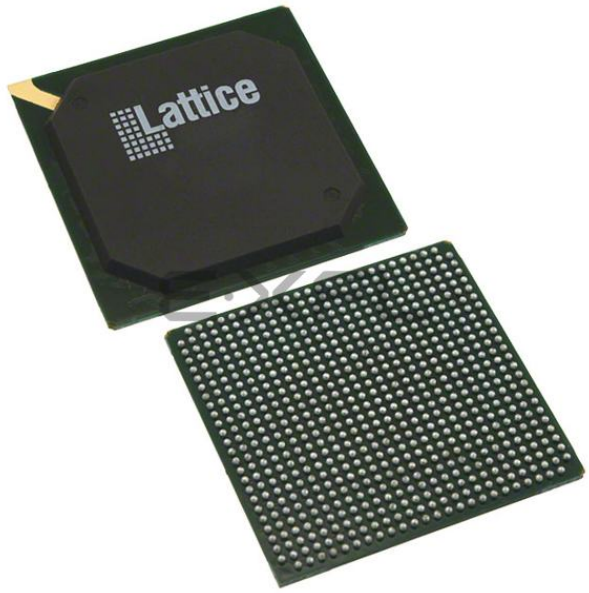
Lattice Semiconductor Corporation - LFE3-70EA-6FN672I Datasheet



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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Active
Number of LABs/CLBs	8375
Number of Logic Elements/Cells	67000
Total RAM Bits	4526080
Number of I/O	380
Number of Gates	-
Voltage - Supply	1.14V ~ 1.26V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 100°C (TJ)
Package / Case	672-BBGA
Supplier Device Package	672-FPBGA (27x27)
Purchase URL	https://www.e-xfl.com/product-detail/lattice-semiconductor/lfe3-70ea-6fn672i

Modes of Operation

Each slice has up to four potential modes of operation: Logic, Ripple, RAM and ROM.

Logic Mode

In this mode, the LUTs in each slice are configured as 4-input combinatorial lookup tables. A LUT4 can have 16 possible input combinations. Any four input logic functions can be generated by programming this lookup table. Since there are two LUT4s per slice, a LUT5 can be constructed within one slice. Larger look-up tables such as LUT6, LUT7 and LUT8 can be constructed by concatenating other slices. Note LUT8 requires more than four slices.

Ripple Mode

Ripple mode supports the efficient implementation of small arithmetic functions. In ripple mode, the following functions can be implemented by each slice:

- Addition 2-bit
- Subtraction 2-bit
- Add/Subtract 2-bit using dynamic control
- Up counter 2-bit
- Down counter 2-bit
- Up/Down counter with asynchronous clear
- Up/Down counter with preload (sync)
- Ripple mode multiplier building block
- Multiplier support
- Comparator functions of A and B inputs
  - A greater-than-or-equal-to B
  - A not-equal-to B
  - A less-than-or-equal-to B

Ripple Mode includes an optional configuration that performs arithmetic using fast carry chain methods. In this configuration (also referred to as CCU2 mode) two additional signals, Carry Generate and Carry Propagate, are generated on a per slice basis to allow fast arithmetic functions to be constructed by concatenating Slices.

RAM Mode

In this mode, a 16x4-bit distributed single port RAM (SPR) can be constructed using each LUT block in Slice 0 and Slice 1 as a 16x1-bit memory. Slice 2 is used to provide memory address and control signals. A 16x2-bit pseudo dual port RAM (PDPR) memory is created by using one Slice as the read-write port and the other companion slice as the read-only port.

LatticeECP3 devices support distributed memory initialization.

The Lattice design tools support the creation of a variety of different size memories. Where appropriate, the software will construct these using distributed memory primitives that represent the capabilities of the PFU. Table 2-3 shows the number of slices required to implement different distributed RAM primitives. For more information about using RAM in LatticeECP3 devices, please see TN1179, LatticeECP3 Memory Usage Guide.

Table 2-3. Number of Slices Required to Implement Distributed RAM

	SPR 16X4	PDPR 16X4
Number of slices	3	3

Note: SPR = Single Port RAM, PDPR = Pseudo Dual Port RAM

PLL/DLL Cascading

LatticeECP3 devices have been designed to allow certain combinations of PLL and DLL cascading. The allowable combinations are:

- PLL to PLL supported
- PLL to DLL supported

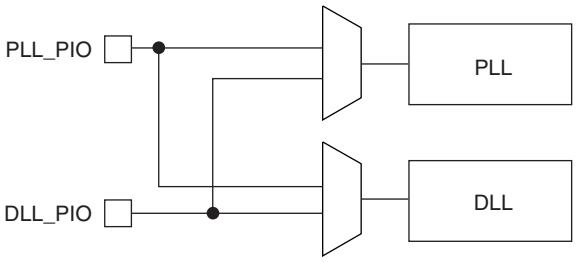
The DLLs in the LatticeECP3 are used to shift the clock in relation to the data for source synchronous inputs. PLLs are used for frequency synthesis and clock generation for source synchronous interfaces. Cascading PLL and DLL blocks allows applications to utilize the unique benefits of both DLLs and PLLs.

For further information about the DLL, please see the list of technical documentation at the end of this data sheet.

PLL/DLL PIO Input Pin Connections

All LatticeECP3 devices contains two DLLs and up to ten PLLs, arranged in quadrants. If a PLL and a DLL are next to each other, they share input pins as shown in the Figure 2-7.

Figure 2-7. Sharing of PIO Pins by PLLs and DLLs in LatticeECP3 Devices

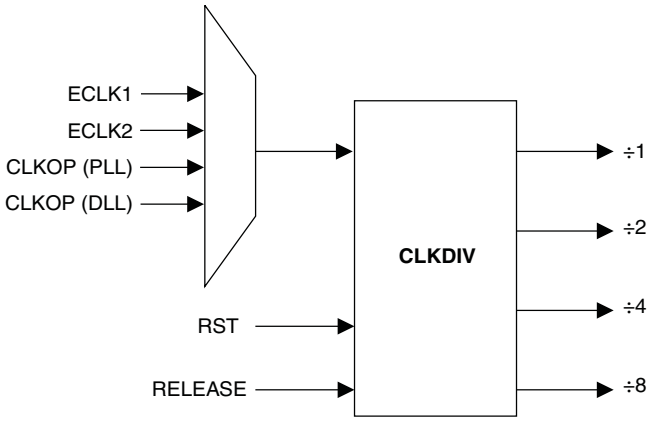


Note: Not every PLL has an associated DLL.

Clock Dividers

LatticeECP3 devices have two clock dividers, one on the left side and one on the right side of the device. These are intended to generate a slower-speed system clock from a high-speed edge clock. The block operates in a ÷2, ÷4 or ÷8 mode and maintains a known phase relationship between the divided down clock and the high-speed clock based on the release of its reset signal. The clock dividers can be fed from selected PLL/DLL outputs, the Slave Delay lines, routing or from an external clock input. The clock divider outputs serve as primary clock sources and feed into the clock distribution network. The Reset (RST) control signal resets input and asynchronously forces all outputs to low. The RELEASE signal releases outputs synchronously to the input clock. For further information on clock dividers, please see TN1178, LatticeECP3 sysCLOCK PLL/DLL Design and Usage Guide. Figure 2-8 shows the clock divider connections.

Figure 2-8. Clock Divider Connections



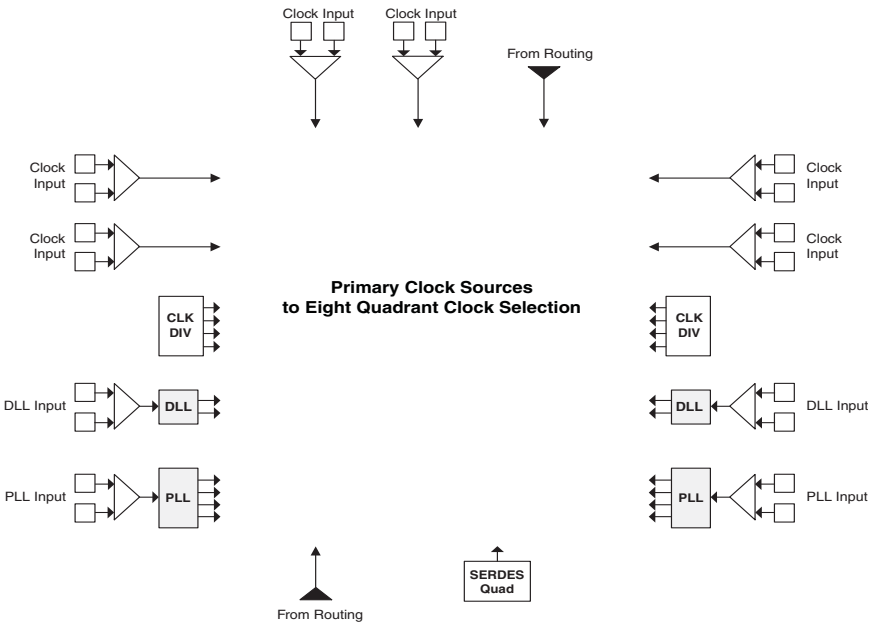
Clock Distribution Network

LatticeECP3 devices have eight quadrant-based primary clocks and eight secondary clock/control sources. Two high performance edge clocks are available on the top, left, and right edges of the device to support high speed interfaces. These clock sources are selected from external I/Os, the sysCLOCK PLLs, DLLs or routing. These clock sources are fed throughout the chip via a clock distribution system.

