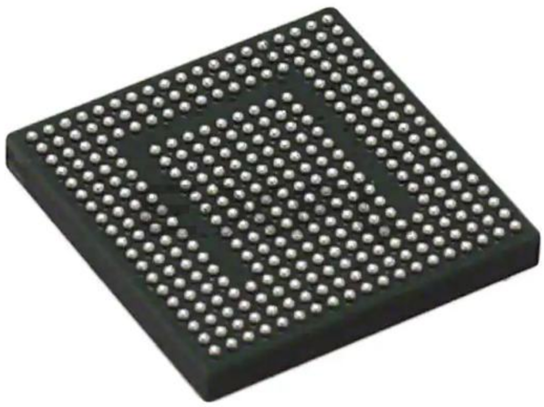

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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

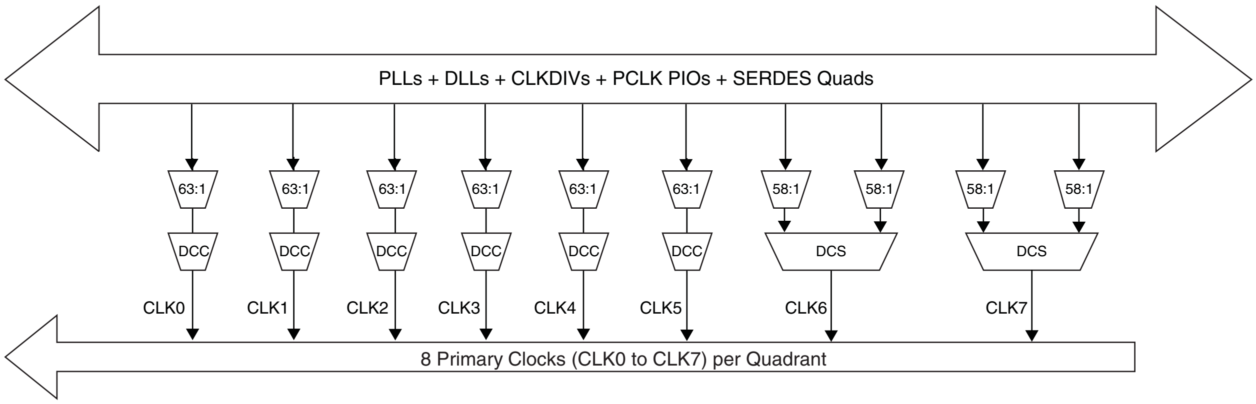
The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications,

Details	
Product Status	Active
Number of LABs/CLBs	2125
Number of Logic Elements/Cells	17000
Total RAM Bits	716800
Number of I/O	116
Number of Gates	-
Voltage - Supply	1.14V ~ 1.26V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	328-LFBGA, CSBGA
Supplier Device Package	328-CSBGA (10x10)
Purchase URL	https://www.e-xfl.com/product-detail/lattice-semiconductor/lfe3-17ea-7mg328c

Primary Clock Routing

The purpose of the primary clock routing is to distribute primary clock sources to the destination quadrants of the device. A global primary clock is a primary clock that is distributed to all quadrants. The clock routing structure in LatticeECP3 devices consists of a network of eight primary clock lines (CLK0 through CLK7) per quadrant. The primary clocks of each quadrant are generated from muxes located in the center of the device. All the clock sources are connected to these muxes. Figure 2-12 shows the clock routing for one quadrant. Each quadrant mux is identical. If desired, any clock can be routed globally.

Figure 2-12. Per Quadrant Primary Clock Selection



Dynamic Clock Control (DCC)

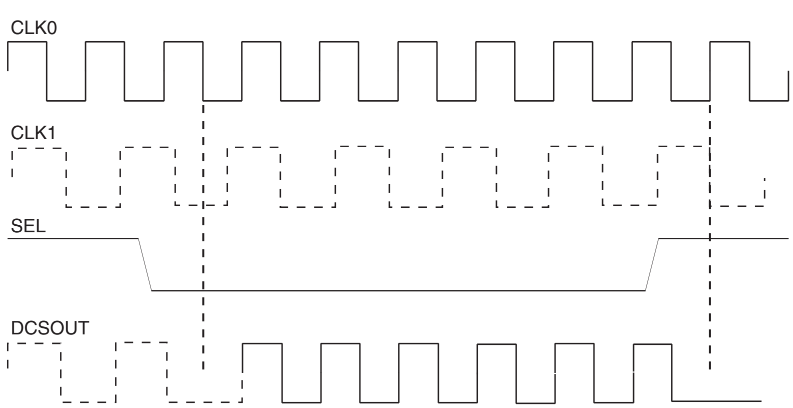
The DCC (Quadrant Clock Enable/Disable) feature allows internal logic control of the quadrant primary clock network. When a clock network is disabled, all the logic fed by that clock does not toggle, reducing the overall power consumption of the device.

Dynamic Clock Select (DCS)

The DCS is a smart multiplexer function available in the primary clock routing. It switches between two independent input clock sources without any glitches or runt pulses. This is achieved regardless of when the select signal is toggled. There are two DCS blocks per quadrant; in total, there are eight DCS blocks per device. The inputs to the DCS block come from the center muxes. The output of the DCS is connected to primary clocks CLK6 and CLK7 (see Figure 2-12).

Figure 2-13 shows the timing waveforms of the default DCS operating mode. The DCS block can be programmed to other modes. For more information about the DCS, please see the list of technical documentation at the end of this data sheet.

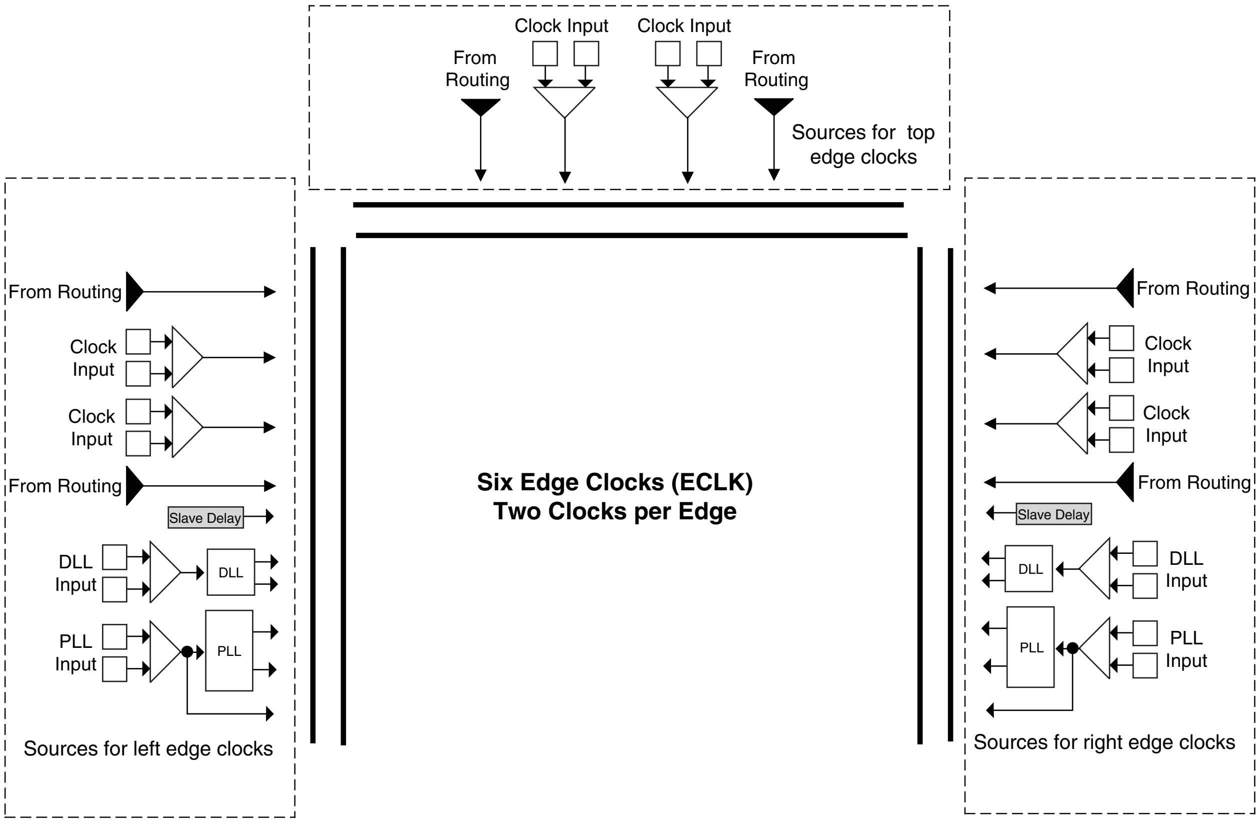
Figure 2-13. DCS Waveforms



Edge Clock Sources

Edge clock resources can be driven from a variety of sources at the same edge. Edge clock resources can be driven from adjacent edge clock PIOs, primary clock PIOs, PLLs, DLLs, Slave Delay and clock dividers as shown in Figure 2-19.

