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Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Active
Core Processor	PowerPC e500
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	667MHz
Co-Processors/DSP	Security; SEC
RAM Controllers	DDR, SDRAM
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100/1000Mbps (2)
SATA	-
USB	-
Voltage - I/O	2.5V, 3.3V
Operating Temperature	-40°C ~ 105°C (TA)
Security Features	Cryptography, Random Number Generator
Package / Case	783-BBGA, FCBGA
Supplier Device Package	783-FCPBGA (29x29)
Purchase URL	https://www.e-xfl.com/pro/item?MUrl=&PartUrl=mpc8541ecpxalf

1 Overview

The following section provides a high-level overview of the MPC8541E features. Figure 1 shows the major functional units within the MPC8541E.

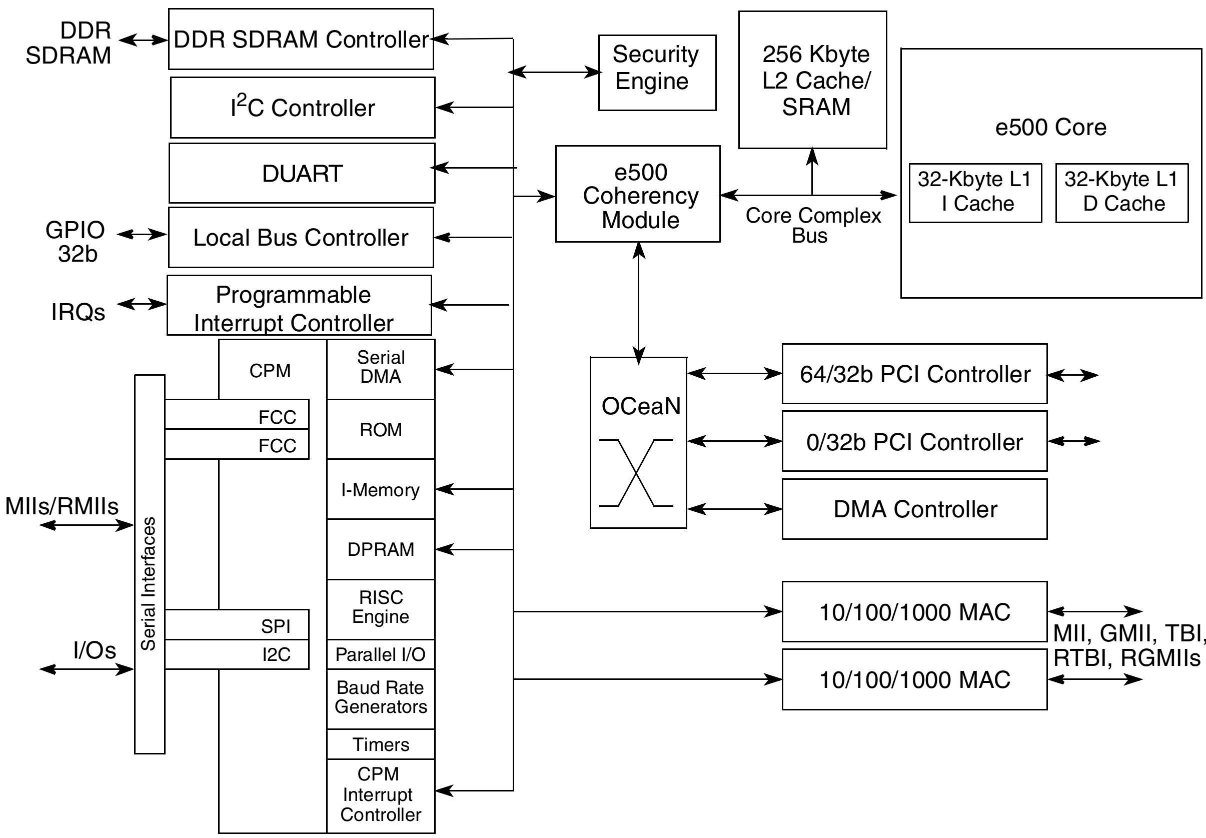

Figure 1. MPC8541E Block Diagram

1.1 Key Features

The following lists an overview of the MPC8541E feature set.

- Embedded e500 Book E-compatible core
  - High-performance, 32-bit Book E-enhanced core that implements the PowerPC architecture
  - Dual-issue superscalar, 7-stage pipeline design
  - 32-Kbyte L1 instruction cache and 32-Kbyte L1 data cache with parity protection
  - Lockable L1 caches—entire cache or on a per-line basis
  - Separate locking for instructions and data
  - Single-precision floating-point operations
  - Memory management unit especially designed for embedded applications
  - Enhanced hardware and software debug support
  - Dynamic power management
  - Performance monitor facility

  - SRAM operation supports relocation and is byte-accessible
  - Cache mode supports instruction caching, data caching, or both
  - External masters can force data to be allocated into the cache through programmed memory ranges or special transaction types (stashing).
  - Eight-way set-associative cache organization (1024 sets of 32-byte cache lines)
  - Supports locking the entire cache or selected lines
    - Individual line locks set and cleared through Book E instructions or by externally mastered transactions
  - Global locking and flash clearing done through writes to L2 configuration registers
  - Instruction and data locks can be flash cleared separately
  - Read and write buffering for internal bus accesses
- Address translation and mapping unit (ATMU)
  - Eight local access windows define mapping within local 32-bit address space
  - Inbound and outbound ATMUs map to larger external address spaces
    - Three inbound windows plus a configuration window on PCI
    - Four inbound windows
    - Four outbound windows plus default translation for PCI
- DDR memory controller
  - Programmable timing supporting first generation DDR SDRAM
  - 64-bit data interface, up to MHz data rate
  - Four banks of memory supported, each up to 1 Gbyte
  - DRAM chip configurations from 64 Mbits to 1 Gbit with x8/x16 data ports
  - Full ECC support
  - Page mode support (up to 16 simultaneous open pages)
  - Contiguous or discontiguous memory mapping
  - Sleep mode support for self refresh DDR SDRAM
  - Supports auto refreshing
  - On-the-fly power management using CKE signal
  - Registered DIMM support
  - Fast memory access via JTAG port
  - 2.5-V SSTL2 compatible I/O
- Programmable interrupt controller (PIC)
  - Programming model is compliant with the OpenPIC architecture
  - Supports 16 programmable interrupt and processor task priority levels
  - Supports 12 discrete external interrupts
  - Supports 4 message interrupts with 32-bit messages
  - Supports connection of an external interrupt controller such as the 8259 programmable interrupt controller

Items on the same line have no ordering requirement with respect to one another. Items on separate lines must be ordered sequentially such that voltage rails on a previous step must reach 90 percent of their value before the voltage rails on the current step reach ten percent of theirs.

NOTE

If the items on line 2 must precede items on line 1, please ensure that the delay does not exceed 500 ms and the power sequence is not done greater than once per day in production environment.