Primary Clock Sources

LatticeECP3 devices derive clocks from six primary source types: PLL outputs, DLL outputs, CLKDIV outputs, dedicated clock inputs, routing and SERDES Quads. LatticeECP3 devices have two to ten sysCLOCK PLLs and two DLLs, located on the left and right sides of the device. There are six dedicated clock inputs: two on the top side, two on the left side and two on the right side of the device. Figures 2-9, 2-10 and 2-11 show the primary clock sources for LatticeECP3 devices.

Figure 2-9. Primary Clock Sources for LatticeECP3-17

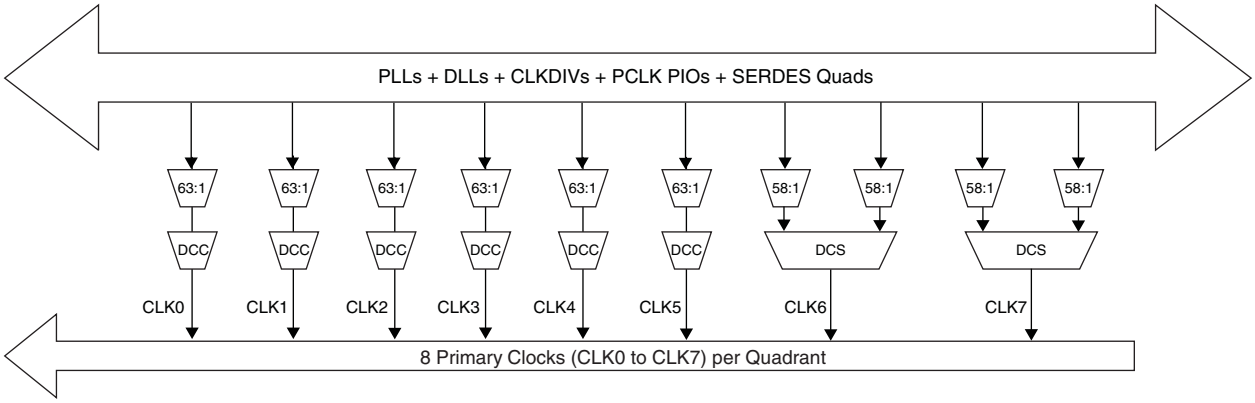


Note: Clock inputs can be configured in differential or single-ended mode.

Primary Clock Routing

The purpose of the primary clock routing is to distribute primary clock sources to the destination quadrants of the device. A global primary clock is a primary clock that is distributed to all quadrants. The clock routing structure in LatticeECP3 devices consists of a network of eight primary clock lines (CLK0 through CLK7) per quadrant. The primary clocks of each quadrant are generated from muxes located in the center of the device. All the clock sources are connected to these muxes. Figure 2-12 shows the clock routing for one quadrant. Each quadrant mux is identical. If desired, any clock can be routed globally.

Figure 2-12. Per Quadrant Primary Clock Selection



Dynamic Clock Control (DCC)

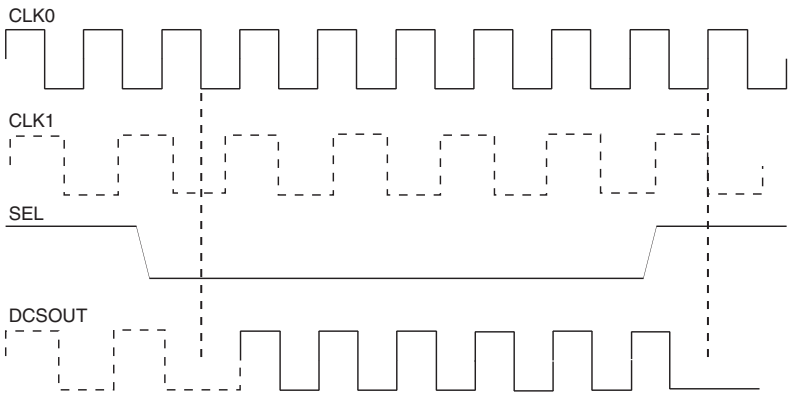
The DCC (Quadrant Clock Enable/Disable) feature allows internal logic control of the quadrant primary clock network. When a clock network is disabled, all the logic fed by that clock does not toggle, reducing the overall power consumption of the device.

Dynamic Clock Select (DCS)

The DCS is a smart multiplexer function available in the primary clock routing. It switches between two independent input clock sources without any glitches or runt pulses. This is achieved regardless of when the select signal is toggled. There are two DCS blocks per quadrant; in total, there are eight DCS blocks per device. The inputs to the DCS block come from the center muxes. The output of the DCS is connected to primary clocks CLK6 and CLK7 (see Figure 2-12).

Figure 2-13 shows the timing waveforms of the default DCS operating mode. The DCS block can be programmed to other modes. For more information about the DCS, please see the list of technical documentation at the end of this data sheet.

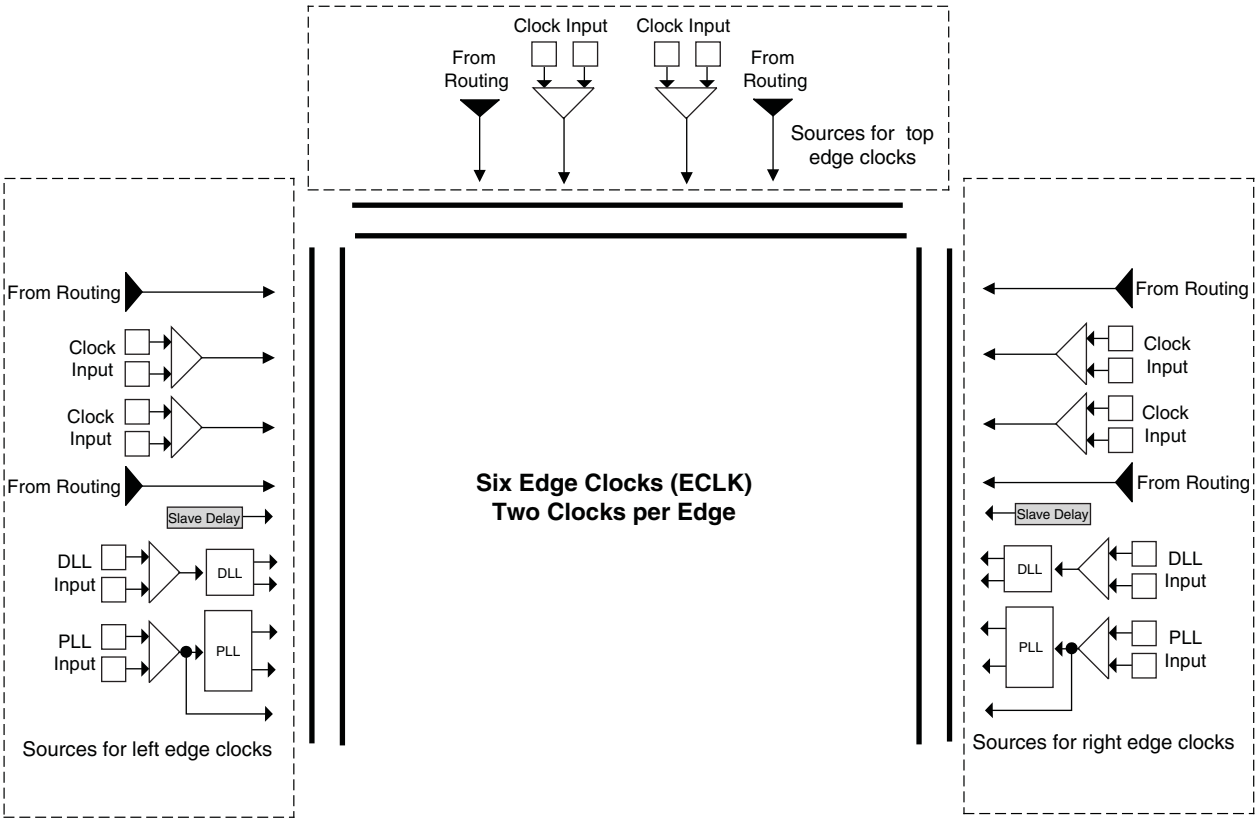
Figure 2-13. DCS Waveforms



Edge Clock Sources

Edge clock resources can be driven from a variety of sources at the same edge. Edge clock resources can be driven from adjacent edge clock PIOs, primary clock PIOs, PLLs, DLLs, Slave Delay and clock dividers as shown in Figure 2-19.

