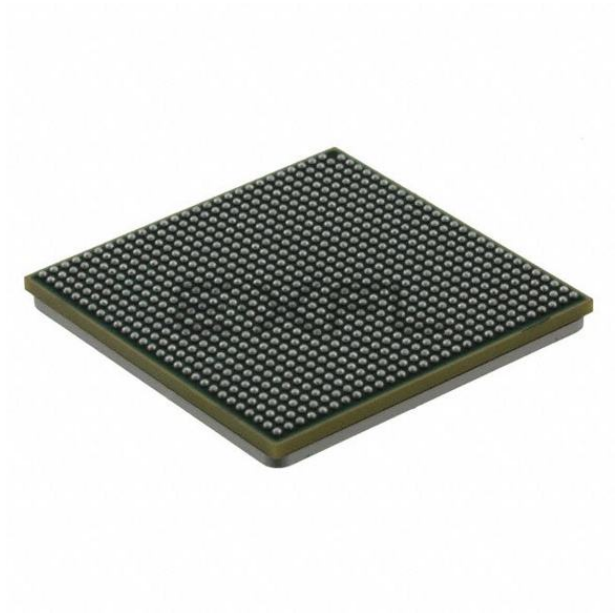

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Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Obsolete
Core Processor	PowerPC e500
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	667MHz
Co-Processors/DSP	Security; SEC
RAM Controllers	DDR, SDRAM
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100/1000Mbps (2)
SATA	-
USB	-
Voltage - I/O	2.5V, 3.3V
Operating Temperature	0°C ~ 105°C (TA)
Security Features	Cryptography, Random Number Generator
Package / Case	783-BBGA, FCBGA
Supplier Device Package	783-FCPBGA (29x29)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mpc8541evtalf

Figure 6 provides the AC test load for the DDR bus.

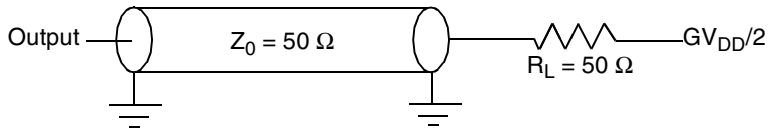


Figure 6. DDR AC Test Load

Table 15. DDR SDRAM Measurement Conditions

Symbol	DDR	Unit	Notes
V_{TH}	MV_{REF} ± 0.31 V	V	1
V_{OUT}	0.5 × GV_{DD}	V	2

Notes:
1. Data input threshold measurement point.
2. Data output measurement point.

7 DUART

This section describes the DC and AC electrical specifications for the DUART interface of the MPC8541E.

7.1 DUART DC Electrical Characteristics

Table 16 provides the DC electrical characteristics for the DUART interface of the MPC8541E.

Table 16. DUART DC Electrical Characteristics

Parameter	Symbol	Test Condition	Min	Max	Unit
High-level input voltage	V_{IH}	$V_{OUT} \geq V_{OH}$ (min) or	2	OV_{DD} + 0.3	V
Low-level input voltage	V_{IL}	$V_{OUT} \leq V_{OL}$ (max)	–0.3	0.8	V
Input current	I_{IN}	V_{IN} [1] = 0 V or $V_{IN} = V_{DD}$	—	±5	μA
High-level output voltage	V_{OH}	OV_{DD} = min, I_{OH} = –100 μA	OV_{DD} – 0.2	—	V
Low-level output voltage	V_{OL}	OV_{DD} = min, I_{OL} = 100 μA	—	0.2	V

Note:
1. Note that the symbol V_{IN}, in this case, represents the OV_{IN} symbol referenced in Table 1 and Table 2.

7.2 DUART AC Electrical Specifications

Table 17 provides the AC timing parameters for the DUART interface of the MPC8541E.

Table 17. DUART AC Timing Specifications

Parameter	Value	Unit	Notes
Minimum baud rate	f_{CCB_CLK} / 1048576	baud	3
Maximum baud rate	f_{CCB_CLK} / 16	baud	1, 3
Oversample rate	16	—	2, 3

Notes:
1. Actual attainable baud rate is limited by the latency of interrupt processing.
2. The middle of a start bit is detected as the 8^{th} sampled 0 after the 1-to-0 transition of the start bit. Subsequent bit values are sampled each 16^{th} sample.
3. Guaranteed by design.

8 Ethernet: Three-Speed, MII Management

This section provides the AC and DC electrical characteristics for three-speed, 10/100/1000, and MII management.

8.1 Three-Speed Ethernet Controller (TSEC) (10/100/1000 Mbps)—GMII/MII/TBI/RGMII/RTBI Electrical Characteristics

The electrical characteristics specified here apply to all GMII (gigabit media independent interface), the MII (media independent interface), TBI (ten-bit interface), RGMII (reduced gigabit media independent interface), and RTBI (reduced ten-bit interface) signals except MDIO (management data input/output) and MDC (management data clock). The RGMII and RTBI interfaces are defined for 2.5 V, while the GMII and TBI interfaces can be operated at 3.3 V or 2.5 V. Whether the GMII, MII, or TBI interface is operated at 3.3 or 2.5 V, the timing is compliant with the IEEE 802.3 standard. The RGMII and RTBI interfaces follow the Hewlett-Packard reduced pin-count interface for Gigabit Ethernet Physical Layer Device Specification Version 1.2a (9/22/2000). The electrical characteristics for MDIO and MDC are specified in Section 8.3, “Ethernet Management Interface Electrical Characteristics.”

8.1.1 TSEC DC Electrical Characteristics

All GMII, MII, TBI, RGMII, and RTBI drivers and receivers comply with the DC parametric attributes specified in Table 18 and Table 19. The potential applied to the input of a GMII, MII, TBI, RGMII, or RTBI receiver may exceed the potential of the receiver’s power supply (for example, a GMII driver powered from a 3.6-V supply driving V_{OH} into a GMII receiver powered from a 2.5-V supply). Tolerance for dissimilar GMII driver and receiver supply potentials is implicit in these specifications. The RGMII and RTBI signals are based on a 2.5-V CMOS interface voltage as defined by JEDEC EIA/JESD8-5.

8.2.2.1 GMII Receive AC Timing Specifications

Table 21 provides the GMII receive AC timing specifications.

Table 21. GMII Receive AC Timing Specifications

At recommended operating conditions with LV_{DD} of 3.3 V ± 5%.

Parameter/Condition	Symbol [1]	Min	Typ	Max	Unit
RX_CLK clock period	t_{GRX}	—	8.0	—	ns
RX_CLK duty cycle	t_{GRXH}/t_{GRX}	40	—	60	%
RXD[7:0], RX_DV, RX_ER setup time to RX_CLK	t_{GRDVKH}	2.0	—	—	ns
RXD[7:0], RX_DV, RX_ER hold time to RX_CLK	t_{GRDXKH}	0.5	—	—	ns
RX_CLK clock rise and fall time	t_{GRXR}, t_{GRXF} [2,3]	—	—	1.0	ns

Note:

1. The symbols used for timing specifications herein follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{GRDVKH} symbolizes GMII receive timing (GR) with respect to the time data input signals (D) reaching the valid state (V) relative to the t_{RX} clock reference (K) going to the high state (H) or setup time. Also, t_{GRDXKL} symbolizes GMII receive timing (GR) with respect to the time data input signals (D) went invalid (X) relative to the t_{GRX} clock reference (K) going to the low (L) state or hold time. Note that, in general, the clock reference symbol representation is based on three letters representing the clock of a particular functional. For example, the subscript of t_{GRX} represents the GMII (G) receive (RX) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).
2. Signal timings are measured at 0.7 V and 1.9 V voltage levels.
3. Guaranteed by design.

