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Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	ARM® Cortex®-M3
Core Size	32-Bit Single-Core
Speed	120MHz
Connectivity	CANbus, I²C, IrDA, LINbus, MMC, SPI, UART/USART, USB OTG
Peripherals	Brown-out Detect/Reset, DMA, I²S, LCD, POR, PWM, WDT
Number of I/O	51
Program Memory Size	256KB (256K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	100K x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.6V
Data Converters	A/D 16x12b; D/A 2x12b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 105°C (TA)
Mounting Type	Surface Mount
Package / Case	64-LQFP
Supplier Device Package	64-LQFP (10x10)
Purchase URL	https://www.e-xfl.com/product-detail/stmicroelectronics/stm32f205rct7

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There are three power modes configured by software when the regulator is ON:

- MR is used in the nominal regulation mode
- LPR is used in Stop modes

  The LP regulator mode is configured by software when entering Stop mode.
- Power-down is used in Standby mode.

  The Power-down mode is activated only when entering Standby mode. The regulator output is in high impedance and the kernel circuitry is powered down, inducing zero consumption. The contents of the registers and SRAM are lost).

Two external ceramic capacitors should be connected on V_{CAP_1} and V_{CAP_2} pin. Refer to *Figure 19: Power supply scheme* and *Table 16: VCAP1/VCAP2 operating conditions*.

All packages have the regulator ON feature.

3.16.2 Regulator OFF

This feature is available only on packages featuring the REGOFF pin. The regulator is disabled by holding REGOFF high. The regulator OFF mode allows to supply externally a V12 voltage source through V_{CAP_1} and V_{CAP_2} pins.

The two 2.2 µF ceramic capacitors should be replaced by two 100 nF decoupling capacitors. Refer to *Figure 19: Power supply scheme*.

When the regulator is OFF, there is no more internal monitoring on V12. An external power supply supervisor should be used to monitor the V12 of the logic power domain. PA0 pin should be used for this purpose, and act as power-on reset on V12 power domain.

In regulator OFF mode, the following features are no more supported:

- PA0 cannot be used as a GPIO pin since it allows to reset the part of the 1.2 V logic power domain which is not reset by the NRST pin.
- As long as PA0 is kept low, the debug mode cannot be used at power-on reset. As a consequence, PA0 and NRST pins must be managed separately if the debug connection at reset or pre-reset is required.

Regulator OFF/internal reset ON

On WLCSP64+2 package, this mode is activated by connecting REGOFF pin to V_{DD} and IRROFF pin to V_{SS}. On UFBGA176 package, only REGOFF must be connected to V_{DD} (IRROFF not available). In this mode, V_{DD}/V_{DDA} minimum value is 1.8 V.

The regulator OFF/internal reset ON mode allows to supply externally a 1.2 V voltage source through V_{CAP_1} and V_{CAP_2} pins, in addition to V_{DD}.

The backup registers are 32-bit registers used to store 80 bytes of user application data when V_{DD} power is not present. Backup registers are not reset by a system, a power reset, or when the device wakes up from the Standby mode (see *Section 3.18: Low-power modes*).

Like backup SRAM, the RTC and backup registers are supplied through a switch that is powered either from the V_{DD} supply when present or the V_{BAT} pin.

3.18 Low-power modes

The STM32F20x family supports three low-power modes to achieve the best compromise between low-power consumption, short startup time and available wakeup sources:

- **Sleep mode**

  In Sleep mode, only the CPU is stopped. All peripherals continue to operate and can wake up the CPU when an interrupt/event occurs.

- **Stop mode**

  The Stop mode achieves the lowest power consumption while retaining the contents of SRAM and registers. All clocks in the 1.2 V domain are stopped, the PLL, the HSI RC and the HSE crystal oscillators are disabled. The voltage regulator can also be put either in normal or in low-power mode.

  The device can be woken up from the Stop mode by any of the EXTI line. The EXTI line source can be one of the 16 external lines, the PVD output, the RTC alarm / wakeup / tamper / time stamp events, the USB OTG FS/HS wakeup or the Ethernet wakeup.

- **Standby mode**

  The Standby mode is used to achieve the lowest power consumption. The internal voltage regulator is switched off so that the entire 1.2 V domain is powered off. The PLL, the HSI RC and the HSE crystal oscillators are also switched off. After entering Standby mode, the SRAM and register contents are lost except for registers in the backup domain and the backup SRAM when selected.

  The device exits the Standby mode when an external reset (NRST pin), an IWDG reset, a rising edge on the WKUP pin, or an RTC alarm / wakeup / tamper /time stamp event occurs.

Note: The RTC, the IWDG, and the corresponding clock sources are not stopped when the device enters the Stop or Standby mode.

3.19 V_{BAT} operation

The V_{BAT} pin allows to power the device V_{BAT} domain from an external battery or an external supercapacitor.

V_{BAT} operation is activated when V_{DD} is not present.

The V_{BAT} pin supplies the RTC, the backup registers and the backup SRAM.

Note: When the microcontroller is supplied from V_{BAT}, external interrupts and RTC alarm/events do not exit it from V_{BAT} operation.

When using WLCSP64+2 package, if IRROFF pin is connected to V_{DD}, the V_{BAT} functionality is no more available and V_{BAT} pin should be connected to V_{DD}.

3.20 Timers and watchdogs

The STM32F20x devices include two advanced-control timers, eight general-purpose timers, two basic timers and two watchdog timers.

All timer counters can be frozen in debug mode.

Table 5 compares the features of the advanced-control, general-purpose and basic timers.

Table 5. Timer feature comparison

Timer type	Timer	Counter resolution	Counter type	Prescaler factor	DMA request generation	Capture/ compare channels	Complementary output	Max interface clock	Max timer clock
Advanced-control	TIM1, TIM8	16-bit	Up, Down, Up/down	Any integer between 1 and 65536	Yes	4	Yes	60 MHz	120 MHz
General purpose	TIM2, TIM5	32-bit	Up, Down, Up/down	Any integer between 1 and 65536	Yes	4	No	30 MHz	60 MHz
	TIM3, TIM4	16-bit	Up, Down, Up/down	Any integer between 1 and 65536	Yes	4	No	30 MHz	60 MHz
Basic	TIM6, TIM7	16-bit	Up	Any integer between 1 and 65536	Yes	0	No	30 MHz	60 MHz
General purpose	TIM9	16-bit	Up	Any integer between 1 and 65536	No	2	No	60 MHz	120 MHz
	TIM10, TIM11	16-bit	Up	Any integer between 1 and 65536	No	1	No	60 MHz	120 MHz
	TIM12	16-bit	Up	Any integer between 1 and 65536	No	2	No	30 MHz	60 MHz
	TIM13, TIM14	16-bit	Up	Any integer between 1 and 65536	No	1	No	30 MHz	60 MHz

3.20.1 Advanced-control timers (TIM1, TIM8)

The advanced-control timers (TIM1, TIM8) can be seen as three-phase PWM generators multiplexed on 6 channels. They have complementary PWM outputs with programmable inserted dead times. They can also be considered as complete general-purpose timers. Their 4 independent channels can be used for:

- Input capture
- Output compare
- PWM generation (edge- or center-aligned modes)
- One-pulse mode output

3.34 ADCs (analog-to-digital converters)

Three 12-bit analog-to-digital converters are embedded and each ADC shares up to 16 external channels, performing conversions in the single-shot or scan mode. In scan mode, automatic conversion is performed on a selected group of analog inputs.

