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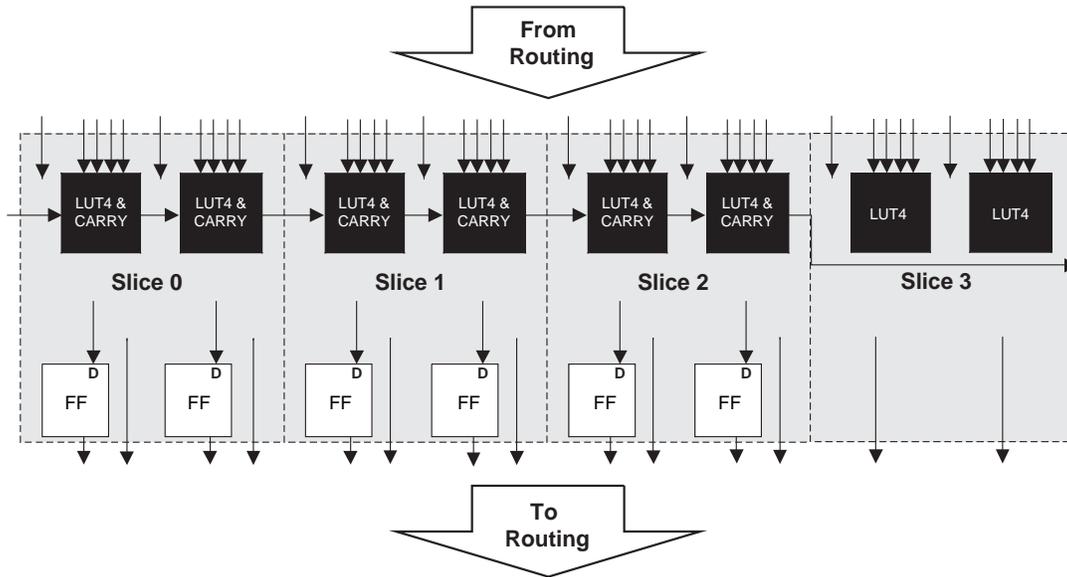
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The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details

Product Status	Not For New Designs
Number of LABs/CLBs	11500
Number of Logic Elements/Cells	92000
Total RAM Bits	4526080
Number of I/O	380
Number of Gates	-
Voltage - Supply	1.14V ~ 1.26V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	672-BBGA
Supplier Device Package	672-FPBGA (27x27)
Purchase URL	https://www.e-xfl.com/product-detail/lattice-semiconductor/lfe3-95ea-9fn672c

Figure 2-2. PFU Diagram



Slice

Slice 0 through Slice 2 contain two LUT4s feeding two registers, whereas Slice 3 contains two LUT4s only. For PFUs, Slice 0 through Slice 2 can be configured as distributed memory, a capability not available in the PFF. Table 2-1 shows the capability of the slices in both PFF and PFU blocks along with the operation modes they enable. In addition, each PFU contains logic that allows the LUTs to be combined to perform functions such as LUT5, LUT6, LUT7 and LUT8. There is control logic to perform set/reset functions (programmable as synchronous/asynchronous), clock select, chip-select and wider RAM/ROM functions.

Table 2-1. Resources and Modes Available per Slice

Slice	PFU BLock		PFF Block	
	Resources	Modes	Resources	Modes
Slice 0	2 LUT4s and 2 Registers	Logic, Ripple, RAM, ROM	2 LUT4s and 2 Registers	Logic, Ripple, ROM
Slice 1	2 LUT4s and 2 Registers	Logic, Ripple, RAM, ROM	2 LUT4s and 2 Registers	Logic, Ripple, ROM
Slice 2	2 LUT4s and 2 Registers	Logic, Ripple, RAM, ROM	2 LUT4s and 2 Registers	Logic, Ripple, ROM
Slice 3	2 LUT4s	Logic, ROM	2 LUT4s	Logic, ROM

Figure 2-3 shows an overview of the internal logic of the slice. The registers in the slice can be configured for positive/negative and edge triggered or level sensitive clocks.

Slices 0, 1 and 2 have 14 input signals: 13 signals from routing and one from the carry-chain (from the adjacent slice or PFU). There are seven outputs: six to routing and one to carry-chain (to the adjacent PFU). Slice 3 has 10 input signals from routing and four signals to routing. Table 2-2 lists the signals associated with Slice 0 to Slice 2.