Figure 8 provides the AC test load for TSEC.

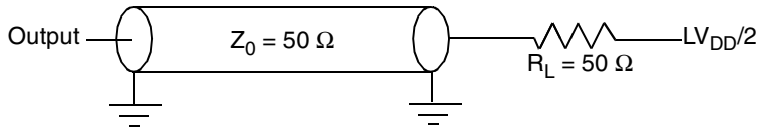


Figure 8. TSEC AC Test Load

Figure 9 shows the GMII receive AC timing diagram.

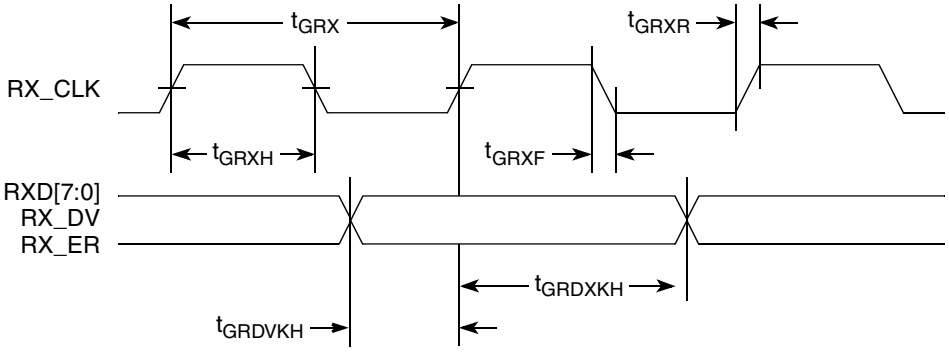


Figure 9. GMII Receive AC Timing Diagram

Figure 14 shows the RBMII and RTBI AC timing and multiplexing diagrams.

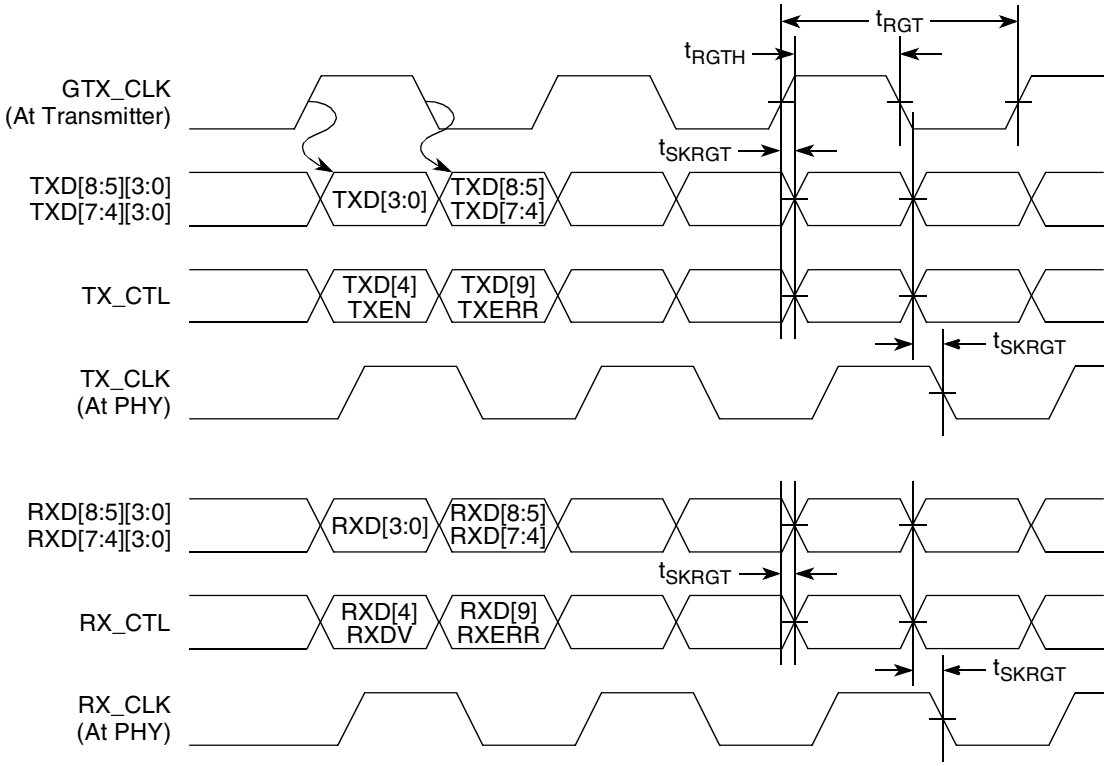


Figure 14. RGMII and RTBI AC Timing and Multiplexing Diagrams

8.3 Ethernet Management Interface Electrical Characteristics

The electrical characteristics specified here apply to MII management interface signals MDIO (management data input/output) and MDC (management data clock). The electrical characteristics for GMII, RGMII, TBI and RTBI are specified in Section 8.1, "Three-Speed Ethernet Controller (TSEC) (10/100/1000 Mbps)—GMII/MII/TBI/RGMII/RTBI Electrical Characteristics."

8.3.1 MII Management DC Electrical Characteristics

The MDC and MDIO are defined to operate at a supply voltage of 3.3 V. The DC electrical characteristics for MDIO and MDC are provided in Table 27.

Table 27. MII Management DC Electrical Characteristics

Parameter	Symbol	Conditions		Min	Max	Unit
Supply voltage (3.3 V)	OV_{DD}	—		3.13	3.47	V
Output high voltage	V_{OH}	I_{OH} = –1.0 mA	LV_{DD} = Min	2.10	LV_{DD} + 0.3	V
Output low voltage	V_{OL}	I_{OL} = 1.0 mA	LV_{DD} = Min	GND	0.50	V
Input high voltage	V_{IH}	—		1.70	—	V
Input low voltage	V_{IL}	—		—	0.90	V

9.2 Local Bus AC Electrical Specifications

Table 30 describes the general timing parameters of the local bus interface of the MPC8541E with the DLL enabled.

