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#### Details

E·XFI

Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	64MHz
Connectivity	I <sup>2</sup> C, LINbus, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, LVD, POR, PWM, WDT
Number of I/O	36
Program Memory Size	64KB (32K x 16)
Program Memory Type	FLASH
EEPROM Size	1K x 8
RAM Size	3.6K x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.6V
Data Converters	A/D 35x10b; D/A 1x5b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	40-UFQFN Exposed Pad
Supplier Device Package	40-UQFN (5x5)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic18lf46k40-i-mv

Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

## PIC18LF26/45/46K40



## 11.3 Data EEPROM Memory

The data EEPROM is a nonvolatile memory array, separate from the data RAM and program memory, which is used for long-term storage of program data. It is not directly mapped in either the register file or program memory space but is indirectly addressed through the Special Function Registers (SFRs). The EEPROM is readable and writable during normal operation over the entire VDD range.

Four SFRs are used to read and write to the data EEPROM as well as the program memory. They are:

- NVMCON1
- NVMCON2
- NVMDAT
- NVMADRL
- NVMADRH(1)

Note 1: NVMADRH register is not implemented on PIC18(L)F45K40.

The data EEPROM allows byte read and write. When interfacing to the data memory block, NVMDAT holds the 8-bit data for read/write and the NVMADRH:NVMADRL register pair hold the address of the EEPROM location being accessed.

The EEPROM data memory is rated for high erase/write cycle endurance. A byte write automatically erases the location and writes the new data (erase-before-write). The write time is controlled by an internal programming timer; it will vary with voltage and temperature as well as from chip-to-chip. Please refer to the Data EEPROM Memory parameters in **Section 37.0 "Electrical Specifications"** for limits.

#### 11.3.1 NVMADRL AND NVMADRH REGISTERS

The NVMADRH:NVMADRL registers are used to address the data EEPROM for read and write operations.

#### 11.3.2 NVMCON1 AND NVMCON2 REGISTERS

Access to the data EEPROM is controlled by two registers: NVMCON1 and NVMCON2. These are the same registers which control access to the program memory and are used in a similar manner for the data EEPROM.

The NVMCON1 register (Register 11-1) is the control register for data and program memory access. Control bits NVMREG<1:0> determine if the access will be to program, Data EEPROM Memory or the User IDs, Configuration bits, Revision ID and Device ID.

The WREN bit, when set, will allow a write operation. On power-up, the WREN bit is clear.

The WRERR bit is set by hardware when the WR bit is set and cleared when the internal programming timer expires and the write operation is complete.

The WR control bit initiates write operations. The bit can be set but not cleared by software. It is cleared only by hardware at the completion of the write operation.

The NVMIF interrupt flag bit of the PIR7 register is set when the write is complete. It must be cleared by software.

Control bits, RD and WR, start read and erase/write operations, respectively. These bits are set by firmware and cleared by hardware at the completion of the operation.

The RD bit cannot be set when accessing program memory (NVMREG<1:0> = 0x10). Program memory is read using table read instructions. See **Section 11.1.1 "Table Reads and Table Writes"** regarding table reads.

R/W-0/0	R/W-0/0	U-0	U-0	U-0	U-0	R/W-0/0	R/W-0/0
OSCFIF	CSWIF <sup>(1)</sup>	_	—	_	—	ADTIF	ADIF
bit 7							bit 0
Legend:							
R = Readable	e bit	W = Writable	bit	U = Unimpler	mented bit, read	l as '0'	
-n = Value at	POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unkr	nown
bit 7	OSCFIF: Osc	illator Fail Inter	rupt Flag bit				
	1 = Device oscillator failed, clock input has changed to HFINTOSC (must be cleared by software)					y software)	
	0 = Device cl	ock operating					
bit 6	CSWIF: Clock	k-Switch Interru	ipt Flag bit <sup>(1)</sup>				
	1 = New osci	llator is ready f	or switch (mu	st be cleared b	oy software) (se	e Figure 4-6 an	d Figure 4-7)
	0 = New oscillator is not ready for switch or has not been started						
bit 5-2	Unimplemen	ted: Read as '	)´				
bit 1	ADTIF: ADC	Threshold Inter	rupt Flag bit				
	1 = ADC Three	eshold interrup	t has occurred	d (must be clea	ared by software	e)	
	0 = ADC Three	eshold event is	not complete	or has not bee	en started		
bit 0	ADIF: ADC In	iterrupt Flag bit					
	1 = An A/D cc	onversion comp	pleted (must b	e cleared by s	oftware)		
	0 = 1  he A/D c	conversion is no	ot complete or	has not been	started		
Note 1: ⊺	he CSWIF interr	upt will not wa	ke the systen	n from Sleep.	The system will	sleep until an	other interrupt

## REGISTER 14-3: PIR1: PERIPHERAL INTERRUPT REQUEST (FLAG) REGISTER 1

causes the wake-up.