Figure 2-19. Edge Clock Sources



Notes:
1. Clock inputs can be configured in differential or single ended mode.
2. The two DLLs can also drive the two top edge clocks.
3. The top left and top right PLL can also drive the two top edge clocks.

Edge Clock Routing

LatticeECP3 devices have a number of high-speed edge clocks that are intended for use with the PIOs in the implementation of high-speed interfaces. There are six edge clocks per device: two edge clocks on each of the top, left, and right edges. Different PLL and DLL outputs are routed to the two muxes on the left and right sides of the device. In addition, the CLKINDEL signal (generated from the DLL Slave Delay Line block) is routed to all the edge clock muxes on the left and right sides of the device. Figure 2-20 shows the selection muxes for these clocks.

Figure 2-20. Sources of Edge Clock (Left and Right Edges)

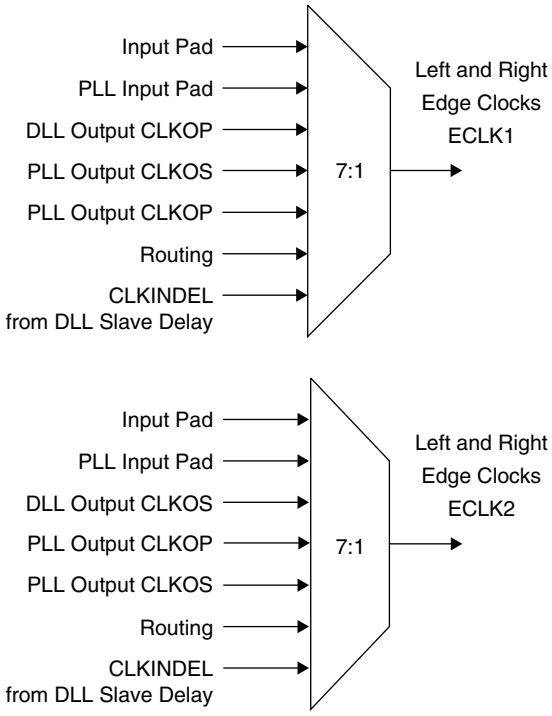
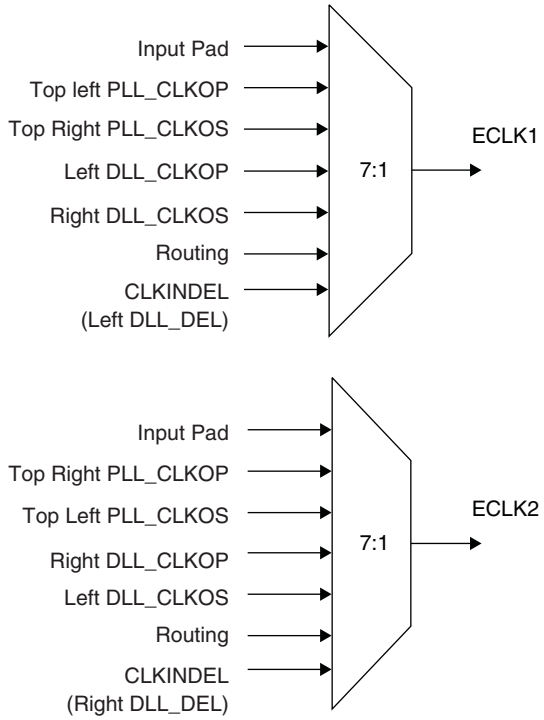


Figure 2-21. Sources of Edge Clock (Top Edge)

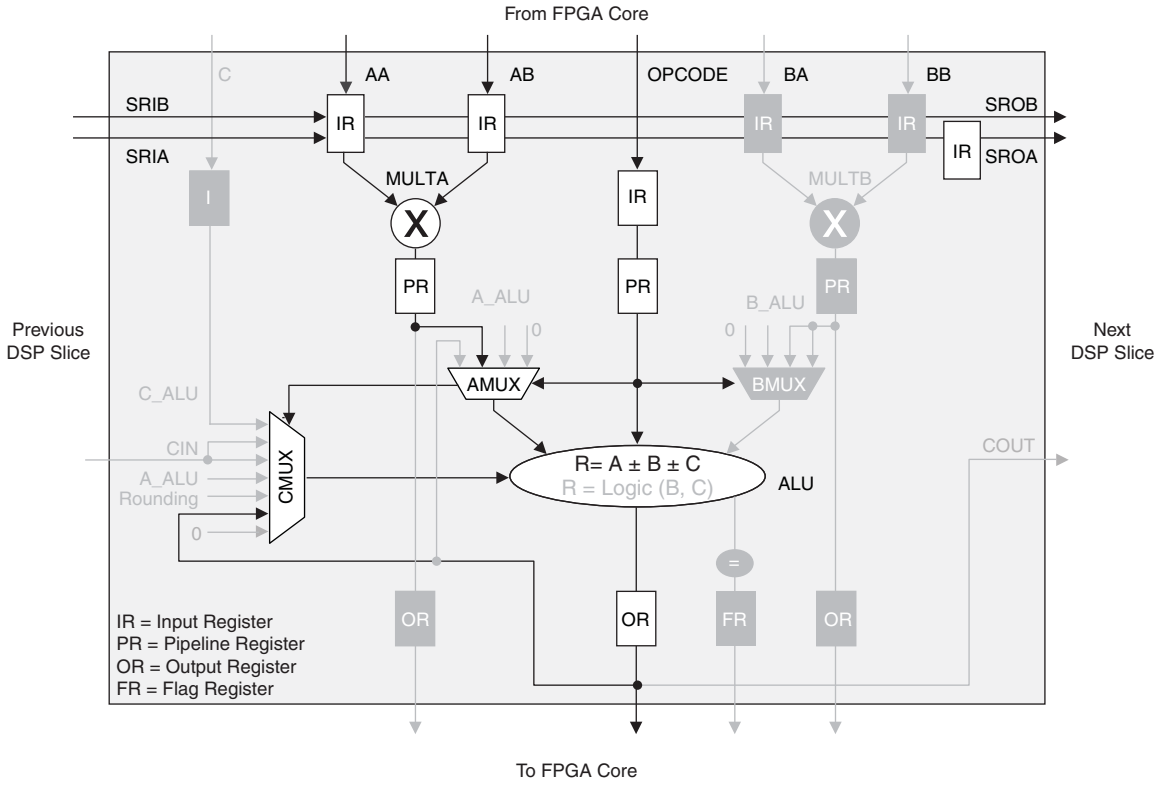


The edge clocks have low injection delay and low skew. They are used to clock the I/O registers and thus are ideal for creating I/O interfaces with a single clock signal and a wide data bus. They are also used for DDR Memory or Generic DDR interfaces.

MAC DSP Element

In this case, the two operands, AA and AB, are multiplied and the result is added with the previous accumulated value. This accumulated value is available at the output. The user can enable the input and pipeline registers, but the output register is always enabled. The output register is used to store the accumulated value. The ALU is configured as the accumulator in the sysDSP slice in the LatticeECP3 family can be initialized dynamically. A registered overflow signal is also available. The overflow conditions are provided later in this document. Figure 2-27 shows the MAC sysDSP element.