NOTE

From a system standpoint, if the I/O power supplies ramp prior to the V_{DD} core supply, the I/Os on the MPC8541E may drive a logic one or zero during power-up.

2.1.3 Recommended Operating Conditions

Table 2 provides the recommended operating conditions for the MPC8541E. Note that the values in Table 2 are the recommended and tested operating conditions. Proper device operation outside of these conditions is not guaranteed.

Table 2. Recommended Operating Conditions

Characteristic		Symbol	Recommended Value	Unit
Core supply voltage		V_{DD}	1.2 V ± 60 mV 1.3 V± 50 mV (for 1 GHz only)	V
PLL supply voltage		AV_{DD}	1.2 V ± 60 mV 1.3 V ± 50 mV (for 1 GHz only)	V
DDR DRAM I/O voltage		GV_{DD}	2.5 V ± 125 mV	V
Three-speed Ethernet I/O voltage		LV_{DD}	3.3 V ± 165 mV 2.5 V ± 125 mV	V
PCI, local bus, DUART, system control and power management, I^2C, and JTAG I/O voltage		OV_{DD}	3.3 V ± 165 mV	V
Input voltage	DDR DRAM signals	MV_{IN}	GND to GV_{DD}	V
	DDR DRAM reference	MV_{REF}	GND to GV_{DD}	V
	Three-speed Ethernet signals	LV_{IN}	GND to LV_{DD}	V
	PCI, local bus, DUART, SYSCLK, system control and power management, I^2C, and JTAG signals	OV_{IN}	GND to OV_{DD}	V
Die-junction Temperature		T_j	0 to 105	°C

Table 5. Typical I/O Power Dissipation

Interface	Parameters	GV_{DD} (2.5 V)	OV_{DD} (3.3 V)	LV_{DD} (3.3 V)	LV_{DD} (2.5 V)	Unit	Comments
DDR I/O	CCB = 200 MHz	0.46	—	—	—	W	—
	CCB = 266 MHz	0.59	—	—	—	W	—
	CCB = 300 MHz	0.66	—	—	—	W	—
	CCB = 333 MHz	0.73	—	—	—	W	—
PCI I/O	64b, 66 MHz	—	0.14	—	—	W	—
	64b, 33 MHz	—	0.08	—	—	W	—
	32b, 66 MHz	—	0.07	—	—	W	Multiply by 2 if using two 32b ports
	32b, 33 MHz	—	0.04	—	—	W	
Local Bus I/O	32b, 167 MHz	—	0.30	—	—	W	—
	32b, 133 MHz	—	0.24	—	—	W	—
	32b, 83 MHz	—	0.16	—	—	W	—
	32b, 66 MHz	—	0.13	—	—	W	—
	32b, 33 MHz	—	0.07	—	—	W	—
TSEC I/O	MII	—	—	0.01	—	W	Multiply by number of interfaces used.
	GMII or TBI	—	—	0.07	—	W	
	RGMII or RTBI	—	—	—	0.04	W	
CPM - FCC	MII	—	0.015	—	—	W	—
	RMII	—	0.013	—	—	W	—
	HDLC 16 Mbps	—	0.009	—	—	W	—

6 DDR SDRAM

This section describes the DC and AC electrical specifications for the DDR SDRAM interface of the MPC8541E.

6.1 DDR SDRAM DC Electrical Characteristics

Table 11 provides the recommended operating conditions for the DDR SDRAM component(s) of the MPC8541E.

Table 11. DDR SDRAM DC Electrical Characteristics

Parameter/Condition	Symbol	Min	Max	Unit	Notes
I/O supply voltage	GV_{DD}	2.375	2.625	V	1
I/O reference voltage	MV_{REF}	$0.49 \times GV_{DD}$	$0.51 \times GV_{DD}$	V	2
I/O termination voltage	V_{TT}	$MV_{REF} – 0.04$	$MV_{REF} + 0.04$	V	3
Input high voltage	V_{IH}	$MV_{REF} + 0.18$	$GV_{DD} + 0.3$	V	—
Input low voltage	V_{IL}	–0.3	$MV_{REF} – 0.18$	V	—
Output leakage current	I_{OZ}	–10	10	μA	4
Output high current (V_{OUT} = 1.95 V)	I_{OH}	–15.2	—	mA	—
Output low current (V_{OUT} = 0.35 V)	I_{OL}	15.2	—	mA	—
MV_{REF} input leakage current	I_{VREF}	—	5	μA	—

Notes:
1. GV_{DD} is expected to be within 50 mV of the DRAM GV_{DD} at all times.
2. MV_{REF} is expected to be equal to $0.5 \times GV_{DD}$, and to track GV_{DD} DC variations as measured at the receiver. Peak-to-peak noise on MV_{REF} may not exceed ±2% of the DC value.
3. V_{TT} is not applied directly to the device. It is the supply to which far end signal termination is made and is expected to be equal to MV_{REF}. This rail should track variations in the DC level of MV_{REF}.
4. Output leakage is measured with all outputs disabled, $0 V \le V_{OUT} \le GV_{DD}$.

Table 12 provides the DDR capacitance.

Table 12. DDR SDRAM Capacitance

Parameter/Condition	Symbol	Min	Max	Unit	Notes
Input/output capacitance: DQ, DQS, MSYNC_IN	C_{IO}	6	8	pF	1
Delta input/output capacitance: DQ, DQS	C_{DIO}	—	0.5	pF	1

Note:
1. This parameter is sampled. GV_{DD} = 2.5 V ± 0.125 V, f = 1 MHz, T_A = 25°C, $V_{OUT} = GV_{DD}/2$, V_{OUT} (peak to peak) = 0.2 V.

Table 14. DDR SDRAM Output AC Timing Specifications for Source Synchronous Mode (continued)

At recommended operating conditions with GV_{DD} of 2.5 V ± 5%.

Parameter	Symbol [1]	Min	Max	Unit	Notes
$\overline{MCS}$(n) output hold with respect to MCK 333 MHz 266 MHz 200 MHz	t_{DDKHCX}	 2.0 2.65 3.8	—	ns	4
MCK to MDQS 333 MHz 266 MHz 200 MHz	t_{DDKHMH}	 –0.9 –1.1 –1.2	 0.3 0.5 0.6	ns	5
MDQ/MECC/MDM output setup with respect to MDQS 333 MHz 266 MHz 200 MHz	t_{DDKHDS}, t_{DDKLDS}	 900 900 1200	—	ps	6
MDQ/MECC/MDM output hold with respect to MDQS 333 MHz 266 MHz 200 MHz	t_{DDKHDX}, t_{DDKLDX}	 900 900 1200	—	ps	6
MDQS preamble start	t_{DDKHMP}	$-0.5 \times t_{MCK} - 0.9$	$-0.5 \times t_{MCK} + 0.3$	ns	7
MDQS epilogue end	t_{DDKLME}	–0.9	0.3	ns	7

Notes:

1. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state) (reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. Output hold time can be read as DDR timing (DD) from the rising or falling edge of the reference clock (KH or KL) until the output went invalid (AX or DX). For example, t_{DDKHAS} symbolizes DDR timing (DD) for the time t_{MCK} memory clock reference (K) goes from the high (H) state until outputs (A) are setup (S) or output valid time. Also, t_{DDKLDX} symbolizes DDR timing (DD) for the time t_{MCK} memory clock reference (K) goes low (L) until data outputs (D) are invalid (X) or data output hold time.
2. All MCK/$\overline{MCK}$ referenced measurements are made from the crossing of the two signals ±0.1 V.
3. In the source synchronous mode, MCK/$\overline{MCK}$ can be shifted in 1/4 applied cycle increments through the Clock Control Register. For the skew measurements referenced for t_{AOSKEW} it is assumed that the clock adjustment is set to align the address/command valid with the rising edge of MCK.
4. ADDR/CMD includes all DDR SDRAM output signals except MCK/$\overline{MCK}$, $\overline{MCS}$, and MDQ/MECC/MDM/MDQS. For the ADDR/CMD setup and hold specifications, it is assumed that the Clock Control register is set to adjust the memory clocks by 1/2 applied cycle. The MCSx pins are separated from the ADDR/CMD (address and command) bus in the HW spec. This was separated because the MCSx pins typically have different loadings than the rest of the address and command bus, even though they have the same timings.
5. Note that t_{DDKHMH} follows the symbol conventions described in note 1. For example, t_{DDKHMH} describes the DDR timing (DD) from the rising edge of the MCK(n) clock (KH) until the MDQS signal is valid (MH). In the source synchronous mode, MDQS can launch later than MCK by 0.3 ns at the maximum. However, MCK may launch later than MDQS by as much as 0.9 ns. t_{DDKHMH} can be modified through control of the DQSS override bits in the TIMING_CFG_2 register. In source synchronous mode, this typically is set to the same delay as the clock adjust in the CLK_CNTL register. The timing parameters listed in the table assume that these two parameters have been set to the same adjustment value. See the *MPC8555E PowerQUICC™ III Integrated Communications Processor Reference Manual* for a description and understanding of the timing modifications enabled by use of these bits.
6. Determined by maximum possible skew between a data strobe (MDQS) and any corresponding bit of data (MDQ), ECC (MECC), or data mask (MDM). The data strobe should be centered inside of the data eye at the pins of the MPC8541E.
7. All outputs are referenced to the rising edge of MCK(n) at the pins of the MPC8541E. Note that t_{DDKHMP} follows the symbol conventions described in note 1.

Figure 4 shows the DDR SDRAM output timing for address skew with respect to any MCK.

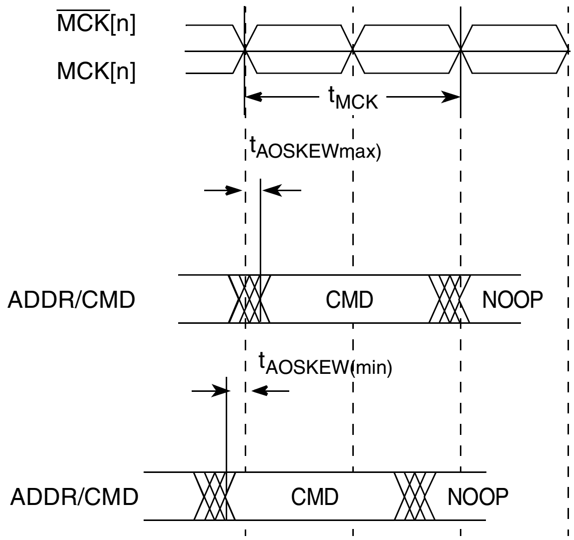

Figure 4. Timing Diagram for t_{AOSKEW} Measurement

Figure 5 shows the DDR SDRAM output timing diagram for the source synchronous mode.

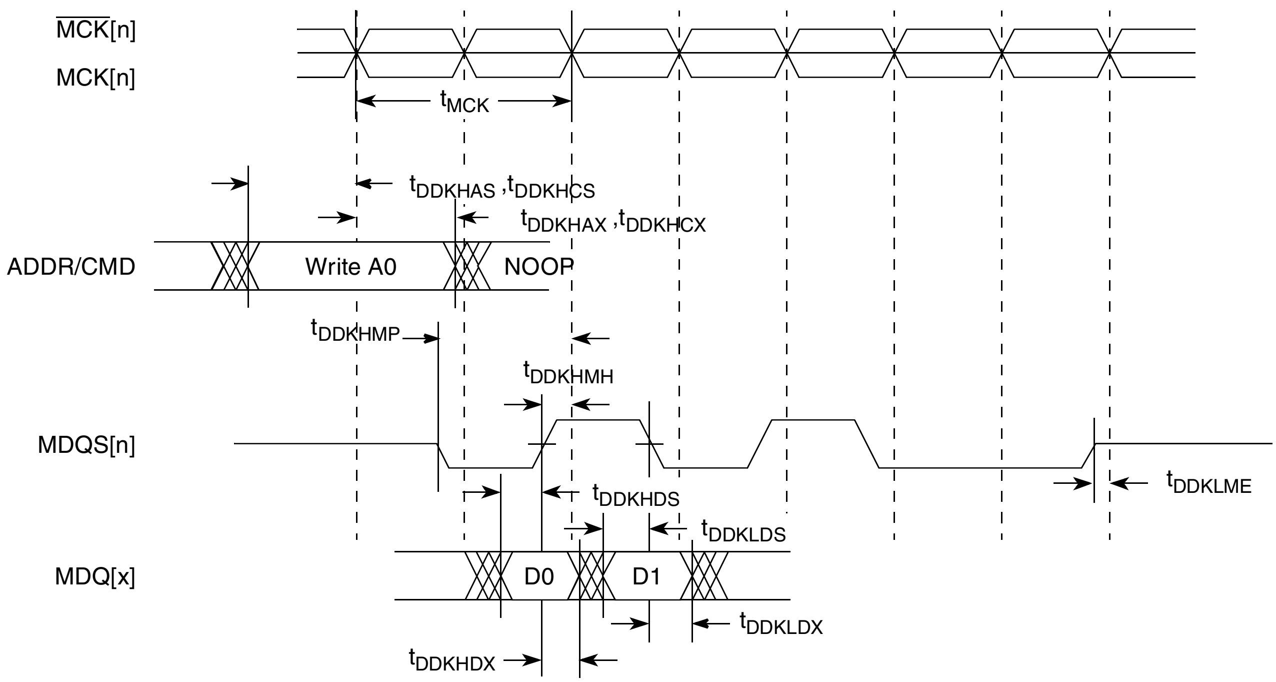

Figure 5. DDR SDRAM Output Timing Diagram for Source Synchronous Mode

Table 18. GMII, MII, and TBI DC Electrical Characteristics

Parameter	Symbol	Conditions		Min	Max	Unit
Supply voltage 3.3 V	LV_{DD}	—		3.13	3.47	V
Output high voltage	V_{OH}	I_{OH} = –4.0 mA	LV_{DD} = Min	2.40	LV_{DD} + 0.3	V
Output low voltage	V_{OL}	I_{OL} = 4.0 mA	LV_{DD} = Min	GND	0.50	V
Input high voltage	V_{IH}	—	—	1.70	LV_{DD} + 0.3	V
Input low voltage	V_{IL}	—	—	–0.3	0.90	V
Input high current	I_{IH}	V_{IN} [1] = LV_{DD}		—	40	μA
Input low current	I_{IL}	V_{IN} [1] = GND		–600	—	μA

Note:
1. The symbol V_{IN}, in this case, represents the LV_{IN} symbol referenced in Table 1 and Table 2.