U-0	U-0	U-0	U-0	U-0	U-0	R/W-0/0	R/W-0/0		
—	—	—	_	—	—	CCP2IE	CCP1IE		
bit 7							bit 0		
Legend:									
R = Readable bit W = Writable bit			bit	U = Unimplemented bit, read as '0'					
-n = Value at POR '1' = Bit is set				'0' = Bit is cleared x = Bit is unknown					
bit 7-2	Unimplemen	ted: Read as '	0'						
bit 1 CCP2IE: ECCP2 Interrupt Enable bit 1 = Enabled 0 = Disabled		nable bit							
bit 0 CCP1IE: ECCP1 Interrupt Enable bit 1 = Enabled 0 = Disabled			nable bit						

#### REGISTER 14-16: PIE6: PERIPHERAL INTERRUPT ENABLE REGISTER 6

REGISTER 17-2:	RxyPPS: PIN Rxy	OUTPUT SOURCE SELECTION REGISTER

U-0	U-0	U-0	R/W-0/u	R/W-0/u	R/W-0/u	R/W-0/u	R/W-0/u
—	—	—			RxyPPS<4:02	>	
bit 7							bit 0

Legend:
---------

-		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-5 Unimplemented: Read as '0'

bit 4-0 RxyPPS<4:0>: Pin Rxy Output Source Selection bits<sup>(1)</sup>

	Din Dyy Output Source	Device Configuration								
KXYPPS<4:02	Pill Kxy Output Source	PIC1	8(L)F2	6K40	PIC18(L)F45/46K40					
5'b1 0111	ADGRDB	А		С	Α	_	С	_	_	
5'b1 0110	ADGRDA	А		С	Α	—	С		_	
5'b1 0101	DSM	А		С	Α	—	—	D		
5'b1 0100	CLKR	_	В	С	—	В	С	—	_	
5'b1 0011	TMR0	_	В	С	—	В	С	—	_	
5'b1 0010	MSSP2 (SDO/SDA)	_	В	С	—	В	_	D	_	
5'b1 0001	MSSP2 (SCK/SCL)		В	С	—	В	—	D		
5'b1 0000	MSSP1 (SDO/SDA)	_	В	С	—	В	С	—	_	
5'b0 1111	MSSP1 (SCK/SCL)	_	В	С	—	В	С	—	_	
5'b0 1110	CMP2	А		С	Α	—	—		Е	
5'b0 1101	CMP1	А		С	Α	—	—	D	_	
5'b0 1100	EUSART2 (RX)	_	В	С	—	В	_	D	_	
5'b0 1011	EUSART2 (TX)	_	В	С	—	В	_	D	_	
5'b0 1010	EUSART1 (RX)	_	В	С	—	В	С	—	_	
5'b0 1001	EUSART1 (TX)	_	В	С	—	В	С	—	_	
5'b0 1000	PWM4	А	—	С	Α		С	—	_	
5'b0 0111	PWM3	А	—	С	Α		_	D	_	
5'b0 0110	CCP2	_	В	С	—	В	С	—	_	
5'b0 0101	CCP1	_	В	С	—	В	С	—	_	
5'b0 0100	CWG1D	_	В	С	—	В	_	D	_	
5'b0 0011	CWG1C		В	С		В		D	—	
5'b0 0010	CWG1B		В	С		В		D	—	
5'b0 0001	CWG1A		В	С		В	С		—	
5'b0 0000	LATxy	А	В	С	А	В	С	D	Е	

