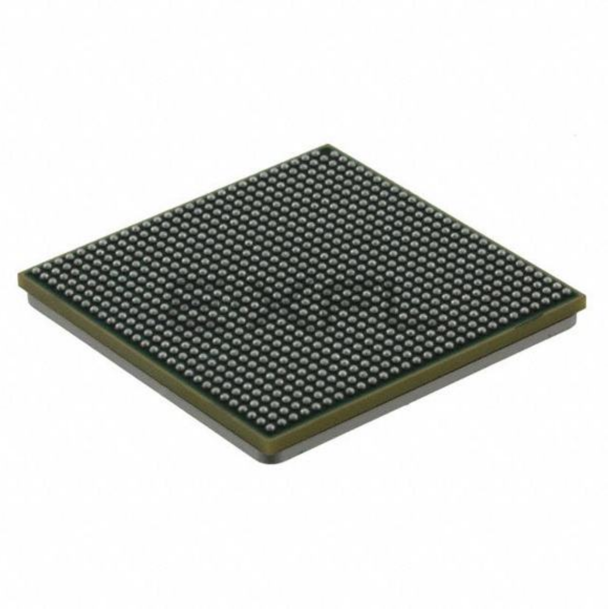
NXP USA Inc. - MPC8541ECPXAJD Datasheet



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Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Obsolete
Core Processor	PowerPC e500
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	533MHz
Co-Processors/DSP	Security; SEC
RAM Controllers	DDR, SDRAM
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100/1000Mbps (2)
SATA	-
USB	-
Voltage - I/O	2.5V, 3.3V
Operating Temperature	-40°C ~ 105°C (TA)
Security Features	Cryptography, Random Number Generator
Package / Case	783-BBGA, FCBGA
Supplier Device Package	783-FCPBGA (29x29)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mpc8541ecpxajd

2.1.1 Absolute Maximum Ratings

Table 1 provides the absolute maximum ratings.

Table 1. Absolute Maximum Ratings [1]

Characteristic		Symbol	Max Value	Unit	Notes
Core supply voltage		V_{DD}	–0.3 to 1.32 0.3 to 1.43 (for 1 GHz only)	V	
PLL supply voltage		AV_{DD}	–0.3 to 1.32 0.3 to 1.43 (for 1 GHz only)	V	
DDR DRAM I/O voltage		GV_{DD}	–0.3 to 3.63	V	
Three-speed Ethernet I/O, MII management voltage		LV_{DD}	–0.3 to 3.63 –0.3 to 2.75	V	
CPM, PCI, local bus, DUART, system control and power management, I^2C, and JTAG I/O voltage		OV_{DD}	–0.3 to 3.63	V	3
Input voltage	DDR DRAM signals	MV_{IN}	–0.3 to (GV_{DD} + 0.3)	V	2, 5
	DDR DRAM reference	MV_{REF}	–0.3 to (GV_{DD} + 0.3)	V	2, 5
	Three-speed Ethernet signals	LV_{IN}	–0.3 to (LV_{DD} + 0.3)	V	4, 5
	CPM, Local bus, DUART, SYSCLK, system control and power management, I^2C, and JTAG signals	OV_{IN}	–0.3 to (OV_{DD} + 0.3)1	V	5
	PCI	OV_{IN}	–0.3 to (OV_{DD} + 0.3)	V	6
Storage temperature range		T_{STG}	–55 to 150	°C	

Notes:
1. Functional and tested operating conditions are given in Table 2. Absolute maximum ratings are stress ratings only, and functional operation at the maximums is not guaranteed. Stresses beyond those listed may affect device reliability or cause permanent damage to the device.
2. **Caution:** MV_{IN} must not exceed GV_{DD} by more than 0.3 V. This limit may be exceeded for a maximum of 20 ms during power-on reset and power-down sequences.
3. **Caution:** OV_{IN} must not exceed OV_{DD} by more than 0.3 V. This limit may be exceeded for a maximum of 20 ms during power-on reset and power-down sequences.
4. **Caution:** LV_{IN} must not exceed LV_{DD} by more than 0.3 V. This limit may be exceeded for a maximum of 20 ms during power-on reset and power-down sequences.
5. (M,L,O)V_{IN} and MV_{REF} may overshoot/undershoot to a voltage and for a maximum duration as shown in Figure 2.
6. OV_{IN} on the PCI interface may overshoot/undershoot according to the PCI Electrical Specification for 3.3-V operation, as shown in Figure 3.

2.1.2 Power Sequencing

The MPC8541Erequires its power rails to be applied in a specific sequence in order to ensure proper device operation. These requirements are as follows for power up:

1. V_{DD}, AV_{DDn}
2. GV_{DD}, LV_{DD}, OV_{DD} (I/O supplies)

Items on the same line have no ordering requirement with respect to one another. Items on separate lines must be ordered sequentially such that voltage rails on a previous step must reach 90 percent of their value before the voltage rails on the current step reach ten percent of theirs.

NOTE

If the items on line 2 must precede items on line 1, please ensure that the delay does not exceed 500 ms and the power sequence is not done greater than once per day in production environment.

NOTE

From a system standpoint, if the I/O power supplies ramp prior to the V_{DD} core supply, the I/Os on the MPC8541E may drive a logic one or zero during power-up.

2.1.3 Recommended Operating Conditions

Table 2 provides the recommended operating conditions for the MPC8541E. Note that the values in Table 2 are the recommended and tested operating conditions. Proper device operation outside of these conditions is not guaranteed.

Table 2. Recommended Operating Conditions

Characteristic		Symbol	Recommended Value	Unit
Core supply voltage		V_{DD}	1.2 V ± 60 mV 1.3 V± 50 mV (for 1 GHz only)	V
PLL supply voltage		AV_{DD}	1.2 V ± 60 mV 1.3 V ± 50 mV (for 1 GHz only)	V
DDR DRAM I/O voltage		GV_{DD}	2.5 V ± 125 mV	V
Three-speed Ethernet I/O voltage		LV_{DD}	3.3 V ± 165 mV 2.5 V ± 125 mV	V
PCI, local bus, DUART, system control and power management, I^2C, and JTAG I/O voltage		OV_{DD}	3.3 V ± 165 mV	V
Input voltage	DDR DRAM signals	MV_{IN}	GND to GV_{DD}	V
	DDR DRAM reference	MV_{REF}	GND to GV_{DD}	V
	Three-speed Ethernet signals	LV_{IN}	GND to LV_{DD}	V
	PCI, local bus, DUART, SYSCLK, system control and power management, I^2C, and JTAG signals	OV_{IN}	GND to OV_{DD}	V
Die-junction Temperature		T_j	0 to 105	°C