Figure 2-10. Primary Clock Sources for LatticeECP3-35

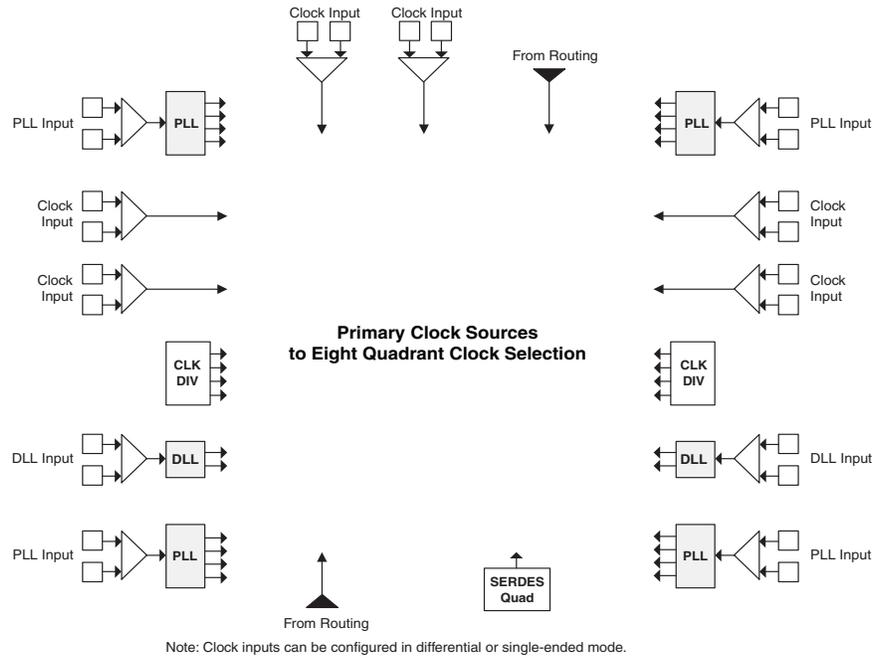


Figure 2-11. Primary Clock Sources for LatticeECP3-70, -95, -150

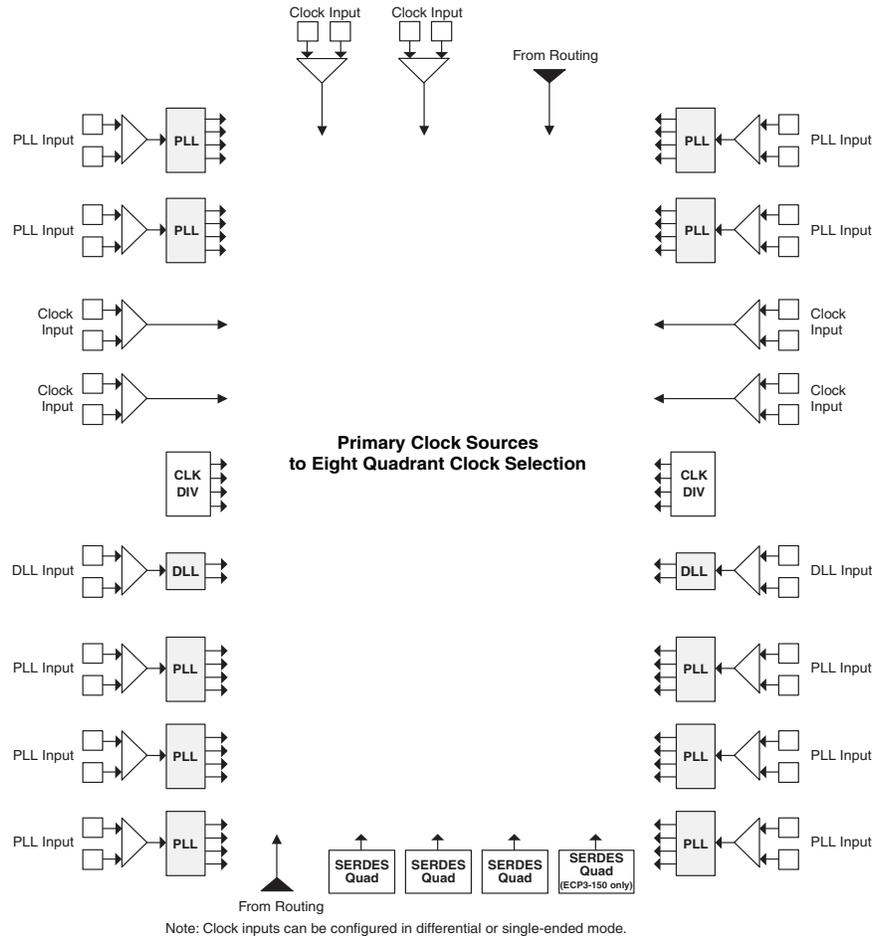
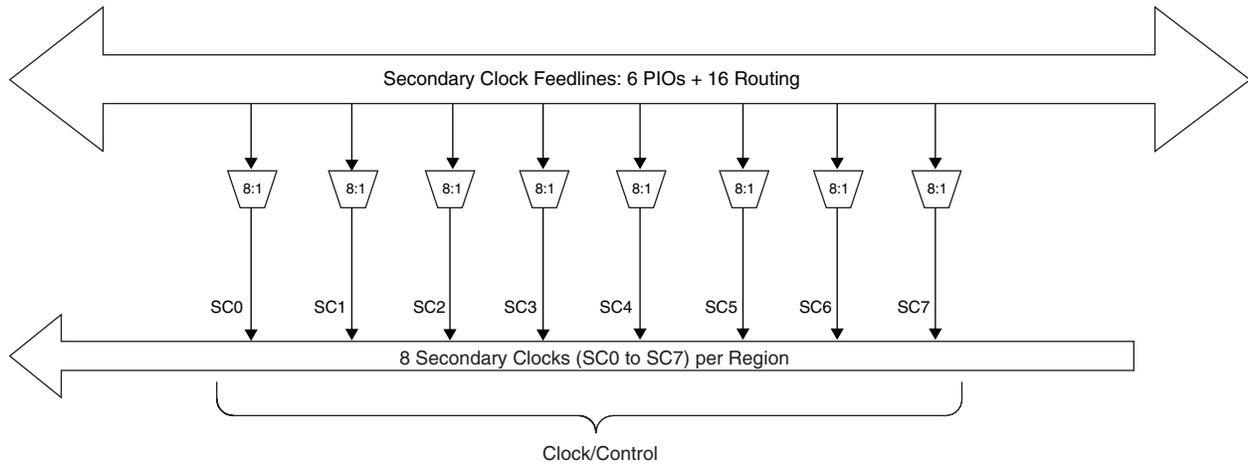


Figure 2-16. Per Region Secondary Clock Selection



Slice Clock Selection

Figure 2-17 shows the clock selections and Figure 2-18 shows the control selections for Slice0 through Slice2. All the primary clocks and seven secondary clocks are routed to this clock selection mux. Other signals can be used as a clock input to the slices via routing. Slice controls are generated from the secondary clocks/controls or other signals connected via routing.

If none of the signals are selected for both clock and control then the default value of the mux output is 1. Slice 3 does not have any registers; therefore it does not have the clock or control muxes.

Figure 2-17. Slice0 through Slice2 Clock Selection

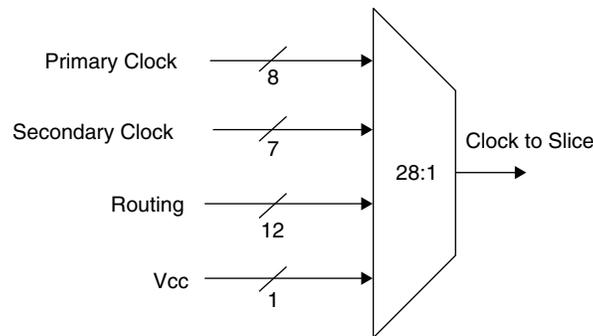
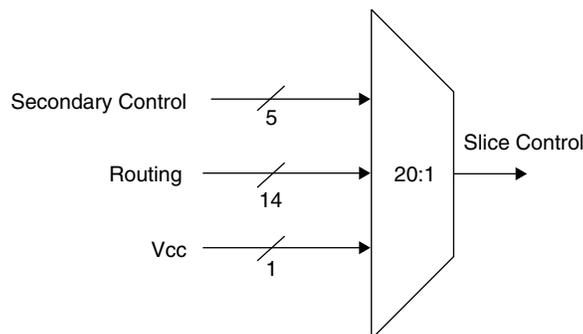


