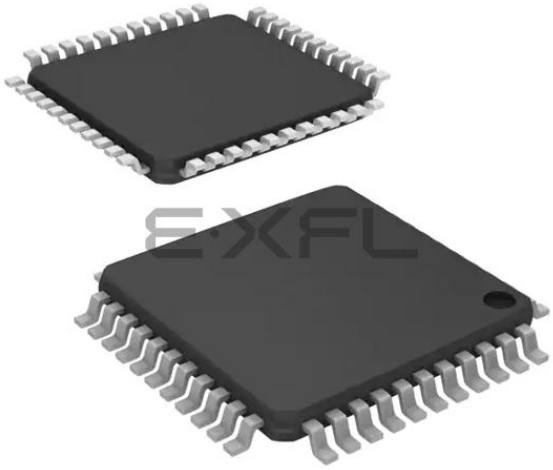

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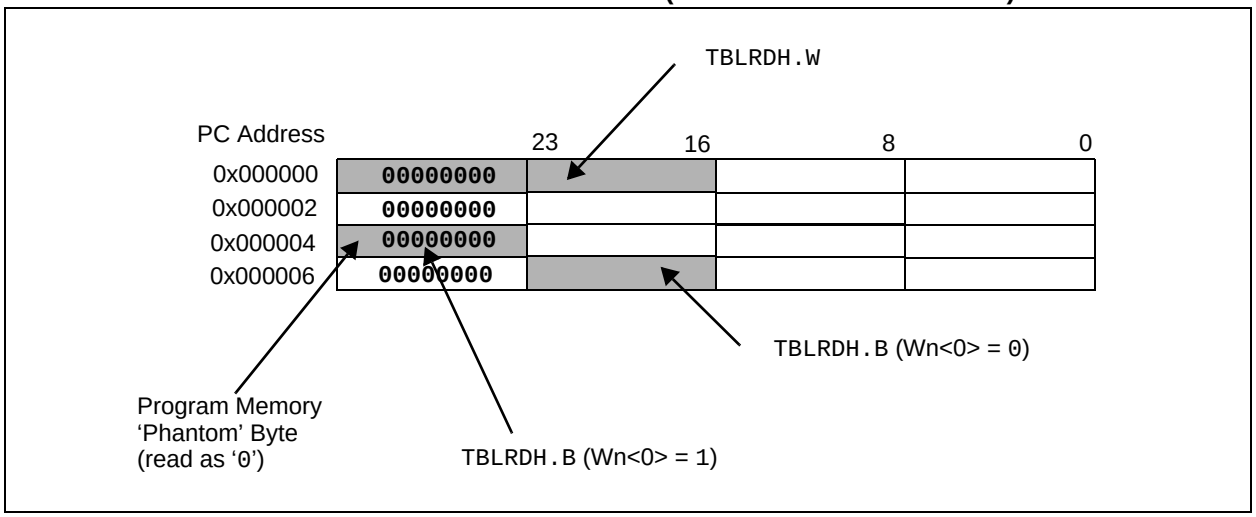
Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	dsPIC
Core Size	16-Bit
Speed	20 MIPS
Connectivity	CANbus, I²C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, Motor Control PWM, QEI, POR, PWM, WDT
Number of I/O	30
Program Memory Size	48KB (16K x 24)
Program Memory Type	FLASH
EEPROM Size	1K x 8
RAM Size	2K x 8
Voltage - Supply (Vcc/Vdd)	2.5V ~ 5.5V
Data Converters	A/D 9x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	44-TQFP
Supplier Device Package	44-TQFP (10x10)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/dspic30f4011t-20i-pt

FIGURE 3-4: PROGRAM DATA TABLE ACCESS (MOST SIGNIFICANT BYTE)



3.1.2 DATA ACCESS FROM PROGRAM MEMORY USING PROGRAM SPACE VISIBILITY

The upper 32 Kbytes of data space may optionally be mapped into any 16K word program space page. This provides transparent access of stored constant data from X data space without the need to use special instructions (i.e., TBLRDL/H, TBLWTL/H instructions).

Program space access through the data space occurs if the MSb of the data space EA is set and program space visibility is enabled by setting the PSV bit in the Core Control register (CORCON). The functions of CORCON are discussed in **Section 2.4 "DSP Engine"**.

Data accesses to this area add an additional cycle to the instruction being executed, since two program memory fetches are required.

Note that the upper half of addressable data space is always part of the X data space. Therefore, when a DSP operation uses program space mapping to access this memory region, Y data space should typically contain state (variable) data for DSP operations, whereas X data space should typically contain coefficient (constant) data.

Although each data space address, 0x8000 and higher, maps directly into a corresponding program memory address (see Figure 3-5), only the lower 16 bits of the 24-bit program word are used to contain the data. The upper 8 bits should be programmed to force an illegal instruction to maintain machine robustness. For information on instruction encoding, refer to the *"16-bit MCU and DSC Programmer's Reference Manual"* (DS70157).

Note that by incrementing the PC by 2 for each program memory word, the Least Significant 15 bits of data space addresses directly map to the Least Significant 15 bits in the corresponding program space addresses. The remaining bits are provided by the Program Space Visibility Page register, PSVPAG<7:0>, as shown in Figure 3-5.

Note:	PSV access is temporarily disabled during table reads/writes.

For instructions that use PSV which are executed outside a REPEAT loop:

- The following instructions require one instruction cycle in addition to the specified execution time:
  - MAC class of instructions with data operand prefetch
  - MOV instructions
  - MOV.D instructions
- All other instructions require two instruction cycles in addition to the specified execution time of the instruction.

For instructions that use PSV which are executed inside a REPEAT loop:

- The following instances require two instruction cycles in addition to the specified execution time of the instruction:
  - Execution in the first iteration
  - Execution in the last iteration
  - Execution prior to exiting the loop due to an interrupt
  - Execution upon re-entering the loop after an interrupt is serviced
- Any other iteration of the REPEAT loop allows the instruction, accessing data using PSV, to execute in a single cycle.

3.2.2 DATA SPACES

The X data space is used by all instructions and supports all addressing modes. There are separate read and write data buses. The X read data bus is the return data path for all instructions that view data space as combined X and Y address space. It is also the X address space data path for the dual operand read instructions (MAC class). The X write data bus is the only write path to data space for all instructions.

The X data space also supports Modulo Addressing for all instructions, subject to addressing mode restrictions. Bit-Reversed Addressing is only supported for writes to X data space.

The Y data space is used in concert with the X data space by the MAC class of instructions (CLR, ED, EDAC, MAC, MOVSAC, MPY, MPY.N and MSC) to provide two concurrent data read paths. No writes occur across the Y bus. This class of instructions dedicates two W register pointers, W10 and W11, to always address Y data space, independent of X data space, whereas W8 and W9 always address X data space. Note that during accumulator write-back, the data address space is considered a combination of X and Y data spaces, so the write occurs across the X bus. Consequently, the write can be to any address in the entire data space.

The Y data space can only be used for the data prefetch operation associated with the MAC class of instructions. It also supports Modulo Addressing for automated circular buffers. Of course, all other instructions can access the Y data address space through the X data path, as part of the composite linear space.

The boundary between the X and Y data spaces is defined as shown in Figure 3-6 and is not user-programmable. Should an EA point to data outside its own assigned address space, or to a location outside physical memory, an all-zero word/byte is returned. For example, although Y address space is visible by all non-MAC instructions using any addressing mode, an attempt by a MAC instruction to fetch data from that space, using W8 or W9 (X Space Pointers), returns 0x0000.

TABLE 3-2: EFFECT OF INVALID MEMORY ACCESSES

Attempted Operation	Data Returned
EA = an unimplemented address	0x0000
W8 or W9 used to access Y data space in a MAC instruction	0x0000
W10 or W11 used to access X data space in a MAC instruction	0x0000

All Effective Addresses (EA) are 16 bits wide and point to bytes within the data space. Therefore, the data space address range is 64 Kbytes or 32K words.