Table 5. Typical I/O Power Dissipation

Interface	Parameters	GV_{DD} (2.5 V)	OV_{DD} (3.3 V)	LV_{DD} (3.3 V)	LV_{DD} (2.5 V)	Unit	Comments
DDR I/O	CCB = 200 MHz	0.46	—	—	—	W	—
	CCB = 266 MHz	0.59	—	—	—	W	—
	CCB = 300 MHz	0.66	—	—	—	W	—
	CCB = 333 MHz	0.73	—	—	—	W	—
PCI I/O	64b, 66 MHz	—	0.14	—	—	W	—
	64b, 33 MHz	—	0.08	—	—	W	—
	32b, 66 MHz	—	0.07	—	—	W	Multiply by 2 if using two 32b ports
	32b, 33 MHz	—	0.04	—	—	W	
Local Bus I/O	32b, 167 MHz	—	0.30	—	—	W	—
	32b, 133 MHz	—	0.24	—	—	W	—
	32b, 83 MHz	—	0.16	—	—	W	—
	32b, 66 MHz	—	0.13	—	—	W	—
	32b, 33 MHz	—	0.07	—	—	W	—
TSEC I/O	MII	—	—	0.01	—	W	Multiply by number of interfaces used.
	GMII or TBI	—	—	0.07	—	W	
	RGMII or RTBI	—	—	—	0.04	W	
CPM - FCC	MII	—	0.015	—	—	W	—
	RMII	—	0.013	—	—	W	—
	HDLC 16 Mbps	—	0.009	—	—	W	—

6.2 DDR SDRAM AC Electrical Characteristics

This section provides the AC electrical characteristics for the DDR SDRAM interface.

6.2.1 DDR SDRAM Input AC Timing Specifications

Table 13 provides the input AC timing specifications for the DDR SDRAM interface.

Table 13. DDR SDRAM Input AC Timing Specifications

At recommended operating conditions with GV_{DD} of 2.5 V ± 5%.

Parameter	Symbol	Min	Max	Unit	Notes
AC input low voltage	V_{IL}	—	$MV_{REF} – 0.31$	V	—
AC input high voltage	V_{IH}	$MV_{REF} + 0.31$	$GV_{DD} + 0.3$	V	—
MDQS—MDQ/MECC input skew per byte For DDR = 333 MHz For DDR ≤ 266 MHz	t_{DISKEW}	—	 750 1125	ps	1

Note:

1. Maximum possible skew between a data strobe (MDQS[n]) and any corresponding bit of data (MDQ[8n + {0...7}] if 0 <= n <= 7) or ECC (MECC[{0...7}] if n = 8).

6.2.2 DDR SDRAM Output AC Timing Specifications

Table 14 and Table 15 provide the output AC timing specifications and measurement conditions for the DDR SDRAM interface.

Table 14. DDR SDRAM Output AC Timing Specifications for Source Synchronous Mode

At recommended operating conditions with GV_{DD} of 2.5 V ± 5%.

Parameter	Symbol [1]	Min	Max	Unit	Notes
MCK[n] cycle time, (MCK[n]/$\overline{\text{MCK}}$[n] crossing)	t_{MCK}	6	10	ns	2
Skew between any MCK to ADDR/CMD 333 MHz 266 MHz 200 MHz	t_{AOSKEW}	 –1000 –1100 –1200	 200 300 400	ps	3
ADDR/CMD output setup with respect to MCK 333 MHz 266 MHz 200 MHz	t_{DDKHAS}	 2.8 3.45 4.6	—	ns	4
ADDR/CMD output hold with respect to MCK 333 MHz 266 MHz 200 MHz	t_{DDKHAX}	 2.0 2.65 3.8	—	ns	4
$\overline{\text{MCS}}$(n) output setup with respect to MCK 333 MHz 266 MHz 200 MHz	t_{DDKHCS}	 2.8 3.45 4.6	—	ns	4

Figure 4 shows the DDR SDRAM output timing for address skew with respect to any MCK.

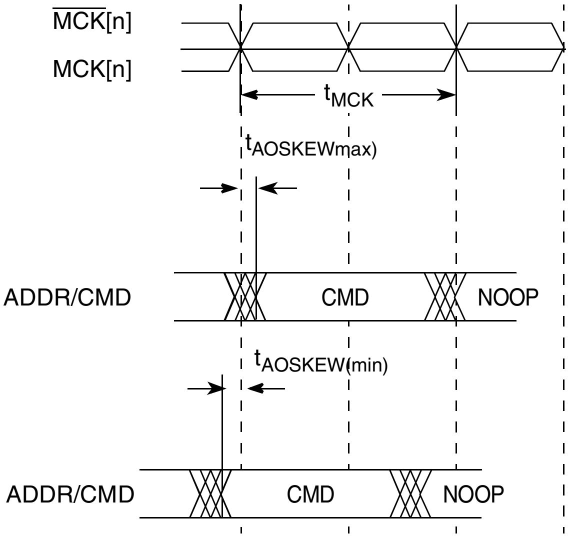


Figure 4. Timing Diagram for t_{AOSKEW} Measurement

Figure 5 shows the DDR SDRAM output timing diagram for the source synchronous mode.

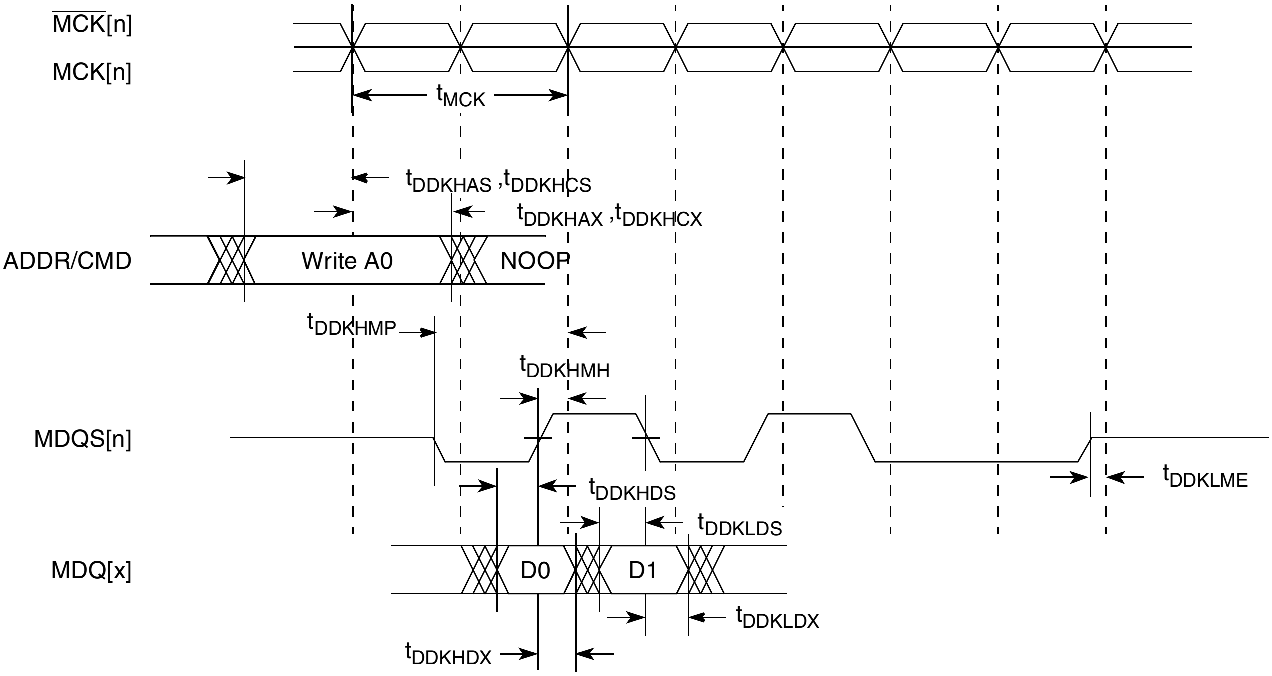


Figure 5. DDR SDRAM Output Timing Diagram for Source Synchronous Mode

Figure 6 provides the AC test load for the DDR bus.