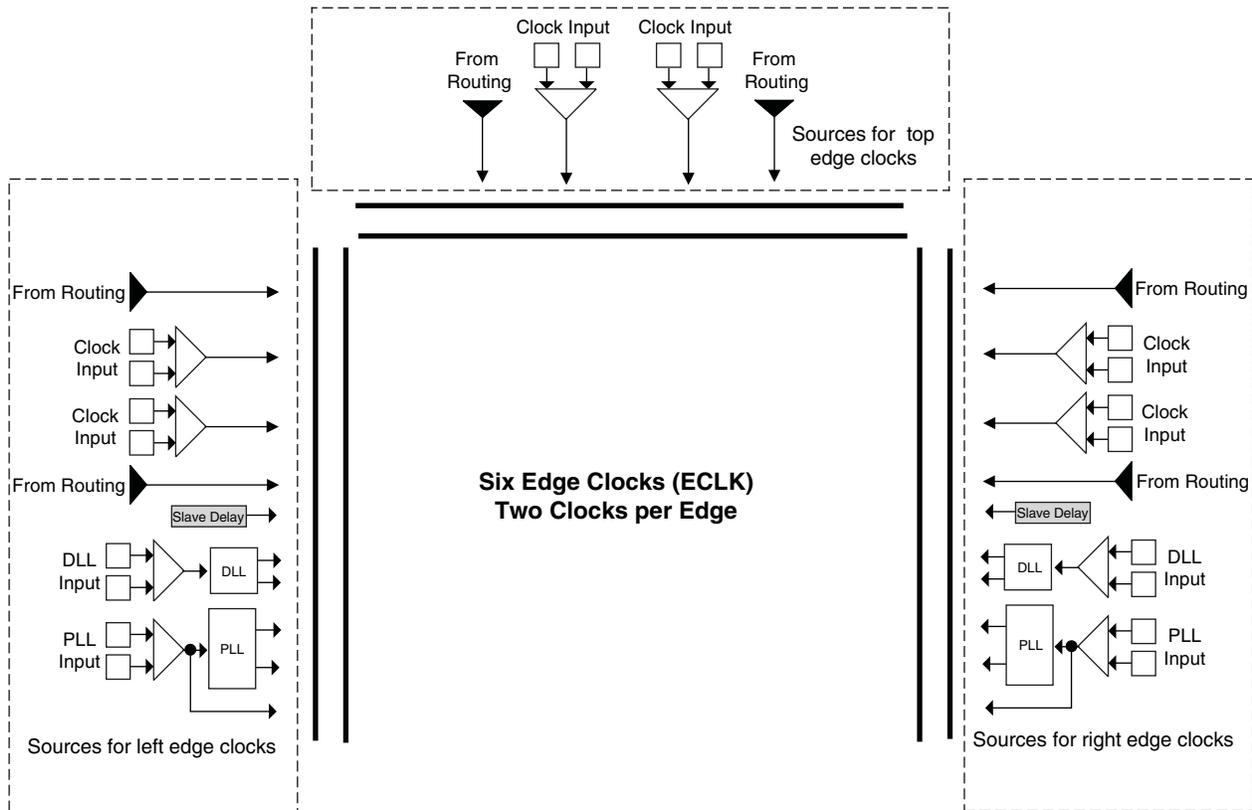
Figure 2-18. Slice0 through Slice2 Control Selection



Edge Clock Sources

Edge clock resources can be driven from a variety of sources at the same edge. Edge clock resources can be driven from adjacent edge clock PIOs, primary clock PIOs, PLLs, DLLs, Slave Delay and clock dividers as shown in Figure 2-19.

Figure 2-19. Edge Clock Sources



Notes:

1. Clock inputs can be configured in differential or single ended mode.
2. The two DLLs can also drive the two top edge clocks.
3. The top left and top right PLL can also drive the two top edge clocks.

Edge Clock Routing

LatticeECP3 devices have a number of high-speed edge clocks that are intended for use with the PIOs in the implementation of high-speed interfaces. There are six edge clocks per device: two edge clocks on each of the top, left, and right edges. Different PLL and DLL outputs are routed to the two muxes on the left and right sides of the device. In addition, the CLKINDEL signal (generated from the DLL Slave Delay Line block) is routed to all the edge clock muxes on the left and right sides of the device. Figure 2-20 shows the selection muxes for these clocks.

Figure 2-20. Sources of Edge Clock (Left and Right Edges)

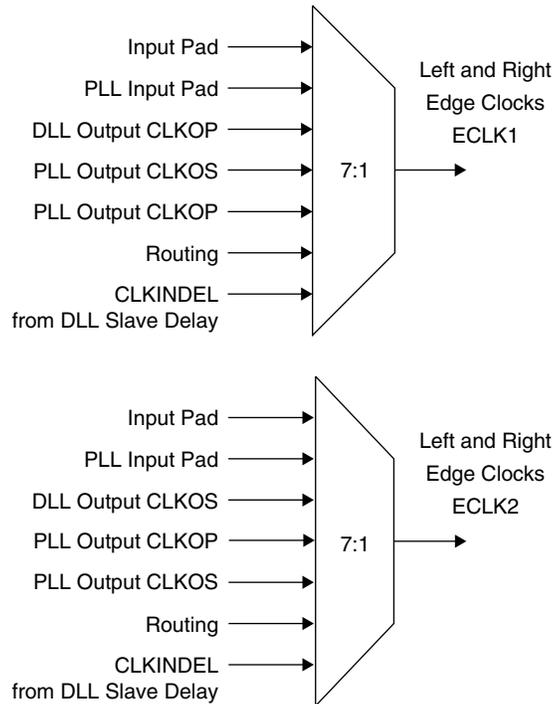
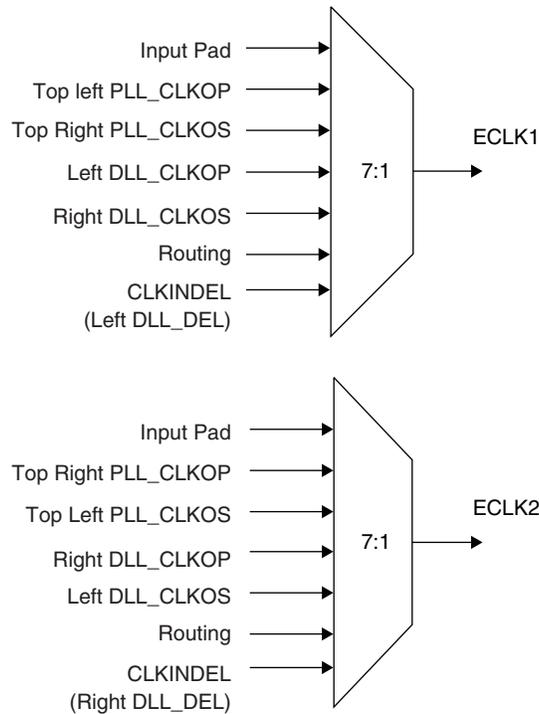


Figure 2-21. Sources of Edge Clock (Top Edge)



The edge clocks have low injection delay and low skew. They are used to clock the I/O registers and thus are ideal for creating I/O interfaces with a single clock signal and a wide data bus. They are also used for DDR Memory or Generic DDR interfaces.

Two adjacent PIOs can be joined to provide a differential I/O pair (labeled as “T” and “C”) as shown in Figure 2-32. The PAD Labels “T” and “C” distinguish the two PIOs. Approximately 50% of the PIO pairs on the left and right edges of the device can be configured as true LVDS outputs. All I/O pairs can operate as LVDS inputs.

