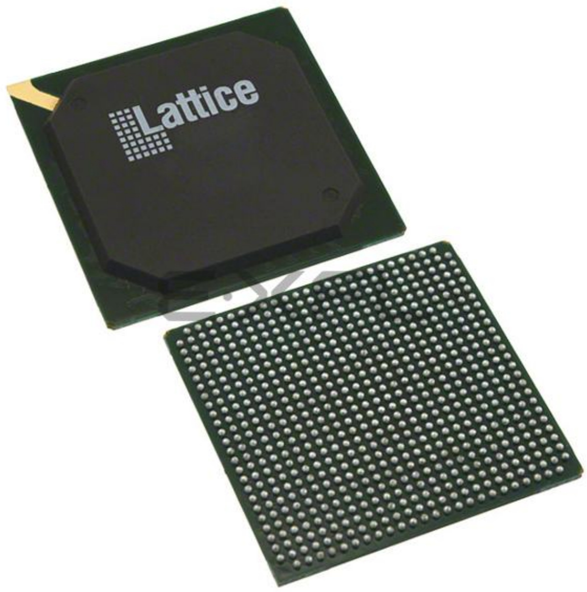
Lattice Semiconductor Corporation - LFE3-150EA-9FN672I Datasheet



Understanding Embedded - FPGAs (Field Programmable Gate Array)

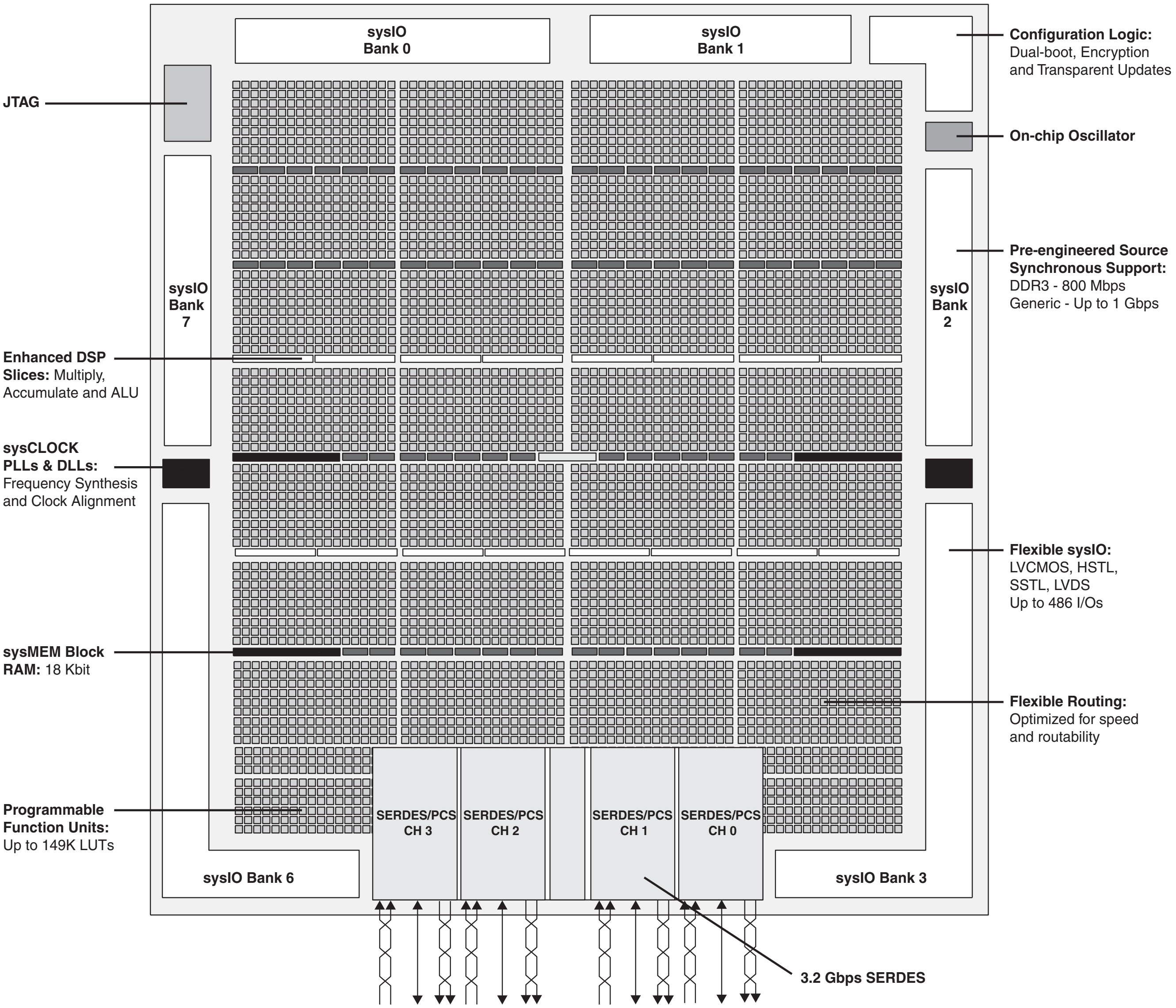
Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications,

Details	
Product Status	Not For New Designs
Number of LABs/CLBs	18625
Number of Logic Elements/Cells	149000
Total RAM Bits	7014400
Number of I/O	380
Number of Gates	-
Voltage - Supply	1.14V ~ 1.26V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 100°C (TJ)
Package / Case	672-BBGA
Supplier Device Package	672-FPBGA (27x27)
Purchase URL	https://www.e-xfl.com/product-detail/lattice-semiconductor/lfe3-150ea-9fn672i

Figure 2-1. Simplified Block Diagram, LatticeECP3-35 Device (Top Level)



Note: There is no Bank 4 or Bank 5 in LatticeECP3 devices.

PFU Blocks

The core of the LatticeECP3 device consists of PFU blocks, which are provided in two forms, the PFU and PFF. The PFUs can be programmed to perform Logic, Arithmetic, Distributed RAM and Distributed ROM functions. PFF blocks can be programmed to perform Logic, Arithmetic and ROM functions. Except where necessary, the remainder of this data sheet will use the term PFU to refer to both PFU and PFF blocks.

Each PFU block consists of four interconnected slices numbered 0-3 as shown in Figure 2-2. Each slice contains two LUTs. All the interconnections to and from PFU blocks are from routing. There are 50 inputs and 23 outputs associated with each PFU block.

ROM Mode

ROM mode uses the LUT logic; hence, Slices 0 through 3 can be used in ROM mode. Preloading is accomplished through the programming interface during PFU configuration.

For more information, please refer to TN1179, LatticeECP3 Memory Usage Guide.

Routing

There are many resources provided in the LatticeECP3 devices to route signals individually or as busses with related control signals. The routing resources consist of switching circuitry, buffers and metal interconnect (routing) segments.

The LatticeECP3 family has an enhanced routing architecture that produces a compact design. The Diamond and ispLEVER design software tool suites take the output of the synthesis tool and places and routes the design.

sysCLOCK PLLs and DLLs

The sysCLOCK PLLs provide the ability to synthesize clock frequencies. The devices in the LatticeECP3 family support two to ten full-featured General Purpose PLLs.

General Purpose PLL

The architecture of the PLL is shown in Figure 2-4. A description of the PLL functionality follows.

CLKI is the reference frequency (generated either from the pin or from routing) for the PLL. CLKI feeds into the Input Clock Divider block. The CLKFB is the feedback signal (generated from CLKOP, CLKOS or from a user clock pin/logic). This signal feeds into the Feedback Divider. The Feedback Divider is used to multiply the reference frequency.