Figure 2-19. Edge Clock Sources



Notes:
1. Clock inputs can be configured in differential or single ended mode.
2. The two DLLs can also drive the two top edge clocks.
3. The top left and top right PLL can also drive the two top edge clocks.

Edge Clock Routing

LatticeECP3 devices have a number of high-speed edge clocks that are intended for use with the PIOs in the implementation of high-speed interfaces. There are six edge clocks per device: two edge clocks on each of the top, left, and right edges. Different PLL and DLL outputs are routed to the two muxes on the left and right sides of the device. In addition, the CLKINDEL signal (generated from the DLL Slave Delay Line block) is routed to all the edge clock muxes on the left and right sides of the device. Figure 2-20 shows the selection muxes for these clocks.

as, overflow, underflow and convergent rounding, etc.
— Flexible cascading across slices to get larger functions

- RTL Synthesis friendly synchronous reset on all registers, while still supporting asynchronous reset for legacy users
- Dynamic MUX selection to allow Time Division Multiplexing (TDM) of resources for applications that require processor-like flexibility that enables different functions for each clock cycle

For most cases, as shown in Figure 2-24, the LatticeECP3 DSP slice is backwards-compatible with the LatticeECP2™ sysDSP block, such that, legacy applications can be targeted to the LatticeECP3 sysDSP slice. The functionality of one LatticeECP2 sysDSP Block can be mapped into two adjacent LatticeECP3 sysDSP slices, as shown in Figure 2-25.

Figure 2-24. Simplified sysDSP Slice Block Diagram

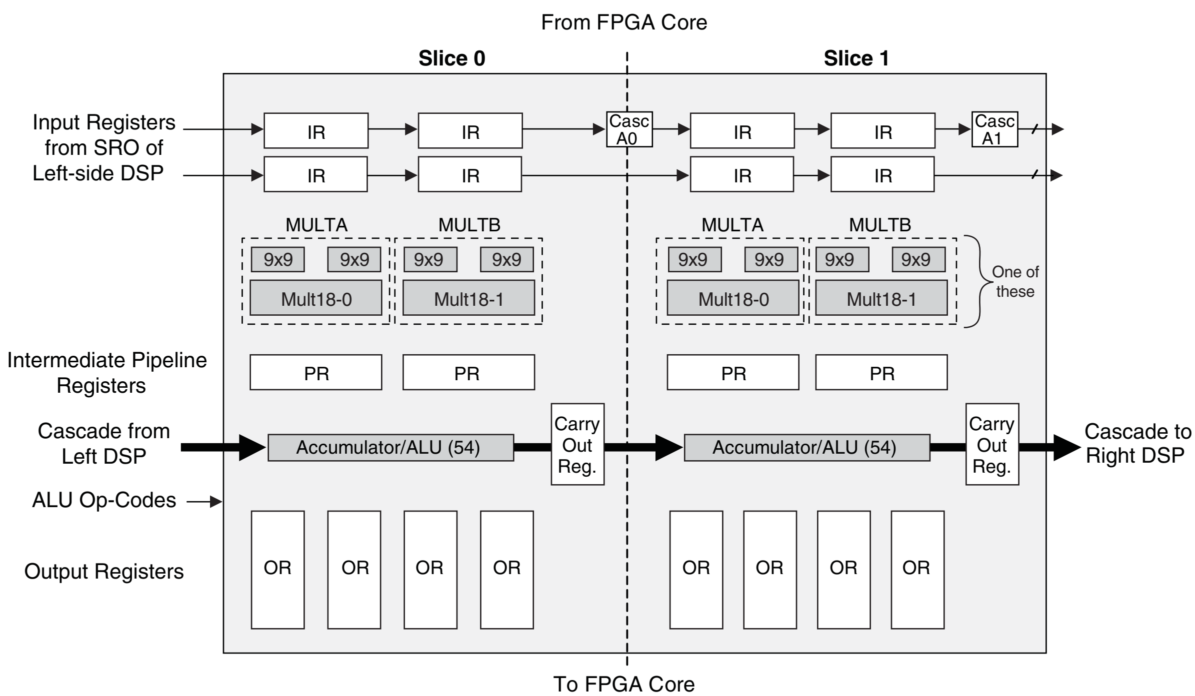
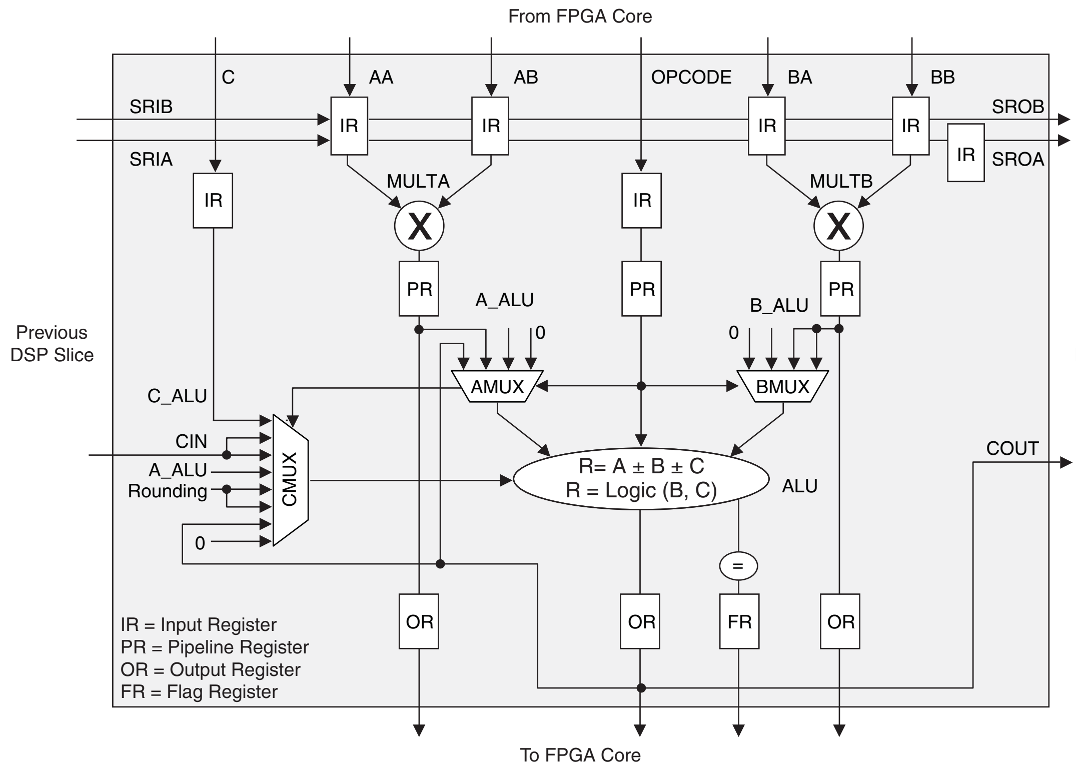


Figure 2-25. Detailed sysDSP Slice Diagram



Note: A_ALU, B_ALU and C_ALU are internal signals generated by combining bits from AA, AB, BA BB and C inputs. See TN1182, LatticeECP3 sysDSP Usage Guide, for further information.

The LatticeECP2 sysDSP block supports the following basic elements.

- MULT (Multiply)
- MAC (Multiply, Accumulate)
- MULTADDSUB (Multiply, Addition/Subtraction)
- MULTADDSUBSUM (Multiply, Addition/Subtraction, Summation)

Table 2-8 shows the capabilities of each of the LatticeECP3 slices versus the above functions.

Table 2-8. Maximum Number of Elements in a Slice

Width of Multiply	x9	x18	x36
MULT	4	2	1/2
MAC	1	1	—
MULTADDSUB	2	1	—
MULTADDSUBSUM	1[1]	1/2	—

1. One slice can implement 1/2 9x9 m9x9addsubsum and two m9x9addsubsum with two slices.

Some options are available in the four elements. The input register in all the elements can be directly loaded or can be loaded as a shift register from previous operand registers. By selecting “dynamic operation” the following operations are possible:

- In the Add/Sub option the Accumulator can be switched between addition and subtraction on every cycle.
- The loading of operands can switch between parallel and serial operations.