**Note 1:** PORTD is present only on the PIC18(L)F45/46K40 devices.

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# PIC18(L)F26/45/46K40

REGISTER 18	3-2: T0CO	N1: TIMER0 (	CONTROL R	EGISTER 1			
R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
	T0CS<2:0>		TOASYNC		TOCKP	S<3:0>	
bit 7							bit 0
Legend:							
R = Readable I	bit	W = Writable	bit	U = Unimple	mented bit, read	as '0'	
u = Bit is uncha	anged	x = Bit is unkr	nown	-n/n = Value	at POR and BO	R/Value at all c	ther Resets
'1' = Bit is set		'0' = Bit is cle	ared				
bit 7-5	<b>TOCS&lt;2:0&gt;:1</b> 111 = Reserv 110 = Reserv 101 = SOSC 100 = LFINT 011 = HFINT 010 = Fosc/2 001 = Pin sel 000 = Pin sel	Fimer0 Clock So /ed OSC OSC 4 lected by T0CK lected by T0CK	UPPS (Inverter UPPS (Non-inv	its d) /erted)			
bit 4	TOASYNC: T	MR0 Input Asy	nchronization	Enable bit	to system clocks		
	0 = The input	it to the TMR0	counter is syn	chronized to F	OSC/4	>	
bit 3-0	<b>TOCKPS&lt;3:0</b> 1111 = 1:327 1110 = 1:163 1101 = 1:819 1100 = 1:409 1011 = 1:204 1010 = 1:102 1001 = 1:512 1000 = 1:256 0111 = 1:128 0100 = 1:4 0011 = 1:8 0010 = 1:4 0000 = 1:1	<ul> <li>&gt;: Prescaler R</li> <li>768</li> <li>384</li> <li>302</li> <li>306</li> <li>48</li> <li>34</li> <li>34</li> <li>35</li> <li>36</li> </ul>	ate Select bit				

# PIC18(L)F26/45/46K40



### 20.4 Timer2 Interrupt

Timer2 can also generate a device interrupt. The interrupt is generated when the postscaler counter matches one of 16 postscale options (from 1:1 through 1:16), which are selected with the postscaler control bits, OUTPS<3:0> of the T2CON register. The interrupt is enabled by setting the TMR2IE interrupt enable bit of the PIE4 register. Interrupt timing is illustrated in Figure 20-3.

#### FIGURE 20-3: TIMER2 PRESCALER, POSTSCALER, AND INTERRUPT TIMING DIAGRAM

	Rex 10.000056. 47/2019
CKPS	0b010
PRx	1
OUTPS	0b0001
TMRx_clk	
TMRx	
TMRx_postscaled	
TMRxIF	(1) (2)
Note 1: 2:	Setting the interrupt flag is synchronized with the instruction clock. Synchronization may take as many as 2 instruction cycles Cleared by software.

#### 20.5.6 EDGE-TRIGGERED ONE-SHOT MODE

The Edge-Triggered One-Shot modes start the timer on an edge from the external signal input, after the ON bit is set, and clear the ON bit when the timer matches the PRx period value. The following edges will start the timer:

- Rising edge (MODE<4:0> = 01001)
- Falling edge (MODE<4:0> = 01010)
- Rising or Falling edge (MODE<4:0> = 01011)

If the timer is halted by clearing the ON bit then another TMRx\_ers edge is required after the ON bit is set to resume counting. Figure 20-9 illustrates operation in the rising edge One-Shot mode.

When Edge-Triggered One-Shot mode is used in conjunction with the CCP then the edge-trigger will activate the PWM drive and the PWM drive will deactivate when the timer matches the CCPRx pulse width value and stay deactivated when the timer halts at the PRx period count match.

#### FIGURE 20-9: EDGE-TRIGGERED ONE-SHOT MODE TIMING DIAGRAM (MODE = 01001)



#### 20.5.7 EDGE-TRIGGERED HARDWARE LIMIT ONE-SHOT MODE

In Edge-Triggered Hardware Limit One-Shot modes the timer starts on the first external signal edge after the ON bit is set and resets on all subsequent edges. Only the first edge after the ON bit is set is needed to start the timer. The counter will resume counting automatically two clocks after all subsequent external Reset edges. Edge triggers are as follows:

- Rising edge start and Reset (MODE<4:0> = 01100)
- Falling edge start and Reset (MODE<4:0> = 01101)

The timer resets and clears the ON bit when the timer value matches the PRx period value. External signal edges will have no effect until after software sets the ON bit. Figure 20-10 illustrates the rising edge hardware limit one-shot operation.