Additional logic functions embedded in the ADC interface allow:

- Simultaneous sample and hold
- Interleaved sample and hold

The ADC can be served by the DMA controller. An analog watchdog feature allows very precise monitoring of the converted voltage of one, some or all selected channels. An interrupt is generated when the converted voltage is outside the programmed thresholds.

The events generated by the timers TIM1, TIM2, TIM3, TIM4, TIM5 and TIM8 can be internally connected to the ADC start trigger and injection trigger, respectively, to allow the application to synchronize A/D conversion and timers.

3.35 DAC (digital-to-analog converter)

The two 12-bit buffered DAC channels can be used to convert two digital signals into two analog voltage signal outputs. The design structure is composed of integrated resistor strings and an amplifier in inverting configuration.

This dual digital Interface supports the following features:

- two DAC converters: one for each output channel
- 8-bit or 12-bit monotonic output
- left or right data alignment in 12-bit mode
- synchronized update capability
- noise-wave generation
- triangular-wave generation
- dual DAC channel independent or simultaneous conversions
- DMA capability for each channel
- external triggers for conversion
- input voltage reference V_{REF+}

Eight DAC trigger inputs are used in the device. The DAC channels are triggered through the timer update outputs that are also connected to different DMA streams.

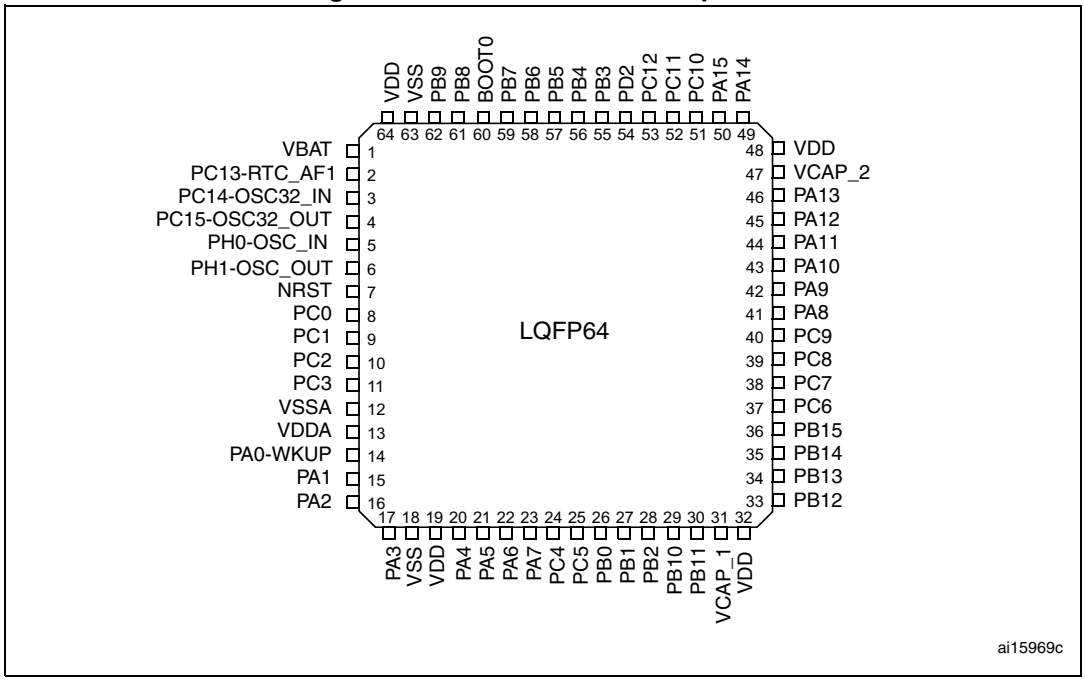
3.36 Temperature sensor

The temperature sensor has to generate a voltage that varies linearly with temperature. The conversion range is between 1.8 and 3.6 V. The temperature sensor is internally connected to the ADC1_IN16 input channel which is used to convert the sensor output voltage into a digital value.

As the offset of the temperature sensor varies from chip to chip due to process variation, the internal temperature sensor is mainly suitable for applications that detect temperature changes instead of absolute temperatures. If an accurate temperature reading is needed, then an external temperature sensor part should be used.

4 Pinouts and pin description

Figure 10. STM32F20x LQFP64 pinout



1. The above figure shows the package top view.

Figure 11. STM32F20x WLCSP64+2 ballout

	1	2	3	4	5	6	7	8	9
A	PA14	PA15	PC12	PB3	PB5	PB7	PB9	VDD	V_{BAT}
B	VSS	PA13	PC10	PB4	PB6	BOOT0	PB8	PC13	PC14
C	PA12	VCAP_2	PC11				PD2	IRROFF	PC15
D	PC9	PA11	PA10				PC2	VSS	VDD
E	VDD	PA8	PA9				PA0	NRST	PH0-OSC_IN
F	VSS	PC7	PC8				VREF+	PC1	PH1-OSC_OUT
G	PB15	PC6	PC5				PA3	PC3	PC0
H	PB14	PB13	PB10	PC4	PA6	PA5	REGOFF	PA1	VSS_5
J	PB12	PB11	VCAP_1	PB2	PB1	PB0	PA7	PA4	PA2

ai18470c

1. The above figure shows the package top view.

Figure 23. Typical current consumption vs. temperature, Run mode, code with data processing running from RAM, and peripherals ON

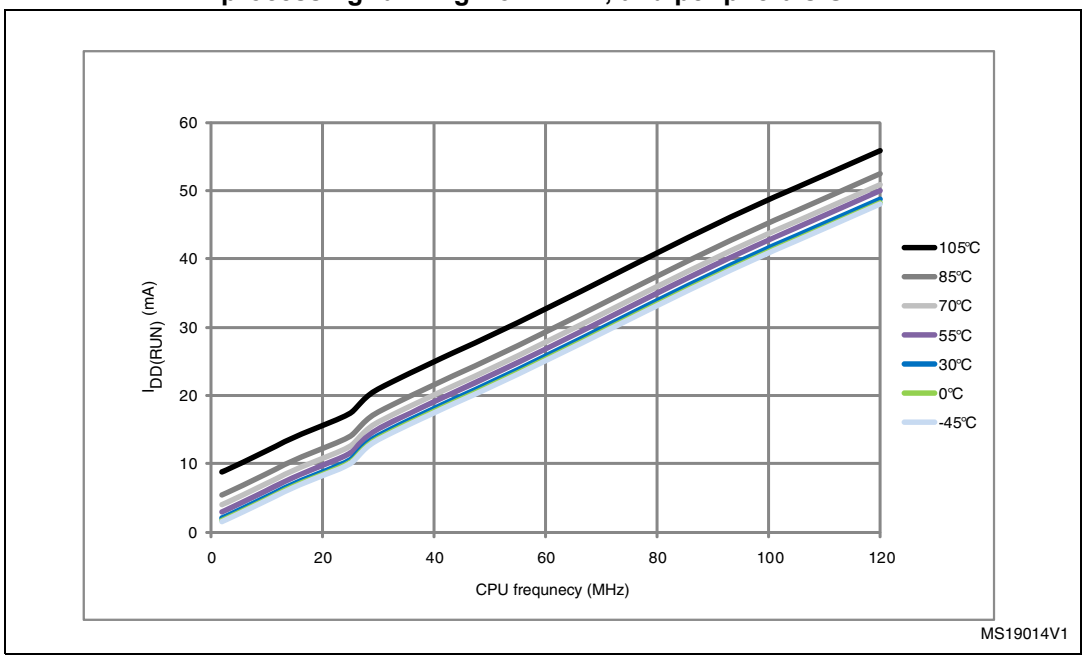


Figure 24. Typical current consumption vs. temperature, Run mode, code with data processing running from RAM, and peripherals OFF

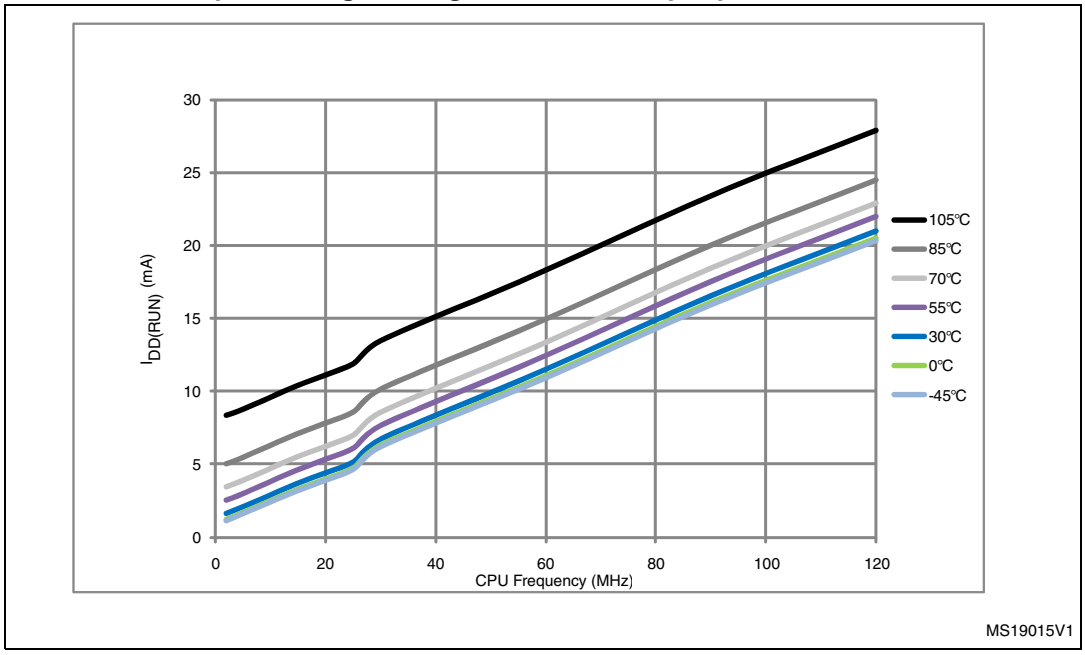


Table 24. Typical and maximum current consumptions in Standby mode

Symbol	Parameter	Conditions	Typ			Max(1)		Unit
			T_A = 25 °C			T_A = 85 °C	T_A = 105 °C	
			V_{DD} = 1.8 V	V_{DD}= 2.4 V	V_{DD} = 3.3 V	V_{DD} = 3.6 V		
I_{DD_STBY}	Supply current in Standby mode	Backup SRAM ON, low-speed oscillator and RTC ON	3.0	3.4	4.0	15.1	25.8	µA
		Backup SRAM OFF, low-speed oscillator and RTC ON	2.4	2.7	3.3	12.4	20.5	
		Backup SRAM ON, RTC OFF	2.4	2.6	3.0	12.5	24.8	
		Backup SRAM OFF, RTC OFF	1.7	1.9	2.2	9.8	19.2	

1. Guaranteed by characterization results, not tested in production.

Table 25. Typical and maximum current consumptions in V_{BAT} mode

Symbol	Parameter	Conditions	Typ			Max(1)		Unit
			T_A = 25 °C			T_A = 85 °C	T_A = 105 °C	
			V_{DD} = 1.8 V	V_{DD}= 2.4 V	V_{DD} = 3.3 V	V_{DD} = 3.6 V		
I_{DD_VBAT}	Backup domain supply current	Backup SRAM ON, low-speed oscillator and RTC ON	1.29	1.42	1.68	12	19	µA
		Backup SRAM OFF, low-speed oscillator and RTC ON	0.62	0.73	0.96	8	10	
		Backup SRAM ON, RTC OFF	0.79	0.81	0.86	9	16	
		Backup SRAM OFF, RTC OFF	0.10	0.10	0.10	5	7	

1. Guaranteed by characterization results, not tested in production.

On-chip peripheral current consumption

The current consumption of the on-chip peripherals is given in *Table 26*. The MCU is placed under the following conditions:

- At startup, all I/O pins are configured as analog inputs by firmware.
- All peripherals are disabled unless otherwise mentioned
- The given value is calculated by measuring the current consumption
  - with all peripherals clocked off
  - with one peripheral clocked on (with only the clock applied)
- The code is running from Flash memory and the Flash memory access time is equal to 3 wait states at 120 MHz
- Prefetch and Cache ON
- When the peripherals are enabled, HCLK = 120MHz, $f_{PCLK1} = f_{HCLK}/4$, and $f_{PCLK2} = f_{HCLK}/2$
- The typical values are obtained for V_{DD} = 3.3 V and T_A= 25 °C, unless otherwise specified.

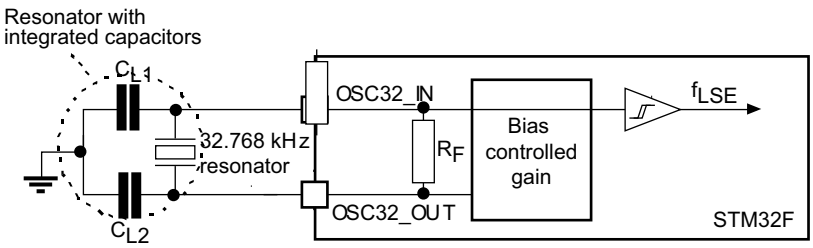
Table 31. LSE oscillator characteristics (f_{LSE} = 32.768 kHz) (1)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
R_F	Feedback resistor	-	-	18.4	-	MΩ
I_{DD}	LSE current consumption	-	-	-	1	µA
g_m	Oscillator Transconductance	-	2.8	-	-	µA/V
$t_{SU(LSE)}$ (2)	startup time	V_{DD} is stabilized	-	2	-	s

1. Guaranteed by design, not tested in production.
2. $t_{SU(LSE)}$ is the startup time measured from the moment it is enabled (by software) to a stabilized 32.768 kHz oscillation is reached. This value is measured for a standard crystal resonator and it can vary significantly with the crystal manufacturer

Note: For information on electing the crystal, refer to the application note AN2867 "Oscillator design guide for ST microcontrollers" available from the ST website www.st.com.

Figure 33. Typical application with a 32.768 kHz crystal



6.3.9 Internal clock source characteristics

The parameters given in *Table 32* and *Table 33* are derived from tests performed under ambient temperature and V_{DD} supply voltage conditions summarized in *Table 14*.

High-speed internal (HSI) RC oscillator

Table 32. HSI oscillator characteristics (1)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
f_{HSI}	Frequency	-	-	16	-	MHz
ACC_{HSI}	HSI user-trimming step(2)	-	-	-	1	%
	Accuracy of the HSI oscillator	T_A = –40 to 105 °C(3)	– 8	-	4.5	%
		T_A = –10 to 85 °C(3)	– 4	-	4	%
		T_A = 25 °C(4)	– 1	-	1	%
$t_{su(HSI)}$ (2)	HSI oscillator startup time	-	-	2.2	4.0	µs
$I_{DD(HSI)}$ (2)	HSI oscillator power consumption	-	-	60	80	µA

1. V_{DD} = 3.3 V, T_A = –40 to 105 °C unless otherwise specified.
2. Guaranteed by design, not tested in production.
3. Guaranteed by characterization results.
4. Factory calibrated, parts not soldered.

Table 35. PLLI2S (audio PLL) characteristics (continued)

Symbol	Parameter	Conditions		Min	Typ	Max	Unit
Jitter(3)	Master I2S clock jitter	Cycle to cycle at 12.288 MHz on 48KHz period, N=432, R=5	RMS	-	90	-	ps
			peak to peak	-	±280	-	
		Average frequency of 12.288 MHz N=432, R=5 on 1000 samples		-	90	-	ps
	WS I2S clock jitter	Cycle to cycle at 48 KHz on 1000 samples		-	400	-	ps
$I_{DD(PLLI2S)}$(4)	PLLI2S power consumption on V_{DD}	VCO freq = 192 MHz VCO freq = 432 MHz		0.15 0.45	-	0.40 0.75	mA
$I_{DDA(PLLI2S)}$(4)	PLLI2S power consumption on V_{DDA}	VCO freq = 192 MHz VCO freq = 432 MHz		0.30 0.55	-	0.40 0.85	mA

1. Take care of using the appropriate division factor M to have the specified PLL input clock values.
2. Guaranteed by design, not tested in production.
3. Value given with main PLL running.
4. Guaranteed by characterization results, not tested in production.

6.3.11 PLL spread spectrum clock generation (SSCG) characteristics

The spread spectrum clock generation (SSCG) feature allows to reduce electromagnetic interferences (see *Table 42: EMI characteristics*). It is available only on the main PLL.

Table 36. SSCG parameters constraint

Symbol	Parameter	Min	Typ	Max(1)	Unit
f_{Mod}	Modulation frequency	-	-	10	KHz
md	Peak modulation depth	0.25	-	2	%
MODEPER * INCSTEP	-	-	-	$2^{15}-1$	-

1. Guaranteed by design, not tested in production.

Equation 1

The frequency modulation period (MODEPER) is given by the equation below:

$$MODEPER = round[f_{PLL\_IN} / (4 \times f_{Mod})]$$

f_{PLL_IN} and f_{Mod} must be expressed in Hz.

As an example:

If f_{PLL_IN} = 1 MHz and f_{MOD} = 1 kHz, the modulation depth (MODEPER) is given by equation 1:

$$MODEPER = round[10^6 / (4 \times 10^3)] = 250$$

Equation 2

Equation 2 allows to calculate the increment step (INCSTEP):

$$INCSTEP = round[((2^{15} - 1) \times md \times PLLN) / (100 \times 5 \times MODEPER)]$$

f_{VCO_OUT} must be expressed in MHz.

With a modulation depth (md) = ±2 % (4 % peak to peak), and PLLN = 240 (in MHz):

$$INCSTEP = round[((2^{15} - 1) \times 2 \times 240) / (100 \times 5 \times 250)] = 126md(quantitazed)\%$$

An amplitude quantization error may be generated because the linear modulation profile is obtained by taking the quantized values (rounded to the nearest integer) of MODPER and INCSTEP. As a result, the achieved modulation depth is quantized. The percentage quantized modulation depth is given by the following formula:

$$md_{quantized}\% = (MODEPER \times INCSTEP \times 100 \times 5) / ((2^{15} - 1) \times PLLN)$$

As a result:

$$md_{quantized}\% = (250 \times 126 \times 100 \times 5) / ((2^{15} - 1) \times 240) = 2.0002\%(peak)$$

Table 48. I/O AC characteristics(1) (continued)

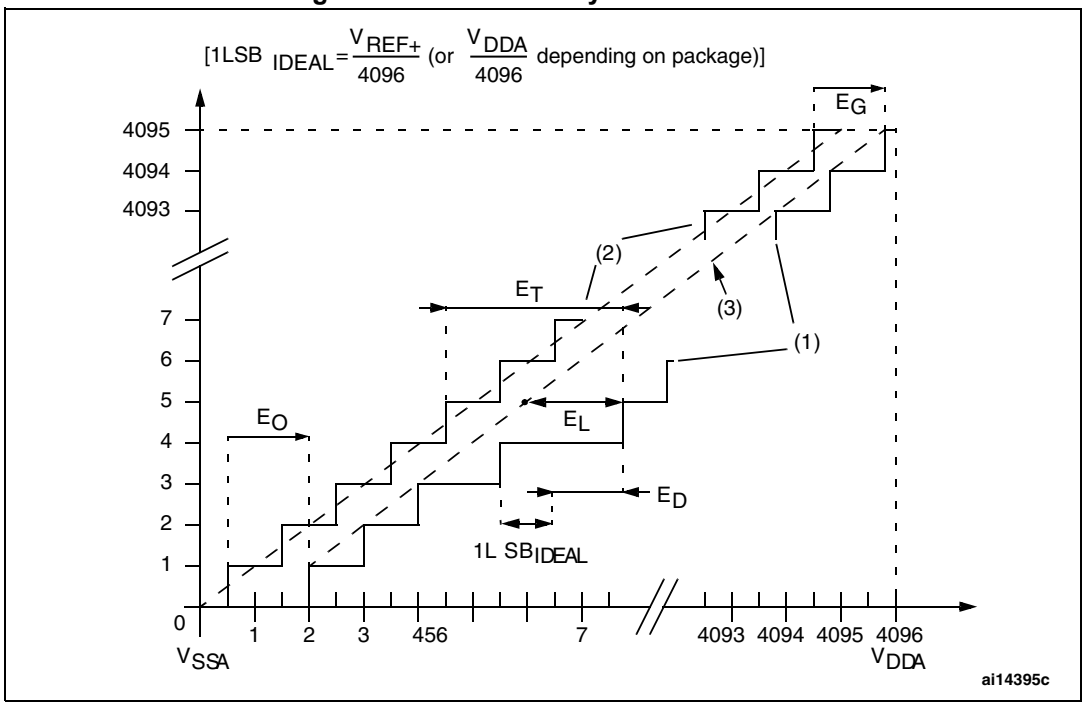
OSPEEDRy [1:0] bit value(1)	Symbol	Parameter	Conditions	Min	Typ	Max	Unit
01	$f_{max(IO)out}$	Maximum frequency(2)	C_L = 50 pF, V_{DD} > 2.70 V	-	-	25	MHz
			C_L = 50 pF, V_{DD} > 1.8 V	-	-	12.5	
			C_L = 10 pF, V_{DD} > 2.70 V	-	-	50(3)	
			C_L = 10 pF, V_{DD} > 1.8 V	-	-	20	
	$t_{f(IO)out}$/ $t_{r(IO)out}$	Output high to low level fall time and output low to high level rise time	C_L = 50 pF, V_{DD} >2.7 V	-	-	10	ns
			C_L = 50 pF, V_{DD} > 1.8 V	-	-	20	
			C_L = 10 pF, V_{DD} > 2.70 V	-	-	6	
			C_L = 10 pF, V_{DD} > 1.8 V	-	-	10	
10	$f_{max(IO)out}$	Maximum frequency(2)	C_L = 40 pF, V_{DD} > 2.70 V	-	-	25	MHz
			C_L = 40 pF, V_{DD} > 1.8 V	-	-	20	
			C_L = 10 pF, V_{DD} > 2.70 V	-	-	100(3)	
			C_L = 10 pF, V_{DD} > 1.8 V	-	-	50(3)	
	$t_{f(IO)out}$/ $t_{r(IO)out}$	Output high to low level fall time and output low to high level rise time	C_L = 40 pF, V_{DD} > 2.70 V	-	-	6	ns
			C_L = 40 pF, V_{DD} > 1.8 V	-	-	10	
			C_L = 10 pF, V_{DD} > 2.70 V	-	-	4	
			C_L = 10 pF, V_{DD} > 1.8 V	-	-3	6	
11	$f_{max(IO)out}$	Maximum frequency(2)	C_L = 30 pF, V_{DD} > 2.70 V	-	-	100(3)	MHz
			C_L = 30 pF, V_{DD} > 1.8 V	-	-	50(3)	
			C_L = 10 pF, V_{DD} > 2.70 V	-	-	120(3)	
			C_L = 10 pF, V_{DD} > 1.8 V	-	-	100(3)	
	$t_{f(IO)out}$/ $t_{r(IO)out}$	Output high to low level fall time and output low to high level rise time	C_L = 30 pF, V_{DD} > 2.70 V	-	-	4	ns
			C_L = 30 pF, V_{DD} > 1.8 V	-	-	6	
			C_L = 10 pF, V_{DD} > 2.70 V	-	-	2.5	
			C_L = 10 pF, V_{DD} > 1.8 V	-	-	4	
-	t_{EXTIpw}	Pulse width of external signals detected by the EXTI controller	-	10	-	-	ns

1. The I/O speed is configured using the OSPEEDRy[1:0] bits. Refer to the STM32F20/21xxx reference manual for a description of the GPIOx_SPEEDR GPIO port output speed register.
2. The maximum frequency is defined in *Figure 39*.
3. For maximum frequencies above 50 MHz and V_{DD} above 2.4 V, the compensation cell should be used.

being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative currents.

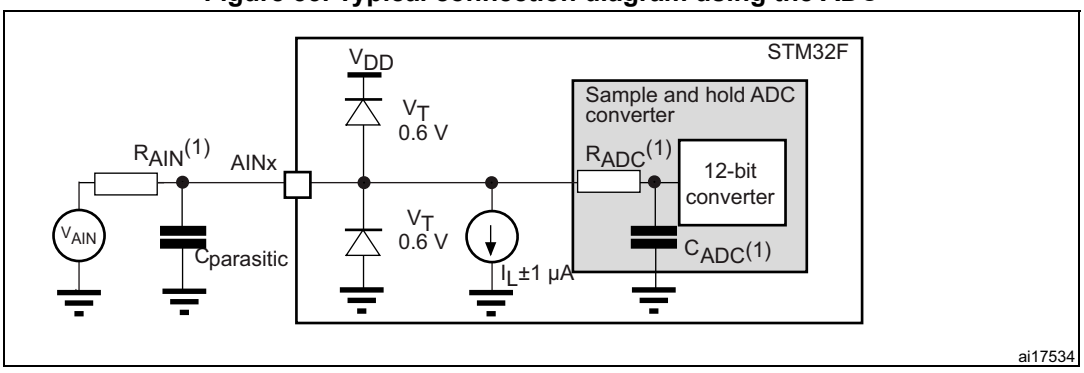
Any positive injection current within the limits specified for $I_{INJ(PIN)}$ and $\Sigma I_{INJ(PIN)}$ in *Section 6.3.16* does not affect the ADC accuracy.

Figure 52. ADC accuracy characteristics



1. Example of an actual transfer curve.
2. Ideal transfer curve.
3. End point correlation line.
4. E_T = Total Unadjusted Error: maximum deviation between the actual and the ideal transfer curves.
   EO = Offset Error: deviation between the first actual transition and the first ideal one.
   EG = Gain Error: deviation between the last ideal transition and the last actual one.
   ED = Differential Linearity Error: maximum deviation between actual steps and the ideal one.
   EL = Integral Linearity Error: maximum deviation between any actual transition and the end point correlation line.

Figure 53. Typical connection diagram using the ADC



1. Refer to *Table 66* for the values of R_{AIN}, R_{ADC} and C_{ADC}.
2. $C_{parasitic}$ represents the capacitance of the PCB (dependent on soldering and PCB layout quality) plus the pad capacitance (roughly 7 pF). A high $C_{parasitic}$ value downgrades conversion accuracy. To remedy this, f_{ADC} should be reduced.

Table 75. Asynchronous multiplexed PSRAM/NOR write timings(1)(2) (continued)

Symbol	Parameter	Min	Max	Unit
$t_{h(A\_NWE)}$	Address hold time after FSMC_NWE high	T_{HCLK}– 0.5	-	ns
$t_{h(BL\_NWE)}$	FSMC_BL hold time after FSMC_NWE high	T_{HCLK}- 1	-	ns
$t_{v(BL\_NE)}$	FSMC_NEx low to FSMC_BL valid	-	0.5	ns
$t_{v(Data\_NADV)}$	FSMC_NADV high to Data valid	-	T_{HCLK}+2	ns
$t_{h(Data\_NWE)}$	Data hold time after FSMC_NWE high	T_{HCLK}– 0.5	-	ns

1. C_L = 30 pF.
2. Guaranteed by characterization results, not tested in production.

Synchronous waveforms and timings

Figure 61 through *Figure 64* represent synchronous waveforms, and *Table 77* through *Table 79* provide the corresponding timings. The results shown in these tables are obtained with the following FSMC configuration:

- BurstAccessMode = FSMC_BurstAccessMode_Enable;
- MemoryType = FSMC_MemoryType_CRAM;
- WriteBurst = FSMC_WriteBurst_Enable;
- CLKDivision = 1; (0 is not supported, see the STM32F20xxx/21xxx reference manual)
- DataLatency = 1 for NOR Flash; DataLatency = 0 for PSRAM

In all timing tables, the T_{HCLK} is the HCLK clock period.

Table 78. Synchronous non-multiplexed NOR/PSRAM read timings(1)(2) (continued)

Symbol	Parameter	Min	Max	Unit
$t_{d(CLKL-NADVH)}$	FSMC_CLK low to FSMC_NADV high	4	-	ns
$t_{d(CLKL-AV)}$	FSMC_CLK low to FSMC_Ax valid (x=16…25)	-	0	ns
$t_{d(CLKL-AIV)}$	FSMC_CLK low to FSMC_Ax invalid (x=16…25)	3	-	ns
$t_{d(CLKH-NOEL)}$	FSMC_CLK high to FSMC_NOE low	-	1	ns
$t_{d(CLKL-NOEH)}$	FSMC_CLK low to FSMC_NOE high	1.5	-	ns
$t_{su(DV-CLKH)}$	FSMC_D[15:0] valid data before FSMC_CLK high	8	-	ns
$t_{h(CLKH-DV)}$	FSMC_D[15:0] valid data after FSMC_CLK high	0	-	ns

1. C_L = 30 pF.
2. Guaranteed by characterization results, not tested in production.

Figure 64. Synchronous non-multiplexed PSRAM write timings

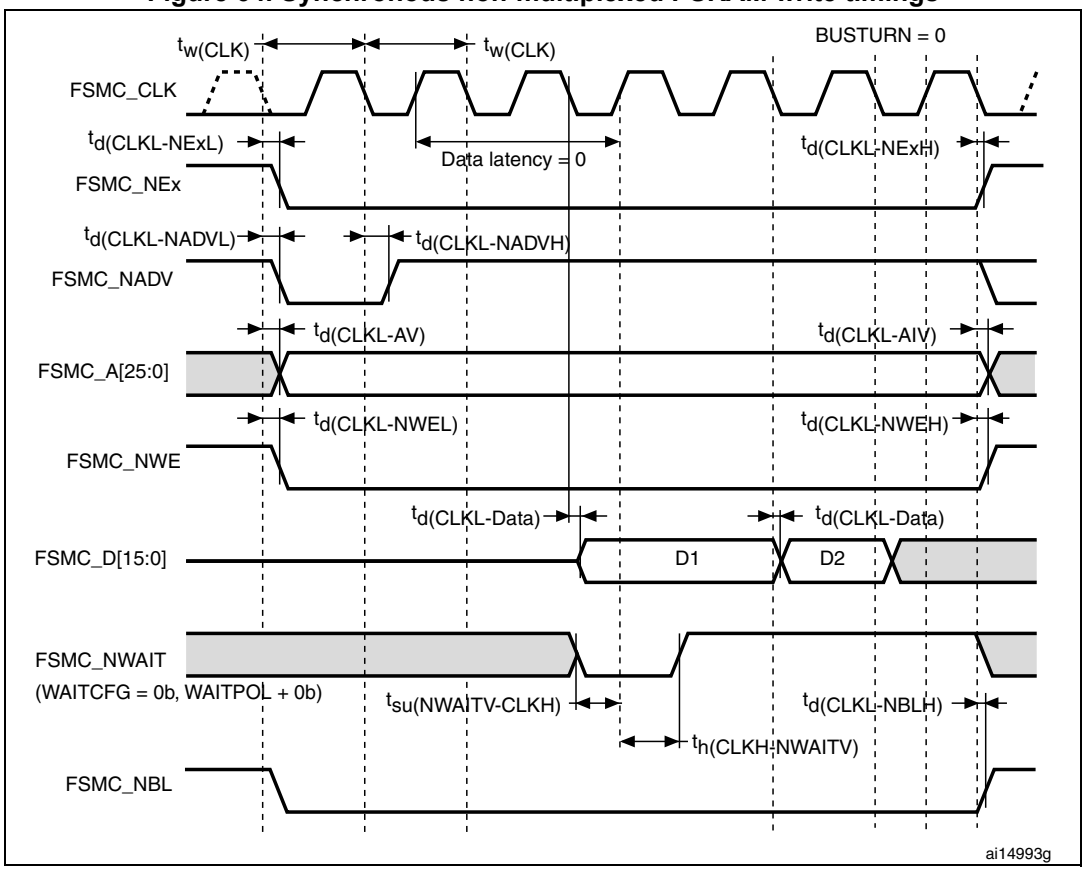
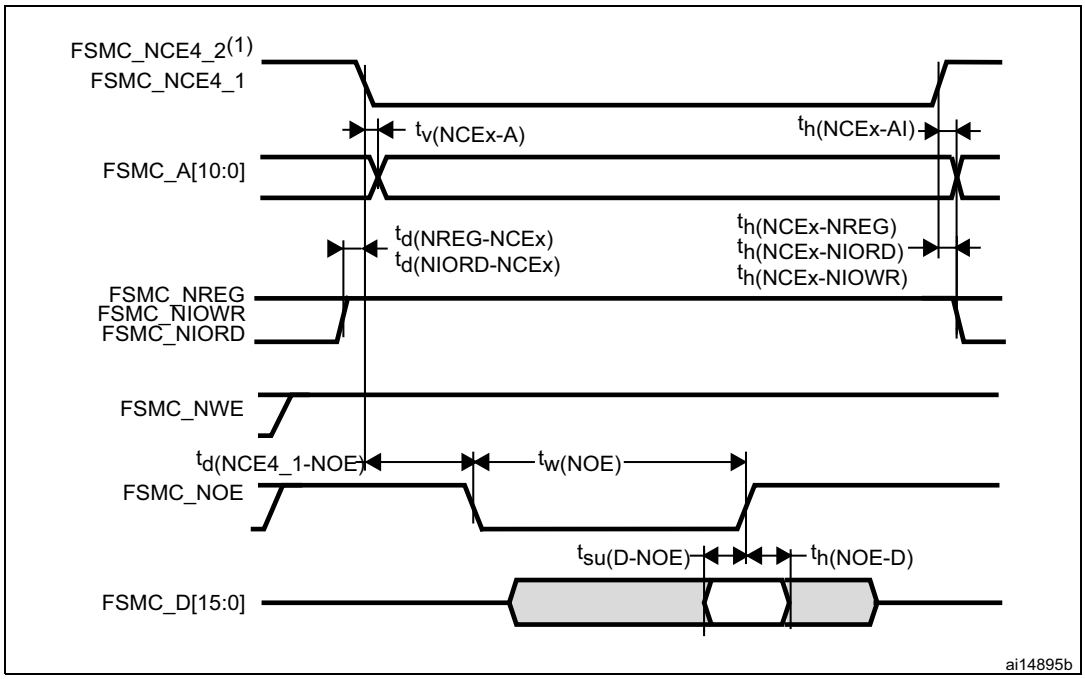


Table 79. Synchronous non-multiplexed PSRAM write timings(1)(2)

Symbol	Parameter	Min	Max	Unit
$t_{w(CLK)}$	FSMC_CLK period	$2T_{HCLK}$- 1	-	ns
$t_{d(CLKL-NExL)}$	FSMC_CLK low to FSMC_NEx low (x=0..2)	-	1	ns
$t_{d(CLKL-NExH)}$	FSMC_CLK low to FSMC_NEx high (x= 0…2)	1	-	ns

Figure 65. PC Card/CompactFlash controller waveforms for common memory read access



1. FSMC_NCE4_2 remains high (inactive during 8-bit access.

Figure 66. PC Card/CompactFlash controller waveforms for common memory write access

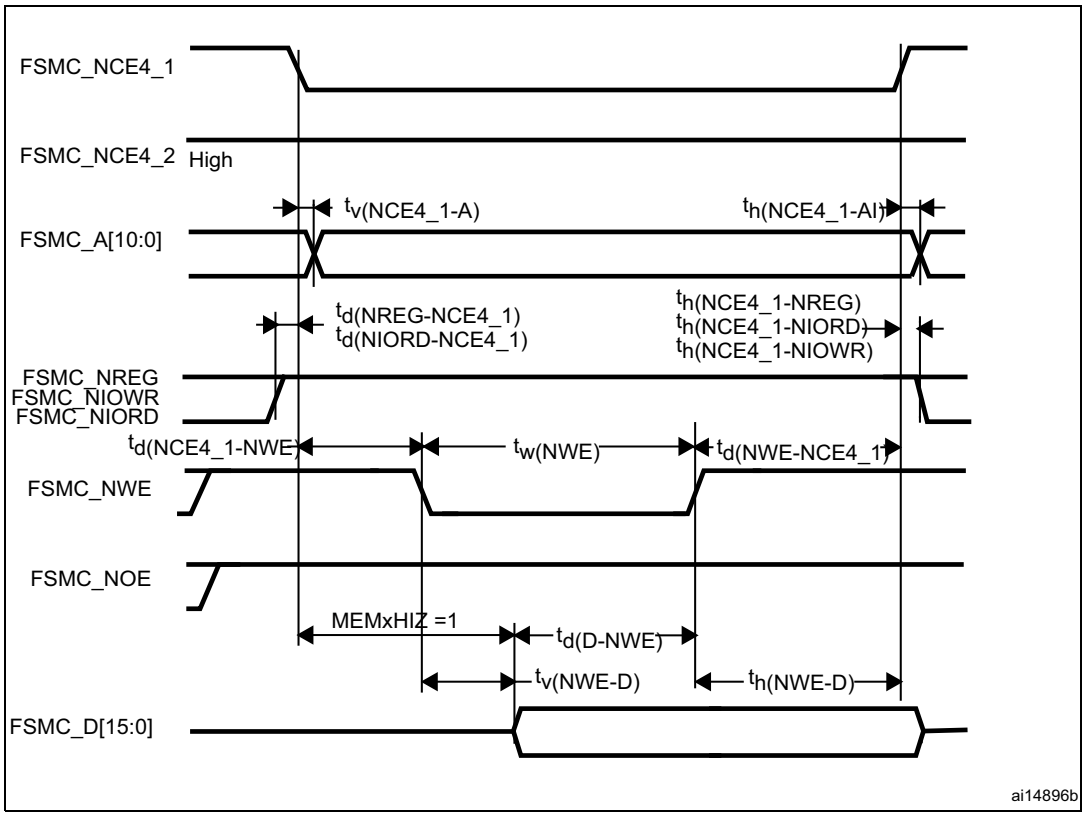
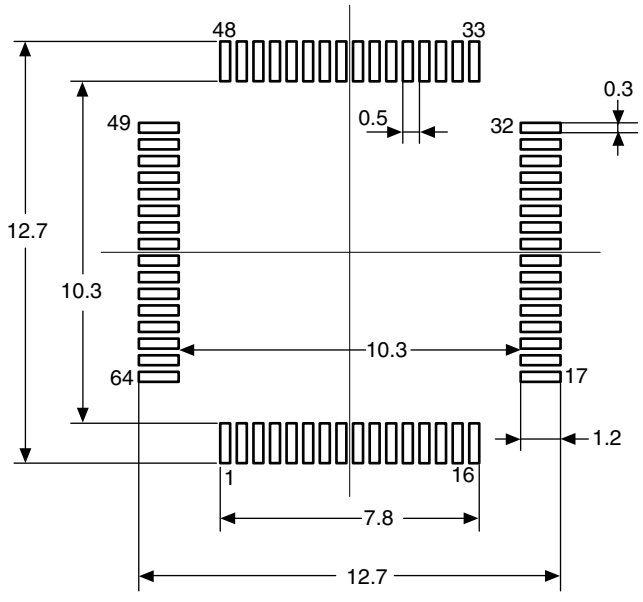


Table 87. LQFP64 - 64-pin, 10 x 10 mm low-profile quad flat package mechanical data (continued)

Symbol	millimeters			inches[1]		
	Min	Typ	Max	Min	Typ	Max
c	0.090	-	0.200	0.0035	-	0.0079
D	-	12.000	-	-	0.4724	-
D1	-	10.000	-	-	0.3937	-
D3	-	7.500	-	-	0.2953	-
E	-	12.000	-	-	0.4724	-
E1	-	10.000	-	-	0.3937	-
E3	-	7.500	-	-	0.2953	-
e	-	0.500	-	-	0.0197	-
K	0°	3.5°	7°	0°	3.5°	7°
L	0.450	0.600	0.750	0.0177	0.0236	0.0295
L1	-	1.000	-	-	0.0394	-
ccc	-	-	0.080	-	-	0.0031

1. Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 78. LQFP64 - 64-pin, 10 x 10 mm low-profile quad flat package recommended footprint



1. Dimensions are expressed in millimeters.

Table 88. WLCSP64+2 - 66-ball, 4.539 x 4.911 mm, 0.4 mm pitch wafer level chip scale package mechanical data (continued)

Symbol	millimeters			inches[1]		
	Min	Typ	Max	Min	Typ	Max
e2	-	3.200	-	-	0.1260	-
F	-	0.220	-	-	0.0087	-
G	-	0.386	-	-	0.0152	-
aaa	-	-	0.100	-	-	0.0039
bbb	-	-	0.100	-	-	0.0039
ccc	-	-	0.100	-	-	0.0039
ddd	-	-	0.050	-	-	0.0020
eee	-	-	0.050	-	-	0.0020

1. Values in inches are converted from mm and rounded to 4 decimal digits.
2. Dimension is measured at the maximum bump diameter parallel to primary datum Z.

Figure 80. WLCSP64+2 - 66-ball, 4.539 x 4.911 mm, 0.4 mm pitch wafer level chip scale package recommended footprint

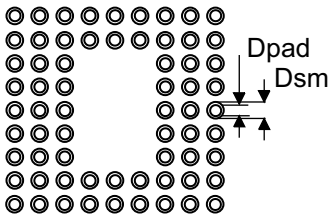


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Table 89. WLCSP64 recommended PCB design rules (0.4 mm pitch)

Dimension	Recommended values
Pitch	0.4
Dpad	0.225 mm
Dsm	0.290 mm typ. (depends on the soldermask registration tolerance)
Stencil opening	0.250 mm
Stencil thickness	0.100 mm

Table 97. Document revision history (continued)

Date	Revision	Changes
14-Jun-2011	7	Added SDIO in *Table 2: STM32F205xx features and peripheral counts*. Updated V_{IN} for 5V tolerant pins in *Table 11: Voltage characteristics*. Updated jitter parameters description in *Table 34: Main PLL characteristics*. Remove jitter values for system clock in *Table 35: PLLI2S (audio PLL) characteristics*. Updated *Table 42: EMI characteristics*. Update *Note 2* in *Table 52: I2C characteristics*. Updated Avg_Slope typical value and T_{S_temp} minimum value in *Table 69: Temperature sensor characteristics*. Updated T_{S_vbat} minimum value in *Table 70: VBAT monitoring characteristics*. Updated $T_{S\_vrefint}$ minimum value in *Table 71: Embedded internal reference voltage*. Added Software option in *Section 8: Ordering information*. In *Table 101: Main applications versus package for STM32F2xxx microcontrollers*, renamed USB1 and USB2, USB OTG FS and USB OTG HS, respectively; and removed USB OTG FS and camera interface for 64-pin package; added USB OTG HS on 64-pin package; added *Note 1* and *Note 2*.
20-Dec-2011	8	Updated SDIO register addresses in *Figure 16: Memory map*. Updated *Figure 3: Compatible board design between STM32F10x and STM32F2xx for LQFP144 package*, *Figure 2: Compatible board design between STM32F10x and STM32F2xx for LQFP100 package*, *Figure 1: Compatible board design between STM32F10x and STM32F2xx for LQFP64 package*, and added *Figure 4: Compatible board design between STM32F10xx and STM32F2xx for LQFP176 package*. Updated *Section 3.3: Memory protection unit*. Updated *Section 3.6: Embedded SRAM*. Updated *Section 3.28: Universal serial bus on-the-go full-speed (OTG_FS)* to remove external FS OTG PHY support. In *Table 8: STM32F20x pin and ball definitions*: changed SPI2_MCK and SPI3_MCK to I2S2_MCK and I2S3_MCK, respectively. Added ETH _RMII_TX_EN alternate function to PG11. Added EVENTOUT in the list of alternate functions for I/O pin/balls. Removed OTG_FS_SDA, OTG_FS_SCL and OTG_FS_INTN alternate functions. In *Table 10: Alternate function mapping*: changed I2S3_SCK to I2S3_MCK for PC7/AF6, added FSMC_NCE3 for PG9, FSMC_NE3 for PG10, and FSMC_NCE2 for PD7. Removed OTG_FS_SDA, OTG_FS_SCL and OTG_FS_INTN alternate functions. Changed I2S3_SCK into I2S3_MCK for PC7/AF6. Updated peripherals corresponding to AF12. Removed CEXT and ESR from *Table 14: General operating conditions*.