Lattice Semiconductor Corporation - LFE3-17EA-8FTN256C Datasheet

Welcome to E-XFL.COM

Understanding Embedded - FPGAs (Field Programmable Gate Array)

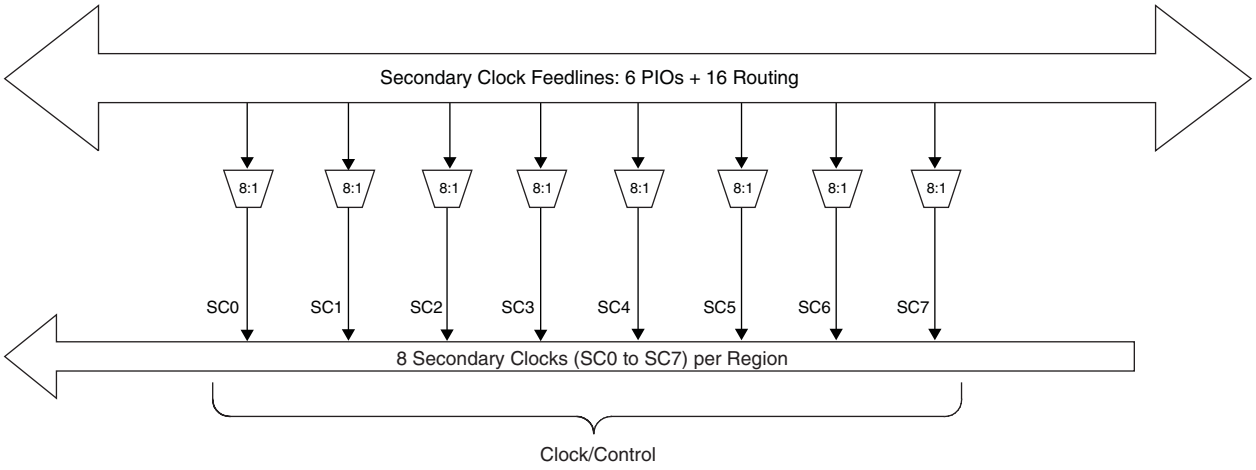
Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Active
Number of LABs/CLBs	2125
Number of Logic Elements/Cells	17000
Total RAM Bits	716800
Number of I/O	133
Number of Gates	-
Voltage - Supply	1.14V ~ 1.26V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	256-BGA
Supplier Device Package	256-FTBGA (17x17)
Purchase URL	https://www.e-xfl.com/product-detail/lattice-semiconductor/lfe3-17ea-8ftn256c

Figure 2-16. Per Region Secondary Clock Selection



Slice Clock Selection

Figure 2-17 shows the clock selections and Figure 2-18 shows the control selections for Slice0 through Slice2. All the primary clocks and seven secondary clocks are routed to this clock selection mux. Other signals can be used as a clock input to the slices via routing. Slice controls are generated from the secondary clocks/controls or other signals connected via routing.

If none of the signals are selected for both clock and control then the default value of the mux output is 1. Slice 3 does not have any registers; therefore it does not have the clock or control muxes.

Figure 2-17. Slice0 through Slice2 Clock Selection

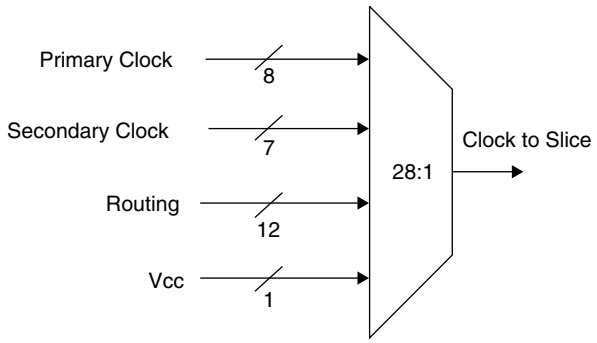
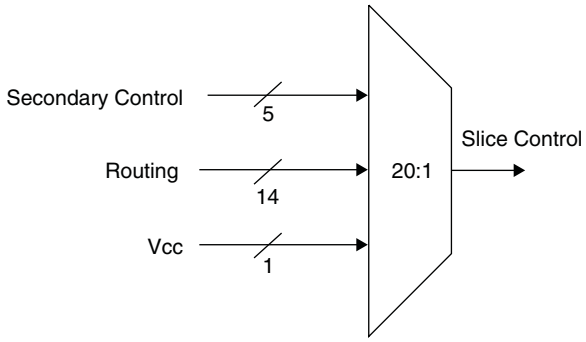
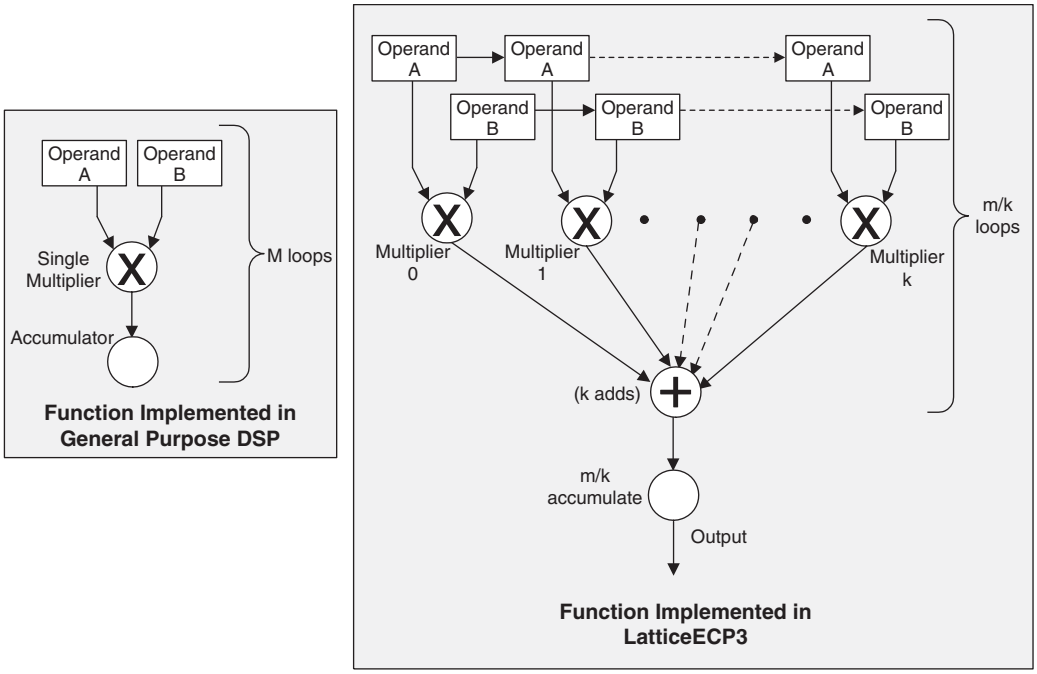


Figure 2-18. Slice0 through Slice2 Control Selection



This allows designers to use highly parallel implementations of DSP functions. Designers can optimize DSP performance vs. area by choosing appropriate levels of parallelism. Figure 2-23 compares the fully serial implementation to the mixed parallel and serial implementation.

Figure 2-23. Comparison of General DSP and LatticeECP3 Approaches



LatticeECP3 sysDSP Slice Architecture Features

The LatticeECP3 sysDSP Slice has been significantly enhanced to provide functions needed for advanced processing applications. These enhancements provide improved flexibility and resource utilization.