Both the input path and feedback signals enter the Phase Frequency Detect Block (PFD) which detects first for the frequency, and then the phase, of the CLKI and CLKFB are the same which then drives the Voltage Controlled Oscillator (VCO) block. In this block the difference between the input path and feedback signals is used to control the frequency and phase of the oscillator. A LOCK signal is generated by the VCO to indicate that the VCO has locked onto the input clock signal. In dynamic mode, the PLL may lose lock after a dynamic delay adjustment and not relock until the t_{LOCK} parameter has been satisfied.

The output of the VCO then enters the CLKOP divider. The CLKOP divider allows the VCO to operate at higher frequencies than the clock output (CLKOP), thereby increasing the frequency range. The Phase/Duty Cycle/Duty Trim block adjusts the phase and duty cycle of the CLKOS signal. The phase/duty cycle setting can be pre-programmed or dynamically adjusted. A secondary divider takes the CLKOP or CLKOS signal and uses it to derive lower frequency outputs (CLKOK).

The primary output from the CLKOP divider (CLKOP) along with the outputs from the secondary dividers (CLKOK and CLKOK2) and Phase/Duty select (CLKOS) are fed to the clock distribution network.

The PLL allows two methods for adjusting the phase of signal. The first is referred to as Fine Delay Adjustment. This inserts up to 16 nominal 125 ps delays to be applied to the secondary PLL output. The number of steps may be set statically or from the FPGA logic. The second method is referred to as Coarse Phase Adjustment. This allows the phase of the rising and falling edge of the secondary PLL output to be adjusted in 22.5 degree steps. The number of steps may be set statically or from the FPGA logic.

PLL/DLL Cascading

LatticeECP3 devices have been designed to allow certain combinations of PLL and DLL cascading. The allowable combinations are:

- PLL to PLL supported
- PLL to DLL supported

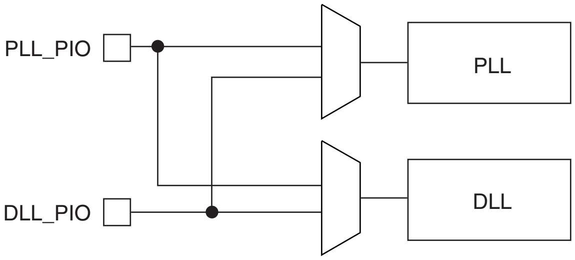
The DLLs in the LatticeECP3 are used to shift the clock in relation to the data for source synchronous inputs. PLLs are used for frequency synthesis and clock generation for source synchronous interfaces. Cascading PLL and DLL blocks allows applications to utilize the unique benefits of both DLLs and PLLs.

For further information about the DLL, please see the list of technical documentation at the end of this data sheet.

PLL/DLL PIO Input Pin Connections

All LatticeECP3 devices contains two DLLs and up to ten PLLs, arranged in quadrants. If a PLL and a DLL are next to each other, they share input pins as shown in the Figure 2-7.

Figure 2-7. Sharing of PIO Pins by PLLs and DLLs in LatticeECP3 Devices



Note: Not every PLL has an associated DLL.

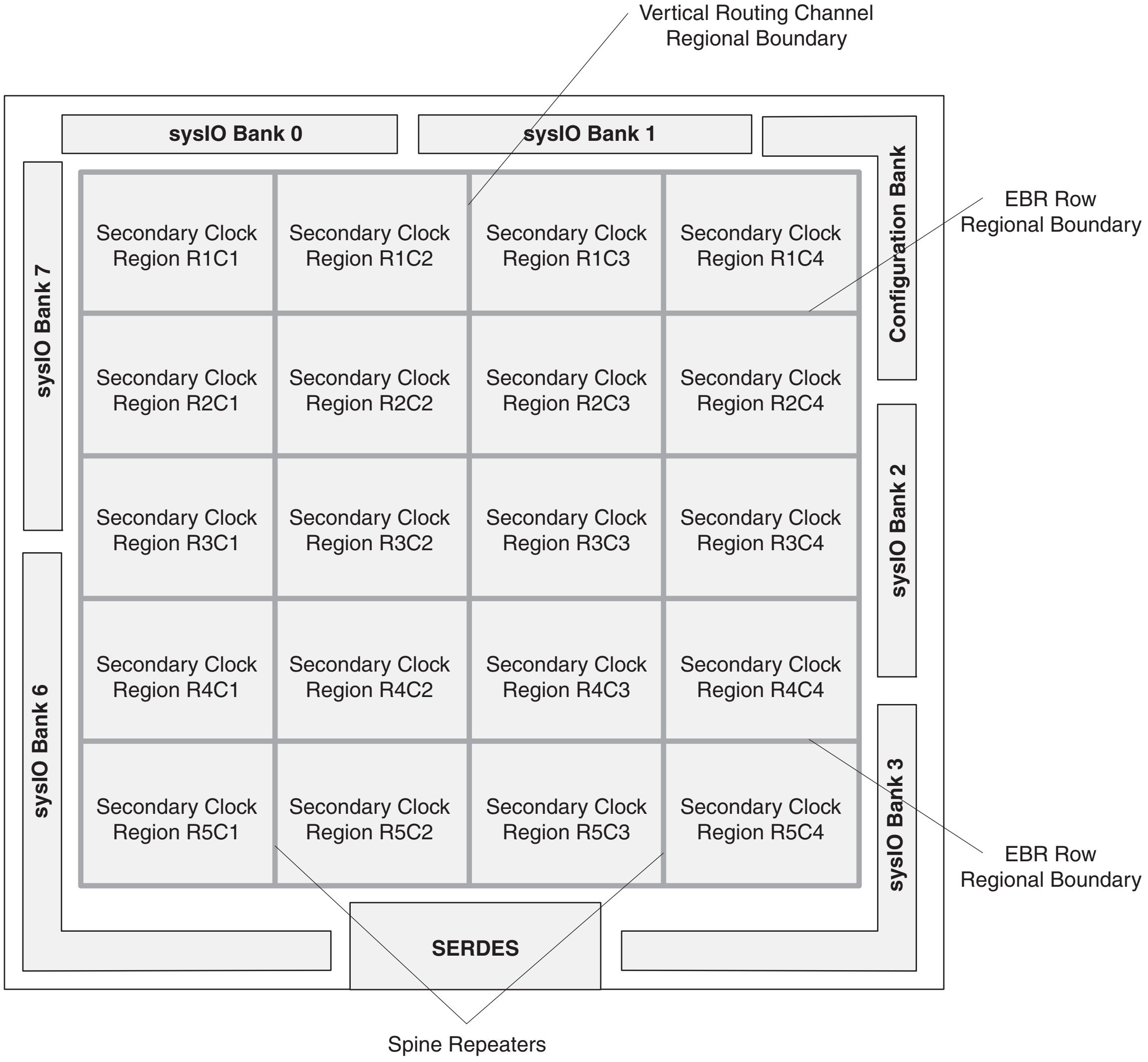
Clock Dividers

LatticeECP3 devices have two clock dividers, one on the left side and one on the right side of the device. These are intended to generate a slower-speed system clock from a high-speed edge clock. The block operates in a ÷2, ÷4 or ÷8 mode and maintains a known phase relationship between the divided down clock and the high-speed clock based on the release of its reset signal. The clock dividers can be fed from selected PLL/DLL outputs, the Slave Delay lines, routing or from an external clock input. The clock divider outputs serve as primary clock sources and feed into the clock distribution network. The Reset (RST) control signal resets input and asynchronously forces all outputs to low. The RELEASE signal releases outputs synchronously to the input clock. For further information on clock dividers, please see TN1178, LatticeECP3 sysCLOCK PLL/DLL Design and Usage Guide. Figure 2-8 shows the clock divider connections.