When this mode is used in conjunction with the CCP then the first starting edge trigger, and all subsequent Reset edges, will activate the PWM drive. The PWM drive will deactivate when the timer matches the CCPRx pulse-width value and stay deactivated until the timer halts at the PRx period match unless an external signal edge resets the timer before the match occurs.

## 21.0 CAPTURE/COMPARE/PWM MODULE

The Capture/Compare/PWM module is a peripheral that allows the user to time and control different events, and to generate Pulse-Width Modulation (PWM) signals. In Capture mode, the peripheral allows the timing of the duration of an event. The Compare mode allows the user to trigger an external event when a predetermined amount of time has expired. The PWM mode can generate Pulse-Width Modulated signals of varying frequency and duty cycle.

This family of devices contains two standard Capture/Compare/PWM modules (CCP1 and CCP2). Each individual CCP module can select the timer source that controls the module. Each module has an independent timer selection which can be accessed using the CxTSEL bits in the CCPTMRS register (Register 21-2). The default timer selection is TMR1 when using Capture/Compare mode and TMR2 when using PWM mode in the CCPx module.

Please note that the Capture/Compare mode operation is described with respect to TMR1 and the PWM mode operation is described with respect to TMR2 in the following sections.

The Capture and Compare functions are identical for all CCP modules.

- Note 1: In devices with more than one CCP module, it is very important to pay close attention to the register names used. A number placed after the module acronym is used to distinguish between separate modules. For example, the CCP1CON and CCP2CON control the same operational aspects of two completely different CCP modules.
  - 2: Throughout this section, generic references to a CCP module in any of its operating modes may be interpreted as being equally applicable to CCPx module. Register names, module signals, I/O pins, and bit names may use the generic designator 'x' to indicate the use of a numeral to distinguish a particular module, when required.

## 21.1 CCP Module Configuration

Each Capture/Compare/PWM module is associated with a control register (CCPxCON), a capture input selection register (CCPxCAP) and a data register (CCPRx). The data register, in turn, is comprised of two 8-bit registers: CCPRxL (low byte) and CCPRxH (high byte).

#### 21.1.1 CCP MODULES AND TIMER RESOURCES

The CCP modules utilize Timers 1 through 6 that vary with the selected mode. Various timers are available to the CCP modules in Capture, Compare or PWM modes, as shown in Table 21-1.

#### TABLE 21-1: CCP MODE – TIMER RESOURCE

CCP Mode	Timer Resource					
Capture						
Compare	Timer1, Timer3 or Timer5					
PWM	Timer2, Timer4 or Timer6					

The assignment of a particular timer to a module is determined by the timer to CCP enable bits in the CCPTMRS register (see Register 21-2) All of the modules may be active at once and may share the same timer resource if they are configured to operate in the same mode (Capture/Compare or PWM) at the same time.

#### 21.1.2 OPEN-DRAIN OUTPUT OPTION

When operating in Output mode (the Compare or PWM modes), the drivers for the CCPx pins can be optionally configured as open-drain outputs. This feature allows the voltage level on the pin to be pulled to a higher level through an external pull-up resistor and allows the output to communicate with external circuits without the need for additional level shifters.





The I<sup>2</sup>C bus specifies two signal connections:

- · Serial Clock (SCL)
- Serial Data (SDA)

Both the SCL and SDA connections are bidirectional open-drain lines, each requiring pull-up resistors for the supply voltage. Pulling the line to ground is considered a logical zero and letting the line float is considered a logical one.

Figure 26-11 shows a typical connection between two processors configured as master and slave devices.

The  $I^2C$  bus can operate with one or more master devices and one or more slave devices.

There are four potential modes of operation for a given device:

- Master Transmit mode
   (master is transmitting data to a slave)
- Master Receive mode
   (master is receiving data from a slave)
- Slave Transmit mode (slave is transmitting data to a master)
- Slave Receive mode
  - (slave is receiving data from the master)

To begin communication, a master device starts out in Master Transmit mode. The master device sends out a Start bit followed by the address byte of the slave it intends to communicate with. This is followed by a single Read/Write bit, which determines whether the master intends to transmit to or receive data from the slave device. If the requested slave exists on the bus, it will respond with an Acknowledge bit, otherwise known as an ACK. The master then continues in either Transmit mode or Receive mode and the slave continues in the complement, either in Receive mode or Transmit mode, respectively.

