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Understanding [Embedded - Microprocessors](#)

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of [Embedded - Microprocessors](#)

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details

Product Status	Obsolete
Core Processor	MPC8xx
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	80MHz
Co-Processors/DSP	Communications; CPM, Security; SEC
RAM Controllers	DRAM
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10Mbps (3), 10/100Mbps (2)
SATA	-
USB	USB 2.0 (1)
Voltage - I/O	3.3V
Operating Temperature	0°C ~ 95°C (TA)
Security Features	Cryptography
Package / Case	357-BBGA
Supplier Device Package	357-PBGA (25x25)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/kmpc885vr80

- 1.8-V core and 3.3-V I/O operation
- The MPC885/MPC880 comes in a 357-pin ball grid array (PBGA) package

The MPC885 block diagram is shown in [Figure 1](#).

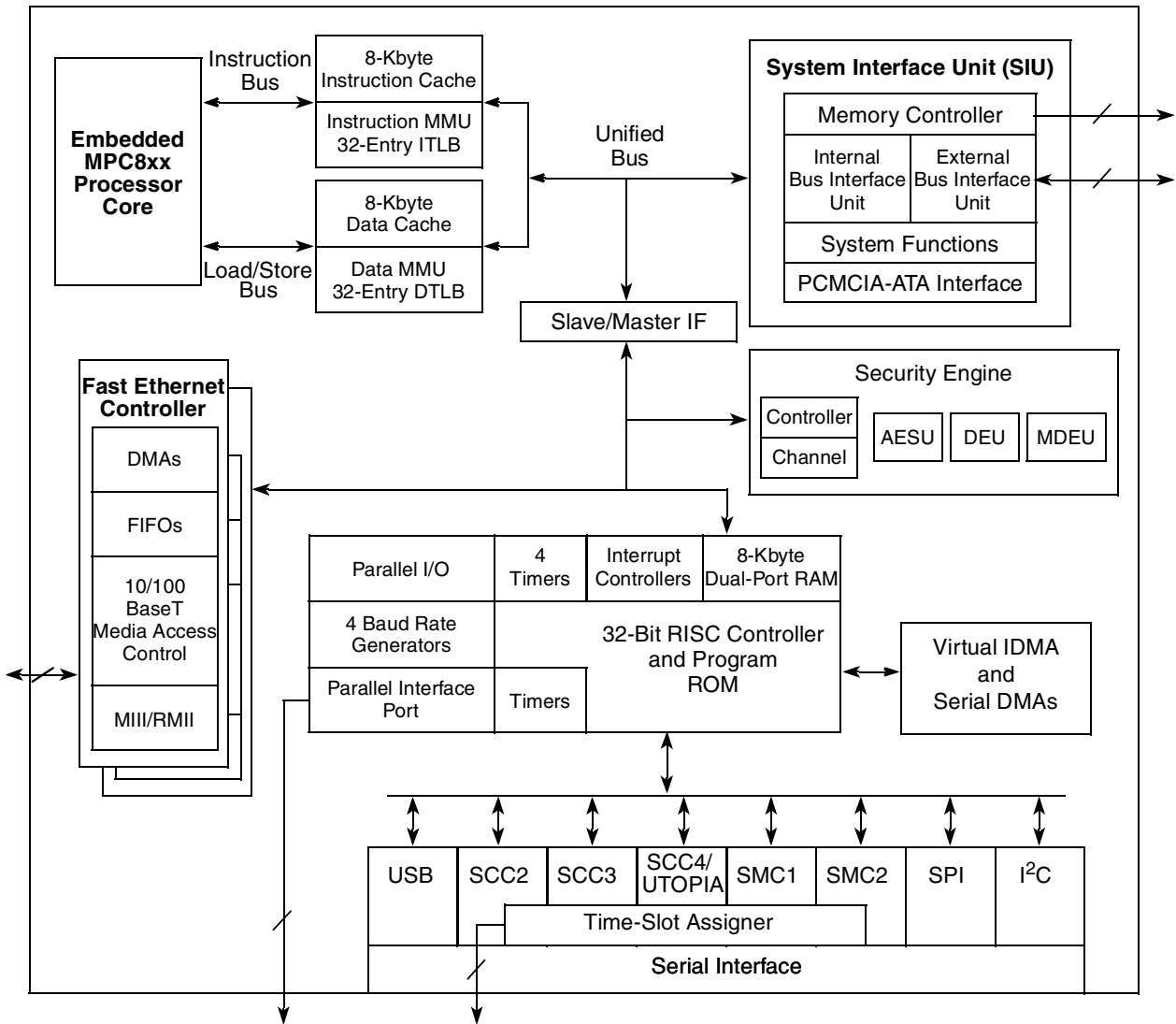


Figure 1. MPC885 Block Diagram

Table 6. DC Electrical Specifications (continued)

Characteristic	Symbol	Min	Max	Unit
Output high voltage, $I_{OH} = -2.0$ mA, except XTAL and open-drain pins	V_{OH}	2.4	—	V
Output low voltage $I_{OL} = 2.0$ mA (CLKOUT) $I_{OL} = 3.2$ mA ⁵ $I_{OL} = 5.3$ mA ⁶ $I_{OL} = 7.0$ mA (TXD1/PA14, TXD2/PA12) $I_{OL} = 8.9$ mA ($\overline{TS}$, $\overline{TA}$, $\overline{TEA}$, $\overline{BI}$, $\overline{BB}$, $\overline{HRESET}$, $\overline{SRESET}$)	V_{OL}	—	0.5	V

¹ The difference between V_{DDL} and V_{DDSYN} cannot be more than 100 mV.

² The signals PA[0:15], PB[14:31], PC[4:15], PD[3:15], PE(14:31), TDI, TDO, TCK, $\overline{TRST}$, TMS, MII1_TXEN, MII_MDIO are 5-V tolerant. The minimum voltage is still 2.0 V.

³ V_{IL} (max) for the I²C interface is 0.8 V rather than the 1.5 V as specified in the I²C standard.

⁴ Input capacitance is periodically sampled.

⁵ A(0:31), TSIZ0/REG, TSIZ1, D(0:31), $\overline{IRQ6}$, RD/ $\overline{WR}$, $\overline{BURST}$, IP_B(3:7), PA(0:11), PA13, PA15, PB(14:31), PC(4:15), PD(3:15), PE(14:31), MII1_CRS, MII_MDIO, MII1_TXEN, and MII1_COL.

⁶ $\overline{BDIP}/\overline{GPL_B}(5)$, $\overline{BR}$, $\overline{BG}$, $\overline{FRZ}/\overline{IRQ6}$, $\overline{CS}(0:7)$, $\overline{WE}(0:3)$, $\overline{BS_A}(0:3)$, $\overline{GPL_A0}/\overline{GPL_B0}$, $\overline{OE}/\overline{GPL_A1}/\overline{GPL_B1}$, $\overline{GPL_A}(2:3)/\overline{GPL_B}(2:3)/\overline{CS}(2:3)$, UPWAITA/ $\overline{GPL_A4}$, UPWAITB/ $\overline{GPL_B4}$, $\overline{GPL_A5}$, $\overline{ALE_A}$, $\overline{CE1_A}$, $\overline{CE2_A}$, OP(0:3), and BADDR(28:30).

7 Thermal Calculation and Measurement

For the following discussions, $P_D = (V_{DDL} \times I_{DDL}) + P_{I/O}$, where $P_{I/O}$ is the power dissipation of the I/O drivers.

NOTE

The V_{DDSYN} power dissipation is negligible.

7.1 Estimation with Junction-to-Ambient Thermal Resistance

An estimation of the chip junction temperature, T_J , in °C can be obtained from the following equation:

$$T_J = T_A + (R_{\theta JA} \times P_D)$$

where:

T_A = ambient temperature (°C)

$R_{\theta JA}$ = package junction-to-ambient thermal resistance (°C/W)

P_D = power dissipation in package

The junction-to-ambient thermal resistance is an industry standard value that provides a quick and easy estimation of thermal performance. However, the answer is only an estimate; test cases have demonstrated that errors of a factor of two (in the quantity $T_J - T_A$) are possible.