Table 2-6. Secondary Clock Regions

Device	Number of Secondary Clock Regions
ECP3-17	16
ECP3-35	16
ECP3-70	20
ECP3-95	20
ECP3-150	36

Figure 2-15. LatticeECP3-70 and LatticeECP3-95 Secondary Clock Regions

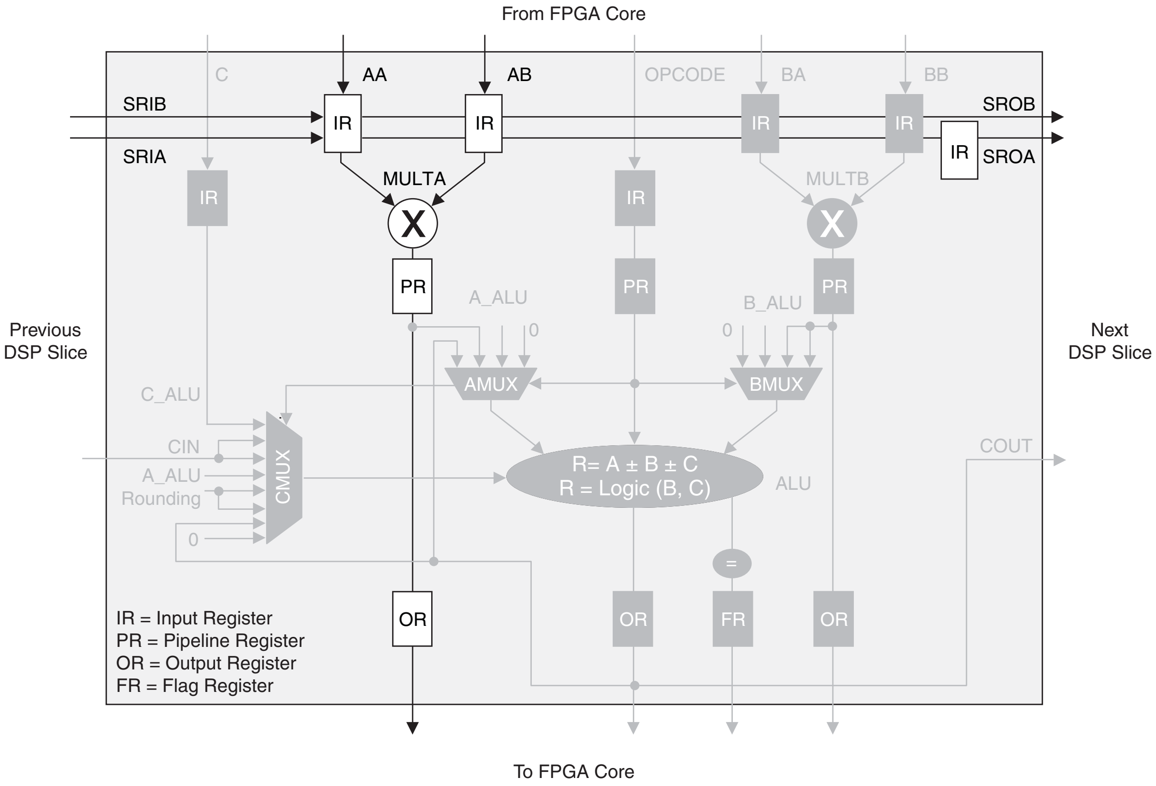


For further information, please refer to TN1182, LatticeECP3 sysDSP Usage Guide.

MULT DSP Element

This multiplier element implements a multiply with no addition or accumulator nodes. The two operands, AA and AB, are multiplied and the result is available at the output. The user can enable the input/output and pipeline registers. Figure 2-26 shows the MULT sysDSP element.

Figure 2-26. MULT sysDSP Element



Two adjacent PIOs can be joined to provide a differential I/O pair (labeled as “T” and “C”) as shown in Figure 2-32. The PAD Labels “T” and “C” distinguish the two PIOs. Approximately 50% of the PIO pairs on the left and right edges of the device can be configured as true LVDS outputs. All I/O pairs can operate as LVDS inputs.

Table 2-11. PIO Signal List

Name	Type	Description
INDD	Input Data	Register bypassed input. This is not the same port as INCK.
IPA, INA, IPB, INB	Input Data	Ports to core for input data
OPOSA, ONEGA[1], OPOSB, ONEGB[1]	Output Data	Output signals from core. An exception is the ONEGB port, used for tristate logic at the DQS pad.
CE	PIO Control	Clock enables for input and output block flip-flops.
SCLK	PIO Control	System Clock (PCLK) for input and output/TS blocks. Connected from clock ISB.
LSR	PIO Control	Local Set/Reset
ECLK1, ECLK2	PIO Control	Edge clock sources. Entire PIO selects one of two sources using mux.
ECLKDQSR[1]	Read Control	From DQS_STROBE, shifted strobe for memory interfaces only.
DDRCLKPOL[1]	Read Control	Ensures transfer from DQS domain to SCLK domain.
DDRLAT[1]	Read Control	Used to guarantee INDDRX2 gearing by selectively enabling a D-Flip-Flop in datapath.
DEL[3:0]	Read Control	Dynamic input delay control bits.
INCK	To Clock Distribution and PLL	PIO treated as clock PIO, path to distribute to primary clocks and PLL.
TS	Tristate Data	Tristate signal from core (SDR)
DQCLK0[1], DQCLK1[1]	Write Control	Two clocks edges, 90 degrees out of phase, used in output gearing.
DQSW[2]	Write Control	Used for output and tristate logic at DQS only.
DYNDEL[7:0]	Write Control	Shifting of write clocks for specific DQS group, using 6:0 each step is approximately 25ps, 128 steps. Bit 7 is an invert (timing depends on input frequency). There is also a static control for this 8-bit setting, enabled with a memory cell.
DCNTL[6:0]	PIO Control	Original delay code from DDR DLL
DATAVALID[1]	Output Data	Status flag from DATAVALID logic, used to indicate when input data is captured in IOLOGIC and valid to core.
READ	For DQS_Strobe	Read signal for DDR memory interface
DQSI	For DQS_Strobe	Unshifted DQS strobe from input pad
PRMBDET	For DQS_Strobe	DQSI biased to go high when DQSI is tristate, goes to input logic block as well as core logic.
GSRN	Control from routing	Global Set/Reset

1. Signals available on left/right/top edges only.
2. Selected PIO.

PIO

The PIO contains four blocks: an input register block, output register block, tristate register block and a control logic block. These blocks contain registers for operating in a variety of modes along with the necessary clock and selection logic.

Input Register Block

The input register blocks for the PIOs, in the left, right and top edges, contain delay elements and registers that can be used to condition high-speed interface signals, such as DDR memory interfaces and source synchronous interfaces, before they are passed to the device core. Figure 2-33 shows the input register block for the left, right and top edges. The input register block for the bottom edge contains one element to register the input signal and no DDR registers. The following description applies to the input register block for PIOs in the left, right and top edges only.

DLL Calibrated DQS Delay Block

Source synchronous interfaces generally require the input clock to be adjusted in order to correctly capture data at the input register. For most interfaces, a PLL is used for this adjustment. However, in DDR memories the clock (referred to as DQS) is not free-running so this approach cannot be used. The DQS Delay block provides the required clock alignment for DDR memory interfaces.