A Start bit is indicated by a high-to-low transition of the SDA line while the SCL line is held high. Address and data bytes are sent out, Most Significant bit (MSb) first. The Read/Write bit is sent out as a logical one when the master intends to read data from the slave, and is sent out as a logical zero when it intends to write data to the slave.

#### FIGURE 26-11: I<sup>2</sup>C MASTER/ SLAVE CONNECTION



The Acknowledge bit  $(\overline{ACK})$  is an active-low signal, which holds the SDA line low to indicate to the transmitter that the slave device has received the transmitted data and is ready to receive more.

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#### 26.8.5 START CONDITION

The  $I^2C$  specification defines a Start condition as a transition of SDA from a high to a low state while SCL line is high. A Start condition is always generated by the master and signifies the transition of the bus from an Idle to an Active state. Figure 26-12 shows wave forms for Start and Stop conditions.

A bus collision can occur on a Start condition if the module samples the SDA line low before asserting it low. This does not conform to the I<sup>2</sup>C Specification that states no bus collision can occur on a Start.

#### 26.8.6 STOP CONDITION

A Stop condition is a transition of the SDA line from low-to-high state while the SCL line is high.

Note: At least one SCL low time must appear before a Stop is valid, therefore, if the SDA line goes low then high again while the SCL line stays high, only the Start condition is detected.

### 26.8.7 RESTART CONDITION

A Restart is valid any time that a Stop would be valid. A master can issue a Restart if it wishes to hold the bus after terminating the current transfer. A Restart has the same effect on the slave that a Start would, resetting all slave logic and preparing it to clock in an address. The master may want to address the same or another slave. Figure 26-13 shows the wave form for a Restart condition.

In 10-bit Addressing Slave mode a Restart is required for the master to clock data out of the addressed slave. Once a slave has been fully addressed, matching both high and low address bytes, the master can issue a Restart and the high address byte with the  $R/\overline{W}$  bit set. The slave logic will then hold the clock and prepare to clock out data.

After a full match with R/W clear in 10-bit mode, a prior match flag is set and maintained until a Stop condition, a high address with R/W clear, or high address match fails.

#### 26.8.8 START/STOP CONDITION INTERRUPT MASKING

The SCIE and PCIE bits of the SSPxCON3 register can enable the generation of an interrupt in Slave modes that do not typically support this function. Slave modes where interrupt on Start and Stop detect are already enabled, these bits will have no effect.







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#### 26.9.2.1 7-bit Addressing Reception

This section describes a standard sequence of events for the MSSP module configured as an  $I^2C$  slave in 7-bit Addressing mode. Figure 26-14 and Figure 26-15 is used as a visual reference for this description.

This is a step by step process of what typically must be done to accomplish  $I^2C$  communication.

- 1. Start bit detected.
- 2. S bit of SSPxSTAT is set; SSPxIF is set if interrupt on Start detect is enabled.
- 3. Matching address with  $R/\overline{W}$  bit clear is received.
- 4. The slave pulls SDA low sending an ACK to the master, and sets SSPxIF bit.
- 5. Software clears the SSPxIF bit.
- 6. Software reads received address from SSPxBUF clearing the BF flag.
- 7. If SEN = 1; Slave software sets CKP bit to release the SCL line.
- 8. The master clocks out a data byte.
- 9. Slave drives SDA low sending an ACK to the master, and sets SSPxIF bit.
- 10. Software clears SSPxIF.
- 11. Software reads the received byte from SSPxBUF clearing BF.
- 12. Steps 8-12 are repeated for all received bytes from the master.
- 13. Master sends Stop condition, setting P bit of SSPxSTAT, and the bus goes idle.

## 26.9.2.2 7-bit Reception with AHEN and DHEN

Slave device reception with AHEN and DHEN set operate the same as without these options with extra interrupts and clock stretching added after the eighth falling edge of SCL. These additional interrupts allow the slave software to decide whether it wants to ACK the receive address or data byte, rather than the hardware. This functionality adds support for PMBus<sup>™</sup> that was not present on previous versions of this module.

This list describes the steps that need to be taken by slave software to use these options for  $I^2C$  communication. Figure 26-16 displays a module using both address and data holding. Figure 26-17 includes the operation with the SEN bit of the SSPxCON2 register set.

- 1. S bit of SSPxSTAT is set; SSPxIF is set if interrupt on Start detect is enabled.
- Matching address with R/W bit clear is clocked in. SSPxIF is set and CKP cleared after the eighth falling edge of SCL.
- 3. Slave clears the SSPxIF.
- Slave can look at the ACKTIM bit of the SSPxCON3 register to <u>determine</u> if the SSPxIF was after or before the ACK.
- 5. Slave reads the address value from SSPxBUF, clearing the BF flag.
- 6. Slave sets ACK value clocked out to the master by setting ACKDT.
- 7. Slave releases the clock by setting CKP.
- 8. SSPxIF is set after an ACK, not after a NACK.
- 9. If SEN = 1 the slave hardware will stretch the clock after the ACK.
- 10. Slave clears SSPxIF.

Note: SSPxIF is still set after the ninth falling edge of SCL even if there is no clock stretching and BF has been cleared. Only if NACK is sent to master is SSPxIF not set

- 11. SSPxIF set and CKP cleared after eighth falling edge of SCL for a received data byte.
- 12. Slave looks at ACKTIM bit of SSPxCON3 to determine the source of the interrupt.
- 13. Slave reads the received data from SSPxBUF clearing BF.
- 14. Steps 7-14 are the same for each received data byte.
- 15. Communication is ended by either the slave sending an ACK = 1, or the master sending a Stop condition. If a Stop is sent and Interrupt on Stop Detect is disabled, the slave will only know by polling the P bit of the SSTSTAT register.





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## 27.3 Clock Accuracy with Asynchronous Operation

The factory calibrates the internal oscillator block output (INTOSC). However, the INTOSC frequency may drift as VDD or temperature changes, and this directly affects the asynchronous baud rate. Two methods may be used to adjust the baud rate clock, but both require a reference clock source of some kind.

The first (preferred) method uses the OSCTUNE register to adjust the INTOSC output. Adjusting the value in the OSCTUNE register allows for fine resolution changes to the system clock source. See **Section 4.3.2.3 "Internal Oscillator Frequency Adjustment"** for more information.

The other method adjusts the value in the Baud Rate Generator. This can be done automatically with the Auto-Baud Detect feature (see **Section 27.4.1 "Auto-Baud Detect"**). There may not be fine enough resolution when adjusting the Baud Rate Generator to compensate for a gradual change in the peripheral clock frequency.

## 27.4.2 AUTO-BAUD OVERFLOW

During the course of automatic baud detection, the ABDOVF bit of the BAUDxCON register will be set if the baud rate counter overflows before the fifth rising edge is detected on the RXx pin. The ABDOVF bit indicates that the counter has exceeded the maximum count that can fit in the 16 bits of the SPxBRGH:SPxBRGL register pair. After the ABDOVF bit has been set, the counter continues to count until the fifth rising edge is detected on the RXx pin. Upon detecting the fifth RX edge, the hardware will set the RCxIF interrupt flag and clear the ABDEN bit of the BAUDxCON register. The RCxIF flag can be subsequently cleared by reading the RCxREG register. The ABDOVF flag of the BAUDxCON register can be cleared by software directly.

To terminate the auto-baud process before the RCxIF flag is set, clear the ABDEN bit then clear the ABDOVF bit of the BAUDxCON register. The ABDOVF bit will remain set if the ABDEN bit is not cleared first.

#### 27.4.3 AUTO-WAKE-UP ON BREAK

During Sleep mode, all clocks to the EUSART are suspended. Because of this, the Baud Rate Generator is inactive and a proper character reception cannot be performed. The Auto-Wake-up feature allows the controller to wake-up due to activity on the RX/DT line. This feature is available only in Asynchronous mode.

The Auto-Wake-up feature is enabled by setting the WUE bit of the BAUDxCON register. Once set, the normal receive sequence on RX/DT is disabled, and the EUSART remains in an Idle state, monitoring for a wake-up event independent of the CPU mode. A wake-up event consists of a high-to-low transition on the RX/DT line. (This coincides with the start of a Sync Break or a wake-up signal character for the LIN protocol.)

The EUSART module generates an RCxIF interrupt coincident with the wake-up event. The interrupt is generated synchronously to the Q clocks in normal CPU operating modes (Figure 27-7), and asynchronously if the device is in Sleep mode (Figure 27-8). The interrupt condition is cleared by reading the RCxREG register.

The WUE bit is automatically cleared by the low-to-high transition on the RX line at the end of the Break. This signals to the user that the Break event is over. At this point, the EUSART module is in Idle mode waiting to receive the next character.

#### 27.4.3.1 Special Considerations

#### Break Character

To avoid character errors or character fragments during a wake-up event, the wake-up character must be all zeros.

When the wake-up is enabled the function works independent of the low time on the data stream. If the WUE bit is set and a valid non-zero character is received, the low time from the Start bit to the first rising edge will be interpreted as the wake-up event. The remaining bits in the character will be received as a fragmented character and subsequent characters can result in framing or overrun errors.

Therefore, the initial character in the transmission must be all '0's. This must be ten or more bit times, 13-bit times recommended for LIN bus, or any number of bit times for standard RS-232 devices.

#### Oscillator Start-up Time

Oscillator start-up time must be considered, especially in applications using oscillators with longer start-up intervals (i.e., LP, XT or HS/PLL mode). The Sync Break (or wake-up signal) character must be of sufficient length, and be followed by a sufficient interval, to allow enough time for the selected oscillator to start and provide proper initialization of the EUSART.

#### WUE Bit

The wake-up event causes a receive interrupt by setting the RCxIF bit. The WUE bit is cleared in hardware by a rising edge on RX/DT. The interrupt condition is then cleared in software by reading the RCxREG register and discarding its contents.

To ensure that no actual data is lost, check the RCIDL bit to verify that a receive operation is not in process before setting the WUE bit. If a receive operation is not occurring, the WUE bit may then be set just prior to entering the Sleep mode.

TABLE 27-7:	SUMMARY OF REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER
	TRANSMISSION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
BAUDxCON	ABDOVF	RCIDL	_	SCKP	BRG16	—	WUE	ABDEN	395
INTCON	GIE/GIEH	PEIE/GIEL	IPEN				INT1EDG	INT0EDG	170
PIE3	RC2IE	TX2IE	RC1IE	TX1IE	BCL2IE	SSP2IE	BCL1IE	SSP1IE	182
PIR3	RC2IF	TX2IF	RC1IF	TX1IF	BCL2IF	SSP2IF	BCL1IF	SSP1IF	174
IPR3	RC2IP	TX2IP	RC1IP	TX1IP	BCL2IP	SSP2IP	BCL1IP	SSP1IP	190
RCxSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	394
RxyPPS		—			218				
TXxPPS		—		TXPPS<4:0>					216
SPxBRGH	EUSARTx Baud Rate Generator, High Byte							404*	
SPxBRGL	EUSARTx Baud Rate Generator, Low Byte							404*	
TXxREG	EUSARTx Transmit Data Register							396*	
TXxSTA	CSRC	TX9	TXEN	SYNC SENDB BRGH TRMT TX9D				393	

Legend: — = unimplemented location, read as '0'. Shaded cells are not used for synchronous master transmission.

\* Page provides register information.









#### TABLE 35-2: INSTRUCTION SET (CONTINUED)

Mnemonic, Operands		Description	Cycles	16-Bit Instruction Word				Status	Notos
				MSb			LSb	Affected	NOTES
LITERAL OPERATIONS									
ADDLW	k	Add literal and WREG	1	0000	1111	kkkk	kkkk	C, DC, Z, OV, N	
ANDLW	k	AND literal with WREG	1	0000	1011	kkkk	kkkk	Z, N	
IORLW	k	Inclusive OR literal with WREG	1	0000	1001	kkkk	kkkk	Z, N	
LFSR	f, k	Move literal (12-bit) 2nd word	2	1110	1110	00ff	kkkk	None	
		to FSR(f) 1st word		1111	0000	kkkk	kkkk		
MOVLB	k	Move literal to BSR<3:0>	1	0000	0001	0000	kkkk	None	
MOVLW	k	Move literal to WREG	1	0000	1110	kkkk	kkkk	None	
MULLW	k	Multiply literal with WREG	1	0000	1101	kkkk	kkkk	None	
RETLW	k	Return with literal in WREG	2	0000	1100	kkkk	kkkk	None	
SUBLW	k	Subtract WREG from literal	1	0000	1000	kkkk	kkkk	C, DC, Z, OV, N	
XORLW	k	Exclusive OR literal with WREG	1	0000	1010	kkkk	kkkk	Z, N	
DATA MEMORY									
TBLRD*		Table Read	2	0000	0000	0000	1000	None	
TBLRD*+		Table Read with post-increment		0000	0000	0000	1001	None	
TBLRD*-		Table Read with post-decrement		0000	0000	0000	1010	None	
TBLRD+*		Table Read with pre-increment		0000	0000	0000	1011	None	
TBLWT*		Table Write	2	0000	0000	0000	1100	None	
TBLWT*+		Table Write with post-increment		0000	0000	0000	1101	None	
TBLWT*-		Table Write with post-decrement		0000	0000	0000	1110	None	
TBLWT+*		Table Write with pre-increment		0000	0000	0000	1111	None	

**Note 1:** When a PORT register is modified as a function of itself (e.g., MOVF PORTB, 1, 0), the value used will be that value present on the pins themselves. For example, if the data latch is '1' for a pin configured as input and is driven low by an external device, the data will be written back with a '0'.

2: If this instruction is executed on the TMR0 register (and where applicable, 'd' = 1), the prescaler will be cleared if assigned.

**3:** If Program Counter (PC) is modified or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOP.

4: Some instructions are two-word instructions. The second word of these instructions will be executed as a NOP unless the first word of the instruction retrieves the information embedded in these 16 bits. This ensures that all program memory locations have a valid instruction.

# PIC18(L)F26/45/46K40

TBLWT	Table Write						
Syntax:	TBLWT ( *; *+; *-; +*)						
Operands:	None						
Operation:	if TBLWT*, (TABLAT) $\rightarrow$ Holding Register; TBLPTR – No Change; if TBLWT*+, (TABLAT) $\rightarrow$ Holding Register; (TBLPTR) + 1 $\rightarrow$ TBLPTR; if TBLWT*-, (TABLAT) $\rightarrow$ Holding Register; (TBLPTR) – 1 $\rightarrow$ TBLPTR; if TBLWT+*, (TABLPTR) + 1 $\rightarrow$ TBLPTR; (TABLAT) $\rightarrow$ Holding Register;						
Status Affected:	None						
Encoding:	0000	0000	0000	11nn nn=0 * =1 *+ =2 *- =3 +*			
	TBLPTR to determine which of the eight holding registers the TABLAT is written to. The holding registers are used to program the contents of Program Memory (P.M.). (Refer to Section 11.1 "Program Flash Memory" for additional details on pro- gramming Flash memory.) The TBLPTR (a 21-bit pointer) points to each byte in the program memory. TBLPTR has a 2-MByte address range. The LSb of the TBLPTR selects which byte of the program memory location to access. TBLPTR[0] = 0: Least Significant Byte of Program Memory Word TBLPTR[0] = 1: Most Significant Byte of Program Memory Word The TBLWT instruction can modify the value of TBLPTR as follows: <ul> <li>no change</li> <li>post-increment</li> <li>post-decrement</li> </ul>						
Words:	1						
Cycles:	2						
Q Cycle Activity:							
	Q1	Q2	Q3	Q4			
	Decode	No	No	No			
	No	No	No	No			
	operation	operation (Read TABLAT)	operation	operation (Write to Holding			

#### TBLWT Table Write (Continued)

Example1:	TBLWT *+;						
Before Instruction							
TABLAT TBLPTF HOLDIN	R R IG REGISTER	=	55h 00A356h				
(00A35	56h)	=	FFh				
After Instruct	ions (table write	completion)					
TABLAT		=	55h				
		=	00A357h				
(00A35	56h)	=	55h				
Example 2:	TBLWT +*;						
Before Instru	ction						
TABLAT		=	34h				
		=	01389Ah				
(01389 HOLDIN	Ah) IG REGISTER	=	FFh				
(01389	Bh)	=	FFh				
After Instruct	on (table write o	comple	ompletion)				
TABLAT		=	34h				
		=	01389Bh				
(01389 HOLDIN	IG REGISTER	=	FFh				
(01389	Bh)	=	34h				

Register)