3.2.3 DATA SPACE WIDTH

The core data width is 16-bits. All internal registers are organized as 16-bit wide words. Data space memory is organized in byte addressable, 16-bit wide blocks.

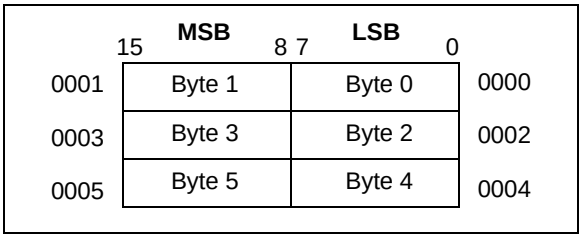
3.2.4 DATA ALIGNMENT

To help maintain backward compatibility with PIC® devices and improve data space memory usage efficiency, the dsPIC30F instruction set supports both word and byte operations. Data is aligned in data memory and registers as words, but all data space EAs resolve to bytes. Data byte reads read the complete word, which contains the byte, using the LSb of any EA to determine which byte to select. The selected byte is placed onto the LSB of the X data path (no byte accesses are possible from the Y data path as the MAC class of instructions can only fetch words). That is, data memory and registers are organized as two parallel byte-wide entities, with shared (word) address decode, but separate write lines. Data byte writes only write to the corresponding side of the array or register which matches the byte address.

As a consequence of this byte accessibility, all Effective Address calculations (including those generated by the DSP operations, which are restricted to word-sized data) are internally scaled to step through word-aligned memory. For example, the core would recognize that Post-Modified Register Indirect Addressing mode, [Ws++], will result in a value of Ws + 1 for byte operations and Ws + 2 for word operations.

All word accesses must be aligned to an even address. Misaligned word data fetches are not supported, so care must be taken when mixing byte and word operations, or translating from 8-bit MCU code. Should a misaligned read or write be attempted, an address error trap is generated. If the error occurred on a read, the instruction underway is completed, whereas if it occurred on a write, the instruction will be executed but the write does not occur. In either case, a trap is then executed, allowing the system and/or user to examine the machine state prior to execution of the address Fault.

FIGURE 3-8: DATA ALIGNMENT



6.4 RTSP Operation

The dsPIC30F Flash program memory is organized into rows and panels. Each row consists of 32 instructions or 96 bytes. Each panel consists of 128 rows or 4K x 24 instructions. RTSP allows the user to erase one row (32 instructions) at a time and to program 32 instructions at one time.

Each panel of program memory contains write latches that hold 32 instructions of programming data. Prior to the actual programming operation, the write data must be loaded into the panel write latches. The data to be programmed into the panel is loaded in sequential order into the write latches: instruction 0, instruction 1, etc. The addresses loaded must always be from a 32 address boundary.

The basic sequence for RTSP programming is to set up a Table Pointer, then do a series of TBLWT instructions to load the write latches. Programming is performed by setting the special bits in the NVMCON register. 32 TBLWTL and 32 TBLWTH instructions are required to load the 32 instructions.

All of the table write operations are single-word writes (2 instruction cycles) because only the table latches are written.

After the latches are written, a programming operation needs to be initiated to program the data.

The Flash program memory is readable, writable and erasable during normal operation over the entire VDD range.

6.5 RTSP Control Registers

The four SFRs used to read and write the program Flash memory are:

- NVMCON
- NVMADR
- NVMADRU
- NVMKEY

6.5.1 NVMCON REGISTER

The NVMCON register controls which blocks are to be erased, which memory type is to be programmed and the start of the programming cycle.

6.5.2 NVMADR REGISTER

The NVMADR register is used to hold the lower two bytes of the Effective Address. The NVMADR register captures the EA<15:0> of the last table instruction that has been executed and selects the row to write.

6.5.3 NVMADRU REGISTER

The NVMADRU register is used to hold the upper byte of the Effective Address. The NVMADRU register captures the EA<23:16> of the last table instruction that has been executed.

6.5.4 NVMKEY REGISTER

NVMKEY is a write-only register that is used for write protection. To start a programming or an erase sequence, the user must consecutively write 0x55 and 0xAA to the NVMKEY register. Refer to **Section 6.6 “Programming Operations”** for further details.

Note:	The user can also directly write to the NVMADR and NVMADRU registers to specify a program memory address for erasing or programming.

TABLE 11-1: TIMER4/5 REGISTER MAP(1)

SFR Name	Addr.	Bit 15	Bit 14	Bit 13	Bit 12	Bit 11	Bit 10	Bit 9	Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset State
TMR4	0114	Timer4 Register																uuuu uuuu uuuu uuuu
TMR5HLD	0116	Timer5 Holding Register (For 32-bit operations only)																uuuu uuuu uuuu uuuu
TMR5	0118	Timer5 Register																uuuu uuuu uuuu uuuu
PR4	011A	Period Register 4																1111 1111 1111 1111
PR5	011C	Period Register 5																1111 1111 1111 1111
T4CON	011E	TON	—	TSIDL	—	—	—	—	—	—	TGATE	TCKPS1	TCKPS0	T45	—	TCS	—	0000 0000 0000 0000
T5CON	0120	TON	—	TSIDL	—	—	—	—	—	—	TGATE	TCKPS1	TCKPS0	—	—	TCS	—	0000 0000 0000 0000

Legend: u = uninitialized bit; — = unimplemented bit, read as '0'

Note 1: Refer to the "*dsPIC30F Family Reference Manual*" (DS70046) for descriptions of register bit fields.

13.4.2 PWM PERIOD

The PWM period is specified by writing to the PRx register. The PWM period can be calculated using Equation 13-1.

EQUATION 13-1: PWM PERIOD

$$PWM\ period = [(PRx) + 1] \cdot 4 \cdot T_{OSC} \cdot (TMRx\ prescale\ value)$$

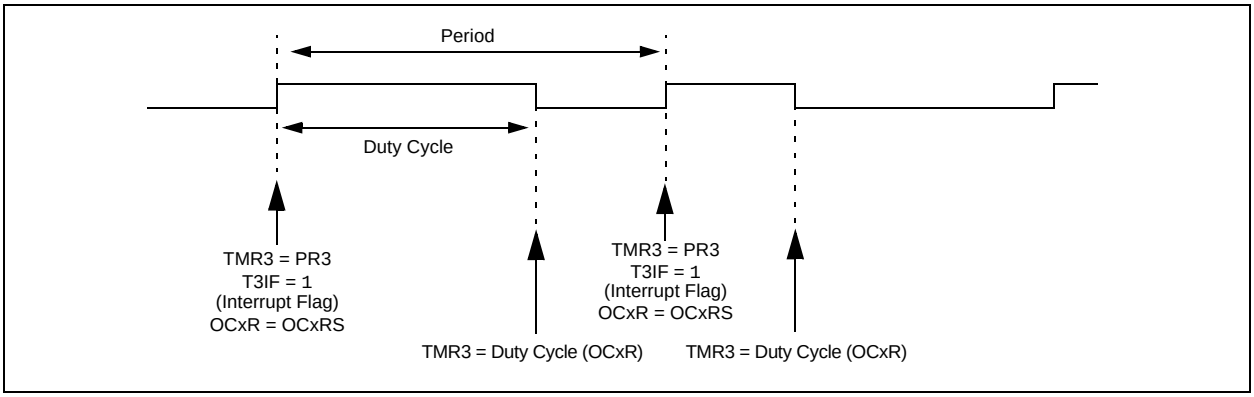
PWM frequency is defined as 1/[PWM period].