The delay required for the DQS signal is generated by two dedicated DLLs (DDR DLL) on opposite side of the device. Each DLL creates DQS delays in its half of the device as shown in Figure 2-36. The DDR DLL on the left side will generate delays for all the DQS Strobe pins on Banks 0, 7 and 6 and DDR DLL on the right will generate delays for all the DQS pins on Banks 1, 2 and 3. The DDR DLL loop compensates for temperature, voltage and process variations by using the system clock and DLL feedback loop. DDR DLL communicates the required delay to the DQS delay block using a 7-bit calibration bus (DCNTL[6:0])

The DQS signal (selected PIOs only, as shown in Figure 2-35) feeds from the PAD through a DQS control logic block to a dedicated DQS routing resource. The DQS control logic block consists of DQS Read Control logic block that generates control signals for the read side and DQS Write Control logic that generates the control signals required for the write side. A more detailed DQS control diagram is shown in Figure 2-37, which shows how the DQS control blocks interact with the data paths.

The DQS Read control logic receives the delay generated by the DDR DLL on its side and delays the incoming DQS signal by 90 degrees. This delayed ECLKDQSR is routed to 10 or 11 DQ pads covered by that DQS signal. This block also contains a polarity control logic that generates a DDRCLKPOL signal, which controls the polarity of the clock to the sync registers in the input register blocks. The DQS Read control logic also generates a DDRLAT signal that is in the input register block to transfer data from the first set of DDR register to the second set of DDR registers when using the DDRX2 gearbox mode for DDR3 memory interface.

The DQS Write control logic block generates the DQCLK0 and DQCLK1 clocks used to control the output gearing in the Output register block which generates the DDR data output and the DQS output. They are also used to control the generation of the DQS output through the DQS output register block. In addition to the DCNTL [6:0] input from the DDR DLL, the DQS Write control block also uses a Dynamic Delay DYN DEL [7:0] attribute which is used to further delay the DQS to accomplish the write leveling found in DDR3 memory. Write leveling is controlled by the DDR memory controller implementation. The DYN DELAY can set 128 possible delay step settings. In addition, the most significant bit will invert the clock for a 180-degree shift of the incoming clock. This will generate the DQSW signal used to generate the DQS output in the DQS output register block.

Figure 2-36 and Figure 2-37 show how the DQS transition signals that are routed to the PIOs.

Please see TN1180, LatticeECP3 High-Speed I/O Interface for more information on this topic.

SCI (SERDES Client Interface) Bus

The SERDES Client Interface (SCI) is an IP interface that allows the SERDES/PCS Quad block to be controlled by registers rather than the configuration memory cells. It is a simple register configuration interface that allows SERDES/PCS configuration without power cycling the device.

The Diamond and ispLEVER design tools support all modes of the PCS. Most modes are dedicated to applications associated with a specific industry standard data protocol. Other more general purpose modes allow users to define their own operation. With these tools, the user can define the mode for each quad in a design.

Popular standards such as 10Gb Ethernet, x4 PCI Express and 4x Serial RapidIO can be implemented using IP (available through Lattice), a single quad (Four SERDES channels and PCS) and some additional logic from the core.

The LatticeECP3 family also supports a wide range of primary and secondary protocols. Within the same quad, the LatticeECP3 family can support mixed protocols with semi-independent clocking as long as the required clock frequencies are integer x1, x2, or x11 multiples of each other. Table 2-15 lists the allowable combination of primary and secondary protocol combinations.

Flexible Quad SERDES Architecture

The LatticeECP3 family SERDES architecture is a quad-based architecture. For most SERDES settings and standards, the whole quad (consisting of four SERDES) is treated as a unit. This helps in silicon area savings, better utilization and overall lower cost.

However, for some specific standards, the LatticeECP3 quad architecture provides flexibility; more than one standard can be supported within the same quad.

Table 2-15 shows the standards can be mixed and matched within the same quad. In general, the SERDES standards whose nominal data rates are either the same or a defined subset of each other, can be supported within the same quad. In Table 2-15, the Primary Protocol column refers to the standard that determines the reference clock and PLL settings. The Secondary Protocol column shows the other standard that can be supported within the same quad.

Furthermore, Table 2-15 also implies that more than two standards in the same quad can be supported, as long as they conform to the data rate and reference clock requirements. For example, a quad may contain PCI Express 1.1, SGMII, Serial RapidIO Type I and Serial RapidIO Type II, all in the same quad.

Table 2-15. LatticeECP3 Primary and Secondary Protocol Support

Primary Protocol	Secondary Protocol
PCI Express 1.1	SGMII
PCI Express 1.1	Gigabit Ethernet
PCI Express 1.1	Serial RapidIO Type I
PCI Express 1.1	Serial RapidIO Type II
Serial RapidIO Type I	SGMII
Serial RapidIO Type I	Gigabit Ethernet
Serial RapidIO Type II	SGMII
Serial RapidIO Type II	Gigabit Ethernet
Serial RapidIO Type II	Serial RapidIO Type I
CPRI-3	CPRI-2 and CPRI-1
3G-SDI	HD-SDI and SD-SDI

There are some restrictions to be aware of when using spread spectrum. When a quad shares a PCI Express x1 channel with a non-PCI Express channel, ensure that the reference clock for the quad is compatible with all protocols within the quad. For example, a PCI Express spread spectrum reference clock is not compatible with most Gigabit Ethernet applications because of tight CTC ppm requirements.

While the LatticeECP3 architecture will allow the mixing of a PCI Express channel and a Gigabit Ethernet, Serial RapidIO or SGMII channel within the same quad, using a PCI Express spread spectrum clocking as the transmit reference clock will cause a violation of the Gigabit Ethernet, Serial RapidIO and SGMII transmit jitter specifications.

For further information on SERDES, please see TN1176, LatticeECP3 SERDES/PCS Usage Guide.

IEEE 1149.1-Compliant Boundary Scan Testability

All LatticeECP3 devices have boundary scan cells that are accessed through an IEEE 1149.1 compliant Test Access Port (TAP). This allows functional testing of the circuit board on which the device is mounted through a serial scan path that can access all critical logic nodes. Internal registers are linked internally, allowing test data to be shifted in and loaded directly onto test nodes, or test data to be captured and shifted out for verification. The test access port consists of dedicated I/Os: TDI, TDO, TCK and TMS. The test access port has its own supply voltage V_{CCJ} and can operate with LVCMOS3.3, 2.5, 1.8, 1.5 and 1.2 standards.

For more information, please see TN1169, LatticeECP3 sysCONFIG Usage Guide.

Device Configuration

All LatticeECP3 devices contain two ports that can be used for device configuration. The Test Access Port (TAP), which supports bit-wide configuration, and the sysCONFIG port, support dual-byte, byte and serial configuration. The TAP supports both the IEEE Standard 1149.1 Boundary Scan specification and the IEEE Standard 1532 In-System Configuration specification. The sysCONFIG port includes seven I/Os used as dedicated pins with the remaining pins used as dual-use pins. See TN1169, LatticeECP3 sysCONFIG Usage Guide for more information about using the dual-use pins as general purpose I/Os.

There are various ways to configure a LatticeECP3 device:

1. JTAG
2. Standard Serial Peripheral Interface (SPI and SPIm modes) - interface to boot PROM memory
3. System microprocessor to drive a x8 CPU port (PCM mode)
4. System microprocessor to drive a serial slave SPI port (SSPI mode)
5. Generic byte wide flash with a MachXO™ device, providing control and addressing

On power-up, the FPGA SRAM is ready to be configured using the selected sysCONFIG port. Once a configuration port is selected, it will remain active throughout that configuration cycle. The IEEE 1149.1 port can be activated any time after power-up by sending the appropriate command through the TAP port.