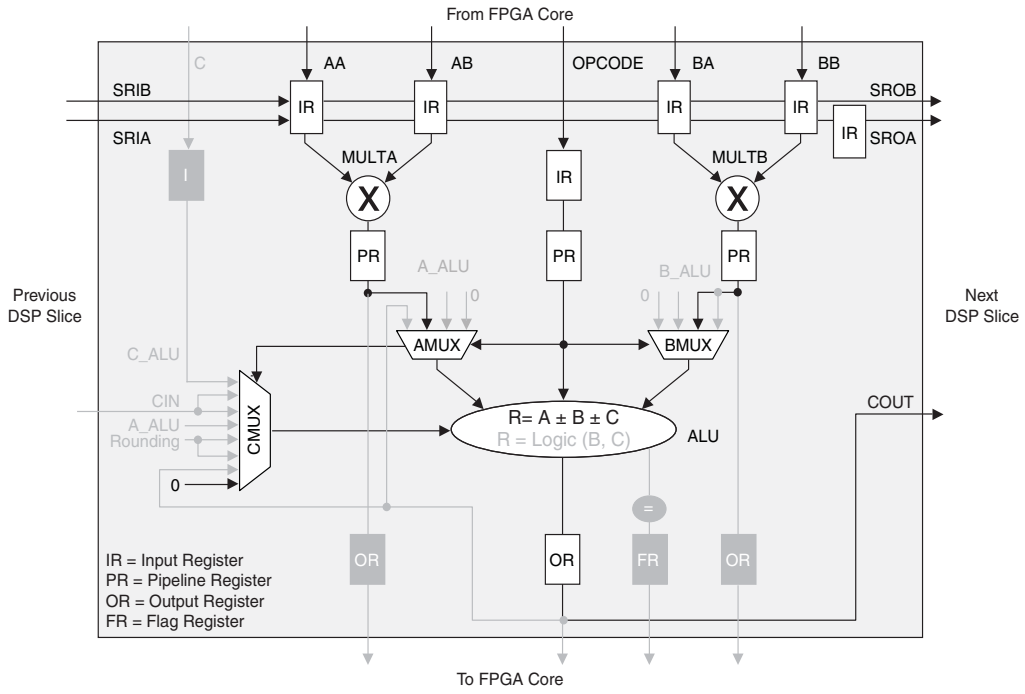
The LatticeECP3 sysDSP Slice supports many functions that include the following:

- Multiply (one 18 x 36, two 18 x 18 or four 9 x 9 Multiplies per Slice)
- Multiply (36 x 36 by cascading across two sysDSP slices)
- Multiply Accumulate (up to 18 x 36 Multipliers feeding an Accumulator that can have up to 54-bit resolution)
- Two Multiplies feeding one Accumulate per cycle for increased processing with lower latency (two 18 x 18 Multiplies feed into an accumulator that can accumulate up to 52 bits)
- Flexible saturation and rounding options to satisfy a diverse set of applications situations
- Flexible cascading across DSP slices
  - Minimizes fabric use for common DSP and ALU functions
  - Enables implementation of FIR Filter or similar structures using dedicated sysDSP slice resources only
  - Provides matching pipeline registers
  - Can be configured to continue cascading from one row of sysDSP slices to another for longer cascade chains
- Flexible and Powerful Arithmetic Logic Unit (ALU) Supports:
  - Dynamically selectable ALU OPCODE
  - Ternary arithmetic (addition/subtraction of three inputs)
  - Bit-wise two-input logic operations (AND, OR, NAND, NOR, XOR and XNOR)
  - Eight flexible and programmable ALU flags that can be used for multiple pattern detection scenarios, such

MULTADDSUBSUM DSP Element

In this case, the operands AA and AB are multiplied and the result is added/subtracted with the result of the multiplier operation of operands BA and BB of Slice 0. Additionally, the operands AA and AB are multiplied and the result is added/subtracted with the result of the multiplier operation of operands BA and BB of Slice 1. The results of both addition/subtractions are added by the second ALU following the slice cascade path. The user can enable the input, output and pipeline registers. Figure 2-30 and Figure 2-31 show the MULTADDSUBSUM sysDSP element.

Figure 2-30. MULTADDSUBSUM Slice 0



Input signals are fed from the sysI/O buffer to the input register block (as signal DI). If desired, the input signal can bypass the register and delay elements and be used directly as a combinatorial signal (INDD), a clock (INCK) and, in selected blocks, the input to the DQS delay block. If an input delay is desired, designers can select either a fixed delay or a dynamic delay DEL[3:0]. The delay, if selected, reduces input register hold time requirements when using a global clock.

The input block allows three modes of operation. In single data rate (SDR) the data is registered with the system clock by one of the registers in the single data rate sync register block.

In DDR mode, two registers are used to sample the data on the positive and negative edges of the modified DQS (ECLKDQSR) in the DDR Memory mode or ECLK signal when using DDR Generic mode, creating two data streams. Before entering the core, these two data streams are synchronized to the system clock to generate two data streams.

A gearbox function can be implemented in each of the input registers on the left and right sides. The gearbox function takes a double data rate signal applied to PIOA and converts it as four data streams, INA, IPA, INB and IPB. The two data streams from the first set of DDR registers are synchronized to the edge clock and then to the system clock before entering the core. Figure 2-30 provides further information on the use of the gearbox function.

The signal DDRCLKPOL controls the polarity of the clock used in the synchronization registers. It ensures adequate timing when data is transferred to the system clock domain from the ECLKDQSR (DDR Memory Interface mode) or ECLK (DDR Generic mode). The DDRLAT signal is used to ensure the data transfer from the synchronization registers to the clock transfer and gearbox registers.

The ECLKDQSR, DDRCLKPOL and DDRLAT signals are generated in the DQS Read Control Logic Block. See Figure 2-37 for an overview of the DQS read control logic.

Further discussion about using the DQS strobe in this module is discussed in the DDR Memory section of this data sheet.

Please see TN1180, LatticeECP3 High-Speed I/O Interface for more information on this topic.

DLL Calibrated DQS Delay Block

Source synchronous interfaces generally require the input clock to be adjusted in order to correctly capture data at the input register. For most interfaces, a PLL is used for this adjustment. However, in DDR memories the clock (referred to as DQS) is not free-running so this approach cannot be used. The DQS Delay block provides the required clock alignment for DDR memory interfaces.

The delay required for the DQS signal is generated by two dedicated DLLs (DDR DLL) on opposite side of the device. Each DLL creates DQS delays in its half of the device as shown in Figure 2-36. The DDR DLL on the left side will generate delays for all the DQS Strobe pins on Banks 0, 7 and 6 and DDR DLL on the right will generate delays for all the DQS pins on Banks 1, 2 and 3. The DDR DLL loop compensates for temperature, voltage and process variations by using the system clock and DLL feedback loop. DDR DLL communicates the required delay to the DQS delay block using a 7-bit calibration bus (DCNTL[6:0])

The DQS signal (selected PIOs only, as shown in Figure 2-35) feeds from the PAD through a DQS control logic block to a dedicated DQS routing resource. The DQS control logic block consists of DQS Read Control logic block that generates control signals for the read side and DQS Write Control logic that generates the control signals required for the write side. A more detailed DQS control diagram is shown in Figure 2-37, which shows how the DQS control blocks interact with the data paths.