MMAC DSP Element

The LatticeECP3 supports a MAC with two multipliers. This is called Multiply Multiply Accumulate or MMAC. In this case, the two operands, AA and AB, are multiplied and the result is added with the previous accumulated value and with the result of the multiplier operation of operands BA and BB. This accumulated value is available at the output. The user can enable the input and pipeline registers, but the output register is always enabled. The output register is used to store the accumulated value. The ALU is configured as the accumulator in the sysDSP slice. A registered overflow signal is also available. The overflow conditions are provided later in this document. Figure 2-28 shows the MMAC sysDSP element.

Figure 2-28. MMAC sysDSP Element

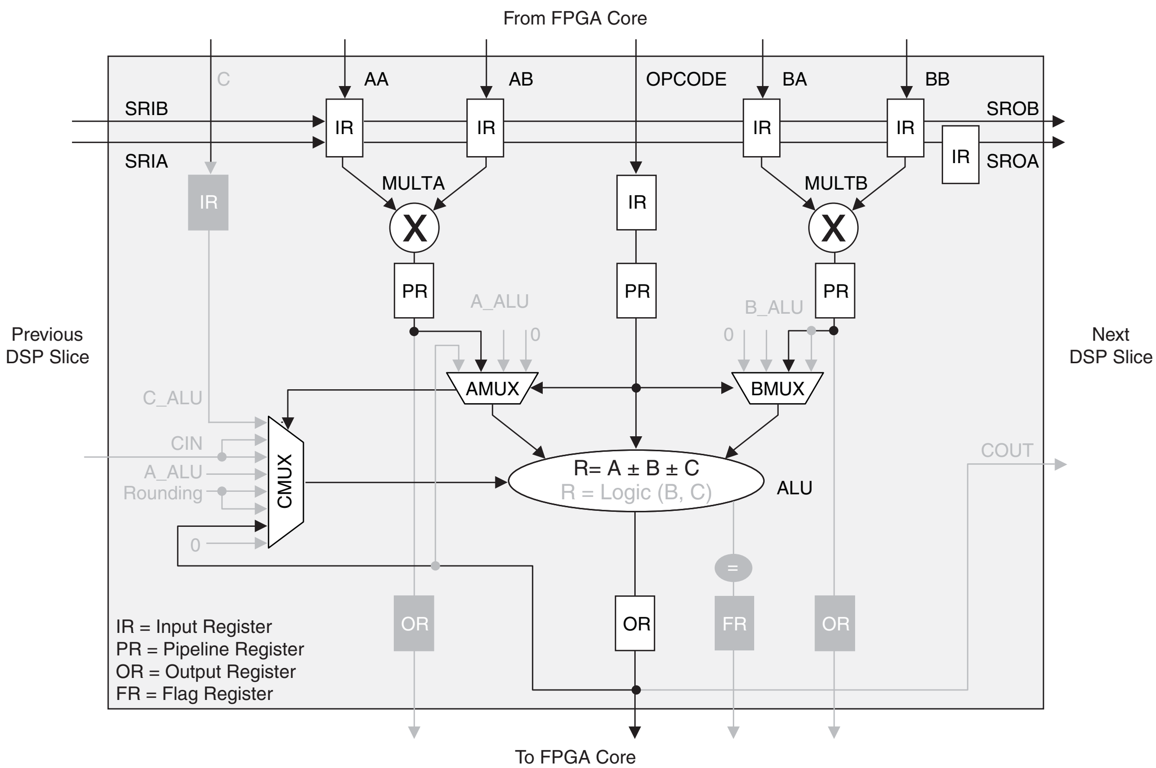
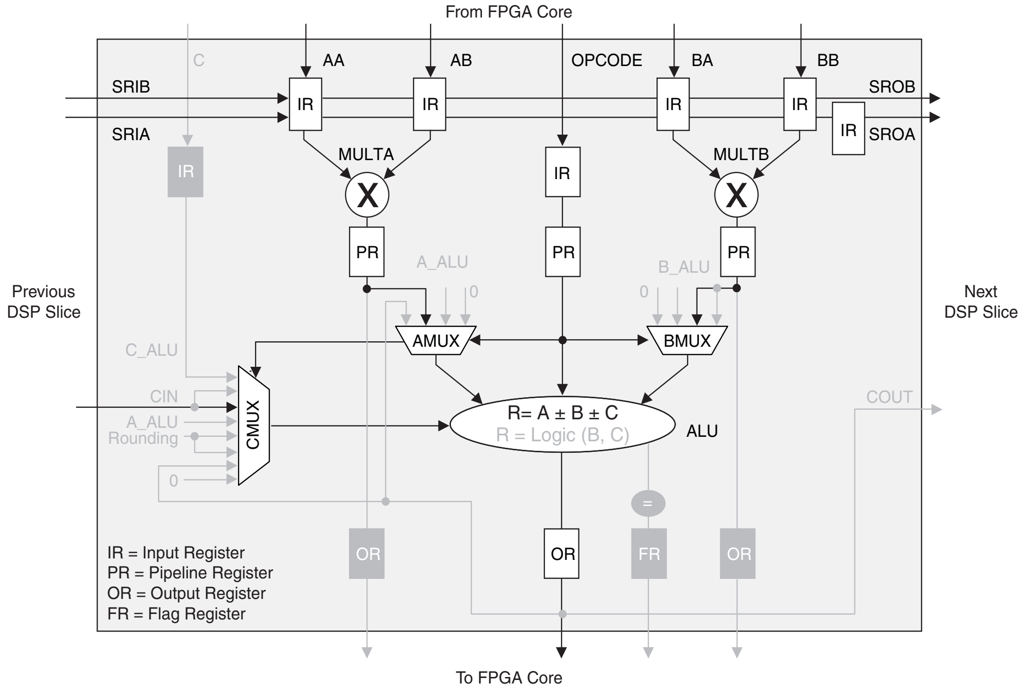


Figure 2-31. MULTADDSUBSUM Slice 1



Advanced sysDSP Slice Features

Cascading

The LatticeECP3 sysDSP slice has been enhanced to allow cascading. Adder trees are implemented fully in sysDSP slices, improving the performance. Cascading of slices uses the signals CIN, COUT and C Mux of the slice.

Addition

The LatticeECP3 sysDSP slice allows for the bypassing of multipliers and cascading of adder logic. High performance adder functions are implemented without the use of LUTs. The maximum width adders that can be implemented are 54-bit.

Rounding

The rounding operation is implemented in the ALU and is done by adding a constant followed by a truncation operation. The rounding methods supported are:

- Rounding to zero (RTZ)
- Rounding to infinity (RTI)
- Dynamic rounding
- Random rounding
- Convergent rounding

Two adjacent PIOs can be joined to provide a differential I/O pair (labeled as “T” and “C”) as shown in Figure 2-32. The PAD Labels “T” and “C” distinguish the two PIOs. Approximately 50% of the PIO pairs on the left and right edges of the device can be configured as true LVDS outputs. All I/O pairs can operate as LVDS inputs.

Table 2-11. PIO Signal List

Name	Type	Description
INDD	Input Data	Register bypassed input. This is not the same port as INCK.
IPA, INA, IPB, INB	Input Data	Ports to core for input data
OPOSA, ONEGA[1], OPOSB, ONEGB[1]	Output Data	Output signals from core. An exception is the ONEGB port, used for tristate logic at the DQS pad.
CE	PIO Control	Clock enables for input and output block flip-flops.
SCLK	PIO Control	System Clock (PCLK) for input and output/TS blocks. Connected from clock ISB.
LSR	PIO Control	Local Set/Reset
ECLK1, ECLK2	PIO Control	Edge clock sources. Entire PIO selects one of two sources using mux.
ECLKDQSR[1]	Read Control	From DQS_STROBE, shifted strobe for memory interfaces only.
DDRCLKPOL[1]	Read Control	Ensures transfer from DQS domain to SCLK domain.
DDRLAT[1]	Read Control	Used to guarantee INDDRX2 gearing by selectively enabling a D-Flip-Flop in datapath.
DEL[3:0]	Read Control	Dynamic input delay control bits.
INCK	To Clock Distribution and PLL	PIO treated as clock PIO, path to distribute to primary clocks and PLL.
TS	Tristate Data	Tristate signal from core (SDR)
DQCLK0[1], DQCLK1[1]	Write Control	Two clocks edges, 90 degrees out of phase, used in output gearing.
DQSW[2]	Write Control	Used for output and tristate logic at DQS only.
DYNDEL[7:0]	Write Control	Shifting of write clocks for specific DQS group, using 6:0 each step is approximately 25ps, 128 steps. Bit 7 is an invert (timing depends on input frequency). There is also a static control for this 8-bit setting, enabled with a memory cell.
DCNTL[6:0]	PIO Control	Original delay code from DDR DLL
DATAVALID[1]	Output Data	Status flag from DATAVALID logic, used to indicate when input data is captured in IOLOGIC and valid to core.
READ	For DQS_Strobe	Read signal for DDR memory interface
DQSI	For DQS_Strobe	Unshifted DQS strobe from input pad
PRMBDET	For DQS_Strobe	DQSI biased to go high when DQSI is tristate, goes to input logic block as well as core logic.
GSRN	Control from routing	Global Set/Reset

1. Signals available on left/right/top edges only.
2. Selected PIO.

PIO

The PIO contains four blocks: an input register block, output register block, tristate register block and a control logic block. These blocks contain registers for operating in a variety of modes along with the necessary clock and selection logic.