Table 2-11. PIO Signal List

Name	Type	Description
INDD	Input Data	Register bypassed input. This is not the same port as INCK.
IPA, INA, IPB, INB	Input Data	Ports to core for input data
OPOSA, ONEGA ¹ , OPOSB, ONEGB ¹	Output Data	Output signals from core. An exception is the ONEGB port, used for tristate logic at the DQS pad.
CE	PIO Control	Clock enables for input and output block flip-flops.
SCLK	PIO Control	System Clock (PCLK) for input and output/TS blocks. Connected from clock ISB.
LSR	PIO Control	Local Set/Reset
ECLK1, ECLK2	PIO Control	Edge clock sources. Entire PIO selects one of two sources using mux.
ECLKDQSR ¹	Read Control	From DQS_STROBE, shifted strobe for memory interfaces only.
DDRCLKPOL ¹	Read Control	Ensures transfer from DQS domain to SCLK domain.
DDRLAT ¹	Read Control	Used to guarantee INDDR2 gearing by selectively enabling a D-Flip-Flop in datapath.
DEL[3:0]	Read Control	Dynamic input delay control bits.
INCK	To Clock Distribution and PLL	PIO treated as clock PIO, path to distribute to primary clocks and PLL.
TS	Tristate Data	Tristate signal from core (SDR)
DQCLK0 ¹ , DQCLK1 ¹	Write Control	Two clocks edges, 90 degrees out of phase, used in output gearing.
DQSW ²	Write Control	Used for output and tristate logic at DQS only.
DYNDEL[7:0]	Write Control	Shifting of write clocks for specific DQS group, using 6:0 each step is approximately 25ps, 128 steps. Bit 7 is an invert (timing depends on input frequency). There is also a static control for this 8-bit setting, enabled with a memory cell.
DCNTL[6:0]	PIO Control	Original delay code from DDR DLL
DATAVALID ¹	Output Data	Status flag from DATAVALID logic, used to indicate when input data is captured in IOLOGIC and valid to core.
READ	For DQS_Strobe	Read signal for DDR memory interface
DQSI	For DQS_Strobe	Unshifted DQS strobe from input pad
PRMBDET	For DQS_Strobe	DQSI biased to go high when DQSI is tristate, goes to input logic block as well as core logic.
GSRN	Control from routing	Global Set/Reset

1. Signals available on left/right/top edges only.

2. Selected PIO.

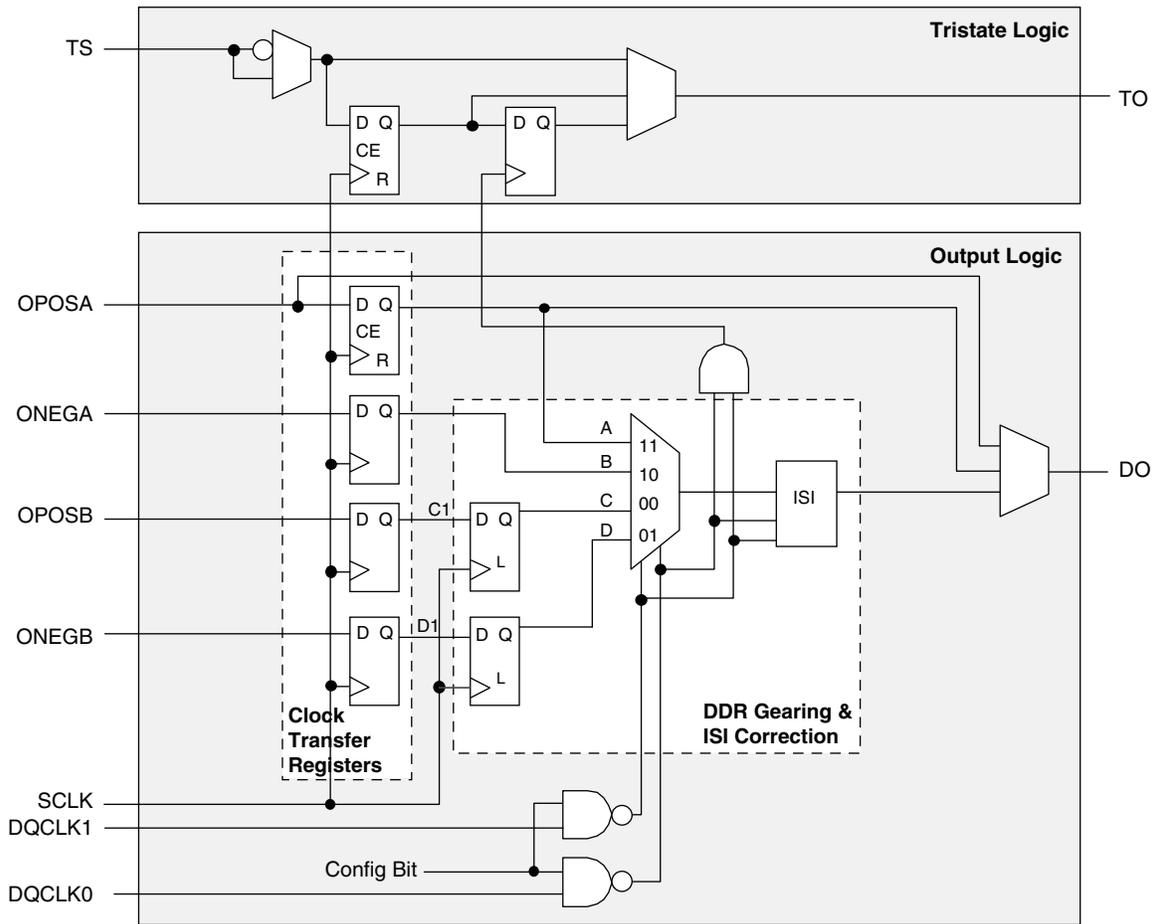
PIO

The PIO contains four blocks: an input register block, output register block, tristate register block and a control logic block. These blocks contain registers for operating in a variety of modes along with the necessary clock and selection logic.

Input Register Block

The input register blocks for the PIOs, in the left, right and top edges, contain delay elements and registers that can be used to condition high-speed interface signals, such as DDR memory interfaces and source synchronous interfaces, before they are passed to the device core. Figure 2-33 shows the input register block for the left, right and top edges. The input register block for the bottom edge contains one element to register the input signal and no DDR registers. The following description applies to the input register block for PIOs in the left, right and top edges only.

Figure 2-34. Output and Tristate Block for Left and Right Edges



Tristate Register Block

The tristate register block registers tri-state control signals from the core of the device before they are passed to the sysI/O buffers. The block contains a register for SDR operation and an additional register for DDR operation.

In SDR and non-gearing DDR modes, TS input feeds one of the flip-flops that then feeds the output. In DDRX2 mode, the register TS input is fed into another register that is clocked using the DQCLK0 and DQCLK1 signals. The output of this register is used as a tristate control.

ISI Calibration

The setting for Inter-Symbol Interference (ISI) cancellation occurs in the output register block. ISI correction is only available in the DDRX2 modes. ISI calibration settings exist once per output register block, so each I/O in a DQS-12 group may have a different ISI calibration setting.

The ISI block extends output signals at certain times, as a function of recent signal history. So, if the output pattern consists of a long strings of 0's to long strings of 1's, there are no delays on output signals. However, if there are quick, successive transitions from 010, the block will stretch out the binary 1. This is because the long trail of 0's will cause these symbols to interfere with the logic 1. Likewise, if there are quick, successive transitions from 101, the block will stretch out the binary 0. This block is controlled by a 3-bit delay control that can be set in the DQS control logic block.

For more information about this topic, please see the list of technical documentation at the end of this data sheet.

2. Left and Right (Banks 2, 3, 6 and 7) sysI/O Buffer Pairs (50% Differential and 100% Single-Ended Outputs)

The sysI/O buffer pairs in the left and right banks of the device consist of two single-ended output drivers, two sets of single-ended input buffers (both ratioed and referenced) and one differential output driver. One of the referenced input buffers can also be configured as a differential input. In these banks the two pads in the pair are described as “true” and “comp”, where the true pad is associated with the positive side of the differential I/O, and the comp (complementary) pad is associated with the negative side of the differential I/O.

In addition, programmable on-chip input termination (parallel or differential, static or dynamic) is supported on these sides, which is required for DDR3 interface. However, there is no support for hot-socketing for the I/O pins located on the left and right side of the device as the PCI clamp is always enabled on these pins.

