

Welcome to E-XFL.COM

Understanding **Embedded - FPGAs (Field Programmable Gate Array)**

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

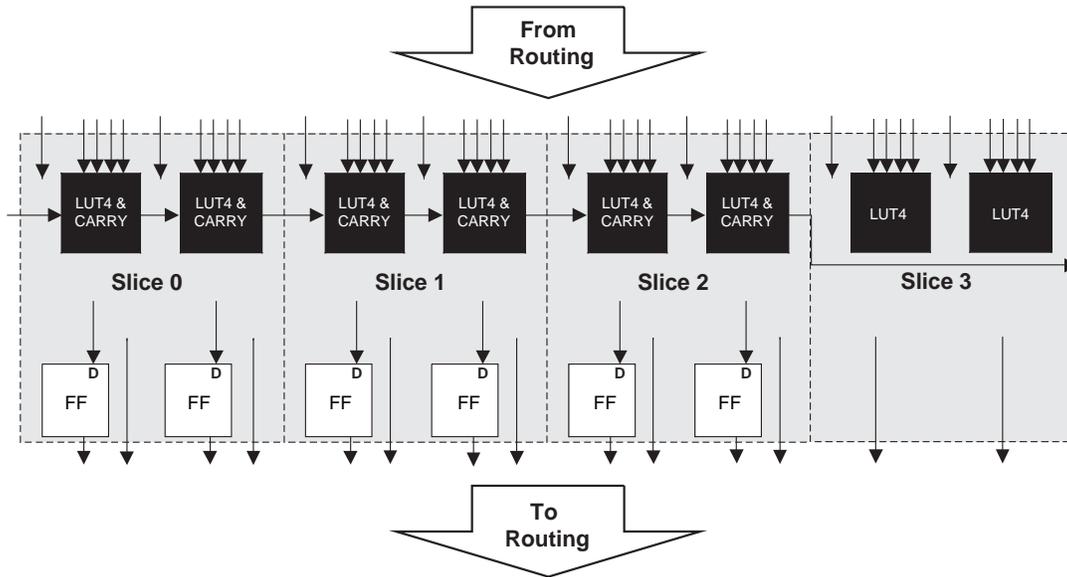
Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details

Product Status	Active
Number of LABs/CLBs	4125
Number of Logic Elements/Cells	33000
Total RAM Bits	1358848
Number of I/O	310
Number of Gates	-
Voltage - Supply	1.14V ~ 1.26V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	672-BBGA
Supplier Device Package	672-FPBGA (27x27)
Purchase URL	https://www.e-xfl.com/product-detail/lattice-semiconductor/lfe3-35ea-8lfn672c

Figure 2-2. PFU Diagram



Slice

Slice 0 through Slice 2 contain two LUT4s feeding two registers, whereas Slice 3 contains two LUT4s only. For PFUs, Slice 0 through Slice 2 can be configured as distributed memory, a capability not available in the PFF. Table 2-1 shows the capability of the slices in both PFF and PFU blocks along with the operation modes they enable. In addition, each PFU contains logic that allows the LUTs to be combined to perform functions such as LUT5, LUT6, LUT7 and LUT8. There is control logic to perform set/reset functions (programmable as synchronous/asynchronous), clock select, chip-select and wider RAM/ROM functions.

Table 2-1. Resources and Modes Available per Slice

Slice	PFU BLock		PFF Block	
	Resources	Modes	Resources	Modes
Slice 0	2 LUT4s and 2 Registers	Logic, Ripple, RAM, ROM	2 LUT4s and 2 Registers	Logic, Ripple, ROM
Slice 1	2 LUT4s and 2 Registers	Logic, Ripple, RAM, ROM	2 LUT4s and 2 Registers	Logic, Ripple, ROM
Slice 2	2 LUT4s and 2 Registers	Logic, Ripple, RAM, ROM	2 LUT4s and 2 Registers	Logic, Ripple, ROM
Slice 3	2 LUT4s	Logic, ROM	2 LUT4s	Logic, ROM

Figure 2-3 shows an overview of the internal logic of the slice. The registers in the slice can be configured for positive/negative and edge triggered or level sensitive clocks.

Slices 0, 1 and 2 have 14 input signals: 13 signals from routing and one from the carry-chain (from the adjacent slice or PFU). There are seven outputs: six to routing and one to carry-chain (to the adjacent PFU). Slice 3 has 10 input signals from routing and four signals to routing. Table 2-2 lists the signals associated with Slice 0 to Slice 2.

ROM Mode

ROM mode uses the LUT logic; hence, Slices 0 through 3 can be used in ROM mode. Preloading is accomplished through the programming interface during PFU configuration.

For more information, please refer to TN1179, [LatticeECP3 Memory Usage Guide](#).

Routing

There are many resources provided in the LatticeECP3 devices to route signals individually or as busses with related control signals. The routing resources consist of switching circuitry, buffers and metal interconnect (routing) segments.

The LatticeECP3 family has an enhanced routing architecture that produces a compact design. The Diamond and ispLEVER design software tool suites take the output of the synthesis tool and places and routes the design.

sysCLOCK PLLs and DLLs

The sysCLOCK PLLs provide the ability to synthesize clock frequencies. The devices in the LatticeECP3 family support two to ten full-featured General Purpose PLLs.

General Purpose PLL

The architecture of the PLL is shown in Figure 2-4. A description of the PLL functionality follows.

CLKI is the reference frequency (generated either from the pin or from routing) for the PLL. CLKI feeds into the Input Clock Divider block. The CLKFB is the feedback signal (generated from CLKOP, CLKOS or from a user clock pin/logic). This signal feeds into the Feedback Divider. The Feedback Divider is used to multiply the reference frequency.

Both the input path and feedback signals enter the Phase Frequency Detect Block (PFD) which detects first for the frequency, and then the phase, of the CLKI and CLKFB are the same which then drives the Voltage Controlled Oscillator (VCO) block. In this block the difference between the input path and feedback signals is used to control the frequency and phase of the oscillator. A LOCK signal is generated by the VCO to indicate that the VCO has locked onto the input clock signal. In dynamic mode, the PLL may lose lock after a dynamic delay adjustment and not relock until the t_{LOCK} parameter has been satisfied.

The output of the VCO then enters the CLKOP divider. The CLKOP divider allows the VCO to operate at higher frequencies than the clock output (CLKOP), thereby increasing the frequency range. The Phase/Duty Cycle/Duty Trim block adjusts the phase and duty cycle of the CLKOS signal. The phase/duty cycle setting can be pre-programmed or dynamically adjusted. A secondary divider takes the CLKOP or CLKOS signal and uses it to derive lower frequency outputs (CLKOK).

The primary output from the CLKOP divider (CLKOP) along with the outputs from the secondary dividers (CLKOK and CLKOK2) and Phase/Duty select (CLKOS) are fed to the clock distribution network.

The PLL allows two methods for adjusting the phase of signal. The first is referred to as Fine Delay Adjustment. This inserts up to 16 nominal 125 ps delays to be applied to the secondary PLL output. The number of steps may be set statically or from the FPGA logic. The second method is referred to as Coarse Phase Adjustment. This allows the phase of the rising and falling edge of the secondary PLL output to be adjusted in 22.5 degree steps. The number of steps may be set statically or from the FPGA logic.

Figure 2-4. General Purpose PLL Diagram

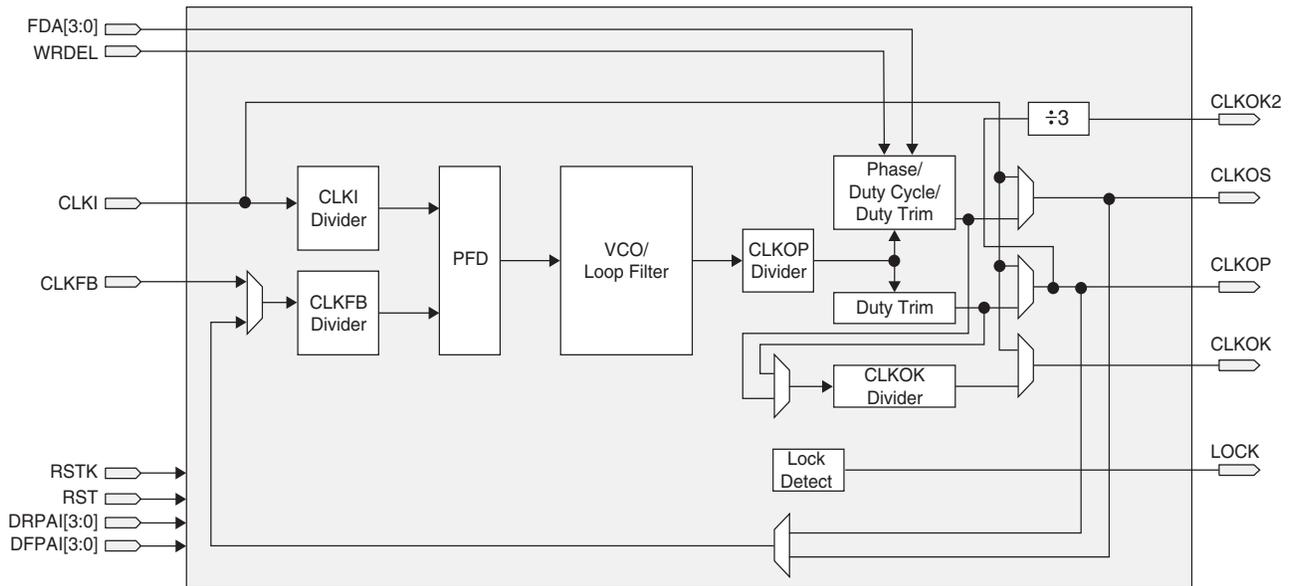


Table 2-4 provides a description of the signals in the PLL blocks.

Table 2-4. PLL Blocks Signal Descriptions

Signal	I/O	Description
CLKI	I	Clock input from external pin or routing
CLKFB	I	PLL feedback input from CLKOP, CLKOS, or from a user clock (pin or logic)
RST	I	"1" to reset PLL counters, VCO, charge pumps and M-dividers
RSTK	I	"1" to reset K-divider
WRDEL	I	DPA Fine Delay Adjust input
CLKOS	O	PLL output to clock tree (phase shifted/duty cycle changed)
CLKOP	O	PLL output to clock tree (no phase shift)
CLKKOK	O	PLL output to clock tree through secondary clock divider
CLKKOK2	O	PLL output to clock tree (CLKOP divided by 3)
LOCK	O	"1" indicates PLL LOCK to CLKI
FDA [3:0]	I	Dynamic fine delay adjustment on CLKOS output
DRPAI[3:0]	I	Dynamic coarse phase shift, rising edge setting
DFPAI[3:0]	I	Dynamic coarse phase shift, falling edge setting

Delay Locked Loops (DLL)

In addition to PLLs, the LatticeECP3 family of devices has two DLLs per device.

CLKI is the input frequency (generated either from the pin or routing) for the DLL. CLKI feeds into the output muxes block to bypass the DLL, directly to the DELAY CHAIN block and (directly or through divider circuit) to the reference input of the Phase Detector (PD) input mux. The reference signal for the PD can also be generated from the Delay Chain signals. The feedback input to the PD is generated from the CLKFB pin or from a tapped signal from the Delay chain.

The PD produces a binary number proportional to the phase and frequency difference between the reference and feedback signals. Based on these inputs, the ALU determines the correct digital control codes to send to the delay

chain in order to better match the reference and feedback signals. This digital code from the ALU is also transmitted via the Digital Control bus (DCNTL) bus to its associated Slave Delay lines (two per DLL). The ALUHOLD input allows the user to suspend the ALU output at its current value. The UDDCNTL signal allows the user to latch the current value on the DCNTL bus.

The DLL has two clock outputs, CLKOP and CLKOS. These outputs can individually select one of the outputs from the tapped delay line. The CLKOS has optional fine delay shift and divider blocks to allow this output to be further modified, if required. The fine delay shift block allows the CLKOS output to phase shifted a further 45, 22.5 or 11.25 degrees relative to its normal position. Both the CLKOS and CLKOP outputs are available with optional duty cycle correction. Divide by two and divide by four frequencies are available at CLKOS. The LOCK output signal is asserted when the DLL is locked. Figure 2-5 shows the DLL block diagram and Table 2-5 provides a description of the DLL inputs and outputs.

The user can configure the DLL for many common functions such as time reference delay mode and clock injection removal mode. Lattice provides primitives in its design tools for these functions.

Figure 2-5. Delay Locked Loop Diagram (DLL)

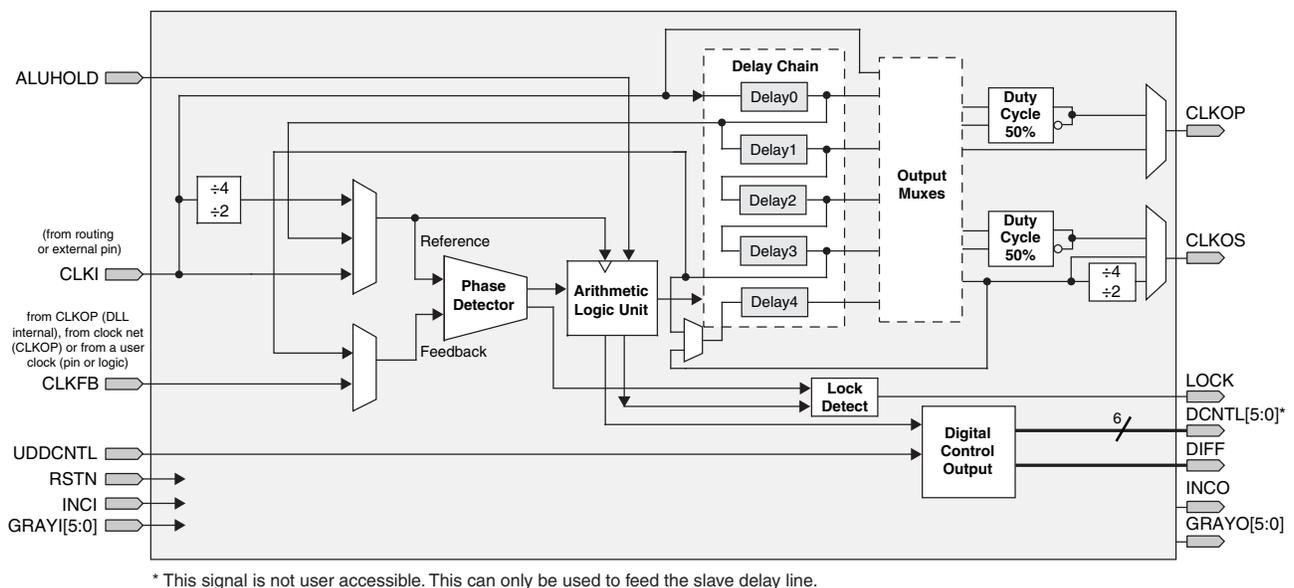
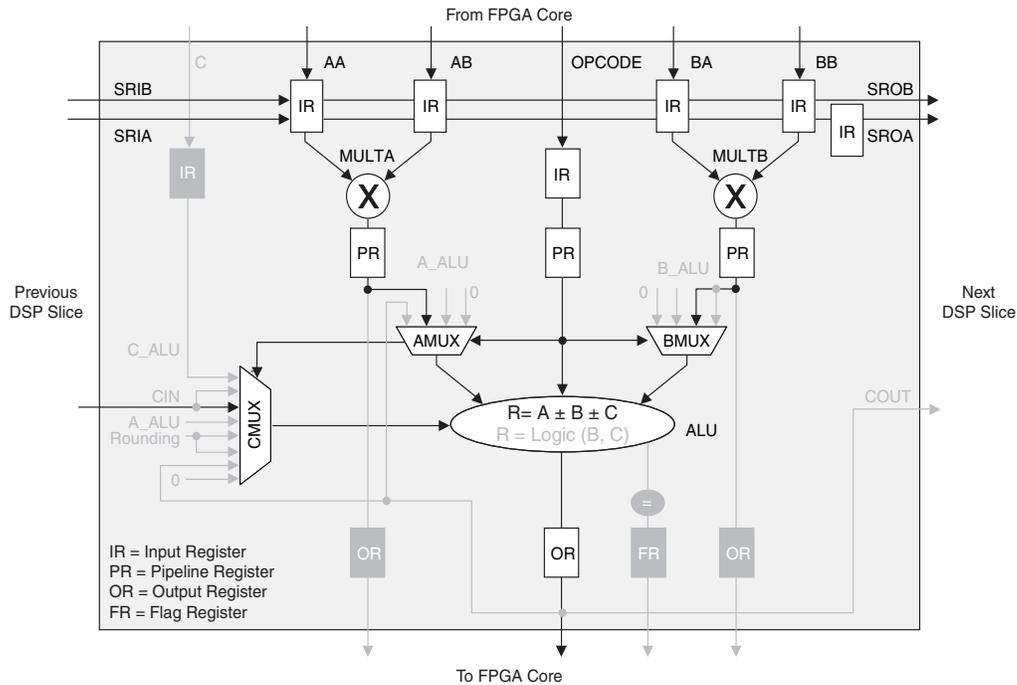


Figure 2-31. MULTADDSUBSUM Slice 1



Advanced sysDSP Slice Features

Cascading

The LatticeECP3 sysDSP slice has been enhanced to allow cascading. Adder trees are implemented fully in sysDSP slices, improving the performance. Cascading of slices uses the signals CIN, COUT and C Mux of the slice.

Addition

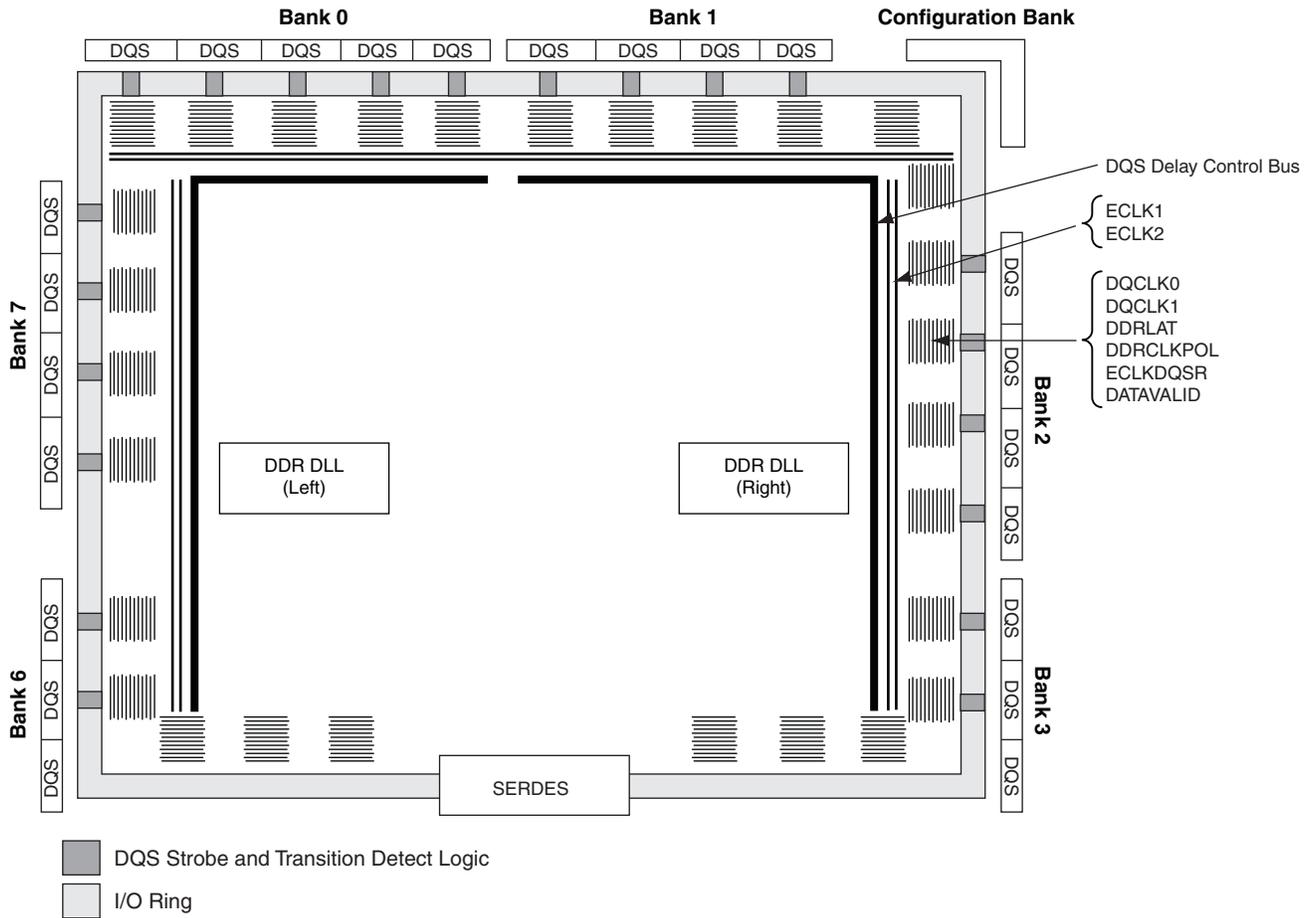
The LatticeECP3 sysDSP slice allows for the bypassing of multipliers and cascading of adder logic. High performance adder functions are implemented without the use of LUTs. The maximum width adders that can be implemented are 54-bit.

Rounding

The rounding operation is implemented in the ALU and is done by adding a constant followed by a truncation operation. The rounding methods supported are:

- Rounding to zero (RTZ)
- Rounding to infinity (RTI)
- Dynamic rounding
- Random rounding
- Convergent rounding

Figure 2-36. Edge Clock, DLL Calibration and DQS Local Bus Distribution



*Includes shared configuration I/Os and dedicated configuration I/Os.

2. Left and Right (Banks 2, 3, 6 and 7) sysI/O Buffer Pairs (50% Differential and 100% Single-Ended Outputs)

The sysI/O buffer pairs in the left and right banks of the device consist of two single-ended output drivers, two sets of single-ended input buffers (both ratioed and referenced) and one differential output driver. One of the referenced input buffers can also be configured as a differential input. In these banks the two pads in the pair are described as “true” and “comp”, where the true pad is associated with the positive side of the differential I/O, and the comp (complementary) pad is associated with the negative side of the differential I/O.

In addition, programmable on-chip input termination (parallel or differential, static or dynamic) is supported on these sides, which is required for DDR3 interface. However, there is no support for hot-socketing for the I/O pins located on the left and right side of the device as the PCI clamp is always enabled on these pins.

LVDS, RSDS, PPLVDS and Mini-LVDS differential output drivers are available on 50% of the buffer pairs on the left and right banks.

3. Configuration Bank sysI/O Buffer Pairs (Single-Ended Outputs, Only on Shared Pins When Not Used by Configuration)

The sysI/O buffers in the Configuration Bank consist of ratioed single-ended output drivers and single-ended input buffers. This bank does not support PCI clamp like the other banks on the top, left, and right sides.

The two pads in the pair are described as “true” and “comp”, where the true pad is associated with the positive side of the differential input buffer and the comp (complementary) pad is associated with the negative side of the differential input buffer.

Programmable PCI clamps are only available on the top banks. PCI clamps are used primarily on inputs and bi-directional pads to reduce ringing on the receiving end.

Typical sysI/O I/O Behavior During Power-up

The internal power-on-reset (POR) signal is deactivated when V_{CC} , V_{CCIO8} and V_{CCAUX} have reached satisfactory levels. After the POR signal is deactivated, the FPGA core logic becomes active. It is the user's responsibility to ensure that all other V_{CCIO} banks are active with valid input logic levels to properly control the output logic states of all the I/O banks that are critical to the application. For more information about controlling the output logic state with valid input logic levels during power-up in LatticeECP3 devices, see the list of technical documentation at the end of this data sheet.

The V_{CC} and V_{CCAUX} supply the power to the FPGA core fabric, whereas the V_{CCIO} supplies power to the I/O buffers. In order to simplify system design while providing consistent and predictable I/O behavior, it is recommended that the I/O buffers be powered-up prior to the FPGA core fabric. V_{CCIO} supplies should be powered-up before or together with the V_{CC} and V_{CCAUX} supplies.

Supported sysI/O Standards

The LatticeECP3 sysI/O buffer supports both single-ended and differential standards. Single-ended standards can be further subdivided into LVCMOS, LVTTTL and other standards. The buffers support the LVTTTL, LVCMOS 1.2 V, 1.5 V, 1.8 V, 2.5 V and 3.3 V standards. In the LVCMOS and LVTTTL modes, the buffer has individual configuration options for drive strength, slew rates, bus maintenance (weak pull-up, weak pull-down, or a bus-keeper latch) and open drain. Other single-ended standards supported include SSTL and HSTL. Differential standards supported include LVDS, BLVDS, LVPECL, MLVDS, RSDS, Mini-LVDS, PPLVDS (point-to-point LVDS), TRLVDS (Transition Reduced LVDS), differential SSTL and differential HSTL. For further information on utilizing the sysI/O buffer to support a variety of standards please see TN1177, [LatticeECP3 sysIO Usage Guide](#).

SCI (SERDES Client Interface) Bus

The SERDES Client Interface (SCI) is an IP interface that allows the SERDES/PCS Quad block to be controlled by registers rather than the configuration memory cells. It is a simple register configuration interface that allows SERDES/PCS configuration without power cycling the device.

The Diamond and ispLEVER design tools support all modes of the PCS. Most modes are dedicated to applications associated with a specific industry standard data protocol. Other more general purpose modes allow users to define their own operation. With these tools, the user can define the mode for each quad in a design.

Popular standards such as 10Gb Ethernet, x4 PCI Express and 4x Serial RapidIO can be implemented using IP (available through Lattice), a single quad (Four SERDES channels and PCS) and some additional logic from the core.

The LatticeECP3 family also supports a wide range of primary and secondary protocols. Within the same quad, the LatticeECP3 family can support mixed protocols with semi-independent clocking as long as the required clock frequencies are integer x1, x2, or x11 multiples of each other. Table 2-15 lists the allowable combination of primary and secondary protocol combinations.

Flexible Quad SERDES Architecture

The LatticeECP3 family SERDES architecture is a quad-based architecture. For most SERDES settings and standards, the whole quad (consisting of four SERDES) is treated as a unit. This helps in silicon area savings, better utilization and overall lower cost.

However, for some specific standards, the LatticeECP3 quad architecture provides flexibility; more than one standard can be supported within the same quad.

Table 2-15 shows the standards can be mixed and matched within the same quad. In general, the SERDES standards whose nominal data rates are either the same or a defined subset of each other, can be supported within the same quad. In Table 2-15, the Primary Protocol column refers to the standard that determines the reference clock and PLL settings. The Secondary Protocol column shows the other standard that can be supported within the same quad.

Furthermore, Table 2-15 also implies that more than two standards in the same quad can be supported, as long as they conform to the data rate and reference clock requirements. For example, a quad may contain PCI Express 1.1, SGMII, Serial RapidIO Type I and Serial RapidIO Type II, all in the same quad.

Table 2-15. LatticeECP3 Primary and Secondary Protocol Support

Primary Protocol	Secondary Protocol
PCI Express 1.1	SGMII
PCI Express 1.1	Gigabit Ethernet
PCI Express 1.1	Serial RapidIO Type I
PCI Express 1.1	Serial RapidIO Type II
Serial RapidIO Type I	SGMII
Serial RapidIO Type I	Gigabit Ethernet
Serial RapidIO Type II	SGMII
Serial RapidIO Type II	Gigabit Ethernet
Serial RapidIO Type II	Serial RapidIO Type I
CPRI-3	CPRI-2 and CPRI-1
3G-SDI	HD-SDI and SD-SDI

Enhanced Configuration Options

LatticeECP3 devices have enhanced configuration features such as: decryption support, TransFR™ I/O and dual-boot image support.

1. TransFR (Transparent Field Reconfiguration)

TransFR I/O (TFR) is a unique Lattice technology that allows users to update their logic in the field without interrupting system operation using a single ispVM command. TransFR I/O allows I/O states to be frozen during device configuration. This allows the device to be field updated with a minimum of system disruption and downtime. See TN1087, [Minimizing System Interruption During Configuration Using TransFR Technology](#) for details.

2. Dual-Boot Image Support

Dual-boot images are supported for applications requiring reliable remote updates of configuration data for the system FPGA. After the system is running with a basic configuration, a new boot image can be downloaded remotely and stored in a separate location in the configuration storage device. Any time after the update the LatticeECP3 can be re-booted from this new configuration file. If there is a problem, such as corrupt data during download or incorrect version number with this new boot image, the LatticeECP3 device can revert back to the original backup golden configuration and try again. This all can be done without power cycling the system. For more information, please see TN1169, [LatticeECP3 sysCONFIG Usage Guide](#).

Soft Error Detect (SED) Support

LatticeECP3 devices have dedicated logic to perform Cycle Redundancy Code (CRC) checks. During configuration, the configuration data bitstream can be checked with the CRC logic block. In addition, the LatticeECP3 device can also be programmed to utilize a Soft Error Detect (SED) mode that checks for soft errors in configuration SRAM. The SED operation can be run in the background during user mode. If a soft error occurs, during user mode (normal operation) the device can be programmed to generate an error signal.

For further information on SED support, please see TN1184, [LatticeECP3 Soft Error Detection \(SED\) Usage Guide](#).

External Resistor

LatticeECP3 devices require a single external, 10 kOhm $\pm 1\%$ value between the XRES pin and ground. Device configuration will not be completed if this resistor is missing. There is no boundary scan register on the external resistor pad.

On-Chip Oscillator

Every LatticeECP3 device has an internal CMOS oscillator which is used to derive a Master Clock (MCCLK) for configuration. The oscillator and the MCCLK run continuously and are available to user logic after configuration is completed. The software default value of the MCCLK is nominally 2.5 MHz. Table 2-16 lists all the available MCCLK frequencies. When a different Master Clock is selected during the design process, the following sequence takes place:

1. Device powers up with a nominal Master Clock frequency of 3.1 MHz.
2. During configuration, users select a different master clock frequency.
3. The Master Clock frequency changes to the selected frequency once the clock configuration bits are received.
4. If the user does not select a master clock frequency, then the configuration bitstream defaults to the MCCLK frequency of 2.5 MHz.

This internal 130 MHz $\pm 15\%$ CMOS oscillator is available to the user by routing it as an input clock to the clock tree. For further information on the use of this oscillator for configuration or user mode, please see TN1169, [LatticeECP3 sysCONFIG Usage Guide](#).

MLVDS25

The LatticeECP3 devices support the differential MLVDS standard. This standard is emulated using complementary LVCMOS outputs in conjunction with a parallel resistor across the driver outputs. The MLVDS input standard is supported by the LVDS differential input buffer. The scheme shown in Figure 3-5 is one possible solution for MLVDS standard implementation. Resistor values in Figure 3-5 are industry standard values for 1% resistors.

Figure 3-5. MLVDS25 (Multipoint Low Voltage Differential Signaling)

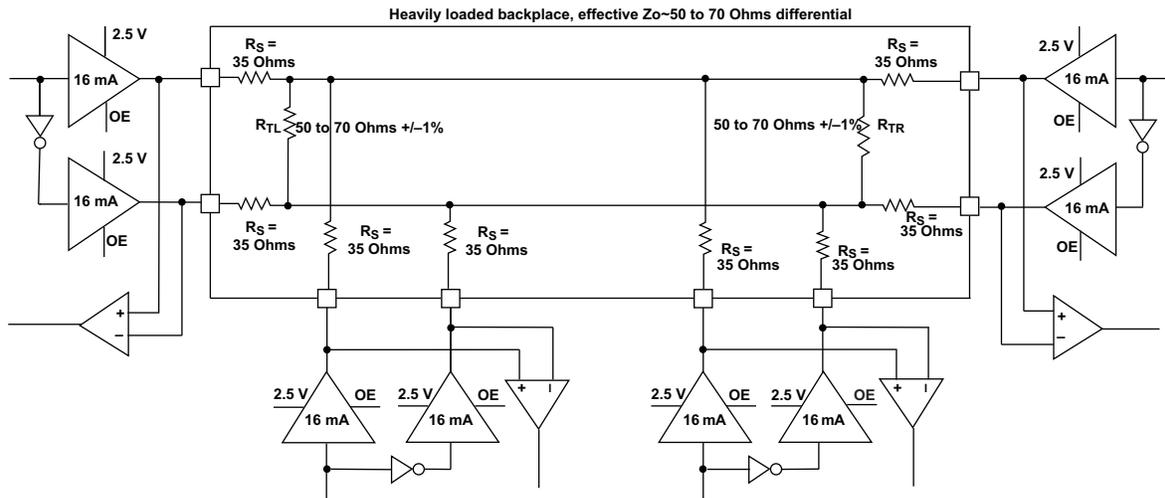


Table 3-5. MLVDS25 DC Conditions¹

Parameter	Description	Typical		Units
		Zo=50Ω	Zo=70Ω	
V _{CCIO}	Output Driver Supply (+/-5%)	2.50	2.50	V
Z _{OUT}	Driver Impedance	10.00	10.00	Ω
R _S	Driver Series Resistor (+/-1%)	35.00	35.00	Ω
R _{TL}	Driver Parallel Resistor (+/-1%)	50.00	70.00	Ω
R _{TR}	Receiver Termination (+/-1%)	50.00	70.00	Ω
V _{OH}	Output High Voltage	1.52	1.60	V
V _{OL}	Output Low Voltage	0.98	0.90	V
V _{OD}	Output Differential Voltage	0.54	0.70	V
V _{CM}	Output Common Mode Voltage	1.25	1.25	V
I _{DC}	DC Output Current	21.74	20.00	mA

1. For input buffer, see LVDS table.

LatticeECP3 External Switching Characteristics ^{1, 2, 3, 13}

Over Recommended Commercial Operating Conditions

Parameter	Description	Device	-8		-7		-6		Units
			Min.	Max.	Min.	Max.	Min.	Max.	
Clocks									
Primary Clock⁶									
f _{MAX_PRI}	Frequency for Primary Clock Tree	ECP3-150EA	—	500	—	420	—	375	MHz
t _{W_PRI}	Clock Pulse Width for Primary Clock	ECP3-150EA	0.8	—	0.9	—	1.0	—	ns
t _{SKEW_PRI}	Primary Clock Skew Within a Device	ECP3-150EA	—	300	—	330	—	360	ps
t _{SKEW_PRIIB}	Primary Clock Skew Within a Bank	ECP3-150EA	—	250	—	280	—	300	ps
f _{MAX_PRI}	Frequency for Primary Clock Tree	ECP3-70EA/95EA	—	500	—	420	—	375	MHz
t _{W_PRI}	Pulse Width for Primary Clock	ECP3-70EA/95EA	0.8	—	0.9	—	1.0	—	ns
t _{SKEW_PRI}	Primary Clock Skew Within a Device	ECP3-70EA/95EA	—	360	—	370	—	380	ps
t _{SKEW_PRIIB}	Primary Clock Skew Within a Bank	ECP3-70EA/95EA	—	310	—	320	—	330	ps
f _{MAX_PRI}	Frequency for Primary Clock Tree	ECP3-35EA	—	500	—	420	—	375	MHz
t _{W_PRI}	Pulse Width for Primary Clock	ECP3-35EA	0.8	—	0.9	—	1.0	—	ns
t _{SKEW_PRI}	Primary Clock Skew Within a Device	ECP3-35EA	—	300	—	330	—	360	ps
t _{SKEW_PRIIB}	Primary Clock Skew Within a Bank	ECP3-35EA	—	250	—	280	—	300	ps
f _{MAX_PRI}	Frequency for Primary Clock Tree	ECP3-17EA	—	500	—	420	—	375	MHz
t _{W_PRI}	Pulse Width for Primary Clock	ECP3-17EA	0.8	—	0.9	—	1.0	—	ns
t _{SKEW_PRI}	Primary Clock Skew Within a Device	ECP3-17EA	—	310	—	340	—	370	ps
t _{SKEW_PRIIB}	Primary Clock Skew Within a Bank	ECP3-17EA	—	220	—	230	—	240	ps
Edge Clock⁶									
f _{MAX_EDGE}	Frequency for Edge Clock	ECP3-150EA	—	500	—	420	—	375	MHz
t _{W_EDGE}	Clock Pulse Width for Edge Clock	ECP3-150EA	0.9	—	1.0	—	1.2	—	ns
t _{SKEW_EDGE_DQS}	Edge Clock Skew Within an Edge of the Device	ECP3-150EA	—	200	—	210	—	220	ps
f _{MAX_EDGE}	Frequency for Edge Clock	ECP3-70EA/95EA	—	500	—	420	—	375	MHz
t _{W_EDGE}	Clock Pulse Width for Edge Clock	ECP3-70EA/95EA	0.9	—	1.0	—	1.2	—	ns
t _{SKEW_EDGE_DQS}	Edge Clock Skew Within an Edge of the Device	ECP3-70EA/95EA	—	200	—	210	—	220	ps
f _{MAX_EDGE}	Frequency for Edge Clock	ECP3-35EA	—	500	—	420	—	375	MHz
t _{W_EDGE}	Clock Pulse Width for Edge Clock	ECP3-35EA	0.9	—	1.0	—	1.2	—	ns
t _{SKEW_EDGE_DQS}	Edge Clock Skew Within an Edge of the Device	ECP3-35EA	—	200	—	210	—	220	ps
f _{MAX_EDGE}	Frequency for Edge Clock	ECP3-17EA	—	500	—	420	—	375	MHz
t _{W_EDGE}	Clock Pulse Width for Edge Clock	ECP3-17EA	0.9	—	1.0	—	1.2	—	ns
t _{SKEW_EDGE_DQS}	Edge Clock Skew Within an Edge of the Device	ECP3-17EA	—	200	—	210	—	220	ps
Generic SDR									
General I/O Pin Parameters Using Dedicated Clock Input Primary Clock Without PLL²									
t _{CO}	Clock to Output - PIO Output Register	ECP3-150EA	—	3.9	—	4.3	—	4.7	ns
t _{SU}	Clock to Data Setup - PIO Input Register	ECP3-150EA	0.0	—	0.0	—	0.0	—	ns
t _H	Clock to Data Hold - PIO Input Register	ECP3-150EA	1.5	—	1.7	—	2.0	—	ns
t _{SU_DEL}	Clock to Data Setup - PIO Input Register with Data Input Delay	ECP3-150EA	1.3	—	1.5	—	1.7	—	ns

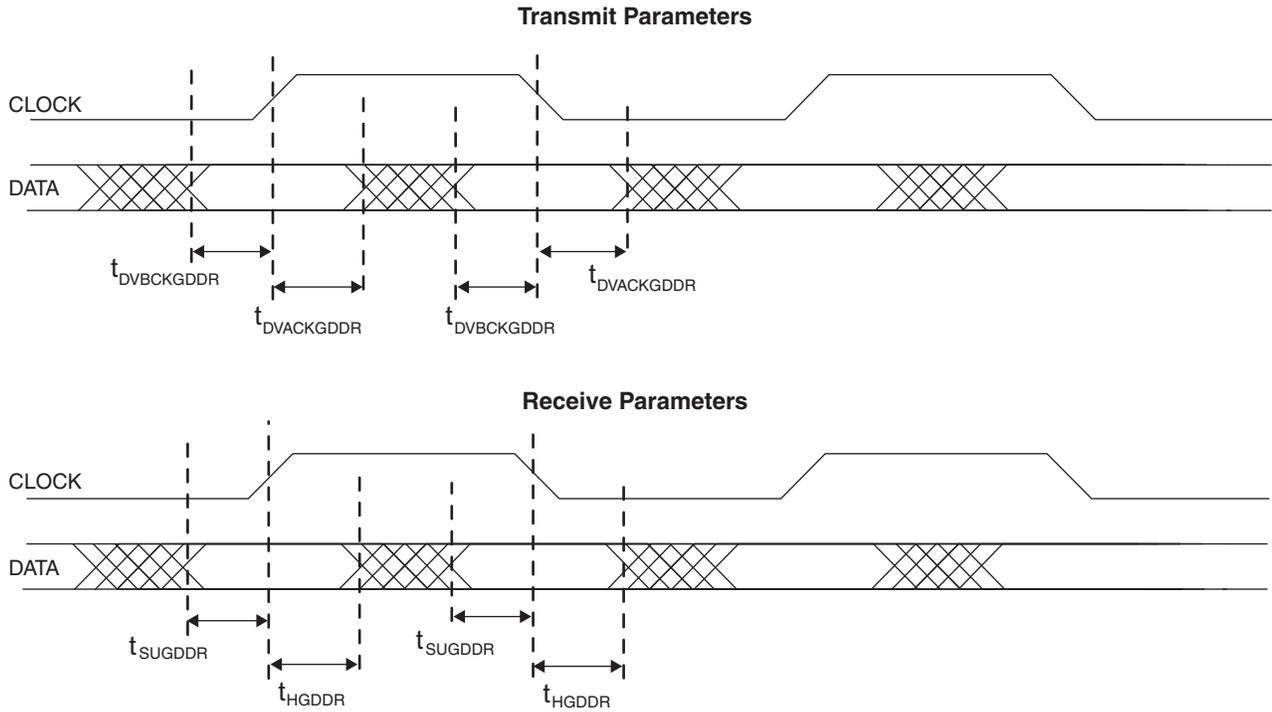
LatticeECP3 External Switching Characteristics (Continued)^{1, 2, 3, 13}

Over Recommended Commercial Operating Conditions

Parameter	Description	Device	-8		-7		-6		Units
			Min.	Max.	Min.	Max.	Min.	Max.	
Generic DDRX2 Output with Clock and Data (>10 Bits Wide) Centered at Pin Using PLL (GDDR2_TX.PLL.Centered)¹⁰									
Left and Right Sides									
t _{DVBGDDR}	Data Valid Before CLK	All ECP3EA Devices	285	—	370	—	431	—	ps
t _{DVAGDDR}	Data Valid After CLK	All ECP3EA Devices	285	—	370	—	432	—	ps
f _{MAX_GDDR}	DDR2 Clock Frequency	All ECP3EA Devices	—	500	—	420	—	375	MHz
Memory Interface									
DDR/DDR2 I/O Pin Parameters (Input Data are Strobe Edge Aligned, Output Strobe Edge is Data Centered)⁴									
t _{DVADQ}	Data Valid After DQS (DDR Read)	All ECP3 Devices	—	0.225	—	0.225	—	0.225	UI
t _{DVEDQ}	Data Hold After DQS (DDR Read)	All ECP3 Devices	0.64	—	0.64	—	0.64	—	UI
t _{DQVBS}	Data Valid Before DQS	All ECP3 Devices	0.25	—	0.25	—	0.25	—	UI
t _{DQVAS}	Data Valid After DQS	All ECP3 Devices	0.25	—	0.25	—	0.25	—	UI
f _{MAX_DDR}	DDR Clock Frequency	All ECP3 Devices	95	200	95	200	95	166	MHz
f _{MAX_DDR2}	DDR2 clock frequency	All ECP3 Devices	125	266	125	200	125	166	MHz
DDR3 (Using PLL for SCLK) I/O Pin Parameters									
t _{DVADQ}	Data Valid After DQS (DDR Read)	All ECP3 Devices	—	0.225	—	0.225	—	0.225	UI
t _{DVEDQ}	Data Hold After DQS (DDR Read)	All ECP3 Devices	0.64	—	0.64	—	0.64	—	UI
t _{DQVBS}	Data Valid Before DQS	All ECP3 Devices	0.25	—	0.25	—	0.25	—	UI
t _{DQVAS}	Data Valid After DQS	All ECP3 Devices	0.25	—	0.25	—	0.25	—	UI
f _{MAX_DDR3}	DDR3 clock frequency	All ECP3 Devices	300	400	266	333	266	300	MHz
DDR3 Clock Timing									
t _{CH} (avg) ⁹	Average High Pulse Width	All ECP3 Devices	0.47	0.53	0.47	0.53	0.47	0.53	UI
t _{CL} (avg) ⁹	Average Low Pulse Width	All ECP3 Devices	0.47	0.53	0.47	0.53	0.47	0.53	UI
t _{JIT} (per, lck) ⁹	Output Clock Period Jitter During DLL Locking Period	All ECP3 Devices	-90	90	-90	90	-90	90	ps
t _{JIT} (cc, lck) ⁹	Output Cycle-to-Cycle Period Jitter During DLL Locking Period	All ECP3 Devices	—	180	—	180	—	180	ps

- Commercial timing numbers are shown. Industrial numbers are typically slower and can be extracted from the Diamond or ispLEVER software.
- General I/O timing numbers based on LVCMOS 2.5, 12mA, Fast Slew Rate, 0pf load.
- Generic DDR timing numbers based on LVDS I/O.
- DDR timing numbers based on SSTL25. DDR2 timing numbers based on SSTL18.
- DDR3 timing numbers based on SSTL15.
- Uses LVDS I/O standard.
- The current version of software does not support per bank skew numbers; this will be supported in a future release.
- Maximum clock frequencies are tested under best case conditions. System performance may vary upon the user environment.
- Using settings generated by IPexpress.
- These numbers are generated using best case PLL located in the center of the device.
- Uses SSTL25 Class II Differential I/O Standard.
- All numbers are generated with ispLEVER 8.1 software.
- For details on -9 speed grade devices, please contact your Lattice Sales Representative.

Figure 3-8. Generic DDRX1/DDR2 (With Clock Center on Data Window)



LatticeECP3 Internal Switching Characteristics^{1, 2, 5} (Continued)

Over Recommended Commercial Operating Conditions

Parameter	Description	-8		-7		-6		Units.
		Min.	Max.	Min.	Max.	Min.	Max.	
t _{HWREN_EBR}	Hold Write/Read Enable to EBR Memory	0.141	—	0.145	—	0.149	—	ns
t _{SUCE_EBR}	Clock Enable Setup Time to EBR Output Register	0.087	—	0.096	—	0.104	—	ns
t _{HCE_EBR}	Clock Enable Hold Time to EBR Output Register	-0.066	—	-0.080	—	-0.094	—	ns
t _{SUBE_EBR}	Byte Enable Set-Up Time to EBR Output Register	-0.071	—	-0.070	—	-0.068	—	ns
t _{HBE_EBR}	Byte Enable Hold Time to EBR Output Register	0.118	—	0.098	—	0.077	—	ns
DSP Block Timing³								
t _{SUI_DSP}	Input Register Setup Time	0.32	—	0.36	—	0.39	—	ns
t _{HI_DSP}	Input Register Hold Time	-0.17	—	-0.19	—	-0.21	—	ns
t _{SUP_DSP}	Pipeline Register Setup Time	2.23	—	2.30	—	2.37	—	ns
t _{HP_DSP}	Pipeline Register Hold Time	-1.02	—	-1.09	—	-1.15	—	ns
t _{SUO_DSP}	Output Register Setup Time	3.09	—	3.22	—	3.34	—	ns
t _{HO_DSP}	Output Register Hold Time	-1.67	—	-1.76	—	-1.84	—	ns
t _{COI_DSP}	Input Register Clock to Output Time	—	3.05	—	3.35	—	3.73	ns
t _{COP_DSP}	Pipeline Register Clock to Output Time	—	1.30	—	1.47	—	1.64	ns
t _{COO_DSP}	Output Register Clock to Output Time	—	0.58	—	0.60	—	0.62	ns
t _{SUOPT_DSP}	Opcode Register Setup Time	0.31	—	0.35	—	0.39	—	ns
t _{HOPT_DSP}	Opcode Register Hold Time	-0.20	—	-0.24	—	-0.27	—	ns
t _{SUDATA_DSP}	Cascade_data through ALU to Output Register Setup Time	1.69	—	1.94	—	2.14	—	ns
t _{HPDATA_DSP}	Cascade_data through ALU to Output Register Hold Time	-0.58	—	-0.80	—	-0.97	—	ns

1. Internal parameters are characterized but not tested on every device.

2. Commercial timing numbers are shown. Industrial timing numbers are typically slower and can be extracted from the Diamond or ispLEVER software.

3. DSP slice is configured in Multiply Add/Sub 18 x 18 mode.

4. The output register is in Flip-flop mode.

5. For details on -9 speed grade devices, please contact your Lattice Sales Representative.

LatticeECP3 Maximum I/O Buffer Speed (Continued)^{1, 2, 3, 4, 5, 6}**Over Recommended Operating Conditions**

Buffer	Description	Max.	Units
PCI33	PCI, $V_{CCIO} = 3.3\text{ V}$	66	MHz

1. These maximum speeds are characterized but not tested on every device.
2. Maximum I/O speed for differential output standards emulated with resistors depends on the layout.
3. LVCMOS timing is measured with the load specified in the Switching Test Conditions table of this document.
4. All speeds are measured at fast slew.
5. Actual system operation may vary depending on user logic implementation.
6. Maximum data rate equals 2 times the clock rate when utilizing DDR.

SERDES/PCS Block Latency

Table 3-8 describes the latency of each functional block in the transmitter and receiver. Latency is given in parallel clock cycles. Figure 3-12 shows the location of each block.

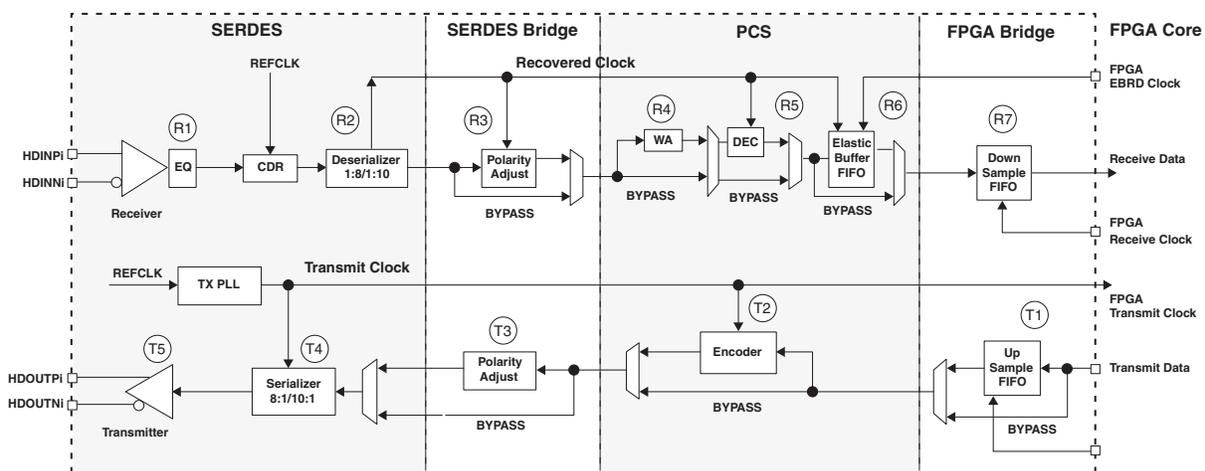
Table 3-8. SERDES/PCS Latency Breakdown

Item	Description	Min.	Avg.	Max.	Fixed	Bypass	Units
Transmit Data Latency¹							
T1	FPGA Bridge - Gearing disabled with different clocks	1	3	5	—	1	word clk
	FPGA Bridge - Gearing disabled with same clocks	—	—	—	3	1	word clk
	FPGA Bridge - Gearing enabled	1	3	5	—	—	word clk
T2	8b10b Encoder	—	—	—	2	1	word clk
T3	SERDES Bridge transmit	—	—	—	2	1	word clk
T4	Serializer: 8-bit mode	—	—	—	15 + Δ1	—	UI + ps
	Serializer: 10-bit mode	—	—	—	18 + Δ1	—	UI + ps
T5	Pre-emphasis ON	—	—	—	1 + Δ2	—	UI + ps
	Pre-emphasis OFF	—	—	—	0 + Δ3	—	UI + ps
Receive Data Latency²							
R1	Equalization ON	—	—	—	Δ1	—	UI + ps
	Equalization OFF	—	—	—	Δ2	—	UI + ps
R2	Deserializer: 8-bit mode	—	—	—	10 + Δ3	—	UI + ps
	Deserializer: 10-bit mode	—	—	—	12 + Δ3	—	UI + ps
R3	SERDES Bridge receive	—	—	—	2	—	word clk
R4	Word alignment	3.1	—	4	—	—	word clk
R5	8b10b decoder	—	—	—	1	—	word clk
R6	Clock Tolerance Compensation	7	15	23	1	1	word clk
R7	FPGA Bridge - Gearing disabled with different clocks	1	3	5	—	1	word clk
	FPGA Bridge - Gearing disabled with same clocks	—	—	—	3	1	word clk
	FPGA Bridge - Gearing enabled	1	3	5	—	—	word clk

1. Δ1 = -245 ps, Δ2 = +88 ps, Δ3 = +112 ps.

2. Δ1 = +118 ps, Δ2 = +132 ps, Δ3 = +700 ps.

Figure 3-12. Transmitter and Receiver Latency Block Diagram



SERDES High Speed Data Receiver

Table 3-9. Serial Input Data Specifications

Symbol	Description	Min.	Typ.	Max.	Units	
RX-CID _S	Stream of nontransitions ¹ (CID = Consecutive Identical Digits) @ 10 ⁻¹² BER	3.125 G	—	—	136	Bits
		2.5 G	—	—	144	
		1.485 G	—	—	160	
		622 M	—	—	204	
		270 M	—	—	228	
		150 M	—	—	296	
V _{RX-DIFF-S}	Differential input sensitivity	150	—	1760	mV, p-p	
V _{RX-IN}	Input levels	0	—	V _{CCA} +0.5 ⁴	V	
V _{RX-CM-DC}	Input common mode range (DC coupled)	0.6	—	V _{CCA}	V	
V _{RX-CM-AC}	Input common mode range (AC coupled) ³	0.1	—	V _{CCA} +0.2	V	
T _{RX-RELOCK}	SCDR re-lock time ²	—	1000	—	Bits	
Z _{RX-TERM}	Input termination 50/75 Ohm/High Z	-20%	50/75/HiZ	+20%	Ohms	
RL _{RX-RL}	Return loss (without package)	10	—	—	dB	

1. This is the number of bits allowed without a transition on the incoming data stream when using DC coupling.
2. This is the typical number of bit times to re-lock to a new phase or frequency within +/- 300 ppm, assuming 8b10b encoded data.
3. AC coupling is used to interface to LVPECL and LVDS. LVDS interfaces are found in laser drivers and Fibre Channel equipment. LVDS interfaces are generally found in 622 Mbps SERDES devices.
4. Up to 1.76 V.

Input Data Jitter Tolerance

A receiver's ability to tolerate incoming signal jitter is very dependent on jitter type. High speed serial interface standards have recognized the dependency on jitter type and have specifications to indicate tolerance levels for different jitter types as they relate to specific protocols. Sinusoidal jitter is considered to be a worst case jitter type.

Table 3-10. Receiver Total Jitter Tolerance Specification

Description	Frequency	Condition	Min.	Typ.	Max.	Units
Deterministic	3.125 Gbps	600 mV differential eye	—	—	0.47	UI, p-p
Random		600 mV differential eye	—	—	0.18	UI, p-p
Total		600 mV differential eye	—	—	0.65	UI, p-p
Deterministic	2.5 Gbps	600 mV differential eye	—	—	0.47	UI, p-p
Random		600 mV differential eye	—	—	0.18	UI, p-p
Total		600 mV differential eye	—	—	0.65	UI, p-p
Deterministic	1.25 Gbps	600 mV differential eye	—	—	0.47	UI, p-p
Random		600 mV differential eye	—	—	0.18	UI, p-p
Total		600 mV differential eye	—	—	0.65	UI, p-p
Deterministic	622 Mbps	600 mV differential eye	—	—	0.47	UI, p-p
Random		600 mV differential eye	—	—	0.18	UI, p-p
Total		600 mV differential eye	—	—	0.65	UI, p-p

Note: Values are measured with CJPAT, all channels operating, FPGA Logic active, I/Os around SERDES pins quiet, voltages are nominal, room temperature.

XAUI/Serial Rapid I/O Type 3/CPRI LV E.30 Electrical and Timing Characteristics

AC and DC Characteristics

Table 3-13. Transmit

Over Recommended Operating Conditions

Symbol	Description	Test Conditions	Min.	Typ.	Max.	Units
T_{RF}	Differential rise/fall time	20%-80%	—	80	—	ps
$Z_{TX_DIFF_DC}$	Differential impedance		80	100	120	Ohms
$J_{TX_DDJ}^{2,3,4}$	Output data deterministic jitter		—	—	0.17	UI
$J_{TX_TJ}^{1,2,3,4}$	Total output data jitter		—	—	0.35	UI

1. Total jitter includes both deterministic jitter and random jitter.
2. Jitter values are measured with each CML output AC coupled into a 50-Ohm impedance (100-Ohm differential impedance).
3. Jitter and skew are specified between differential crossings of the 50% threshold of the reference signal.
4. Values are measured at 2.5 Gbps.

Table 3-14. Receive and Jitter Tolerance

Over Recommended Operating Conditions

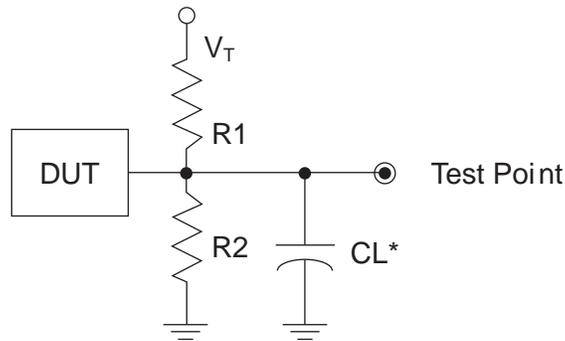
Symbol	Description	Test Conditions	Min.	Typ.	Max.	Units
RL_{RX_DIFF}	Differential return loss	From 100 MHz to 3.125 GHz	10	—	—	dB
RL_{RX_CM}	Common mode return loss	From 100 MHz to 3.125 GHz	6	—	—	dB
Z_{RX_DIFF}	Differential termination resistance		80	100	120	Ohms
$J_{RX_DJ}^{1,2,3}$	Deterministic jitter tolerance (peak-to-peak)		—	—	0.37	UI
$J_{RX_RJ}^{1,2,3}$	Random jitter tolerance (peak-to-peak)		—	—	0.18	UI
$J_{RX_SJ}^{1,2,3}$	Sinusoidal jitter tolerance (peak-to-peak)		—	—	0.10	UI
$J_{RX_TJ}^{1,2,3}$	Total jitter tolerance (peak-to-peak)		—	—	0.65	UI
T_{RX_EYE}	Receiver eye opening		0.35	—	—	UI

1. Total jitter includes deterministic jitter, random jitter and sinusoidal jitter. The sinusoidal jitter tolerance mask is shown in Figure 3-18.
2. Jitter values are measured with each high-speed input AC coupled into a 50-Ohm impedance.
3. Jitter and skew are specified between differential crossings of the 50% threshold of the reference signal.
4. Jitter tolerance parameters are characterized when Full Rx Equalization is enabled.
5. Values are measured at 2.5 Gbps.

Switching Test Conditions

Figure 3-33 shows the output test load that is used for AC testing. The specific values for resistance, capacitance, voltage, and other test conditions are shown in Table 3-23.

Figure 3-33. Output Test Load, LVTTTL and LVCMOS Standards



*CL Includes Test Fixture and Probe Capacitance

Table 3-23. Test Fixture Required Components, Non-Terminated Interfaces

Test Condition	R ₁	R ₂	C _L	Timing Ref.	V _T
LVTTTL and other LVCMOS settings (L -> H, H -> L)	∞	∞	0 pF	LVCMOS 3.3 = 1.5V	—
				LVCMOS 2.5 = V _{CCIO} /2	—
				LVCMOS 1.8 = V _{CCIO} /2	—
				LVCMOS 1.5 = V _{CCIO} /2	—
				LVCMOS 1.2 = V _{CCIO} /2	—
LVCMOS 2.5 I/O (Z -> H)	∞	1MΩ	0 pF	V _{CCIO} /2	—
LVCMOS 2.5 I/O (Z -> L)	1 MΩ	∞	0 pF	V _{CCIO} /2	V _{CCIO}
LVCMOS 2.5 I/O (H -> Z)	∞	100	0 pF	V _{OH} - 0.10	—
LVCMOS 2.5 I/O (L -> Z)	100	∞	0 pF	V _{OL} + 0.10	V _{CCIO}

Note: Output test conditions for all other interfaces are determined by the respective standards.

Signal Descriptions (Cont.)

Signal Name	I/O	Description
[LOC]DQS[num]	I/O	DQ input/output pads: T (top), R (right), B (bottom), L (left), DQS, num = ball function number.
[LOC]DQ[num]	I/O	DQ input/output pads: T (top), R (right), B (bottom), L (left), DQ, associated DQS number.
Test and Programming (Dedicated Pins)		
TMS	I	Test Mode Select input, used to control the 1149.1 state machine. Pull-up is enabled during configuration.
TCK	I	Test Clock input pin, used to clock the 1149.1 state machine. No pull-up enabled.
TDI	I	Test Data in pin. Used to load data into device using 1149.1 state machine. After power-up, this TAP port can be activated for configuration by sending appropriate command. (Note: once a configuration port is selected it is locked. Another configuration port cannot be selected until the power-up sequence). Pull-up is enabled during configuration.
TDO	O	Output pin. Test Data Out pin used to shift data out of a device using 1149.1.
VCCJ	—	Power supply pin for JTAG Test Access Port.
Configuration Pads (Used During sysCONFIG)		
CFG[2:0]	I	Mode pins used to specify configuration mode values latched on rising edge of INITN. During configuration, a pull-up is enabled. These are dedicated pins.
INITN	I/O	Open Drain pin. Indicates the FPGA is ready to be configured. During configuration, a pull-up is enabled. It is a dedicated pin.
PROGRAMN	I	Initiates configuration sequence when asserted low. This pin always has an active pull-up. It is a dedicated pin.
DONE	I/O	Open Drain pin. Indicates that the configuration sequence is complete, and the startup sequence is in progress. It is a dedicated pin.
CCLK	I	Input Configuration Clock for configuring an FPGA in Slave SPI, Serial, and CPU modes. It is a dedicated pin.
MCLK	I/O	Output Configuration Clock for configuring an FPGA in SPI, SPIm, and Master configuration modes.
BUSY/SISPI	O	Parallel configuration mode busy indicator. SPI/SPIm mode data output.
CSN/SN/OEN	I/O	Parallel configuration mode active-low chip select. Slave SPI chip select. Parallel burst Flash output enable.
CS1N/HOLDN/RDY	I	Parallel configuration mode active-low chip select. Slave SPI hold input.
WRITEN	I	Write enable for parallel configuration modes.
DOUT/CSN/CSSPI1N	O	Serial data output. Chip select output. SPI/SPIm mode chip select.
D[0]/SPIFASTN	I/O	sysCONFIG Port Data I/O for Parallel mode. Open drain during configuration. sysCONFIG Port Data I/O for SPI or SPIm. When using the SPI or SPIm mode, this pin should either be tied high or low, must not be left floating. Open drain during configuration.
D1	I/O	Parallel configuration I/O. Open drain during configuration.
D2	I/O	Parallel configuration I/O. Open drain during configuration.
D3/SI	I/O	Parallel configuration I/O. Slave SPI data input. Open drain during configuration.
D4/SO	I/O	Parallel configuration I/O. Slave SPI data output. Open drain during configuration.
D5	I/O	Parallel configuration I/O. Open drain during configuration.
D6/SPID1	I/O	Parallel configuration I/O. SPI/SPIm data input. Open drain during configuration.