Input Register Block

The input register blocks for the PIOs, in the left, right and top edges, contain delay elements and registers that can be used to condition high-speed interface signals, such as DDR memory interfaces and source synchronous interfaces, before they are passed to the device core. Figure 2-33 shows the input register block for the left, right and top edges. The input register block for the bottom edge contains one element to register the input signal and no DDR registers. The following description applies to the input register block for PIOs in the left, right and top edges only.

Control Logic Block

The control logic block allows the selection and modification of control signals for use in the PIO block.

DDR Memory Support

Certain PICs have additional circuitry to allow the implementation of high-speed source synchronous and DDR, DDR2 and DDR3 memory interfaces. The support varies by the edge of the device as detailed below.

Left and Right Edges

The left and right sides of the PIC have fully functional elements supporting DDR, DDR2, and DDR3 memory interfaces. One of every 12 PIOs supports the dedicated DQS pins with the DQS control logic block. Figure 2-35 shows the DQS bus spanning 11 I/O pins. Two of every 12 PIOs support the dedicated DQS and DQS# pins with the DQS control logic block.

Bottom Edge

PICs on the bottom edge of the device do not support DDR memory and Generic DDR interfaces.

Top Edge

PICs on the top side are similar to the PIO elements on the left and right sides but do not support gearing on the output registers. Hence, the modes to support output/tristate DDR3 memory are removed on the top side.

The exact DQS pins are shown in a dual function in the Logic Signal Connections table in this data sheet. Additional detail is provided in the Signal Descriptions table. The DQS signal from the bus is used to strobe the DDR data from the memory into input register blocks. Interfaces on the left, right and top edges are designed for DDR memories that support 10 bits of data.

Figure 2-35. DQS Grouping on the Left, Right and Top Edges

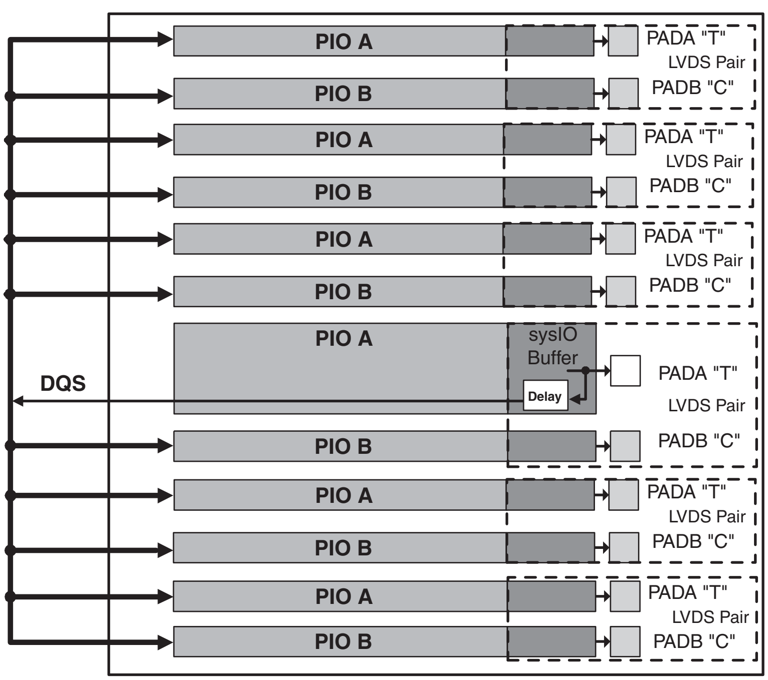
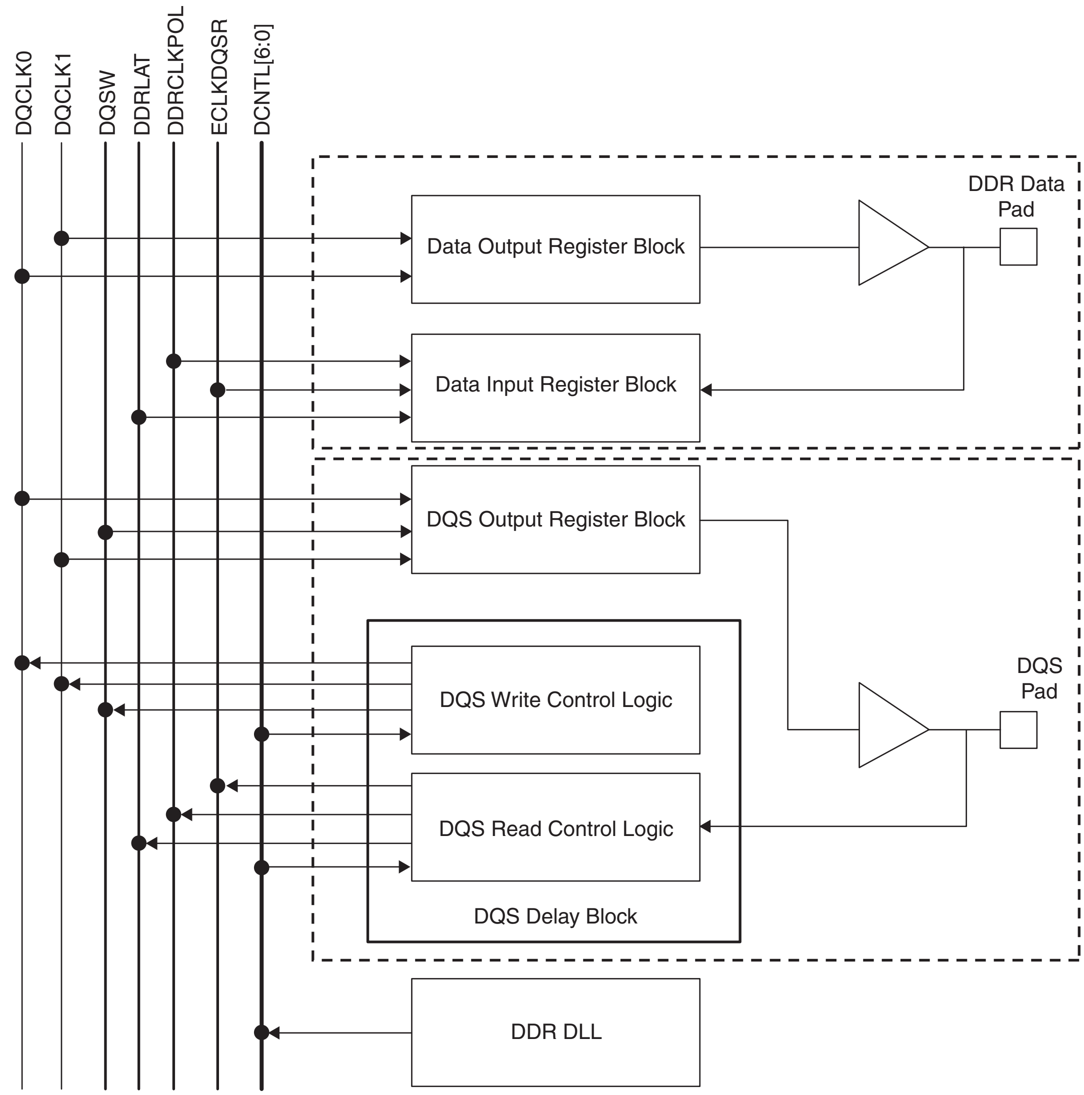


Figure 2-37. DQS Local Bus



Polarity Control Logic

In a typical DDR Memory interface design, the phase relationship between the incoming delayed DQS strobe and the internal system clock (during the READ cycle) is unknown. The LatticeECP3 family contains dedicated circuits to transfer data between these domains. A clock polarity selector is used to prevent set-up and hold violations at the domain transfer between DQS (delayed) and the system clock. This changes the edge on which the data is registered in the synchronizing registers in the input register block. This requires evaluation at the start of each READ cycle for the correct clock polarity.