Figure 2-27. MAC DSP Element



MMAC DSP Element

The LatticeECP3 supports a MAC with two multipliers. This is called Multiply Multiply Accumulate or MMAC. In this case, the two operands, AA and AB, are multiplied and the result is added with the previous accumulated value and with the result of the multiplier operation of operands BA and BB. This accumulated value is available at the output. The user can enable the input and pipeline registers, but the output register is always enabled. The output register is used to store the accumulated value. The ALU is configured as the accumulator in the sysDSP slice. A registered overflow signal is also available. The overflow conditions are provided later in this document. Figure 2-28 shows the MMAC sysDSP element.

Figure 2-28. MMAC sysDSP Element

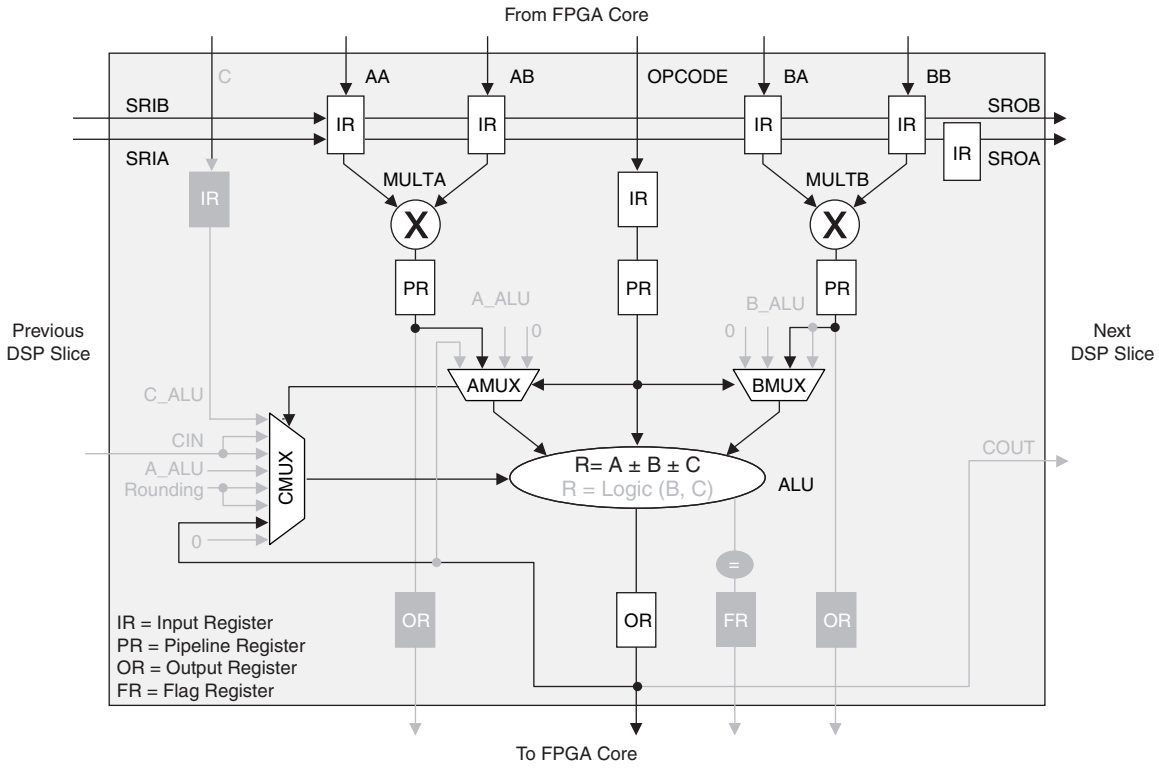
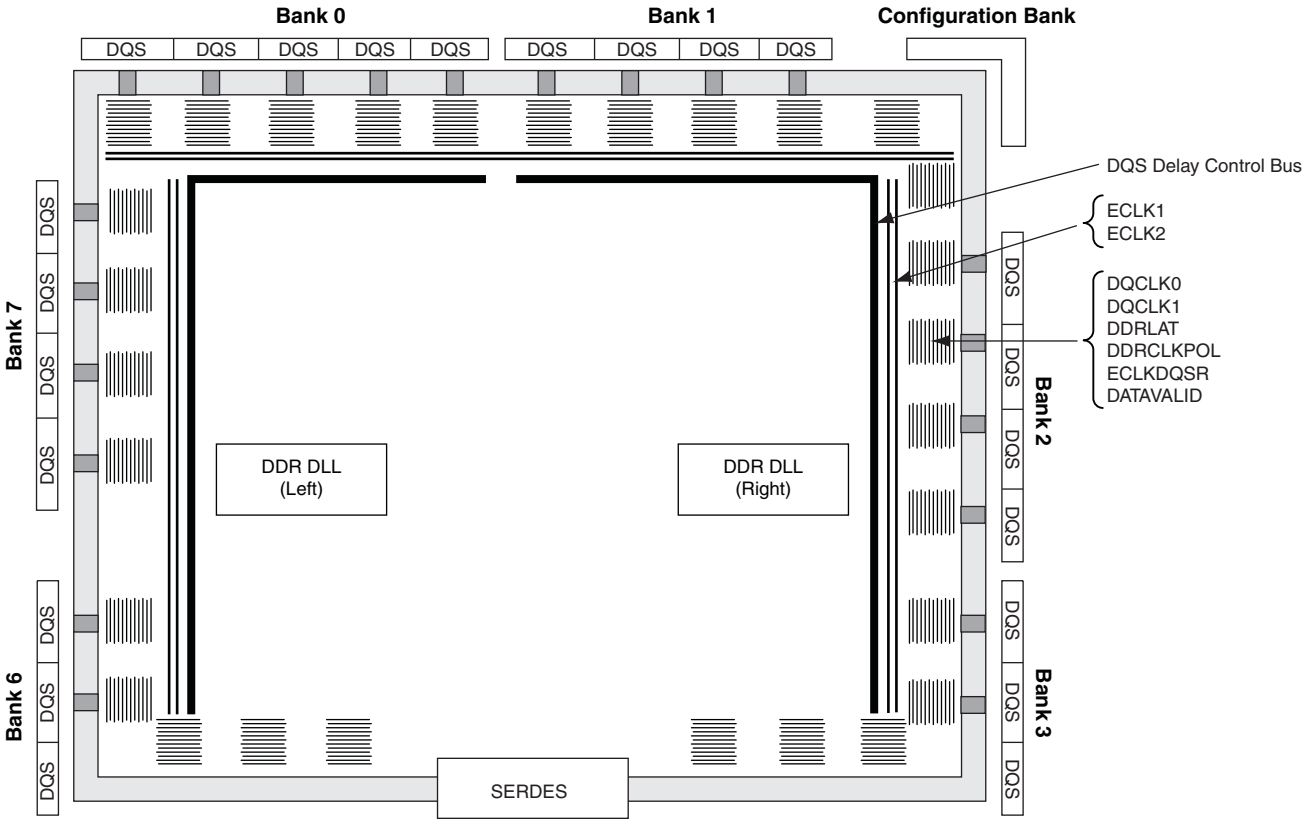


Figure 2-36. Edge Clock, DLL Calibration and DQS Local Bus Distribution



*Includes shared configuration I/Os and dedicated configuration I/Os.

Enhanced Configuration Options

LatticeECP3 devices have enhanced configuration features such as: decryption support, TransFR™ I/O and dual-boot image support.

1. **TransFR (Transparent Field Reconfiguration)**
   TransFR I/O (TFR) is a unique Lattice technology that allows users to update their logic in the field without interrupting system operation using a single ispVM command. TransFR I/O allows I/O states to be frozen during device configuration. This allows the device to be field updated with a minimum of system disruption and downtime. See TN1087, Minimizing System Interruption During Configuration Using TransFR Technology for details.

2. **Dual-Boot Image Support**
   Dual-boot images are supported for applications requiring reliable remote updates of configuration data for the system FPGA. After the system is running with a basic configuration, a new boot image can be downloaded remotely and stored in a separate location in the configuration storage device. Any time after the update the LatticeECP3 can be re-booted from this new configuration file. If there is a problem, such as corrupt data during download or incorrect version number with this new boot image, the LatticeECP3 device can revert back to the original backup golden configuration and try again. This all can be done without power cycling the system. For more information, please see TN1169, LatticeECP3 sysCONFIG Usage Guide.