LVDS, RSDS, PPLVDS and Mini-LVDS differential output drivers are available on 50% of the buffer pairs on the left and right banks.

3. Configuration Bank sysI/O Buffer Pairs (Single-Ended Outputs, Only on Shared Pins When Not Used by Configuration)

The sysI/O buffers in the Configuration Bank consist of ratioed single-ended output drivers and single-ended input buffers. This bank does not support PCI clamp like the other banks on the top, left, and right sides.

The two pads in the pair are described as “true” and “comp”, where the true pad is associated with the positive side of the differential input buffer and the comp (complementary) pad is associated with the negative side of the differential input buffer.

Programmable PCI clamps are only available on the top banks. PCI clamps are used primarily on inputs and bi-directional pads to reduce ringing on the receiving end.

Typical sysI/O I/O Behavior During Power-up

The internal power-on-reset (POR) signal is deactivated when V_{CC} , V_{CCIO8} and V_{CCAUX} have reached satisfactory levels. After the POR signal is deactivated, the FPGA core logic becomes active. It is the user's responsibility to ensure that all other V_{CCIO} banks are active with valid input logic levels to properly control the output logic states of all the I/O banks that are critical to the application. For more information about controlling the output logic state with valid input logic levels during power-up in LatticeECP3 devices, see the list of technical documentation at the end of this data sheet.

The V_{CC} and V_{CCAUX} supply the power to the FPGA core fabric, whereas the V_{CCIO} supplies power to the I/O buffers. In order to simplify system design while providing consistent and predictable I/O behavior, it is recommended that the I/O buffers be powered-up prior to the FPGA core fabric. V_{CCIO} supplies should be powered-up before or together with the V_{CC} and V_{CCAUX} supplies.

Supported sysI/O Standards

The LatticeECP3 sysI/O buffer supports both single-ended and differential standards. Single-ended standards can be further subdivided into LVCMOS, LVTTTL and other standards. The buffers support the LVTTTL, LVCMOS 1.2 V, 1.5 V, 1.8 V, 2.5 V and 3.3 V standards. In the LVCMOS and LVTTTL modes, the buffer has individual configuration options for drive strength, slew rates, bus maintenance (weak pull-up, weak pull-down, or a bus-keeper latch) and open drain. Other single-ended standards supported include SSTL and HSTL. Differential standards supported include LVDS, BLVDS, LVPECL, MLVDS, RSDS, Mini-LVDS, PPLVDS (point-to-point LVDS), TRLVDS (Transition Reduced LVDS), differential SSTL and differential HSTL. For further information on utilizing the sysI/O buffer to support a variety of standards please see TN1177, [LatticeECP3 sysIO Usage Guide](#).

Please see TN1177, [LatticeECP3 sysIO Usage Guide](#) for on-chip termination usage and value ranges.

Equalization Filter

Equalization filtering is available for single-ended inputs on both true and complementary I/Os, and for differential inputs on the true I/Os on the left, right, and top sides. Equalization is required to compensate for the difficulty of sampling alternating logic transitions with a relatively slow slew rate. It is considered the most useful for the Input DDRX2 modes, used in DDR3 memory, LVDS, or TRLVDS signaling. Equalization filter acts as a tunable filter with settings to determine the level of correction. In the LatticeECP3 devices, there are four settings available: 0 (none), 1, 2 and 3. The default setting is 0. The equalization logic resides in the sysI/O buffers, the two bits of setting is set uniquely in each input IOLOGIC block. Therefore, each sysI/O can have a unique equalization setting within a DQS-12 group.

Hot Socketing

LatticeECP3 devices have been carefully designed to ensure predictable behavior during power-up and power-down. During power-up and power-down sequences, the I/Os remain in tri-state until the power supply voltage is high enough to ensure reliable operation. In addition, leakage into I/O pins is controlled within specified limits. Please refer to the Hot Socketing Specifications in the DC and Switching Characteristics in this data sheet.

SERDES and PCS (Physical Coding Sublayer)

LatticeECP3 devices feature up to 16 channels of embedded SERDES/PCS arranged in quads at the bottom of the devices supporting up to 3.2Gbps data rate. Figure 2-40 shows the position of the quad blocks for the LatticeECP3-150 devices. Table 2-14 shows the location of available SERDES Quads for all devices.

The LatticeECP3 SERDES/PCS supports a range of popular serial protocols, including:

- PCI Express 1.1
- Ethernet (XAUI, GbE - 1000 Base CS/SX/LX and SGMII)
- Serial RapidIO
- SMPTE SDI (3G, HD, SD)
- CPRI
- SONET/SDH (STS-3, STS-12, STS-48)

Each quad contains four dedicated SERDES for high speed, full duplex serial data transfer. Each quad also has a PCS block that interfaces to the SERDES channels and contains protocol specific digital logic to support the standards listed above. The PCS block also contains interface logic to the FPGA fabric. All PCS logic for dedicated protocol support can also be bypassed to allow raw 8-bit or 10-bit interfaces to the FPGA fabric.

Even though the SERDES/PCS blocks are arranged in quads, multiple baud rates can be supported within a quad with the use of dedicated, per channel $\div 1$, $\div 2$ and $\div 11$ rate dividers. Additionally, multiple quads can be arranged together to form larger data pipes.

For information on how to use the SERDES/PCS blocks to support specific protocols, as well on how to combine multiple protocols and baud rates within a device, please refer to TN1176, [LatticeECP3 SERDES/PCS Usage Guide](#).

Table 2-14. Available SERDES Quads per LatticeECP3 Devices

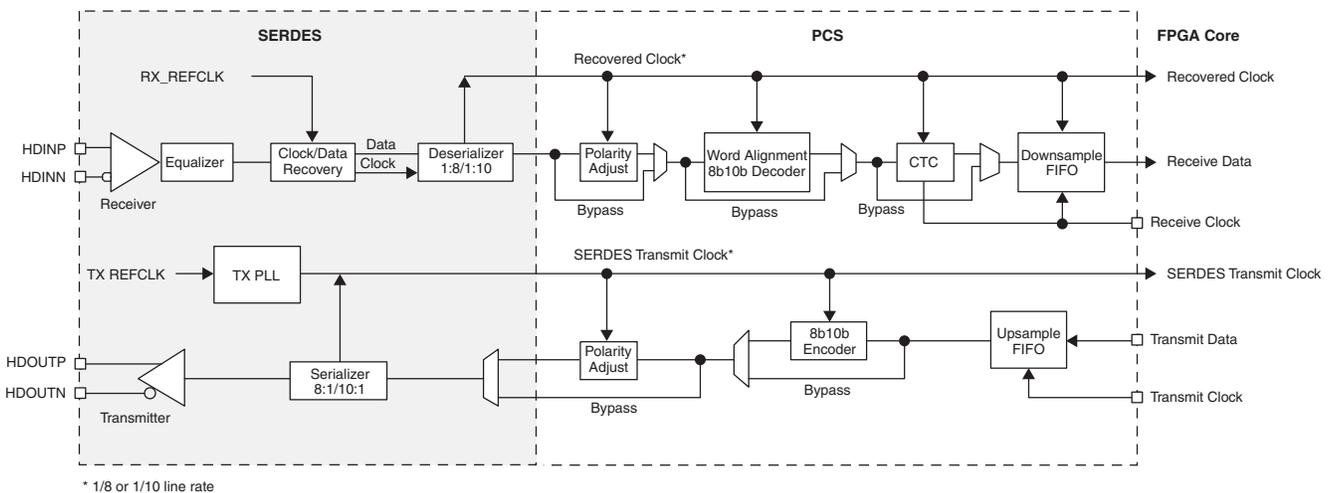
Package	ECP3-17	ECP3-35	ECP3-70	ECP3-95	ECP3-150
256 ftBGA	1	1	—	—	—
328 csBGA	2 channels	—	—	—	—
484 fpBGA	1	1	1	1	
672 fpBGA	—	1	2	2	2
1156 fpBGA	—	—	3	3	4