When the selected TMRx is equal to its respective Period register, PRx, the following four events occur on the next increment cycle:

- TMRx is cleared
- The OCx pin is set
  - Exception 1: If PWM duty cycle is 0x0000, the OCx pin will remain low.
  - Exception 2: If duty cycle is greater than PRx, the pin will remain high.
- The PWM duty cycle is latched from OCxRS into OCxR
- The corresponding timer interrupt flag is set

See Figure 13-1 for key PWM period comparisons. Timer3 is referred to in the figure for clarity.

FIGURE 13-1: PWM OUTPUT TIMING



14.1 Quadrature Encoder Interface Logic

A typical, incremental (a.k.a. optical) encoder has three outputs: Phase A, Phase B and an index pulse. These signals are useful and often required in position and speed control of ACIM and SR motors.

The two channels, Phase A (QEA) and Phase B (QEB), have a unique relationship. If Phase A leads Phase B, then the direction (of the motor) is deemed positive or forward. If Phase A lags Phase B, then the direction (of the motor) is deemed negative or reverse.

A third channel, termed index pulse, occurs once per revolution and is used as a reference to establish an absolute position. The index pulse coincides with Phase A and Phase B, both low.

14.2 16-bit Up/Down Position Counter Mode

The 16-bit Up/Down Counter counts up or down on every count pulse, which is generated by the difference of the Phase A and Phase B input signals. The counter acts as an integrator, whose count value is proportional to position. The direction of the count is determined by the UPDN signal which is generated by the Quadrature Encoder Interface Logic.

14.2.1 POSITION COUNTER ERROR CHECKING

Position counter error checking in the QEI is provided for and indicated by the CNTERR bit (QEICON<15>). The error checking only applies when the position counter is configured for Reset on the Index Pulse modes (QEIM<2:0> = 110 or 100). In these modes, the contents of the POSCNT register are compared with the values (0xFFFF or MAXCNT + 1, depending on direction). If these values are detected, an error condition is generated by setting the CNTERR bit and a QEI count error interrupt is generated. The QEI count error interrupt can be disabled by setting the CEID bit (DFLTCON<8>). The position counter continues to count encoder edges after an error has been detected. The POSCNT register continues to count up/down until a natural rollover/underflow. No interrupt is generated for the natural rollover/underflow event. The CNTERR bit is a read/write bit and reset in software by the user.

14.2.2 POSITION COUNTER RESET

The Position Counter Reset Enable bit, POSRES (QEICON<2>) controls whether the position counter is reset when the index pulse is detected. This bit is only applicable when QEIM<2:0> = 100 or 110.

If the POSRES bit is set to '1', then the position counter is reset when the index pulse is detected. If the POSRES bit is set to '0', then the position counter is not reset when the index pulse is detected. The position counter will continue counting up or down and will be reset on the rollover or underflow condition.

When selecting the INDX signal to reset the position counter (POSCNT), the user has to specify the states on the QEA and QEB input pins. These states have to be matched in order for a Reset to occur. These states are selected by the IMV<1:0> bits (DFLTCON<10:9>).

The Index Match Value bits (IMV<1:0>) allow the user to specify the state of the QEA and QEB input pins, during an index pulse, when the POSCNT register is to be reset.

In 4x Quadrature Count mode:
IMV1 = Required state of Phase B input signal for match on index pulse
IMV0 = Required state of Phase A input signal for match on index pulse

In 2x Quadrature Count mode:
IMV1 = Selects phase input signal for index state match (0 = Phase A, 1 = Phase B)
IMV0 = Required state of the selected phase input signal for match on index pulse

The interrupt is still generated on the detection of the index pulse and not on the position counter overflow/underflow.

14.2.3 COUNT DIRECTION STATUS

As mentioned in the previous section, the QEI logic generates an UPDN signal, based upon the relationship between Phase A and Phase B. In addition to the output pin, the state of this internal UPDN signal is supplied to an SFR bit, UPDN (QEICON<11>), as a read-only bit.

Note:	QEI pins are multiplexed with analog inputs. The user must insure that all QEI associated pins are set as digital inputs in the ADPCFG register.

15.13 PWM Update Lockout

For a complex PWM application, the user may need to write up to three duty cycle registers and the Time Base Period register, PTPER, at a given time. In some applications, it is important that all buffer registers be written before the new duty cycle and period values are loaded for use by the module.

The PWM update lockout feature is enabled by setting the UDIS control bit in the PWMCON2 SFR. The UDIS bit affects all duty cycle buffer registers and the PWM Time Base Period buffer, PTPER. No duty cycle changes or period value changes will have effect while UDIS = 1.

15.14 PWM Special Event Trigger

The PWM module has a special event trigger that allows A/D conversions to be synchronized to the PWM time base. The A/D sampling and conversion time may be programmed to occur at any point within the PWM period. The special event trigger allows the user to minimize the delay between the time when A/D conversion results are acquired and the time when the duty cycle value is updated.

The PWM special event trigger has an SFR named SEVTCMP and five control bits to control its operation. The PTMR value for which a special event trigger should occur is loaded into the SEVTCMP register. When the PWM time base is in a Continuous Up/Down Count mode, an additional control bit is required to specify the counting phase for the special event trigger. The count phase is selected using the SEVTDIR control bit in the SEVTCMP SFR. If the SEVTDIR bit is cleared, the special event trigger will occur on the upward counting cycle of the PWM time base. If the SEVTDIR bit is set, the special event trigger will occur on the downward count cycle of the PWM time base. The SEVTDIR control bit has no effect unless the PWM time base is configured for a Continuous Up/Down Count mode.

15.14.1 SPECIAL EVENT TRIGGER POSTSCALER

The PWM special event trigger has a postscaler that allows a 1:1 to 1:16 postscale ratio. The postscaler is configured by writing the SEVOPS<3:0> control bits in the PWMCON2 SFR.

The special event output postscaler is cleared on the following events:

- Any write to the SEVTCMP register
- Any device Reset

15.15 PWM Operation During CPU Sleep Mode

The Fault A input pin has the ability to wake the CPU from Sleep mode. The PWM module generates an interrupt if the Fault pin is driven low while in Sleep.

15.16 PWM Operation During CPU Idle Mode

The PTCON SFR contains a PTSIDL control bit. This bit determines if the PWM module will continue to operate or stop when the device enters Idle mode. If PTSIDL = 0, the module will continue to operate. If PTSIDL = 1, the module will stop operation as long as the CPU remains in Idle mode.

17.2 I²C Module Addresses

The I2CADD register contains the Slave mode addresses. The register is a 10-bit register.

If the A10M bit (I2CCON<10>) is '0', the address is interpreted by the module as a 7-bit address. When an address is received, it is compared to the 7 LSbs of the I2CADD register.

If the A10M bit is '1', the address is assumed to be a 10-bit address. When an address is received, it will be compared with the binary value, '11110 A9 A8' (where A9 and A8 are two Most Significant bits of I2CADD). If that value matches, the next address will be compared with the Least Significant 8 bits of I2CADD, as specified in the 10-bit addressing protocol.

The 7-bit I²C slave addresses supported by the dsPIC30F are shown in Table 17-1.

TABLE 17-1: 7-BIT I²C™ SLAVE ADDRESSES

Address	Description
0x00	General call address or Start byte
0x01-0x03	Reserved
0x04-0x07	HS mode master codes
0x08-0x77	Valid 7-bit addresses
0x78-0x7B	Valid 10-bit addresses (lower 7 bits)
0x7C-0x7F	Reserved

17.3 I²C 7-bit Slave Mode Operation

Once enabled (I2CEN = 1), the slave module will wait for a Start bit to occur (i.e., the I²C module is 'Idle'). Following the detection of a Start bit, 8 bits are shifted into I2CRSR and the address is compared against I2CADD. In 7-bit mode (A10M = 0), I2CADD<6:0> bits are compared against I2CRSR<7:1>, and I2CRSR<0> is the R_W bit. All incoming bits are sampled on the rising edge of SCL.