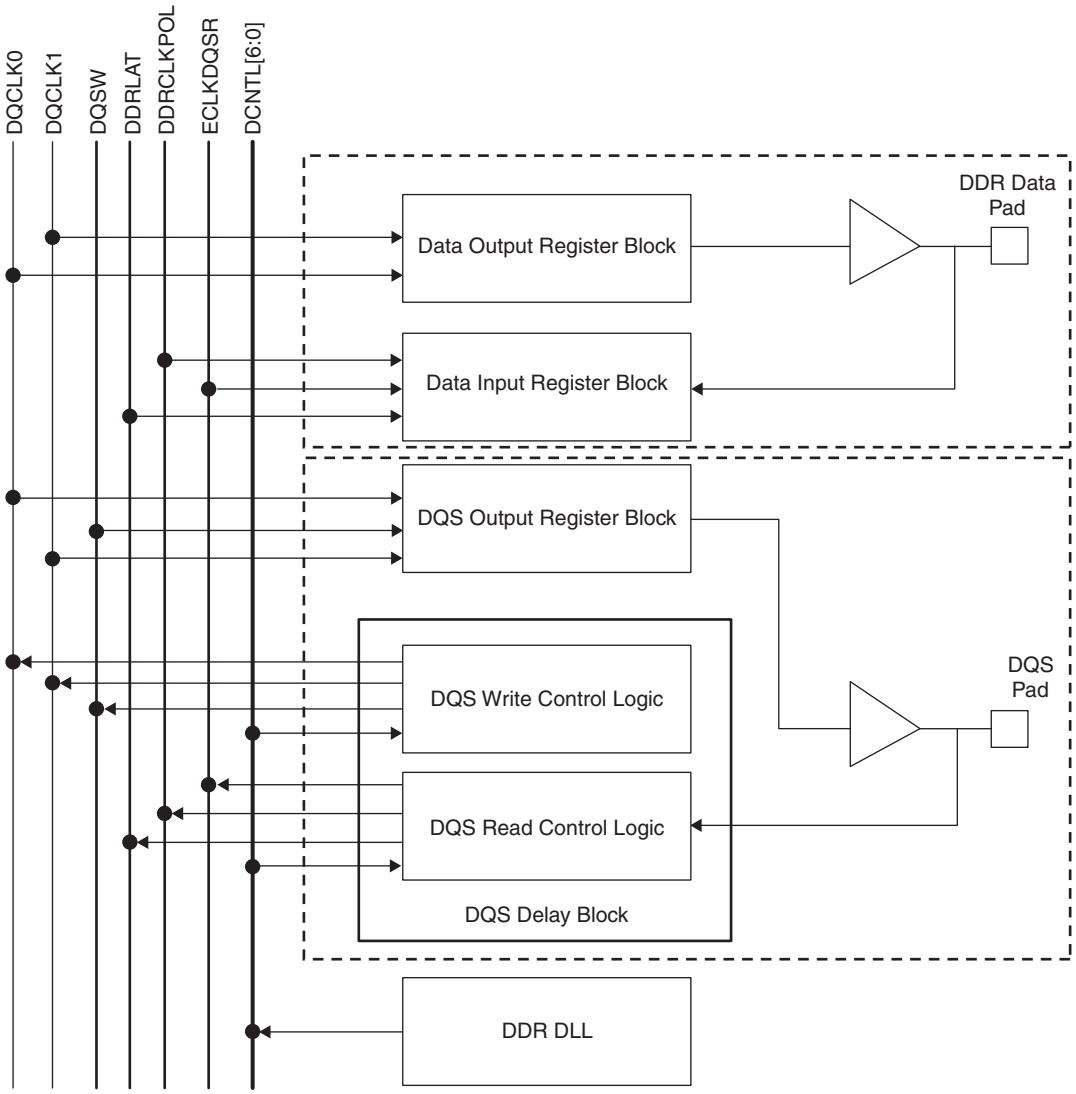
The DQS Read control logic receives the delay generated by the DDR DLL on its side and delays the incoming DQS signal by 90 degrees. This delayed ECLKDQSR is routed to 10 or 11 DQ pads covered by that DQS signal. This block also contains a polarity control logic that generates a DDRCLKPOL signal, which controls the polarity of the clock to the sync registers in the input register blocks. The DQS Read control logic also generates a DDRLAT signal that is in the input register block to transfer data from the first set of DDR register to the second set of DDR registers when using the DDRX2 gearbox mode for DDR3 memory interface.

The DQS Write control logic block generates the DQCLK0 and DQCLK1 clocks used to control the output gearing in the Output register block which generates the DDR data output and the DQS output. They are also used to control the generation of the DQS output through the DQS output register block. In addition to the DCNTL [6:0] input from the DDR DLL, the DQS Write control block also uses a Dynamic Delay DYN DEL [7:0] attribute which is used to further delay the DQS to accomplish the write leveling found in DDR3 memory. Write leveling is controlled by the DDR memory controller implementation. The DYN DELAY can set 128 possible delay step settings. In addition, the most significant bit will invert the clock for a 180-degree shift of the incoming clock. This will generate the DQSW signal used to generate the DQS output in the DQS output register block.

Figure 2-36 and Figure 2-37 show how the DQS transition signals that are routed to the PIOs.

Please see TN1180, LatticeECP3 High-Speed I/O Interface for more information on this topic.

Figure 2-37. DQS Local Bus



Polarity Control Logic

In a typical DDR Memory interface design, the phase relationship between the incoming delayed DQS strobe and the internal system clock (during the READ cycle) is unknown. The LatticeECP3 family contains dedicated circuits to transfer data between these domains. A clock polarity selector is used to prevent set-up and hold violations at the domain transfer between DQS (delayed) and the system clock. This changes the edge on which the data is registered in the synchronizing registers in the input register block. This requires evaluation at the start of each READ cycle for the correct clock polarity.

Prior to the READ operation in DDR memories, DQS is in tristate (pulled by termination). The DDR memory device drives DQS low at the start of the preamble state. A dedicated circuit detects the first DQS rising edge after the preamble state. This signal is used to control the polarity of the clock to the synchronizing registers.

DDR3 Memory Support

LatticeECP3 supports the read and write leveling required for DDR3 memory interfaces.

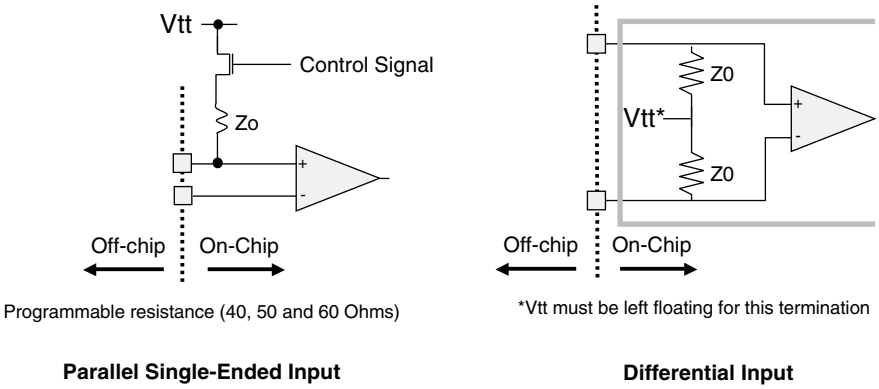
Read leveling is supported by the use of the DDRCLKPOL and the DDRLAT signals generated in the DQS Read Control logic block. These signals dynamically control the capture of the data with respect to the DQS at the input register block.

On-Chip Programmable Termination

The LatticeECP3 supports a variety of programmable on-chip terminations options, including:

- Dynamically switchable Single-Ended Termination with programmable resistor values of 40, 50, or 60 Ohms. External termination to Vtt should be used for DDR2 and DDR3 memory controller implementation.
- Common mode termination of 80, 100, 120 Ohms for differential inputs

Figure 2-39. On-Chip Termination



See Table 2-12 for termination options for input modes.

Table 2-12. On-Chip Termination Options for Input Modes

IO_TYPE	TERMINATE to VTT[1, 2]	DIFFERENTIAL TERMINATION RESISTOR[1]
LVDS25	þ	80, 100, 120
BLVDS25	þ	80, 100, 120
MLVDS	þ	80, 100, 120
HSTL18_I	40, 50, 60	þ
HSTL18_II	40, 50, 60	þ
HSTL18D_I	40, 50, 60	þ
HSTL18D_II	40, 50, 60	þ
HSTL15_I	40, 50, 60	þ
HSTL15D_I	40, 50, 60	þ
SSTL25_I	40, 50, 60	þ
SSTL25_II	40, 50, 60	þ
SSTL25D_I	40, 50, 60	þ
SSTL25D_II	40, 50, 60	þ
SSTL18_I	40, 50, 60	þ
SSTL18_II	40, 50, 60	þ
SSTL18D_I	40, 50, 60	þ
SSTL18D_II	40, 50, 60	þ
SSTL15	40, 50, 60	þ
SSTL15D	40, 50, 60	þ

1. TERMINATE to VTT and DIFFRENTIAL TERMINATION RESISTOR when turned on can only have one setting per bank. Only left and right banks have this feature.
   Use of TERMINATE to VTT and DIFFRENTIAL TERMINATION RESISTOR are mutually exclusive in an I/O bank.
   On-chip termination tolerance +/– 20%
2. External termination to VTT should be used when implementing DDR2 and DDR3 memory controller.

SCI (SERDES Client Interface) Bus

The SERDES Client Interface (SCI) is an IP interface that allows the SERDES/PCS Quad block to be controlled by registers rather than the configuration memory cells. It is a simple register configuration interface that allows SERDES/PCS configuration without power cycling the device.

The Diamond and ispLEVER design tools support all modes of the PCS. Most modes are dedicated to applications associated with a specific industry standard data protocol. Other more general purpose modes allow users to define their own operation. With these tools, the user can define the mode for each quad in a design.

Popular standards such as 10Gb Ethernet, x4 PCI Express and 4x Serial RapidIO can be implemented using IP (available through Lattice), a single quad (Four SERDES channels and PCS) and some additional logic from the core.

The LatticeECP3 family also supports a wide range of primary and secondary protocols. Within the same quad, the LatticeECP3 family can support mixed protocols with semi-independent clocking as long as the required clock frequencies are integer x1, x2, or x11 multiples of each other. Table 2-15 lists the allowable combination of primary and secondary protocol combinations.

Flexible Quad SERDES Architecture

The LatticeECP3 family SERDES architecture is a quad-based architecture. For most SERDES settings and standards, the whole quad (consisting of four SERDES) is treated as a unit. This helps in silicon area savings, better utilization and overall lower cost.

However, for some specific standards, the LatticeECP3 quad architecture provides flexibility; more than one standard can be supported within the same quad.

Table 2-15 shows the standards can be mixed and matched within the same quad. In general, the SERDES standards whose nominal data rates are either the same or a defined subset of each other, can be supported within the same quad. In Table 2-15, the Primary Protocol column refers to the standard that determines the reference clock and PLL settings. The Secondary Protocol column shows the other standard that can be supported within the same quad.

Furthermore, Table 2-15 also implies that more than two standards in the same quad can be supported, as long as they conform to the data rate and reference clock requirements. For example, a quad may contain PCI Express 1.1, SGMII, Serial RapidIO Type I and Serial RapidIO Type II, all in the same quad.

Table 2-15. LatticeECP3 Primary and Secondary Protocol Support

Primary Protocol	Secondary Protocol
PCI Express 1.1	SGMII
PCI Express 1.1	Gigabit Ethernet
PCI Express 1.1	Serial RapidIO Type I
PCI Express 1.1	Serial RapidIO Type II
Serial RapidIO Type I	SGMII
Serial RapidIO Type I	Gigabit Ethernet
Serial RapidIO Type II	SGMII
Serial RapidIO Type II	Gigabit Ethernet
Serial RapidIO Type II	Serial RapidIO Type I
CPRI-3	CPRI-2 and CPRI-1
3G-SDI	HD-SDI and SD-SDI

LVPECL33

The LatticeECP3 devices support the differential LVPECL standard. This standard is emulated using complementary LVCMOS outputs in conjunction with a parallel resistor across the driver outputs. The LVPECL input standard is supported by the LVDS differential input buffer. The scheme shown in Figure 3-3 is one possible solution for point-to-point signals.

Figure 3-3. Differential LVPECL33

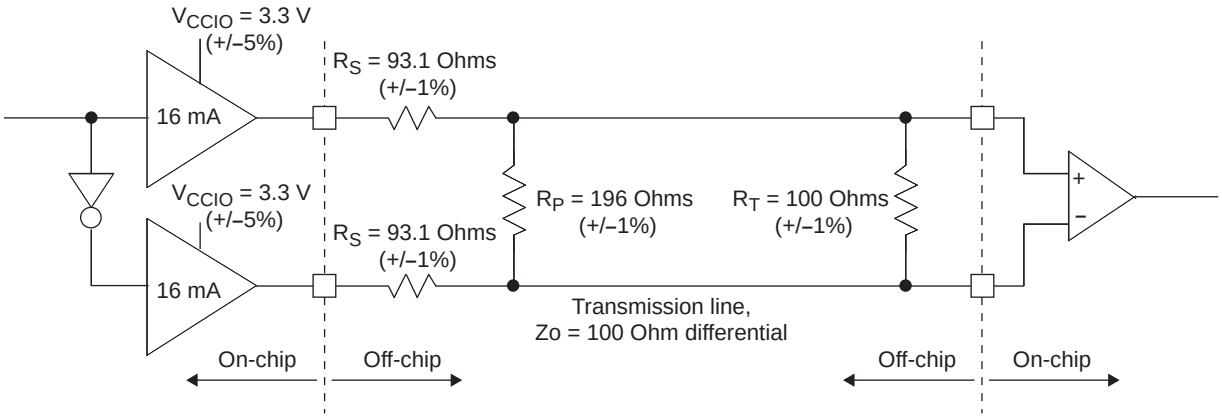


Table 3-3. LVPECL33 DC Conditions[1]

Over Recommended Operating Conditions

Parameter	Description	Typical	Units
V_{CCIO}	Output Driver Supply (+/–5%)	3.30	V
Z_{OUT}	Driver Impedance	10	Ω
R_S	Driver Series Resistor (+/–1%)	93	Ω
R_P	Driver Parallel Resistor (+/–1%)	196	Ω
R_T	Receiver Termination (+/–1%)	100	Ω
V_{OH}	Output High Voltage	2.05	V
V_{OL}	Output Low Voltage	1.25	V
V_{OD}	Output Differential Voltage	0.80	V
V_{CM}	Output Common Mode Voltage	1.65	V
Z_{BACK}	Back Impedance	100.5	Ω
I_{DC}	DC Output Current	12.11	mA

1. For input buffer, see LVDS table.

Typical Building Block Function Performance

Pin-to-Pin Performance (LVCMOS25 12 mA Drive)[1, 2, 3]

Function	–8 Timing	Units
Basic Functions		
16-bit Decoder	4.7	ns
32-bit Decoder	4.7	ns
64-bit Decoder	5.7	ns
4:1 MUX	4.1	ns
8:1 MUX	4.3	ns
16:1 MUX	4.7	ns
32:1 MUX	4.8	ns

1. These functions were generated using the ispLEVER design tool. Exact performance may vary with device and tool version. The tool uses internal parameters that have been characterized but are not tested on every device.
2. Commercial timing numbers are shown. Industrial numbers are typically slower and can be extracted from the Diamond or ispLEVER software.

Register-to-Register Performance[1, 2, 3]

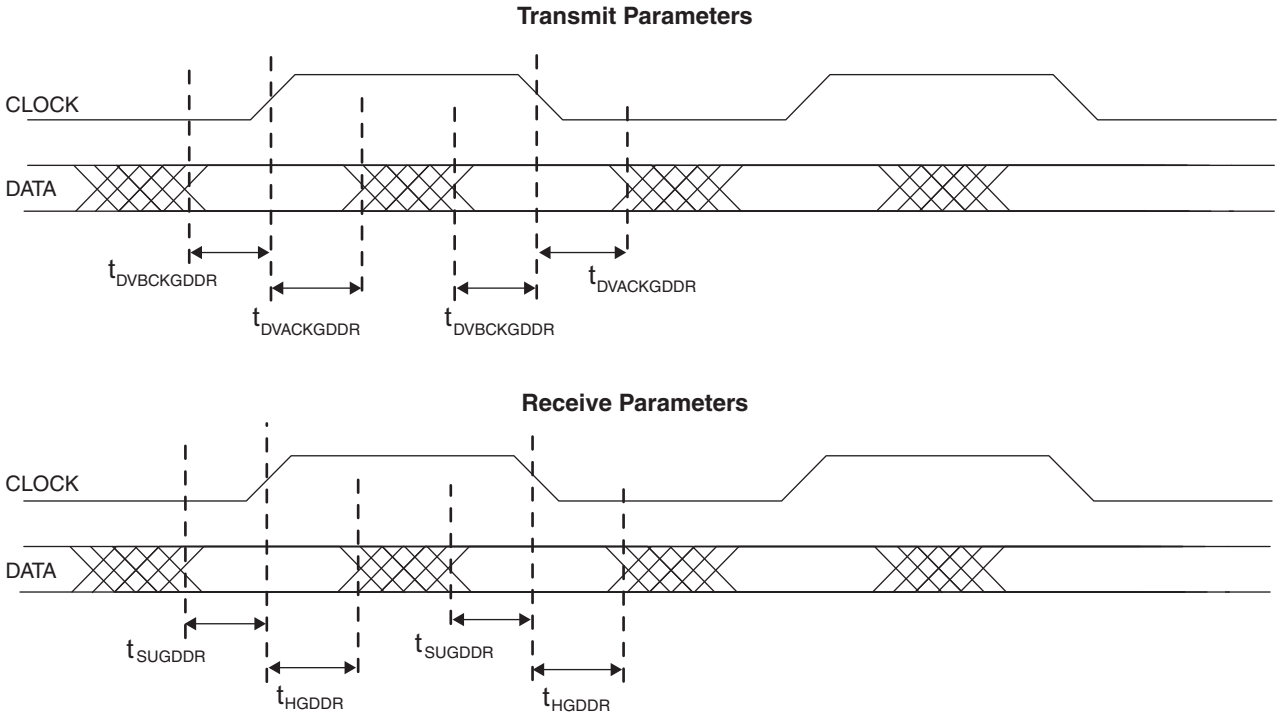
Function	–8 Timing	Units
Basic Functions		
16-bit Decoder	500	MHz
32-bit Decoder	500	MHz
64-bit Decoder	500	MHz
4:1 MUX	500	MHz
8:1 MUX	500	MHz
16:1 MUX	500	MHz
32:1 MUX	445	MHz
8-bit adder	500	MHz
16-bit adder	500	MHz
64-bit adder	305	MHz
16-bit counter	500	MHz
32-bit counter	460	MHz
64-bit counter	320	MHz
64-bit accumulator	315	MHz
Embedded Memory Functions		
512x36 Single Port RAM, EBR Output Registers	340	MHz
1024x18 True-Dual Port RAM (Write Through or Normal, EBR Output Registers)	340	MHz
1024x18 True-Dual Port RAM (Read-Before-Write, EBR Output Registers	130	MHz
1024x18 True-Dual Port RAM (Write Through or Normal, PLC Output Registers)	245	MHz
Distributed Memory Functions		
16x4 Pseudo-Dual Port RAM (One PFU)	500	MHz
32x4 Pseudo-Dual Port RAM	500	MHz
64x8 Pseudo-Dual Port RAM	400	MHz
DSP Function		
18x18 Multiplier (All Registers)	400	MHz
9x9 Multiplier (All Registers)	400	MHz
36x36 Multiply (All Registers)	260	MHz

LatticeECP3 External Switching Characteristics (Continued)[1, 2, 3, 13]

Over Recommended Commercial Operating Conditions

Parameter	Description	Device	–8		–7		–6		Units
			Min.	Max.	Min.	Max.	Min.	Max.	
t_{H_DEL}	Clock to Data Hold - PIO Input Register with Input Data Delay	ECP3-150EA	0.0	—	0.0	—	0.0	—	ns
f_{MAX_IO}	Clock Frequency of I/O and PFU Register	ECP3-150EA	—	500	—	420	—	375	MHz
t_{CO}	Clock to Output - PIO Output Register	ECP3-70EA/95EA	—	3.8	—	4.2	—	4.6	ns
t_{SU}	Clock to Data Setup - PIO Input Register	ECP3-70EA/95EA	0.0	—	0.0	—	0.0	—	ns
t_{H}	Clock to Data Hold - PIO Input Register	ECP3-70EA/95EA	1.4	—	1.6	—	1.8	—	ns
t_{SU_DEL}	Clock to Data Setup - PIO Input Register with Data Input Delay	ECP3-70EA/95EA	1.3	—	1.5	—	1.7	—	ns
t_{H_DEL}	Clock to Data Hold - PIO Input Register with Input Data Delay	ECP3-70EA/95EA	0.0	—	0.0	—	0.0	—	ns
f_{MAX_IO}	Clock Frequency of I/O and PFU Register	ECP3-70EA/95EA	—	500	—	420	—	375	MHz
t_{CO}	Clock to Output - PIO Output Register	ECP3-35EA	—	3.7	—	4.1	—	4.5	ns
t_{SU}	Clock to Data Setup - PIO Input Register	ECP3-35EA	0.0	—	0.0	—	0.0	—	ns
t_{H}	Clock to Data Hold - PIO Input Register	ECP3-35EA	1.2	—	1.4	—	1.6	—	ns
t_{SU_DEL}	Clock to Data Setup - PIO Input Register with Data Input Delay	ECP3-35EA	1.3	—	1.4	—	1.5	—	ns
t_{H_DEL}	Clock to Data Hold - PIO Input Register with Input Data Delay	ECP3-35EA	0.0	—	0.0	—	0.0	—	ns
f_{MAX_IO}	Clock Frequency of I/O and PFU Register	ECP3-35EA	—	500	—	420	—	375	MHz
t_{CO}	Clock to Output - PIO Output Register	ECP3-17EA	—	3.5	—	3.9	—	4.3	ns
t_{SU}	Clock to Data Setup - PIO Input Register	ECP3-17EA	0.0	—	0.0	—	0.0	—	ns
t_{H}	Clock to Data Hold - PIO Input Register	ECP3-17EA	1.3	—	1.5	—	1.6	—	ns
t_{SU_DEL}	Clock to Data Setup - PIO Input Register with Data Input Delay	ECP3-17EA	1.3	—	1.4	—	1.5	—	ns
t_{H_DEL}	Clock to Data Hold - PIO Input Register with Input Data Delay	ECP3-17EA	0.0	—	0.0	—	0.0	—	ns
f_{MAX_IO}	Clock Frequency of I/O and PFU Register	ECP3-17EA	—	500	—	420	—	375	MHz
General I/O Pin Parameters Using Dedicated Clock Input Primary Clock with PLL with Clock Injection Removal Setting[2]									
t_{COPLL}	Clock to Output - PIO Output Register	ECP3-150EA	—	3.3	—	3.6	—	39	ns
t_{SUPLL}	Clock to Data Setup - PIO Input Register	ECP3-150EA	0.7	—	0.8	—	0.9	—	ns
t_{HPLL}	Clock to Data Hold - PIO Input Register	ECP3-150EA	0.8	—	0.9	—	1.0	—	ns
t_{SU_DELPLL}	Clock to Data Setup - PIO Input Register with Data Input Delay	ECP3-150EA	1.6	—	1.8	—	2.0	—	ns
t_{H_DELPLL}	Clock to Data Hold - PIO Input Register with Input Data Delay	ECP3-150EA	—	0.0	—	0.0	—	0.0	ns
t_{COPLL}	Clock to Output - PIO Output Register	ECP3-70EA/95EA	—	3.3	—	3.5	—	3.8	ns
t_{SUPLL}	Clock to Data Setup - PIO Input Register	ECP3-70EA/95EA	0.7	—	0.8	—	0.9	—	ns

Figure 3-8. Generic DDRX1/DDRX2 (With Clock Center on Data Window)



LatticeECP3 Maximum I/O Buffer Speed [1, 2, 3, 4, 5, 6]

Over Recommended Operating Conditions

Buffer	Description	Max.	Units
Maximum Input Frequency			
LVDS25	LVDS, V_{CCIO} = 2.5 V	400	MHz
MLVDS25	MLVDS, Emulated, V_{CCIO} = 2.5 V	400	MHz
BLVDS25	BLVDS, Emulated, V_{CCIO} = 2.5 V	400	MHz
PPLVDS	Point-to-Point LVDS	400	MHz
TRLVDS	Transition-Reduced LVDS	612	MHz
Mini LVDS	Mini LVDS	400	MHz
LVPECL33	LVPECL, Emulated, V_{CCIO} = 3.3 V	400	MHz
HSTL18 (all supported classes)	HSTL_18 class I, II, V_{CCIO} = 1.8 V	400	MHz
HSTL15	HSTL_15 class I, V_{CCIO} = 1.5 V	400	MHz
SSTL33 (all supported classes)	SSTL_3 class I, II, V_{CCIO} = 3.3 V	400	MHz
SSTL25 (all supported classes)	SSTL_2 class I, II, V_{CCIO} = 2.5 V	400	MHz
SSTL18 (all supported classes)	SSTL_18 class I, II, V_{CCIO} = 1.8 V	400	MHz
LVTTL33	LVTTL, V_{CCIO} = 3.3 V	166	MHz
LVCMOS33	LVCMOS, V_{CCIO} = 3.3 V	166	MHz
LVCMOS25	LVCMOS, V_{CCIO} = 2.5 V	166	MHz
LVCMOS18	LVCMOS, V_{CCIO} = 1.8 V	166	MHz
LVCMOS15	LVCMOS 1.5, V_{CCIO} = 1.5 V	166	MHz
LVCMOS12	LVCMOS 1.2, V_{CCIO} = 1.2 V	166	MHz
PCI33	PCI, V_{CCIO} = 3.3 V	66	MHz
Maximum Output Frequency			
LVDS25E	LVDS, Emulated, V_{CCIO} = 2.5 V	300	MHz
LVDS25	LVDS, V_{CCIO} = 2.5 V	612	MHz
MLVDS25	MLVDS, Emulated, V_{CCIO} = 2.5 V	300	MHz
RSDS25	RSDS, Emulated, V_{CCIO} = 2.5 V	612	MHz
BLVDS25	BLVDS, Emulated, V_{CCIO} = 2.5 V	300	MHz
PPLVDS	Point-to-point LVDS	612	MHz
LVPECL33	LVPECL, Emulated, V_{CCIO} = 3.3 V	612	MHz
Mini-LVDS	Mini LVDS	612	MHz
HSTL18 (all supported classes)	HSTL_18 class I, II, V_{CCIO} = 1.8 V	200	MHz
HSTL15 (all supported classes)	HSTL_15 class I, V_{CCIO} = 1.5 V	200	MHz
SSTL33 (all supported classes)	SSTL_3 class I, II, V_{CCIO} = 3.3 V	233	MHz
SSTL25 (all supported classes)	SSTL_2 class I, II, V_{CCIO} = 2.5 V	233	MHz
SSTL18 (all supported classes)	SSTL_18 class I, II, V_{CCIO} = 1.8 V	266	MHz
LVTTL33	LVTTL, V_{CCIO} = 3.3 V	166	MHz
LVCMOS33 (For all drives)	LVCMOS, 3.3 V	166	MHz
LVCMOS25 (For all drives)	LVCMOS, 2.5 V	166	MHz
LVCMOS18 (For all drives)	LVCMOS, 1.8 V	166	MHz
LVCMOS15 (For all drives)	LVCMOS, 1.5 V	166	MHz
LVCMOS12 (For all drives except 2 mA)	LVCMOS, V_{CCIO} = 1.2 V	166	MHz
LVCMOS12 (2 mA drive)	LVCMOS, V_{CCIO} = 1.2 V	100	MHz

LatticeECP3 Maximum I/O Buffer Speed (Continued)[1, 2, 3, 4, 5, 6]

Over Recommended Operating Conditions

Buffer	Description	Max.	Units
PCI33	PCI, V_{CCIO} = 3.3 V	66	MHz

1. These maximum speeds are characterized but not tested on every device.
2. Maximum I/O speed for differential output standards emulated with resistors depends on the layout.
3. LVCMOS timing is measured with the load specified in the Switching Test Conditions table of this document.
4. All speeds are measured at fast slew.
5. Actual system operation may vary depending on user logic implementation.
6. Maximum data rate equals 2 times the clock rate when utilizing DDR.

sysCLOCK PLL Timing

Over Recommended Operating Conditions

Parameter	Descriptions	Conditions	Clock	Min.	Typ.	Max.	Units
f_{IN}	Input clock frequency (CLKI, CLKFB)		Edge clock	2	—	500	MHz
			Primary clock[4]	2	—	420	MHz
f_{OUT}	Output clock frequency (CLKOP, CLKOS)		Edge clock	4	—	500	MHz
			Primary clock[4]	4	—	420	MHz
f_{OUT1}	K-Divider output frequency	CLKOK		0.03125	—	250	MHz
f_{OUT2}	K2-Divider output frequency	CLKOK2		0.667	—	166	MHz
f_{VCO}	PLL VCO frequency			500	—	1000	MHz
f_{PFD}[3]	Phase detector input frequency		Edge clock	2	—	500	MHz
			Primary clock[4]	2	—	420	MHz
AC Characteristics							
t_{PA}	Programmable delay unit			65	130	260	ps
t_{DT}	Output clock duty cycle (CLKOS, at 50% setting)		Edge clock	45	50	55	%
		$f_{OUT} \leq 250$ MHz	Primary clock	45	50	55	%
		$f_{OUT} > 250$ MHz	Primary clock	30	50	70	%
t_{CPA}	Coarse phase shift error (CLKOS, at all settings)			-5	0	+5	% of period
t_{OPW}	Output clock pulse width high or low (CLKOS)			1.8	—	—	ns
t_{OPJIT}[1]	Output clock period jitter	$f_{OUT} \geq 420$ MHz		—	—	200	ps
		420 MHz > $f_{OUT} \geq 100$ MHz		—	—	250	ps
		$f_{OUT} < 100$ MHz		—	—	0.025	UIPP
t_{SK}	Input clock to output clock skew when N/M = integer			—	—	500	ps
t_{LOCK}[2]	Lock time	2 to 25 MHz		—	—	200	us
		25 to 500 MHz		—	—	50	us
t_{UNLOCK}	Reset to PLL unlock time to ensure fast reset			—	—	50	ns
t_{HI}	Input clock high time	90% to 90%		0.5	—	—	ns
t_{LO}	Input clock low time	10% to 10%		0.5	—	—	ns
t_{IPJIT}	Input clock period jitter			—	—	400	ps
t_{RST}	Reset signal pulse width high, RSTK			10	—	—	ns
	Reset signal pulse width high, RST			500	—	—	ns

1. Jitter sample is taken over 10,000 samples of the primary PLL output with clean reference clock with no additional I/O toggling.
2. Output clock is valid after t_{LOCK} for PLL reset and dynamic delay adjustment.
3. Period jitter and cycle-to-cycle jitter numbers are guaranteed for $f_{PFD} > 4$ MHz. For $f_{PFD} < 4$ MHz, the jitter numbers may not be met in certain conditions. Please contact the factory for $f_{PFD} < 4$ MHz.
4. When using internal feedback, maximum can be up to 500 MHz.

SERDES High-Speed Data Transmitter[1]

Table 3-6. Serial Output Timing and Levels

Symbol	Description	Frequency	Min.	Typ.	Max.	Units
$V_{TX-DIFF-P-P-1.44}$	Differential swing (1.44 V setting)[1, 2]	0.15 to 3.125 Gbps	1150	1440	1730	mV, p-p
$V_{TX-DIFF-P-P-1.35}$	Differential swing (1.35 V setting)[1, 2]	0.15 to 3.125 Gbps	1080	1350	1620	mV, p-p
$V_{TX-DIFF-P-P-1.26}$	Differential swing (1.26 V setting)[1, 2]	0.15 to 3.125 Gbps	1000	1260	1510	mV, p-p
$V_{TX-DIFF-P-P-1.13}$	Differential swing (1.13 V setting)[1, 2]	0.15 to 3.125 Gbps	840	1130	1420	mV, p-p
$V_{TX-DIFF-P-P-1.04}$	Differential swing (1.04 V setting)[1, 2]	0.15 to 3.125 Gbps	780	1040	1300	mV, p-p
$V_{TX-DIFF-P-P-0.92}$	Differential swing (0.92 V setting)[1, 2]	0.15 to 3.125 Gbps	690	920	1150	mV, p-p
$V_{TX-DIFF-P-P-0.87}$	Differential swing (0.87 V setting)[1, 2]	0.15 to 3.125 Gbps	650	870	1090	mV, p-p
$V_{TX-DIFF-P-P-0.78}$	Differential swing (0.78 V setting)[1, 2]	0.15 to 3.125 Gbps	585	780	975	mV, p-p
$V_{TX-DIFF-P-P-0.64}$	Differential swing (0.64 V setting)[1, 2]	0.15 to 3.125 Gbps	480	640	800	mV, p-p
V_{OCM}	Output common mode voltage	—	V_{CCOB} –0.75	V_{CCOB} –0.60	V_{CCOB} –0.45	V
T_{TX-R}	Rise time (20% to 80%)	—	145	185	265	ps
T_{TX-F}	Fall time (80% to 20%)	—	145	185	265	ps
$Z_{TX-OI-SE}$	Output Impedance 50/75/HiZ Ohms (single ended)	—	–20%	50/75/ Hi Z	+20%	Ohms
R_{LTX-RL}	Return loss (with package)	—	10			dB
$T_{TX-INTRASKEW}$	Lane-to-lane TX skew within a SERDES quad block (intra-quad)	—	—	—	200	ps
$T_{TX-INTERSKEW}$[3]	Lane-to-lane skew between SERDES quad blocks (inter-quad)	—	—	—	1UI +200	ps

1. All measurements are with 50 Ohm impedance.
2. See TN1176, LatticeECP3 SERDES/PCS Usage Guide for actual binary settings and the min-max range.
3. Inter-quad skew is between all SERDES channels on the device and requires the use of a low skew internal reference clock.

SERDES/PCS Block Latency

Table 3-8 describes the latency of each functional block in the transmitter and receiver. Latency is given in parallel clock cycles. Figure 3-12 shows the location of each block.

Table 3-8. SERDES/PCS Latency Breakdown

Item	Description	Min.	Avg.	Max.	Fixed	Bypass	Units
Transmit Data Latency[1]							
T1	FPGA Bridge - Gearing disabled with different clocks	1	3	5	—	1	word clk
	FPGA Bridge - Gearing disabled with same clocks	—	—	—	3	1	word clk
	FPGA Bridge - Gearing enabled	1	3	5	—	—	word clk
T2	8b10b Encoder	—	—	—	2	1	word clk
T3	SERDES Bridge transmit	—	—	—	2	1	word clk
T4	Serializer: 8-bit mode	—	—	—	15 + Δ1	—	UI + ps
	Serializer: 10-bit mode	—	—	—	18 + Δ1	—	UI + ps
T5	Pre-emphasis ON	—	—	—	1 + Δ2	—	UI + ps
	Pre-emphasis OFF	—	—	—	0 + Δ3	—	UI + ps
Receive Data Latency[2]							
R1	Equalization ON	—	—	—	Δ1	—	UI + ps
	Equalization OFF	—	—	—	Δ2	—	UI + ps
R2	Deserializer: 8-bit mode	—	—	—	10 + Δ3	—	UI + ps
	Deserializer: 10-bit mode	—	—	—	12 + Δ3	—	UI + ps
R3	SERDES Bridge receive	—	—	—	2	—	word clk
R4	Word alignment	3.1	—	4	—	—	word clk
R5	8b10b decoder	—	—	—	1	—	word clk
R6	Clock Tolerance Compensation	7	15	23	1	1	word clk
R7	FPGA Bridge - Gearing disabled with different clocks	1	3	5	—	1	word clk
	FPGA Bridge - Gearing disabled with same clocks	—	—	—	3	1	word clk
	FPGA Bridge - Gearing enabled	1	3	5	—	—	word clk

1. Δ1 = –245 ps, Δ2 = +88 ps, Δ3 = +112 ps.
2. Δ1 = +118 ps, Δ2 = +132 ps, Δ3 = +700 ps.