If the board temperature is known, an estimate of the junction temperature in the environment can be made using the following equation:

$$T_J = T_B + (R_{\theta JB} \times P_D)$$

where:

$R_{\theta JB}$ = junction-to-board thermal resistance ($^{\circ}\text{C}/\text{W}$)

T_B = board temperature ($^{\circ}\text{C}$)

P_D = power dissipation in package

If the board temperature is known and the heat loss from the package case to the air can be ignored, acceptable predictions of junction temperature can be made. For this method to work, the board and board mounting must be similar to the test board used to determine the junction-to-board thermal resistance, namely a 2s2p (board with a power and a ground plane) and vias attaching the thermal balls to the ground plane.

7.4 Estimation Using Simulation

When the board temperature is not known, a thermal simulation of the application is needed. The simple two resistor model can be used with the thermal simulation of the application [2], or a more accurate and complex model of the package can be used in the thermal simulation.

7.5 Experimental Determination

To determine the junction temperature of the device in the application after prototypes are available, the thermal characterization parameter (Ψ_{JT}) can be used to determine the junction temperature with a measurement of the temperature at the top center of the package case using the following equation:

$$T_J = T_T + (\Psi_{JT} \times P_D)$$

where:

Ψ_{JT} = thermal characterization parameter

T_T = thermocouple temperature on top of package

P_D = power dissipation in package

The thermal characterization parameter is measured per the JESD51-2 specification published by JEDEC using a 40 gauge type T thermocouple epoxied to the top center of the package case. The thermocouple should be positioned so that the thermocouple junction rests on the package. A small amount of epoxy is placed over the thermocouple junction and over about 1 mm of wire extending from the junction. The thermocouple wire is placed flat against the package case to avoid measurement errors caused by the cooling effects of the thermocouple wire.

Table 7. Frequency Ranges for Standard Part Frequencies (1:1 Bus Mode)

Part Frequency	66 MHz		80 MHz	
	Min	Max	Min	Max
Core frequency	40	66.67	40	80
Bus frequency	40	66.67	40	80

Table 8. Frequency Ranges for Standard Part Frequencies (2:1 Bus Mode)

Part Frequency	66 MHz		80 MHz		133 MHz	
	Min	Max	Min	Max	Min	Max
Core frequency	40	66.67	40	80	40	133
Bus frequency	20	33.33	20	40	20	66

[Table 9](#) provides the timings for the MPC885/MPC880 at 33-, 40-, 66-, and 80-MHz bus operation.

The timing for the MPC885/MPC880 bus shown assumes a 50-pF load for maximum delays and a 0-pF load for minimum delays. CLKOUT assumes a 100-pF load for maximum delays and a 50-pF load for minimum delays.

Table 9. Bus Operation Timings

Num	Characteristic	33 MHz		40 MHz		66 MHz		80 MHz		Unit
		Min	Max	Min	Max	Min	Max	Min	Max	
B1	Bus period (CLKOUT), see Table 7	—	—	—	—	—	—	—	—	ns
B1a	EXTCLK to CLKOUT phase skew - If CLKOUT is an integer multiple of EXTCLK, then the rising edge of EXTCLK is aligned with the rising edge of CLKOUT. For a non-integer multiple of EXTCLK, this synchronization is lost, and the rising edges of EXTCLK and CLKOUT have a continuously varying phase skew.	-2	+2	-2	+2	-2	+2	-2	+2	ns
B1b	CLKOUT frequency jitter peak-to-peak	—	1	—	1	—	1	—	1	ns
B1c	Frequency jitter on EXTCLK	—	0.50	—	0.50	—	0.50	—	0.50	%
B1d	CLKOUT phase jitter peak-to-peak for OSCLK ≥ 15 MHz	—	4	—	4	—	4	—	4	ns
	CLKOUT phase jitter peak-to-peak for OSCLK < 15 MHz	—	5	—	5	—	5	—	5	ns
B2	CLKOUT pulse width low (MIN = 0.4 × B1, MAX = 0.6 × B1)	12.1	18.2	10.0	15.0	6.1	9.1	5.0	7.5	ns
B3	CLKOUT pulse width high (MIN = 0.4 × B1, MAX = 0.6 × B1)	12.1	18.2	10.0	15.0	6.1	9.1	5.0	7.5	ns
B4	CLKOUT rise time	—	4.00	—	4.00	—	4.00	—	4.00	ns

Figure 12 provides the timing for the input data controlled by the UPM for data beats where DLT3 = 1 in the UPM RAM words. (This is only the case where data is latched on the falling edge of CLKOUT.)

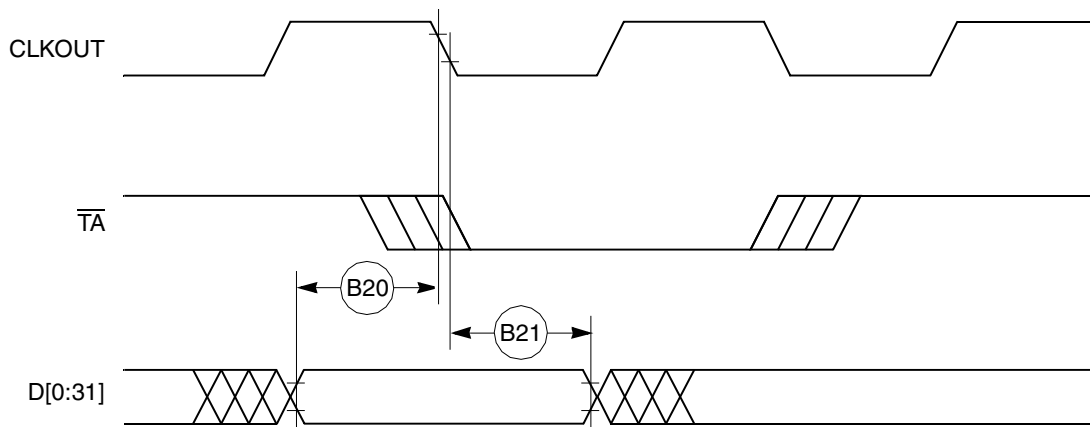


Figure 12. Input Data Timing when Controlled by UPM in the Memory Controller and DLT3 = 1

Figure 13 through Figure 16 provide the timing for the external bus read controlled by various GPCM factors.

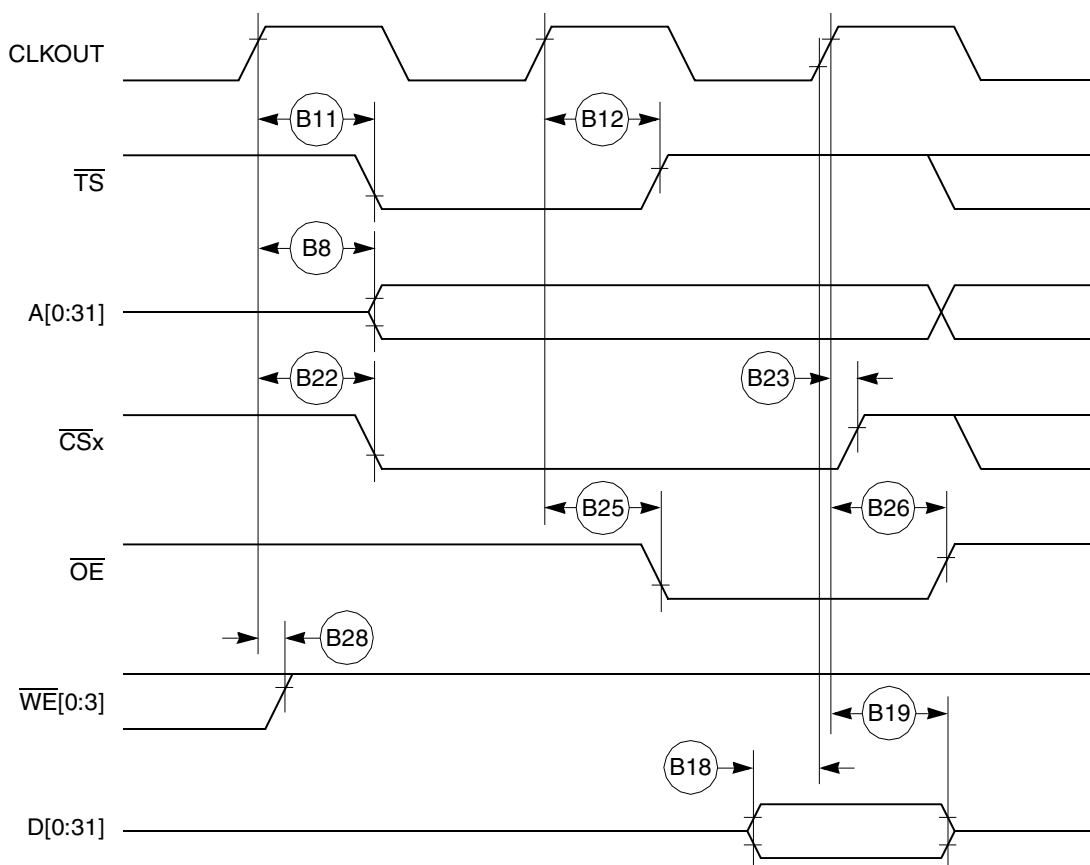


Figure 13. External Bus Read Timing (GPCM Controlled—ACS = 00)