If an address match occurs, an Acknowledgement will be sent, and the Slave Event Interrupt Flag (SI2CIF) is set on the falling edge of the ninth (ACK) bit. The address match does not affect the contents of the I2CRCV buffer or the RBF bit.

17.3.1 SLAVE TRANSMISSION

If the R_W bit received is a '1', then the serial port will go into Transmit mode. It will send ACK on the ninth bit and then hold SCL to '0' until the CPU responds by writing to I2CTRN. SCL is released by setting the SCLREL bit and 8 bits of data are shifted out. Data bits are shifted out on the falling edge of SCL, such that SDA is valid during SCL high (see timing diagram). The interrupt pulse is sent on the falling edge of the ninth clock pulse, regardless of the status of the ACK received from the master.

17.3.2 SLAVE RECEPTION

If the R_W bit received is a '0' during an address match, then Receive mode is initiated. Incoming bits are sampled on the rising edge of SCL. After 8 bits are received, if I2CRCV is not full or I2COV is not set, I2CRSR is transferred to I2CRCV. ACK is sent on the ninth clock.

If the RBF flag is set, indicating that I2CRCV is still holding data from a previous operation (RBF = 1), then ACK is not sent; however, the interrupt pulse is generated. In the case of an overflow, the contents of the I2CRSR are not loaded into the I2CRCV.

Note:	The I2CRCV will be loaded if the I2COV bit = 1 and the RBF flag = 0. In this case, a read of the I2CRCV was performed but the user did not clear the state of the I2COV bit before the next receive occurred. The Acknowledgement is not sent (ACK = 1) and the I2CRCV is updated.

17.4 I²C 10-bit Slave Mode Operation

In 10-bit mode, the basic receive and transmit operations are the same as in the 7-bit mode. However, the criteria for address match is more complex.

The I²C specification dictates that a slave must be addressed for a write operation, with two address bytes following a Start bit.

The A10M bit is a control bit that signifies that the address in I2CADD is a 10-bit address rather than a 7-bit address. The address detection protocol for the first byte of a message address is identical for 7-bit and 10-bit messages but the bits being compared are different.

I2CADD holds the entire 10-bit address. Upon receiving an address following a Start bit, I2CRSR<7:3> is compared against a literal '11110' (the default 10-bit address) and I2CRSR<2:1> are compared against I2CADD<9:8>. If a match occurs and if R_W = 0, the interrupt pulse is sent. The ADD10 bit will be cleared to indicate a partial address match. If a match fails or R_W = 1, the ADD10 bit is cleared and the module returns to the Idle state.

The low byte of the address is then received and compared with I2CADD<7:0>. If an address match occurs, the interrupt pulse is generated and the ADD10 bit is set, indicating a complete 10-bit address match. If an address match did not occur, the ADD10 bit is cleared and the module returns to the Idle state.

18.3.4 TRANSMIT INTERRUPT

The Transmit Interrupt Flag (U1TXIF or U2TXIF) is located in the corresponding interrupt flag register.

The transmitter generates an edge to set the UxTXIF bit. The condition for generating the interrupt depends on UTXISEL control bit:

a) If UTXISEL = 0, an interrupt is generated when a word is transferred from the transmit buffer to the Transmit Shift register (UxTSR). This implies that the transmit buffer has at least one empty word.
b) If UTXISEL = 1, an interrupt is generated when a word is transferred from the transmit buffer to the Transmit Shift register (UxTSR) and the transmit buffer is empty.

Switching between the two interrupt modes during operation is possible and sometimes offers more flexibility.

18.3.5 TRANSMIT BREAK

Setting the UTXBRK bit (UxSTA<11>) will cause the UxTX line to be driven to logic '0'. The UTXBRK bit overrides all transmission activity. Therefore, the user should generally wait for the transmitter to be Idle before setting UTXBRK.

To send a Break character, the UTXBRK bit must be set by software and must remain set for a minimum of 13 baud clock cycles. The UTXBRK bit is then cleared by software to generate Stop bits. The user must wait for a duration of at least one or two baud clock cycles in order to ensure a valid Stop bit(s) before reloading the UxTXB or starting other transmitter activity. Transmission of a Break character does not generate a transmit interrupt.

18.4 Receiving Data

18.4.1 RECEIVING IN 8-BIT OR 9-BIT DATA MODE

The following steps must be performed while receiving 8-bit or 9-bit data:

1. Set up the UART (see **Section 18.3.1 "Transmitting in 8-bit Data Mode"**).
2. Enable the UART (see **Section 18.3.1 "Transmitting in 8-bit Data Mode"**).
3. A receive interrupt will be generated when one or more data words have been received, depending on the receive interrupt settings specified by the URXISEL bits (UxSTA<7:6>).
4. Read the OERR bit to determine if an overrun error has occurred. The OERR bit must be reset in software.
5. Read the received data from UxRXREG. The act of reading UxRXREG will move the next word to the top of the receive FIFO and the PERR and FERR values will be updated.

18.4.2 RECEIVE BUFFER (UxRXB)

The receive buffer is 4 words deep. Including the Receive Shift register (UxRSR), the user effectively has a 5-word deep FIFO buffer.

URXDA (UxSTA<0>) = 1 indicates that the receive buffer has data available. URXDA = 0 implies that the buffer is empty. If a user attempts to read an empty buffer, the old values in the buffer will be read and no data shift will occur within the FIFO.

The FIFO is reset during any device Reset. It is not affected when the device enters or wakes up from a power-saving mode.

18.4.3 RECEIVE INTERRUPT

The Receive Interrupt Flag (U1RXIF or U2RXIF) can be read from the corresponding interrupt flag register. The interrupt flag is set by an edge generated by the receiver. The condition for setting the receive interrupt flag depends on the settings specified by the URXISEL<1:0> (UxSTA<7:6>) control bits.

a) If URXISEL<1:0> = 00 or 01, an interrupt is generated every time a data word is transferred from the Receive Shift register (UxRSR) to the receive buffer. There may be one or more characters in the receive buffer.
b) If URXISEL<1:0> = 10, an interrupt is generated when a word is transferred from the Receive Shift register (UxRSR) to the receive buffer which, as a result of the transfer, contains 3 characters.
c) If URXISEL<1:0> = 11, an interrupt is set when a word is transferred from the Receive Shift register (UxRSR) to the receive buffer which, as a result of the transfer, contains 4 characters (i.e., becomes full).

Switching between the interrupt modes during operation is possible, though generally not advisable during normal operation.

18.5 Reception Error Handling

18.5.1 RECEIVE BUFFER OVERRUN ERROR (OERR BIT)

The OERR bit (UxSTA<1>) is set if all of the following conditions occur:

a) The receive buffer is full.
b) The Receive Shift register is full, but unable to transfer the character to the receive buffer.
c) The Stop bit of the character in the UxRSR is detected, indicating that the UxRSR needs to transfer the character to the buffer.

Once OERR is set, no further data is shifted in UxRSR (until the OERR bit is cleared in software or a Reset occurs). The data held in UxRSR and UxRXREG remains valid.

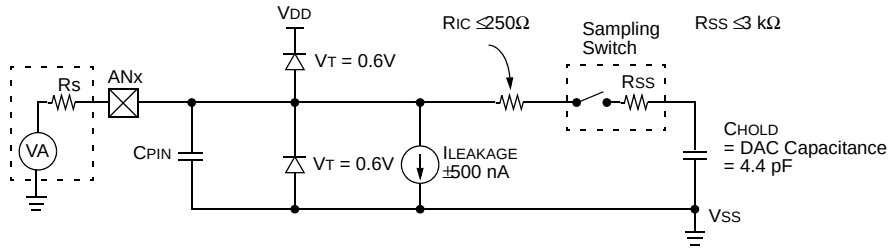
20.8 A/D Acquisition Requirements

The analog input model of the 10-bit ADC is shown in Figure 20-3. The total sampling time for the ADC is a function of the internal amplifier settling time, device VDD and the holding capacitor charge time.