Soft Error Detect (SED) Support

LatticeECP3 devices have dedicated logic to perform Cycle Redundancy Code (CRC) checks. During configuration, the configuration data bitstream can be checked with the CRC logic block. In addition, the LatticeECP3 device can also be programmed to utilize a Soft Error Detect (SED) mode that checks for soft errors in configuration SRAM. The SED operation can be run in the background during user mode. If a soft error occurs, during user mode (normal operation) the device can be programmed to generate an error signal.

For further information on SED support, please see TN1184, LatticeECP3 Soft Error Detection (SED) Usage Guide.

External Resistor

LatticeECP3 devices require a single external, 10 kOhm ±1% value between the XRES pin and ground. Device configuration will not be completed if this resistor is missing. There is no boundary scan register on the external resistor pad.

On-Chip Oscillator

Every LatticeECP3 device has an internal CMOS oscillator which is used to derive a Master Clock (MCCLK) for configuration. The oscillator and the MCCLK run continuously and are available to user logic after configuration is completed. The software default value of the MCCLK is nominally 2.5 MHz. Table 2-16 lists all the available MCCLK frequencies. When a different Master Clock is selected during the design process, the following sequence takes place:

1. Device powers up with a nominal Master Clock frequency of 3.1 MHz.
2. During configuration, users select a different master clock frequency.
3. The Master Clock frequency changes to the selected frequency once the clock configuration bits are received.
4. If the user does not select a master clock frequency, then the configuration bitstream defaults to the MCCLK frequency of 2.5 MHz.

This internal 130 MHz +/– 15% CMOS oscillator is available to the user by routing it as an input clock to the clock tree. For further information on the use of this oscillator for configuration or user mode, please see TN1169, LatticeECP3 sysCONFIG Usage Guide.

Table 2-16. Selectable Master Clock (MCCLK) Frequencies During Configuration (Nominal)

MCCLK (MHz)	MCCLK (MHz)
	10
2.5[1]	13
4.3	15[2]
5.4	20
6.9	26
8.1	33[3]
9.2	

1. Software default MCCLK frequency. Hardware default is 3.1 MHz.
2. Maximum MCCLK with encryption enabled.
3. Maximum MCCLK without encryption.

Density Shifting

The LatticeECP3 family is designed to ensure that different density devices in the same family and in the same package have the same pinout. Furthermore, the architecture ensures a high success rate when performing design migration from lower density devices to higher density devices. In many cases, it is also possible to shift a lower utilization design targeted for a high-density device to a lower density device. However, the exact details of the final resource utilization will impact the likelihood of success in each case. An example is that some user I/Os may become No Connects in smaller devices in the same package. Refer to the LatticeECP3 Pin Migration Tables and Diamond software for specific restrictions and limitations.

Hot Socketing Specifications[1, 2, 3]

Symbol	Parameter	Condition	Min.	Typ.	Max.	Units
IDK_HS[4]	Input or I/O Leakage Current	$0 \le V_{IN} \le V_{IH}$ (Max.)	—	—	+/–1	mA
IDK[5]	Input or I/O Leakage Current	$0 \le V_{IN} < V_{CCIO}$	—	—	+/–1	mA
		$V_{CCIO} \le V_{IN} \le V_{CCIO}$ + 0.5V	—	18	—	mA

1. V_{CC}, V_{CCAUX} and V_{CCIO} should rise/fall monotonically.
2. I_{DK} is additive to I_{PU}, I_{PD} or I_{BH}.
3. LVCMOS and LVTTL only.
4. Applicable to general purpose I/O pins located on the top and bottom sides of the device.
5. Applicable to general purpose I/O pins located on the left and right sides of the device.

Hot Socketing Requirements[1, 2]

Description	Min.	Typ.	Max.	Units
Input current per SERDES I/O pin when device is powered down and inputs driven.	—	—	8	mA

1. Assumes the device is powered down, all supplies grounded, both P and N inputs driven by CML driver with maximum allowed VCCOB (1.575 V), 8b10b data, internal AC coupling.
2. Each P and N input must have less than the specified maximum input current. For a 16-channel device, the total input current would be 8 mA*16 channels *2 input pins per channel = 256 mA

ESD Performance

Please refer to the LatticeECP3 Product Family Qualification Summary for complete qualification data, including ESD performance.

sysI/O Recommended Operating Conditions

Standard	V_{CCIO}			V_{REF} (V)		
	Min.	Typ.	Max.	Min.	Typ.	Max.
LVCMOS33[2]	3.135	3.3	3.465	—	—	—
LVCMOS33D	3.135	3.3	3.465	—	—	—
LVCMOS25[2]	2.375	2.5	2.625	—	—	—
LVCMOS18	1.71	1.8	1.89	—	—	—
LVCMOS15	1.425	1.5	1.575	—	—	—
LVCMOS12[2]	1.14	1.2	1.26	—	—	—
LVTTL33[2]	3.135	3.3	3.465	—	—	—
PCI33	3.135	3.3	3.465	—	—	—
SSTL15[3]	1.43	1.5	1.57	0.68	0.75	0.9
SSTL18_I, II[2]	1.71	1.8	1.89	0.833	0.9	0.969
SSTL25_I, II[2]	2.375	2.5	2.625	1.15	1.25	1.35
SSTL33_I, II[2]	3.135	3.3	3.465	1.3	1.5	1.7
HSTL15_I[2]	1.425	1.5	1.575	0.68	0.75	0.9
HSTL18_I, II[2]	1.71	1.8	1.89	0.816	0.9	1.08
LVDS25[2]	2.375	2.5	2.625	—	—	—
LVDS25E	2.375	2.5	2.625	—	—	—
MLVDS[1]	2.375	2.5	2.625	—	—	—
LVPECL33[1, 2]	3.135	3.3	3.465	—	—	—
Mini LVDS	2.375	2.5	2.625	—	—	—
BLVDS25[1, 2]	2.375	2.5	2.625	—	—	—
RSDS[2]	2.375	2.5	2.625	—	—	—
RSDSE[1, 2]	2.375	2.5	2.625	—	—	—
TRLVDS	3.14	3.3	3.47	—	—	—
PPLVDS	3.14/2.25	3.3/2.5	3.47/2.75	—	—	—
SSTL15D[3]	1.43	1.5	1.57	—	—	—
SSTL18D_I[2, 3], II[2, 3]	1.71	1.8	1.89	—	—	—
SSTL25D_ I[2], II[2]	2.375	2.5	2.625	—	—	—
SSTL33D_ I[2], II[2]	3.135	3.3	3.465	—	—	—
HSTL15D_ I[2]	1.425	1.5	1.575	—	—	—
HSTL18D_ I[2], II[2]	1.71	1.8	1.89	—	—	—

1. Inputs on chip. Outputs are implemented with the addition of external resistors.
2. For input voltage compatibility, see TN1177, LatticeECP3 sysIO Usage Guide.
3. VREF is required when using Differential SSTL to interface to DDR memory.

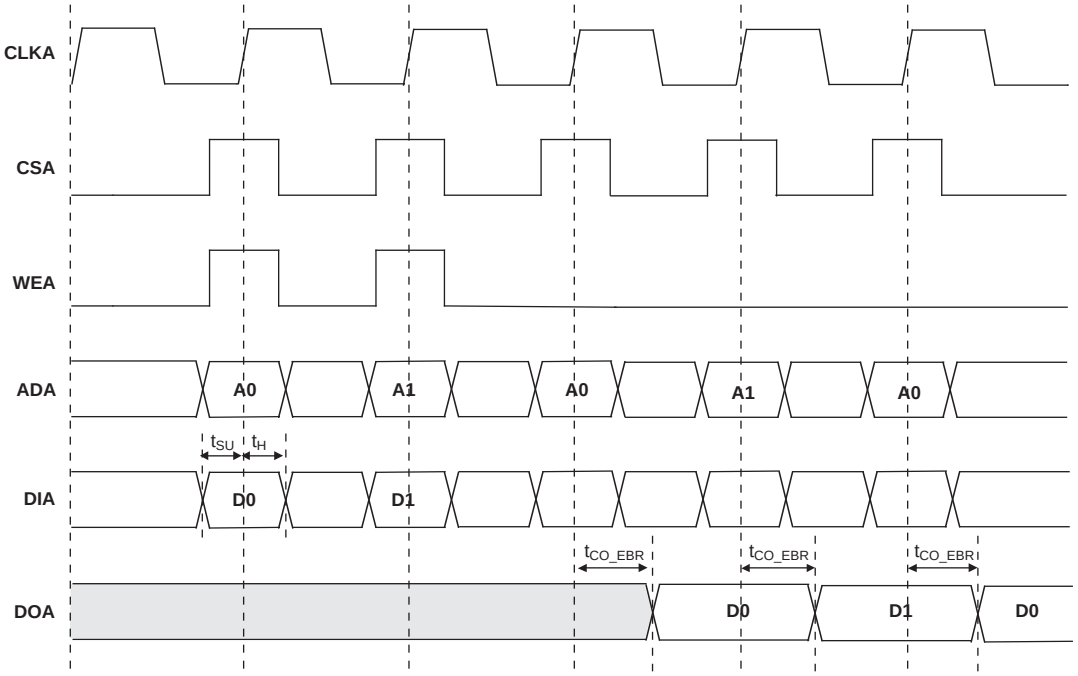
LatticeECP3 Internal Switching Characteristics[1, 2, 5]