Prior to the READ operation in DDR memories, DQS is in tristate (pulled by termination). The DDR memory device drives DQS low at the start of the preamble state. A dedicated circuit detects the first DQS rising edge after the preamble state. This signal is used to control the polarity of the clock to the synchronizing registers.

DDR3 Memory Support

LatticeECP3 supports the read and write leveling required for DDR3 memory interfaces.

Read leveling is supported by the use of the DDRCLKPOL and the DDRLAT signals generated in the DQS Read Control logic block. These signals dynamically control the capture of the data with respect to the DQS at the input register block.

LVDS25E

The top and bottom sides of LatticeECP3 devices support LVDS outputs via emulated complementary LVCMOS outputs in conjunction with a parallel resistor across the driver outputs. The scheme shown in Figure 3-1 is one possible solution for point-to-point signals.

Figure 3-1. LVDS25E Output Termination Example

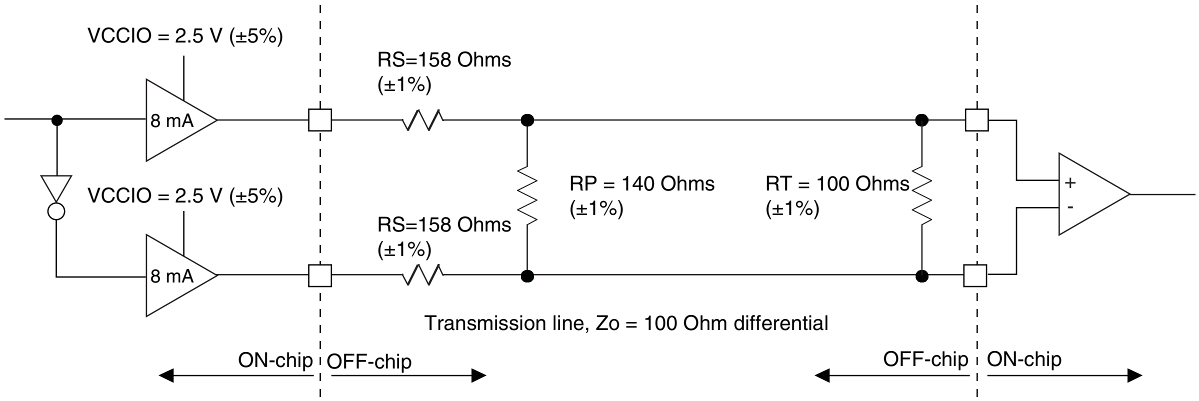


Table 3-1. LVDS25E DC Conditions

Parameter	Description	Typical	Units
V_{CCIO}	Output Driver Supply (+/–5%)	2.50	V
Z_{OUT}	Driver Impedance	20	Ω
R_S	Driver Series Resistor (+/–1%)	158	Ω
R_P	Driver Parallel Resistor (+/–1%)	140	Ω
R_T	Receiver Termination (+/–1%)	100	Ω
V_{OH}	Output High Voltage	1.43	V
V_{OL}	Output Low Voltage	1.07	V
V_{OD}	Output Differential Voltage	0.35	V
V_{CM}	Output Common Mode Voltage	1.25	V
Z_{BACK}	Back Impedance	100.5	Ω
I_{DC}	DC Output Current	6.03	mA

LVCMOS33D

All I/O banks support emulated differential I/O using the LVCMOS33D I/O type. This option, along with the external resistor network, provides the system designer the flexibility to place differential outputs on an I/O bank with 3.3 V V_{CCIO}. The default drive current for LVCMOS33D output is 12 mA with the option to change the device strength to 4 mA, 8 mA, 16 mA or 20 mA. Follow the LVCMOS33 specifications for the DC characteristics of the LVCMOS33D.

LatticeECP3 External Switching Characteristics (Continued)[1, 2, 3, 13]

Over Recommended Commercial Operating Conditions

Parameter	Description	Device	–8		–7		–6		Units
			Min.	Max.	Min.	Max.	Min.	Max.	
t_{H_DEL}	Clock to Data Hold - PIO Input Register with Input Data Delay	ECP3-150EA	0.0	—	0.0	—	0.0	—	ns
f_{MAX_IO}	Clock Frequency of I/O and PFU Register	ECP3-150EA	—	500	—	420	—	375	MHz
t_{CO}	Clock to Output - PIO Output Register	ECP3-70EA/95EA	—	3.8	—	4.2	—	4.6	ns
t_{SU}	Clock to Data Setup - PIO Input Register	ECP3-70EA/95EA	0.0	—	0.0	—	0.0	—	ns
t_H	Clock to Data Hold - PIO Input Register	ECP3-70EA/95EA	1.4	—	1.6	—	1.8	—	ns
t_{SU_DEL}	Clock to Data Setup - PIO Input Register with Data Input Delay	ECP3-70EA/95EA	1.3	—	1.5	—	1.7	—	ns
t_{H_DEL}	Clock to Data Hold - PIO Input Register with Input Data Delay	ECP3-70EA/95EA	0.0	—	0.0	—	0.0	—	ns
f_{MAX_IO}	Clock Frequency of I/O and PFU Register	ECP3-70EA/95EA	—	500	—	420	—	375	MHz
t_{CO}	Clock to Output - PIO Output Register	ECP3-35EA	—	3.7	—	4.1	—	4.5	ns
t_{SU}	Clock to Data Setup - PIO Input Register	ECP3-35EA	0.0	—	0.0	—	0.0	—	ns
t_H	Clock to Data Hold - PIO Input Register	ECP3-35EA	1.2	—	1.4	—	1.6	—	ns
t_{SU_DEL}	Clock to Data Setup - PIO Input Register with Data Input Delay	ECP3-35EA	1.3	—	1.4	—	1.5	—	ns
t_{H_DEL}	Clock to Data Hold - PIO Input Register with Input Data Delay	ECP3-35EA	0.0	—	0.0	—	0.0	—	ns
f_{MAX_IO}	Clock Frequency of I/O and PFU Register	ECP3-35EA	—	500	—	420	—	375	MHz
t_{CO}	Clock to Output - PIO Output Register	ECP3-17EA	—	3.5	—	3.9	—	4.3	ns
t_{SU}	Clock to Data Setup - PIO Input Register	ECP3-17EA	0.0	—	0.0	—	0.0	—	ns
t_H	Clock to Data Hold - PIO Input Register	ECP3-17EA	1.3	—	1.5	—	1.6	—	ns
t_{SU_DEL}	Clock to Data Setup - PIO Input Register with Data Input Delay	ECP3-17EA	1.3	—	1.4	—	1.5	—	ns
t_{H_DEL}	Clock to Data Hold - PIO Input Register with Input Data Delay	ECP3-17EA	0.0	—	0.0	—	0.0	—	ns
f_{MAX_IO}	Clock Frequency of I/O and PFU Register	ECP3-17EA	—	500	—	420	—	375	MHz
General I/O Pin Parameters Using Dedicated Clock Input Primary Clock with PLL with Clock Injection Removal Setting[2]									
t_{COPLL}	Clock to Output - PIO Output Register	ECP3-150EA	—	3.3	—	3.6	—	39	ns
t_{SUPLL}	Clock to Data Setup - PIO Input Register	ECP3-150EA	0.7	—	0.8	—	0.9	—	ns
t_{HPLL}	Clock to Data Hold - PIO Input Register	ECP3-150EA	0.8	—	0.9	—	1.0	—	ns
t_{SU_DELPLL}	Clock to Data Setup - PIO Input Register with Data Input Delay	ECP3-150EA	1.6	—	1.8	—	2.0	—	ns
t_{H_DELPLL}	Clock to Data Hold - PIO Input Register with Input Data Delay	ECP3-150EA	—	0.0	—	0.0	—	0.0	ns
t_{COPLL}	Clock to Output - PIO Output Register	ECP3-70EA/95EA	—	3.3	—	3.5	—	3.8	ns
t_{SUPLL}	Clock to Data Setup - PIO Input Register	ECP3-70EA/95EA	0.7	—	0.8	—	0.9	—	ns

LatticeECP3 External Switching Characteristics (Continued)[1, 2, 3, 13]