For the ADC to meet its specified accuracy, the charge holding capacitor (CHOLD) must be allowed to fully charge to the voltage level on the analog input pin. The source impedance (RS), the interconnect impedance (RIC) and the internal sampling switch (RSS) impedance combine to directly affect the time required to charge the capacitor CHOLD. The combined impedance of the analog sources must therefore be small enough to fully charge the holding capacitor within the chosen sample time. To minimize the effects of pin leakage currents on the accuracy of the ADC, the maximum recommended source impedance, RS, is 5 kΩ After the analog input channel is selected (changed), this sampling function must be completed prior to starting the conversion. The internal holding capacitor will be in a discharged state prior to each sample operation.

The user must allow at least 1 TAD period of sampling time, TSAMP, between conversions to allow each sample to be acquired. This sample time may be controlled manually in software by setting/clearing the SAMP bit, or it may be automatically controlled by the ADC. In an automatic configuration, the user must allow enough time between conversion triggers so that the minimum sample time can be satisfied. Refer to **Section 24.0 "Electrical Characteristics"** for TAD and sample time requirements.

FIGURE 20-3: A/D CONVERTER ANALOG INPUT MODEL



Legend: CPIN = Input Capacitance
VT = Threshold Voltage
I leakage = Leakage Current at the pin due to various junctions
RIC = Interconnect Resistance
RSS = Sampling Switch Resistance
CHOLD = Sample/Hold Capacitance (from DAC)

Note: CPIN value depends on device package and is not tested. Effect of CPIN negligible if Rs ≤5 kΩ

TABLE 20-2: ADC REGISTER MAP(1)

SFR Name	Addr.	Bit 15	Bit 14	Bit 13	Bit 12	Bit 11	Bit 10	Bit 9	Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset State
ADCBUF0	0280	—	—	—	—	—	—	ADC Data Buffer 0										0000 00uu uuuu uuuu
ADCBUF1	0282	—	—	—	—	—	—	ADC Data Buffer 1										0000 00uu uuuu uuuu
ADCBUF2	0284	—	—	—	—	—	—	ADC Data Buffer 2										0000 00uu uuuu uuuu
ADCBUF3	0286	—	—	—	—	—	—	ADC Data Buffer 3										0000 00uu uuuu uuuu
ADCBUF4	0288	—	—	—	—	—	—	ADC Data Buffer 4										0000 00uu uuuu uuuu
ADCBUF5	028A	—	—	—	—	—	—	ADC Data Buffer 5										0000 00uu uuuu uuuu
ADCBUF6	028C	—	—	—	—	—	—	ADC Data Buffer 6										0000 00uu uuuu uuuu
ADCBUF7	028E	—	—	—	—	—	—	ADC Data Buffer 7										0000 00uu uuuu uuuu
ADCBUF8	0290	—	—	—	—	—	—	ADC Data Buffer 8										0000 00uu uuuu uuuu
ADCBUF9	0292	—	—	—	—	—	—	ADC Data Buffer 9										0000 00uu uuuu uuuu
ADCBUFA	0294	—	—	—	—	—	—	ADC Data Buffer 10										0000 00uu uuuu uuuu
ADCBUFB	0296	—	—	—	—	—	—	ADC Data Buffer 11										0000 00uu uuuu uuuu
ADCBUFC	0298	—	—	—	—	—	—	ADC Data Buffer 12										0000 00uu uuuu uuuu
ADCBUFD	029A	—	—	—	—	—	—	ADC Data Buffer 13										0000 00uu uuuu uuuu
ADCBUFE	029C	—	—	—	—	—	—	ADC Data Buffer 14										0000 00uu uuuu uuuu
ADCBUFF	029E	—	—	—	—	—	—	ADC Data Buffer 15										0000 00uu uuuu uuuu
ADCON1	02A0	ADON	—	ADSIDL	—	—	—	FORM<1:0>		SSRC<2:0>			—	SIMSAM	ASAM	SAMP	DONE	0000 0000 0000 0000
ADCON2	02A2	VCFG<2:0>			—	—	CSCNA	CHPS<1:0>		BUFS	—	SMPI<3:0>				BUFM	ALTS	0000 0000 0000 0000
ADCON3	02A4	—	—	—	SAMC<4:0>					ADRC	—	ADCS<5:0>						0000 0000 0000 0000
ADCHS	02A6	CH123NB<1:0>		CH123SB	CH0NB	CH0SB<3:0>				CH123NA<1:0>		CH123SA	CH0NA	CH0SA<3:0>				0000 0000 0000 0000
ADPCFG	02A8	—	—	—	—	—	—	—	PCFG8(2)	PCFG7(2)	PCFG6(2)	PCFG5	PCFG4	PCFG3	PCFG2	PCFG1	PCFG0	0000 0000 0000 0000
ADCSSL	02AA	—	—	—	—	—	—	—	CSSL8(2)	CSSL7(2)	CSSL6(2)	CSSL5	CSSL4	CSSL3	CSSL2	CSSL1	CSSL0	0000 0000 0000 0000

Legend: u = uninitialized bit; — = unimplemented bit, read as '0'

Note 1: Refer to the "*dsPIC30F Family Reference Manual*" (DS70046) for descriptions of register bit fields.

2: These bits are not available on dsPIC30F4012 devices.

TABLE 22-2: INSTRUCTION SET OVERVIEW (CONTINUED)