Table 19. GMII, MII, RGMII RTBI, and TBI DC Electrical Characteristics

Parameters	Symbol	Min	Max	Unit
Supply voltage 2.5 V	LV_{DD}	2.37	2.63	V
Output high voltage (LV_{DD} = Min, I_{OH} = –1.0 mA)	V_{OH}	2.00	LV_{DD} + 0.3	V
Output low voltage (LV_{DD} = Min, I_{OL} = 1.0 mA)	V_{OL}	GND – 0.3	0.40	V
Input high voltage (LV_{DD} = Min)	V_{IH}	1.70	LV_{DD} + 0.3	V
Input low voltage (LV_{DD} = Min)	V_{IL}	–0.3	0.70	V
Input high current (V_{IN} [1] = LV_{DD})	I_{IH}	—	10	μA
Input low current (V_{IN} [1] = GND)	I_{IL}	–15	—	μA

Note:
1. Note that the symbol V_{IN}, in this case, represents the LV_{IN} symbol referenced in Table 1and Table 2.

8.2.4.2 TBI Receive AC Timing Specifications

Table 25 provides the TBI receive AC timing specifications.

Table 25. TBI Receive AC Timing Specifications

At recommended operating conditions with LV_{DD} of 3.3 V ± 5%.

Parameter/Condition	Symbol [1]	Min	Typ	Max	Unit
RX_CLK clock period	t_{TRX}		16.0		ns
RX_CLK skew	t_{SKTRX}	7.5	—	8.5	ns
RX_CLK duty cycle	t_{TRXH}/t_{TRX}	40	—	60	%
RCG[9:0] setup time to rising RX_CLK	t_{TRDVKH}	2.5	—	—	ns
RCG[9:0] hold time to rising RX_CLK	t_{TRDXKH}	1.5	—	—	ns
RX_CLK clock rise time and fall time	t_{TRXR}, t_{TRXF} [2,3]	0.7	—	2.4	ns

Note:

1. The symbols used for timing specifications herein follow the pattern of $t_{(first\ two\ letters\ of\ functional\ block)(signal)(state)(reference)(state)}$ for inputs and $t_{(first\ two\ letters\ of\ functional\ block)(reference)(state)(signal)(state)}$ for outputs. For example, t_{TRDVKH} symbolizes TBI receive timing (TR) with respect to the time data input signals (D) reach the valid state (V) relative to the t_{TRX} clock reference (K) going to the high (H) state or setup time. Also, t_{TRDXKH} symbolizes TBI receive timing (TR) with respect to the time data input signals (D) went invalid (X) relative to the t_{TRX} clock reference (K) going to the high (H) state. Note that, in general, the clock reference symbol representation is based on three letters representing the clock of a particular functional. For example, the subscript of t_{TRX} represents the TBI (T) receive (RX) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall). For symbols representing skews, the subscript is skew (SK) followed by the clock that is being skewed (TRX).
2. Guaranteed by design.

Figure 13 shows the TBI receive AC timing diagram.

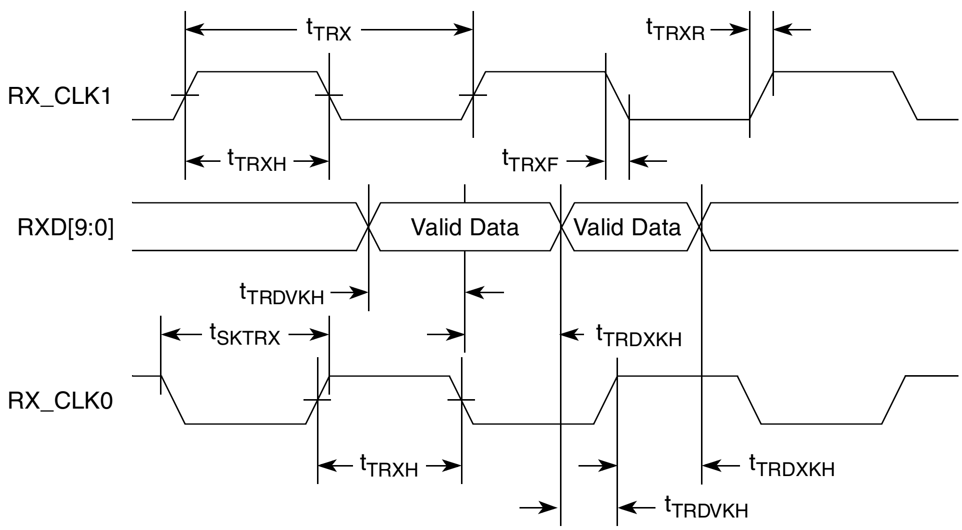

Figure 13. TBI Receive AC Timing Diagram

8.2.5 RGMII and RTBI AC Timing Specifications

Table 26 presents the RGMII and RTBI AC timing specifications.

Table 26. RGMII and RTBI AC Timing Specifications

At recommended operating conditions with LV_{DD} of 2.5 V ± 5%.

Parameter/Condition	Symbol [1]	Min	Typ	Max	Unit
Data to clock output skew (at transmitter)	t_{SKRGT} [5]	–500	0	500	ps
Data to clock input skew (at receiver) [2]	t_{SKRGT}	1.0	—	2.8	ns
Clock cycle duration [3]	t_{RGT} [6]	7.2	8.0	8.8	ns
Duty cycle for 1000Base-T [4]	t_{RGTH}/t_{RGT} [6]	45	50	55	%
Duty cycle for 10BASE-T and 100BASE-TX [3]	t_{RGTH}/t_{RGT} [6]	40	50	60	%
Rise and fall times	t_{RGTR} [6,7], t_{RGTF} [6,7]	—	—	0.75	ns

Notes:

1. Note that, in general, the clock reference symbol representation for this section is based on the symbols RGT to represent RGMII and RTBI timing. For example, the subscript of t_{RGT} represents the TBI (T) receive (RX) clock. Note also that the notation for rise (R) and fall (F) times follows the clock symbol that is being represented. For symbols representing skews, the subscript is skew (SK) followed by the clock that is being skewed (RGT).
2. The RGMII specification requires that PC board designer add 1.5 ns or greater in trace delay to the RX_CLK in order to meet this specification. However, as stated above, this device functions with only 1.0 ns of delay.
3. For 10 and 100 Mbps, t_{RGT} scales to 400 ns ± 40 ns and 40 ns ± 4 ns, respectively.
4. Duty cycle may be stretched/shrunk during speed changes or while transitioning to a received packet's clock domains as long as the minimum duty cycle is not violated and stretching occurs for no more than three t_{RGT} of the lowest speed transitioned between.
5. Guaranteed by characterization.
6. Guaranteed by design.
7. Signal timings are measured at 0.5 and 2.0 V voltage levels.