Over Recommended Commercial Operating Conditions

Parameter	Description	Device	–8 Min.	–8 Max.	–7 Min.	–7 Max.	–6 Min.	–6 Max.	Units
f_{MAX_GDDR}	DDRX1 Clock Frequency	ECP3-70EA/95EA	—	250	—	250	—	250	MHz
$t_{DVBGDDR}$	Data Valid Before CLK	ECP3-35EA	683	—	688	—	690	—	ps
$t_{DVAGDDR}$	Data Valid After CLK	ECP3-35EA	683	—	688	—	690	—	ps
f_{MAX_GDDR}	DDRX1 Clock Frequency	ECP3-35EA	—	250	—	250	—	250	MHz
$t_{DVBGDDR}$	Data Valid Before CLK	ECP3-17EA	683	—	688	—	690	—	ps
$t_{DVAGDDR}$	Data Valid After CLK	ECP3-17EA	683	—	688	—	690	—	ps
f_{MAX_GDDR}	DDRX1 Clock Frequency	ECP3-17EA	—	250	—	250	—	250	MHz
Generic DDRX1 Output with Clock and Data Aligned at Pin (GDDRX1_TX.SCLK.Aligned)[10]									
$t_{DIBGDDR}$	Data Invalid Before Clock	ECP3-150EA	—	335	—	338	—	341	ps
$t_{DIAGDDR}$	Data Invalid After Clock	ECP3-150EA	—	335	—	338	—	341	ps
f_{MAX_GDDR}	DDRX1 Clock Frequency	ECP3-150EA	—	250	—	250	—	250	MHz
$t_{DIBGDDR}$	Data Invalid Before Clock	ECP3-70EA/95EA	—	339	—	343	—	347	ps
$t_{DIAGDDR}$	Data Invalid After Clock	ECP3-70EA/95EA	—	339	—	343	—	347	ps
f_{MAX_GDDR}	DDRX1 Clock Frequency	ECP3-70EA/95EA	—	250	—	250	—	250	MHz
$t_{DIBGDDR}$	Data Invalid Before Clock	ECP3-35EA	—	322	—	320	—	321	ps
$t_{DIAGDDR}$	Data Invalid After Clock	ECP3-35EA	—	322	—	320	—	321	ps
f_{MAX_GDDR}	DDRX1 Clock Frequency	ECP3-35EA	—	250	—	250	—	250	MHz
$t_{DIBGDDR}$	Data Invalid Before Clock	ECP3-17EA	—	322	—	320	—	321	ps
$t_{DIAGDDR}$	Data Invalid After Clock	ECP3-17EA	—	322	—	320	—	321	ps
f_{MAX_GDDR}	DDRX1 Clock Frequency	ECP3-17EA	—	250	—	250	—	250	MHz
Generic DDRX1 Output with Clock and Data (<10 Bits Wide) Centered at Pin (GDDRX1_TX.DQS.Centered)[10]									
Left and Right Sides									
$t_{DVBGDDR}$	Data Valid Before CLK	ECP3-150EA	670	—	670	—	670	—	ps
$t_{DVAGDDR}$	Data Valid After CLK	ECP3-150EA	670	—	670	—	670	—	ps
f_{MAX_GDDR}	DDRX1 Clock Frequency	ECP3-150EA	—	250	—	250	—	250	MHz
$t_{DVBGDDR}$	Data Valid Before CLK	ECP3-70EA/95EA	657	—	652	—	650	—	ps
$t_{DVAGDDR}$	Data Valid After CLK	ECP3-70EA/95EA	657	—	652	—	650	—	ps
f_{MAX_GDDR}	DDRX1 Clock Frequency	ECP3-70EA/95EA	—	250	—	250	—	250	MHz
$t_{DVBGDDR}$	Data Valid Before CLK	ECP3-35EA	670	—	675	—	676	—	ps
$t_{DVAGDDR}$	Data Valid After CLK	ECP3-35EA	670	—	675	—	676	—	ps
f_{MAX_GDDR}	DDRX1 Clock Frequency	ECP3-35EA	—	250	—	250	—	250	MHz
$t_{DVBGDDR}$	Data Valid Before CLK	ECP3-17EA	670	—	670	—	670	—	ps
$t_{DVAGDDR}$	Data Valid After CLK	ECP3-17EA	670	—	670	—	670	—	ps
f_{MAX_GDDR}	DDRX1 Clock Frequency	ECP3-17EA	—	250	—	250	—	250	MHz
Generic DDRX2 Output with Clock and Data (>10 Bits Wide) Aligned at Pin (GDDRX2_TX.Aligned)									
Left and Right Sides									
$t_{DIBGDDR}$	Data Invalid Before Clock	All ECP3EA Devices	—	200	—	210	—	220	ps
$t_{DIAGDDR}$	Data Invalid After Clock	All ECP3EA Devices	—	200	—	210	—	220	ps
f_{MAX_GDDR}	DDRX2 Clock Frequency	All ECP3EA Devices	—	500	—	420	—	375	MHz
Generic DDRX2 Output with Clock and Data (>10 Bits Wide) Centered at Pin Using DQSDLL (GDDRX2_TX.DQSDLL.Centered)[11]									
Left and Right Sides									
$t_{DVBGDDR}$	Data Valid Before CLK	All ECP3EA Devices	400	—	400	—	431	—	ps
$t_{DVAGDDR}$	Data Valid After CLK	All ECP3EA Devices	400	—	400	—	432	—	ps
f_{MAX_GDDR}	DDRX2 Clock Frequency	All ECP3EA Devices	—	400	—	400	—	375	MHz

DLL Timing

Over Recommended Operating Conditions

Parameter	Description	Condition	Min.	Typ.	Max.	Units
f_{REF}	Input reference clock frequency (on-chip or off-chip)		133	—	500	MHz
f_{FB}	Feedback clock frequency (on-chip or off-chip)		133	—	500	MHz
f_{CLKOP}[1]	Output clock frequency, CLKOP		133	—	500	MHz
f_{CLKOS}[2]	Output clock frequency, CLKOS		33.3	—	500	MHz
t_{PJIT}	Output clock period jitter (clean input)			—	200	ps p-p
t_{DUTY}	Output clock duty cycle (at 50% levels, 50% duty cycle input clock, 50% duty cycle circuit turned off, time reference delay mode)	Edge Clock	40		60	%
		Primary Clock	30		70	%
$t_{DUTYTRD}$	Output clock duty cycle (at 50% levels, arbitrary duty cycle input clock, 50% duty cycle circuit enabled, time reference delay mode)	Primary Clock < 250 MHz	45		55	%
		Primary Clock ≥ 250 MHz	30		70	%
		Edge Clock	45		55	%
$t_{DUTYCIR}$	Output clock duty cycle (at 50% levels, arbitrary duty cycle input clock, 50% duty cycle circuit enabled, clock injection removal mode) with DLL cascading	Primary Clock < 250 MHz	40		60	%
		Primary Clock ≥ 250 MHz	30		70	%
		Edge Clock	45		55	%
t_{SKEW}[3]	Output clock to clock skew between two outputs with the same phase setting		—	—	100	ps
t_{PHASE}	Phase error measured at device pads between off-chip reference clock and feedback clocks		—	—	+/-400	ps
t_{PWH}	Input clock minimum pulse width high (at 80% level)		550	—	—	ps
t_{PWL}	Input clock minimum pulse width low (at 20% level)		550	—	—	ps
t_{INSTB}	Input clock period jitter		—	—	500	ps
t_{LOCK}	DLL lock time		8	—	8200	cycles
t_{RSWD}	Digital reset minimum pulse width (at 80% level)		3	—	—	ns
t_{DEL}	Delay step size		27	45	70	ps
t_{RANGE1}	Max. delay setting for single delay block (64 taps)		1.9	3.1	4.4	ns
t_{RANGE4}	Max. delay setting for four chained delay blocks		7.6	12.4	17.6	ns

1. CLKOP runs at the same frequency as the input clock.
2. CLKOS minimum frequency is obtained with divide by 4.
3. This is intended to be a “path-matching” design guideline and is not a measurable specification.

Serial Rapid I/O Type 2/CPRI LV E.24 Electrical and Timing Characteristics

AC and DC Characteristics

Table 3-15. Transmit

Symbol	Description	Test Conditions	Min.	Typ.	Max.	Units
T_{RF}[1]	Differential rise/fall time	20%-80%	—	80	—	ps
$Z_{TX\_DIFF\_DC}$	Differential impedance		80	100	120	Ohms
J_{TX_DDJ}[3, 4, 5]	Output data deterministic jitter		—	—	0.17	UI
J_{TX_TJ}[2, 3, 4, 5]	Total output data jitter		—	—	0.35	UI

1. Rise and Fall times measured with board trace, connector and approximately 2.5pf load.
2. Total jitter includes both deterministic jitter and random jitter. The random jitter is the total jitter minus the actual deterministic jitter.
3. Jitter values are measured with each CML output AC coupled into a 50-Ohm impedance (100-Ohm differential impedance).
4. Jitter and skew are specified between differential crossings of the 50% threshold of the reference signal.
5. Values are measured at 2.5 Gbps.