Table 30. Local Bus General Timing Parameters—DLL Enabled

Parameter	Configuration [7]	Symbol [1]	Min	Max	Unit	Notes
Local bus cycle time		t_{LBK}	6.0	—	ns	2
LCLK[n] skew to LCLK[m] or LSYNC_OUT		$t_{LBKSKEW}$	—	150	ps	7, 9
Input setup to local bus clock (except LUPWAIT)		$t_{LBIVKH1}$	1.8	—	ns	3, 4, 8
LUPWAIT input setup to local bus clock		$t_{LBIVKH2}$	1.7	—	ns	3, 4
Input hold from local bus clock (except LUPWAIT)		$t_{LBIXKH1}$	0.5	—	ns	3, 4, 8
LUPWAIT input hold from local bus clock		$t_{LBIXKH2}$	1.0	—	ns	3, 4
LALE output transition to LAD/LDP output transition (LATCH hold time)		t_{LBOTOT}	1.5	—	ns	6
Local bus clock to output valid (except LAD/LDP and LALE)	$\overline{LWE[0:1]}$ = 00	$t_{LBKHOV1}$	—	2.3	ns	3, 8
	$\overline{LWE[0:1]}$ = 11 (default)			3.8		
Local bus clock to data valid for LAD/LDP	$\overline{LWE[0:1]}$ = 00	$t_{LBKHOV2}$	—	2.5	ns	3, 8
	$\overline{LWE[0:1]}$ = 11 (default)			4.0		
Local bus clock to address valid for LAD	$\overline{LWE[0:1]}$ = 00	$t_{LBKHOV3}$	—	2.6	ns	3, 8
	$\overline{LWE[0:1]}$ = 11 (default)			4.1		
Output hold from local bus clock (except LAD/LDP and LALE)	$\overline{LWE[0:1]}$ = 00	$t_{LBKHOX1}$	0.7	—	ns	3, 8
	$\overline{LWE[0:1]}$ = 11 (default)		1.6			
Output hold from local bus clock for LAD/LDP	$\overline{LWE[0:1]}$ = 00	$t_{LBKHOX2}$	0.7	—	ns	3, 8
	$\overline{LWE[0:1]}$ = 11 (default)		1.6			
Local bus clock to output high Impedance (except LAD/LDP and LALE)	$\overline{LWE[0:1]}$ = 00	$t_{LBKHOZ1}$	—	2.8	ns	5, 9
	$\overline{LWE[0:1]}$ = 11 (default)			4.2		

Figure 17 to Figure 22 show the local bus signals.

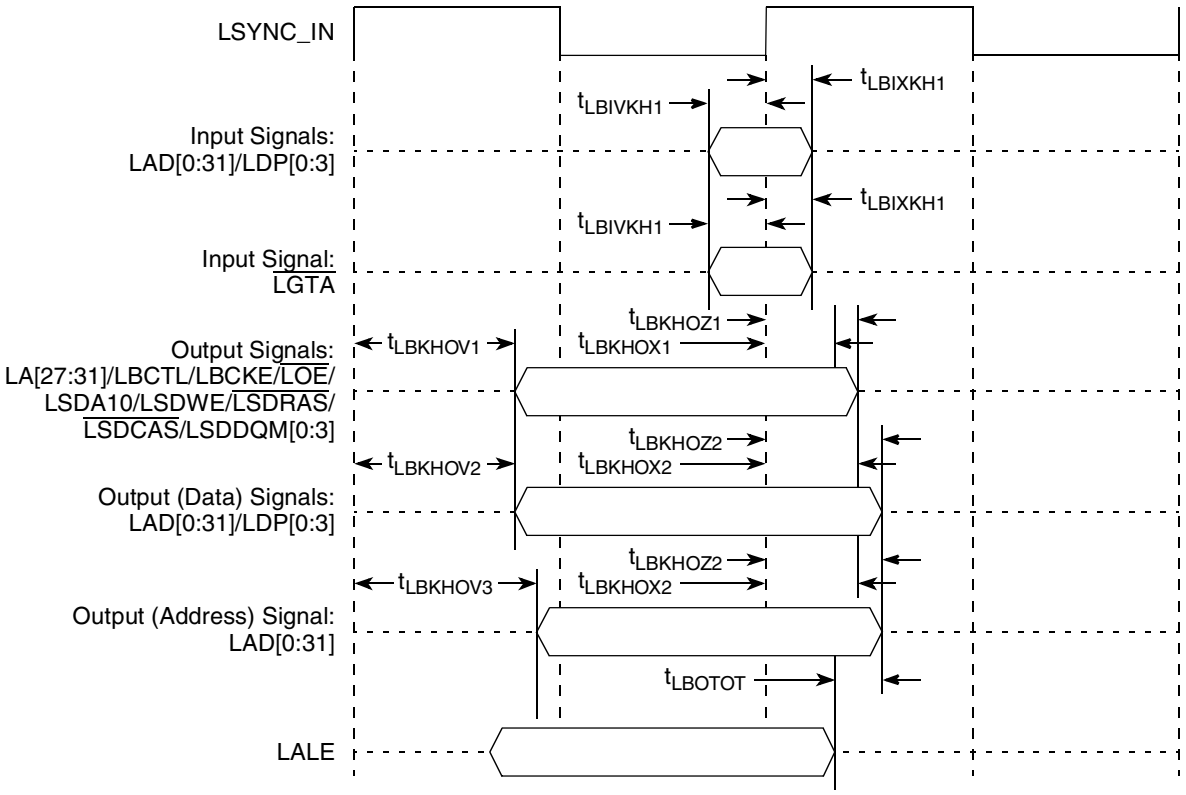


Figure 17. Local Bus Signals, Nonspecial Signals Only (DLL Enabled)