Table 31. Local Bus General Timing Parameters—DLL Bypassed (continued)

Parameter	Configuration [7]	Symbol [1]	Min	Max	Unit	Notes
Local bus clock to address valid for LAD	$\overline{LWE[0:1]}$ = 00	$t_{LBKLOV3}$	—	0.8	ns	3
	$\overline{LWE[0:1]}$ = 11 (default)			2.3		
Output hold from local bus clock (except LAD/LDP and LALE)	$\overline{LWE[0:1]}$ = 00	$t_{LBKLOX1}$	–2.7	—	ns	3
	$\overline{LWE[0:1]}$ = 11 (default)		–1.8			
Output hold from local bus clock for LAD/LDP	$\overline{LWE[0:1]}$ = 00	$t_{LBKLOX2}$	–2.7	—	ns	3
	$\overline{LWE[0:1]}$ = 11 (default)		–1.8			
Local bus clock to output high Impedance (except LAD/LDP and LALE)	$\overline{LWE[0:1]}$ = 00	$t_{LBKLOZ1}$	—	1.0	ns	5
	$\overline{LWE[0:1]}$ = 11 (default)			2.4		
Local bus clock to output high impedance for LAD/LDP	$\overline{LWE[0:1]}$ = 00	$t_{LBKLOZ2}$	—	1.0	ns	5
	$\overline{LWE[0:1]}$ = 11 (default)			2.4		

Notes:
1. The symbols used for timing specifications herein follow the pattern of $t_{\text{(First two letters of functional block)(signal)(state) (reference)(state)}}$ for inputs and $t_{\text{(First two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, $t_{LBIXKH1}$ symbolizes local bus timing (LB) for the input (I) to go invalid (X) with respect to the time the t_{LBK} clock reference (K) goes high (H), in this case for clock one(1). Also, t_{LBKHOX} symbolizes local bus timing (LB) for the t_{LBK} clock reference (K) to go high (H), with respect to the output (O) going invalid (X) or output hold time.
2. All timings are in reference to LSYNC_IN for DLL enabled mode.
3. All signals are measured from $OV_{DD}/2$ of the rising edge of local bus clock for DLL bypass mode to $0.4 \times OV_{DD}$ of the signal in question for 3.3-V signaling levels.
4. Input timings are measured at the pin.
5. For purposes of active/float timing measurements, the Hi-Z or off state is defined to be when the total current delivered through the component pin is less than or equal to the leakage current specification.
6. The value of t_{LBOTOT} is defined as the sum of 1/2 or 1 ccb_clk cycle as programmed by LBCR[AHD], and the number of local bus buffer delays used as programmed at power-on reset with configuration pins $\overline{LWE[0:1]}$.
7. Maximum possible clock skew between a clock LCLK[m] and a relative clock LCLK[n]. Skew measured between complementary signals at $OV_{DD}/2$.
8. Guaranteed by characterization.
9. Guaranteed by design.

Figure 16 provides the AC test load for the local bus.

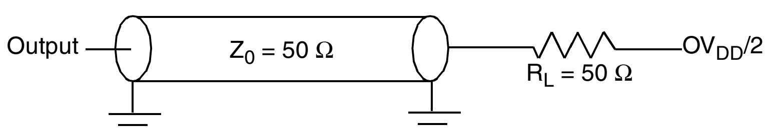

Figure 16. Local Bus C Test Load

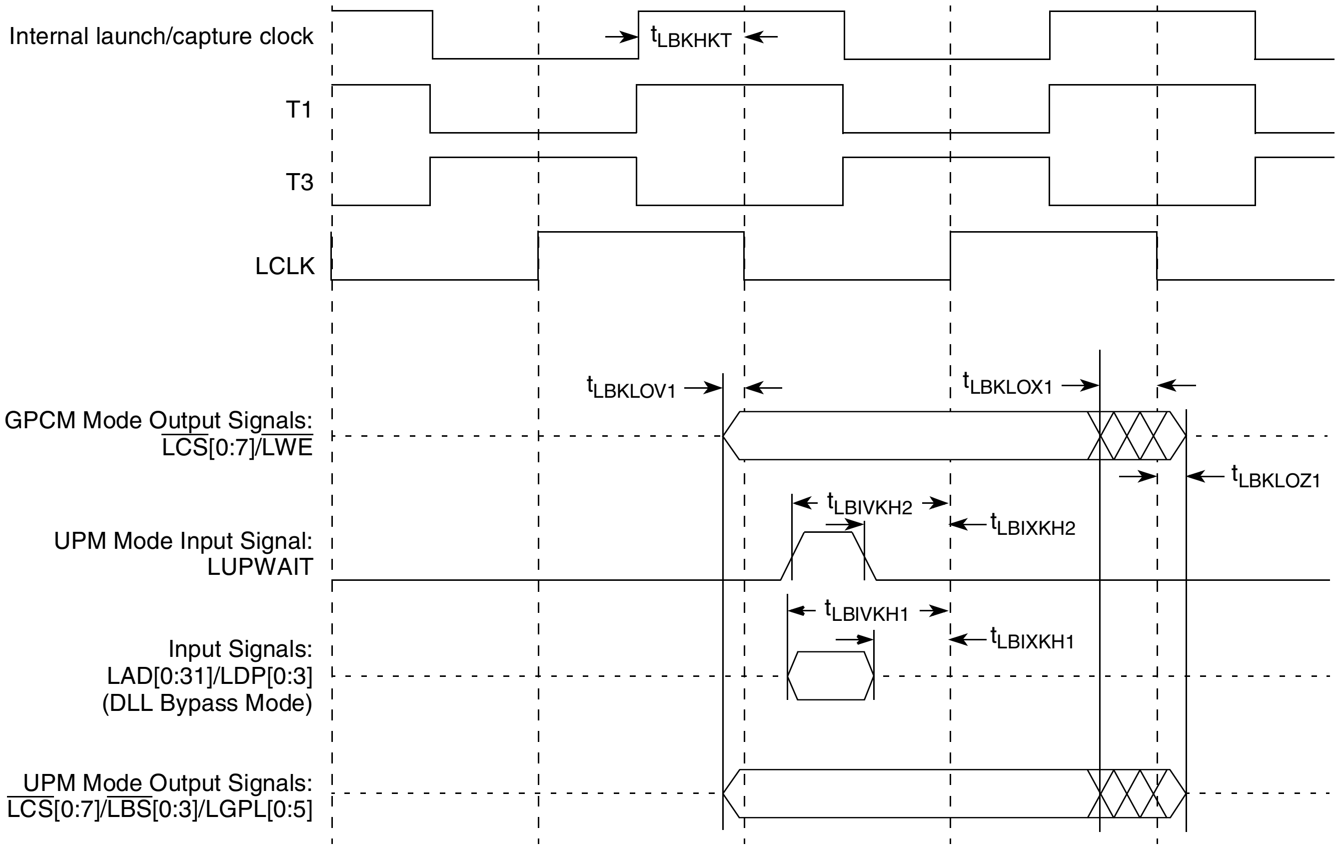

Figure 20. Local Bus Signals, GPCM/UPM Signals for LCCR[CLKDIV] = 2 (DLL Bypass Mode)

Figure 24 through Figure 29 represent the AC timing from Table 33 and Table 34. Note that although the specifications generally reference the rising edge of the clock, these AC timing diagrams also apply when the falling edge is the active edge.

Figure 24 shows the FCC internal clock.

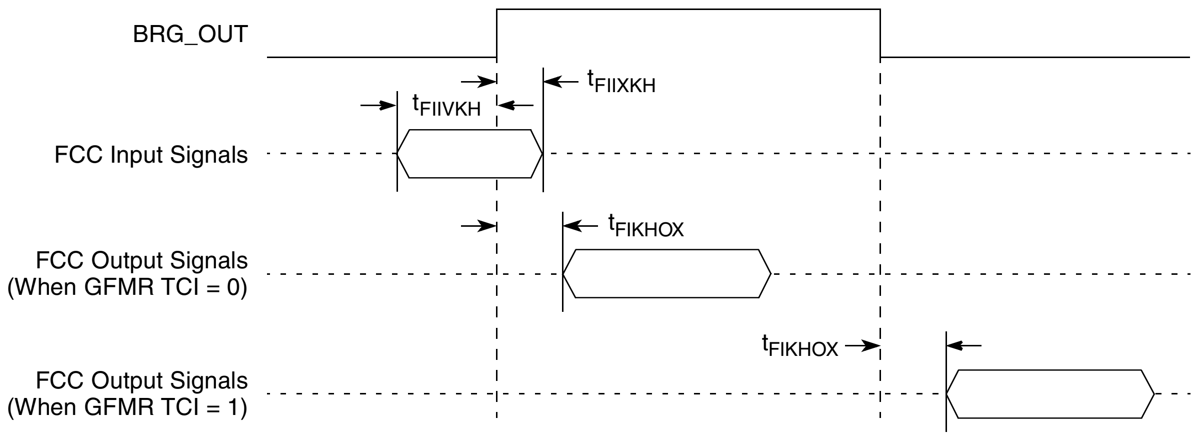

Figure 24. FCC Internal AC Timing Clock Diagram

Figure 25 shows the FCC external clock.

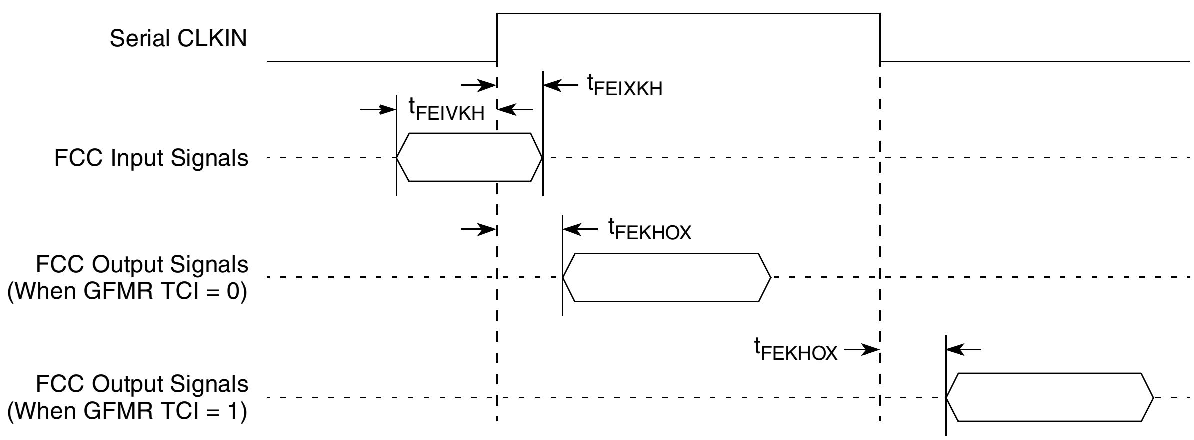

Figure 25. FCC External AC Timing Clock Diagram

Figure 26 shows Ethernet collision timing on FCCs.

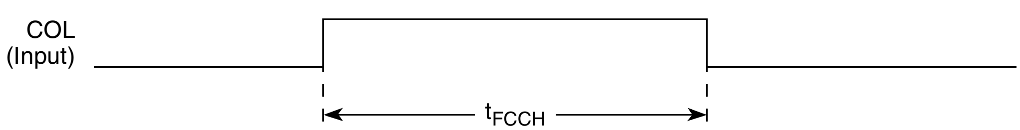

Figure 26. Ethernet Collision AC Timing Diagram (FCC)

14.2 Mechanical Dimensions of the FC-PBGA

Figure 41 the mechanical dimensions and bottom surface nomenclature of the MPC8541E 783 FC-PBGA package.

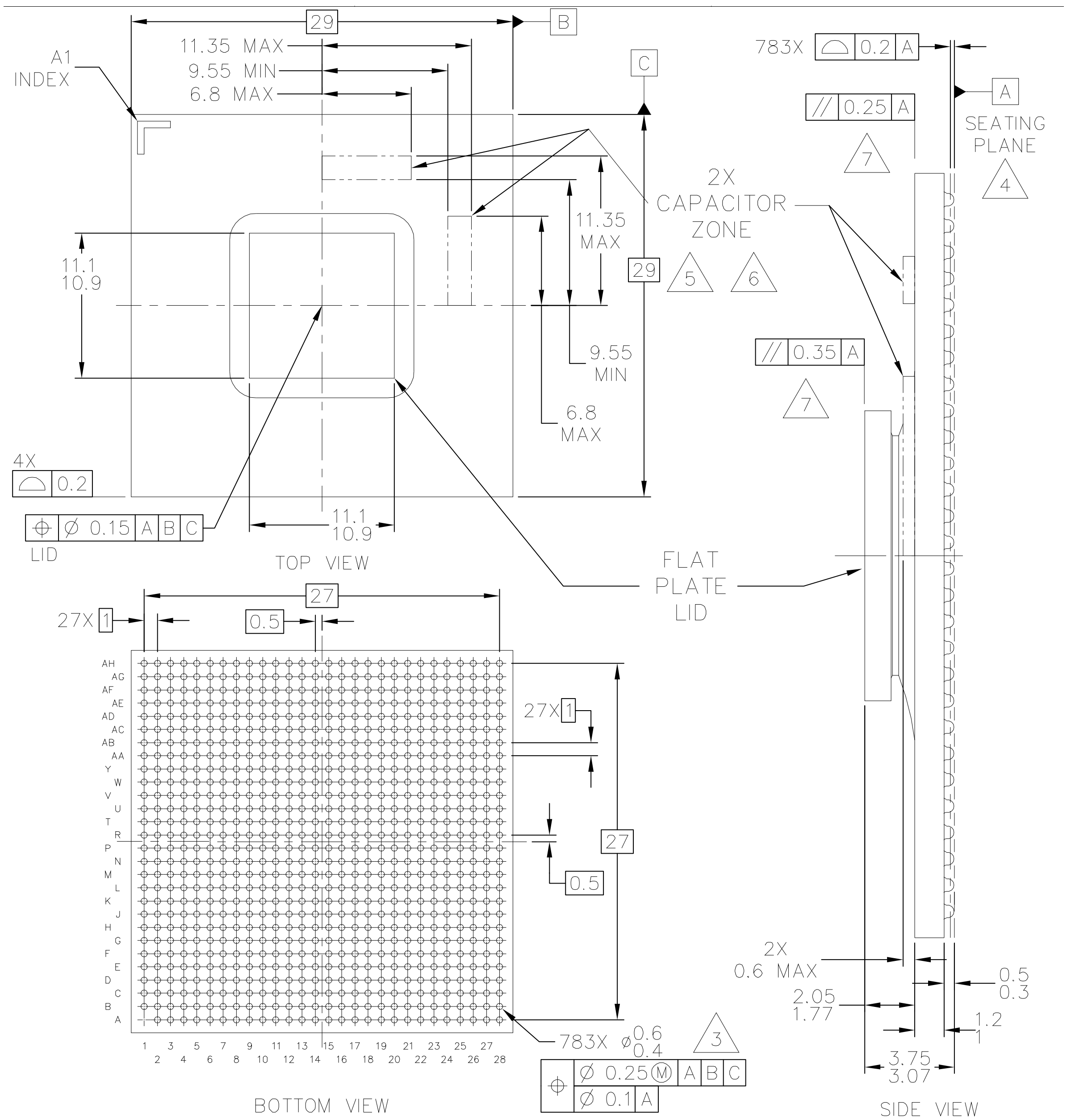

Figure 41. Mechanical Dimensions and Bottom Surface Nomenclature of the FC-PBGA

Notes:

1. All dimensions are in millimeters.
2. Dimensions and tolerances per ASME Y14.5M-1994.
3. Maximum solder ball diameter measured parallel to datum A.
4. Datum A, the seating plane, is defined by the spherical crowns of the solder balls.
5. Capacitors may not be present on all devices.
6. Caution must be taken not to short capacitors or exposed metal capacitor pads on package top.
7. The socket lid must always be oriented to A1.

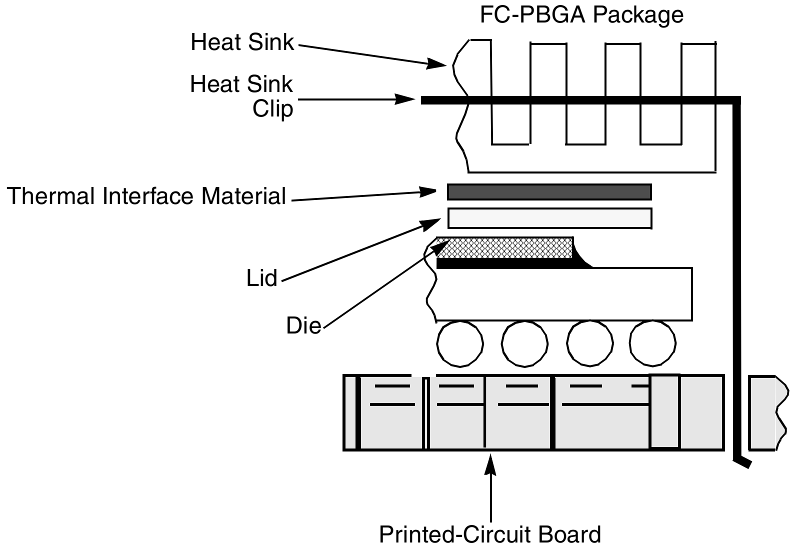

Figure 42. Package Exploded Cross-Sectional View with Several Heat Sink Options

The system board designer can choose between several types of heat sinks to place on the MPC8541E. There are several commercially-available heat sinks from the following vendors:

Aavid Thermalloy 80 Commercial St. Concord, NH 03301 Internet: www.aavidthermalloy.com	603-224-9988
Alpha Novatech 473 Sapena Ct. #15 Santa Clara, CA 95054 Internet: www.alphanovatech.com	408-749-7601
International Electronic Research Corporation (IERC) 413 North Moss St. Burbank, CA 91502 Internet: www.ctscorp.com	818-842-7277
Millennium Electronics (MEI) Loroco Sites 671 East Brokaw Road San Jose, CA 95112 Internet: www.mei-millennium.com	408-436-8770
Tyco Electronics Chip Coolers™ P.O. Box 3668 Harrisburg, PA 17105-3668 Internet: www.chipcoolers.com	800-522-6752
Wakefield Engineering 33 Bridge St. Pelham, NH 03076 Internet: www.wakefield.com	603-635-5102

Chanhassen, MN 55317
Internet: www.bergquistcompany.com

Thermagon Inc. 4707 Detroit Ave. Cleveland, OH 44102 Internet: www.thermagon.com	888-246-9050

16.2.4 Heat Sink Selection Examples

The following section provides a heat sink selection example using one of the commercially available heat sinks.

16.2.4.1 Case 1

For preliminary heat sink sizing, the die-junction temperature can be expressed as follows:

$$T_J = T_I + T_R + (\theta_{JC} + \theta_{INT} + \theta_{SA}) \times P_D$$

where

T_J is the die-junction temperature

T_I is the inlet cabinet ambient temperature

T_R is the air temperature rise within the computer cabinet

θ_{JC} is the junction-to-case thermal resistance

θ_{INT} is the adhesive or interface material thermal resistance

θ_{SA} is the heat sink base-to-ambient thermal resistance

P_D is the power dissipated by the device. See Table 4 and Table 5.

During operation the die-junction temperatures (T_J) should be maintained within the range specified in Table 2. The temperature of air cooling the component greatly depends on the ambient inlet air temperature and the air temperature rise within the electronic cabinet. An electronic cabinet inlet-air temperature (T_A) may range from 30° to 40°C. The air temperature rise within a cabinet (T_R) may be in the range of 5° to 10°C. The thermal resistance of some thermal interface material (θ_{INT}) may be about 1°C/W. For the purposes of this example, the θ_{JC} value given in Table 49 that includes the thermal grease interface and is documented in note 4 is used. If a thermal pad is used, θ_{INT} must be added.

Assuming a T_I of 30°C, a T_R of 5°C, a FC-PBGA package θ_{JC} = 0.96, and a power consumption (P_D) of 8.0 W, the following expression for T_J is obtained:

$$\text{Die-junction temperature: } T_J = 30°C + 5°C + (0.96°C/W + \theta_{SA}) \times 8.0\ W$$

The heat sink-to-ambient thermal resistance (θ_{SA}) versus airflow velocity for a Thermalloy heat sink #2328B is shown in Figure 46.

Assuming an air velocity of 2 m/s, we have an effective θ_{SA+} of about 3.3°C/W, thus

$$T_J = 30°C + 5°C + (0.96°C/W + 3.3°C/W) \times 8.0\ W,$$

resulting in a die-junction temperature of approximately 69°C which is well within the maximum operating temperature of the component.

17.6 Configuration Pin Multiplexing

The MPC8541E provides the user with power-on configuration options which can be set through the use of external pull-up or pull-down resistors of 4.7 kΩ on certain output pins (see customer visible configuration pins). These pins are generally used as output only pins in normal operation.

While $\overline{\text{HRESET}}$ is asserted however, these pins are treated as inputs. The value presented on these pins while $\overline{\text{HRESET}}$ is asserted, is latched when $\overline{\text{HRESET}}$ deasserts, at which time the input receiver is disabled and the I/O circuit takes on its normal function. Most of these sampled configuration pins are equipped with an on-chip gated resistor of approximately 20 kΩ. This value should permit the 4.7-kΩ resistor to pull the configuration pin to a valid logic low level. The pull-up resistor is enabled only during $\overline{\text{HRESET}}$ (and for platform/system clocks after $\overline{\text{HRESET}}$ deassertion to ensure capture of the reset value). When the input receiver is disabled the pull-up is also, thus allowing functional operation of the pin as an output with minimal signal quality or delay disruption. The default value for all configuration bits treated this way has been encoded such that a high voltage level puts the device into the default state and external resistors are needed only when non-default settings are required by the user.

Careful board layout with stubless connections to these pull-down resistors coupled with the large value of the pull-down resistor should minimize the disruption of signal quality or speed for output pins thus configured.

The platform PLL ratio and e500 PLL ratio configuration pins are not equipped with these default pull-up devices.

17.7 Pull-Up Resistor Requirements

The MPC8541E requires high resistance pull-up resistors (10 kΩ is recommended) on open drain type pins.

Correct operation of the JTAG interface requires configuration of a group of system control pins as demonstrated in Figure 52. Care must be taken to ensure that these pins are maintained at a valid deasserted state under normal operating conditions as most have asynchronous behavior and spurious assertion give unpredictable results.

TSEC1_TXD[3:0] must not be pulled low during reset. Some PHY chips have internal pulldowns that could cause this to happen. If such PHY chips are used, then a pullup must be placed on these signals strong enough to restore these signals to a logical 1 during reset.

Refer to the PCI 2.2 specification for all pull-ups required for PCI.

17.8 JTAG Configuration Signals

Boundary-scan testing is enabled through the JTAG interface signals. The $\overline{\text{TRST}}$ signal is optional in the IEEE 1149.1 specification, but is provided on all processors that implement the Power Architecture. The device requires $\overline{\text{TRST}}$ to be asserted during reset conditions to ensure the JTAG boundary logic does not interfere with normal chip operation. While it is possible to force the TAP controller to the reset state using only the TCK and TMS signals, generally systems assert $\overline{\text{TRST}}$ during the power-on reset flow. Simply tying $\overline{\text{TRST}}$ to $\overline{\text{HRESET}}$ is not practical because the JTAG interface is also used for accessing the common on-chip processor (COP) function.

17.8.1 Termination of Unused Signals

If the JTAG interface and COP header are not used, Freescale recommends the following connections:

- $\overline{\text{TRST}}$ should be tied to $\overline{\text{HRESET}}$ through a 0 kΩ isolation resistor so that it is asserted when the system reset signal ($\overline{\text{HRESET}}$) is asserted, ensuring that the JTAG scan chain is initialized during the power-on reset flow. Freescale recommends that the COP header be designed into the system as shown in Figure 52. If this is not possible, the isolation resistor allows future access to $\overline{\text{TRST}}$ in case a JTAG interface may need to be wired onto the system in future debug situations.
- Tie TCK to OV_{DD} through a 10 kΩ resistor. This prevents TCK from changing state and reading incorrect data into the device.
- No connection is required for TDI, TMS, or TDO.

18 Document Revision History

Table 51 provides a revision history for this hardware specification.

Table 51. Document Revision History

Rev. No.	Date	Substantive Change(s)
4.2	1/2008	Added "Note: Rise/Fall Time on CPM Input Pins" and following note text to Section 10.2, "CPM AC Timing Specifications."
4.1	07/2007	Inserted Figure 3, ""Maximum AC Waveforms on PCI interface for 3.3-V Signaling."
4	12/2006	Updated Section 2.1.2, "Power Sequencing." Updated back page information.
3.2	11/2006	Updated Section 2.1.2, "Power Sequencing." Replaced Section 17.8, "JTAG Configuration Signals."
3.1	10/2005	Table 4: Added footnote 2 about junction temperature. Table 4: Added max. power values for 1000 MHz core frequency. Removed Figure 3, "Maximum AC Waveforms on PCI Interface for 3.3-V Signaling." Table 30: Modified note to $t_{LBKSKEW}$ from 8 to 9 Table 30: Changed $t_{LBKHOZ1}$ and $t_{LBKHOV2}$ values. Table 30: Added note 3 to $t_{LBKHOV1}$. Table 30 and Table 31: Modified note 3. Table 31: Added note 3 to $t_{LBKLOV1}$. Table 31: Modified values for t_{LBKHKT}, $t_{LBKLOV1}$, $t_{LBKLOV2}$, $t_{LBKLOV3}$, $t_{LBKLOZ1}$, and $t_{LBKLOZ2}$. Figure 21: Changed Input Signals: LAD[0:31]/LDP[0:3]. Table 43: Modified note for signal CLK_OUT. Table 43: PCI1_CLK and PCI2_CLK changed from I/O to I. Table 52: Added column for Encryption Acceleration.
3	8/29/2005	Table 4: Modified max. power values. Table 43: Modified notes for signals TSEC1_TXD[3:0], TSEC2_TXD[3:0], TRIG_OUT/READY, MSRCID4, and MDVAL.
2	8/2005	Previous revision's history listed incorrect cross references. Table 2 is now correctly listed as Table 27 and Table 31 is now listed as Table 31. Table 7: Added note 2. Table 14: Modified min and max values for t_{DDKHMP}
1	6/2005	Table 27: Changed LV_{dd} to OV_{dd} for the supply voltage Ethernet management interface. Table 4: Modified footnote 4 and changed typical power for the 1000MHz core frequency. Table 31: Corrected symbols for body rows 9–15, effectively changing them from a high state to a low state.
0	6/2005	Initial Release.

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