Table 3-16. Receive and Jitter Tolerance

Symbol	Description	Test Conditions	Min.	Typ.	Max.	Units
RL_{RX_DIFF}	Differential return loss	From 100 MHz to 2.5 GHz	10	—	—	dB
RL_{RX_CM}	Common mode return loss	From 100 MHz to 2.5 GHz	6	—	—	dB
Z_{RX_DIFF}	Differential termination resistance		80	100	120	Ohms
J_{RX_DJ}[2, 3, 4, 5]	Deterministic jitter tolerance (peak-to-peak)		—	—	0.37	UI
J_{RX_RJ}[2, 3, 4, 5]	Random jitter tolerance (peak-to-peak)		—	—	0.18	UI
J_{RX_SJ}[2, 3, 4, 5]	Sinusoidal jitter tolerance (peak-to-peak)		—	—	0.10	UI
J_{RX_TJ}[1, 2, 3, 4, 5]	Total jitter tolerance (peak-to-peak)		—	—	0.65	UI
T_{RX_EYE}	Receiver eye opening		0.35	—	—	UI

1. Total jitter includes deterministic jitter, random jitter and sinusoidal jitter. The sinusoidal jitter tolerance mask is shown in Figure 3-18.
2. Jitter values are measured with each high-speed input AC coupled into a 50-Ohm impedance.
3. Jitter and skew are specified between differential crossings of the 50% threshold of the reference signal.
4. Jitter tolerance, Differential Input Sensitivity and Receiver Eye Opening parameters are characterized when Full Rx Equalization is enabled.
5. Values are measured at 2.5 Gbps.

LatticeECP3 sysCONFIG Port Timing Specifications (Continued)

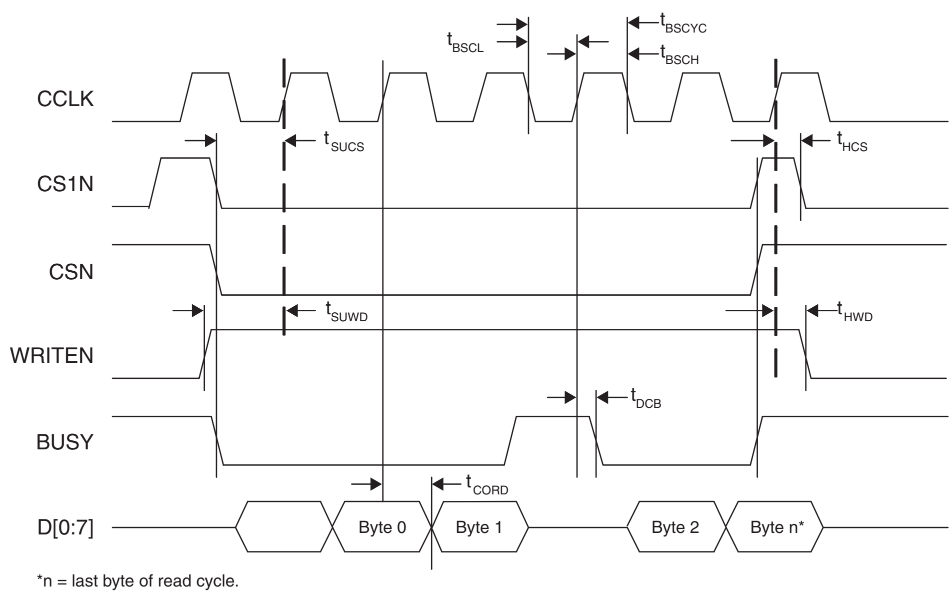
Over Recommended Operating Conditions

Parameter	Description	Min.	Max.	Units
t_{SSCL}	CCLK Minimum Low Pulse	5	—	ns
t_{HLCH}	HOLDN Low Setup Time (Relative to CCLK)	5	—	ns
t_{CHHH}	HOLDN Low Hold Time (Relative to CCLK)	5	—	ns
Master and Slave SPI (Continued)				
t_{CHHL}	HOLDN High Hold Time (Relative to CCLK)	5	—	ns
t_{HHCH}	HOLDN High Setup Time (Relative to CCLK)	5	—	ns
t_{HLQZ}	HOLDN to Output High-Z	—	9	ns
t_{HHQX}	HOLDN to Output Low-Z	—	9	ns

1. Re-toggling the PROGRAMN pin is not permitted until the INITN pin is high. Avoid consecutive toggling of the PROGRAMN.

Parameter	Min.	Max.	Units
Master Clock Frequency	Selected value - 15%	Selected value + 15%	MHz
Duty Cycle	40	60	%

Figure 3-20. sysCONFIG Parallel Port Read Cycle



Signal Descriptions (Cont.)

Signal Name	I/O	Description
[LOC]DQS[num]	I/O	DQ input/output pads: T (top), R (right), B (bottom), L (left), DQS, num = ball function number.
[LOC]DQ[num]	I/O	DQ input/output pads: T (top), R (right), B (bottom), L (left), DQ, associated DQS number.
Test and Programming (Dedicated Pins)		
TMS	I	Test Mode Select input, used to control the 1149.1 state machine. Pull-up is enabled during configuration.
TCK	I	Test Clock input pin, used to clock the 1149.1 state machine. No pull-up enabled.
TDI	I	Test Data in pin. Used to load data into device using 1149.1 state machine. After power-up, this TAP port can be activated for configuration by sending appropriate command. (Note: once a configuration port is selected it is locked. Another configuration port cannot be selected until the power-up sequence). Pull-up is enabled during configuration.
TDO	O	Output pin. Test Data Out pin used to shift data out of a device using 1149.1.
VCCJ	—	Power supply pin for JTAG Test Access Port.
Configuration Pads (Used During sysCONFIG)		
CFG[2:0]	I	Mode pins used to specify configuration mode values latched on rising edge of INITN. During configuration, a pull-up is enabled. These are dedicated pins.
INITN	I/O	Open Drain pin. Indicates the FPGA is ready to be configured. During configuration, a pull-up is enabled. It is a dedicated pin.
PROGRAMN	I	Initiates configuration sequence when asserted low. This pin always has an active pull-up. It is a dedicated pin.
DONE	I/O	Open Drain pin. Indicates that the configuration sequence is complete, and the startup sequence is in progress. It is a dedicated pin.
CCLK	I	Input Configuration Clock for configuring an FPGA in Slave SPI, Serial, and CPU modes. It is a dedicated pin.
MCLK	I/O	Output Configuration Clock for configuring an FPGA in SPI, SPIm, and Master configuration modes.
BUSY/SISPI	O	Parallel configuration mode busy indicator. SPI/SPIm mode data output.
CSN/SN/OEN	I/O	Parallel configuration mode active-low chip select. Slave SPI chip select. Parallel burst Flash output enable.
CS1N/HOLDN/RDY	I	Parallel configuration mode active-low chip select. Slave SPI hold input.
WRITEN	I	Write enable for parallel configuration modes.
DOUT/CSON/CSSPI1N	O	Serial data output. Chip select output. SPI/SPIm mode chip select.
D[0]/SPIFASTN	I/O	sysCONFIG Port Data I/O for Parallel mode. Open drain during configuration.
		sysCONFIG Port Data I/O for SPI or SPIm. When using the SPI or SPIm mode, this pin should either be tied high or low, must not be left floating. Open drain during configuration.
D1	I/O	Parallel configuration I/O. Open drain during configuration.
D2	I/O	Parallel configuration I/O. Open drain during configuration.
D3/SI	I/O	Parallel configuration I/O. Slave SPI data input. Open drain during configuration.
D4/SO	I/O	Parallel configuration I/O. Slave SPI data output. Open drain during configuration.
D5	I/O	Parallel configuration I/O. Open drain during configuration.
D6/SPID1	I/O	Parallel configuration I/O. SPI/SPIm data input. Open drain during configuration.

Pin Information Summary (Cont.)

Pin Information Summary		ECP3-70EA		
Pin Type		484 fpBGA	672 fpBGA	1156 fpBGA
Emulated Differential I/O per Bank	Bank 0	21	30	43
	Bank 1	18	24	39
	Bank 2	8	12	13
	Bank 3	20	23	33
	Bank 6	22	25	33
	Bank 7	11	16	18
	Bank 8	12	12	12
High-Speed Differential I/O per Bank	Bank 0	0	0	0
	Bank 1	0	0	0
	Bank 2	6	9	9
	Bank 3	9	12	16
	Bank 6	11	14	16
	Bank 7	9	12	13
	Bank 8	0	0	0
Total Single-Ended/ Total Differential I/O per Bank	Bank 0	42/21	60/30	86/43
	Bank 1	36/18	48/24	78/39
	Bank 2	28/14	42/21	44/22
	Bank 3	58/29	71/35	98/49
	Bank 6	67/33	78/39	98/49
	Bank 7	40/20	56/28	62/31
	Bank 8	24/12	24/12	24/12
DDR Groups Bonded per Bank[1]	Bank 0	3	5	7
	Bank 1	3	4	7
	Bank 2	2	3	3
	Bank 3	3	4	5
	Bank 6	4	4	5
	Bank 7	3	4	4
	Configuration Bank 8	0	0	0
SERDES Quads		1	2	3

1. Some DQS groups may not support DQS-12. Refer to the device pinout (.csv) file.

Pin Information Summary (Cont.)