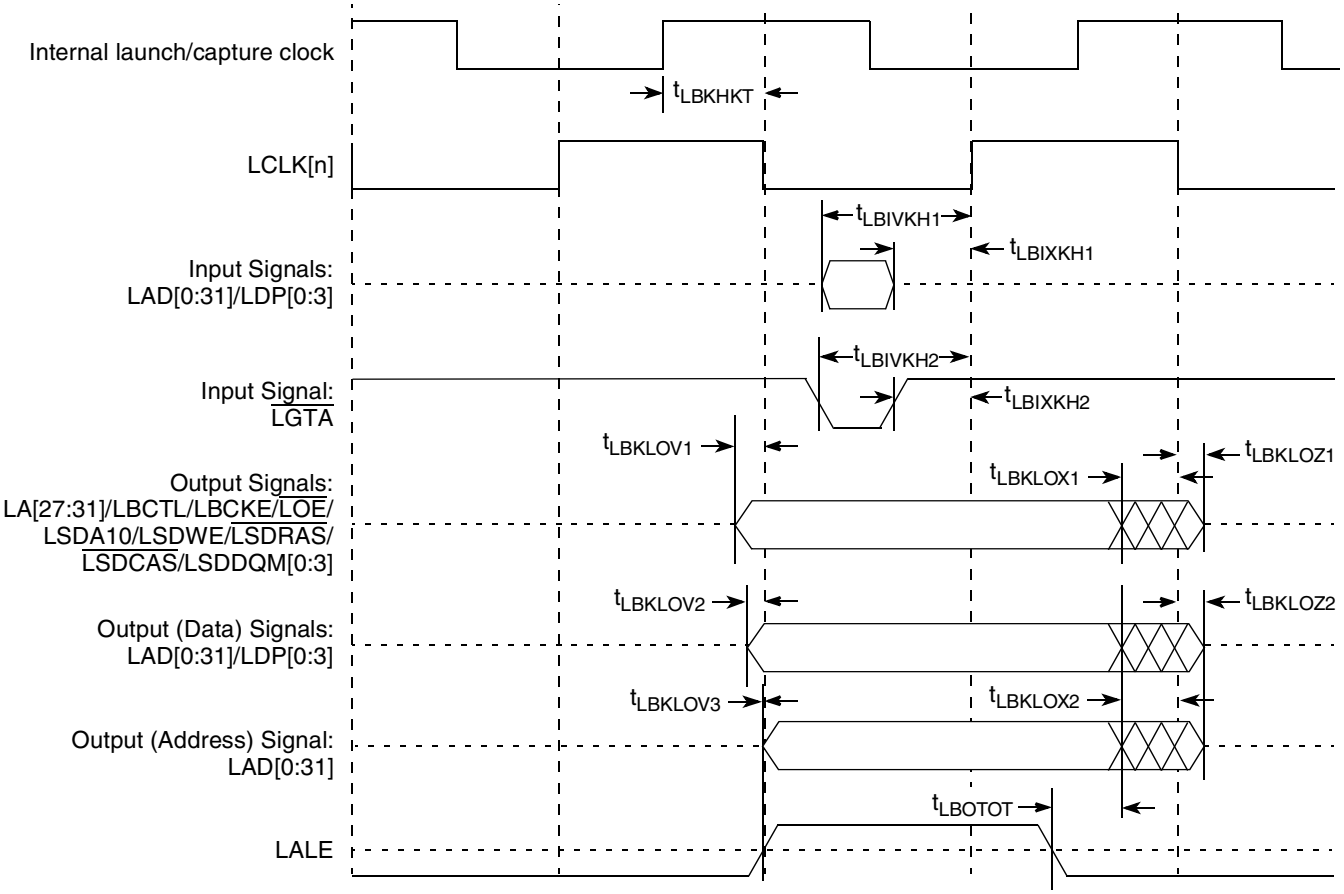


Figure 18. Local Bus Signals, Nonspecial Signals Only (DLL Bypass Mode)

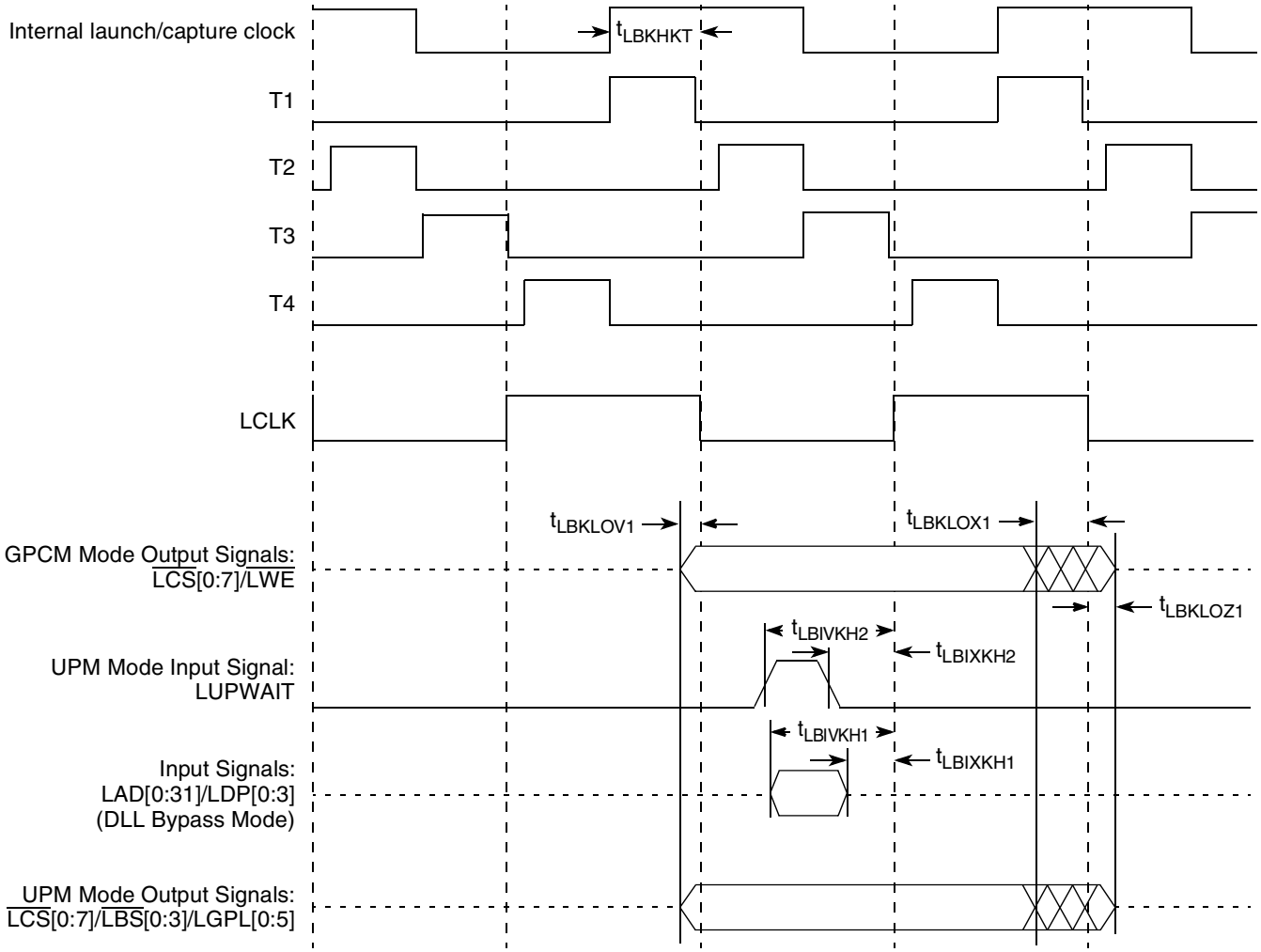


Figure 22. Local Bus Signals, GPCM/UPM Signals for LCCR[CLKDIV] = 4 or 8 (DLL Bypass Mode)

Table 33. CPM Input AC Timing Specifications [1] (continued)

Characteristic	Symbol [2]	Min[3]	Unit
PIO inputs—input hold time	t_{PIIXKH}	1	ns
COL width high (FCC)	t_{FCCH}	1.5	CLK

Notes:
1. Input specifications are measured from the 50% level of the signal to the 50% level of the rising edge of CLKIN. Timings are measured at the pin.
2. The symbols used for timing specifications herein follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{FIIVKH} symbolizes the FCC inputs internal timing (FI) with respect to the time the input signals (I) reaching the valid state (V) relative to the reference clock t_{FCC} (K) going to the high (H) state or setup time.
3. PIO and TIMER inputs and outputs are asynchronous to SYSCLK or any other externally visible clock. PIO/TIMER inputs are internally synchronized to the CPM internal clock. PIO/TIMER outputs should be treated as asynchronous.

Table 34. CPM Output AC Timing Specifications [1]

Characteristic	Symbol [2]	Min	Max	Unit
FCC outputs—internal clock (NMSI) delay	t_{FIKHOX}	1	5.5	ns
FCC outputs—external clock (NMSI) delay	t_{FEKHOX}	2	8	ns
SPI outputs—internal clock (NMSI) delay	t_{NIKHOX}	0.5	10	ns
SPI outputs—external clock (NMSI) delay	t_{NEKHOX}	2	8	ns
PIO outputs delay	t_{PIKHOX}	1	11	ns

Notes:
1. Output specifications are measured from the 50% level of the rising edge of CLKIN to the 50% level of the signal. Timings are measured at the pin.
2. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state) (reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{FIKHOX} symbolizes the FCC inputs internal timing (FI) for the time t_{FCC} memory clock reference (K) goes from the high state (H) until outputs (O) are invalid (X).

Figure 23 provides the AC test load for the CPM.

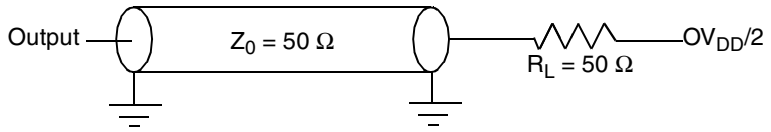


Figure 23. CPM AC Test Load

14.3 Pinout Listings

Table 43 provides the pin-out listing for the MPC8541E, 783 FC-PBGA package.