Base Instr #	Assembly Mnemonic	Assembly Syntax		Description	# of words	# of cycles	Status Flags Affected
48	MPY	MPY	Wm*Wn,Acc,Wx,Wxd,Wy,Wyd	Multiply Wm by Wn to Accumulator	1	1	OA, OB, OAB, SA, SB, SAB
		MPY	Wm*Wm,Acc,Wx,Wxd,Wy,Wyd	Square Wm to Accumulator	1	1	OA, OB, OAB, SA, SB, SAB
49	MPY.N	MPY.N	Wm*Wn,Acc,Wx,Wxd,Wy,Wyd	-(Multiply Wm by Wn) to Accumulator	1	1	None
50	MSC	MSC	Wm*Wm,Acc,Wx,Wxd,Wy,Wyd, AWB	Multiply and Subtract from Accumulator	1	1	OA, OB, OAB, SA, SB, SAB
51	MUL	MUL.SS	Wb,Ws,Wnd	{Wnd+1, Wnd} = signed(Wb) * signed(Ws)	1	1	None
		MUL.SU	Wb,Ws,Wnd	{Wnd+1, Wnd} = signed(Wb) * unsigned(Ws)	1	1	None
		MUL.US	Wb,Ws,Wnd	{Wnd+1, Wnd} = unsigned(Wb) * signed(Ws)	1	1	None
		MUL.UU	Wb,Ws,Wnd	{Wnd+1, Wnd} = unsigned(Wb) * unsigned(Ws)	1	1	None
		MUL.SU	Wb,#lit5,Wnd	{Wnd+1, Wnd} = signed(Wb) * unsigned(lit5)	1	1	None
		MUL.UU	Wb,#lit5,Wnd	{Wnd+1, Wnd} = unsigned(Wb) * unsigned(lit5)	1	1	None
		MUL	f	W3:W2 = f * WREG	1	1	None
52	NEG	NEG	Acc	Negate Accumulator	1	1	OA, OB, OAB, SA, SB, SAB
		NEG	f	f = $\bar{f}$ + 1	1	1	C, DC, N, OV, Z
		NEG	f,WREG	WREG = $\bar{f}$ + 1	1	1	C, DC, N, OV, Z
		NEG	Ws,Wd	Wd = $\overline{Ws}$ + 1	1	1	C, DC, N, OV, Z
53	NOP	NOP		No Operation	1	1	None
		NOPR		No Operation	1	1	None
54	POP	POP	f	Pop f from Top-of-Stack (TOS)	1	1	None
		POP	Wdo	Pop from Top-of-Stack (TOS) to Wdo	1	1	None
		POP.D	Wnd	Pop from Top-of-Stack (TOS) to W(nd):W(nd +1)	1	2	None
		POP.S		Pop Shadow Registers	1	1	All
55	PUSH	PUSH	f	Push f to Top-of-Stack (TOS)	1	1	None
		PUSH	Wso	Push Wso to Top-of-Stack (TOS)	1	1	None
		PUSH.D	Wns	Push W(ns):W(ns + 1) to Top-of-Stack (TOS)	1	2	None
		PUSH.S		Push Shadow Registers	1	1	None
56	PWRSAV	PWRSAV	#lit1	Go into Sleep or Idle mode	1	1	WDTO, Sleep
57	RCALL	RCALL	Expr	Relative Call	1	2	None
		RCALL	Wn	Computed Call	1	2	None
58	REPEAT	REPEAT	#lit14	Repeat Next Instruction lit14 + 1 time	1	1	None
		REPEAT	Wn	Repeat Next Instruction (Wn) + 1 time	1	1	None
59	RESET	RESET		Software device Reset	1	1	None
60	RETFIE	RETFIE		Return from interrupt	1	3 (2)	None
61	RETLW	RETLW	#lit10,Wn	Return with literal in Wn	1	3 (2)	None
62	RETURN	RETURN		Return from Subroutine	1	3 (2)	None
63	RLC	RLC	f	f = Rotate Left through Carry f	1	1	C, N, Z
		RLC	f,WREG	WREG = Rotate Left through Carry f	1	1	C, N, Z
		RLC	Ws,Wd	Wd = Rotate Left through Carry Ws	1	1	C, N, Z
64	RLNC	RLNC	f	f = Rotate Left (No Carry) f	1	1	N, Z
		RLNC	f,WREG	WREG = Rotate Left (No Carry) f	1	1	N, Z
		RLNC	Ws,Wd	Wd = Rotate Left (No Carry) Ws	1	1	N, Z
65	RRC	RRC	f	f = Rotate Right through Carry f	1	1	C, N, Z
		RRC	f,WREG	WREG = Rotate Right through Carry f	1	1	C, N, Z
		RRC	Ws,Wd	Wd = Rotate Right through Carry Ws	1	1	C, N, Z

23.7 MPLAB SIM Software Simulator

The MPLAB SIM Software Simulator allows code development in a PC-hosted environment by simulating the PIC MCUs and dsPIC® DSCs on an instruction level. On any given instruction, the data areas can be examined or modified and stimuli can be applied from a comprehensive stimulus controller. Registers can be logged to files for further run-time analysis. The trace buffer and logic analyzer display extend the power of the simulator to record and track program execution, actions on I/O, most peripherals and internal registers.

The MPLAB SIM Software Simulator fully supports symbolic debugging using the MPLAB C Compilers, and the MPASM and MPLAB Assemblers. The software simulator offers the flexibility to develop and debug code outside of the hardware laboratory environment, making it an excellent, economical software development tool.

23.8 MPLAB REAL ICE In-Circuit Emulator System

MPLAB REAL ICE In-Circuit Emulator System is Microchip's next generation high-speed emulator for Microchip Flash DSC and MCU devices. It debugs and programs PIC® Flash MCUs and dsPIC® Flash DSCs with the easy-to-use, powerful graphical user interface of the MPLAB Integrated Development Environment (IDE), included with each kit.

The emulator is connected to the design engineer's PC using a high-speed USB 2.0 interface and is connected to the target with either a connector compatible with in-circuit debugger systems (RJ11) or with the new high-speed, noise tolerant, Low-Voltage Differential Signal (LVDS) interconnection (CAT5).

The emulator is field upgradable through future firmware downloads in MPLAB IDE. In upcoming releases of MPLAB IDE, new devices will be supported, and new features will be added. MPLAB REAL ICE offers significant advantages over competitive emulators including low-cost, full-speed emulation, run-time variable watches, trace analysis, complex breakpoints, a ruggedized probe interface and long (up to three meters) interconnection cables.

23.9 MPLAB ICD 3 In-Circuit Debugger System

MPLAB ICD 3 In-Circuit Debugger System is Microchip's most cost effective high-speed hardware debugger/programmer for Microchip Flash Digital Signal Controller (DSC) and microcontroller (MCU) devices. It debugs and programs PIC® Flash microcontrollers and dsPIC® DSCs with the powerful, yet easy-to-use graphical user interface of MPLAB Integrated Development Environment (IDE).

The MPLAB ICD 3 In-Circuit Debugger probe is connected to the design engineer's PC using a high-speed USB 2.0 interface and is connected to the target with a connector compatible with the MPLAB ICD 2 or MPLAB REAL ICE systems (RJ-11). MPLAB ICD 3 supports all MPLAB ICD 2 headers.

23.10 PICkit 3 In-Circuit Debugger/ Programmer and PICkit 3 Debug Express

The MPLAB PICkit 3 allows debugging and programming of PIC® and dsPIC® Flash microcontrollers at a most affordable price point using the powerful graphical user interface of the MPLAB Integrated Development Environment (IDE). The MPLAB PICkit 3 is connected to the design engineer's PC using a full speed USB interface and can be connected to the target via an Microchip debug (RJ-11) connector (compatible with MPLAB ICD 3 and MPLAB REAL ICE). The connector uses two device I/O pins and the reset line to implement in-circuit debugging and In-Circuit Serial Programming™.

The PICkit 3 Debug Express include the PICkit 3, demo board and microcontroller, hookup cables and CDROM with user's guide, lessons, tutorial, compiler and MPLAB IDE software.

TABLE 24-13: EXTERNAL CLOCK TIMING REQUIREMENTS

AC CHARACTERISTICS			Standard Operating Conditions: 2.5V to 5.5V (unless otherwise stated) Operating temperature -40°C ≤TA ≤+85°C for Industrial -40°C ≤TA ≤+125°C for Extended				
Param No.	**Symbol**	**Characteristic**	**Min**	**Typ(1)**	**Max**	**Units**	**Conditions**
OS10	FOSC	External CLKI Frequency (external clocks allowed only in EC mode)(2)	DC 4 4 4	— — — —	40 10 10 7.5	MHz MHz MHz MHz	EC EC with 4x PLL EC with 8x PLL EC with 16x PLL
		Oscillator Frequency(2)	DC 0.4 4 4 4 4 10 31 — —	— — — — — — — — 7.37 512	4 4 10 10 10 7.5 25 33 — —	MHz MHz MHz MHz MHz MHz MHz kHz MHz kHz	RC XTL XT XT with 4x PLL XT with 8x PLL XT with 16x PLL HS LP FRC internal LPRC internal
OS20	TOSC	TOSC = 1/FOSC	—	—	—	—	See parameter OS10 for FOSC value
OS25	TCY	Instruction Cycle Time(2,3)	33	—	DC	ns	See Table 24-16
OS30	TosL, TosH	External Clock in (OSC1) High or Low Time(2)	.45 x TOSC	—	—	ns	EC
OS31	TosR, TosF	External Clock in (OSC1) Rise or Fall Time(2)	—	—	20	ns	EC
OS40	TckR	CLKO Rise Time(2,4)	—	—	—	ns	See parameter DO31
OS41	TckF	CLKO Fall Time(2,4)	—	—	—	ns	See parameter DO32

Note 1: Data in “Typ” column is at 5V, 25°C unless otherwise stated. Parameters are for design guidance only and are not tested.