Pin Information Summary		ECP3-95EA			ECP3-150EA	
Pin Type		484 fpBGA	672 fpBGA	1156 fpBGA	672 fpBGA	1156 fpBGA
General Purpose Inputs/Outputs per bank	Bank 0	42	60	86	60	94
	Bank 1	36	48	78	48	86
	Bank 2	24	34	36	34	58
	Bank 3	54	59	86	59	104
	Bank 6	63	67	86	67	104
	Bank 7	36	48	54	48	76
	Bank 8	24	24	24	24	24
General Purpose Inputs per Bank	Bank 0	0	0	0	0	0
	Bank 1	0	0	0	0	0
	Bank 2	4	8	8	8	8
	Bank 3	4	12	12	12	12
	Bank 6	4	12	12	12	12
	Bank 7	4	8	8	8	8
	Bank 8	0	0	0	0	0
General Purpose Outputs per Bank	Bank 0	0	0	0	0	0
	Bank 1	0	0	0	0	0
	Bank 2	0	0	0	0	0
	Bank 3	0	0	0	0	0
	Bank 6	0	0	0	0	0
	Bank 7	0	0	0	0	0
	Bank 8	0	0	0	0	0
Total Single-Ended User I/O		295	380	490	380	586
VCC		16	32	32	32	32
VCCAUX		8	12	16	12	16
VTT		4	4	8	4	8
VCCA		4	8	16	8	16
VCCPLL		4	4	4	4	4
VCCIO	Bank 0	2	4	4	4	4
	Bank 1	2	4	4	4	4
	Bank 2	2	4	4	4	4
	Bank 3	2	4	4	4	4
	Bank 6	2	4	4	4	4
	Bank 7	2	4	4	4	4
	Bank 8	2	2	2	2	2
VCCJ		1	1	1	1	1
TAP		4	4	4	4	4
GND, GNDIO		98	139	233	139	233
NC		0	0	238	0	116
Reserved[1]		2	2	2	2	2
SERDES		26	52	78	52	104
Miscellaneous Pins		8	8	8	8	8
Total Bonded Pins		484	672	1156	672	1156

LatticeECP3 Family Data Sheet
Ordering Information

LatticeECP3 Part Number Description

LFE3 – XXX XX – X X XXXXXX X

Device Family
ECP3 (LatticeECP3 FPGA + SERDES)

Logic Capacity
17 = 17K LUTs
35 = 33K LUTs
70 = 67K LUTs
95 = 92K LUTs
150 = 149K LUTs

Supply Voltage
EA = 1.2 V

Speed
6 = Slowest
7
8 = Fastest for LatticeECP3-17EA

Power
Blank = Standard Power
L = Low Power

Package
FTN256 = 256-ball Lead-Free ftBGA
FN484 = 484-ball Lead-Free fpBGA
FN672 = 672-ball Lead-Free fpBGA
FN1156 = 1156-ball Lead-Free fpBGA
MG328 = 328-ball Green csBGA[1]

Grade
C = Commercial
I = Industrial

1. Green = Halogen free and lead free.

Ordering Information

LatticeECP3 devices have top-side markings, for commercial and industrial grades, as shown below:

Commercial	Industrial
LATTICE LFE3-95EA 7FN672C Datecode	LATTICE LFE3-95EA 7FN672I Datecode

Note: See PCN 05A-12 for information regarding a change to the top-side mark logo.

Part Number	Voltage	Grade[1]	Power	Package	Pins	Temp.	LUTs (K)
LFE3-70EA-6FN484I	1.2 V	–6	STD	Lead-Free fpBGA	484	IND	67
LFE3-70EA-7FN484I	1.2 V	–7	STD	Lead-Free fpBGA	484	IND	67
LFE3-70EA-8FN484I	1.2 V	–8	STD	Lead-Free fpBGA	484	IND	67
LFE3-70EA-6LFN484I	1.2 V	–6	LOW	Lead-Free fpBGA	484	IND	67
LFE3-70EA-7LFN484I	1.2 V	–7	LOW	Lead-Free fpBGA	484	IND	67
LFE3-70EA-8LFN484I	1.2 V	–8	LOW	Lead-Free fpBGA	484	IND	67
LFE3-70EA-6FN672I	1.2 V	–6	STD	Lead-Free fpBGA	672	IND	67
LFE3-70EA-7FN672I	1.2 V	–7	STD	Lead-Free fpBGA	672	IND	67
LFE3-70EA-8FN672I	1.2 V	–8	STD	Lead-Free fpBGA	672	IND	67
LFE3-70EA-6LFN672I	1.2 V	–6	LOW	Lead-Free fpBGA	672	IND	67
LFE3-70EA-7LFN672I	1.2 V	–7	LOW	Lead-Free fpBGA	672	IND	67
LFE3-70EA-8LFN672I	1.2 V	–8	LOW	Lead-Free fpBGA	672	IND	67
LFE3-70EA-6FN1156I	1.2 V	–6	STD	Lead-Free fpBGA	1156	IND	67
LFE3-70EA-7FN1156I	1.2 V	–7	STD	Lead-Free fpBGA	1156	IND	67
LFE3-70EA-8FN1156I	1.2 V	–8	STD	Lead-Free fpBGA	1156	IND	67
LFE3-70EA-6LFN1156I	1.2 V	–6	LOW	Lead-Free fpBGA	1156	IND	67
LFE3-70EA-7LFN1156I	1.2 V	–7	LOW	Lead-Free fpBGA	1156	IND	67
LFE3-70EA-8LFN1156I	1.2 V	–8	LOW	Lead-Free fpBGA	1156	IND	67

1. For ordering information on -9 speed grade devices, please contact your Lattice Sales Representative.

Part Number	Voltage	Grade[1]	Power	Package	Pins	Temp.	LUTs (K)
LFE3-95EA-6FN484I	1.2 V	–6	STD	Lead-Free fpBGA	484	IND	92
LFE3-95EA-7FN484I	1.2 V	–7	STD	Lead-Free fpBGA	484	IND	92
LFE3-95EA-8FN484I	1.2 V	–8	STD	Lead-Free fpBGA	484	IND	92
LFE3-95EA-6LFN484I	1.2 V	–6	LOW	Lead-Free fpBGA	484	IND	92
LFE3-95EA-7LFN484I	1.2 V	–7	LOW	Lead-Free fpBGA	484	IND	92
LFE3-95EA-8LFN484I	1.2 V	–8	LOW	Lead-Free fpBGA	484	IND	92
LFE3-95EA-6FN672I	1.2 V	–6	STD	Lead-Free fpBGA	672	IND	92
LFE3-95EA-7FN672I	1.2 V	–7	STD	Lead-Free fpBGA	672	IND	92
LFE3-95EA-8FN672I	1.2 V	–8	STD	Lead-Free fpBGA	672	IND	92
LFE3-95EA-6LFN672I	1.2 V	–6	LOW	Lead-Free fpBGA	672	IND	92
LFE3-95EA-7LFN672I	1.2 V	–7	LOW	Lead-Free fpBGA	672	IND	92
LFE3-95EA-8LFN672I	1.2 V	–8	LOW	Lead-Free fpBGA	672	IND	92
LFE3-95EA-6FN1156I	1.2 V	–6	STD	Lead-Free fpBGA	1156	IND	92
LFE3-95EA-7FN1156I	1.2 V	–7	STD	Lead-Free fpBGA	1156	IND	92
LFE3-95EA-8FN1156I	1.2 V	–8	STD	Lead-Free fpBGA	1156	IND	92
LFE3-95EA-6LFN1156I	1.2 V	–6	LOW	Lead-Free fpBGA	1156	IND	92
LFE3-95EA-7LFN1156I	1.2 V	–7	LOW	Lead-Free fpBGA	1156	IND	92
LFE3-95EA-8LFN1156I	1.2 V	–8	LOW	Lead-Free fpBGA	1156	IND	92

1. For ordering information on -9 speed grade devices, please contact your Lattice Sales Representative.