SERDES Block

A SERDES receiver channel may receive the serial differential data stream, equalize the signal, perform Clock and Data Recovery (CDR) and de-serialize the data stream before passing the 8- or 10-bit data to the PCS logic. The SERDES transmitter channel may receive the parallel 8- or 10-bit data, serialize the data and transmit the serial bit stream through the differential drivers. Figure 2-41 shows a single-channel SERDES/PCS block. Each SERDES channel provides a recovered clock and a SERDES transmit clock to the PCS block and to the FPGA core logic.

Each transmit channel, receiver channel, and SERDES PLL shares the same power supply (VCCA). The output and input buffers of each channel have their own independent power supplies (VCCOB and VCCIB).

Figure 2-41. Simplified Channel Block Diagram for SERDES/PCS Block



PCS

As shown in Figure 2-41, the PCS receives the parallel digital data from the deserializer and selects the polarity, performs word alignment, decodes (8b/10b), provides Clock Tolerance Compensation and transfers the clock domain from the recovered clock to the FPGA clock via the Down Sample FIFO.

For the transmit channel, the PCS block receives the parallel data from the FPGA core, encodes it with 8b/10b, selects the polarity and passes the 8/10 bit data to the transmit SERDES channel.

The PCS also provides bypass modes that allow a direct 8-bit or 10-bit interface from the SERDES to the FPGA logic. The PCS interface to the FPGA can also be programmed to run at 1/2 speed for a 16-bit or 20-bit interface to the FPGA logic.

There are some restrictions to be aware of when using spread spectrum. When a quad shares a PCI Express x1 channel with a non-PCI Express channel, ensure that the reference clock for the quad is compatible with all protocols within the quad. For example, a PCI Express spread spectrum reference clock is not compatible with most Gigabit Ethernet applications because of tight CTC ppm requirements.

While the LatticeECP3 architecture will allow the mixing of a PCI Express channel and a Gigabit Ethernet, Serial RapidIO or SGMII channel within the same quad, using a PCI Express spread spectrum clocking as the transmit reference clock will cause a violation of the Gigabit Ethernet, Serial RapidIO and SGMII transmit jitter specifications.

For further information on SERDES, please see TN1176, [LatticeECP3 SERDES/PCS Usage Guide](#).

IEEE 1149.1-Compliant Boundary Scan Testability

All LatticeECP3 devices have boundary scan cells that are accessed through an IEEE 1149.1 compliant Test Access Port (TAP). This allows functional testing of the circuit board on which the device is mounted through a serial scan path that can access all critical logic nodes. Internal registers are linked internally, allowing test data to be shifted in and loaded directly onto test nodes, or test data to be captured and shifted out for verification. The test access port consists of dedicated I/Os: TDI, TDO, TCK and TMS. The test access port has its own supply voltage V_{CCJ} and can operate with LVCMOS3.3, 2.5, 1.8, 1.5 and 1.2 standards.

For more information, please see TN1169, [LatticeECP3 sysCONFIG Usage Guide](#).

Device Configuration

All LatticeECP3 devices contain two ports that can be used for device configuration. The Test Access Port (TAP), which supports bit-wide configuration, and the sysCONFIG port, support dual-byte, byte and serial configuration. The TAP supports both the IEEE Standard 1149.1 Boundary Scan specification and the IEEE Standard 1532 In-System Configuration specification. The sysCONFIG port includes seven I/Os used as dedicated pins with the remaining pins used as dual-use pins. See TN1169, [LatticeECP3 sysCONFIG Usage Guide](#) for more information about using the dual-use pins as general purpose I/Os.

There are various ways to configure a LatticeECP3 device:

1. JTAG
2. Standard Serial Peripheral Interface (SPI and SPIm modes) - interface to boot PROM memory
3. System microprocessor to drive a x8 CPU port (PCM mode)
4. System microprocessor to drive a serial slave SPI port (SSPI mode)
5. Generic byte wide flash with a MachXO™ device, providing control and addressing

On power-up, the FPGA SRAM is ready to be configured using the selected sysCONFIG port. Once a configuration port is selected, it will remain active throughout that configuration cycle. The IEEE 1149.1 port can be activated any time after power-up by sending the appropriate command through the TAP port.

LatticeECP3 devices also support the Slave SPI Interface. In this mode, the FPGA behaves like a SPI Flash device (slave mode) with the SPI port of the FPGA to perform read-write operations.

Table 2-16. Selectable Master Clock (MCCLK) Frequencies During Configuration (Nominal)

MCCLK (MHz)	MCCLK (MHz)
	10
2.5 ¹	13
4.3	15 ²
5.4	20
6.9	26
8.1	33 ³
9.2	

1. Software default MCCLK frequency. Hardware default is 3.1 MHz.

2. Maximum MCCLK with encryption enabled.

3. Maximum MCCLK without encryption.

Density Shifting

The LatticeECP3 family is designed to ensure that different density devices in the same family and in the same package have the same pinout. Furthermore, the architecture ensures a high success rate when performing design migration from lower density devices to higher density devices. In many cases, it is also possible to shift a lower utilization design targeted for a high-density device to a lower density device. However, the exact details of the final resource utilization will impact the likelihood of success in each case. An example is that some user I/Os may become No Connects in smaller devices in the same package. Refer to the [LatticeECP3 Pin Migration Tables](#) and Diamond software for specific restrictions and limitations.

Absolute Maximum Ratings^{1, 2, 3}

Supply Voltage V_{CC}	-0.5 V to 1.32 V
Supply Voltage V_{CCAUX}	-0.5 V to 3.75 V
Supply Voltage V_{CCJ}	-0.5 V to 3.75 V
Output Supply Voltage V_{CCIO}	-0.5 V to 3.75 V
Input or I/O Tristate Voltage Applied ⁴	-0.5 V to 3.75 V
Storage Temperature (Ambient)	-65 V to 150 °C
Junction Temperature (T_J)	+125 °C

1. Stress above those listed under the “Absolute Maximum Ratings” may cause permanent damage to the device. Functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.
2. Compliance with the Lattice [Thermal Management](#) document is required.
3. All voltages referenced to GND.
4. Overshoot and undershoot of -2 V to ($V_{IHMAX} + 2$) volts is permitted for a duration of <20 ns.