2: These parameters are characterized but not tested in manufacturing.

3: Instruction cycle period (TCY) equals four times the input oscillator time base period. All specified values are based on characterization data for that particular oscillator type under standard operating conditions with the device executing code. Exceeding these specified limits may result in an unstable oscillator operation and/or higher than expected current consumption. All devices are tested to operate at “min.” values with an external clock applied to the OSC1/CLKI pin. When an external clock input is used, the “Max.” cycle time limit is “DC” (no clock) for all devices.

4: Measurements are taken in EC or ERC modes. The CLKO signal is measured on the OSC2 pin. CLKO is low for the Q1-Q2 period (1/2 TCY) and high for the Q3-Q4 period (1/2 TCY).

TABLE 24-14: PLL CLOCK TIMING SPECIFICATIONS (V_{DD} = 2.5 TO 5.5V)

AC CHARACTERISTICS			Standard Operating Conditions: 2.5V to 5.5V (unless otherwise stated) Operating temperature -40°C ≤TA ≤+85°C for Industrial -40°C ≤TA ≤+125°C for Extended				
Param No.	**Symbol**	**Characteristic(1)**	**Min**	**Typ(2)**	**Max**	**Units**	**Conditions**
OS50	FPLLI	PLL Input Frequency Range(2)	4 4 4 4 4 4	— — — — — —	10 10 7.5(3) 10 10 7.5(3)	MHz MHz MHz MHz MHz MHz	EC with 4x PLL EC with 8x PLL EC with 16x PLL XT with 4x PLL XT with 8x PLL XT with 16x PLL
OS51	FSYS	On-Chip PLL Output(2)	16	—	120	MHz	EC, XT with PLL
OS52	TLOC	PLL Start-up Time (Lock Time)	—	20	50	µs	

Note 1: These parameters are characterized but not tested in manufacturing.

2: Data in "Typ" column is at 5V, 25°C unless otherwise stated. Parameters are for design guidance only and are not tested.

3: Limited by device operating frequency range.

TABLE 24-15: PLL JITTER

AC CHARACTERISTICS		Standard Operating Conditions: 2.5V to 5.5V (unless otherwise stated) Operating temperature -40°C ≤TA ≤+85°C for Industrial -40°C ≤TA ≤+125°C for Extended					
Param No.	**Characteristic**	**Min**	**Typ(1)**	**Max**	**Units**	**Conditions**	
OS61	x4 PLL	—	0.251	0.413	%	-40°C ≤TA ≤+85°C	VDD = 3.0 to 3.6V
		—	0.251	0.413	%	-40°C ≤TA ≤+125°C	VDD = 3.0 to 3.6V
		—	0.256	0.47	%	-40°C ≤TA ≤+85°C	VDD = 4.5 to 5.5V
		—	0.256	0.47	%	-40°C ≤TA ≤+125°C	VDD = 4.5 to 5.5V
	x8 PLL	—	0.355	0.584	%	-40°C ≤TA ≤+85°C	VDD = 3.0 to 3.6V
		—	0.355	0.584	%	-40°C ≤TA ≤+125°C	VDD = 3.0 to 3.6V
		—	0.362	0.664	%	-40°C ≤TA ≤+85°C	VDD = 4.5 to 5.5V
		—	0.362	0.664	%	-40°C ≤TA ≤+125°C	VDD = 4.5 to 5.5V
	x16 PLL	—	0.67	0.92	%	-40°C ≤TA ≤+85°C	VDD = 3.0 to 3.6V
		—	0.632	0.956	%	-40°C ≤TA ≤+85°C	VDD = 4.5 to 5.5V
		—	0.632	0.956	%	-40°C ≤TA ≤+125°C	VDD = 4.5 to 5.5V

Note 1: These parameters are characterized but not tested in manufacturing.

TABLE 24-17: AC CHARACTERISTICS: INTERNAL FRC ACCURACY

AC CHARACTERISTICS		Standard Operating Conditions: 2.5V to 5.5V (unless otherwise stated) Operating temperature -40°C ≤TA ≤+85°C for Industrial -40°C ≤TA ≤+125°C for Extended					
Param No.	**Characteristic**	**Min**	**Typ**	**Max**	**Units**	**Conditions**	
Internal FRC Accuracy @ FRC Freq. = 7.37 MHz(1)							
OS63	FRC	—	—	±2.00	%	-40°C ≤TA ≤+85°C	VDD = 3.0-5.5V
		—	—	±5.00	%	-40°C ≤TA ≤+125°C	VDD = 3.0-5.5V

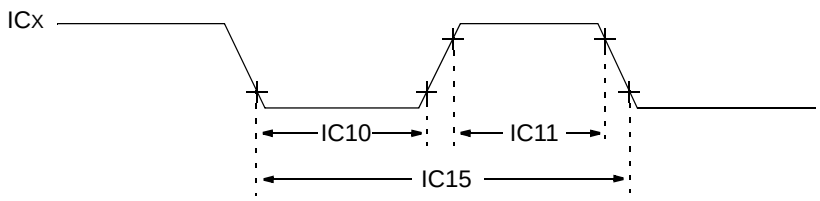
Note 1: Frequency calibrated at 7.372 MHz ±2%, 25°C and 5V. TUN<3:0> bits can be used to compensate for temperature drift.

TABLE 24-18: AC CHARACTERISTICS: INTERNAL LPRC ACCURACY

AC CHARACTERISTICS		Standard Operating Conditions: 2.5V to 5.5V (unless otherwise stated) Operating temperature -40°C ≤TA ≤+85°C for Industrial -40°C ≤TA ≤+125°C for Extended				
Param No.	**Characteristic**	**Min**	**Typ**	**Max**	**Units**	**Conditions**
LPRC @ Freq. = 512 kHz(1)						
OS65A		-50	—	+50	%	VDD = 5.0V, ±10%
OS65B		-60	—	+60	%	VDD = 3.3V, ±10%
OS65C		-70	—	+70	%	VDD = 2.5V

Note 1: Change of LPRC frequency as VDD changes.

FIGURE 24-9: INPUT CAPTURE TIMING CHARACTERISTICS



Note: Refer to Figure 24-2 for load conditions.

TABLE 24-26: INPUT CAPTURE TIMING REQUIREMENTS

AC CHARACTERISTICS			**Standard Operating Conditions: 2.5V to 5.5V (unless otherwise stated)** Operating temperature -40°C ≤TA ≤+85°C for Industrial -40°C ≤TA ≤+125°C for Extended				
Param No.	**Symbol**	**Characteristic(1)**		**Min**	**Max**	**Units**	**Conditions**
IC10	TccL	ICx Input Low Time	No prescaler	0.5 TCY + 20	—	ns	
			With prescaler	10	—	ns	
IC11	TccH	ICx Input High Time	No prescaler	0.5 TCY + 20	—	ns	
			With prescaler	10	—	ns	
IC15	TccP	ICx Input Period		(2 TCY + 40)/N	—	ns	N = prescale value (1, 4, 16)

Note 1: These parameters are characterized but not tested in manufacturing.