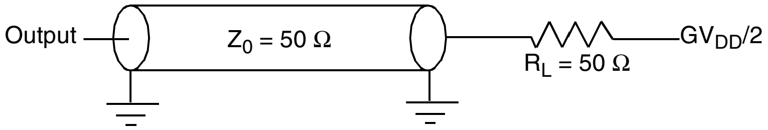


Figure 6. DDR AC Test Load

Table 15. DDR SDRAM Measurement Conditions

Symbol	DDR	Unit	Notes
V_{TH}	MV_{REF} ± 0.31 V	V	1
V_{OUT}	$0.5 \times GV_{DD}$	V	2

Notes:
1. Data input threshold measurement point.
2. Data output measurement point.

7 DUART

This section describes the DC and AC electrical specifications for the DUART interface of the MPC8541E.

7.1 DUART DC Electrical Characteristics

Table 16 provides the DC electrical characteristics for the DUART interface of the MPC8541E.

Table 16. DUART DC Electrical Characteristics

Parameter	Symbol	Test Condition	Min	Max	Unit
High-level input voltage	V_{IH}	$V_{OUT} \geq V_{OH}$ (min) or $V_{OUT} \leq V_{OL}$ (max)	2	OV_{DD} + 0.3	V
Low-level input voltage	V_{IL}		–0.3	0.8	V
Input current	I_{IN}	V_{IN} [1] = 0 V or $V_{IN} = V_{DD}$	—	±5	μA
High-level output voltage	V_{OH}	OV_{DD} = min, I_{OH} = –100 μA	OV_{DD} – 0.2	—	V
Low-level output voltage	V_{OL}	OV_{DD} = min, I_{OL} = 100 μA	—	0.2	V

Note:
1. Note that the symbol V_{IN}, in this case, represents the OV_{IN} symbol referenced in Table 1 and Table 2.

8.2 GMII, MII, TBI, RGMII, and RTBI AC Timing Specifications

The AC timing specifications for GMII, MII, TBI, RGMII, and RTBI are presented in this section.

8.2.1 GMII AC Timing Specifications

This section describes the GMII transmit and receive AC timing specifications.

8.2.2 GMII Transmit AC Timing Specifications

Table 20 provides the GMII transmit AC timing specifications.

Table 20. GMII Transmit AC Timing Specifications

At recommended operating conditions with LV_{DD} of 3.3 V ± 5%.

Parameter/Condition	Symbol [1]	Min	Typ	Max	Unit
GTX_CLK clock period	t_{GTX}	—	8.0	—	ns
GTX_CLK duty cycle	t_{GTXH}/t_{GTX}	40	—	60	%
GMII data TXD[7:0], TX_ER, TX_EN setup time	t_{GTKHDV}	2.5	—	—	ns
GTX_CLK to GMII data TXD[7:0], TX_ER, TX_EN delay	t_{GTKHDX}	0.5	—	5.0	ns
GTX_CLK data clock rise and fall times	t_{GTXR}[3], t_{GTXR}[2,4]	—	—	1.0	ns

Notes:

1. The symbols used for timing specifications herein follow the pattern $t_{\text{(first two letters of functional block)(signal)(state) (reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{GTKHDV} symbolizes GMII transmit timing (GT) with respect to the t_{GTX} clock reference (K) going to the high state (H) relative to the time date input signals (D) reaching the valid state (V) to state or setup time. Also, t_{GTKHDX} symbolizes GMII transmit timing (GT) with respect to the t_{GTX} clock reference (K) going to the high state (H) relative to the time date input signals (D) going invalid (X) or hold time. Note that, in general, the clock reference symbol representation is based on three letters representing the clock of a particular functional. For example, the subscript of t_{GTX} represents the GMII(G) transmit (TX) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).
2. Signal timings are measured at 0.7 V and 1.9 V voltage levels.
3. Guaranteed by characterization.
4. Guaranteed by design.

Figure 7 shows the GMII transmit AC timing diagram.

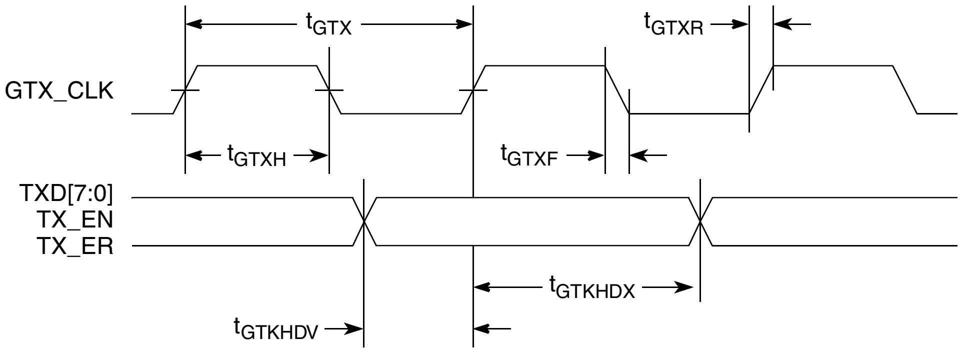


Figure 7. GMII Transmit AC Timing Diagram

8.2.5 RGMII and RTBI AC Timing Specifications

Table 26 presents the RGMII and RTBI AC timing specifications.

Table 26. RGMII and RTBI AC Timing Specifications

At recommended operating conditions with LV_{DD} of 2.5 V ± 5%.

Parameter/Condition	Symbol [1]	Min	Typ	Max	Unit
Data to clock output skew (at transmitter)	t_{SKRGT} [5]	–500	0	500	ps
Data to clock input skew (at receiver) [2]	t_{SKRGT}	1.0	—	2.8	ns
Clock cycle duration [3]	t_{RGT} [6]	7.2	8.0	8.8	ns
Duty cycle for 1000Base-T [4]	t_{RGTH}/t_{RGT} [6]	45	50	55	%
Duty cycle for 10BASE-T and 100BASE-TX [3]	t_{RGTH}/t_{RGT} [6]	40	50	60	%
Rise and fall times	t_{RGTR} [6,7], t_{RGTF} [6,7]	—	—	0.75	ns

Notes:

1. Note that, in general, the clock reference symbol representation for this section is based on the symbols RGT to represent RGMII and RTBI timing. For example, the subscript of t_{RGT} represents the TBI (T) receive (RX) clock. Note also that the notation for rise (R) and fall (F) times follows the clock symbol that is being represented. For symbols representing skews, the subscript is skew (SK) followed by the clock that is being skewed (RGT).
2. The RGMII specification requires that PC board designer add 1.5 ns or greater in trace delay to the RX_CLK in order to meet this specification. However, as stated above, this device functions with only 1.0 ns of delay.
3. For 10 and 100 Mbps, t_{RGT} scales to 400 ns ± 40 ns and 40 ns ± 4 ns, respectively.
4. Duty cycle may be stretched/shrunk during speed changes or while transitioning to a received packet's clock domains as long as the minimum duty cycle is not violated and stretching occurs for no more than three t_{RGT} of the lowest speed transitioned between.
5. Guaranteed by characterization.
6. Guaranteed by design.
7. Signal timings are measured at 0.5 and 2.0 V voltage levels.