Recommended Operating Conditions¹

Symbol	Parameter	Min.	Max.	Units
V_{CC}^2	Core Supply Voltage	1.14	1.26	V
$V_{CCAUX}^{2,4}$	Auxiliary Supply Voltage, Terminating Resistor Switching Power Supply (SERDES)	3.135	3.465	V
V_{CCPLL}	PLL Supply Voltage	3.135	3.465	V
$V_{CCIO}^{2,3}$	I/O Driver Supply Voltage	1.14	3.465	V
V_{CCJ}^2	Supply Voltage for IEEE 1149.1 Test Access Port	1.14	3.465	V
V_{REF1} and V_{REF2}	Input Reference Voltage	0.5	1.7	V
V_{TT}^5	Termination Voltage	0.5	1.3125	V
t_{JCOM}	Junction Temperature, Commercial Operation	0	85	°C
t_{JIND}	Junction Temperature, Industrial Operation	-40	100	°C
SERDES External Power Supply⁶				
V_{CCIB}	Input Buffer Power Supply (1.2 V)	1.14	1.26	V
	Input Buffer Power Supply (1.5 V)	1.425	1.575	V
V_{CCOB}	Output Buffer Power Supply (1.2 V)	1.14	1.26	V
	Output Buffer Power Supply (1.5 V)	1.425	1.575	V
V_{CCA}	Transmit, Receive, PLL and Reference Clock Buffer Power Supply	1.14	1.26	V

1. For correct operation, all supplies except V_{REF} and V_{TT} must be held in their valid operation range. This is true independent of feature usage.
2. If V_{CCIO} or V_{CCJ} is set to 1.2 V, they must be connected to the same power supply as V_{CC} . If V_{CCIO} or V_{CCJ} is set to 3.3 V, they must be connected to the same power supply as V_{CCAUX} .
3. See recommended voltages by I/O standard in subsequent table.
4. V_{CCAUX} ramp rate must not exceed 30 mV/μs during power-up when transitioning between 0 V and 3.3 V.
5. If not used, V_{TT} should be left floating.
6. See TN1176, [LatticeECP3 SERDES/PCS Usage Guide](#) for information on board considerations for SERDES power supplies.

DC Electrical Characteristics

Over Recommended Operating Conditions

Symbol	Parameter	Condition	Min.	Typ.	Max.	Units
$I_{IL}, I_{IH}^{1,4}$	Input or I/O Low Leakage	$0 \leq V_{IN} \leq (V_{CCIO} - 0.2 \text{ V})$	—	—	10	μA
$I_{IH}^{1,3}$	Input or I/O High Leakage	$(V_{CCIO} - 0.2 \text{ V}) < V_{IN} \leq 3.6 \text{ V}$	—	—	150	μA
I_{PU}	I/O Active Pull-up Current	$0 \leq V_{IN} \leq 0.7 V_{CCIO}$	-30	—	-210	μA
I_{PD}	I/O Active Pull-down Current	$V_{IL} (\text{MAX}) \leq V_{IN} \leq V_{CCIO}$	30	—	210	μA
I_{BHLS}	Bus Hold Low Sustaining Current	$V_{IN} = V_{IL} (\text{MAX})$	30	—	—	μA
I_{BHHS}	Bus Hold High Sustaining Current	$V_{IN} = 0.7 V_{CCIO}$	-30	—	—	μA
I_{BHLO}	Bus Hold Low Overdrive Current	$0 \leq V_{IN} \leq V_{CCIO}$	—	—	210	μA
I_{BHHO}	Bus Hold High Overdrive Current	$0 \leq V_{IN} \leq V_{CCIO}$	—	—	-210	μA
V_{BHT}	Bus Hold Trip Points	$0 \leq V_{IN} \leq V_{IH} (\text{MAX})$	$V_{IL} (\text{MAX})$	—	$V_{IH} (\text{MIN})$	V
C1	I/O Capacitance ²	$V_{CCIO} = 3.3 \text{ V}, 2.5 \text{ V}, 1.8 \text{ V}, 1.5 \text{ V}, 1.2 \text{ V},$ $V_{CC} = 1.2 \text{ V}, V_{IO} = 0 \text{ to } V_{IH} (\text{MAX})$	—	5	8	pf
C2	Dedicated Input Capacitance ²	$V_{CCIO} = 3.3 \text{ V}, 2.5 \text{ V}, 1.8 \text{ V}, 1.5 \text{ V}, 1.2 \text{ V},$ $V_{CC} = 1.2 \text{ V}, V_{IO} = 0 \text{ to } V_{IH} (\text{MAX})$	—	5	7	pf

1. Input or I/O leakage current is measured with the pin configured as an input or as an I/O with the output driver tri-stated. It is not measured with the output driver active. Bus maintenance circuits are disabled.

2. T_A 25 °C, $f = 1.0 \text{ MHz}$.

3. Applicable to general purpose I/Os in top and bottom banks.

4. When used as V_{REF} maximum leakage = 25 μA .

Register-to-Register Performance^{1, 2, 3}

Function	-8 Timing	Units
18x18 Multiply/Accumulate (Input & Output Registers)	200	MHz
18x18 Multiply-Add/Sub (All Registers)	400	MHz

1. These timing numbers were generated using ispLEVER tool. Exact performance may vary with device and tool version. The tool uses internal parameters that have been characterized but are not tested on every device.
2. Commercial timing numbers are shown. Industrial numbers are typically slower and can be extracted from the Diamond or ispLEVER software.
3. For details on -9 speed grade devices, please contact your Lattice Sales Representative.

Derating Timing Tables

Logic timing provided in the following sections of this data sheet and the Diamond and ispLEVER design tools are worst case numbers in the operating range. Actual delays at nominal temperature and voltage for best case process, can be much better than the values given in the tables. The Diamond and ispLEVER design tools can provide logic timing numbers at a particular temperature and voltage.

LatticeECP3 Maximum I/O Buffer Speed (Continued)^{1, 2, 3, 4, 5, 6}**Over Recommended Operating Conditions**

Buffer	Description	Max.	Units
PCI33	PCI, $V_{CCIO} = 3.3\text{ V}$	66	MHz

1. These maximum speeds are characterized but not tested on every device.
2. Maximum I/O speed for differential output standards emulated with resistors depends on the layout.
3. LVCMOS timing is measured with the load specified in the Switching Test Conditions table of this document.
4. All speeds are measured at fast slew.
5. Actual system operation may vary depending on user logic implementation.
6. Maximum data rate equals 2 times the clock rate when utilizing DDR.