Figure 3-12. Transmitter and Receiver Latency Block Diagram

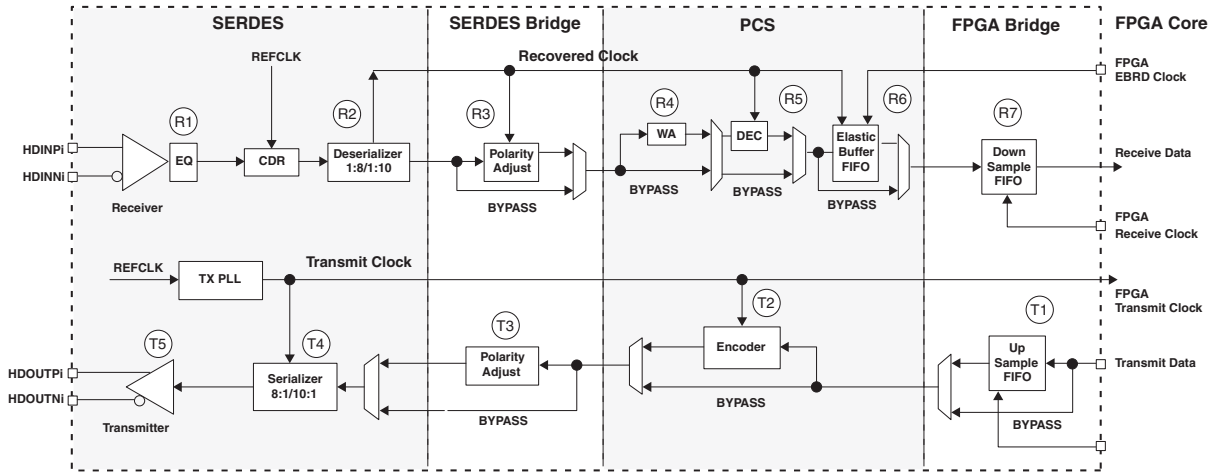
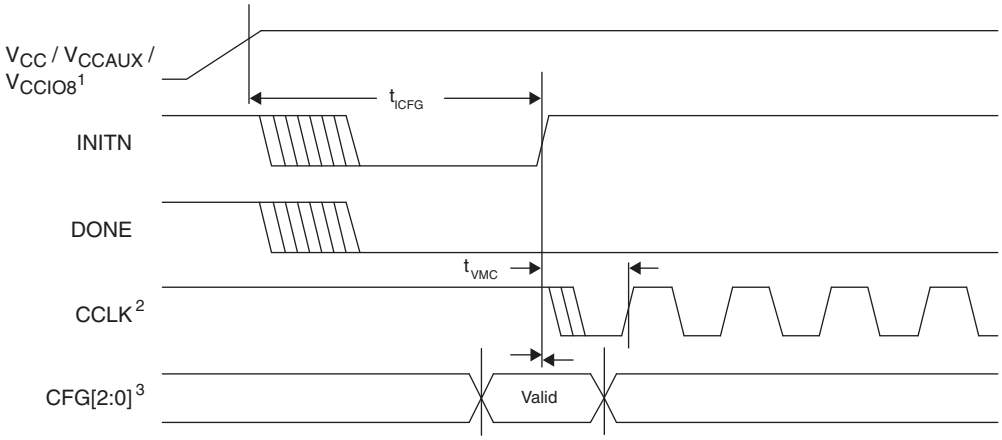
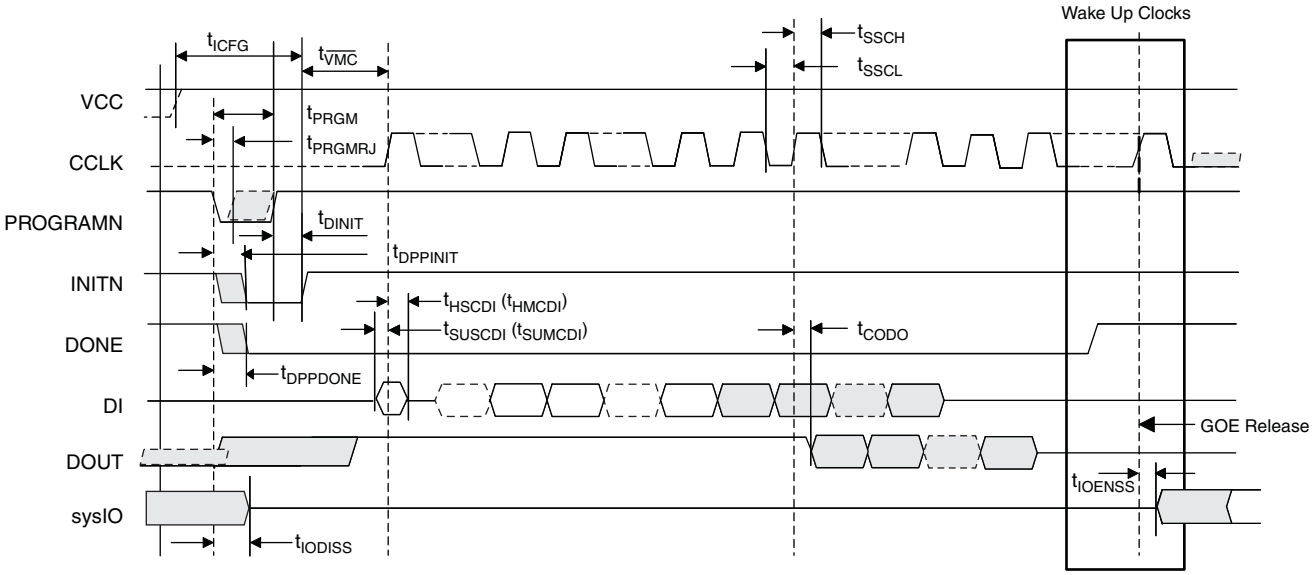


Figure 3-24. Power-On-Reset (POR) Timing



1. Time taken from V_{CC}, V_{CCAUX} or V_{CCIO8}, whichever is the last to cross the POR trip point.
2. Device is in a Master Mode (SPI, SPIm).
3. The CFG pins are normally static (hard wired).

Figure 3-25. sysCONFIG Port Timing



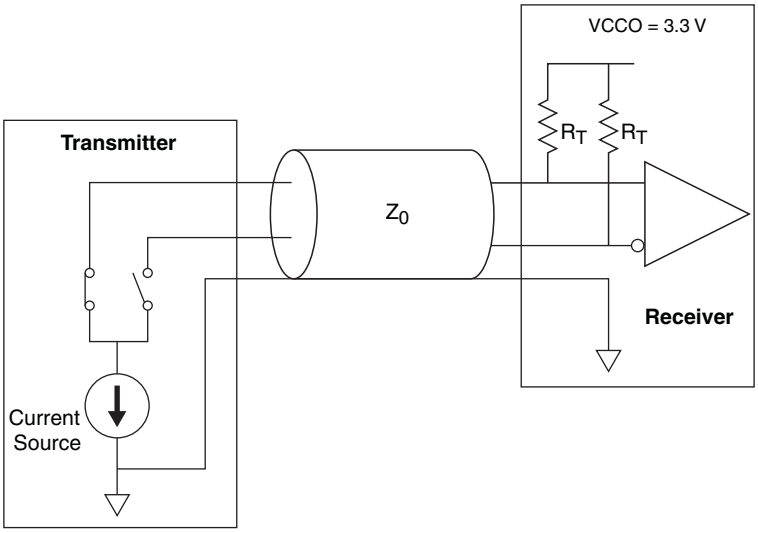
sysI/O Differential Electrical Characteristics

Transition Reduced LVDS (TRLVDS DC Specification)

Over Recommended Operating Conditions

Symbol	Description	Min.	Nom.	Max.	Units
V_{CCO}	Driver supply voltage (+/– 5%)	3.14	3.3	3.47	V
V_{ID}	Input differential voltage	150	—	1200	mV
V_{ICM}	Input common mode voltage	3	—	3.265	V
V_{CCO}	Termination supply voltage	3.14	3.3	3.47	V
R_T	Termination resistance (off-chip)	45	50	55	Ohms

Note: LatticeECP3 only supports the TRLVDS receiver.



Mini LVDS

Over Recommended Operating Conditions

Parameter Symbol	Description	Min.	Typ.	Max.	Units
Z_O	Single-ended PCB trace impedance	30	50	75	Ohms
R_T	Differential termination resistance	50	100	150	Ohms
V_{OD}	Output voltage, differential, $\|V_{OP} - V_{OM}\|$	300	—	600	mV
V_{OS}	Output voltage, common mode, $\|V_{OP} + V_{OM}\|/2$	1	1.2	1.4	V
ΔV_{OD}	Change in V_{OD}, between H and L	—	—	50	mV
ΔV_{ID}	Change in V_{OS}, between H and L	—	—	50	mV
V_{THD}	Input voltage, differential, $\|V_{INP} - V_{INM}\|$	200	—	600	mV
V_{CM}	Input voltage, common mode, $\|V_{INP} + V_{INM}\|/2$	$0.3+(V_{THD}/2)$	—	$2.1-(V_{THD}/2)$	
T_R, T_F	Output rise and fall times, 20% to 80%	—	—	550	ps
T_{ODUTY}	Output clock duty cycle	40	—	60	%

Note: Data is for 6 mA differential current drive. Other differential driver current options are available.

LatticeECP3 Family Data Sheet Supplemental Information

For Further Information

A variety of technical notes for the LatticeECP3 family are available on the Lattice website at www.latticesemi.com.

- TN1169, LatticeECP3 sysCONFIG Usage Guide
- TN1176, LatticeECP3 SERDES/PCS Usage Guide
- TN1177, LatticeECP3 sysIO Usage Guide
- TN1178, LatticeECP3 sysCLOCK PLL/DLL Design and Usage Guide
- TN1179, LatticeECP3 Memory Usage Guide
- TN1180, LatticeECP3 High-Speed I/O Interface
- TN1181, Power Consumption and Management for LatticeECP3 Devices
- TN1182, LatticeECP3 sysDSP Usage Guide
- TN1184, LatticeECP3 Soft Error Detection (SED) Usage Guide
- TN1189, LatticeECP3 Hardware Checklist
- TN1215, LatticeECP2MS and LatticeECP2S Devices
- TN1216, LatticeECP2/M and LatticeECP3 Dual Boot Feature Advanced Security Encryption Key Programming Guide for LatticeECP3
- TN1222, LatticeECP3 Slave SPI Port User's Guide

For further information on interface standards refer to the following websites:

- JEDEC Standards (LVTTL, LVCMOS, SSTL, HSTL): www.jedec.org
- PCI: www.pcisig.com