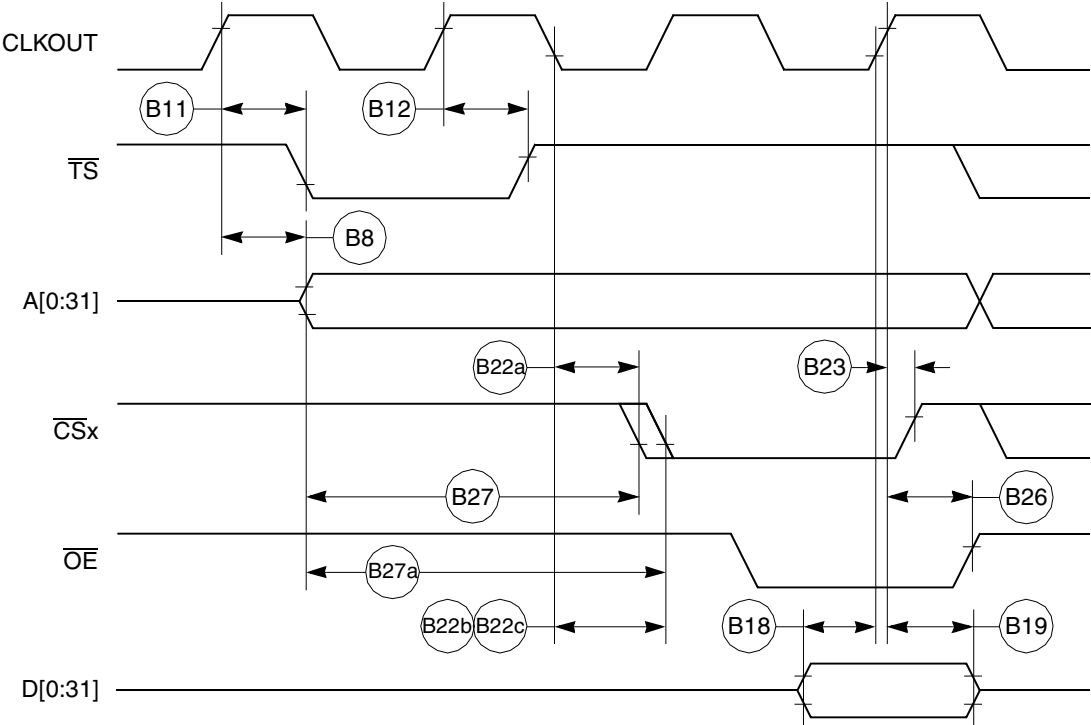


Figure 16. External Bus Read Timing (GPCM Controlled—TRLX = 1, ACS = 10, ACS = 11)

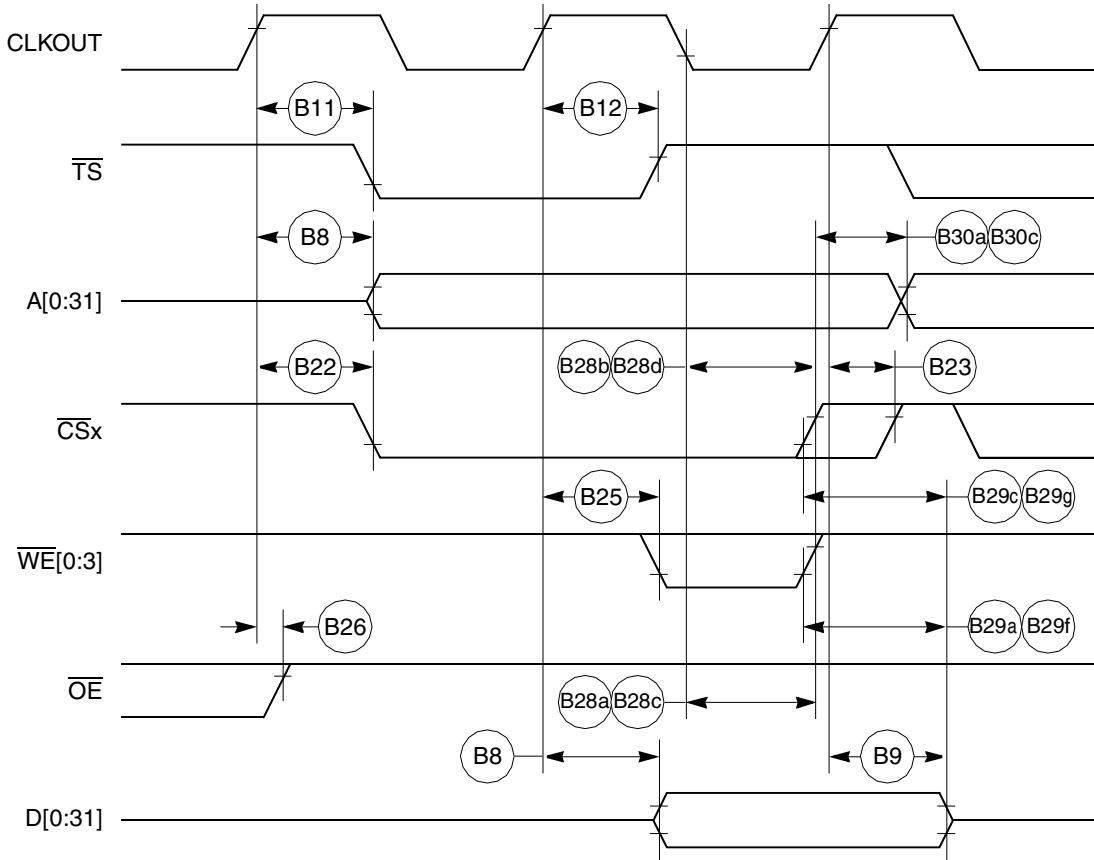


Figure 18. External Bus Write Timing (GPCM Controlled—TRLX = 0, CSNT = 1)

Figure 20 provides the timing for the external bus controlled by the UPM.

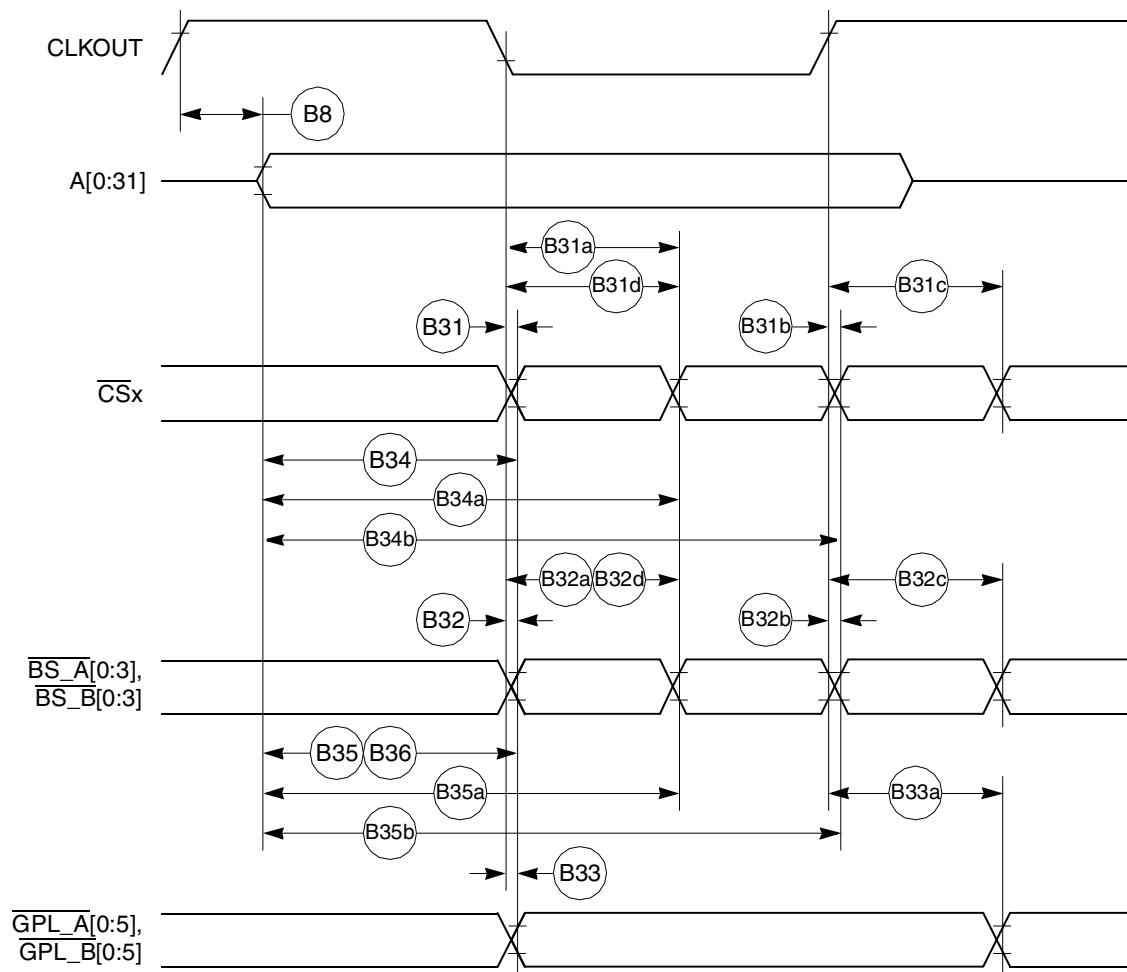


Figure 20. External Bus Timing (UPM-Controlled Signals)

Table 14 shows the reset timing for the MPC885/MPC880.

Table 14. Reset Timing

Num	Characteristic	33 MHz		40 MHz		66 MHz		80 MHz		Unit
		Min	Max	Min	Max	Min	Max	Min	Max	
R69	CLKOUT to $\overline{\text{HRESET}}$ high impedance (MAX = $0.00 \times B1 + 20.00$)	—	20.00	—	20.00	—	20.00	—	20.00	ns
R70	CLKOUT to $\overline{\text{SRESET}}$ high impedance (MAX = $0.00 \times B1 + 20.00$)	—	20.00	—	20.00	—	20.00	—	20.00	ns
R71	$\overline{\text{RSTCONF}}$ pulse width (MIN = $17.00 \times B1$)	515.20	—	425.00	—	257.60	—	212.50	—	ns
R72	—	—	—	—	—	—	—	—	—	—
R73	Configuration data to $\overline{\text{HRESET}}$ rising edge setup time (MIN = $15.00 \times B1 + 50.00$)	504.50	—	425.00	—	277.30	—	237.50	—	ns
R74	Configuration data to $\overline{\text{RSTCONF}}$ rising edge setup time (MIN = $0.00 \times B1 + 350.00$)	350.00	—	350.00	—	350.00	—	350.00	—	ns
R75	Configuration data hold time after $\overline{\text{RSTCONF}}$ negation (MIN = $0.00 \times B1 + 0.00$)	0.00	—	0.00	—	0.00	—	0.00	—	ns
R76	Configuration data hold time after $\overline{\text{HRESET}}$ negation (MIN = $0.00 \times B1 + 0.00$)	0.00	—	0.00	—	0.00	—	0.00	—	ns
R77	$\overline{\text{HRESET}}$ and $\overline{\text{RSTCONF}}$ asserted to data out drive (MAX = $0.00 \times B1 + 25.00$)	—	25.00	—	25.00	—	25.00	—	25.00	ns
R78	$\overline{\text{RSTCONF}}$ negated to data out high impedance (MAX = $0.00 \times B1 + 25.00$)	—	25.00	—	25.00	—	25.00	—	25.00	ns
R79	CLKOUT of last rising edge before chip three-states $\overline{\text{HRESET}}$ to data out high impedance (MAX = $0.00 \times B1 + 25.00$)	—	25.00	—	25.00	—	25.00	—	25.00	ns
R80	DSDI, DSCK setup (MIN = $3.00 \times B1$)	90.90	—	75.00	—	45.50	—	37.50	—	ns
R81	DSDI, DSCK hold time (MIN = $0.00 \times B1 + 0.00$)	0.00	—	0.00	—	0.00	—	0.00	—	ns
R82	$\overline{\text{SRESET}}$ negated to CLKOUT rising edge for DSDI and DSCK sample (MIN = $8.00 \times B1$)	242.40	—	200.00	—	121.20	—	100.00	—	ns

11 IEEE 1149.1 Electrical Specifications

Table 15 provides the JTAG timings for the MPC885/MPC880 shown in Figure 38 through Figure 41.

Table 15. JTAG Timing

Num	Characteristic	All Frequencies		Unit
		Min	Max	
J82	TCK cycle time	100.00	—	ns
J83	TCK clock pulse width measured at 1.5 V	40.00	—	ns
J84	TCK rise and fall times	0.00	10.00	ns
J85	TMS, TDI data setup time	5.00	—	ns
J86	TMS, TDI data hold time	25.00	—	ns
J87	TCK low to TDO data valid	—	27.00	ns
J88	TCK low to TDO data invalid	0.00	—	ns
J89	TCK low to TDO high impedance	—	20.00	ns
J90	$\overline{\text{TRST}}$ assert time	100.00	—	ns
J91	$\overline{\text{TRST}}$ setup time to TCK low	40.00	—	ns
J92	TCK falling edge to output valid	—	50.00	ns
J93	TCK falling edge to output valid out of high impedance	—	50.00	ns
J94	TCK falling edge to output high impedance	—	50.00	ns
J95	Boundary scan input valid to TCK rising edge	50.00	—	ns
J96	TCK rising edge to boundary scan input invalid	50.00	—	ns

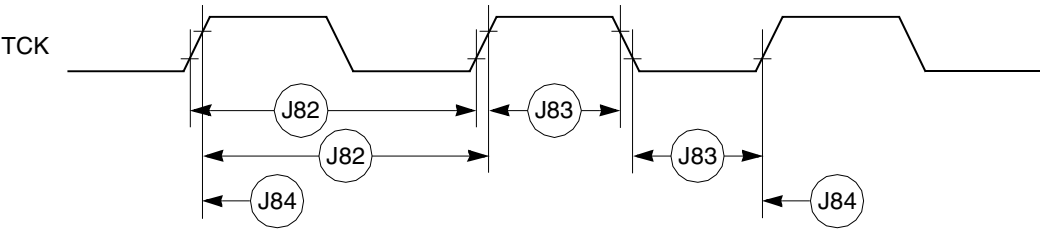


Figure 38. JTAG Test Clock Input Timing

Figure 59 through Figure 61 show the NMSI timings.

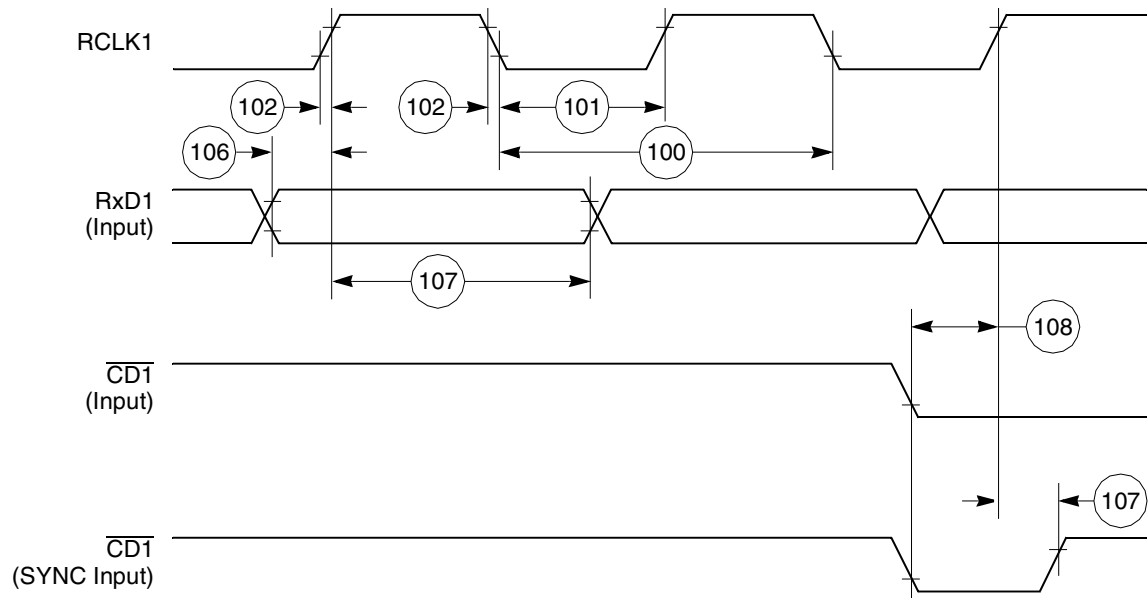


Figure 59. SCC NMSI Receive Timing Diagram

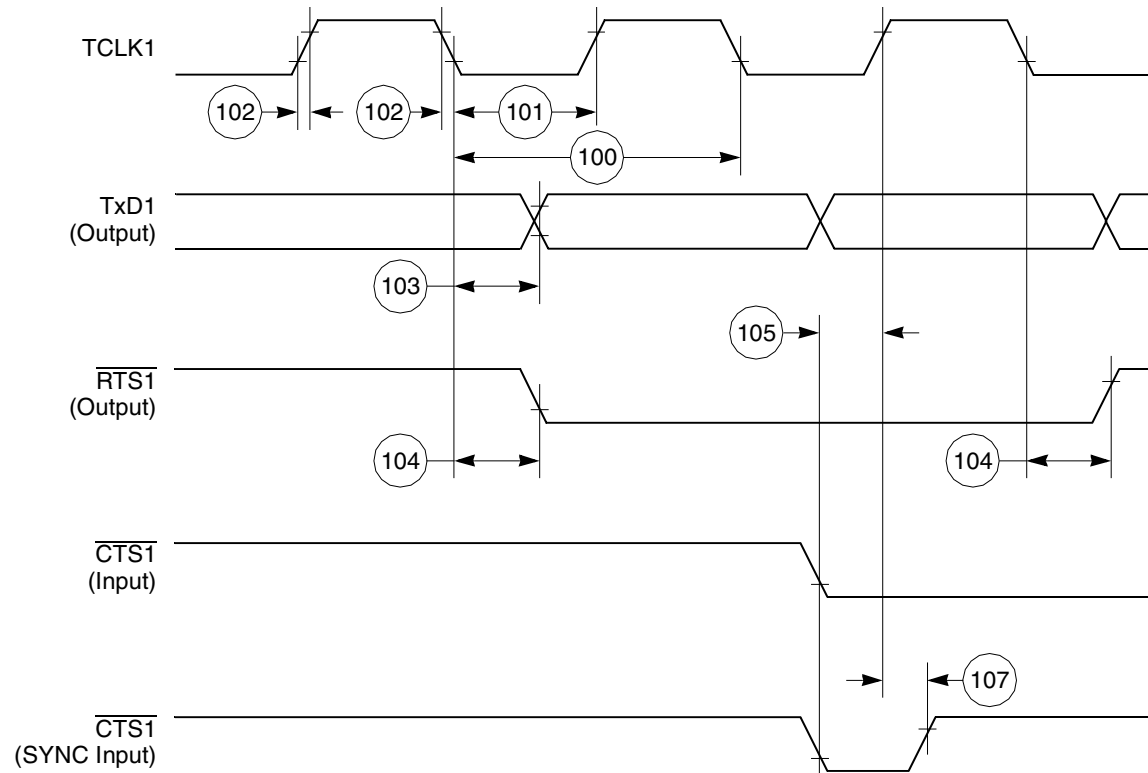


Figure 60. SCC NMSI Transmit Timing Diagram

12.11 SPI Slave AC Electrical Specifications

Table 27 provides the SPI slave timings as shown in Figure 68 and Figure 69.

Table 27. SPI Slave Timing

Num	Characteristic	All Frequencies		Unit
		Min	Max	
170	Slave cycle time	2	—	t_{cyc}
171	Slave enable lead time	15	—	ns
172	Slave enable lag time	15	—	ns
173	Slave clock (SPICLK) high or low time	1	—	t_{cyc}
174	Slave sequential transfer delay (does not require deselect)	1	—	t_{cyc}
175	Slave data setup time (inputs)	20	—	ns
176	Slave data hold time (inputs)	20	—	ns
177	Slave access time	—	50	ns

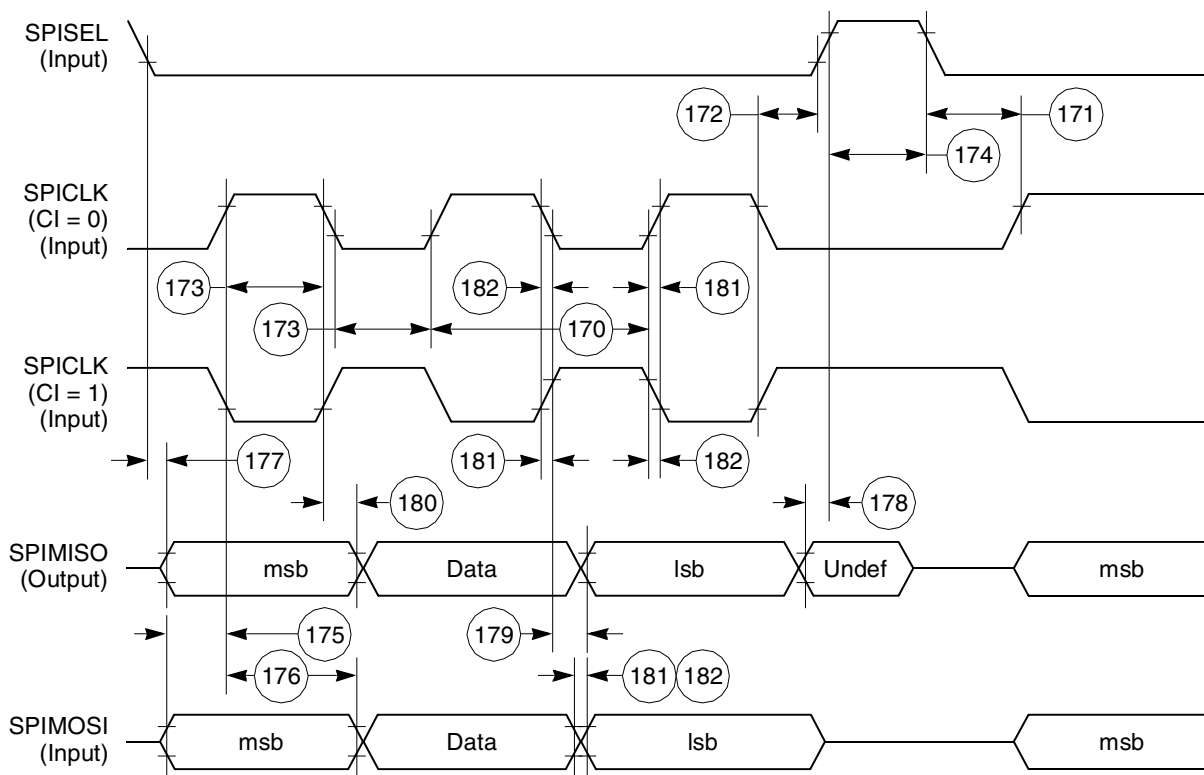


Figure 68. SPI Slave (CP = 0) Timing Diagram

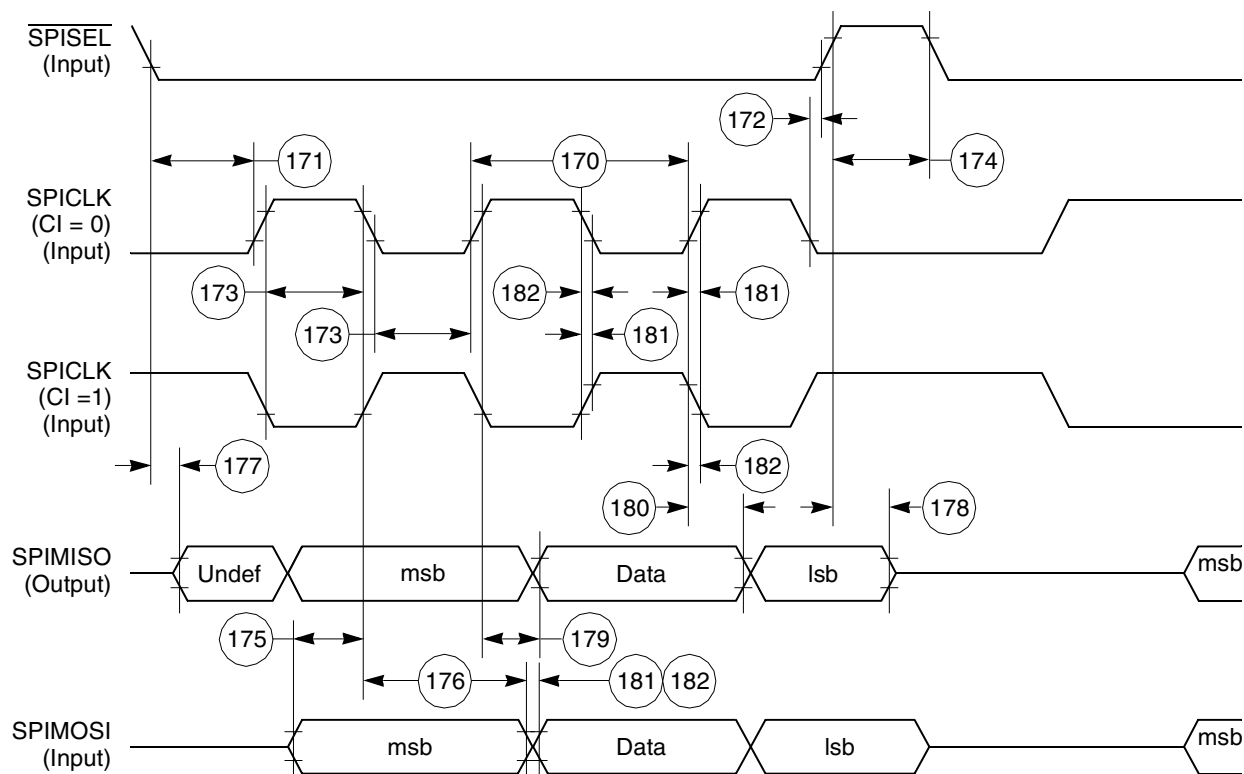


Figure 69. SPI Slave (CP = 1) Timing Diagram

12.12 I²C AC Electrical Specifications

Table 28 provides the I²C (SCL < 100 kHz) timings.

Table 28. I²C Timing (SCL < 100 kHz)

Num	Characteristic	All Frequencies		Unit
		Min	Max	
200	SCL clock frequency (slave)	0	100	kHz
200	SCL clock frequency (master) ¹	1.5	100	kHz
202	Bus free time between transmissions	4.7	—	μs
203	Low period of SCL	4.7	—	μs
204	High period of SCL	4.0	—	μs
205	Start condition setup time	4.7	—	μs
206	Start condition hold time	4.0	—	μs
207	Data hold time	0	—	μs
208	Data setup time	250	—	ns
209	SDL/SCL rise time	—	1	μs

Table 28. I²C Timing (SCL < 100 kHz) (continued)

Num	Characteristic	All Frequencies		Unit
		Min	Max	
210	SDL/SCL fall time	—	300	ns
211	Stop condition setup time	4.7	—	μs

¹ SCL frequency is given by $SCL = BRGCLK_frequency / ((BRG\ register + 3) \times pre_scaler \times 2)$.
The ratio $SyncClk / (BRGCLK/pre_scaler)$ must be greater or equal to 4/1.

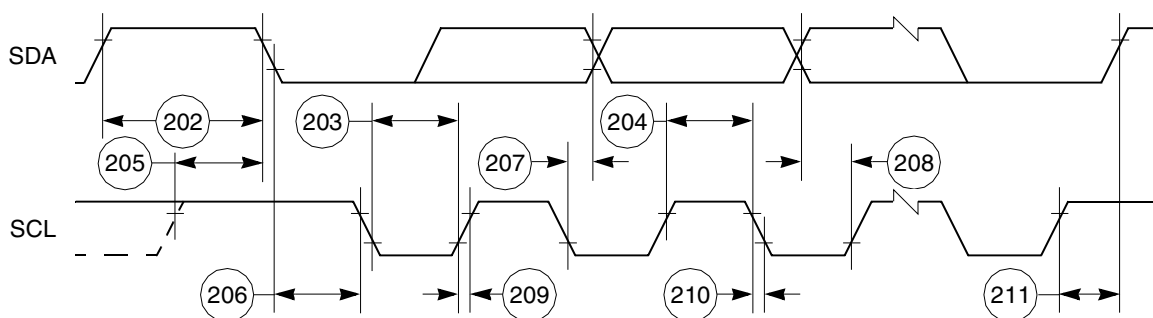
Table 29 provides the I²C (SCL > 100 kHz) timings.

Table 29. I²C Timing (SCL > 100 kHz)

Num	Characteristic	Expression	All Frequencies		Unit
			Min	Max	
200	SCL clock frequency (slave)	fSCL	0	BRGCLK/48	Hz
200	SCL clock frequency (master) ¹	fSCL	BRGCLK/16512	BRGCLK/48	Hz
202	Bus free time between transmissions	—	$1/(2.2 \times fSCL)$	—	s
203	Low period of SCL	—	$1/(2.2 \times fSCL)$	—	s
204	High period of SCL	—	$1/(2.2 \times fSCL)$	—	s
205	Start condition setup time	—	$1/(2.2 \times fSCL)$	—	s
206	Start condition hold time	—	$1/(2.2 \times fSCL)$	—	s
207	Data hold time	—	0	—	s
208	Data setup time	—	$1/(40 \times fSCL)$	—	s
209	SDL/SCL rise time	—	—	$1/(10 \times fSCL)$	s
210	SDL/SCL fall time	—	—	$1/(33 \times fSCL)$	s
211	Stop condition setup time	—	$1/2(2.2 \times fSCL)$	—	s

¹ SCL frequency is given by $SCL = BrgClk_frequency / ((BRG\ register + 3) \times pre_scaler \times 2)$.
The ratio $SyncClk / (Brg_Clk/pre_scaler)$ must be greater or equal to 4/1.

Figure 70 shows the I²C bus timing.


Figure 70. I²C Bus Timing Diagram

13 UTOPIA AC Electrical Specifications

Table 30, Table 31, and Table 32, show the AC electrical specifications for the UTOPIA interface.

Table 30. UTOPIA Master (Muxed Mode) Electrical Specifications

Num	Signal Characteristic	Direction	Min	Max	Unit
U1	UtpClk rise/fall time (internal clock option)	Output		4	ns
	Duty cycle		50	50	%
	Frequency			33	MHz
U2	UTPB, SOC, $\overline{\text{RxEnb}}$, $\overline{\text{TxEnb}}$, RxAddr, and TxAddr $\overline{\text{}}$ active delay (PHREQ and PHSEL active delay in multi-PHY mode)	Output	2	16	ns
U3	UTPB, SOC, Rxclav, and Txclav setup time	Input	4		ns
U4	UTPB, SOC, Rxclav, and Txclav hold time	Input	1		ns

Table 31. UTOPIA Master (Split Bus Mode) Electrical Specifications

Num	Signal Characteristic	Direction	Min	Max	Unit
U1	UtpClk rise/fall time (Internal clock option)	Output		4	ns
	Duty cycle		50	50	%
	Frequency			33	MHz
U2	UTPB, SOC, $\overline{\text{RxEnb}}$, $\overline{\text{TxEnb}}$, RxAddr, and TxAddr active delay (PHREQ and PHSEL active delay in multi-PHY mode)	Output	2	16	ns
U3	UTPB_Aux, SOC_Aux, Rxclav, and Txclav setup time	Input	4		ns
U4	UTPB_Aux, SOC_Aux, Rxclav, and Txclav hold time	Input	1		ns

Table 32. UTOPIA Slave (Split Bus Mode) Electrical Specifications

Num	Signal Characteristic	Direction	Min	Max	Unit
U1	UtpClk rise/fall time (external clock option)	Input		4	ns
	Duty cycle		40	60	%
	Frequency			33	MHz
U2	UTPB, SOC, Rxclav, and Txclav active delay	Output	2	16	ns
U3	UTPB_AUX, SOC_Aux, $\overline{\text{RxEnb}}$, $\overline{\text{TxEnb}}$, RxAddr, and TxAddr setup time	Input	4		ns
U4	UTPB_AUX, SOC_Aux, $\overline{\text{RxEnb}}$, $\overline{\text{TxEnb}}$, RxAddr, and TxAddr hold time	Input	1		ns

Figure 74 shows the MII transmit signal timing diagram.

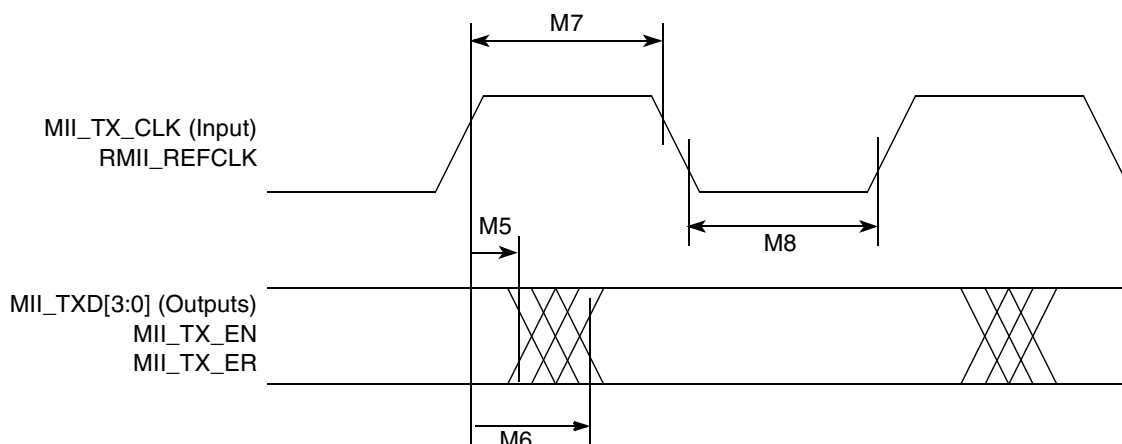


Figure 74. MII Transmit Signal Timing Diagram

15.3 MII Async Inputs Signal Timing (MII_CRSS, MII_COL)

Table 36 provides information on the MII async inputs signal timing.

Table 36. MII Async Inputs Signal Timing

Num	Characteristic	Min	Max	Unit
M9	MII_CRSS, MII_COL minimum pulse width	1.5	—	MII_TX_CLK period

Figure 75 shows the MII asynchronous inputs signal timing diagram.

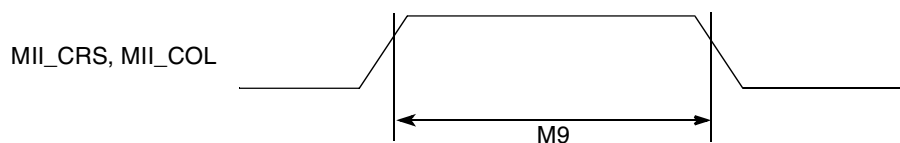


Figure 75. MII Async Inputs Timing Diagram

15.4 MII Serial Management Channel Timing (MII_MDIO, MII_MDC)

Table 37 provides information on the MII serial management channel signal timing. The FEC functions correctly with a maximum MDC frequency in excess of 2.5 MHz.

Table 37. MII Serial Management Channel Timing

Num	Characteristic	Min	Max	Unit
M10	MII_MDC falling edge to MII_MDIO output invalid (minimum propagation delay)	0	—	ns
M11	MII_MDC falling edge to MII_MDIO output valid (max prop delay)	—	25	ns
M12	MII_MDIO (input) to MII_MDC rising edge setup	10	—	ns
M13	MII_MDIO (input) to MII_MDC rising edge hold	0	—	ns

Table 37. MII Serial Management Channel Timing (continued)

Num	Characteristic	Min	Max	Unit
M14	MII_MDC pulse width high	40%	60%	MII_MDC period
M15	MII_MDC pulse width low	40%	60%	MII_MDC period

Figure 76 shows the MII serial management channel timing diagram.

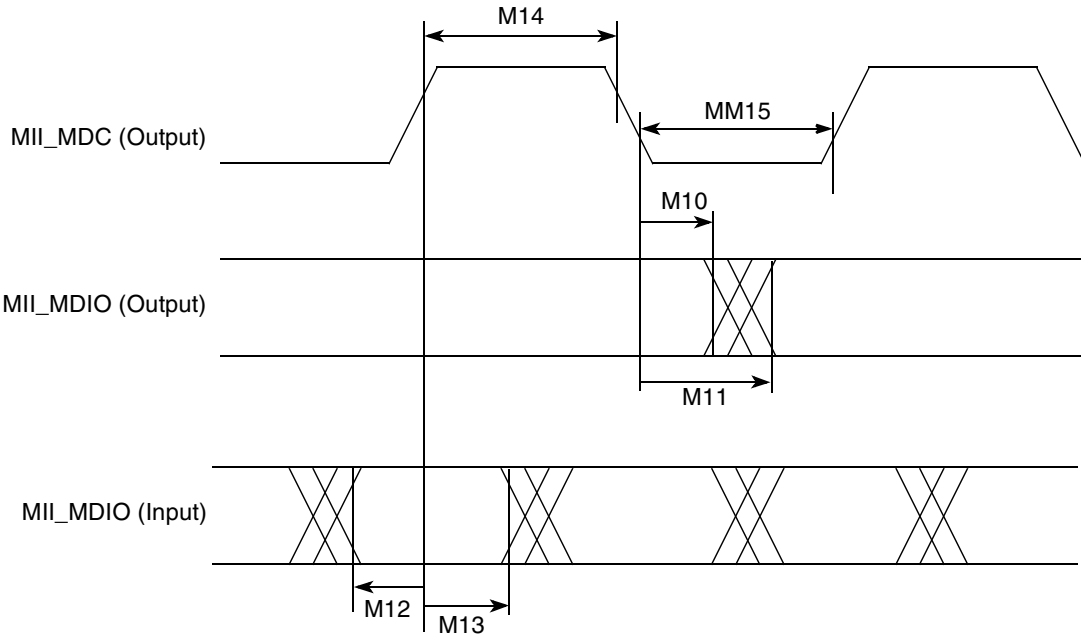


Figure 76. MII Serial Management Channel Timing Diagram

16.1 Pin Assignments

Figure 77 shows the top-view pinout of the PBGA package. For additional information, see the *MPC885 PowerQUICC™ Family Reference Manual*.

NOTE: This is the top view of the device.

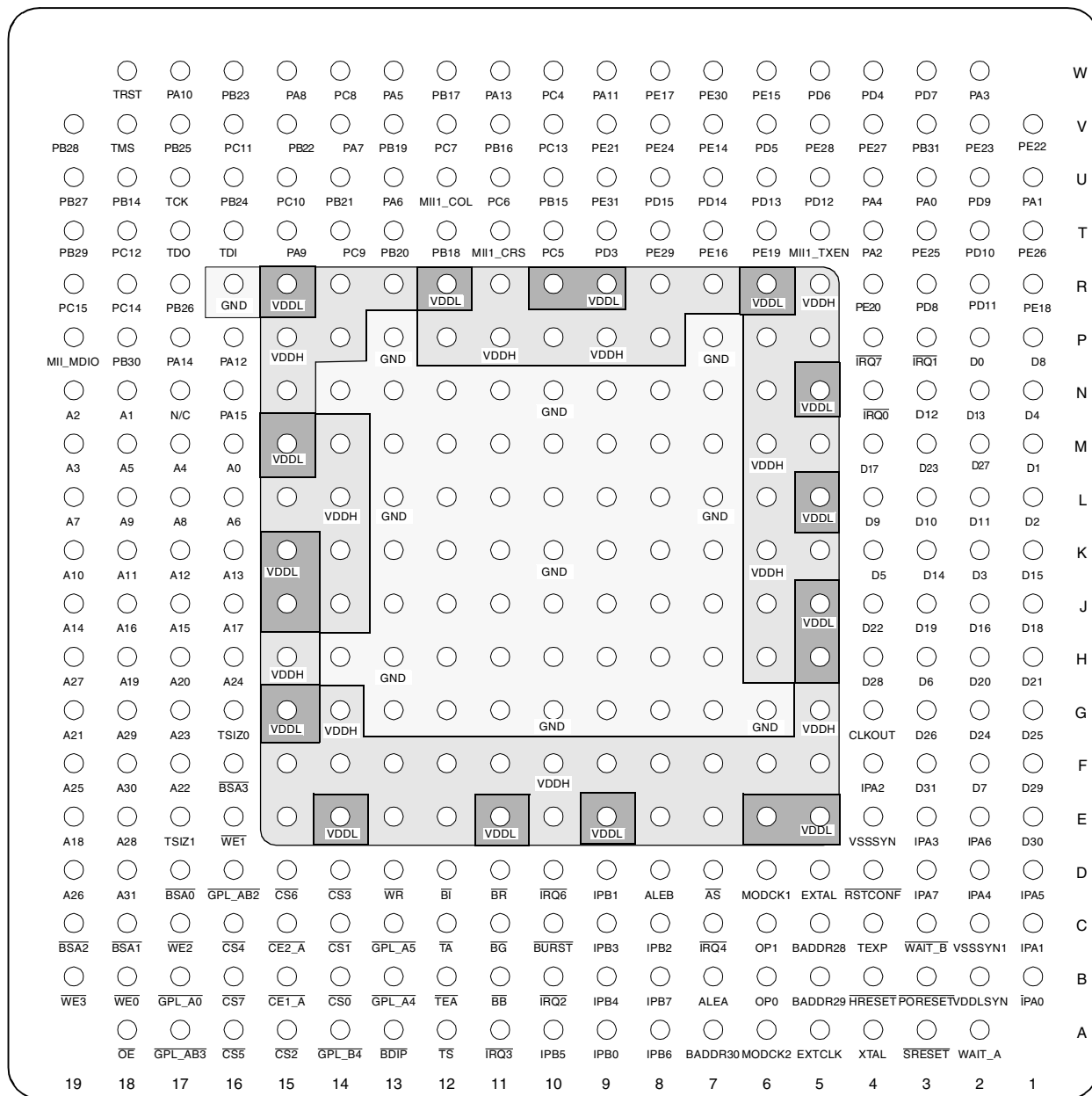


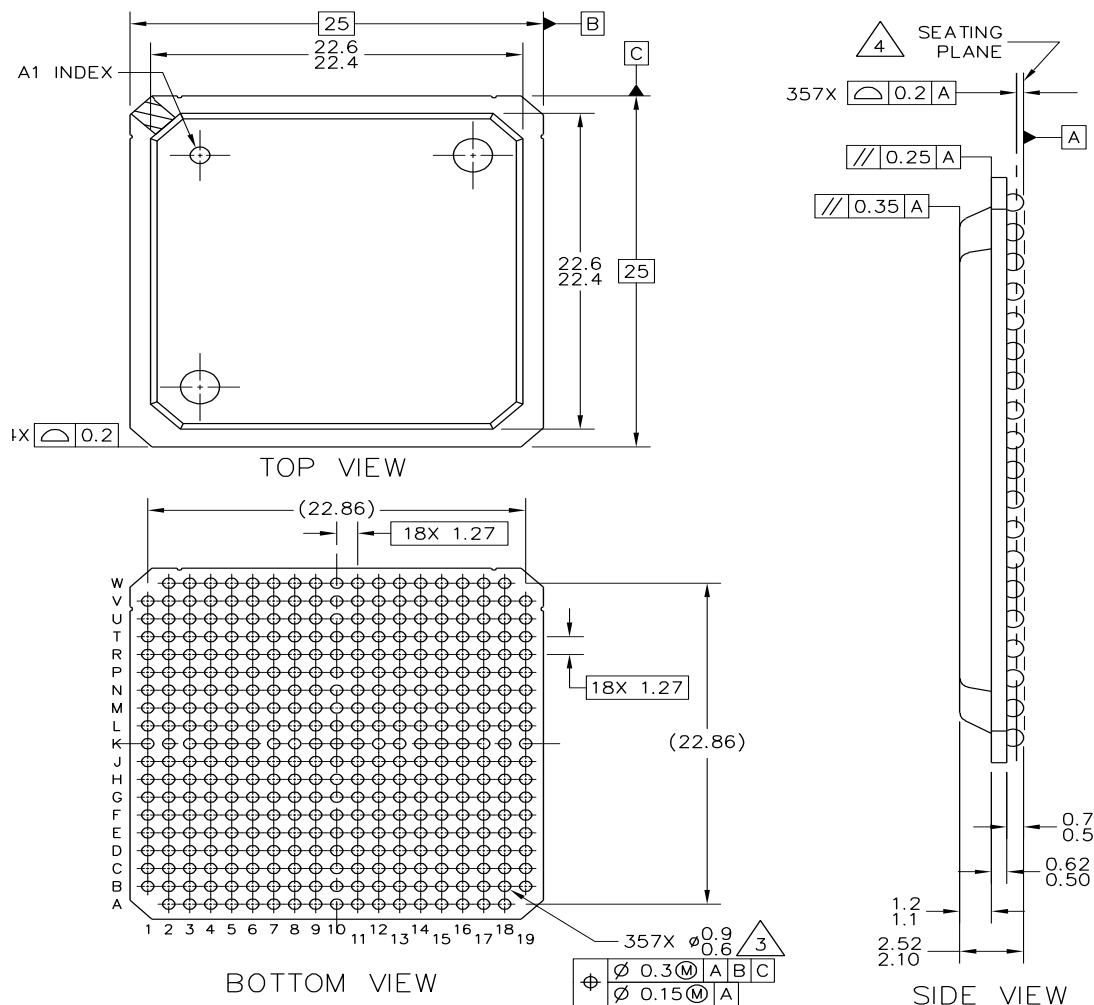
Figure 77. Pinout of the PBGA Package

Table 39. Pin Assignments (continued)

Name	Pin Number	Type
PA4, $\overline{\text{CTS4}}$, MII1-TXD1, RMII1-TXD1	U4	Bidirectional
PA3, MII1-RXER, RMII1-RXER, BRGO3	W2	Bidirectional
PA2, MII1-RXDV, RMII1-CRS_DV, TXD4	T4	Bidirectional
PA1, MII1-RXD0, RMII1-RXD0, BRGO4	U1	Bidirectional
PA0, MII1-RXD1, RMII1-RXD1, $\overline{\text{TOUT4}}$	U3	Bidirectional
PB31, $\overline{\text{SPISEL}}$, MII1-TXCLK, RMII1-REFCLK	V3	Bidirectional (Optional: open-drain)
PB30, SPICLK	P18	Bidirectional (Optional: open-drain)
PB29, SPIMOSI	T19	Bidirectional (Optional: open-drain)
PB28, SPIMISO, BRGO4	V19	Bidirectional (Optional: open-drain)
PB27, I2CSDA, BRGO1	U19	Bidirectional (Optional: open-drain)
PB26, I2CSCL, BRGO2	R17	Bidirectional (Optional: open-drain)
PB25, RXADDR3 ¹ , TXADDR3, SMTXD1	V17	Bidirectional (Optional: open-drain)
PB24, TXADDR3 ¹ , RXADDR3, SMRXD1	U16	Bidirectional (Optional: open-drain)
PB23, TXADDR2 ¹ , RXADDR2, $\overline{\text{SDACK1}}$, SMSYN1	W16	Bidirectional (Optional: open-drain)
PB22, TXADDR4 ¹ , RXADDR4, $\overline{\text{SDACK2}}$, SMSYN2	V15	Bidirectional (Optional: open-drain)
PB21, SMTXD2, TXADDR1 ¹ , BRG01, RXADDR1, PHSEL[1]	U14	Bidirectional (Optional: open-drain)
PB20, SMRXD2, L1CLKOA, TXADDR0 ¹ , RXADDR0, PHSEL[0]	T13	Bidirectional (Optional: open-drain)
PB19, MII1-RXD3, $\overline{\text{RTS4}}$	V13	Bidirectional (Optional: open-drain)
PB18, RXADDR4 ¹ , TXADDR4, $\overline{\text{RTS2}}$, L1ST2	T12	Bidirectional (Optional: open-drain)

16.2 Mechanical Dimensions of the PBGA Package

Figure 78 shows the mechanical dimensions of the PBGA package.



NOTES:

1. ALL DIMENSIONS ARE IN MILLIMETERS.
2. INTERPRET DIMENSIONS AND TOLERANCES PER ASME Y14.5M—1994.
3. MAXIMUM SOLDER BALL DIAMETER MEASURED PARALLEL TO DATUM A.
4. DATUM A, THE SEATING PLANE, IS DEFINED BY THE SPHERICAL CROWNS OF THE SOLDER BALLS.

Figure 78. Mechanical Dimensions and Bottom Surface Nomenclature of the PBGA Package