SERDES High Speed Data Receiver

Table 3-9. Serial Input Data Specifications

Symbol	Description	Min.	Typ.	Max.	Units	
RX-CID _S	Stream of nontransitions ¹ (CID = Consecutive Identical Digits) @ 10 ⁻¹² BER	3.125 G	—	—	136	Bits
		2.5 G	—	—	144	
		1.485 G	—	—	160	
		622 M	—	—	204	
		270 M	—	—	228	
		150 M	—	—	296	
V _{RX-DIFF-S}	Differential input sensitivity	150	—	1760	mV, p-p	
V _{RX-IN}	Input levels	0	—	V _{CCA} +0.5 ⁴	V	
V _{RX-CM-DC}	Input common mode range (DC coupled)	0.6	—	V _{CCA}	V	
V _{RX-CM-AC}	Input common mode range (AC coupled) ³	0.1	—	V _{CCA} +0.2	V	
T _{RX-RELOCK}	SCDR re-lock time ²	—	1000	—	Bits	
Z _{RX-TERM}	Input termination 50/75 Ohm/High Z	-20%	50/75/HiZ	+20%	Ohms	
RL _{RX-RL}	Return loss (without package)	10	—	—	dB	

1. This is the number of bits allowed without a transition on the incoming data stream when using DC coupling.
2. This is the typical number of bit times to re-lock to a new phase or frequency within +/- 300 ppm, assuming 8b10b encoded data.
3. AC coupling is used to interface to LVPECL and LVDS. LVDS interfaces are found in laser drivers and Fibre Channel equipment. LVDS interfaces are generally found in 622 Mbps SERDES devices.
4. Up to 1.76 V.

Input Data Jitter Tolerance

A receiver's ability to tolerate incoming signal jitter is very dependent on jitter type. High speed serial interface standards have recognized the dependency on jitter type and have specifications to indicate tolerance levels for different jitter types as they relate to specific protocols. Sinusoidal jitter is considered to be a worst case jitter type.

Table 3-10. Receiver Total Jitter Tolerance Specification

Description	Frequency	Condition	Min.	Typ.	Max.	Units
Deterministic	3.125 Gbps	600 mV differential eye	—	—	0.47	UI, p-p
Random		600 mV differential eye	—	—	0.18	UI, p-p
Total		600 mV differential eye	—	—	0.65	UI, p-p
Deterministic	2.5 Gbps	600 mV differential eye	—	—	0.47	UI, p-p
Random		600 mV differential eye	—	—	0.18	UI, p-p
Total		600 mV differential eye	—	—	0.65	UI, p-p
Deterministic	1.25 Gbps	600 mV differential eye	—	—	0.47	UI, p-p
Random		600 mV differential eye	—	—	0.18	UI, p-p
Total		600 mV differential eye	—	—	0.65	UI, p-p
Deterministic	622 Mbps	600 mV differential eye	—	—	0.47	UI, p-p
Random		600 mV differential eye	—	—	0.18	UI, p-p
Total		600 mV differential eye	—	—	0.65	UI, p-p

Note: Values are measured with CJPAT, all channels operating, FPGA Logic active, I/Os around SERDES pins quiet, voltages are nominal, room temperature.

Gigabit Ethernet/Serial Rapid I/O Type 1/SGMII/CPRI LV E.12 Electrical and Timing Characteristics

AC and DC Characteristics

Table 3-17. Transmit

Symbol	Description	Test Conditions	Min.	Typ.	Max.	Units
T_{RF}	Differential rise/fall time	20%-80%	—	80	—	ps
$Z_{TX_DIFF_DC}$	Differential impedance		80	100	120	Ohms
$J_{TX_DDJ}^{3,4,5}$	Output data deterministic jitter		—	—	0.10	UI
$J_{TX_TJ}^{2,3,4,5}$	Total output data jitter		—	—	0.24	UI

1. Rise and fall times measured with board trace, connector and approximately 2.5 pF load.
2. Total jitter includes both deterministic jitter and random jitter. The random jitter is the total jitter minus the actual deterministic jitter.
3. Jitter values are measured with each CML output AC coupled into a 50-Ohm impedance (100-Ohm differential impedance).
4. Jitter and skew are specified between differential crossings of the 50% threshold of the reference signal.
5. Values are measured at 1.25 Gbps.

Table 3-18. Receive and Jitter Tolerance

Symbol	Description	Test Conditions	Min.	Typ.	Max.	Units
RL_{RX_DIFF}	Differential return loss	From 100 MHz to 1.25 GHz	10	—	—	dB
RL_{RX_CM}	Common mode return loss	From 100 MHz to 1.25 GHz	6	—	—	dB
Z_{RX_DIFF}	Differential termination resistance		80	100	120	Ohms
$J_{RX_DJ}^{1,2,3,4,5}$	Deterministic jitter tolerance (peak-to-peak)		—	—	0.34	UI
$J_{RX_RJ}^{1,2,3,4,5}$	Random jitter tolerance (peak-to-peak)		—	—	0.26	UI
$J_{RX_SJ}^{1,2,3,4,5}$	Sinusoidal jitter tolerance (peak-to-peak)		—	—	0.11	UI
$J_{RX_TJ}^{1,2,3,4,5}$	Total jitter tolerance (peak-to-peak)		—	—	0.71	UI
T_{RX_EYE}	Receiver eye opening		0.29	—	—	UI

1. Total jitter includes deterministic jitter, random jitter and sinusoidal jitter. The sinusoidal jitter tolerance mask is shown in Figure 3-18.
2. Jitter values are measured with each high-speed input AC coupled into a 50-Ohm impedance.
3. Jitter and skew are specified between differential crossings of the 50% threshold of the reference signal.
4. Jitter tolerance, Differential Input Sensitivity and Receiver Eye Opening parameters are characterized when Full Rx Equalization is enabled.
5. Values are measured at 1.25 Gbps.

HDMI (High-Definition Multimedia Interface) Electrical and Timing Characteristics

AC and DC Characteristics

Table 3-22. Transmit and Receive^{1,2}

Symbol	Description	Spec. Compliance		Units
		Min. Spec.	Max. Spec.	
Transmit				
Intra-pair Skew		—	75	ps
Inter-pair Skew		—	800	ps
TMDS Differential Clock Jitter		—	0.25	UI
Receive				
R_T	Termination Resistance	40	60	Ohms
V_{ICM}	Input AC Common Mode Voltage (50-Ohm Setting)	—	50	mV
TMDS Clock Jitter	Clock Jitter Tolerance	—	0.25	UI

1. Output buffers must drive a translation device. Max. speed is 2 Gbps. If translation device does not modify rise/fall time, the maximum speed is 1.5 Gbps.
2. Input buffers must be AC coupled in order to support the 3.3 V common mode. Generally, HDMI inputs are terminated by an external cable equalizer before data/clock is forwarded to the LatticeECP3 device.

LatticeECP3 sysCONFIG Port Timing Specifications (Continued)

Over Recommended Operating Conditions

Parameter	Description	Min.	Max.	Units
t_{SSCL}	CCLK Minimum Low Pulse	5	—	ns
t_{HLCH}	HOLDN Low Setup Time (Relative to CCLK)	5	—	ns
t_{CHHH}	HOLDN Low Hold Time (Relative to CCLK)	5	—	ns
Master and Slave SPI (Continued)				
t_{CHHL}	HOLDN High Hold Time (Relative to CCLK)	5	—	ns
t_{HHCH}	HOLDN High Setup Time (Relative to CCLK)	5	—	ns
t_{HLQZ}	HOLDN to Output High-Z	—	9	ns
t_{HHQX}	HOLDN to Output Low-Z	—	9	ns

1. Re-toggling the PROGRAMN pin is not permitted until the INITN pin is high. Avoid consecutive toggling of the PROGRAMN.

Parameter	Min.	Max.	Units
Master Clock Frequency	Selected value - 15%	Selected value + 15%	MHz
Duty Cycle	40	60	%

Figure 3-20. sysCONFIG Parallel Port Read Cycle