LatticeECP3 devices also support the Slave SPI Interface. In this mode, the FPGA behaves like a SPI Flash device (slave mode) with the SPI port of the FPGA to perform read-write operations.

Table 2-16. Selectable Master Clock (MCCLK) Frequencies During Configuration (Nominal)

MCCLK (MHz)	MCCLK (MHz)
	10
2.5[1]	13
4.3	15[2]
5.4	20
6.9	26
8.1	33[3]
9.2	

1. Software default MCCLK frequency. Hardware default is 3.1 MHz.
2. Maximum MCCLK with encryption enabled.
3. Maximum MCCLK without encryption.

Density Shifting

The LatticeECP3 family is designed to ensure that different density devices in the same family and in the same package have the same pinout. Furthermore, the architecture ensures a high success rate when performing design migration from lower density devices to higher density devices. In many cases, it is also possible to shift a lower utilization design targeted for a high-density device to a lower density device. However, the exact details of the final resource utilization will impact the likelihood of success in each case. An example is that some user I/Os may become No Connects in smaller devices in the same package. Refer to the LatticeECP3 Pin Migration Tables and Diamond software for specific restrictions and limitations.

LatticeECP3 Family Timing Adders[1, 2, 3, 4, 5, 7] (Continued)

Over Recommended Commercial Operating Conditions

Buffer Type	Description	–8	–7	–6	Units
RSDS25	RSDS, VCCIO = 2.5 V	–0.07	–0.04	–0.01	ns
PPLVDS	Point-to-Point LVDS, True LVDS, VCCIO = 2.5 V or 3.3 V	–0.22	–0.19	–0.16	ns
LVPECL33	LVPECL, Emulated, VCCIO = 3.3 V	0.67	0.76	0.86	ns
HSTL18_I	HSTL_18 class I 8mA drive, VCCIO = 1.8 V	1.20	1.34	1.47	ns
HSTL18_II	HSTL_18 class II, VCCIO = 1.8 V	0.89	1.00	1.11	ns
HSTL18D_I	Differential HSTL 18 class I 8 mA drive	1.20	1.34	1.47	ns
HSTL18D_II	Differential HSTL 18 class II	0.89	1.00	1.11	ns
HSTL15_I	HSTL_15 class I 4 mA drive, VCCIO = 1.5 V	1.67	1.83	1.99	ns
HSTL15D_I	Differential HSTL 15 class I 4 mA drive	1.67	1.83	1.99	ns
SSTL33_I	SSTL_3 class I, VCCIO = 3.3 V	1.12	1.17	1.21	ns
SSTL33_II	SSTL_3 class II, VCCIO = 3.3 V	1.08	1.12	1.15	ns
SSTL33D_I	Differential SSTL_3 class I	1.12	1.17	1.21	ns
SSTL33D_II	Differential SSTL_3 class II	1.08	1.12	1.15	ns
SSTL25_I	SSTL_2 class I 8 mA drive, VCCIO = 2.5 V	1.06	1.19	1.31	ns
SSTL25_II	SSTL_2 class II 16 mA drive, VCCIO = 2.5 V	1.04	1.17	1.31	ns
SSTL25D_I	Differential SSTL_2 class I 8 mA drive	1.06	1.19	1.31	ns
SSTL25D_II	Differential SSTL_2 class II 16 mA drive	1.04	1.17	1.31	ns
SSTL18_I	SSTL_1.8 class I, VCCIO = 1.8 V	0.70	0.84	0.97	ns
SSTL18_II	SSTL_1.8 class II 8 mA drive, VCCIO = 1.8 V	0.70	0.84	0.97	ns
SSTL18D_I	Differential SSTL_1.8 class I	0.70	0.84	0.97	ns
SSTL18D_II	Differential SSTL_1.8 class II 8 mA drive	0.70	0.84	0.97	ns
SSTL15	SSTL_1.5, VCCIO = 1.5 V	1.22	1.35	1.48	ns
SSTL15D	Differential SSTL_15	1.22	1.35	1.48	ns
LVTTL33_4mA	LVTTL 4 mA drive, VCCIO = 3.3V	0.25	0.24	0.23	ns
LVTTL33_8mA	LVTTL 8 mA drive, VCCIO = 3.3V	–0.06	–0.06	–0.07	ns
LVTTL33_12mA	LVTTL 12 mA drive, VCCIO = 3.3V	–0.01	–0.02	–0.02	ns
LVTTL33_16mA	LVTTL 16 mA drive, VCCIO = 3.3V	–0.07	–0.07	–0.08	ns
LVTTL33_20mA	LVTTL 20 mA drive, VCCIO = 3.3V	–0.12	–0.13	–0.14	ns
LVCMOS33_4mA	LVCMOS 3.3 4 mA drive, fast slew rate	0.25	0.24	0.23	ns
LVCMOS33_8mA	LVCMOS 3.3 8 mA drive, fast slew rate	–0.06	–0.06	–0.07	ns
LVCMOS33_12mA	LVCMOS 3.3 12 mA drive, fast slew rate	–0.01	–0.02	–0.02	ns
LVCMOS33_16mA	LVCMOS 3.3 16 mA drive, fast slew rate	–0.07	–0.07	–0.08	ns
LVCMOS33_20mA	LVCMOS 3.3 20 mA drive, fast slew rate	–0.12	–0.13	–0.14	ns
LVCMOS25_4mA	LVCMOS 2.5 4 mA drive, fast slew rate	0.12	0.10	0.09	ns
LVCMOS25_8mA	LVCMOS 2.5 8 mA drive, fast slew rate	–0.05	–0.06	–0.07	ns
LVCMOS25_12mA	LVCMOS 2.5 12 mA drive, fast slew rate	0.00	0.00	0.00	ns
LVCMOS25_16mA	LVCMOS 2.5 16 mA drive, fast slew rate	–0.12	–0.13	–0.14	ns
LVCMOS25_20mA	LVCMOS 2.5 20 mA drive, fast slew rate	–0.12	–0.13	–0.14	ns
LVCMOS18_4mA	LVCMOS 1.8 4 mA drive, fast slew rate	0.11	0.12	0.14	ns
LVCMOS18_8mA	LVCMOS 1.8 8 mA drive, fast slew rate	0.11	0.12	0.14	ns
LVCMOS18_12mA	LVCMOS 1.8 12 mA drive, fast slew rate	–0.04	–0.03	–0.03	ns
LVCMOS18_16mA	LVCMOS 1.8 16 mA drive, fast slew rate	–0.04	–0.03	–0.03	ns

Table 3-11. Periodic Receiver Jitter Tolerance Specification

Description	Frequency	Condition	Min.	Typ.	Max.	Units
Periodic	2.97 Gbps	600 mV differential eye	—	—	0.24	UI, p-p
Periodic	2.5 Gbps	600 mV differential eye	—	—	0.22	UI, p-p
Periodic	1.485 Gbps	600 mV differential eye	—	—	0.24	UI, p-p
Periodic	622 Mbps	600 mV differential eye	—	—	0.15	UI, p-p
Periodic	150 Mbps	600 mV differential eye	—	—	0.5	UI, p-p

Note: Values are measured with PRBS 2^7–1, all channels operating, FPGA Logic active, I/Os around SERDES pins quiet, voltages are nominal, room temperature.

XAUI/Serial Rapid I/O Type 3/CPRI LV E.30 Electrical and Timing Characteristics

AC and DC Characteristics

Table 3-13. Transmit

Over Recommended Operating Conditions

Symbol	Description	Test Conditions	Min.	Typ.	Max.	Units
T_{RF}	Differential rise/fall time	20%-80%	—	80	—	ps
$Z_{TX\_DIFF\_DC}$	Differential impedance		80	100	120	Ohms
J_{TX_DDJ}[2, 3, 4]	Output data deterministic jitter		—	—	0.17	UI
J_{TX_TJ}[1, 2, 3, 4]	Total output data jitter		—	—	0.35	UI

1. Total jitter includes both deterministic jitter and random jitter.
2. Jitter values are measured with each CML output AC coupled into a 50-Ohm impedance (100-Ohm differential impedance).
3. Jitter and skew are specified between differential crossings of the 50% threshold of the reference signal.
4. Values are measured at 2.5 Gbps.

Table 3-14. Receive and Jitter Tolerance

Over Recommended Operating Conditions

Symbol	Description	Test Conditions	Min.	Typ.	Max.	Units
RL_{RX_DIFF}	Differential return loss	From 100 MHz to 3.125 GHz	10	—	—	dB
RL_{RX_CM}	Common mode return loss	From 100 MHz to 3.125 GHz	6	—	—	dB
Z_{RX_DIFF}	Differential termination resistance		80	100	120	Ohms
J_{RX_DJ}[1, 2, 3]	Deterministic jitter tolerance (peak-to-peak)		—	—	0.37	UI
J_{RX_RJ}[1, 2, 3]	Random jitter tolerance (peak-to-peak)		—	—	0.18	UI
J_{RX_SJ}[1, 2, 3]	Sinusoidal jitter tolerance (peak-to-peak)		—	—	0.10	UI
J_{RX_TJ}[1, 2, 3]	Total jitter tolerance (peak-to-peak)		—	—	0.65	UI
T_{RX_EYE}	Receiver eye opening		0.35	—	—	UI

1. Total jitter includes deterministic jitter, random jitter and sinusoidal jitter. The sinusoidal jitter tolerance mask is shown in Figure 3-18.
2. Jitter values are measured with each high-speed input AC coupled into a 50-Ohm impedance.
3. Jitter and skew are specified between differential crossings of the 50% threshold of the reference signal.
4. Jitter tolerance parameters are characterized when Full Rx Equalization is enabled.
5. Values are measured at 2.5 Gbps.

Signal Descriptions (Cont.)

Signal Name	I/O	Description
[LOC]DQS[num]	I/O	DQ input/output pads: T (top), R (right), B (bottom), L (left), DQS, num = ball function number.
[LOC]DQ[num]	I/O	DQ input/output pads: T (top), R (right), B (bottom), L (left), DQ, associated DQS number.
Test and Programming (Dedicated Pins)		
TMS	I	Test Mode Select input, used to control the 1149.1 state machine. Pull-up is enabled during configuration.
TCK	I	Test Clock input pin, used to clock the 1149.1 state machine. No pull-up enabled.
TDI	I	Test Data in pin. Used to load data into device using 1149.1 state machine. After power-up, this TAP port can be activated for configuration by sending appropriate command. (Note: once a configuration port is selected it is locked. Another configuration port cannot be selected until the power-up sequence). Pull-up is enabled during configuration.
TDO	O	Output pin. Test Data Out pin used to shift data out of a device using 1149.1.
VCCJ	—	Power supply pin for JTAG Test Access Port.
Configuration Pads (Used During sysCONFIG)		
CFG[2:0]	I	Mode pins used to specify configuration mode values latched on rising edge of INITN. During configuration, a pull-up is enabled. These are dedicated pins.
INITN	I/O	Open Drain pin. Indicates the FPGA is ready to be configured. During configuration, a pull-up is enabled. It is a dedicated pin.
PROGRAMN	I	Initiates configuration sequence when asserted low. This pin always has an active pull-up. It is a dedicated pin.
DONE	I/O	Open Drain pin. Indicates that the configuration sequence is complete, and the startup sequence is in progress. It is a dedicated pin.
CCLK	I	Input Configuration Clock for configuring an FPGA in Slave SPI, Serial, and CPU modes. It is a dedicated pin.
MCLK	I/O	Output Configuration Clock for configuring an FPGA in SPI, SPIm, and Master configuration modes.
BUSY/SISPI	O	Parallel configuration mode busy indicator. SPI/SPIm mode data output.
CSN/SN/OEN	I/O	Parallel configuration mode active-low chip select. Slave SPI chip select. Parallel burst Flash output enable.
CS1N/HOLDN/RDY	I	Parallel configuration mode active-low chip select. Slave SPI hold input.
WRITEN	I	Write enable for parallel configuration modes.
DOUT/CSON/CSSPI1N	O	Serial data output. Chip select output. SPI/SPIm mode chip select.
D[0]/SPIFASTN	I/O	sysCONFIG Port Data I/O for Parallel mode. Open drain during configuration.
		sysCONFIG Port Data I/O for SPI or SPIm. When using the SPI or SPIm mode, this pin should either be tied high or low, must not be left floating. Open drain during configuration.
D1	I/O	Parallel configuration I/O. Open drain during configuration.
D2	I/O	Parallel configuration I/O. Open drain during configuration.
D3/SI	I/O	Parallel configuration I/O. Slave SPI data input. Open drain during configuration.
D4/SO	I/O	Parallel configuration I/O. Slave SPI data output. Open drain during configuration.
D5	I/O	Parallel configuration I/O. Open drain during configuration.
D6/SPID1	I/O	Parallel configuration I/O. SPI/SPIm data input. Open drain during configuration.

PICs and DDR Data (DQ) Pins Associated with the DDR Strobe (DQS) Pin

PICs Associated with DQS Strobe	PIO Within PIC	DDR Strobe (DQS) and Data (DQ) Pins
For Left and Right Edges of the Device		
P[Edge] [n-3]	A	DQ
	B	DQ
P[Edge] [n-2]	A	DQ
	B	DQ
P[Edge] [n-1]	A	DQ
	B	DQ
P[Edge] [n]	A	[Edge]DQSn
	B	DQ
P[Edge] [n+1]	A	DQ
	B	DQ
P[Edge] [n+2]	A	DQ
	B	DQ
For Top Edge of the Device		
P[Edge] [n-3]	A	DQ
	B	DQ
P[Edge] [n-2]	A	DQ
	B	DQ
P[Edge] [n-1]	A	DQ
	B	DQ
P[Edge] [n]	A	[Edge]DQSn
	B	DQ
P[Edge] [n+1]	A	DQ
	B	DQ
P[Edge] [n+2]	A	DQ
	B	DQ

Note: “n” is a row PIC number.

Pin Information Summary (Cont.)

Pin Information Summary		ECP3-70EA		
Pin Type		484 fpBGA	672 fpBGA	1156 fpBGA
Emulated Differential I/O per Bank	Bank 0	21	30	43
	Bank 1	18	24	39
	Bank 2	8	12	13
	Bank 3	20	23	33
	Bank 6	22	25	33
	Bank 7	11	16	18
	Bank 8	12	12	12
High-Speed Differential I/O per Bank	Bank 0	0	0	0
	Bank 1	0	0	0
	Bank 2	6	9	9
	Bank 3	9	12	16
	Bank 6	11	14	16
	Bank 7	9	12	13
	Bank 8	0	0	0
Total Single-Ended/ Total Differential I/O per Bank	Bank 0	42/21	60/30	86/43
	Bank 1	36/18	48/24	78/39
	Bank 2	28/14	42/21	44/22
	Bank 3	58/29	71/35	98/49
	Bank 6	67/33	78/39	98/49
	Bank 7	40/20	56/28	62/31
	Bank 8	24/12	24/12	24/12
DDR Groups Bonded per Bank[1]	Bank 0	3	5	7
	Bank 1	3	4	7
	Bank 2	2	3	3
	Bank 3	3	4	5
	Bank 6	4	4	5
	Bank 7	3	4	4
	Configuration Bank 8	0	0	0
SERDES Quads		1	2	3

1. Some DQS groups may not support DQS-12. Refer to the device pinout (.csv) file.

Part Number	Voltage	Grade[1]	Power	Package	Pins	Temp.	LUTs (K)
LFE3-70EA-6FN484C	1.2 V	–6	STD	Lead-Free fpBGA	484	COM	67
LFE3-70EA-7FN484C	1.2 V	–7	STD	Lead-Free fpBGA	484	COM	67
LFE3-70EA-8FN484C	1.2 V	–8	STD	Lead-Free fpBGA	484	COM	67
LFE3-70EA-6LFN484C	1.2 V	–6	LOW	Lead-Free fpBGA	484	COM	67
LFE3-70EA-7LFN484C	1.2 V	–7	LOW	Lead-Free fpBGA	484	COM	67
LFE3-70EA-8LFN484C	1.2 V	–8	LOW	Lead-Free fpBGA	484	COM	67
LFE3-70EA-6FN672C	1.2 V	–6	STD	Lead-Free fpBGA	672	COM	67
LFE3-70EA-7FN672C	1.2 V	–7	STD	Lead-Free fpBGA	672	COM	67
LFE3-70EA-8FN672C	1.2 V	–8	STD	Lead-Free fpBGA	672	COM	67
LFE3-70EA-6LFN672C	1.2 V	–6	LOW	Lead-Free fpBGA	672	COM	67
LFE3-70EA-7LFN672C	1.2 V	–7	LOW	Lead-Free fpBGA	672	COM	67
LFE3-70EA-8LFN672C	1.2 V	–8	LOW	Lead-Free fpBGA	672	COM	67
LFE3-70EA-6FN1156C	1.2 V	–6	STD	Lead-Free fpBGA	1156	COM	67
LFE3-70EA-7FN1156C	1.2 V	–7	STD	Lead-Free fpBGA	1156	COM	67
LFE3-70EA-8FN1156C	1.2 V	–8	STD	Lead-Free fpBGA	1156	COM	67
LFE3-70EA-6LFN1156C	1.2 V	–6	LOW	Lead-Free fpBGA	1156	COM	67
LFE3-70EA-7LFN1156C	1.2 V	–7	LOW	Lead-Free fpBGA	1156	COM	67
LFE3-70EA-8LFN1156C	1.2 V	–8	LOW	Lead-Free fpBGA	1156	COM	67

1. For ordering information on -9 speed grade devices, please contact your Lattice Sales Representative.

Part Number	Voltage	Grade[1]	Power	Package	Pins	Temp.	LUTs (K)
LFE3-95EA-6FN484C	1.2 V	–6	STD	Lead-Free fpBGA	484	COM	92
LFE3-95EA-7FN484C	1.2 V	–7	STD	Lead-Free fpBGA	484	COM	92
LFE3-95EA-8FN484C	1.2 V	–8	STD	Lead-Free fpBGA	484	COM	92
LFE3-95EA-6LFN484C	1.2 V	–6	LOW	Lead-Free fpBGA	484	COM	92
LFE3-95EA-7LFN484C	1.2 V	–7	LOW	Lead-Free fpBGA	484	COM	92
LFE3-95EA-8LFN484C	1.2 V	–8	LOW	Lead-Free fpBGA	484	COM	92
LFE3-95EA-6FN672C	1.2 V	–6	STD	Lead-Free fpBGA	672	COM	92
LFE3-95EA-7FN672C	1.2 V	–7	STD	Lead-Free fpBGA	672	COM	92
LFE3-95EA-8FN672C	1.2 V	–8	STD	Lead-Free fpBGA	672	COM	92
LFE3-95EA-6LFN672C	1.2 V	–6	LOW	Lead-Free fpBGA	672	COM	92
LFE3-95EA-7LFN672C	1.2 V	–7	LOW	Lead-Free fpBGA	672	COM	92
LFE3-95EA-8LFN672C	1.2 V	–8	LOW	Lead-Free fpBGA	672	COM	92
LFE3-95EA-6FN1156C	1.2 V	–6	STD	Lead-Free fpBGA	1156	COM	92
LFE3-95EA-7FN1156C	1.2 V	–7	STD	Lead-Free fpBGA	1156	COM	92
LFE3-95EA-8FN1156C	1.2 V	–8	STD	Lead-Free fpBGA	1156	COM	92
LFE3-95EA-6LFN1156C	1.2 V	–6	LOW	Lead-Free fpBGA	1156	COM	92
LFE3-95EA-7LFN1156C	1.2 V	–7	LOW	Lead-Free fpBGA	1156	COM	92
LFE3-95EA-8LFN1156C	1.2 V	–8	LOW	Lead-Free fpBGA	1156	COM	92

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Part Number	Voltage	Grade[1]	Power	Package	Pins	Temp.	LUTs (K)
LFE3-150EA-6FN672C	1.2 V	–6	STD	Lead-Free fpBGA	672	COM	149
LFE3-150EA-7FN672C	1.2 V	–7	STD	Lead-Free fpBGA	672	COM	149
LFE3-150EA-8FN672C	1.2 V	–8	STD	Lead-Free fpBGA	672	COM	149
LFE3-150EA-6LFN672C	1.2 V	–6	LOW	Lead-Free fpBGA	672	COM	149
LFE3-150EA-7LFN672C	1.2 V	–7	LOW	Lead-Free fpBGA	672	COM	149
LFE3-150EA-8LFN672C	1.2 V	–8	LOW	Lead-Free fpBGA	672	COM	149
LFE3-150EA-6FN1156C	1.2 V	–6	STD	Lead-Free fpBGA	1156	COM	149
LFE3-150EA-7FN1156C	1.2 V	–7	STD	Lead-Free fpBGA	1156	COM	149
LFE3-150EA-8FN1156C	1.2 V	–8	STD	Lead-Free fpBGA	1156	COM	149
LFE3-150EA-6LFN1156C	1.2 V	–6	LOW	Lead-Free fpBGA	1156	COM	149
LFE3-150EA-7LFN1156C	1.2 V	–7	LOW	Lead-Free fpBGA	1156	COM	149
LFE3-150EA-8LFN1156C	1.2 V	–8	LOW	Lead-Free fpBGA	1156	COM	149

1. For ordering information on -9 speed grade devices, please contact your Lattice Sales Representative.

Part Number	Voltage	Grade	Power	Package	Pins	Temp.	LUTs (K)
LFE3-150EA-6FN672CTW[1]	1.2 V	–6	STD	Lead-Free fpBGA	672	COM	149
LFE3-150EA-7FN672CTW[1]	1.2 V	–7	STD	Lead-Free fpBGA	672	COM	149
LFE3-150EA-8FN672CTW[1]	1.2 V	–8	STD	Lead-Free fpBGA	672	COM	149
LFE3-150EA-6FN1156CTW[1]	1.2 V	–6	STD	Lead-Free fpBGA	1156	COM	149
LFE3-150EA-7FN1156CTW[1]	1.2 V	–7	STD	Lead-Free fpBGA	1156	COM	149
LFE3-150EA-8FN1156CTW[1]	1.2 V	–8	STD	Lead-Free fpBGA	1156	COM	149

1. Note: Specifications for the LFE3-150EA-*sp*FN*pkg*CTW and LFE3-150EA-*sp*FN*pkg*ITW devices, (where *sp* is the speed and *pkg* is the package), are the same as the LFE3-150EA-*sp*FN*pkg*C and LFE3-150EA-*sp*FN*pkg*I devices respectively, except as specified below.
   - The CTC (Clock Tolerance Circuit) inside the SERDES hard PCS in the TW device is not functional but it can be bypassed and implemented in soft IP.
   - The SERDES XRES pin on the TW device passes CDM testing at 250 V.

Part Number	Voltage	Grade[1]	Power	Package	Pins	Temp.	LUTs (K)
LFE3-70EA-6FN484I	1.2 V	–6	STD	Lead-Free fpBGA	484	IND	67
LFE3-70EA-7FN484I	1.2 V	–7	STD	Lead-Free fpBGA	484	IND	67
LFE3-70EA-8FN484I	1.2 V	–8	STD	Lead-Free fpBGA	484	IND	67
LFE3-70EA-6LFN484I	1.2 V	–6	LOW	Lead-Free fpBGA	484	IND	67
LFE3-70EA-7LFN484I	1.2 V	–7	LOW	Lead-Free fpBGA	484	IND	67
LFE3-70EA-8LFN484I	1.2 V	–8	LOW	Lead-Free fpBGA	484	IND	67
LFE3-70EA-6FN672I	1.2 V	–6	STD	Lead-Free fpBGA	672	IND	67
LFE3-70EA-7FN672I	1.2 V	–7	STD	Lead-Free fpBGA	672	IND	67
LFE3-70EA-8FN672I	1.2 V	–8	STD	Lead-Free fpBGA	672	IND	67
LFE3-70EA-6LFN672I	1.2 V	–6	LOW	Lead-Free fpBGA	672	IND	67
LFE3-70EA-7LFN672I	1.2 V	–7	LOW	Lead-Free fpBGA	672	IND	67
LFE3-70EA-8LFN672I	1.2 V	–8	LOW	Lead-Free fpBGA	672	IND	67
LFE3-70EA-6FN1156I	1.2 V	–6	STD	Lead-Free fpBGA	1156	IND	67
LFE3-70EA-7FN1156I	1.2 V	–7	STD	Lead-Free fpBGA	1156	IND	67
LFE3-70EA-8FN1156I	1.2 V	–8	STD	Lead-Free fpBGA	1156	IND	67
LFE3-70EA-6LFN1156I	1.2 V	–6	LOW	Lead-Free fpBGA	1156	IND	67
LFE3-70EA-7LFN1156I	1.2 V	–7	LOW	Lead-Free fpBGA	1156	IND	67
LFE3-70EA-8LFN1156I	1.2 V	–8	LOW	Lead-Free fpBGA	1156	IND	67

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Part Number	Voltage	Grade[1]	Power	Package	Pins	Temp.	LUTs (K)
LFE3-95EA-6FN484I	1.2 V	–6	STD	Lead-Free fpBGA	484	IND	92
LFE3-95EA-7FN484I	1.2 V	–7	STD	Lead-Free fpBGA	484	IND	92
LFE3-95EA-8FN484I	1.2 V	–8	STD	Lead-Free fpBGA	484	IND	92
LFE3-95EA-6LFN484I	1.2 V	–6	LOW	Lead-Free fpBGA	484	IND	92
LFE3-95EA-7LFN484I	1.2 V	–7	LOW	Lead-Free fpBGA	484	IND	92
LFE3-95EA-8LFN484I	1.2 V	–8	LOW	Lead-Free fpBGA	484	IND	92
LFE3-95EA-6FN672I	1.2 V	–6	STD	Lead-Free fpBGA	672	IND	92
LFE3-95EA-7FN672I	1.2 V	–7	STD	Lead-Free fpBGA	672	IND	92
LFE3-95EA-8FN672I	1.2 V	–8	STD	Lead-Free fpBGA	672	IND	92
LFE3-95EA-6LFN672I	1.2 V	–6	LOW	Lead-Free fpBGA	672	IND	92
LFE3-95EA-7LFN672I	1.2 V	–7	LOW	Lead-Free fpBGA	672	IND	92
LFE3-95EA-8LFN672I	1.2 V	–8	LOW	Lead-Free fpBGA	672	IND	92
LFE3-95EA-6FN1156I	1.2 V	–6	STD	Lead-Free fpBGA	1156	IND	92
LFE3-95EA-7FN1156I	1.2 V	–7	STD	Lead-Free fpBGA	1156	IND	92
LFE3-95EA-8FN1156I	1.2 V	–8	STD	Lead-Free fpBGA	1156	IND	92
LFE3-95EA-6LFN1156I	1.2 V	–6	LOW	Lead-Free fpBGA	1156	IND	92
LFE3-95EA-7LFN1156I	1.2 V	–7	LOW	Lead-Free fpBGA	1156	IND	92
LFE3-95EA-8LFN1156I	1.2 V	–8	LOW	Lead-Free fpBGA	1156	IND	92

1. For ordering information on -9 speed grade devices, please contact your Lattice Sales Representative.

Date	Version	Section	Change Summary
			Updated Frequency to 150 Mbps in Table 3-11 Periodic Receiver Jitter Tolerance Specification
December 2010	01.7EA	Multiple	Data sheet made final. Removed “preliminary” headings.
			Removed data for 70E and 95E devices. A separate data sheet is available for these specific devices.
			Updated for Lattice Diamond design software.
		Introduction	Corrected number of user I/Os
		Architecture	Corrected the package type in Table 2-14 Available SERDES Quad per LatticeECP3 Devices.
			Updated description of General Purpose PLL
			Added additional information in the Flexible Quad SERDES Architecture section.
			Added footnotes and corrected the information in Table 2-16 Selectable master Clock (MCCLK) Frequencies During Configuration (Nominal).
			Updated Figure 2-16, Per Region Secondary Clock Selection.
			Updated description for On-Chip Programmable Termination.
			Added information about number of rows of DSP slices.
			Updated footnote 2 for Table 2-12, On-Chip Termination Options for Input Modes.
			Updated information for sysIO buffer pairs.
			Corrected minimum number of General Purpose PLLs (was 4, now 2).
		DC and Switching Characteristics	Regenerated sysCONFIG Port Timing figure.
			Added t_W (clock pulse width) in External Switching Characteristics table.
			Corrected units, revised and added data, and corrected footnote 1 in External Switching Characteristics table.
			Added Jitter Transfer figures in SERDES External Reference Clock section.
			Corrected capacitance information in the DC Electrical Characteristics table.
			Corrected data in the Register-to-Register Performance table.
			Corrected GDDR Parameter name HOGDDR.
			Corrected RSDS25 -7 data in Family Timing Adders table.
			Added footnotes 10-12 to DDR data information in the External Switching Characteristics table.
			Corrected titles for Figures 3-7 (DDR/DDR2/DDR3 Parameters) and 3-8 (Generic DDR/DDRX2 Parameters).
			Updated titles for Figures 3-5 (MLVDS25 (Multipoint Low Voltage Differential Signaling)) and 3-6 (Generic DDRX1/DDRX2 (With Clock and Data Edges Aligned)).
			Updated Supply Current table.
			Added GDDR interface information to the External Switching and Characteristics table.
			Added footnote to sysIO Recommended Operating Conditions table.
			Added footnote to LVDS25 table.
			Corrected DDR section footnotes and references.
			Corrected Hot Socketing support from “top and bottom banks” to “top and bottom I/O pins”.
		Pinout Information	Updated description for VTTx.