Over Recommended Commercial Operating Conditions

Parameter	Description	–8		–7		–6		Units.
		Min.	Max.	Min.	Max.	Min.	Max.	
PFU/PFF Logic Mode Timing								
t_{LUT4_PFU}	LUT4 delay (A to D inputs to F output)	—	0.147	—	0.163	—	0.179	ns
t_{LUT6_PFU}	LUT6 delay (A to D inputs to OFX output)	—	0.281	—	0.335	—	0.379	ns
t_{LSR_PFU}	Set/Reset to output of PFU (Asynchronous)	—	0.593	—	0.674	—	0.756	ns
t_{LSRREC_PFU}	Asynchronous Set/Reset recovery time for PFU Logic		0.298		0.345		0.391	ns
t_{SUM_PFU}	Clock to Mux (M0,M1) Input Setup Time	0.134	—	0.144	—	0.153	—	ns
t_{HM_PFU}	Clock to Mux (M0,M1) Input Hold Time	–0.097	—	–0.103	—	–0.109	—	ns
t_{SUD_PFU}	Clock to D input setup time	0.061	—	0.068	—	0.075	—	ns
t_{HD_PFU}	Clock to D input hold time	0.019	—	0.013	—	0.015	—	ns
t_{CK2Q_PFU}	Clock to Q delay, (D-type Register Configuration)	—	0.243	—	0.273	—	0.303	ns
PFU Dual Port Memory Mode Timing								
t_{CORAM_PFU}	Clock to Output (F Port)	—	0.710	—	0.803	—	0.897	ns
t_{SUDATA_PFU}	Data Setup Time	–0.137	—	–0.155	—	–0.174	—	ns
t_{HDATA_PFU}	Data Hold Time	0.188	—	0.217	—	0.246	—	ns
t_{SUADDR_PFU}	Address Setup Time	–0.227	—	–0.257	—	–0.286	—	ns
t_{HADDR_PFU}	Address Hold Time	0.240	—	0.275	—	0.310	—	ns
t_{SUWREN_PFU}	Write/Read Enable Setup Time	–0.055	—	–0.055	—	–0.063	—	ns
t_{HWREN_PFU}	Write/Read Enable Hold Time	0.059	—	0.059	—	0.071	—	ns
PIC Timing								
PIO Input/Output Buffer Timing								
t_{IN_PIO}	Input Buffer Delay (LVCMOS25)	—	0.423	—	0.466	—	0.508	ns
t_{OUT_PIO}	Output Buffer Delay (LVCMOS25)	—	1.241	—	1.301	—	1.361	ns
IOLOGIC Input/Output Timing								
t_{SUI_PIO}	Input Register Setup Time (Data Before Clock)	0.956	—	1.124	—	1.293	—	ns
t_{HI_PIO}	Input Register Hold Time (Data after Clock)	0.225	—	0.184	—	0.240	—	ns
t_{COO_PIO}	Output Register Clock to Output Delay[4]	-	1.09	-	1.16	-	1.23	ns
t_{SUCE_PIO}	Input Register Clock Enable Setup Time	0.220	—	0.185	—	0.150	—	ns
t_{HCE_PIO}	Input Register Clock Enable Hold Time	–0.085	—	–0.072	—	–0.058	—	ns
t_{SULSR_PIO}	Set/Reset Setup Time	0.117	—	0.103	—	0.088	—	ns
t_{HLSR_PIO}	Set/Reset Hold Time	–0.107	—	–0.094	—	–0.081	—	ns
EBR Timing								
t_{CO_EBR}	Clock (Read) to output from Address or Data	—	2.78	—	2.89	—	2.99	ns
t_{COO_EBR}	Clock (Write) to output from EBR output Register	—	0.31	—	0.32	—	0.33	ns
t_{SUDATA_EBR}	Setup Data to EBR Memory	–0.218	—	–0.227	—	–0.237	—	ns
t_{HDATA_EBR}	Hold Data to EBR Memory	0.249	—	0.257	—	0.265	—	ns
t_{SUADDR_EBR}	Setup Address to EBR Memory	–0.071	—	–0.070	—	–0.068	—	ns
t_{HADDR_EBR}	Hold Address to EBR Memory	0.118	—	0.098	—	0.077	—	ns
t_{SUWREN_EBR}	Setup Write/Read Enable to EBR Memory	–0.107	—	–0.106	—	–0.106	—	ns

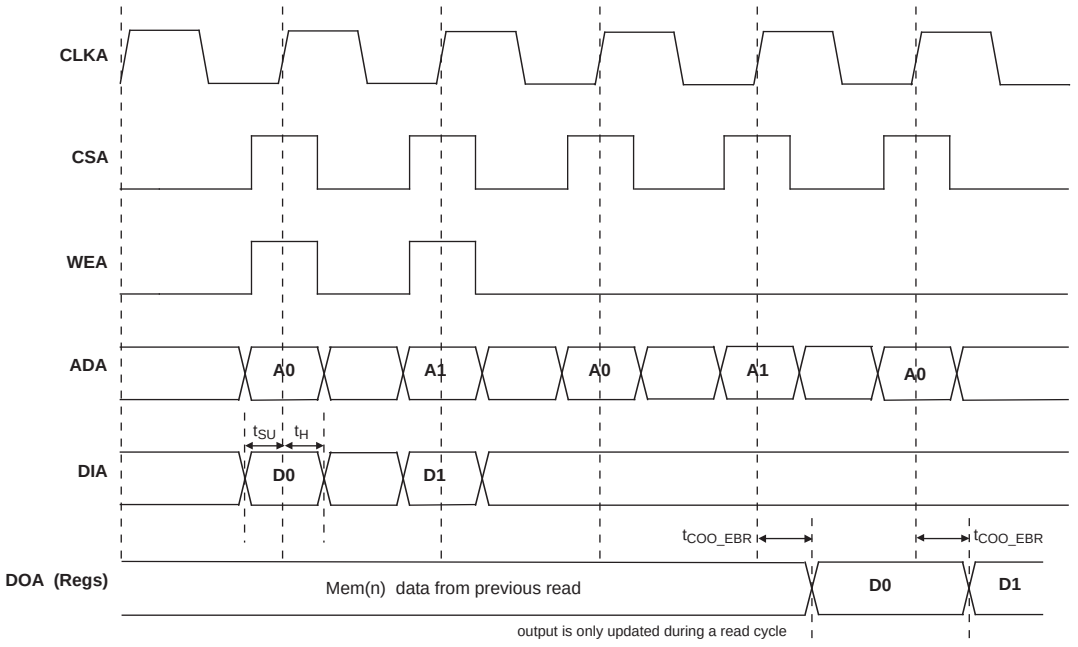
Timing Diagrams

Figure 3-9. Read/Write Mode (Normal)



Note: Input data and address are registered at the positive edge of the clock and output data appears after the positive edge of the clock.

Figure 3-10. Read/Write Mode with Input and Output Registers



LatticeECP3 Family Timing Adders[1, 2, 3, 4, 5, 7]

Over Recommended Commercial Operating Conditions

Buffer Type	Description	–8	–7	–6	Units
Input Adjusters					
LVDS25E	LVDS, Emulated, VCCIO = 2.5 V	0.03	–0.01	–0.03	ns
LVDS25	LVDS, VCCIO = 2.5 V	0.03	0.00	–0.04	ns
BLVDS25	BLVDS, Emulated, VCCIO = 2.5 V	0.03	0.00	–0.04	ns
MLVDS25	MLVDS, Emulated, VCCIO = 2.5 V	0.03	0.00	–0.04	ns
RSDS25	RSDS, VCCIO = 2.5 V	0.03	–0.01	–0.03	ns
PPLVDS	Point-to-Point LVDS	0.03	–0.01	–0.03	ns
TRLVDS	Transition-Reduced LVDS	0.03	0.00	–0.04	ns
Mini MLVDS	Mini LVDS	0.03	–0.01	–0.03	ns
LVPECL33	LVPECL, Emulated, VCCIO = 3.3 V	0.17	0.23	0.28	ns
HSTL18_I	HSTL_18 class I, VCCIO = 1.8 V	0.20	0.17	0.13	ns
HSTL18_II	HSTL_18 class II, VCCIO = 1.8 V	0.20	0.17	0.13	ns
HSTL18D_I	Differential HSTL 18 class I	0.20	0.17	0.13	ns
HSTL18D_II	Differential HSTL 18 class II	0.20	0.17	0.13	ns
HSTL15_I	HSTL_15 class I, VCCIO = 1.5 V	0.10	0.12	0.13	ns
HSTL15D_I	Differential HSTL 15 class I	0.10	0.12	0.13	ns
SSTL33_I	SSTL_3 class I, VCCIO = 3.3 V	0.17	0.23	0.28	ns
SSTL33_II	SSTL_3 class II, VCCIO = 3.3 V	0.17	0.23	0.28	ns
SSTL33D_I	Differential SSTL_3 class I	0.17	0.23	0.28	ns
SSTL33D_II	Differential SSTL_3 class II	0.17	0.23	0.28	ns
SSTL25_I	SSTL_2 class I, VCCIO = 2.5 V	0.12	0.14	0.16	ns
SSTL25_II	SSTL_2 class II, VCCIO = 2.5 V	0.12	0.14	0.16	ns
SSTL25D_I	Differential SSTL_2 class I	0.12	0.14	0.16	ns
SSTL25D_II	Differential SSTL_2 class II	0.12	0.14	0.16	ns
SSTL18_I	SSTL_18 class I, VCCIO = 1.8 V	0.08	0.06	0.04	ns
SSTL18_II	SSTL_18 class II, VCCIO = 1.8 V	0.08	0.06	0.04	ns
SSTL18D_I	Differential SSTL_18 class I	0.08	0.06	0.04	ns
SSTL18D_II	Differential SSTL_18 class II	0.08	0.06	0.04	ns
SSTL15	SSTL_15, VCCIO = 1.5 V	0.087	0.059	0.032	ns
SSTL15D	Differential SSTL_15	0.087	0.059	0.032	ns
LVTTL33	LVTTL, VCCIO = 3.3 V	0.07	0.07	0.07	ns
LVCMOS33	LVCMOS, VCCIO = 3.3 V	0.07	0.07	0.07	ns
LVCMOS25	LVCMOS, VCCIO = 2.5 V	0.00	0.00	0.00	ns
LVCMOS18	LVCMOS, VCCIO = 1.8 V	–0.13	–0.13	–0.13	ns
LVCMOS15	LVCMOS, VCCIO = 1.5 V	–0.07	–0.07	–0.07	ns
LVCMOS12	LVCMOS, VCCIO = 1.2 V	–0.20	–0.19	–0.19	ns
PCI33	PCI, VCCIO = 3.3 V	0.07	0.07	0.07	ns
Output Adjusters					
LVDS25E	LVDS, Emulated, VCCIO = 2.5 V	1.02	1.14	1.26	ns
LVDS25	LVDS, VCCIO = 2.5 V	–0.11	–0.07	–0.03	ns
BLVDS25	BLVDS, Emulated, VCCIO = 2.5 V	1.01	1.13	1.25	ns
MLVDS25	MLVDS, Emulated, VCCIO = 2.5 V	1.01	1.13	1.25	ns

Gigabit Ethernet/Serial Rapid I/O Type 1/SGMII/CPRI LV E.12 Electrical and Timing Characteristics

AC and DC Characteristics

Table 3-17. Transmit

Symbol	Description	Test Conditions	Min.	Typ.	Max.	Units
T_{RF}	Differential rise/fall time	20%-80%	—	80	—	ps
$Z_{TX\_DIFF\_DC}$	Differential impedance		80	100	120	Ohms
J_{TX_DDJ}[3, 4, 5]	Output data deterministic jitter		—	—	0.10	UI
J_{TX_TJ}[2, 3, 4, 5]	Total output data jitter		—	—	0.24	UI

1. Rise and fall times measured with board trace, connector and approximately 2.5 pf load.
2. Total jitter includes both deterministic jitter and random jitter. The random jitter is the total jitter minus the actual deterministic jitter.
3. Jitter values are measured with each CML output AC coupled into a 50-Ohm impedance (100-Ohm differential impedance).
4. Jitter and skew are specified between differential crossings of the 50% threshold of the reference signal.
5. Values are measured at 1.25 Gbps.

Table 3-18. Receive and Jitter Tolerance

Symbol	Description	Test Conditions	Min.	Typ.	Max.	Units
RL_{RX_DIFF}	Differential return loss	From 100 MHz to 1.25 GHz	10	—	—	dB
RL_{RX_CM}	Common mode return loss	From 100 MHz to 1.25 GHz	6	—	—	dB
Z_{RX_DIFF}	Differential termination resistance		80	100	120	Ohms
J_{RX_DJ}[1, 2, 3, 4, 5]	Deterministic jitter tolerance (peak-to-peak)		—	—	0.34	UI
J_{RX_RJ}[1, 2, 3, 4, 5]	Random jitter tolerance (peak-to-peak)		—	—	0.26	UI
J_{RX_SJ}[1, 2, 3, 4, 5]	Sinusoidal jitter tolerance (peak-to-peak)		—	—	0.11	UI
J_{RX_TJ}[1, 2, 3, 4, 5]	Total jitter tolerance (peak-to-peak)		—	—	0.71	UI
T_{RX_EYE}	Receiver eye opening		0.29	—	—	UI

1. Total jitter includes deterministic jitter, random jitter and sinusoidal jitter. The sinusoidal jitter tolerance mask is shown in Figure 3-18.
2. Jitter values are measured with each high-speed input AC coupled into a 50-Ohm impedance.
3. Jitter and skew are specified between differential crossings of the 50% threshold of the reference signal.
4. Jitter tolerance, Differential Input Sensitivity and Receiver Eye Opening parameters are characterized when Full Rx Equalization is enabled.
5. Values are measured at 1.25 Gbps.

LatticeECP3 sysCONFIG Port Timing Specifications

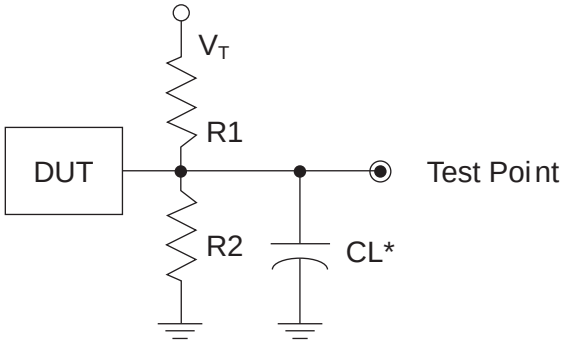
Over Recommended Operating Conditions

Parameter	Description		Min.	Max.	Units
POR, Configuration Initialization, and Wakeup					
t_{ICFG}	Time from the Application of V_{CC}, V_{CCAUX} or V_{CCIO8}* (Whichever is the Last to Cross the POR Trip Point) to the Rising Edge of INITN	Master mode	—	23	ms
		Slave mode	—	6	ms
t_{VMC}	Time from t_{ICFG} to the Valid Master MCLK		—	5	µs
t_{PRGM}	PROGRAMN Low Time to Start Configuration		25	—	ns
t_{PRGMRJ}	PROGRAMN Pin Pulse Rejection		—	10	ns
$t_{DPPINIT}$	Delay Time from PROGRAMN Low to INITN Low		—	37	ns
$t_{DPPDONE}$	Delay Time from PROGRAMN Low to DONE Low		—	37	ns
t_{DINIT}[1]	PROGRAMN High to INITN High Delay		—	1	ms
t_{MWC}	Additional Wake Master Clock Signals After DONE Pin is High		100	500	cycles
t_{CZ}	MCLK From Active To Low To High-Z		—	300	ns
t_{IODISS}	User I/O Disable from PROGRAMN Low		—	100	ns
t_{IOENSS}	User I/O Enabled Time from CCLK Edge During Wake-up Sequence		—	100	ns
All Configuration Modes					
t_{SUCDI}	Data Setup Time to CCLK/MCLK		5	—	ns
t_{HCDI}	Data Hold Time to CCLK/MCLK		1	—	ns
t_{CODO}	CCLK/MCLK to DOUT in Flowthrough Mode		-0.2	12	ns
Slave Serial					
t_{SSCH}	CCLK Minimum High Pulse		5	—	ns
t_{SSCL}	CCLK Minimum Low Pulse		5	—	ns
f_{CCLK}	CCLK Frequency	Without encryption	—	33	MHz
		With encryption	—	20	MHz
Master and Slave Parallel					
t_{SUCS}	CSN[1:0] Setup Time to CCLK/MCLK		7	—	ns
t_{HCS}	CSN[1:0] Hold Time to CCLK/MCLK		1	—	ns
t_{SUWD}	WRITEN Setup Time to CCLK/MCLK		7	—	ns
t_{HWD}	WRITEN Hold Time to CCLK/MCLK		1	—	ns
t_{DCB}	CCLK/MCLK to BUSY Delay Time		—	12	ns
t_{CORD}	CCLK to Out for Read Data		—	12	ns
t_{BSCH}	CCLK Minimum High Pulse		6	—	ns
t_{BSCL}	CCLK Minimum Low Pulse		6	—	ns
t_{BSCYC}	Byte Slave Cycle Time		30	—	ns
f_{CCLK}	CCLK/MCLK Frequency	Without encryption	—	33	MHz
		With encryption	—	20	MHz
Master and Slave SPI					
t_{CFGX}	INITN High to MCLK Low		—	80	ns
t_{CSSPI}	INITN High to CSSPIN Low		0.2	2	µs
t_{SOCDO}	MCLK Low to Output Valid		—	15	ns
t_{CSPID}	CSSPIN[0:1] Low to First MCLK Edge Setup Time		0.3		µs
f_{CCLK}	CCLK Frequency	Without encryption	—	33	MHz
		With encryption	—	20	MHz
t_{SSCH}	CCLK Minimum High Pulse		5	—	ns

Switching Test Conditions

Figure 3-33 shows the output test load that is used for AC testing. The specific values for resistance, capacitance, voltage, and other test conditions are shown in Table 3-23.

Figure 3-33. Output Test Load, LVTTL and LVCMOS Standards



*CL Includes Test Fixture and Probe Capacitance

Table 3-23. Test Fixture Required Components, Non-Terminated Interfaces

Test Condition	R_1	R_2	C_L	Timing Ref.	V_T
LVTTL and other LVCMOS settings (L -> H, H -> L)	∞	∞	0 pF	LVCMOS 3.3 = 1.5V	—
				LVCMOS 2.5 = $V_{CCIO}/2$	—
				LVCMOS 1.8 = $V_{CCIO}/2$	—
				LVCMOS 1.5 = $V_{CCIO}/2$	—
				LVCMOS 1.2 = $V_{CCIO}/2$	—
LVCMOS 2.5 I/O (Z -> H)	∞	1MΩ	0 pF	$V_{CCIO}/2$	—
LVCMOS 2.5 I/O (Z -> L)	1 MΩ	∞	0 pF	$V_{CCIO}/2$	V_{CCIO}
LVCMOS 2.5 I/O (H -> Z)	∞	100	0 pF	V_{OH} - 0.10	—
LVCMOS 2.5 I/O (L -> Z)	100	∞	0 pF	V_{OL} + 0.10	V_{CCIO}

Note: Output test conditions for all other interfaces are determined by the respective standards.

Part Number	Voltage	Grade[1]	Power	Package	Pins	Temp.	LUTs (K)
LFE3-70EA-6FN484C	1.2 V	–6	STD	Lead-Free fpBGA	484	COM	67
LFE3-70EA-7FN484C	1.2 V	–7	STD	Lead-Free fpBGA	484	COM	67
LFE3-70EA-8FN484C	1.2 V	–8	STD	Lead-Free fpBGA	484	COM	67
LFE3-70EA-6LFN484C	1.2 V	–6	LOW	Lead-Free fpBGA	484	COM	67
LFE3-70EA-7LFN484C	1.2 V	–7	LOW	Lead-Free fpBGA	484	COM	67
LFE3-70EA-8LFN484C	1.2 V	–8	LOW	Lead-Free fpBGA	484	COM	67
LFE3-70EA-6FN672C	1.2 V	–6	STD	Lead-Free fpBGA	672	COM	67
LFE3-70EA-7FN672C	1.2 V	–7	STD	Lead-Free fpBGA	672	COM	67
LFE3-70EA-8FN672C	1.2 V	–8	STD	Lead-Free fpBGA	672	COM	67
LFE3-70EA-6LFN672C	1.2 V	–6	LOW	Lead-Free fpBGA	672	COM	67
LFE3-70EA-7LFN672C	1.2 V	–7	LOW	Lead-Free fpBGA	672	COM	67
LFE3-70EA-8LFN672C	1.2 V	–8	LOW	Lead-Free fpBGA	672	COM	67
LFE3-70EA-6FN1156C	1.2 V	–6	STD	Lead-Free fpBGA	1156	COM	67
LFE3-70EA-7FN1156C	1.2 V	–7	STD	Lead-Free fpBGA	1156	COM	67
LFE3-70EA-8FN1156C	1.2 V	–8	STD	Lead-Free fpBGA	1156	COM	67
LFE3-70EA-6LFN1156C	1.2 V	–6	LOW	Lead-Free fpBGA	1156	COM	67
LFE3-70EA-7LFN1156C	1.2 V	–7	LOW	Lead-Free fpBGA	1156	COM	67
LFE3-70EA-8LFN1156C	1.2 V	–8	LOW	Lead-Free fpBGA	1156	COM	67

1. For ordering information on -9 speed grade devices, please contact your Lattice Sales Representative.

Part Number	Voltage	Grade[1]	Power	Package	Pins	Temp.	LUTs (K)
LFE3-95EA-6FN484C	1.2 V	–6	STD	Lead-Free fpBGA	484	COM	92
LFE3-95EA-7FN484C	1.2 V	–7	STD	Lead-Free fpBGA	484	COM	92
LFE3-95EA-8FN484C	1.2 V	–8	STD	Lead-Free fpBGA	484	COM	92
LFE3-95EA-6LFN484C	1.2 V	–6	LOW	Lead-Free fpBGA	484	COM	92
LFE3-95EA-7LFN484C	1.2 V	–7	LOW	Lead-Free fpBGA	484	COM	92
LFE3-95EA-8LFN484C	1.2 V	–8	LOW	Lead-Free fpBGA	484	COM	92
LFE3-95EA-6FN672C	1.2 V	–6	STD	Lead-Free fpBGA	672	COM	92
LFE3-95EA-7FN672C	1.2 V	–7	STD	Lead-Free fpBGA	672	COM	92
LFE3-95EA-8FN672C	1.2 V	–8	STD	Lead-Free fpBGA	672	COM	92
LFE3-95EA-6LFN672C	1.2 V	–6	LOW	Lead-Free fpBGA	672	COM	92
LFE3-95EA-7LFN672C	1.2 V	–7	LOW	Lead-Free fpBGA	672	COM	92
LFE3-95EA-8LFN672C	1.2 V	–8	LOW	Lead-Free fpBGA	672	COM	92
LFE3-95EA-6FN1156C	1.2 V	–6	STD	Lead-Free fpBGA	1156	COM	92
LFE3-95EA-7FN1156C	1.2 V	–7	STD	Lead-Free fpBGA	1156	COM	92
LFE3-95EA-8FN1156C	1.2 V	–8	STD	Lead-Free fpBGA	1156	COM	92
LFE3-95EA-6LFN1156C	1.2 V	–6	LOW	Lead-Free fpBGA	1156	COM	92
LFE3-95EA-7LFN1156C	1.2 V	–7	LOW	Lead-Free fpBGA	1156	COM	92
LFE3-95EA-8LFN1156C	1.2 V	–8	LOW	Lead-Free fpBGA	1156	COM	92

1. For ordering information on -9 speed grade devices, please contact your Lattice Sales Representative.