Figure 14 shows the RBMII and RTBI AC timing and multiplexing diagrams.

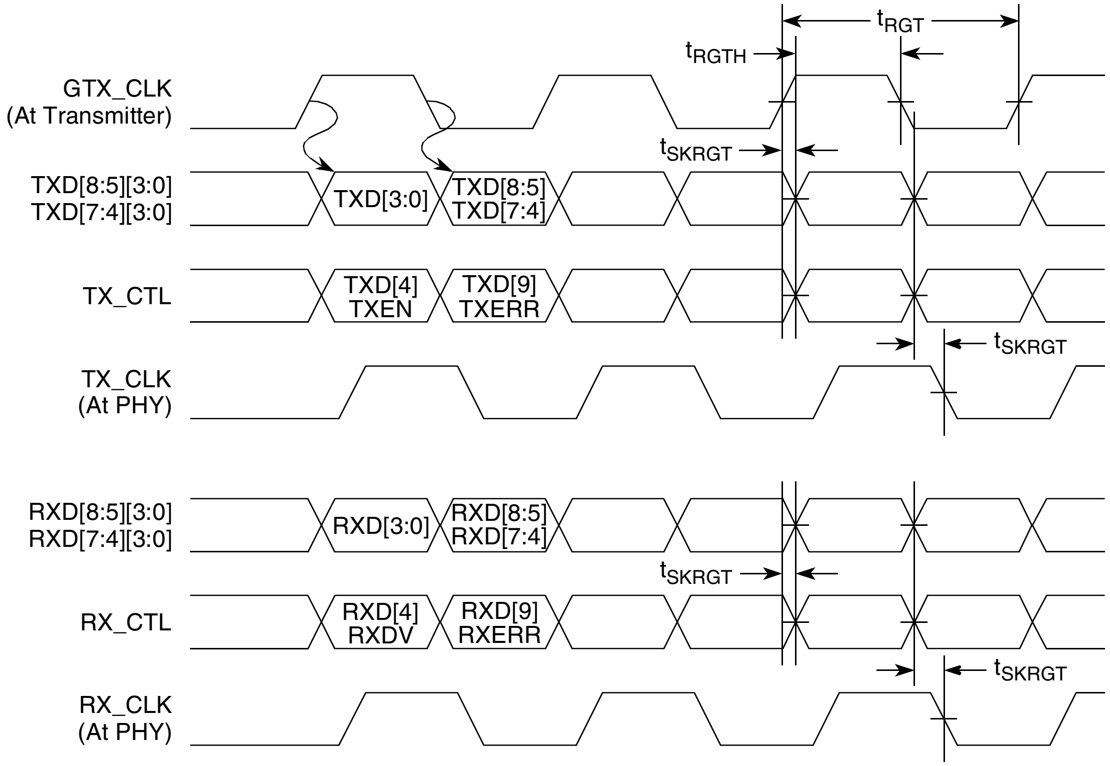


Figure 14. RGMII and RTBI AC Timing and Multiplexing Diagrams

8.3 Ethernet Management Interface Electrical Characteristics

The electrical characteristics specified here apply to MII management interface signals MDIO (management data input/output) and MDC (management data clock). The electrical characteristics for GMII, RGMII, TBI and RTBI are specified in Section 8.1, “Three-Speed Ethernet Controller (TSEC) (10/100/1000 Mbps)—GMII/MII/TBI/RGMII/RTBI Electrical Characteristics.”

8.3.1 MII Management DC Electrical Characteristics

The MDC and MDIO are defined to operate at a supply voltage of 3.3 V. The DC electrical characteristics for MDIO and MDC are provided in Table 27.

Table 27. MII Management DC Electrical Characteristics

Parameter	Symbol	Conditions		Min	Max	Unit
Supply voltage (3.3 V)	OV_{DD}	—		3.13	3.47	V
Output high voltage	V_{OH}	I_{OH} = –1.0 mA	LV_{DD} = Min	2.10	LV_{DD} + 0.3	V
Output low voltage	V_{OL}	I_{OL} = 1.0 mA	LV_{DD} = Min	GND	0.50	V
Input high voltage	V_{IH}	—		1.70	—	V
Input low voltage	V_{IL}	—		—	0.90	V

Table 30. Local Bus General Timing Parameters—DLL Enabled (continued)

Parameter	Configuration [7]	Symbol [1]	Min	Max	Unit	Notes
Local bus clock to output high impedance for LAD/LDP	$\overline{\text{LWE}[0:1]}$ = 00	$t_{LBKHOZ2}$	—	2.8	ns	5, 9
	$\overline{\text{LWE}[0:1]}$ = 11 (default)			4.2		

Notes:

1. The symbols used for timing specifications herein follow the pattern of $t_{\text{(First two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(First two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, $t_{LBIXKH1}$ symbolizes local bus timing (LB) for the input (I) to go invalid (X) with respect to the time the t_{LBK} clock reference (K) goes high (H), in this case for clock one(1). Also, t_{LBKHOX} symbolizes local bus timing (LB) for the t_{LBK} clock reference (K) to go high (H), with respect to the output (O) going invalid (X) or output hold time.
2. All timings are in reference to LSYNC_IN for DLL enabled mode.
3. All signals are measured from $OV_{DD}/2$ of the rising edge of LSYNC_IN for DLL enabled to $0.4 \times OV_{DD}$ of the signal in question for 3.3-V signaling levels.
4. Input timings are measured at the pin.
5. For purposes of active/float timing measurements, the Hi-Z or off state is defined to be when the total current delivered through the component pin is less than or equal to the leakage current specification.
6. The value of t_{LBOTOT} is defined as the sum of 1/2 or 1 ccb_clk cycle as programmed by LBCR[AHD], and the number of local bus buffer delays used as programmed at power-on reset with configuration pins $\overline{\text{LWE}[0:1]}$.
7. Maximum possible clock skew between a clock LCLK[m] and a relative clock LCLK[n]. Skew measured between complementary signals at $OV_{DD}/2$.
8. Guaranteed by characterization.
9. Guaranteed by design.

Table 31 describes the general timing parameters of the local bus interface of the MPC8541E with the DLL bypassed.

Table 31. Local Bus General Timing Parameters—DLL Bypassed

Parameter	Configuration [7]	Symbol [1]	Min	Max	Unit	Notes
Local bus cycle time	—	t_{LBK}	6.0	—	ns	2
Internal launch/capture clock to LCLK delay	—	t_{LBKHKT}	1.8	3.4	ns	8
LCLK[n] skew to LCLK[m] or LSYNC_OUT	—	$t_{LBKSKEW}$	—	150	ps	7, 9
Input setup to local bus clock (except LUPWAIT)	—	$t_{LBIVKH1}$	5.2	—	ns	3, 4
LUPWAIT input setup to local bus clock	—	$t_{LBIVKH2}$	5.1	—	ns	3, 4
Input hold from local bus clock (except LUPWAIT)	—	$t_{LBIXKH1}$	–1.3	—	ns	3, 4
LUPWAIT input hold from local bus clock	—	$t_{LBIXKH2}$	–0.8	—	ns	3, 4
LALE output transition to LAD/LDP output transition (LATCH hold time)	—	t_{LBOTOT}	1.5	—	ns	6
Local bus clock to output valid (except LAD/LDP and LALE)	$\overline{\text{LWE}[0:1]}$ = 00	$t_{LBKLOV1}$	—	0.5	ns	3
	$\overline{\text{LWE}[0:1]}$ = 11 (default)			2.0		
Local bus clock to data valid for LAD/LDP	$\overline{\text{LWE}[0:1]}$ = 00	$t_{LBKLOV2}$	—	0.7	ns	3
	$\overline{\text{LWE}[0:1]}$ = 11 (default)			2.2		

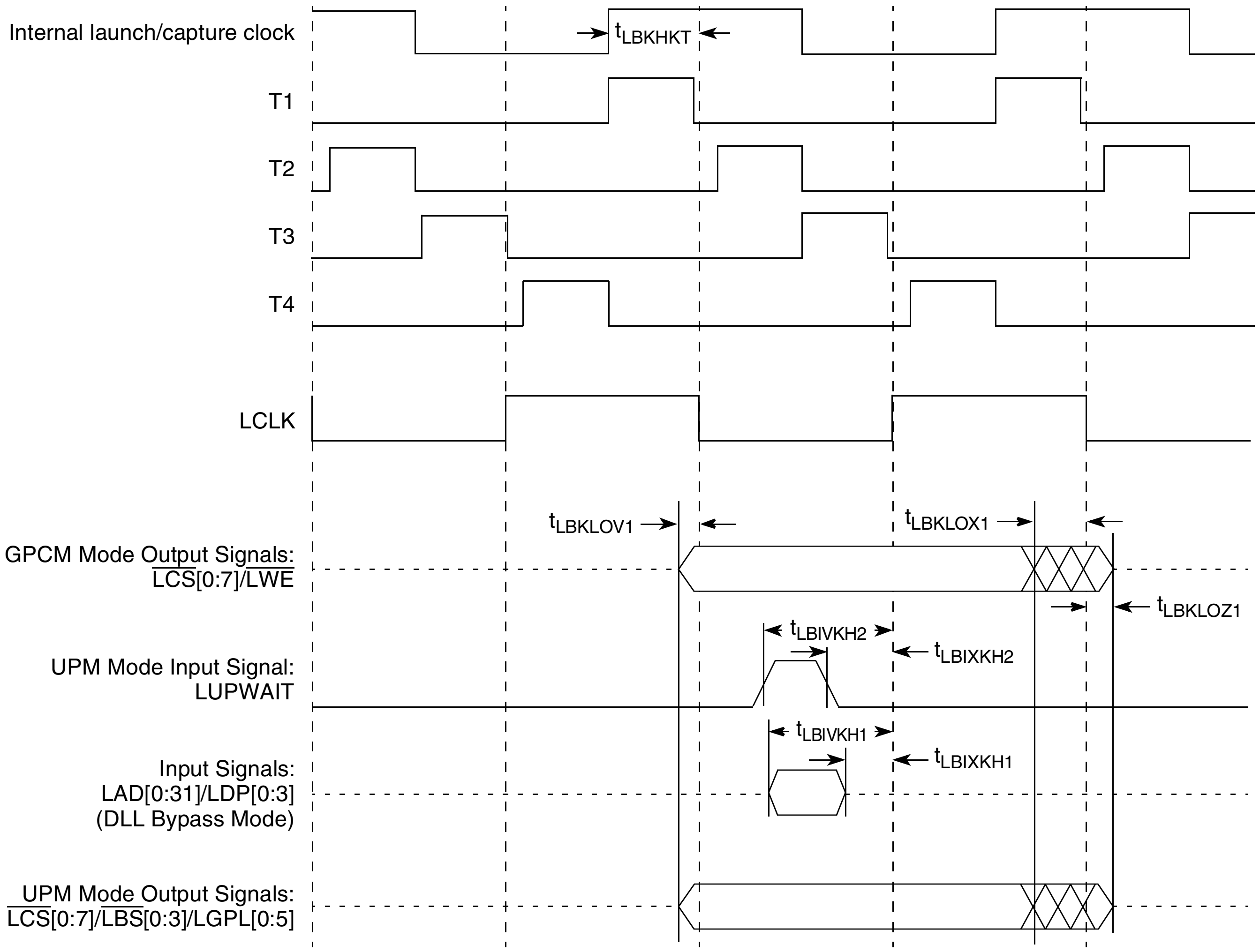


Figure 22. Local Bus Signals, GPCM/UPM Signals for LCCR[CLKDIV] = 4 or 8 (DLL Bypass Mode)

Table 33. CPM Input AC Timing Specifications [1] (continued)

Characteristic	Symbol [2]	Min[3]	Unit
PIO inputs—input hold time	t_{PIIXKH}	1	ns
COL width high (FCC)	t_{FCCH}	1.5	CLK

Notes:

1. Input specifications are measured from the 50% level of the signal to the 50% level of the rising edge of CLKIN. Timings are measured at the pin.
2. The symbols used for timing specifications herein follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{FIIVKH} symbolizes the FCC inputs internal timing (FI) with respect to the time the input signals (I) reaching the valid state (V) relative to the reference clock t_{FCC} (K) going to the high (H) state or setup time.
3. PIO and TIMER inputs and outputs are asynchronous to SYSCLK or any other externally visible clock. PIO/TIMER inputs are internally synchronized to the CPM internal clock. PIO/TIMER outputs should be treated as asynchronous.

Table 34. CPM Output AC Timing Specifications [1]

Characteristic	Symbol [2]	Min	Max	Unit
FCC outputs—internal clock (NMSI) delay	t_{FIKHOX}	1	5.5	ns
FCC outputs—external clock (NMSI) delay	t_{FEKHOX}	2	8	ns
SPI outputs—internal clock (NMSI) delay	t_{NIKHOX}	0.5	10	ns
SPI outputs—external clock (NMSI) delay	t_{NEKHOX}	2	8	ns
PIO outputs delay	t_{PIKHOX}	1	11	ns

Notes:

1. Output specifications are measured from the 50% level of the rising edge of CLKIN to the 50% level of the signal. Timings are measured at the pin.
2. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state) (reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{FIKHOX} symbolizes the FCC inputs internal timing (FI) for the time t_{FCC} memory clock reference (K) goes from the high state (H) until outputs (O) are invalid (X).

Figure 23 provides the AC test load for the CPM.

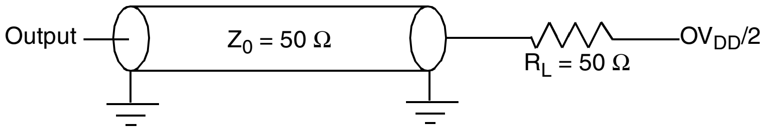


Figure 23. CPM AC Test Load

10.3 CPM I2C AC Specification

Table 35. I2C Timing

Characteristic	Expression	All Frequencies		Unit
		Min	Max	
SCL clock frequency (slave)	f_{SCL}	0	F_{MAX}(1)	Hz
SCL clock frequency (master)	f_{SCL}	BRGCLK/16512	BRGCLK/48	Hz
Bus free time between transmissions	t_{SDHDL}	$1/(2.2 * f_{SCL})$	—	s
Low period of SCL	t_{SCLCH}	$1/(2.2 * f_{SCL})$	—	s
High period of SCL	t_{SCHCL}	$1/(2.2 * f_{SCL})$	—	s
Start condition setup time[2]	t_{SCHDL}	$2/(divider * f_{SCL})$	(2)	s
Start condition hold time[2]	t_{SDLCL}	$3/(divider * f_{SCL})$	—	s
Data hold time [2]	t_{SCLDX}	$2/(divider * f_{SCL})$	—	s
Data setup time[2]	t_{SDVCH}	$3/(divider * f_{SCL})$	—	s
SDA/SCL rise time	t_{SRISE}	—	$1/(10 * f_{SCL})$	s
SDA/SCL fall time	t_{SFALL}	—	$1/(33 * f_{SCL})$	s
Stop condition setup time	t_{SCHDH}	$2/(divider * f_{SCL})$	—	s

Notes:

1. F_{MAX} = BRGCLK/(min_divider*prescale. Where prescaler=25-I2MODE[PDIV]; and min_divider=12 if digital filter disabled and 18 if enabled.
   Example #1: if I2MODE[PDIV]=11 (prescaler=4) and I2MODE[FLT]=0 (digital filter disabled) then FMAX=BRGCLK/48
   Example #2: if I2MODE[PDIV]=00 (prescaler=32) and I2MODE[FLT]=1 (digital filter enabled) then FMAX=BRGCLK/576
2. divider = f_{SCL}/prescaler.
   In master mode: divider=BRGCLK/(f_{SCL}*prescaler)=2*(I2BRG[DIV]+3)
   In slave mode: divider=BRGCLK/(f_{SCL}*prescaler)

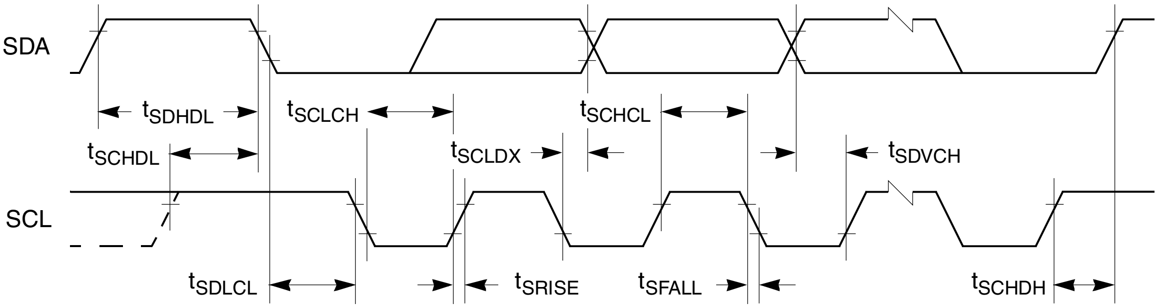


Figure 30. CPM I2C Bus Timing Diagram

12.2 I^{2}C AC Electrical Specifications

Table 40 provides the AC timing parameters for the I^{2}C interface of the MPC8541E.

Table 40. I^{2}C AC Electrical Specifications

All values refer to V_{IH} (min) and V_{IL} (max) levels (see Table 39).

Parameter	Symbol [1]	Min	Max	Unit
SCL clock frequency	f_{I2C}	0	400	kHz
Low period of the SCL clock	t_{I2CL} [6]	1.3	—	µs
High period of the SCL clock	t_{I2CH} [6]	0.6	—	µs
Setup time for a repeated START condition	t_{I2SVKH} [6]	0.6	—	µs
Hold time (repeated) START condition (after this period, the first clock pulse is generated)	t_{I2SXKL} [6]	0.6	—	µs
Data setup time	t_{I2DVKH} [6]	100	—	ns
Data hold time: CBUS compatible masters I^{2}C bus devices	t_{I2DXKL}	 — 0 [2]	 — 0.9 [3]	µs
Rise time of both SDA and SCL signals	t_{I2CR}	$20 + 0.1\ C_b$ [4]	300	ns
Fall time of both SDA and SCL signals	t_{I2CF}	$20 + 0.1\ C_b$ [4]	300	ns
Set-up time for STOP condition	t_{I2PVKH}	0.6	—	µs
Bus free time between a STOP and START condition	t_{I2KHDX}	1.3	—	µs
Noise margin at the LOW level for each connected device (including hysteresis)	V_{NL}	$0.1 \times OV_{DD}$	—	V
Noise margin at the HIGH level for each connected device (including hysteresis)	V_{NH}	$0.2 \times OV_{DD}$	—	V

Notes:

1. The symbols used for timing specifications herein follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state) (reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{I2DVKH} symbolizes I^{2}C timing (I2) with respect to the time data input signals (D) reach the valid state (V) relative to the t_{I2C} clock reference (K) going to the high (H) state or setup time. Also, t_{I2SXKL} symbolizes I^{2}C timing (I2) for the time that the data with respect to the start condition (S) went invalid (X) relative to the t_{I2C} clock reference (K) going to the low (L) state or hold time. Also, t_{I2PVKH} symbolizes I^{2}C timing (I2) for the time that the data with respect to the stop condition (P) reaching the valid state (V) relative to the t_{I2C} clock reference (K) going to the high (H) state or setup time. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).
2. MPC8541E provides a hold time of at least 300 ns for the SDA signal (referred to the V_{IHmin} of the SCL signal) to bridge the undefined region of the falling edge of SCL.
3. The maximum t_{I2DVKH} has only to be met if the device does not stretch the LOW period (t_{I2CL}) of the SCL signal.
4. C_B = capacitance of one bus line in pF.
5. Guaranteed by design.

13.2 PCI AC Electrical Specifications

This section describes the general AC timing parameters of the PCI bus of the MPC8541E. Note that the SYSCLK signal is used as the PCI input clock. Table 42 provides the PCI AC timing specifications at 66 MHz.

NOTE

PCI Clock can be PCI1_CLK or SYSCLK based on POR config input.

NOTE

The input setup time does not meet the PCI specification.

Table 42. PCI AC Timing Specifications at 66 MHz

Parameter	Symbol [1]	Min	Max	Unit	Notes
Clock to output valid	t_{PCKHOV}	—	6.0	ns	2, 3
Output hold from Clock	t_{PCKHOX}	2.0	—	ns	2, 9
Clock to output high impedance	t_{PCKHOZ}	—	14	ns	2, 3, 10
Input setup to Clock	t_{PCIVKH}	3.3	—	ns	2, 4, 9
Input hold from Clock	t_{PCIXKH}	0	—	ns	2, 4, 9
$\overline{\text{REQ64}}$ to $\overline{\text{HRESET}}$ [9] setup time	t_{PCRVRH}	$10 \times t_{SYS}$	—	clocks	5, 6, 10
$\overline{\text{HRESET}}$ to $\overline{\text{REQ64}}$ hold time	t_{PCRHRX}	0	50	ns	6, 10
$\overline{\text{HRESET}}$ high to first $\overline{\text{FRAME}}$ assertion	t_{PCRHFV}	10	—	clocks	7, 10

Notes:

1. Note that the symbols used for timing specifications herein follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{PCIVKH} symbolizes PCI timing (PC) with respect to the time the input signals (I) reach the valid state (V) relative to the SYSCLK clock, t_{SYS}, reference (K) going to the high (H) state or setup time. Also, t_{PCRHFV} symbolizes PCI timing (PC) with respect to the time hard reset (R) went high (H) relative to the frame signal (F) going to the valid (V) state.
2. See the timing measurement conditions in the *PCI 2.2 Local Bus Specifications*.
3. For purposes of active/float timing measurements, the Hi-Z or off state is defined to be when the total current delivered through the component pin is less than or equal to the leakage current specification.
4. Input timings are measured at the pin.
5. The timing parameter t_{SYS} indicates the minimum and maximum CLK cycle times for the various specified frequencies. The system clock period must be kept within the minimum and maximum defined ranges. For values see Section 15, "Clocking."
6. The setup and hold time is with respect to the rising edge of $\overline{\text{HRESET}}$.
7. The timing parameter t_{PCRHFV} is a minimum of 10 clocks rather than the minimum of 5 clocks in the *PCI 2.2 Local Bus Specifications*.
8. The reset assertion timing requirement for $\overline{\text{HRESET}}$ is 100 µs.
9. Guaranteed by characterization.
10. Guaranteed by design.

Figure 16 provides the AC test load for PCI.

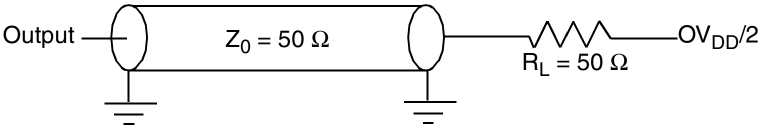


Figure 38. PCI AC Test Load

16.2.2 Internal Package Conduction Resistance

For the packaging technology, shown in Table 49, the intrinsic internal conduction thermal resistance paths are as follows:

- The die junction-to-case thermal resistance
- The die junction-to-board thermal resistance

Figure 44 depicts the primary heat transfer path for a package with an attached heat sink mounted to a printed-circuit board.

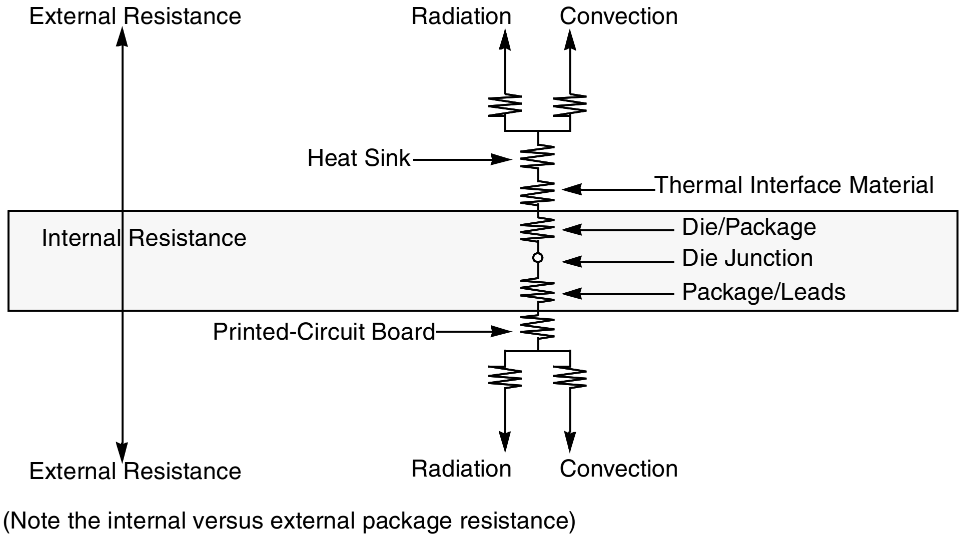


Figure 44. Package with Heat Sink Mounted to a Printed-Circuit Board

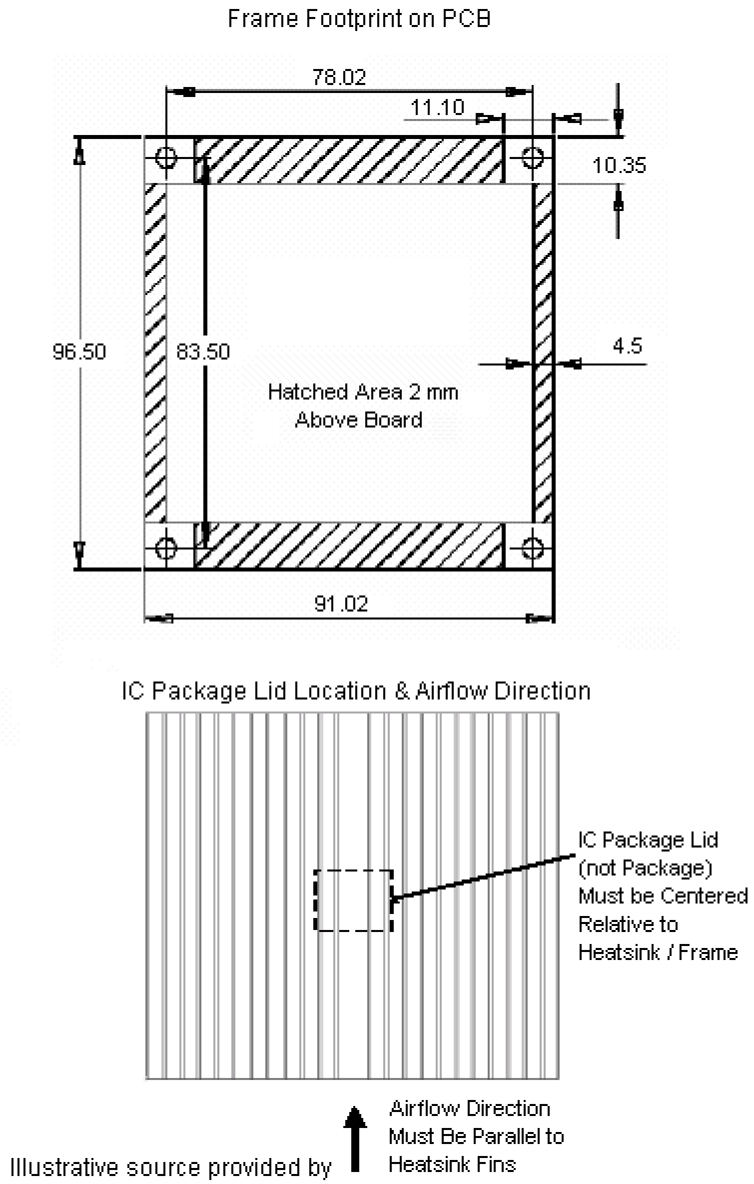
The heat sink removes most of the heat from the device. Heat generated on the active side of the chip is conducted through the silicon and through the lid, then through the heat sink attach material (or thermal interface material), and finally to the heat sink. The junction-to-case thermal resistance is low enough that the heat sink attach material and heat sink thermal resistance are the dominant terms.

16.2.3 Thermal Interface Materials

A thermal interface material is required at the package-to-heat sink interface to minimize the thermal contact resistance. For those applications where the heat sink is attached by spring clip mechanism, Figure 45 shows the thermal performance of three thin-sheet thermal-interface materials (silicone, graphite/oil, floroether oil), a bare joint, and a joint with thermal grease as a function of contact pressure. As shown, the performance of these thermal interface materials improves with increasing contact pressure. The use of thermal grease significantly reduces the interface thermal resistance. The bare joint results in a thermal resistance approximately six times greater than the thermal grease joint.

Heat sinks are attached to the package by means of a spring clip to holes in the printed-circuit board (see Figure 41). Therefore, the synthetic grease offers the best thermal performance, especially at the low interface pressure.

When removing the heat sink for re-work, it is preferable to slide the heat sink off slowly until the thermal interface material loses its grip. If the support fixture around the package prevents sliding off the heat sink, The spring mounting should be designed to apply the force only directly above the die. By localizing the force, rocking of the heat sink is minimized. One suggested mounting method attaches a plastic fence to the board to provide the structure on which the heat sink spring clips. The plastic fence also provides the opportunity to minimize the holes in the printed-circuit board and to locate them at the corners of the package. Figure 47 and provide exploded views of the plastic fence, heat sink, and spring clip.



Illustrative source provided by
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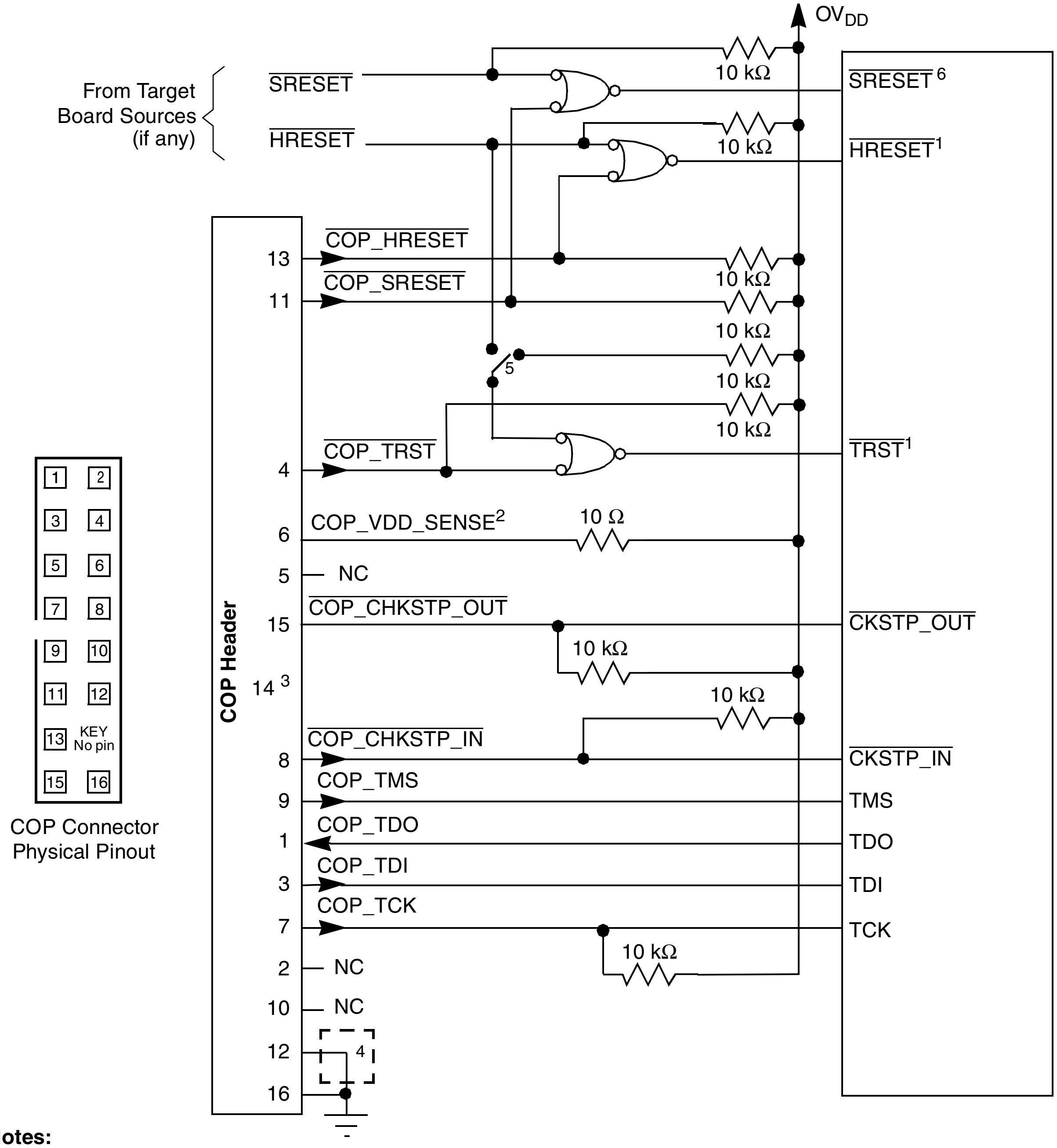
Figure 47. Exploded Views (1) of a Heat Sink Attachment using a Plastic Fence



Notes:

1. The COP port and target board should be able to independently assert HRESET and TRST to the processor in order to fully control the processor as shown here.
2. Populate this with a 10 Ω resistor for short-circuit/current-limiting protection.
3. The KEY location (pin 14) is not physically present on the COP header.
4. Although pin 12 is defined as a No-Connect, some debug tools may use pin 12 as an additional GND pin for improved signal integrity.
5. This switch is included as a precaution for BSDL testing. The switch should be open during BSDL testing to avoid accidentally asserting the TRST line. If BSDL testing is not being performed, this switch should be closed or removed.
6. Asserting SRESET causes a machine check interrupt to the e500 core.

Figure 52. JTAG Interface Connection

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