Table 43. MPC8541E Pinout Listing

Signal	Package Pin Number	Pin Type	Power Supply	Notes
PCI1 and PCI2 (one 64-bit or two 32-bit)				
PCI1_AD[63:32], PCI2_AD[31:0]	AA14, AB14, AC14, AD14, AE14, AF14, AG14, AH14, V15, W15, Y15, AA15, AB15, AC15, AD15, AG15, AH15, V16, W16, AB16, AC16, AD16, AE16, AF16, V17, W17, Y17, AA17, AB17, AE17, AF17, AF18	I/O	OV_{DD}	17
PCI1_AD[31:0]	AH6, AD7, AE7, AH7, AB8, AC8, AF8, AG8, AD9, AE9, AF9, AG9, AH9, W10, Y10, AA10, AE11, AF11, AG11, AH11, V12, W12, Y12, AB12, AD12, AE12, AG12, AH12, V13, Y13, AB13, AC13	I/O	OV_{DD}	17
$\overline{\text{PCI_C_BE64}}$[7:4] $\overline{\text{PCI2_C_BE}}$[3:0]	AG13, AH13, V14, W14	I/O	OV_{DD}	17
$\overline{\text{PCI_C_BE64}}$[3:0] $\overline{\text{PCI1_C_BE}}$[3:0]	AH8, AB10, AD11, AC12	I/O	OV_{DD}	17
PCI1_PAR	AA11	I/O	OV_{DD}	—
PCI1_PAR64/PCI2_PAR	Y14	I/O	OV_{DD}	—
$\overline{\text{PCI1_FRAME}}$	AC10	I/O	OV_{DD}	2
$\overline{\text{PCI1_TRDY}}$	AG10	I/O	OV_{DD}	2
$\overline{\text{PCI1_IRDY}}$	AD10	I/O	OV_{DD}	2
$\overline{\text{PCI1_STOP}}$	V11	I/O	OV_{DD}	2
$\overline{\text{PCI1_DEVSEL}}$	AH10	I/O	OV_{DD}	2
PCI1_IDSEL	AA9	I	OV_{DD}	—
$\overline{\text{PCI1_REQ64}}$/$\overline{\text{PCI2_FRAME}}$	AE13	I/O	OV_{DD}	5, 10
$\overline{\text{PCI1_ACK64}}$/$\overline{\text{PCI2_DEVSEL}}$	AD13	I/O	OV_{DD}	2
$\overline{\text{PCI1_PERR}}$	W11	I/O	OV_{DD}	2
$\overline{\text{PCI1_SERR}}$	Y11	I/O	OV_{DD}	2, 4
$\overline{\text{PCI1_REQ}}$[0]	AF5	I/O	OV_{DD}	—
$\overline{\text{PCI1_REQ}}$[1:4]	AF3, AE4, AG4, AE5	I	OV_{DD}	—
$\overline{\text{PCI1_GNT}}$[0]	AE6	I/O	OV_{DD}	—
$\overline{\text{PCI1_GNT}}$[1:4]	AG5, AH5, AF6, AG6	O	OV_{DD}	5, 9
PCI1_CLK	AH25	I	OV_{DD}	—
PCI2_CLK	AH27	I	OV_{DD}	—
$\overline{\text{PCI2_GNT}}$[0]	AC18	I/O	OV_{DD}	—

Table 43. MPC8541E Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
TSEC2_CRS	D9	I	LV_{DD}	—
TSEC2_COL	F8	I	LV_{DD}	—
TSEC2_RXD[7:0]	F9, E9, C9, B9, A9, H9, G10, F10	I	LV_{DD}	—
TSEC2_RX_DV	H8	I	LV_{DD}	—
TSEC2_RX_ER	A8	I	LV_{DD}	—
TSEC2_RX_CLK	E10	I	LV_{DD}	—
DUART				
$\overline{\text{UART\_CTS}}$[0,1]	Y2, Y3	I	OV_{DD}	—
$\overline{\text{UART\_RTS}}$[0,1]	Y1, AD1	O	OV_{DD}	—
UART_SIN[0,1]	P11, AD5	I	OV_{DD}	—
UART_SOUT[0,1]	N6, AD2	O	OV_{DD}	—
I^2C interface				
IIC_SDA	AH22	I/O	OV_{DD}	4, 19
IIC_SCL	AH23	I/O	OV_{DD}	4, 19
System Control				
$\overline{\text{HRESET}}$	AH16	I	OV_{DD}	—
$\overline{\text{HRESET\_REQ}}$	AG20	O	OV_{DD}	18
$\overline{\text{SRESET}}$	AF20	I	OV_{DD}	—
$\overline{\text{CKSTP\_IN}}$	M11	I	OV_{DD}	—
$\overline{\text{CKSTP\_OUT}}$	G1	O	OV_{DD}	2, 4
Debug				
TRIG_IN	N12	I	OV_{DD}	—
TRIG_OUT/READY	G2	O	OV_{DD}	6, 9, 18
MSRCID[0:1]	J9, G3	O	OV_{DD}	5, 6, 9
MSRCID[2:3]	F3, F5	O	OV_{DD}	6
MSRCID4	F2	O	OV_{DD}	6
MDVAL	F4	O	OV_{DD}	6
Clock				
SYSCLK	AH21	I	OV_{DD}	—
RTC	AB23	I	OV_{DD}	—
CLK_OUT	AF22	O	OV_{DD}	—

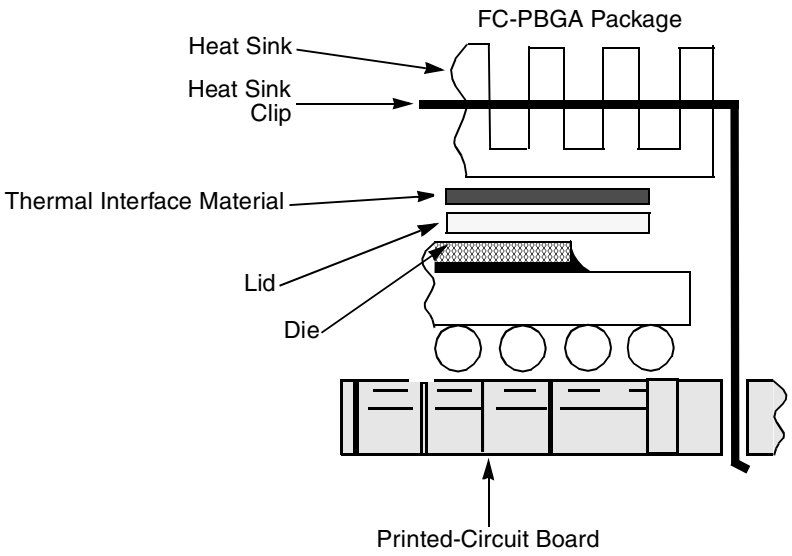


Figure 42. Package Exploded Cross-Sectional View with Several Heat Sink Options

The system board designer can choose between several types of heat sinks to place on the MPC8541E. There are several commercially-available heat sinks from the following vendors:

Aavid Thermalloy 80 Commercial St. Concord, NH 03301 Internet: www.aavidthermalloy.com	603-224-9988
Alpha Novatech 473 Sapena Ct. #15 Santa Clara, CA 95054 Internet: www.alphanovatech.com	408-749-7601
International Electronic Research Corporation (IERC) 413 North Moss St. Burbank, CA 91502 Internet: www.ctscorp.com	818-842-7277
Millennium Electronics (MEI) Loroco Sites 671 East Brokaw Road San Jose, CA 95112 Internet: www.mei-millennium.com	408-436-8770
Tyco Electronics Chip Coolers™ P.O. Box 3668 Harrisburg, PA 17105-3668 Internet: www.chipcoolers.com	800-522-6752
Wakefield Engineering 33 Bridge St. Pelham, NH 03076 Internet: www.wakefield.com	603-635-5102

The spring mounting should be designed to apply the force only directly above the die. By localizing the force, rocking of the heat sink is minimized. One suggested mounting method attaches a plastic fence to the board to provide the structure on which the heat sink spring clips. The plastic fence also provides the opportunity to minimize the holes in the printed-circuit board and to locate them at the corners of the package. Figure 47 and provide exploded views of the plastic fence, heat sink, and spring clip.

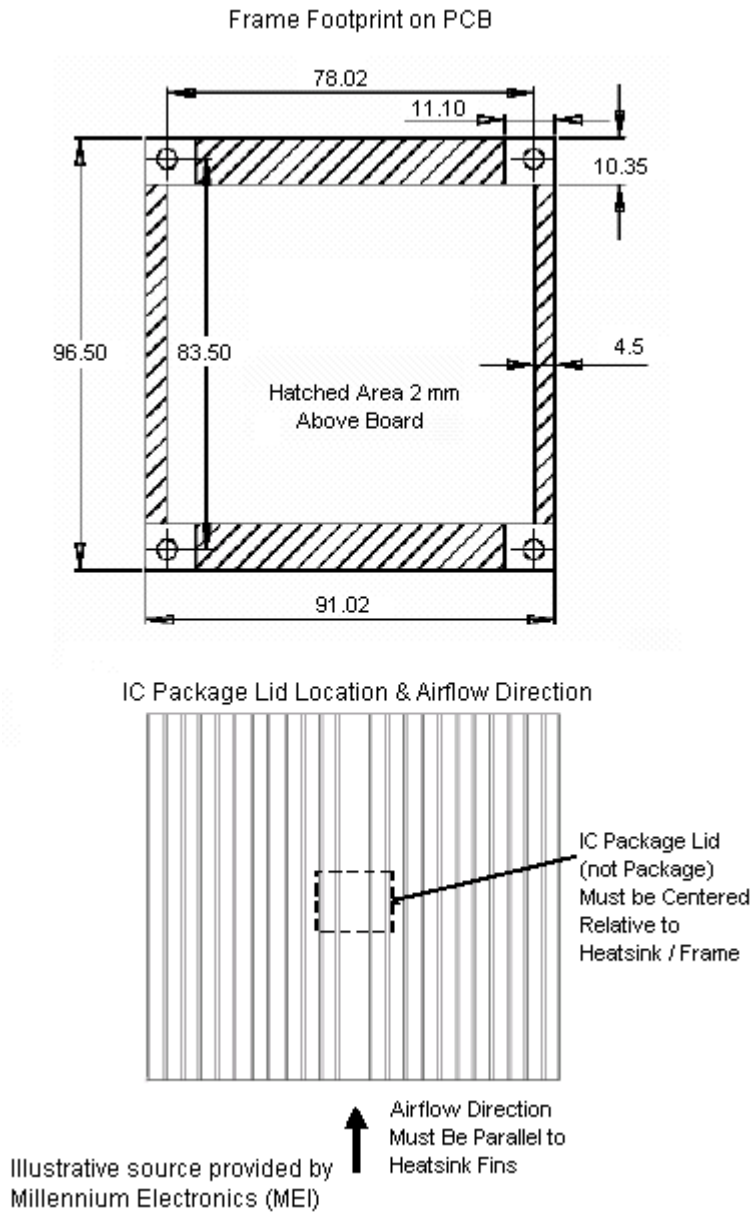
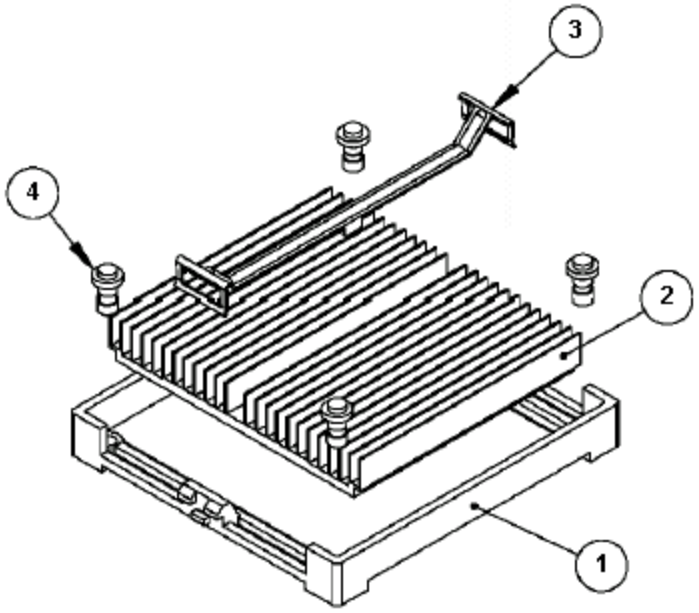


Figure 47. Exploded Views (1) of a Heat Sink Attachment using a Plastic Fence

Item No	QTY	MEI PN	Description
1	1	MFRAME-2000	HEATSINK FRAME
2	1	MSNK-1120	EXTRUDED HEATSINK
3	1	MCLIP-1013	CLIP
4	4	MPPINS-1000	FRAME ATTACHMENT PINS



Illustrative source provided by
Millennium Electronics (MEI)

Figure 48. Exploded Views (2) of a Heat Sink Attachment using a Plastic Force

The die junction-to-ambient and the heat sink-to-ambient thermal resistances are common figure-of-merits used for comparing the thermal performance of various microelectronic packaging technologies, one should exercise caution when only using this metric in determining thermal management because no single parameter can adequately describe three-dimensional heat flow. The final die-junction operating temperature is not only a function of the component-level thermal resistance, but the system level design and its operating conditions. In addition to the component's power consumption, a number of factors affect the final operating die-junction temperature: airflow, board population (local heat flux of adjacent components), system air temperature rise, altitude, etc.

Due to the complexity and the many variations of system-level boundary conditions for today's microelectronic equipment, the combined effects of the heat transfer mechanisms (radiation convection and conduction) may vary widely. For these reasons, we recommend using conjugate heat transfer models for the boards, as well as, system-level designs.

Figure 49 shows the PLL power supply filter circuit.

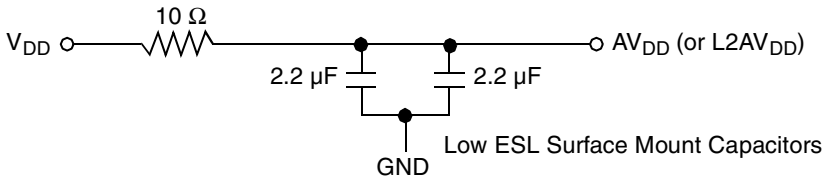


Figure 49. PLL Power Supply Filter Circuit

17.3 Decoupling Recommendations

Due to large address and data buses, and high operating frequencies, the MPC8541E can generate transient power surges and high frequency noise in its power supply, especially while driving large capacitive loads. This noise must be prevented from reaching other components in the MPC8541E system, and the MPC8541E itself requires a clean, tightly regulated source of power. Therefore, it is recommended that the system designer place at least one decoupling capacitor at each V_{DD}, OV_{DD}, GV_{DD}, and LV_{DD} pins of the MPC8541E. These decoupling capacitors should receive their power from separate V_{DD}, OV_{DD}, GV_{DD}, LV_{DD}, and GND power planes in the PCB, utilizing short traces to minimize inductance. Capacitors may be placed directly under the device using a standard escape pattern. Others may surround the part.

These capacitors should have a value of 0.01 or 0.1 µF. Only ceramic SMT (surface mount technology) capacitors should be used to minimize lead inductance, preferably 0402 or 0603 sizes.

In addition, it is recommended that there be several bulk storage capacitors distributed around the PCB, feeding the V_{DD}, OV_{DD}, GV_{DD}, and LV_{DD} planes, to enable quick recharging of the smaller chip capacitors. These bulk capacitors should have a low ESR (equivalent series resistance) rating to ensure the quick response time necessary. They should also be connected to the power and ground planes through two vias to minimize inductance. Suggested bulk capacitors—100–330 µF (AVX TPS tantalum or Sanyo OSCON).

17.4 Connection Recommendations

To ensure reliable operation, it is highly recommended to connect unused inputs to an appropriate signal level. Unused active low inputs should be tied to OV_{DD}, GV_{DD}, or LV_{DD} as required. Unused active high inputs should be connected to GND. All NC (no-connect) signals must remain unconnected.

Power and ground connections must be made to all external V_{DD}, GV_{DD}, LV_{DD}, OV_{DD}, and GND pins of the MPC8541E.

17.5 Output Buffer DC Impedance

The MPC8541E drivers are characterized over process, voltage, and temperature. For all buses, the driver is a push-pull single-ended driver type (open drain for I^2C).

To measure Z_0 for the single-ended drivers, an external resistor is connected from the chip pad to OV_{DD} or GND. Then, the value of each resistor is varied until the pad voltage is $OV_{DD}/2$ (see Figure 50). The output impedance is the average of two components, the resistances of the pull-up and pull-down devices.

When data is held high, SW1 is closed (SW2 is open) and R_P is trimmed until the voltage at the pad equals $OV_{DD}/2$. R_P then becomes the resistance of the pull-up devices. R_P and R_N are designed to be close to each other in value. Then, $Z_0 = (R_P + R_N)/2$.

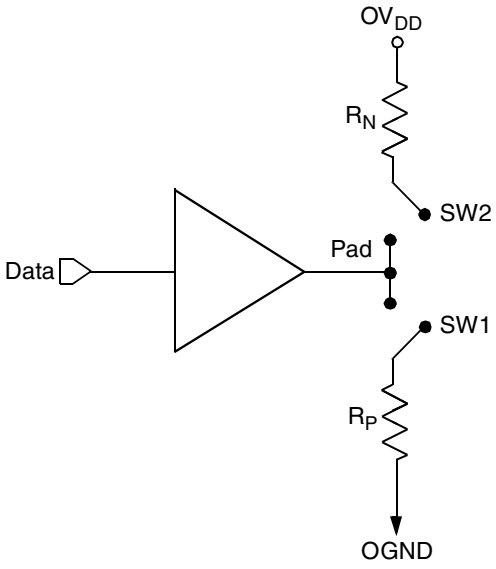


Figure 50. Driver Impedance Measurement

The value of this resistance and the strength of the driver's current source can be found by making two measurements. First, the output voltage is measured while driving logic 1 without an external differential termination resistor. The measured voltage is $V_1 = R_{source} \times I_{source}$. Second, the output voltage is measured while driving logic 1 with an external precision differential termination resistor of value R_{term}. The measured voltage is $V_2 = 1/(1/R_1 + 1/R_2)) \times I_{source}$. Solving for the output impedance gives $R_{source} = R_{term} \times (V_1/V_2 - 1)$. The drive current is then $I_{source} = V_1/R_{source}$.

Table 50 summarizes the signal impedance targets. The driver impedance are targeted at minimum V_{DD}, nominal OV_{DD}, 105°C.

Table 50. Impedance Characteristics

Impedance	Local Bus, Ethernet, DUART, Control, Configuration, Power Management	PCI	DDR DRAM	Symbol	Unit
R_N	43 Target	25 Target	20 Target	Z_0	Ω
R_P	43 Target	25 Target	20 Target	Z_0	Ω
Differential	NA	NA	NA	Z_{DIFF}	Ω

Note: Nominal supply voltages. See Table 1, T_j = 105°C.

The COP function of these processors allow a remote computer system (typically, a PC with dedicated hardware and debugging software) to access and control the internal operations of the processor. The COP interface connects primarily through the JTAG port of the processor, with some additional status monitoring signals. The COP port requires the ability to independently assert $\overline{\text{HRESET}}$ or $\overline{\text{TRST}}$ in order to fully control the processor. If the target system has independent reset sources, such as voltage monitors, watchdog timers, power supply failures, or push-button switches, then the COP reset signals must be merged into these signals with logic.

The arrangement shown in Figure 51 allows the COP port to independently assert $\overline{\text{HRESET}}$ or $\overline{\text{TRST}}$, while ensuring that the target can drive $\overline{\text{HRESET}}$ as well.

The COP interface has a standard header, shown in Figure 51, for connection to the target system, and is based on the 0.025" square-post, 0.100" centered header assembly (often called a Berg header). The connector typically has pin 14 removed as a connector key.

The COP header adds many benefits such as breakpoints, watchpoints, register and memory examination/modification, and other standard debugger features. An inexpensive option can be to leave the COP header unpopulated until needed.

There is no standardized way to number the COP header; consequently, many different pin numbers have been observed from emulator vendors. Some are numbered top-to-bottom then left-to-right, while others use left-to-right then top-to-bottom, while still others number the pins counter clockwise from pin 1 (as with an IC). Regardless of the numbering, the signal placement recommended in Figure 51 is common to all known emulators.

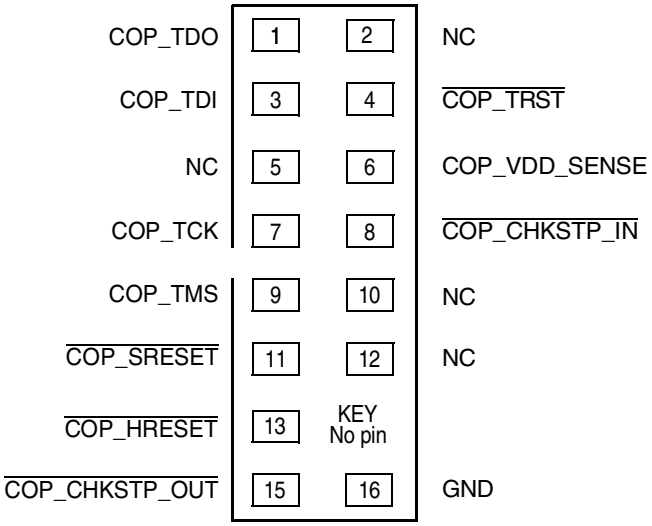
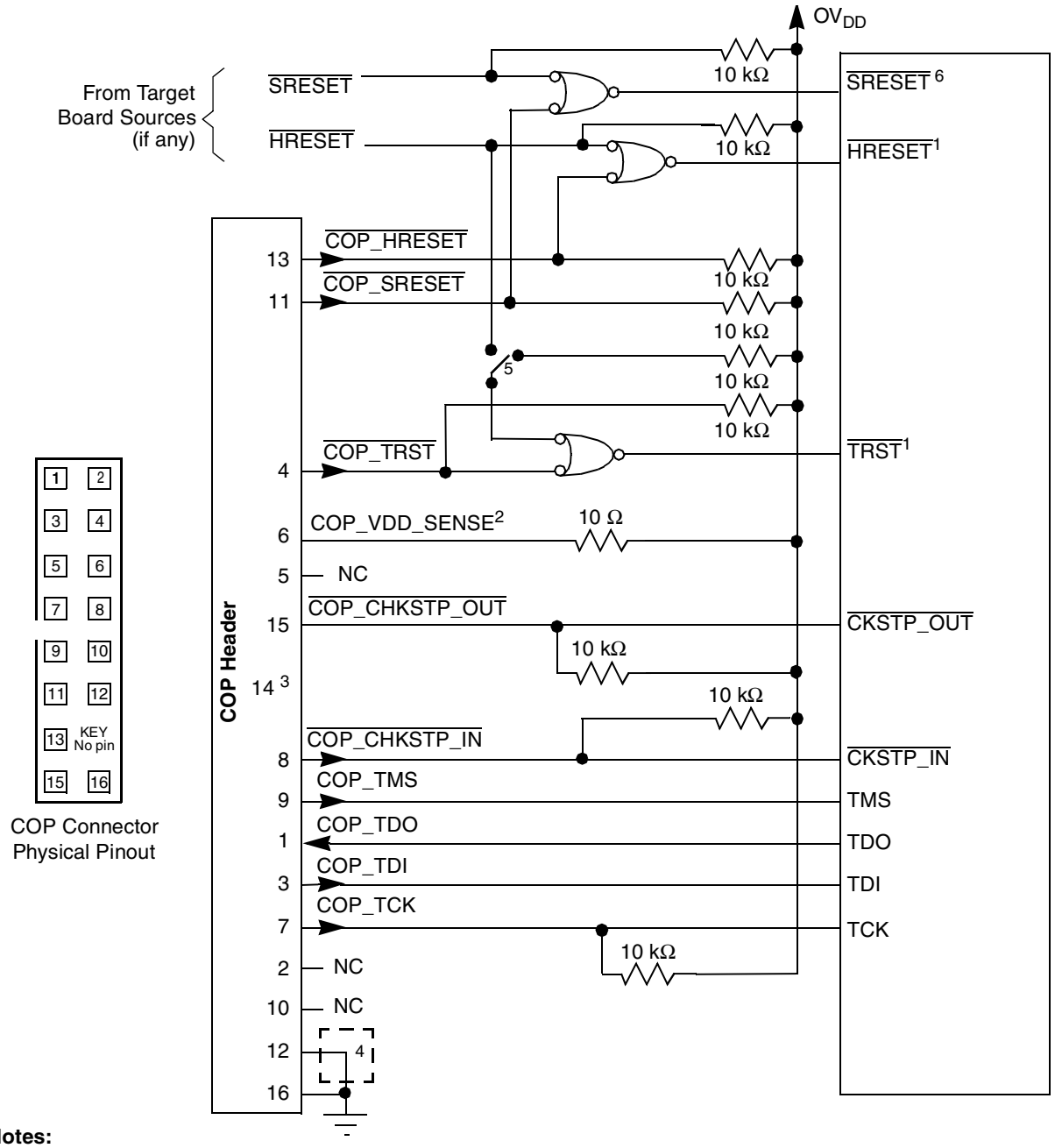


Figure 51. COP Connector Physical Pinout



Notes:

1. The COP port and target board should be able to independently assert $\overline{\text{HRESET}}$ and $\overline{\text{TRST}}$ to the processor in order to fully control the processor as shown here.
2. Populate this with a 10 Ω resistor for short-circuit/current-limiting protection.
3. The KEY location (pin 14) is not physically present on the COP header.
4. Although pin 12 is defined as a No-Connect, some debug tools may use pin 12 as an additional GND pin for improved signal integrity.
5. This switch is included as a precaution for BSDL testing. The switch should be open during BSDL testing to avoid accidentally asserting the $\overline{\text{TRST}}$ line. If BSDL testing is not being performed, this switch should be closed or removed.
6. Asserting $\overline{\text{SRESET}}$ causes a machine check interrupt to the e500 core.

Figure 52. JTAG Interface Connection

18 Document Revision History

Table 51 provides a revision history for this hardware specification.

Table 51. Document Revision History

Rev. No.	Date	Substantive Change(s)
4.2	1/2008	Added “Note: Rise/Fall Time on CPM Input Pins” and following note text to Section 10.2, “CPM AC Timing Specifications.”
4.1	07/2007	Inserted Figure 3, ““Maximum AC Waveforms on PCI interface for 3.3-V Signaling.”
4	12/2006	Updated Section 2.1.2, “Power Sequencing.” Updated back page information.
3.2	11/2006	Updated Section 2.1.2, “Power Sequencing.” Replaced Section 17.8, “JTAG Configuration Signals.”
3.1	10/2005	Table 4: Added footnote 2 about junction temperature. Table 4: Added max. power values for 1000 MHz core frequency. Removed Figure 3, “Maximum AC Waveforms on PCI Interface for 3.3-V Signaling.” Table 30: Modified note to $t_{LBKSKEW}$ from 8 to 9 Table 30: Changed $t_{LBKHOZ1}$ and $t_{LBKHOV2}$ values. Table 30: Added note 3 to $t_{LBKHOV1}$. Table 30 and Table 31: Modified note 3. Table 31: Added note 3 to $t_{LBKLOV1}$. Table 31: Modified values for t_{LBKHKT}, $t_{LBKLOV1}$, $t_{LBKLOV2}$, $t_{LBKLOV3}$, $t_{LBKLOZ1}$, and $t_{LBKLOZ2}$. Figure 21: Changed Input Signals: LAD[0:31]/LDP[0:3]. Table 43: Modified note for signal CLK_OUT. Table 43: PCI1_CLK and PCI2_CLK changed from I/O to I. Table 52: Added column for Encryption Acceleration.
3	8/29/2005	Table 4: Modified max. power values. Table 43: Modified notes for signals TSEC1_TXD[3:0], TSEC2_TXD[3:0], TRIG_OUT/READY, MSRCID4, and MDVAL.
2	8/2005	Previous revision’s history listed incorrect cross references. Table 2 is now correctly listed as Table 27 and Table 31 is now listed as Table 31. Table 7: Added note 2. Table 14: Modified min and max values for t_{DDKHMP}
1	6/2005	Table 27: Changed LV_{dd} to OV_{dd} for the supply voltage Ethernet management interface. Table 4: Modified footnote 4 and changed typical power for the 1000MHz core frequency. Table 31: Corrected symbols for body rows 9–15, effectively changing them from a high state to a low state.
0	6/2005	Initial Release.

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