FIGURE 24-14: QEA/QEB INPUT CHARACTERISTICS

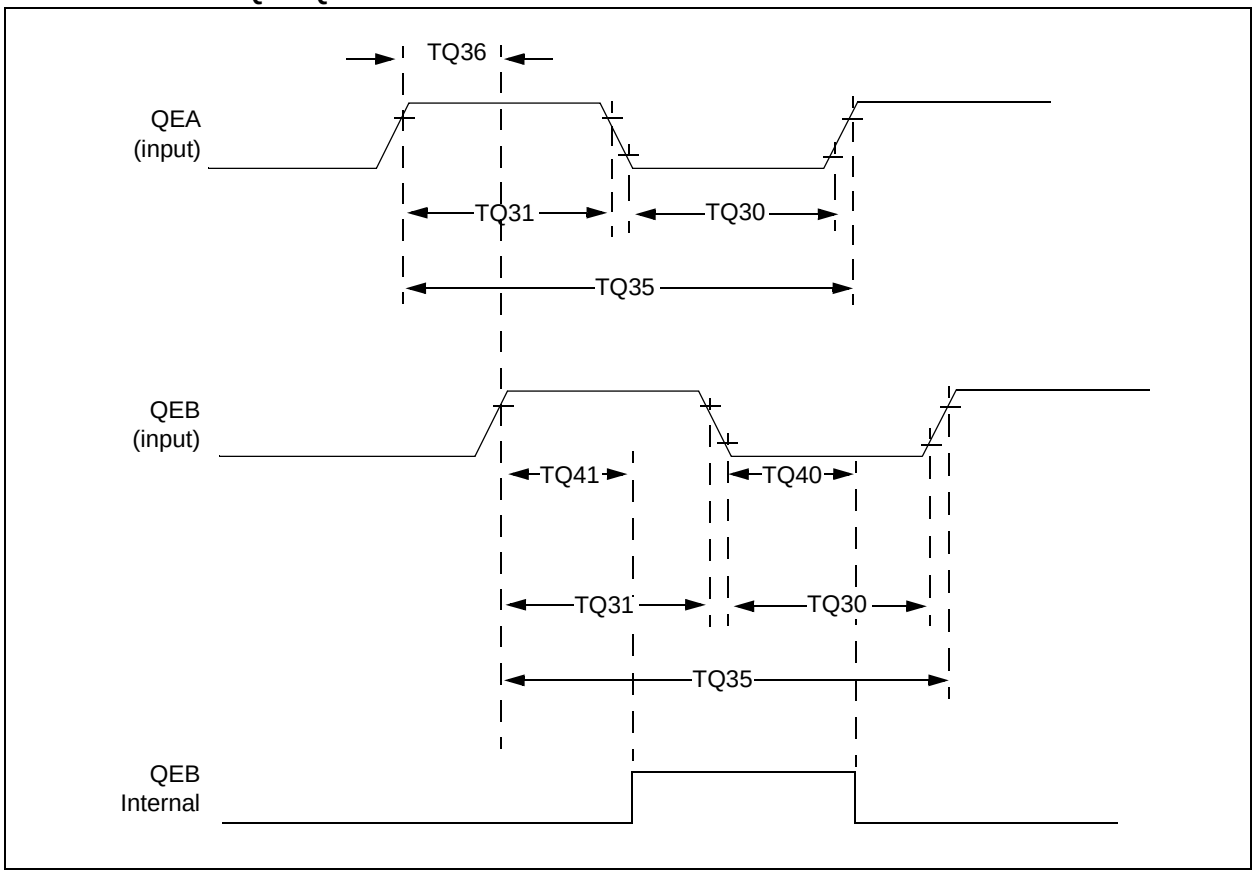


TABLE 24-30: QUADRATURE DECODER TIMING REQUIREMENTS

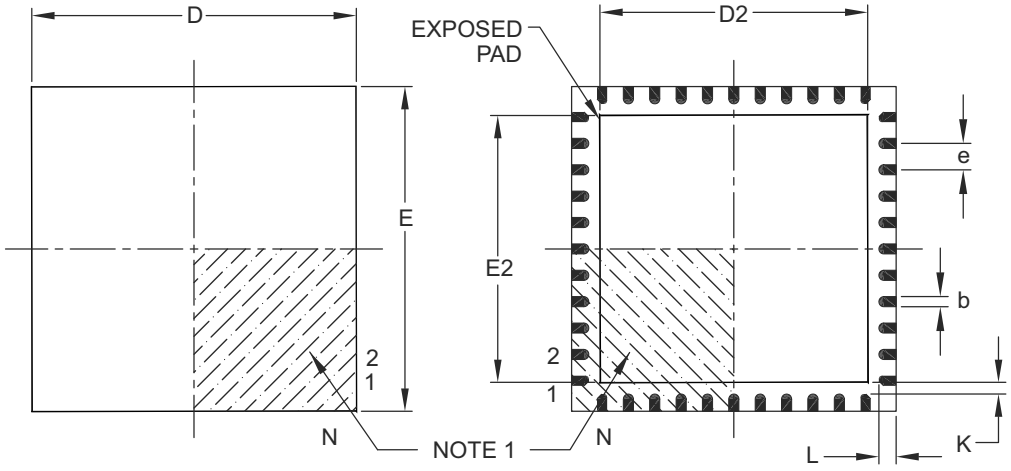
AC CHARACTERISTICS			Standard Operating Conditions: 2.5V to 5.5V (unless otherwise stated) Operating temperature -40°C ≤TA ≤+85°C for Industrial -40°C ≤TA ≤+125°C for Extended			
Param No.	**Symbol**	**Characteristic(1)**	**Typ(2)**	**Max**	**Units**	**Conditions**
TQ30	TQUL	Quadrature Input Low Time	6 TCY	—	ns	
TQ31	TQUH	Quadrature Input High Time	6 TCY	—	ns	
TQ35	TQUIN	Quadrature Input Period	12 TCY	—	ns	
TQ36	TQUP	Quadrature Phase Period	3 TCY	—	ns	
TQ40	TQUFL	Filter Time to Recognize Low, with Digital Filter	3 * N * TCY	—	ns	N = 1, 2, 4, 16, 32, 64, 128 and 256 **(Note 2)**
TQ41	TQUFH	Filter Time to Recognize High, with Digital Filter	3 * N * TCY	—	ns	N = 1, 2, 4, 16, 32, 64, 128 and 256 **(Note 2)**

Note 1: These parameters are characterized but not tested in manufacturing.

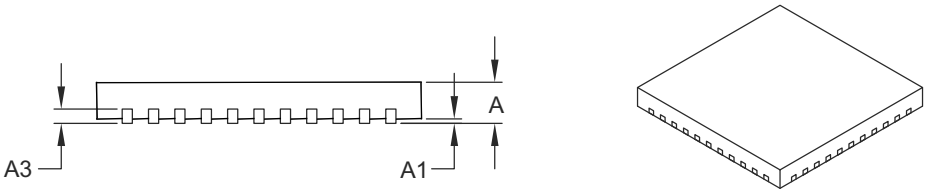
2: N = Index Channel Digital Filter Clock Divide Select bits. Refer to **Section 16. "Quadrature Encoder Interface (QEI)"** in the "*dsPIC30F Family Reference Manual*" (DS70046).

44-Lead Plastic Quad Flat, No Lead Package (ML) – 8x8 mm Body [QFN]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



TOP VIEW **BOTTOM VIEW**



Units		MILLIMETERS		
Dimension Limits		MIN	NOM	MAX
Number of Pins	N		44	
Pitch	e		0.65 BSC	
Overall Height	A	0.80	0.90	1.00
Standoff	A1	0.00	0.02	0.05
Contact Thickness	A3		0.20 REF	
Overall Width	E		8.00 BSC	
Exposed Pad Width	E2	6.30	6.45	6.80
Overall Length	D		8.00 BSC	
Exposed Pad Length	D2	6.30	6.45	6.80
Contact Width	b	0.25	0.30	0.38
Contact Length	L	0.30	0.40	0.50
Contact-to-Exposed Pad	K	0.20	–	–

Notes:

1. Pin 1 visual index feature may vary, but must be located within the hatched area.
2. Package is saw singulated.
3. Dimensioning and tolerancing per ASME Y14.5M.

BSC: Basic Dimension. Theoretically exact value shown without tolerances.
REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-103B

NOTES: